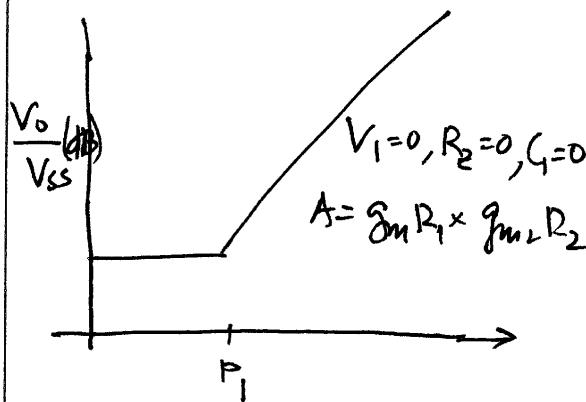
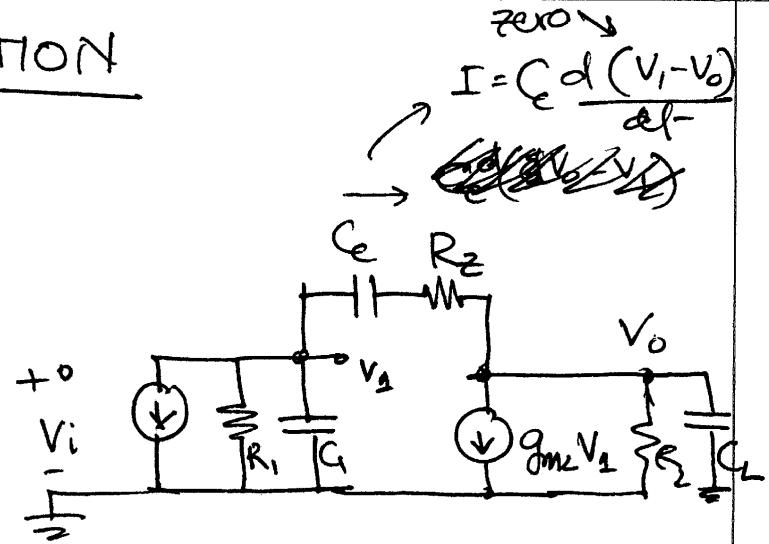
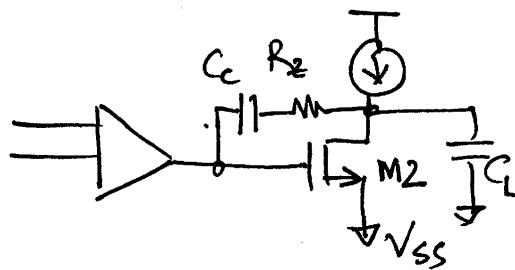


0

AHUJA COMPENSATION

Classic Compensation:



$$P_1 = \frac{1}{(1 + g_{m2}R_2)C_e R_1}$$

$$P_2 = \frac{g_{m2}C_e}{C_L C_e + C_1 C_e + C_L C_1} \approx \frac{g_{m2}}{C_L}$$

$$Z_1 = \frac{1}{C_e \left(\frac{1}{g_{m2}} - R_2 \right)}$$

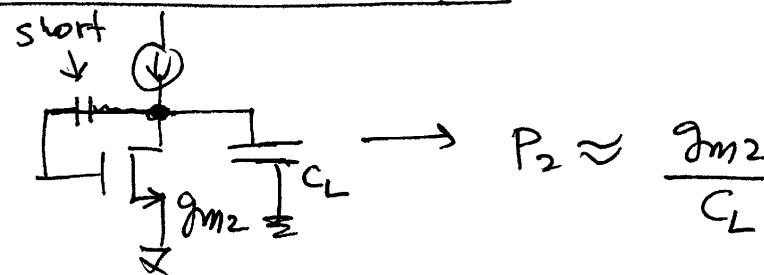
Two issues:

- 1) Cannot drive large loads.

To be stable : $C_L < \frac{g_{m2}C_e}{g_{m1}}$

- 2) Poor negative supply rejection.

Intuitive second pole location

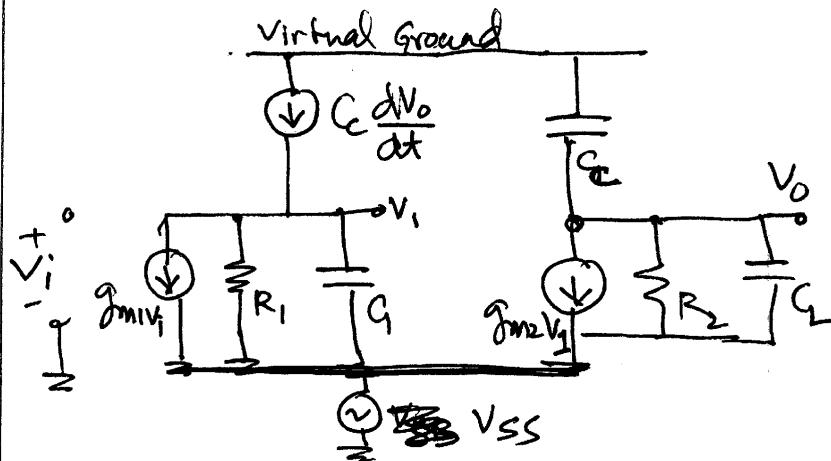


$$P_2 \approx \frac{g_{m2}}{C_L}$$

(2)

