

# **Automatic Control**

**Equilibrium solution of dynamical systems**  
**Linearization of nonlinear dynamical systems**

# Equilibrium solution of dynamical systems

# Equilibrium solution of dynamical systems

A constant state  $\bar{x}$  is an **equilibrium solution (or equilibrium state)** of the nonlinear dynamical system

$$\dot{x}(t) = f(x(t), u(t)), \quad y(t) = g(x(t), u(t))$$

if, in the presence of the constant input  $u(t) = \bar{u}$  and the initial condition  $x(0) = \bar{x}$ , it results

$$x(t) = \bar{x} \quad \forall t \geq 0$$

- the input  $\bar{u}$  is said **equilibrium input**
- the couple  $(\bar{x}, \bar{u})$  is said **equilibrium point**
- the output  $\bar{y} = g(\bar{x}, \bar{u})$  is said **equilibrium output**

An equilibrium point  $(\bar{x}, \bar{u})$  of the dynamical system

$$\dot{x}(t) = f(x(t), u(t))$$

satisfies the equilibrium condition:

$$f(\bar{x}, \bar{u}) = 0$$

In fact:

$$x(t) = \bar{x}, \forall t \geq 0 \Rightarrow \dot{x}(t) = \dot{\bar{x}} = f(\bar{x}, \bar{u}) = 0, \forall t \geq 0$$

For LTI systems

$$\dot{x}(t) = Ax(t) + Bu(t)$$

the equilibrium condition

$$f(\bar{x}, \bar{u}) = 0$$

becomes

$$A\bar{x} + B\bar{u} = 0$$

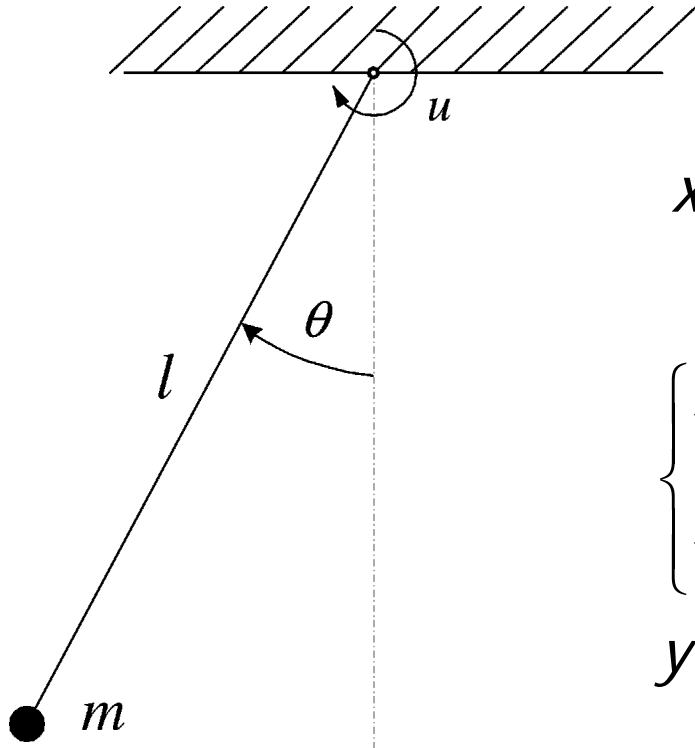
$$B\bar{u} \in \text{Im}(A\bar{x})$$

$$\Rightarrow \bar{x} \stackrel{\uparrow}{=} -A^{-1}B\bar{u}$$

if  $\det(A) \neq 0$



## Example: pendulum



$$x(t) = \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} \theta(t) \\ \dot{\theta}(t) \end{bmatrix}$$

$$\begin{cases} \dot{x}_1(t) = x_2(t) \\ \dot{x}_2(t) = -\frac{g}{l} \sin(x_1(t)) - \frac{\beta}{ml^2} x_2(t) + \frac{1}{ml^2} u(t) \end{cases}$$

$$y(t) = x_1(t)$$



## Example: pendulum

$$\begin{cases} \dot{x}_1(t) = x_2(t) \\ \dot{x}_2(t) = -\frac{g}{l} \sin(x_1(t)) - \frac{\beta}{ml^2} x_2(t) + \frac{1}{ml^2} u(t) \end{cases}$$

Find all the equilibrium states with respect to the equilibrium input  $\bar{u} = 0$

By using the equilibrium condition  $\dot{x}(t) = \dot{\bar{x}} = 0 = f(\bar{x}, \bar{u})$ , we get

$$\begin{cases} 0 = \bar{x}_2 \\ 0 = -\frac{g}{l} \sin \bar{x}_1 - \frac{\beta \bar{x}_2}{ml^2} + \frac{\bar{u}}{ml^2} = -\frac{g}{l} \sin \bar{x}_1 \end{cases} \Rightarrow \begin{cases} \sin \bar{x}_1 = 0 \\ \bar{x}_2 = 0 \end{cases}$$



- Since at the equilibrium the following relations hold

$$\begin{cases} \sin \bar{x}_1 = 0 \\ \bar{x}_2 = 0 \end{cases} \Rightarrow \begin{cases} \bar{x}_1 = k\pi, \quad k = 0, \pm 1, \dots \\ \bar{x}_2 = 0 \end{cases}$$

there exist infinite equilibrium points of the form:

$$x = \begin{bmatrix} \bar{x}_1 \\ \bar{x}_2 \end{bmatrix} = \begin{bmatrix} k\pi \\ 0 \end{bmatrix}, \quad k = 0, \pm 1, \dots$$

- At the equilibrium,

$$y(t) = \bar{y} = g(\bar{x}, \bar{u}), \quad \forall t \geq 0$$

$$\bar{y} = \bar{x}_1 = k\pi, \quad k = 0, \pm 1, \dots$$





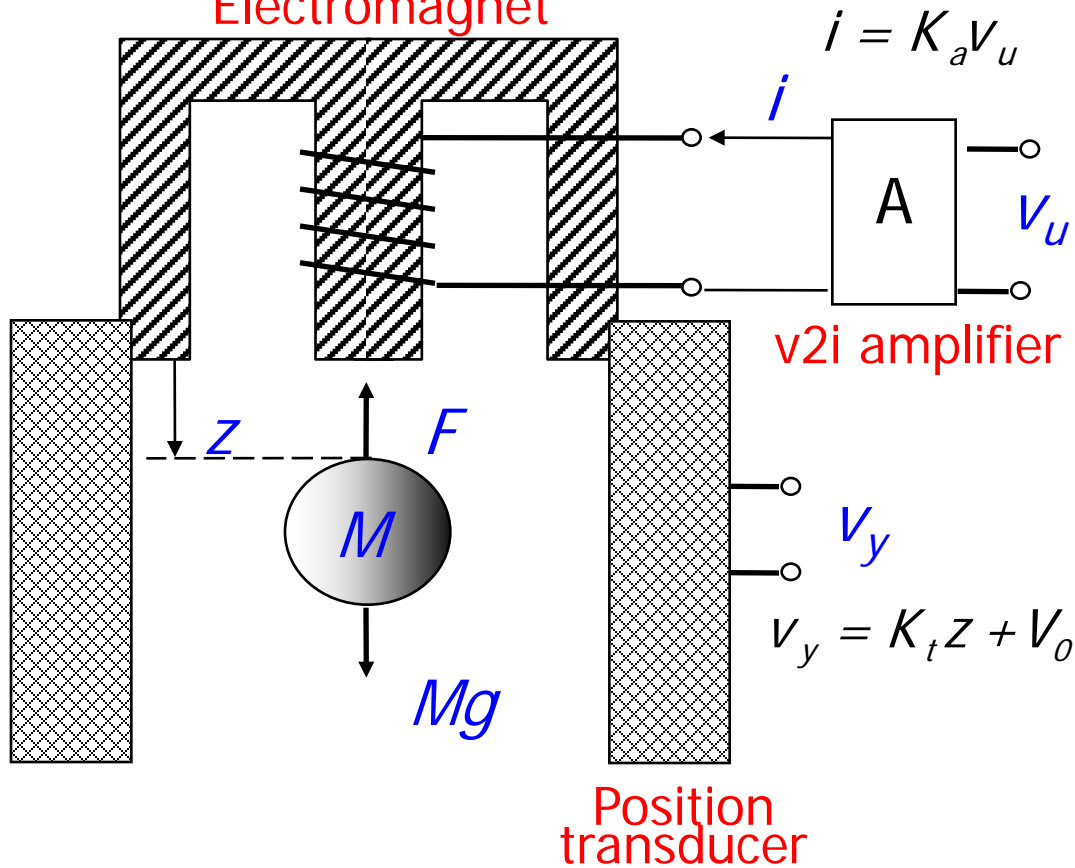
# Example: magnetic levitator





# Example: magnetic levitator

Electromagnet



$z \rightarrow$  ball position

$v_y \rightarrow$  position transducer  
voltage (output)

$v_u \rightarrow$  amplifier input  
voltage (input)

$$\underbrace{M \ddot{z} = Mg - F}_{\text{2nd Newton's law}}$$

$$\underbrace{F = \frac{B_m i^2}{z^2}}_{\text{electromagnetic force}}$$



## Example: magnetic levitator

State space representation

$$\begin{aligned} M \ddot{z} &= Mg - F = Mg - \frac{B_m i^2}{z^2} = \\ &= Mg - \frac{B_m (K_a v_u)^2}{z^2} \\ \ddot{z} &= g - \frac{K_1 v_u^2}{z^2}, K_1 = \frac{B_m K_a^2}{M} \end{aligned}$$

$$x(t) = \begin{bmatrix} z(t) \\ \dot{z}(t) \end{bmatrix} = \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix}$$

$$u(t) = v_u(t)$$

$$y(t) = v_y(t) = K_t z(t) + V_0$$

$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = g - K_1 \frac{u^2}{x_1^2} \\ y = K_2 x_1 + K_3, K_2 = K_t, K_3 = V_0 \end{cases}$$



## Example: magnetic levitator

Compute all the equilibrium points with respect to the input  $\bar{u} \neq 0$

By using the equilibrium condition,

$$\dot{x}(t) = \dot{\bar{x}} = 0 = f(\bar{x}, \bar{u})$$

$$\begin{cases} 0 = \bar{x}_2 \\ 0 = g - (K_1) \bar{u}^2 / \bar{x}_1^2 \end{cases} \Rightarrow \begin{cases} \bar{x}_1 = \sqrt{\frac{K_1}{g}} |\bar{u}| \\ \bar{x}_2 = 0 \end{cases} \Rightarrow \bar{x} = \begin{bmatrix} \sqrt{\frac{K_1}{g}} |\bar{u}| \\ 0 \end{bmatrix}$$

$$\bar{y} = K_2 \bar{x}_1 + K_3 = K_2 \sqrt{\frac{K_1}{g}} |\bar{u}| + K_3$$

# **Equilibrium and nonlinear electronic devices analysis**

## Motivating example: nonlinear electronic devices

In order to establish proper operating conditions in electronic components, biasing networks are usually used.

Many electronic devices that process time-varying signals (AC), also require steady-state current or steady-state voltage (DC) to operate correctly.

The AC signal is superimposed on DC bias current or voltage.

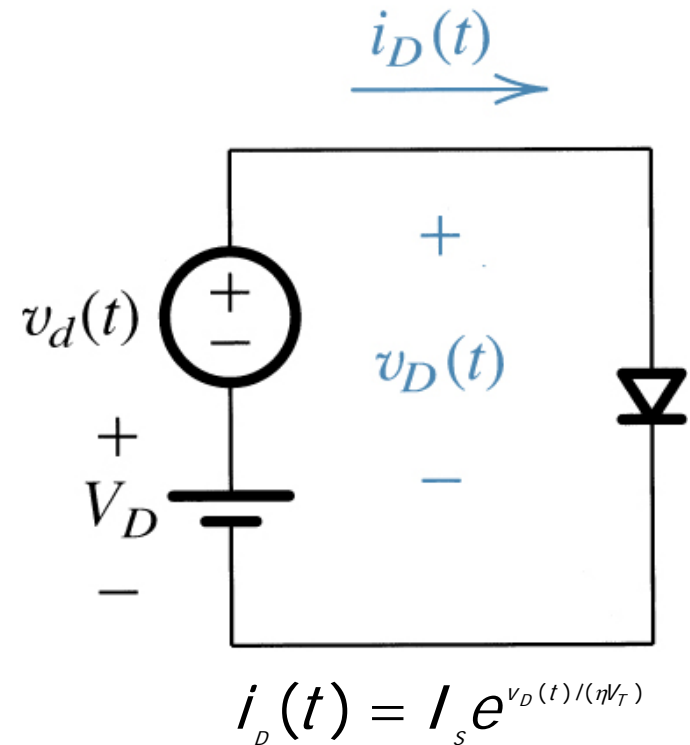
The operating point of a device is also known as bias point, quiescent point, or Q-point.

The operating point (OP) of a device is the steady-state voltage or steady-state current at a specified terminal of an active device with no input AC signal applied.

OP can be seen as an equilibrium point of the device

The operating point (also known as bias point), is determined by the steady-state voltage  $V_D$ .

The AC signal superimposed on DC bias voltage is the time-varying voltage  $v_d(t)$ .



# Example: diode model - small signal approximation

We may be interested in looking for a model in terms of a resistor whose value is the reciprocal of the slope of the " $i$ - $v$ " curve.

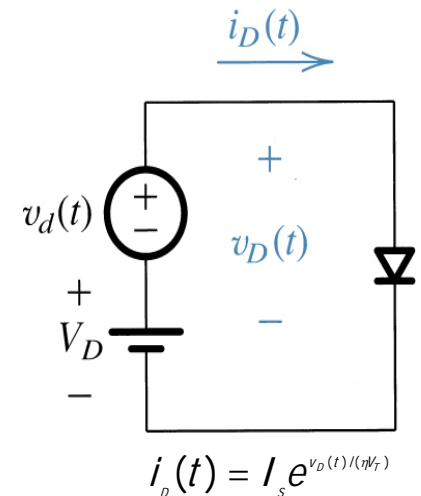
We can do this by means of a linearization around bias point  $V_D$ , where the time varying signal  $v_d(t)$  is superimposed.

Current  $I_D$  due to  $V_D$  is, approximatively, given by:

$$i_D(t) = I_s e^{v_D(t)/(\eta V_T)} \rightarrow I_D = I_s e^{V_D/(\eta V_T)}$$

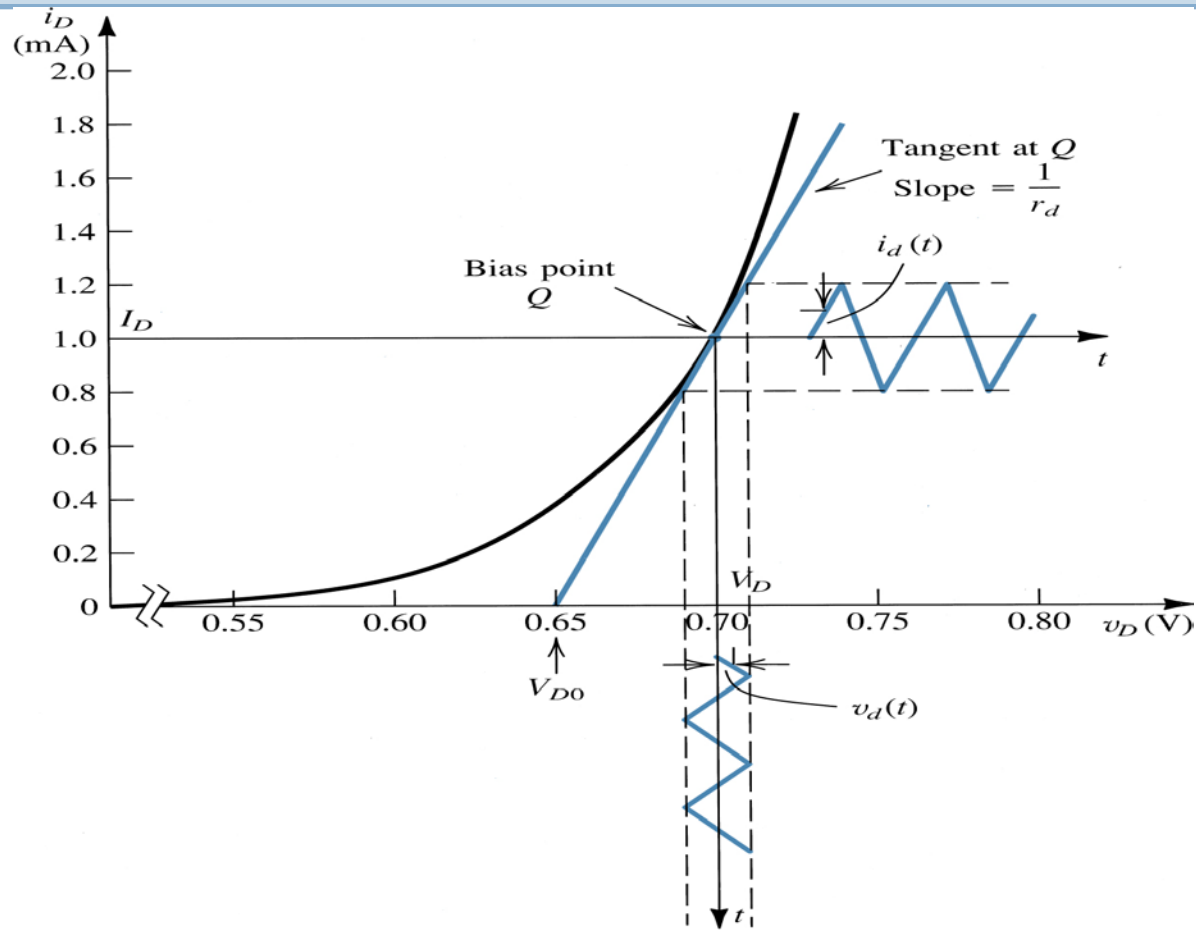
$$\begin{aligned} v_D(t) &= V_D + v_d(t) \Rightarrow i_D(t) = I_s e^{(V_D + v_d(t))/(\eta V_T)} = \\ &= I_s e^{V_D/(\eta V_T)} \cdot e^{v_d(t)/(\eta V_T)} = I_D \cdot e^{v_d(t)/(\eta V_T)} \end{aligned}$$

$$\frac{v_d(t)}{\eta V_T} \ll 1 \rightarrow e^{v_d(t)/(\eta V_T)} \underset{\substack{\uparrow \\ \text{Taylor} \\ \text{polynomial} \\ \text{degree 1}}}{\approx} 1 + \frac{v_d(t)}{\eta V_T} \rightarrow i_D(t) \approx I_D \cdot \left( 1 + \frac{v_d(t)}{\eta V_T} \right)$$





# Example: diode model - small signal approximation



If the AC signal  $v_d(t)$  is "small", the diode current  $i_D(t)$  can be approximated as the sum of the DC bias signal and the superimposed AC signal

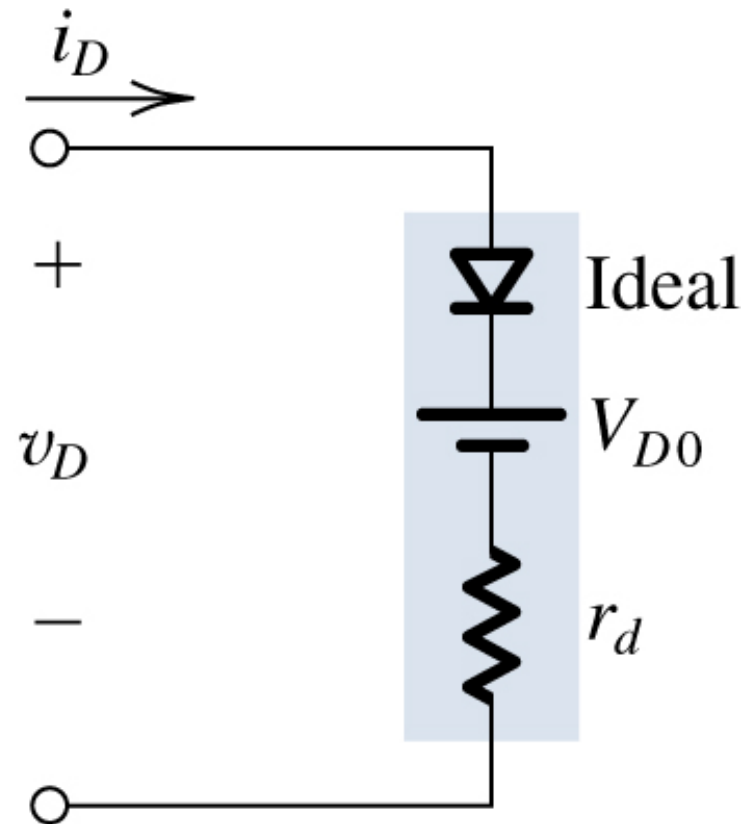
## Example: diode model - small signal approximation

For small changes around the bias point  $Q$ , the equivalent diode model for small variation is given by:

$$r_d = 1 / \left[ \frac{\partial i_D}{\partial v_D} \right]_{i_D = I_D}$$

Where  $V_{D0}$  is the intercept of the tangent on the  $v_D$  axis

$$i_D(t) = \frac{1}{r_d} (v_D(t) - V_{D0})$$



# Linearization of dynamical nonlinear systems

# Linearization of a nonlinear system

Real world systems are, in general, nonlinear

Anyway, their behavior can be approximated in the neighbourhood of a given solution (e.g. an equilibrium solution) through suitable linear models known as **linearized models**

Objective: compute a linear dynamical model able to accurately approximate the behavior of a nonlinear system in the neighbourhood of an equilibrium solution

Idea: use a similar procedure as done in scalar function approximation by means of Taylor series expansions

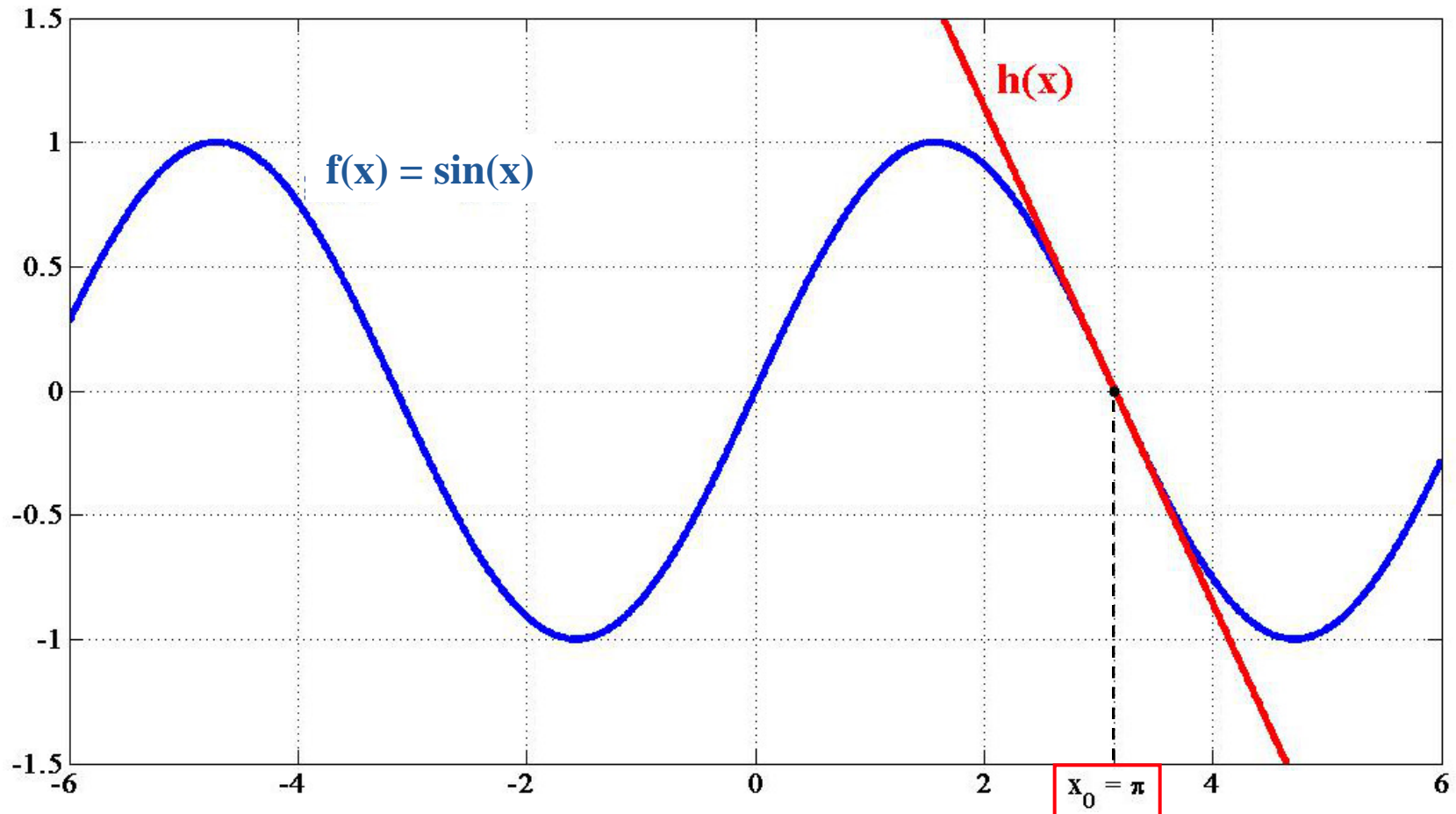
A function  $f(x): \mathbb{R} \rightarrow \mathbb{R}$  can be approximated in a small neighborhood of  $x_0 \in \text{dom}(f(x))$  with radius  $\delta x$ , by a Taylor series defined as:

$$\begin{aligned} f(x) &= f(x_0 + \delta x) = \\ &= f(x_0) + \left. \frac{df(x)}{dx} \right|_{x=x_0} \delta x + \frac{1}{2!} \left. \frac{d^2 f(x)}{dx^2} \right|_{x=x_0} \delta x^2 + \dots \end{aligned}$$

The linear approximation of  $f(x): \mathbb{R} \rightarrow \mathbb{R}$  in a small neighborhood of  $x_0$  is the first order truncation of the Taylor series expansion

$$f(x) = f(x_0 + \delta x) \cong f(x_0) + \left. \frac{df(x)}{dx} \right|_{x=x_0} \delta x = h(x)$$

# Function linearization



The smaller is the radius of the neighborhood of  $x_0$ , the better is the approximation accuracy provided by the linear function  $h(x)$

# Linearization of a nonlinear system: motivation

Problem: given the nonlinear dynamical system:

$$\begin{aligned}\dot{x}(t) &= f(x(t), u(t)) \\ y(t) &= g(x(t), u(t))\end{aligned}$$

compute the solution in the presence of an input of the form:

$$u(t) = \bar{u} + \delta u(t), \quad \|\delta u(t)\| \ll \|\bar{u}\|$$

and of the following initial condition

$$x(0) = \bar{x} + \delta x(0), \quad \|\delta x(0)\| \ll \|\bar{x}\|$$

where  $(\bar{x}, \bar{u})$  is a known equilibrium point of the considered system

# Linearization of a nonlinear system: motivation

In order to compute the required solution, the following Cauchy problem has to be solved:

$$\dot{x}(t) = f(x(t), u(t)), u(t) = \bar{u} + \delta u(t), x(0) = \bar{x} + \delta x(0)$$

Indeed, the solution of a nonlinear Cauchy problem<sup>(\*)</sup> is, in general, a hard task (the superposition principle does not hold)

A viable approach is to compute an approximation  $\tilde{x}(t)$  of the solution  $x(t)$  as:

$$\tilde{x}(t) = \bar{x} + \delta x(t) \approx x(t)$$

(\*) The considered Cauchy problem admits a unique solution if  $f(x, u)$  is Lipschitz continuous in both  $x$  and  $u$ .



# Linearization of a nonlinear system: preliminaries

$$\tilde{x}(t) = \bar{x} + \delta x(t) \approx x(t)$$

In the solution we are looking for, the term  $\delta x(t)$  is obtained solving a suitable LTI system of the form

$$\delta \dot{x}(t) = A \delta x(t) + B \delta u(t)$$

in the presence of the input  $\delta u(t)$  and of the initial condition  $\delta x(0)$

Such a linear system will be derived through a linearization procedure applied at the nonlinear state description:

$$\dot{x}(t) = f(x(t), u(t))$$

$$y(t) = g(x(t), u(t))$$

in the neighbourhood of the equilibrium point  $(\bar{x}, \bar{u})$

→ **linearized system**

# Definition of system perturbations

In order to compute the **linearized system**, we need to define the concept of perturbation of a system wrt an equilibrium solution

$$\delta x(t) = x(t) - \bar{x} = \textbf{state perturbation} \in \mathbb{R}^n$$

$$\Rightarrow x(t) = \bar{x} + \delta x(t)$$

$$\delta u(t) = u(t) - \bar{u} = \textbf{input perturbation} \in \mathbb{R}^p$$

$$\Rightarrow u(t) = \bar{u} + \delta u(t)$$

$$\delta y(t) = y(t) - \bar{y} = \textbf{output perturbation} \in \mathbb{R}^q$$

$$\Rightarrow y(t) = \bar{y}(t) + \delta y(t)$$

Next step: formally derive the differential equation which describes the dynamic behavior of the state perturbation  $\delta x(t)$

# Computing the system perturbation

The dynamic behavior of the state perturbation  $\delta x(t)$  is governed by the differential equation:

$$\delta \dot{x}(t) = \dot{x}(t) - \dot{\bar{x}} = f(x(t), u(t)) - 0 = f(x(t), u(t))$$

Supposing that the vector valued function  $f(x(t), u(t))$  can be expanded in Taylor series in a neighbourhood of  $(\bar{x}, \bar{u})^{(*)}$ , we have:

$$\begin{aligned} f(x(t), u(t)) &= f(\bar{x} + \delta x(t), \bar{u} + \delta u(t)) = \\ &= f(\bar{x}, \bar{u}) + \frac{\partial f(x, u)}{\partial x} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} (x(t) - \bar{x}) + \frac{\partial f(x, u)}{\partial u} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} (u(t) - \bar{u}) + \dots = \\ &\stackrel{\substack{= \\ \uparrow \\ f(\bar{x}, \bar{u})=0}}{=} \frac{\partial f(x, u)}{\partial x} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} \delta x(t) + \frac{\partial f(x, u)}{\partial u} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} \delta u(t) + \dots \end{aligned}$$

(\*) This is true if  $f(x, u)$  is Lipschitz continuous in both  $x$  and  $u$ .

## Computing the system perturbation

...  $f(x(t), u(t))$  can be then approximated by its Taylor expansion truncated at the linear term:

$$f(x(t), u(t)) \cong \left. \frac{\partial f(x, u)}{\partial x} \right|_{\substack{x=\tilde{x} \\ u=\tilde{u}}} \delta x(t) + \left. \frac{\partial f(x, u)}{\partial u} \right|_{\substack{x=\tilde{x} \\ u=\tilde{u}}} \delta u(t)$$

Therefore, the approximate behavior of the state perturbation  $\delta x(t)$  can be computed through the solution to the following linear differential equation

$$\begin{aligned} \delta \dot{x}(t) &\cong \left. \frac{\partial f(x, u)}{\partial x} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}} \delta x(t) + \left. \frac{\partial f(x, u)}{\partial u} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}} \delta u(t) = \\ &= A \delta x(t) + B \delta u(t) \end{aligned}$$

# Computing the system perturbation

Note that:

$$A = \frac{\partial f(x, u)}{\partial x} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \left[ \begin{array}{ccc} \partial f_1 / \partial x_1 & \cdots & \partial f_1 / \partial x_n \\ \vdots & \ddots & \vdots \\ \partial f_n / \partial x_1 & \cdots & \partial f_n / \partial x_n \end{array} \right] \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} \in \mathbb{R}^{n \times n} : \begin{array}{l} \text{Jacobian of } f \\ \text{wrt } x \end{array}$$

$$B = \frac{\partial f(x, u)}{\partial u} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \left[ \begin{array}{ccc} \partial f_1 / \partial u_1 & \cdots & \partial f_1 / \partial u_p \\ \vdots & \ddots & \vdots \\ \partial f_n / \partial u_1 & \cdots & \partial f_n / \partial u_p \end{array} \right] \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} \in \mathbb{R}^{n \times p} : \begin{array}{l} \text{Jacobian of } f \\ \text{wrt } u \end{array}$$

# Computing the system perturbation

A similar procedure can be applied for the computation of  $\delta y(t)$ :

$$\begin{aligned}\delta y(t) &= y(t) - \bar{y} = g(x(t), u(t)) - g(\bar{x}, \bar{u}) \cong \\ &\cong \left. \frac{\partial g(x, u)}{\partial x} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}} \delta x(t) + \left. \frac{\partial g(x, u)}{\partial u} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}} \delta u(t) = \\ &= C \delta x(t) + D \delta u(t)\end{aligned}$$

$$C = \left. \frac{\partial g(x, u)}{\partial x} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \left[ \begin{array}{ccc} \partial g_1 / \partial x_1 & \cdots & \partial g_1 / \partial x_n \\ \vdots & \ddots & \vdots \\ \partial g_q / \partial x_1 & \cdots & \partial g_q / \partial x_n \end{array} \right] \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} \in \mathbb{R}^{q \times n} : \text{Jacobian of } g \text{ wrt } x$$

$$D = \left. \frac{\partial g(x, u)}{\partial u} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \left[ \begin{array}{ccc} \partial g_1 / \partial u_1 & \cdots & \partial g_1 / \partial u_p \\ \vdots & \ddots & \vdots \\ \partial g_q / \partial u_1 & \cdots & \partial g_q / \partial u_p \end{array} \right] \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} \in \mathbb{R}^{q \times p} : \text{Jacobian of } g \text{ wrt } u$$

The linearized system is then given by:

$$\begin{aligned}\delta \dot{x}(t) &= A \delta x(t) + B \delta u(t), & \delta x(0) &= x(0) - \bar{x} \\ \delta y(t) &= C \delta x(t) + D \delta u(t), & \delta u(t) &= u(t) - \bar{u}\end{aligned}$$

$$A = \left. \frac{\partial f(x, u)}{\partial x} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}}, B = \left. \frac{\partial f(x, u)}{\partial u} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}}, C = \left. \frac{\partial g(x, u)}{\partial x} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}}, D = \left. \frac{\partial g(x, u)}{\partial u} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}}$$



## Example: pendulum

### Linearization

$$\begin{cases} \dot{x}_1 = x_2 & = f_1(x, u) \\ \dot{x}_2 = -\frac{g}{l} \sin x_1 - \frac{\beta x_2}{Ml^2} + \frac{u}{Ml^2} & = f_2(x, u) \\ y = x_1 & = g(x, u) \end{cases} \quad \left( \bar{x} = \begin{bmatrix} k\pi \\ 0 \end{bmatrix}, \bar{u} = 0 \right)$$

$$A = \frac{\partial f(x, u)}{\partial x} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \left[ \begin{array}{cc} \frac{\partial f_1}{\partial x_1} & \frac{\partial f_1}{\partial x_2} \\ \frac{\partial f_2}{\partial x_1} & \frac{\partial f_2}{\partial x_2} \end{array} \right] \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} 0 & 1 \\ -\frac{g}{l} \cos \bar{x}_1 & -\frac{\beta}{Ml^2} \end{bmatrix}$$

$$k \text{ even} \Rightarrow A = \begin{bmatrix} 0 & 1 \\ -\frac{g}{l} & -\frac{\beta}{Ml^2} \end{bmatrix}, \quad k \text{ odd} \Rightarrow A = \begin{bmatrix} 0 & 1 \\ \frac{g}{l} & -\frac{\beta}{Ml^2} \end{bmatrix}$$





## Example: pendulum

### Linearization

$$\begin{cases} \dot{x}_1 = x_2 & = f_1(x, u) \\ \dot{x}_2 = -\frac{g}{l} \sin x_1 - \frac{\beta x_2}{Ml^2} + \frac{u}{Ml^2} & = f_2(x, u) \\ y = x_1 & = g(x, u) \end{cases}$$

$$\left( \bar{x} = \begin{bmatrix} k\pi \\ 0 \end{bmatrix}, \bar{u} = 0 \right)$$

$$B = \frac{\partial f(x, u)}{\partial u} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} \frac{\partial f_1}{\partial u} \\ \frac{\partial f_2}{\partial u} \end{bmatrix} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} 0 \\ \frac{1}{Ml^2} \end{bmatrix}$$

$$C = \frac{\partial g(x, u)}{\partial x} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} \frac{\partial g}{\partial x_1} & \frac{\partial g}{\partial x_2} \end{bmatrix} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} 1 & 0 \end{bmatrix} \quad D = \frac{\partial g(x, u)}{\partial u} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} \frac{\partial g}{\partial u} \end{bmatrix} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} 0 \end{bmatrix}$$



## Example: magnetic levitator

### Linearization

$$\begin{cases} \dot{x}_1 = x_2 = f_1(x, u) \\ \dot{x}_2 = g - K_1 \frac{u^2}{x_1^2} = f_2(x, u) \\ y = K_2 x_1 + K_3 = g(x, u) \end{cases}$$

$$\left( \bar{x} = \begin{bmatrix} \sqrt{\frac{K_1}{g}} |\bar{u}| \\ 0 \end{bmatrix}, \bar{u} \neq 0 \right)$$

$$A = \frac{\partial f(x, u)}{\partial x} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} \frac{\partial f_1}{\partial x_1} & \frac{\partial f_1}{\partial x_2} \\ \frac{\partial f_2}{\partial x_1} & \frac{\partial f_2}{\partial x_2} \end{bmatrix} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} 0 & 1 \\ \frac{2K_1 \bar{u}^2}{\bar{x}_1^3} & 0 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ \frac{2g}{|\bar{u}|} \sqrt{\frac{g}{K_1}} & 0 \end{bmatrix}$$

$$B = \frac{\partial f(x, u)}{\partial u} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} \frac{\partial f_1}{\partial u} \\ \frac{\partial f_2}{\partial u} \end{bmatrix} \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} 0 \\ -\frac{2K_1 |\bar{u}|}{\bar{x}_1^2} \end{bmatrix} = \begin{bmatrix} 0 \\ -\frac{2g}{|\bar{u}|} \end{bmatrix}$$



## Example: magnetic levitator

Linearization

$$\begin{cases} \dot{x}_1 = x_2 = f_1(x, u) \\ \dot{x}_2 = g - K_1 \frac{u^2}{x_1^2} = f_2(x, u) \\ y = K_2 x_1 + K_3 = g(x, u) \end{cases}$$

$$\left( \bar{x} = \begin{bmatrix} \sqrt{\frac{K_1}{g}} |\bar{u}| \\ 0 \end{bmatrix}, \bar{u} \neq 0 \right)$$

$$C = \left. \frac{\partial g(x, u)}{\partial x} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \left[ \frac{\partial g}{\partial x_1} \quad \frac{\partial g}{\partial x_2} \right] \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \begin{bmatrix} K_2 & 0 \end{bmatrix}$$

$$D = \left. \frac{\partial g(x, u)}{\partial u} \right|_{\substack{x=\bar{x} \\ u=\bar{u}}} = \left[ \frac{\partial g}{\partial u} \right] \bigg|_{\substack{x=\bar{x} \\ u=\bar{u}}} = [0]$$

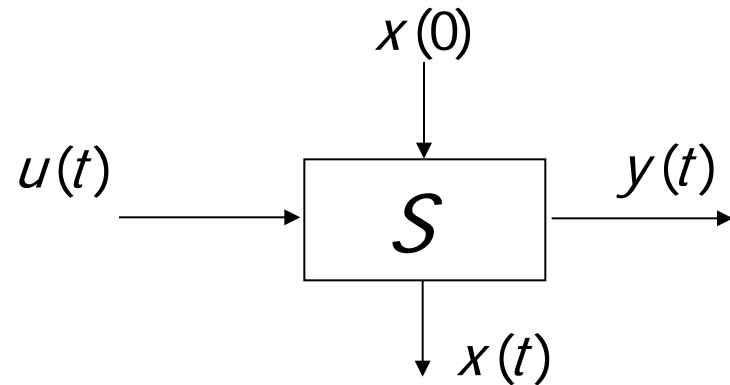
# Resume via block diagram representation

Nonlinear system:

$$S : \begin{cases} \dot{x}(t) = f(x(t), u(t)) \\ y(t) = g(x(t), u(t)) \end{cases}$$

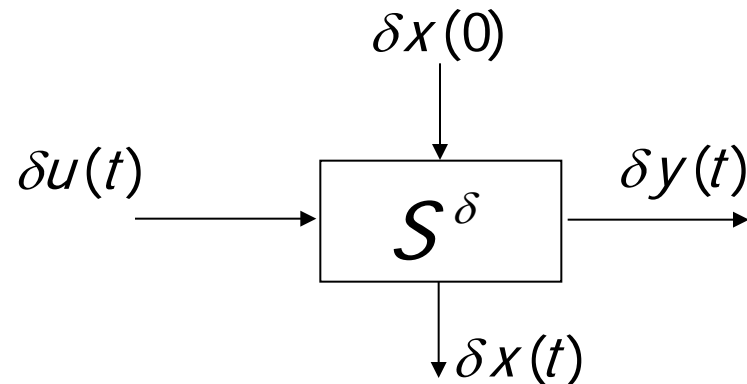
$$u(t) = \bar{u} + \delta u(t)$$

$$x(0) = \bar{x} + \delta x(0)$$



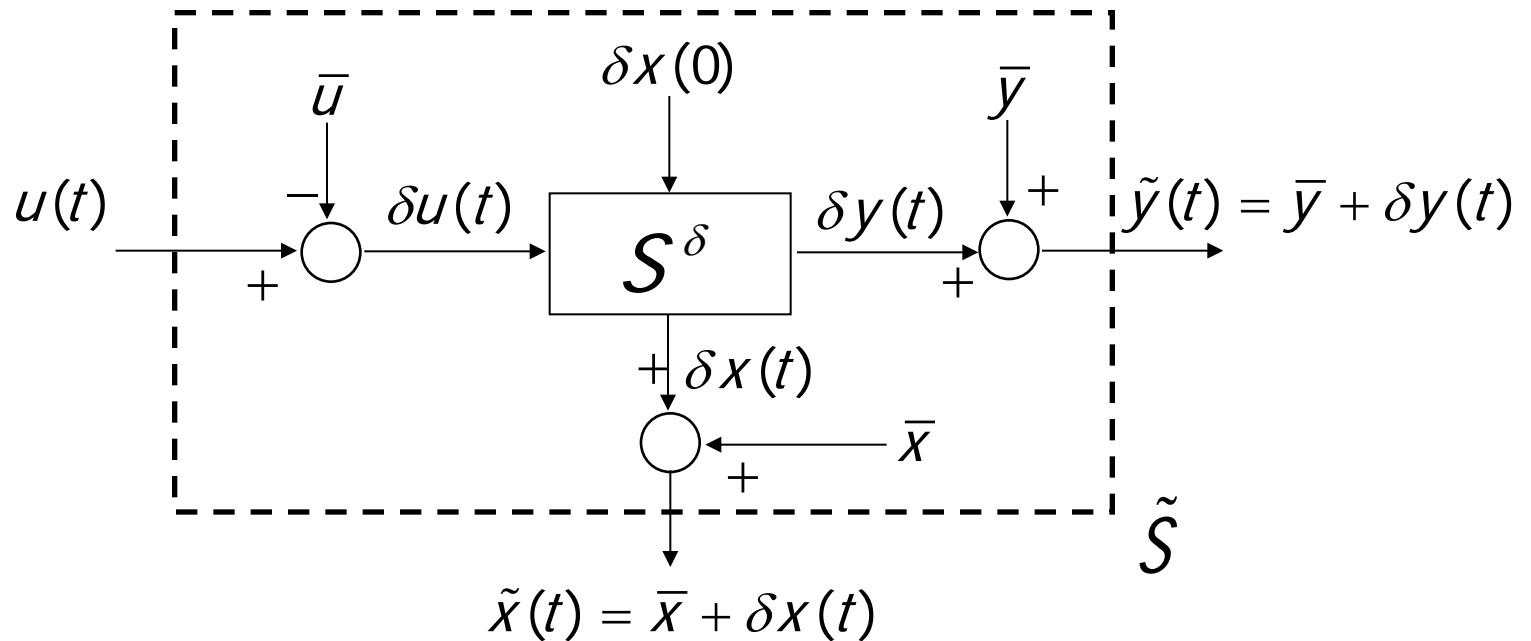
Linearized system:

$$S^\delta : \begin{cases} \delta \dot{x}(t) = A \delta x(t) + B \delta u(t) \\ \delta y(t) = C \delta x(t) + D \delta u(t) \end{cases}$$



# Resume via block diagram representation

Approximating system  $\tilde{S}$



$$u(t) = \bar{u} + \delta u(t) \quad S^\delta : \begin{cases} \delta \dot{x}(t) = A\delta x(t) + B\delta u(t) \\ \delta y(t) = C\delta x(t) + D\delta u(t) \end{cases}$$