Radar Signal Detection

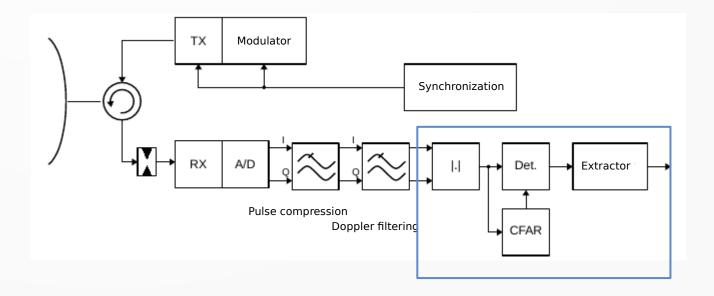
Radar systems (Radarové systémy)

Radar signal processing

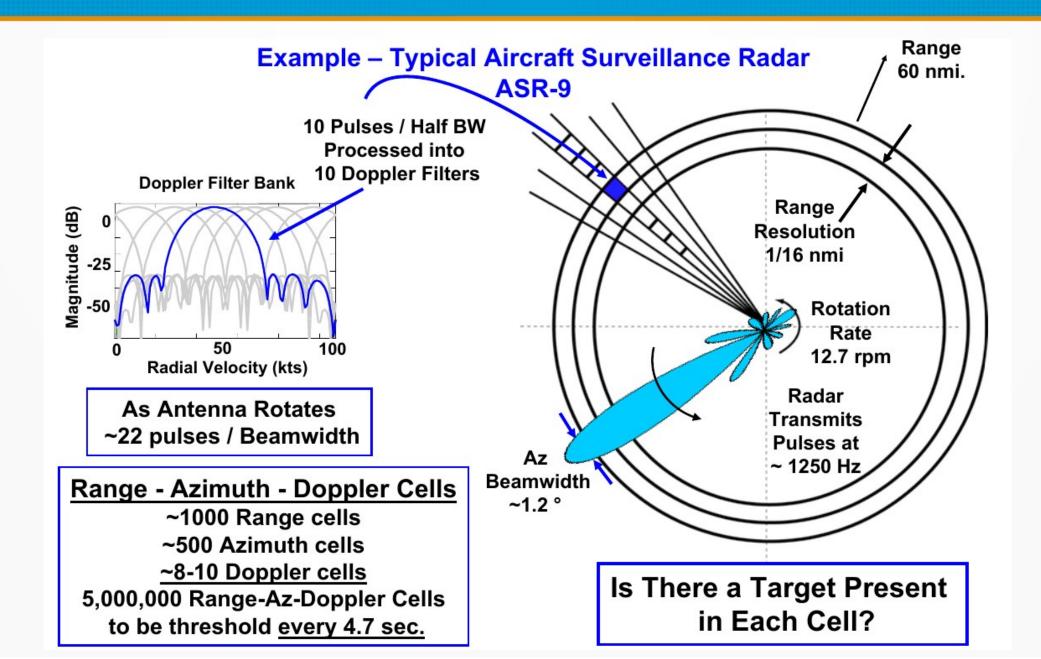
- Primary processing
 - Processing of signal within one antenna revolution one PRI or several adjacent PRIs
 - Pulse compression
 - Doppler filtration
 - Target detection
- Secondary processing
 - Processsing of several revolutions
 - Target tracking
 - Clutter mapping for adaptive processing
- Tertial processing
 - Data fusion signal processing from several radar sites

Radar signal processing

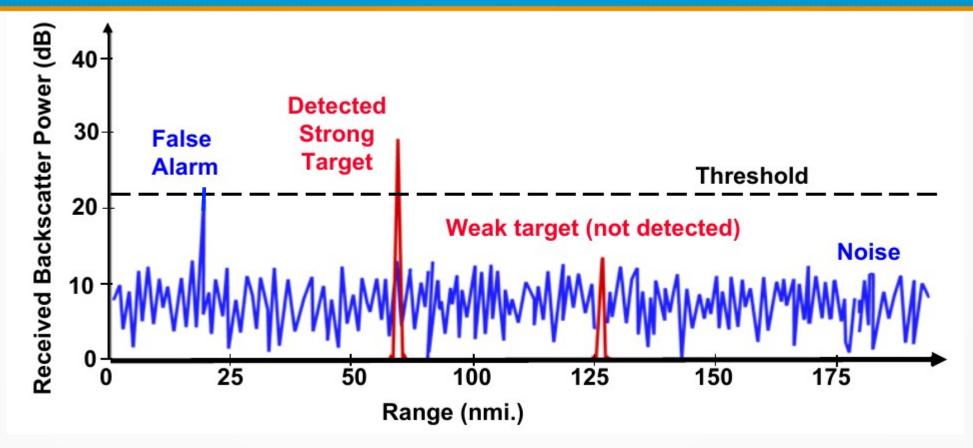
- Primary processing
 - Signal detection
 - Extraction



Range-Azimuth-Doppler Cells to Be Thresholded



Target Detection in Noise



- Received background noise and target echo fluctuate randomly → Both are random variables
- To decide if a target is present, at a given range, we need to set a threshold (constant or variable)
- Detection performance (**Probability of Detection**, P_D) depends of the strength of the target relative to that of the noise and the threshold setting (**Signal-To Noise Ratio** and **Probability of False Alarm**, P_{fa})

Signal Detection

- Decision about traget presence for each evaluated cell from measurement x
- Two hypotheses H₀, H₁
- Neyman-Paerson theorem/detector
- – Maximize P_D for chosen P_{fa}
- Based on information about characteristics of noise, clutter, signal

Statistics Theory	Radar Theory
Tested statistics $U(x)$ and threshold y	Detector
Zero hypothesis H_0	Target absent, Noise only $(x=n)$, PDF: $p(x H_0)$
Alternative hypothesis H_1	Target present, Signal+noise ($x=s+n$), PDF: $p(x H_1)$
1st case error (decision H_1 for real H_0)	False alarm, (P_{fa})
2nd case error (decision H_0 for real H_1)	Missed detection, $(1-P_D)$

Optimum Threshold Test

		Decision	
		H_{0}	H ₁
Reality	H_{0}	Do not report	False alarm
	H ₁	Missed detection	Detection

Probability of Detection:

 P_D Probability we choose H_1 for reality H_1

Probability of False alarm:

 P_{fa} Probability we choose H_1 for reality H_0

Neyman-Pearson criterion:

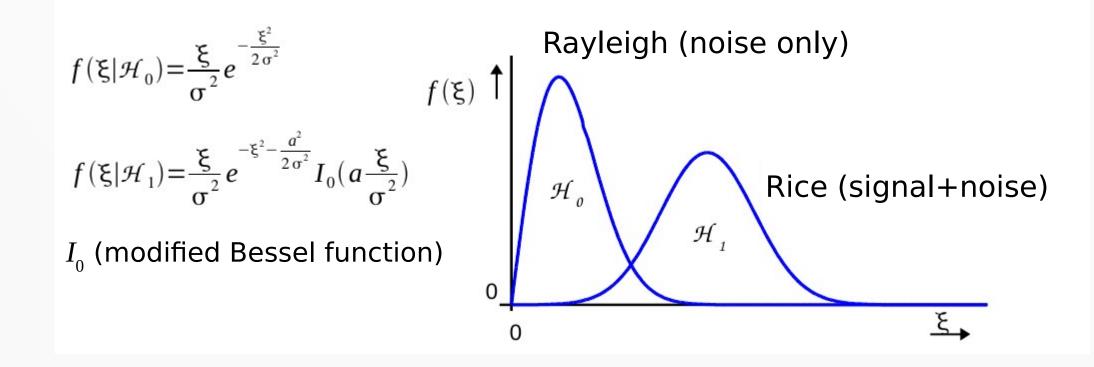
Maximize P_D for P_{fa} not greater than specified value

Likelihood function (ratio)

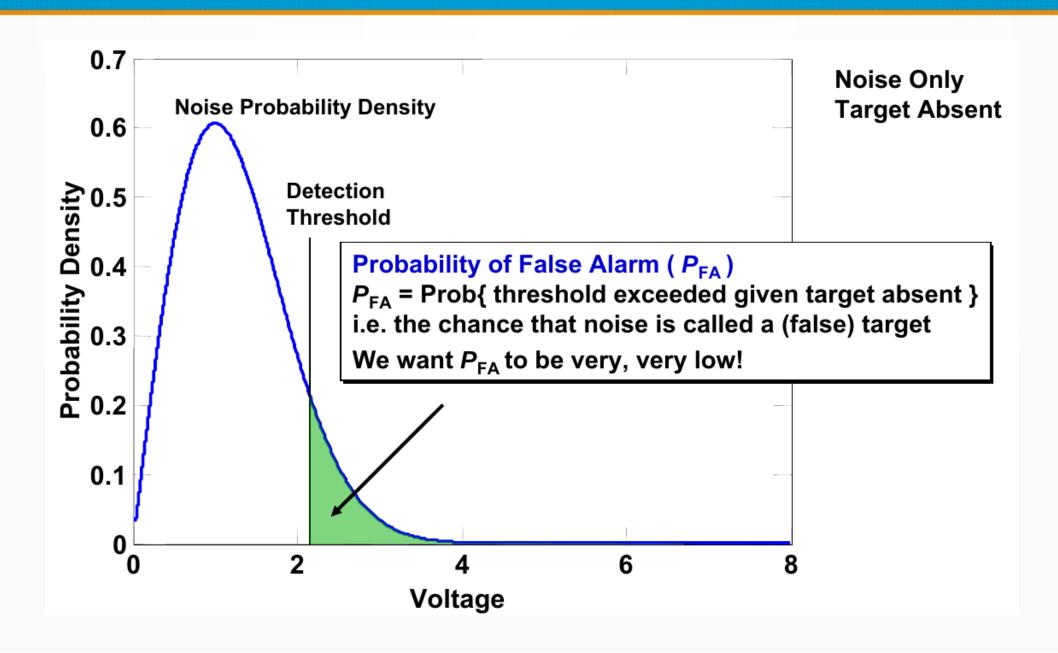
$$L(x) = \frac{p(x|H_1)}{p(x|H_0)}: \quad \stackrel{\geq T \to H_1}{< T \to H_0}$$
Threshold

Statistic description

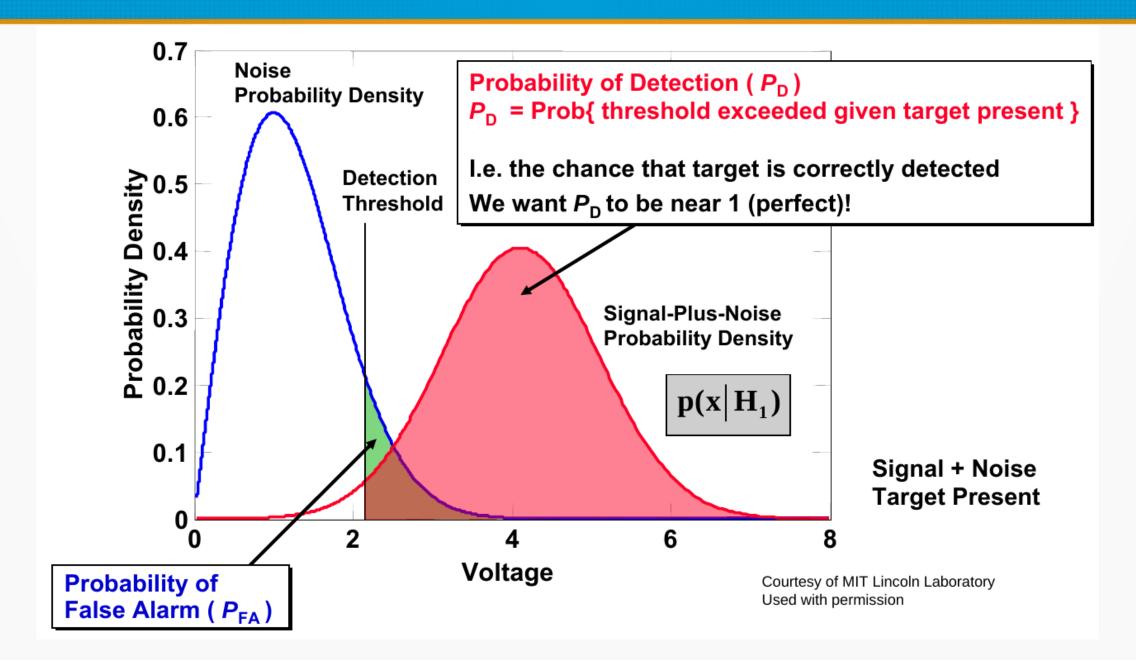
- Rayleigh distribution noise probability density function (e.g. voltages)
- Rice distribution signal+noise probability density function
- AWGN envelope and impacts of environment, non-fluctuating target



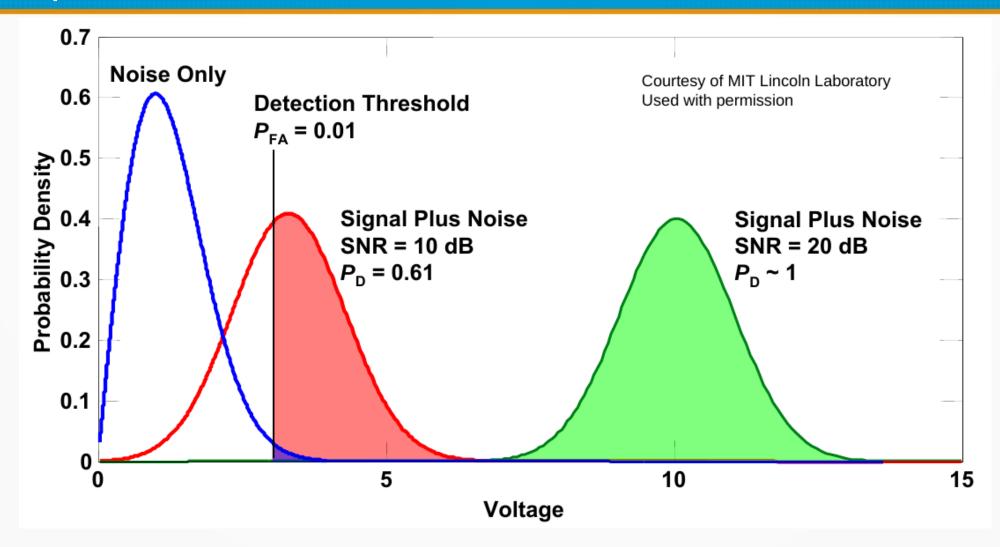
Detection statistics



Detection statistics



Impact of SNR



 P_{D} increases with SNR for given P_{fa} (threshold)

Non-fluctuating target detection

$$L(x) = \frac{P(x|\mathcal{H}_1)}{P(x|\mathcal{H}_0)}^{\mathcal{H}_1} > T$$

$$P_{fa} = Pr\{L(x) > T; \mathcal{H}_0\} = \int_{T}^{\infty} p(\xi|\mathcal{H}_0) d\xi$$

$$P_d = Pr\{L(x) > T; \mathcal{H}_1\} = \int_{T}^{\infty} p(\xi|\mathcal{H}_1) d\xi$$

$$P_{fa} = \int_{T}^{\infty} \frac{\xi}{\sigma^2} e^{\frac{-\xi^2}{2\sigma^2}} d\xi = e^{\frac{-T^2}{2\sigma^2}}$$

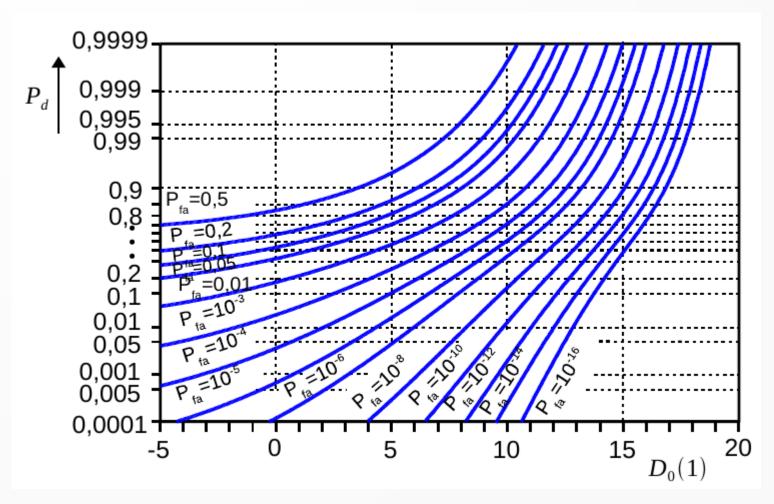
$$T = \sigma \sqrt{-2 \ln \left(P_{fa} \right)}$$

$$S/N = \frac{s^2}{\sigma^2}$$

$$P_{d} = \int_{T}^{\infty} \frac{\xi}{\sigma^{2}} e^{\frac{-\xi^{2} + s^{2}}{2\sigma^{2}}} I_{0}(\frac{\xi s}{\sigma^{2}}) d\xi$$

Detection quality

- System parameters for required detection parameters
- Detection coeficient (D₀(1) for pro SW0 non-fluctuating target, single pulse detection)
 - Necessary SNR to reach P_d for given P_{fa}
 - Usual conditions
 - Non-fluctuating
 - $P_d = 0.9$; $P_{fa} = 10^{-6}$
 - Fluctuating
 - $P_d = 0.8$; $P_{fa} = 10^{-6}$



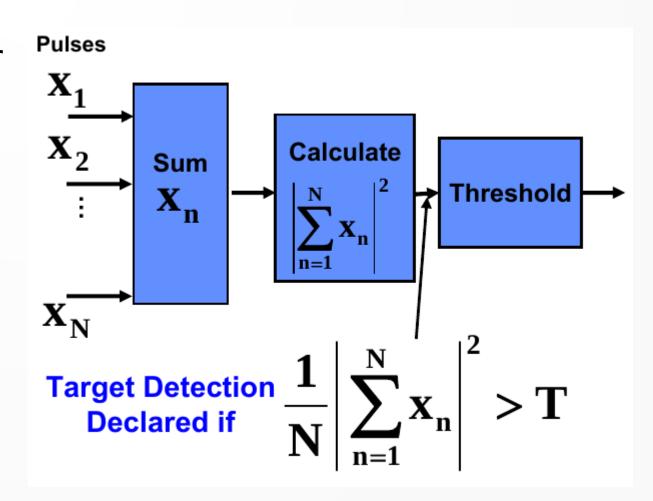
e.g.
$$P_d = 0.9$$
, $P_{fa} = 10^{-6}$, D_0 : SNR=13.2 dB

Improvement by integration

- Use multiple pulses from same target
- Impact of statistical dependence of samples
- Motivation
 - Increase Pd
 - Decrease Pfa
 - Decrease of necessary SNR
- Detection
 - Single pulse
 - Multiple pulses
 - Coherent integration
 - Non-coherent (envelope, video) integration
 - Binary Integration (M of N)
 - Cumulative detection (1 of N)

Coherent integration

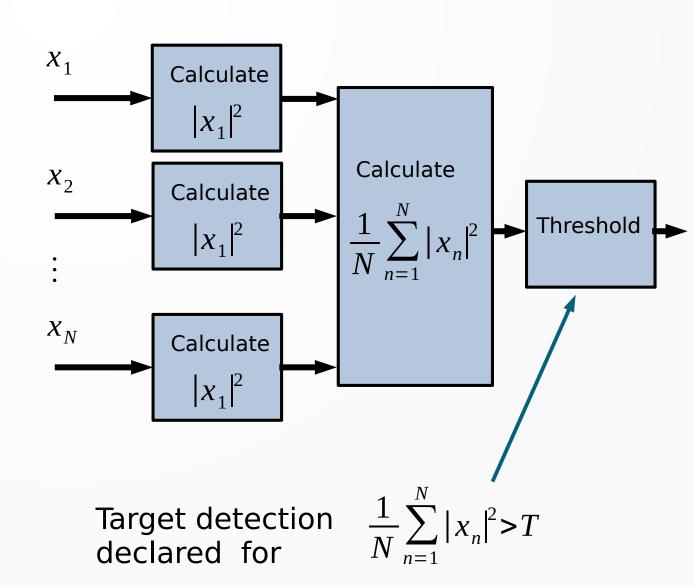
- Doppler compensation prior to integration (MTD)
- Integration (sum) od N pulses – samples of complex envelope
- Amplitude increase N-times
- Noise power (uncorrelated) increase by N-times
- SNR increase N-times
- Integration gain $G_i = N$



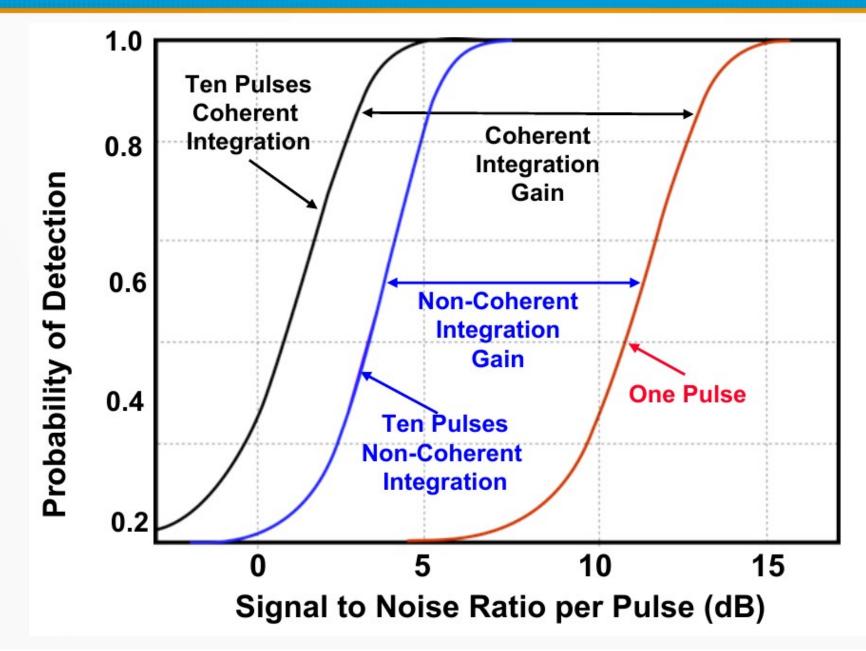
$$D_0(n) = \frac{D_0(1)}{G_i} = \frac{D_0(1)}{n}$$

Non-coherent Integration of envelope (video signal)

- Sum of amplitudes (powers)
- Ignores phase simplification
- Envelope detector
- Higher SNR compared to coherent case
- Integration loses w.r.t. ideal (coherent) integrator



Coherent vs. Non-coherent Integration



Steady Target

 $P_{FA} = 10^{-6}$

Integration of binary output (M of N)

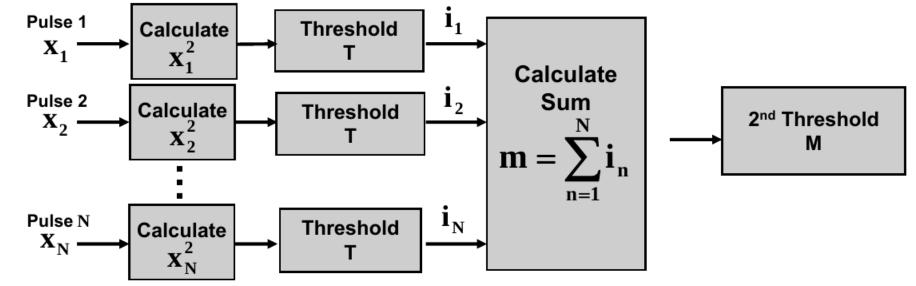
- Simplification of Detector realization
- Integration behind detector, count threshold binary outputs
- Secondary threshold for binary counts (M of N filter)
- First Detector can be set for higher P_{fa}

$$\begin{split} P_{d} &= \sum_{k=m}^{n} \frac{n!}{k!(n-k)!} P_{d1}^{k} (1 - P_{d1})^{n-k} \\ P_{fa} &= \sum_{k=m}^{n} \frac{n!}{k!(n-k)!} P_{fa1}^{k} (1 - P_{fa1})^{n-k} \\ P_{fa} &\approx \frac{n!}{m!(n-m)!} P_{fa1}^{m}; P_{fa1} \ll 1 \end{split}$$

Can be found optimal M for given N

$$m_{opt} \approx 1.5 \sqrt{n}$$

M of N integration



Individual pulse detectors:

$$\left| \mathbf{x}_{\mathbf{n}} \right|^{2} \ge \mathbf{T}, \quad \mathbf{i}_{\mathbf{n}} = \mathbf{1}$$
 $\left| \mathbf{x}_{\mathbf{n}} \right|^{2} < \mathbf{T}, \quad \mathbf{i}_{\mathbf{n}} = \mathbf{0}$

2nd thresholding:

 $m \geq M$, target present m < M , target absent

Target present if at least M detections in N pulses

Binary Integration

Cumulative Detection

$$\begin{array}{ll} \text{At Least} & \\ \text{M of N} & P_{M \, / \, N} = \sum_{k=M}^{N} \frac{N!}{k! \big(N\!-\!k\big)!} p^k \big(1\!-\!p\big)^{N-k} \end{array}$$
 Detections

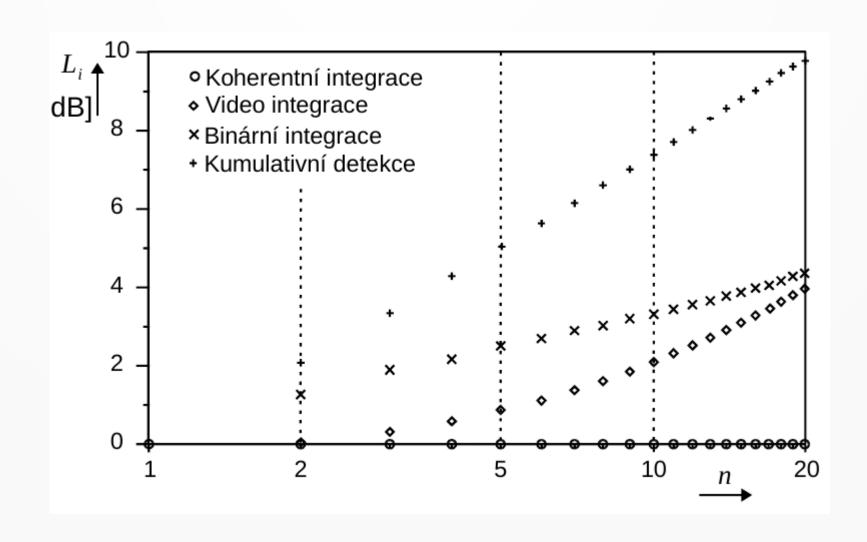
At Least
$$P_C = 1 - (1-p)^N$$

1 of N Detection – cummulative detection

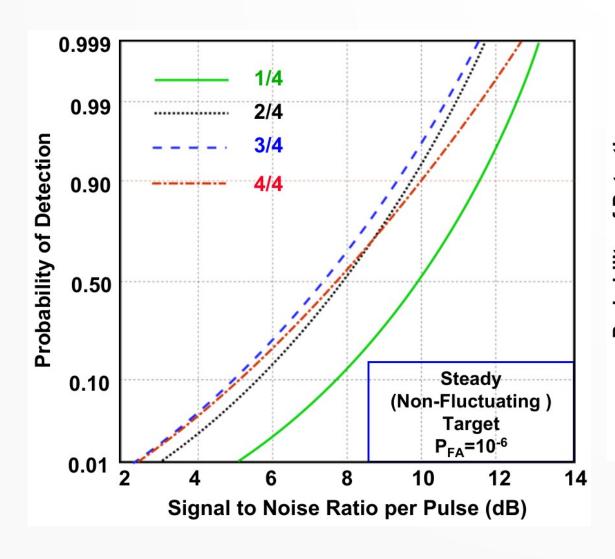
- N independent decisions for N pulses
- Detection positive if at least one of N decisions positive
- Higher requirements to P_{fa} setup in comparison to coherent and non-coherent integrations
- Little improvement of P_d compared to single pulse
 - → low integration gain

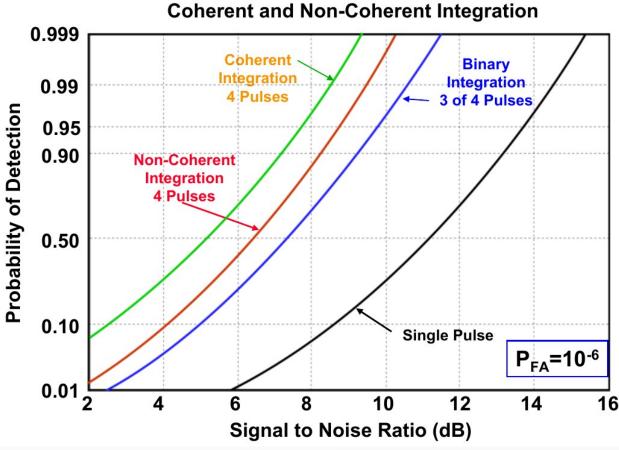
$$P_D = 1 - (1 - P_{d1})^N$$

Loss by integration



Detection comparison





Fluctuation of target (Fluctuation of RCS)

- Swerling models
- SW0 non-fluctuating (steady) target
- SW1 slow fluctuating target (scan to scan)
 - Rayleigh distribution of amplitudes, Exponential distribution of powers (RCS)
 - Mono-frequency radar observations example: Target consists from many comparable scatterers, no dominance
- SW2 fast fluctuating target (pulse to pulse)
 - SW1 target observed by agile radar
- SW3 slow fluctuating target (scan to scan)
 - Power distribution (RCS) χ2 with 4 degrees of freedom (DOF)
 - Mono-frequency radar observations example: Target consists from one dominant scatterer and many secondary scatterers
- SW4 fast fluctuating target (pulse to pulse)
 - SW3 target observed by agile radar

Swerling Target Models

Natura of	RCS	Fluctuation Rate	
Nature of Scattering	Model	Slow Fluctuation "Scan-to-Scan"	Fast Fluctuation "Pulse-to-Pulse"
Similar amplitudes	Exponential (Chi-Squared DOF=2) $p(\sigma) = \frac{1}{\overline{\sigma}} exp\left(-\frac{\sigma}{\overline{\sigma}}\right)$	Swerling I	Swerling II
One scatterer much Larger than others	(Chi-Squared DOF=4) $p(\sigma) = \frac{4 \sigma}{\overline{\sigma}^2} exp\left(-\frac{2 \sigma}{\overline{\sigma}}\right)$	Swerling III	Swerling IV

 $\overline{\sigma}$ = Average RCS (m²)

Swerling Target Models

Noture of	Amplitude	Fluctuation Rate	
Nature of Scattering	Model	Slow Fluctuation "Scan-to-Scan"	Fast Fluctuation "Pulse-to-Pulse"
Similar amplitudes	Rayleigh		
837	$p(a) = \frac{2a}{\overline{\sigma}} exp\left(-\frac{a^2}{\overline{\sigma}}\right)$	Swerling I	Swerling II
One scatterer much Larger than others	Central Rayleigh, DOF=4 $p(a) = \frac{8 a^{3}}{\overline{\sigma}^{2}} exp\left(-\frac{2 a^{2}}{\overline{\sigma}}\right)$	Swerling III	Swerling IV
	$\left[\frac{\mathbf{p}(\mathbf{a}) - \overline{\mathbf{\sigma}^2}}{\overline{\mathbf{\sigma}^2}}\right]$		

 $\overline{\sigma}$ = Average RCS (m²)

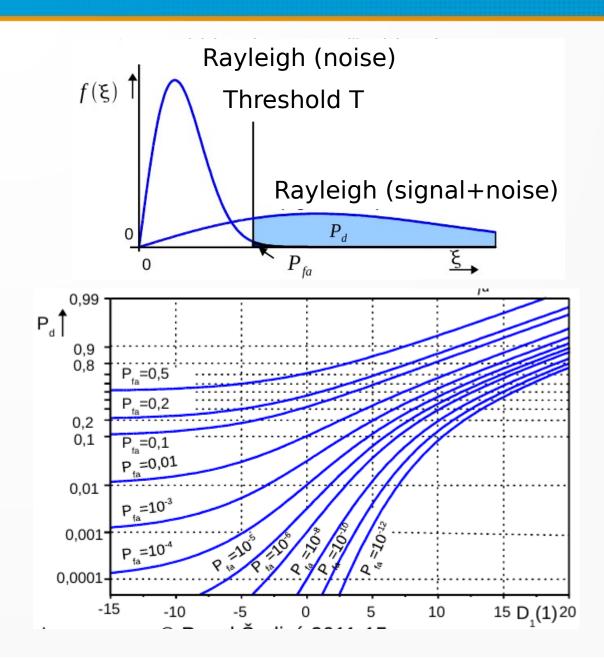
Impact of fluctuations to Probability of detection

 PSD of fluctuating target echo – Rayleigh (non-fluctuating - Rice)

$$f_{s}(\xi|\mathcal{H}_{1}) = \frac{\xi}{\sigma^{2} + \alpha^{2}} e^{\frac{-\xi^{2}}{2(\sigma^{2} + \alpha^{2})}}$$
$$P_{d} = e^{\frac{-T^{2}}{2(S+N)}} = e^{\frac{\ln(P_{fa})}{1+S/N}}$$

Detection cefficient

$$D_1 = \frac{\ln P_{fa}}{P_d} - 1$$

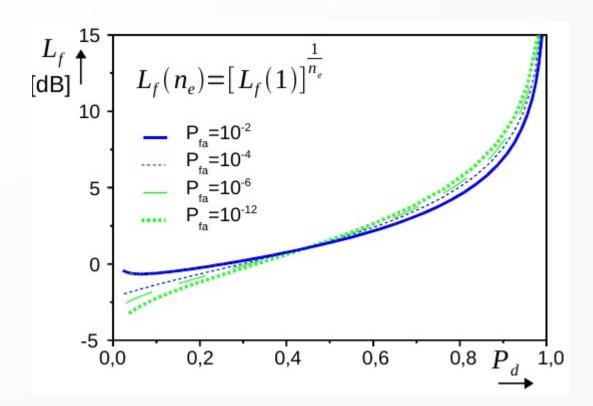


Fluctuation loss

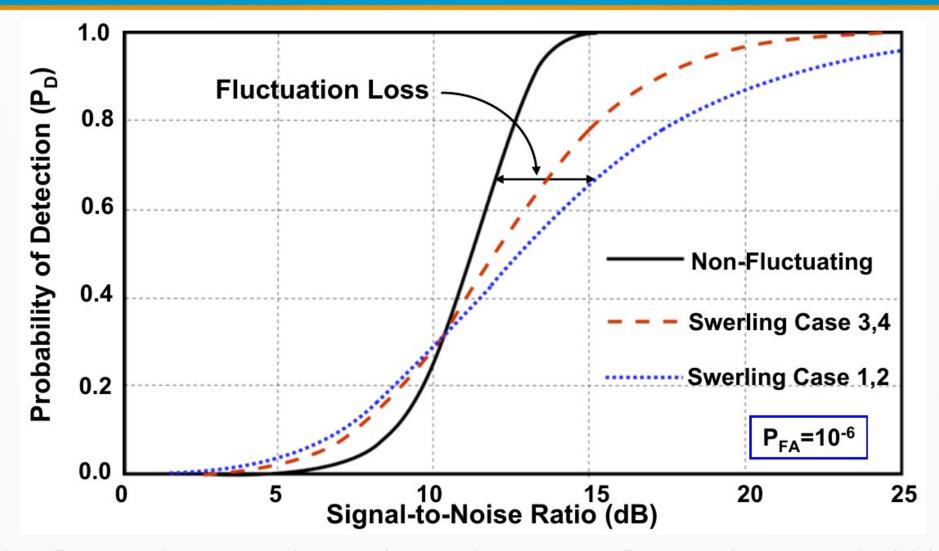
• SNR difference necessary to achieve same P_d for SW0 and (e.g.) SW1

$$L_f = \frac{D_1(1)}{D_0(1)}$$

- Integration of echoes from fluctuating targets
 - Dependent on mutual correlation of samples
- Diversity gain
 - time observation time is longer than correlation length of samples
 - frequency (detection on more frequencies)

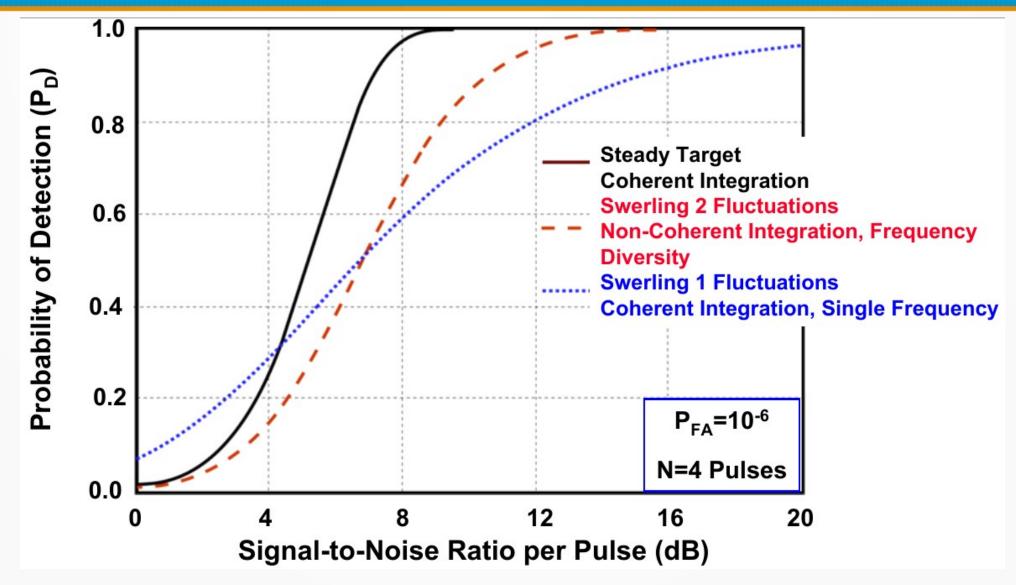


Fluctuation Loss



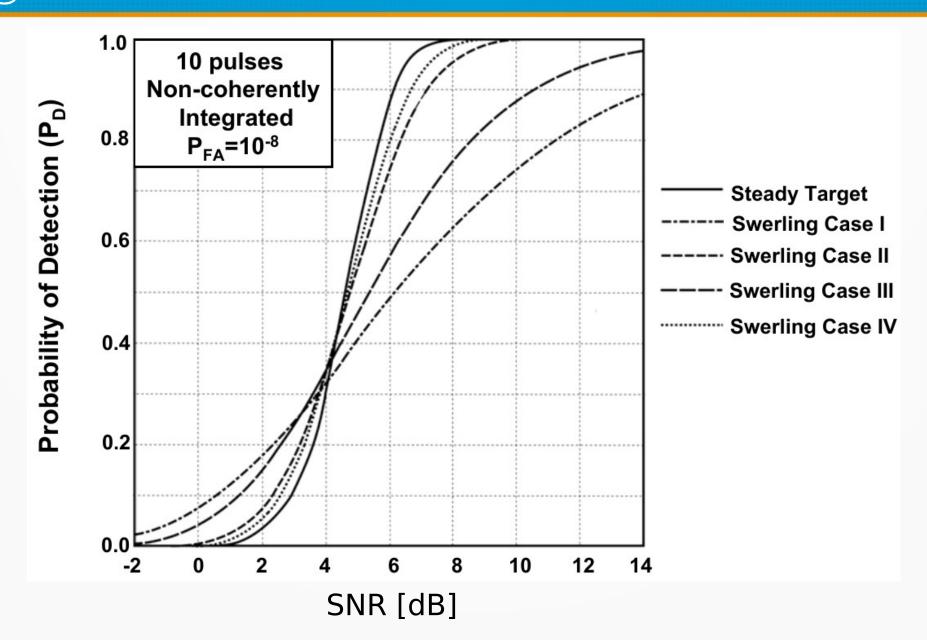
The fluctuation loss depends on the target fluctuations, probability of detection, and probability of false alarm.

Fluctuation Loss - Integration impact



For some targets (fast fluctuating), the non-coherent integration is better

Non-coherent integration for fluctuating targets



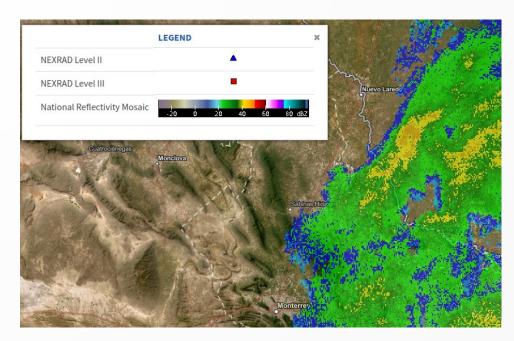
Other Fluctuation Models

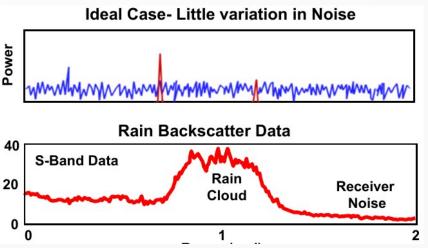
- Detection Statistics Calculations
 - Steady and Swerling 1,2,3,4 Targets in Gaussian Noise
 - Chi- Square Targets in Gaussian Noise
 - Log Normal Targets in Gaussian Noise
 - Steady Targets in Log Normal Noise
 - Log Normal Targets in Log Normal Noise
 - Weibel Targets in Gaussian Noise
- Chi Square, Log Normal and Weibel Distributions have long tails
 - One more parameter to specify distribution
 - Mean to median ratio for log normal distribution
- When used
 - Ground clutter Weibel
 - Sea Clutter Log Normal
 - HF noise Log Normal
 - Birds Log Normal

Adaptive detection - CFAR

Constant False Alarm Rate, CFAR

- Adaptive estimation of the threshold to keep consistent (constant) P_{fa}
- Problems
 - Clutter borders
 - Clutter residuals
 - More targets
 - Interference, jamming
 - High sensitivity to model accuracy (statistics)



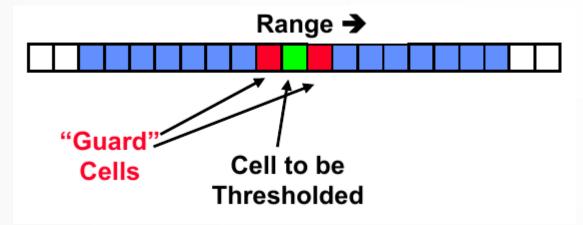


Adaptive detection - CFAR

- Realization
 - Parametric e.g. CA (Cell Averaging) CFAR for Rayleigh
 - Non-parametric ad hoc structures logCACFAR, time averaging, median-based detector
- Rayleigh model suitable for
 - Amplitude based AWGN detection
 - Water reflections (vertical polarization)
 - Terrain reflections

CFAR Window

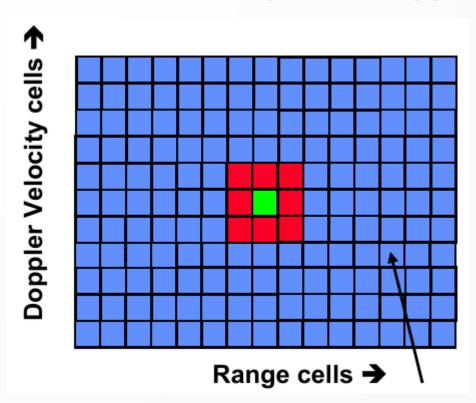
CFAR Window – Range cells



Estimate background mean from Range data only or Range/Doppler space data

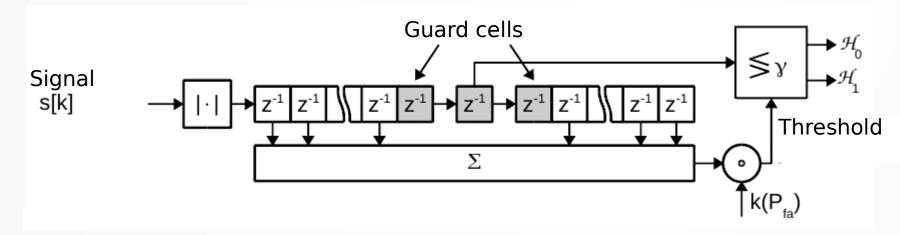
Set threshold as background mean multiplied by a constant (to adjust P_{fa})

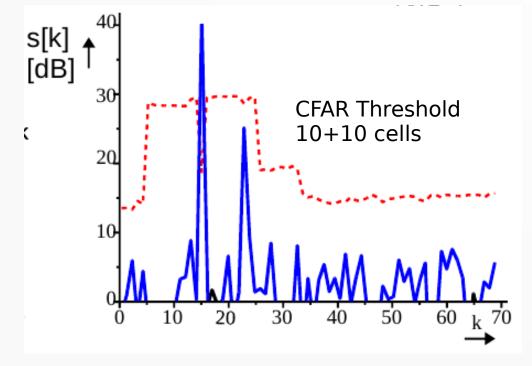
CFAR Window - Range and Doppler cells



Blue cells for estimation of background and threshold

Adaptive detection - CA CFAR

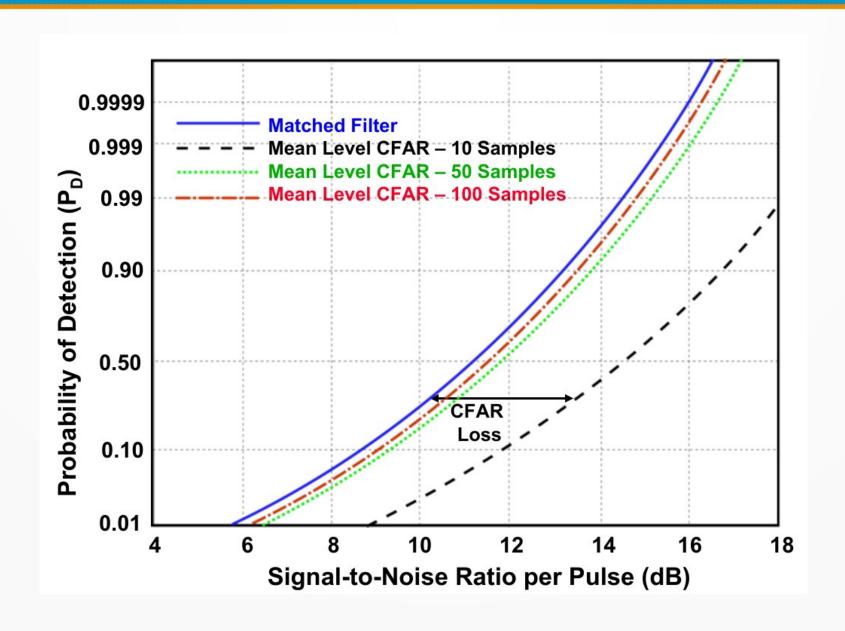


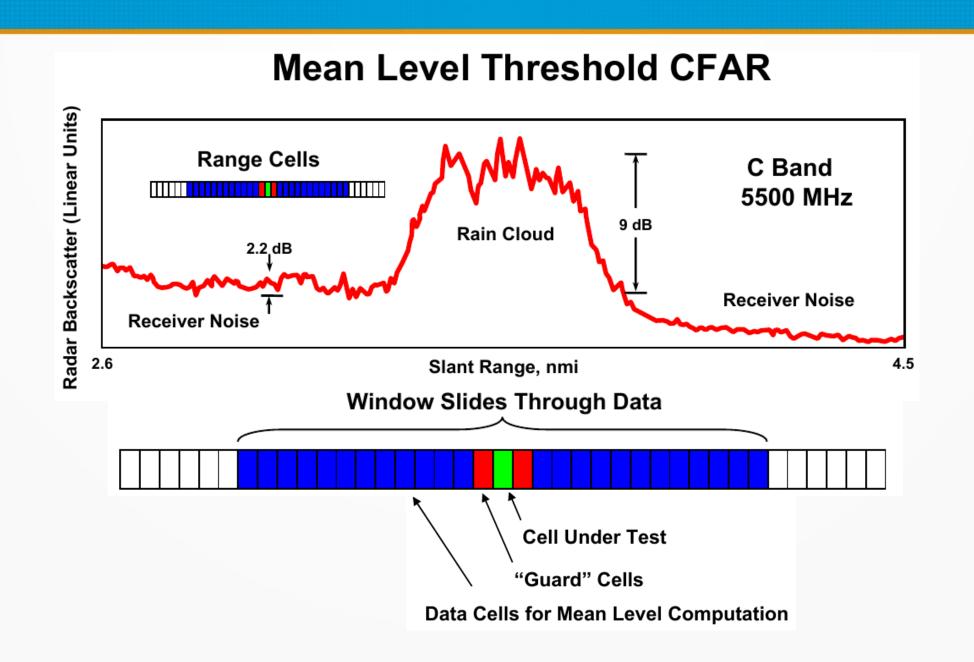


$$E[\xi] = \sigma \frac{\sqrt{\pi}}{2} \qquad P_{fa} = \int_{T}^{\infty} \frac{\xi}{\sigma^{2}} e^{\frac{-\xi^{2}}{2\sigma^{2}}} d\xi = e^{\frac{-T^{2}}{2\sigma^{2}}}$$

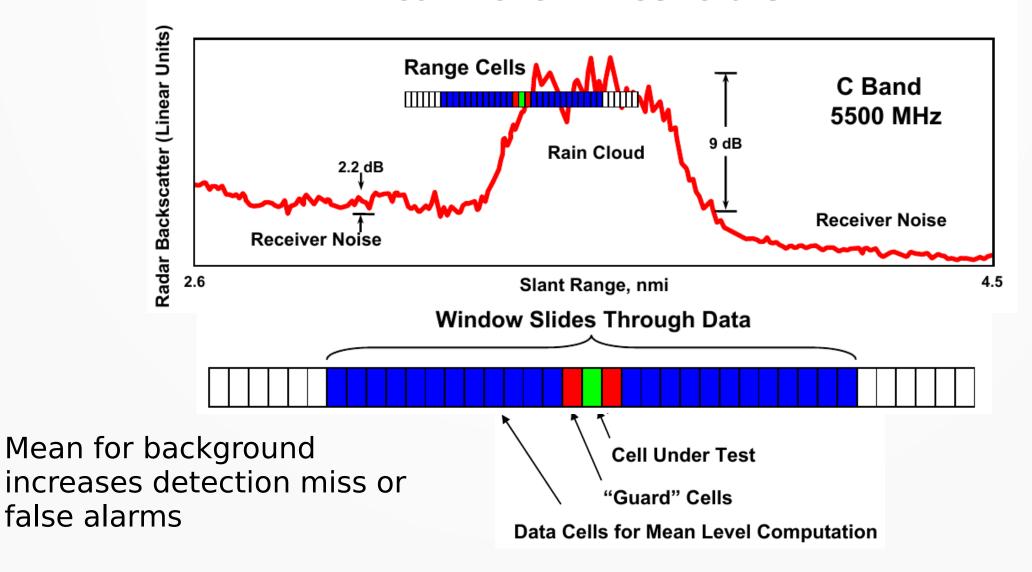
$$T = \sigma \sqrt{-2\ln(P_{fa})} = E[\xi] \sqrt{\frac{-4\ln(P_{fa})}{\pi}}$$

CA CFAR – impact of number of samples (cells)



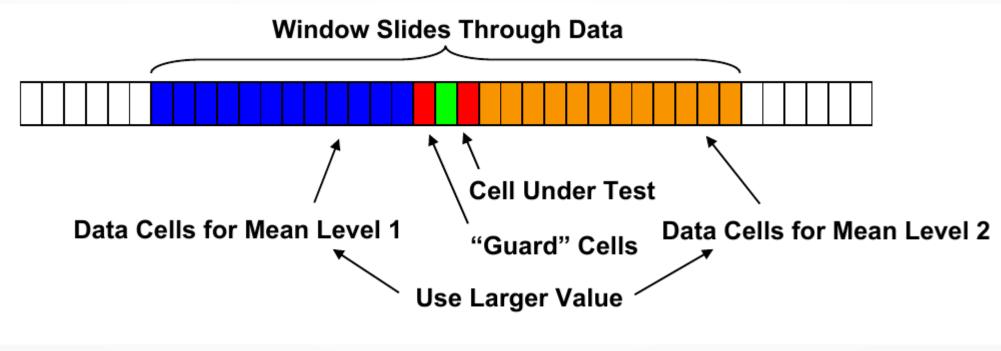


Mean Level Threshold CFAR



Greatest of Means CFAR

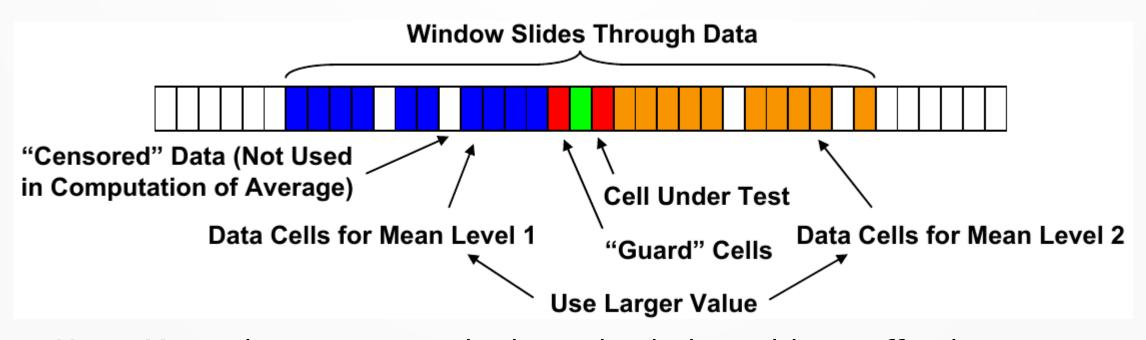
Compare mean value of N/2 cells before and after test cell and use larger value to determine threshold



- Helps reduce false alarms near sharp clutter or interference boundaries
- Nearby targets still raise threshold and suppress detection

Censored Greatest of Means CFAR

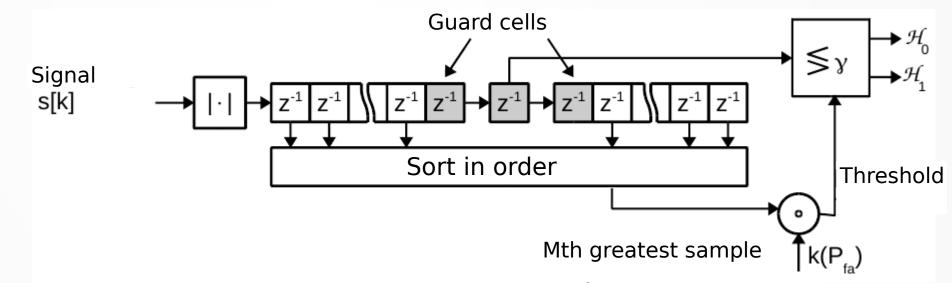
Compute and use noise estimates as in Greatest-of, but remove the largest M samples before computing each average

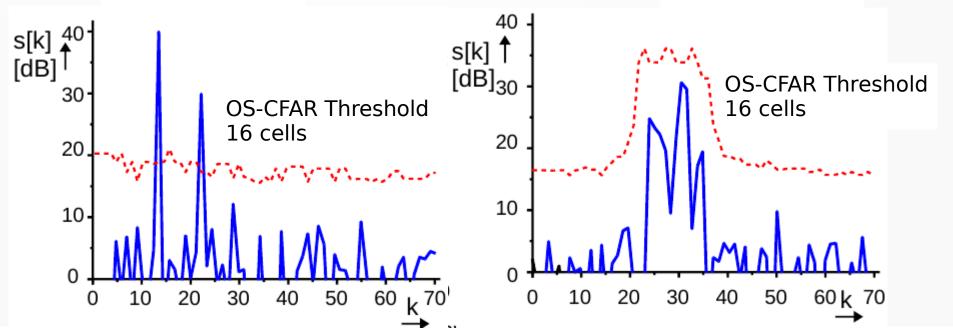


- Up to M nearby targets can be in each window without affecting threshold – dependent on target statistics
- Ordering the samples from each window is more computationally expensive than plain averaging

Adaptive detection - OS CFAR

- Ordered statistics (OS)
- Better in case of multiple targets or at clutter border
- Lower gain





References

- P. Šedivý Rádiové systémy, lectures, CTU FEE Prague 2011-2016
- M. O'Donnell Introduction to Radar Systems, MIT Lincoln Laboratory, set of lectures, 2002
- M. Skolnik Radar Handbook, McGraw-Hill 2008
- M. Richards Fundamentals of Radar Signal Processing, McGraw-Hill, New York, 2005

Thank You!