Improved Compensation:

Conceptual small signal:



No forward path
so no zero.

$$P_1 \approx \frac{1}{g_m R_2 C_c R_1}$$

$$P_2 \approx \frac{g_m C_c}{A_1 (C_c + C_L)}$$

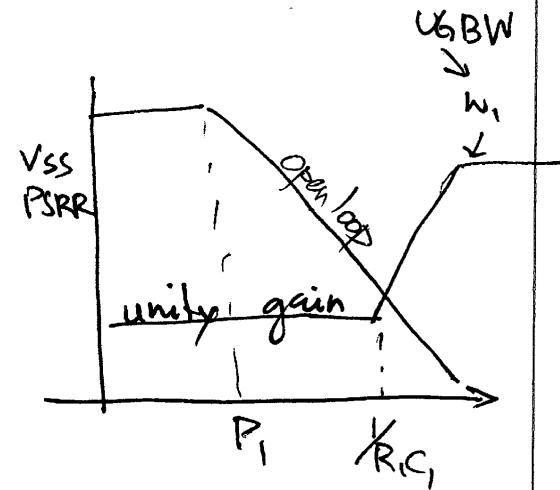
Negative PSRR

For open loop:

$$\frac{V_o}{V_{SS}} \approx \frac{1 + sC_1 R_1}{(1 + s/P_1)(1 + s/P_2)}$$

For unity gain mode:

$$\frac{V_o}{V_{SS}} \approx \frac{1 + sG R_1}{A_1 A_2 (1 + s/W_1)}$$

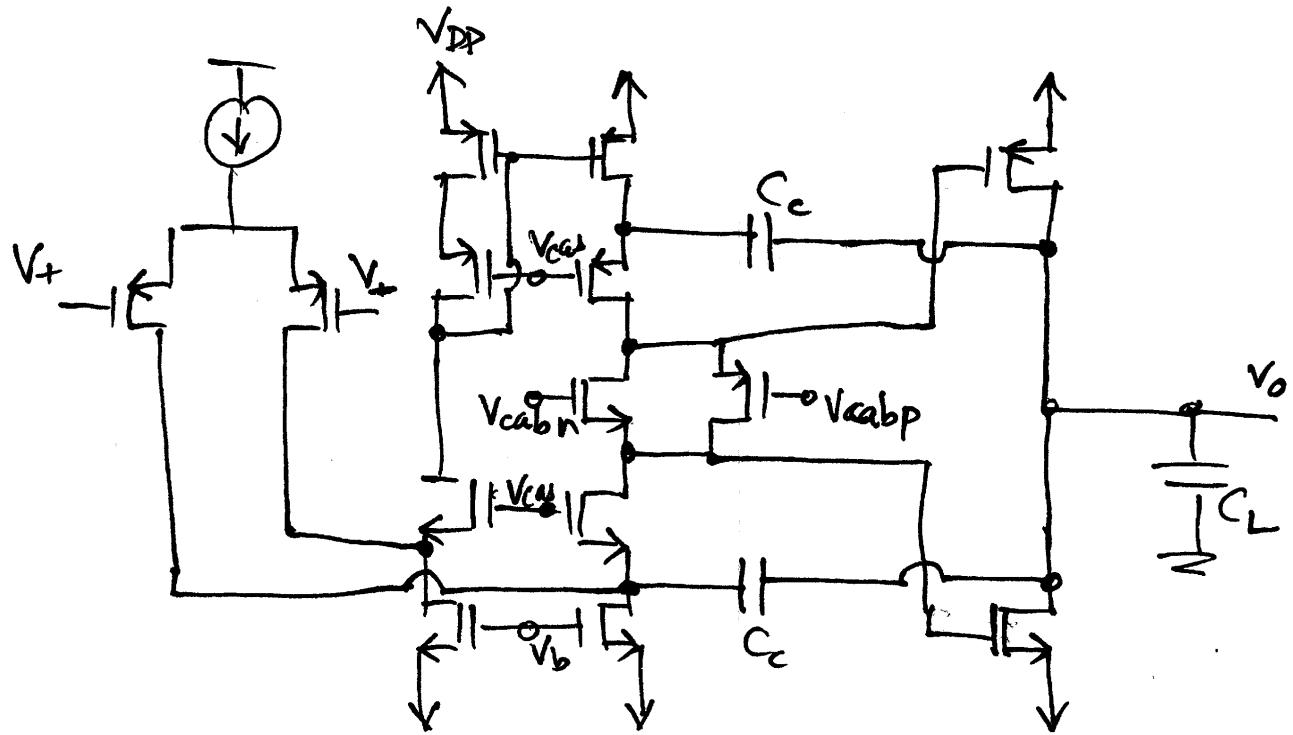


Observation:

- * P_1 is the same as before.
- * P_2 moves out $\approx \frac{g_m}{C_1}$
- * NO zeros
- * V_{SS} PSRR improves by pushing zero to $\frac{1}{R_1 C_1}$.

(3)

Implementation in Class AB Amplifier



Intuitive way to see the second pole:

Let's simplify by assuming a class A operation.
So the output stage looks like:

