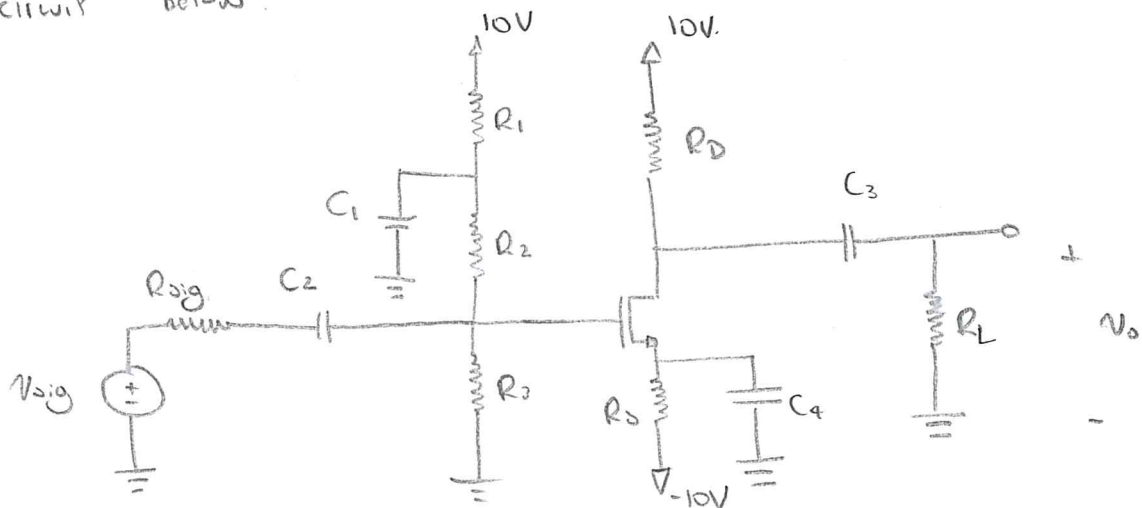


Experiment 1 ENCI Analog Devices

Consider the amplifier circuit below:

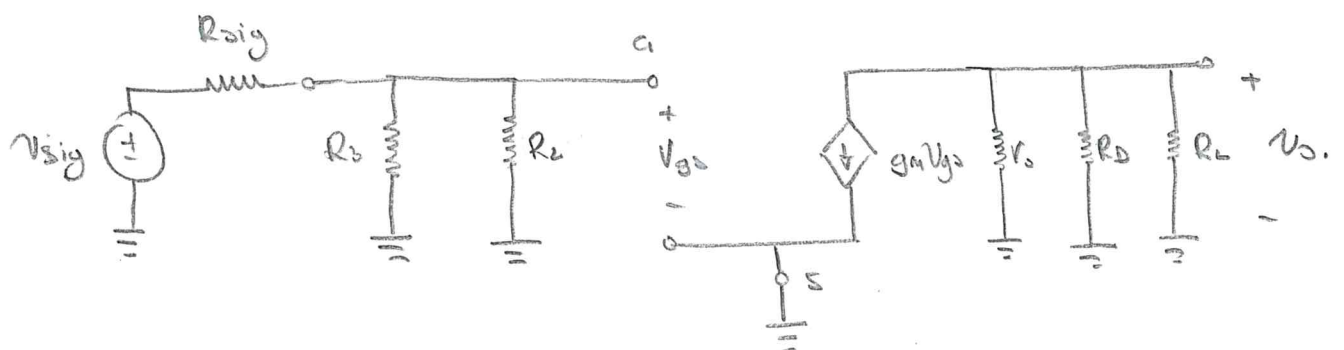


If we consider a small signal model, the capacitors act as short circuits, since

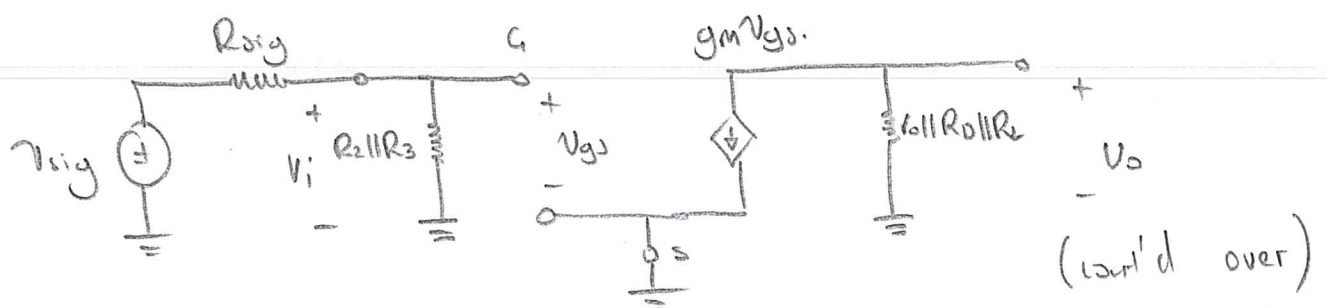
$$Z_c = \frac{1}{j\omega C} \quad \text{as } \omega \rightarrow \infty \quad Z_c \rightarrow 0.$$

We ground the DC voltage sources.

The small signal model is as follows:



this can be simplified:



Experiment 1 ENG Analog Devices

We want to find the voltage gain:

$$A_v = \frac{v_o}{v_i}$$

Now, by voltage divider:

$$v_i = \left[\frac{R_2 \parallel R_3}{R_2 \parallel R_3 + R_{sig}} \right] \cdot v_{sig}$$

Also, we note that $v_i = v_{gs}$.

$$\text{Now, } v_o = -g_m v_{gs} (r_o \parallel R_D \parallel R_L)$$

$$v_o = -g_m v_i (r_o \parallel R_D \parallel R_L).$$

$$\therefore A_v = \frac{v_o}{v_i} = -g_m (r_o \parallel R_D \parallel R_L)$$

$$\text{Voltage Gain : } A_v = -g_m (r_o \parallel R_D \parallel R_L).$$

Now, overall voltage gain

$$G_{vo} = \frac{v_o}{v_{sig}} = -g_m (r_o \parallel R_D \parallel R_L) \left[\frac{R_2 \parallel R_3}{R_{sig} + R_2 \parallel R_3} \right].$$

$$\text{Overall Voltage Gain : } G_{vo} = -g_m (r_o \parallel R_D \parallel R_L) \left[\frac{R_2 \parallel R_3}{R_{sig} + R_2 \parallel R_3} \right].$$

To calculate both voltage gain + overall voltage gain, we need the parameters for r_o + g_m , which we find by working out the DC bias point.

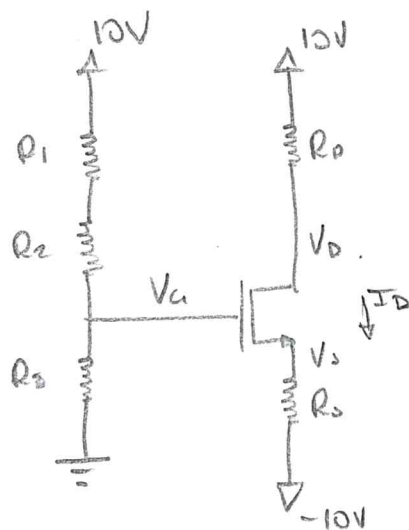
(cont'd over)

Experiment 1 ENCA Analog Devices

If we consider large signals only then the impedance for the capacitor gets very large.

$$X_c = \frac{1}{j\omega C} \quad X_c \rightarrow \infty \text{ as } \omega \rightarrow 0.$$

Hence, the capacitors act as short circuits and we get the following large signal model.



We want to find V_a

$$V_a = \left[\frac{R_3}{R_1 + R_2 + R_3} \right] \cdot 10V.$$

$$\therefore V_a = \frac{145.5}{500} \cdot 10$$

$$V_a = 2.91V.$$

$$\text{Now, } -V_s + I_D R_5 - 10V = 0.$$

$$V_s = I_D R_5 - 10 \quad \text{--- (1)}$$

Further, if the device is in saturation, we get that:

$$I_D = \frac{1}{2} k_n (V_{as} - V_T)^2.$$

$$\therefore I_D = \frac{1}{2} k_n (V_a - V_s - V_T)^2 \quad \text{--- (2)}$$

Sub in eqn 1) to eqn 2) to get:

$$I_D = \frac{1}{2} k_n (2.91 - I_D \cdot R_5 + 10 - 2)^2$$

$$I_D = \frac{2.4 \times 10^{-3}}{2} (2.91 - 10k \cdot I_D + 10 - 2)^2 \quad \text{Cont'd over}$$

Experiment 1 ENG Analog Device.

Large signal Analysis.

$$\therefore \frac{I_D}{1.2e-3} = (10.91 - 10k \cdot I_D)^2.$$

$$833.33 I_D = 119.02 - 218200 I_D + 100e6 I_D^2$$

$$100e6 I_D^2 - 219033.33 I_D + 119.02 = 0.$$

$$I_D = 9.999 e-4 A$$

$$\therefore I_D = 10 e-4 A$$

$$I_D = 1mA.$$

Note: $I_D < 0$ for the other solⁿ & was thus discarded.

If $I_D = 1mA$,

$$\text{then } V_D = 10 - 1mA \cdot 7k$$

$$V_D = 10 - 7$$

$$V_D = 3$$

$$\text{also, } V_S = -10 + 1mA \cdot 10k.$$

$$V_S = 0V.$$

$$\therefore V_D - V_S > V_a - V_s - V_T$$

\Rightarrow Device is actually in saturation, w Bias current $I_D = 1mA$.

Hence,

$$g_m = \mu_n C_{ox} \frac{W}{L} V_{ov} = 2.4mA/V^2 \cdot (2.91 - 2)$$

$$g_m = 2.4e-3 (0.91) \text{ S}$$

$$= 2.184e-3 \text{ S}$$

(cont'd over).

Experiment 1

ENCI Analog Devices

Finally, ignoring the effect of r_o the
Voltage gain is:

$$A_v = -g_m (R_D \parallel R_L)$$

$$= -2.184 \times 10^{-3} \left(\frac{1}{7k} + \frac{1}{10k} \right)^{-1}$$

$$= -8.9929.$$

$$A_v = -9 \frac{V}{V}$$

Overall voltage gain is:

$$G_{vo} = -g_m (R_D \parallel R_L) \left[\frac{R_2 \parallel R_3}{R_{sig} + R_2 \parallel R_3} \right]$$

$$= -9 \cdot \left[\frac{79.5k}{2k + 79.5k} \right]$$

$$= -8.77 \frac{V}{V}$$

$$G_{vo} = -8.77 \frac{V}{V}$$

The max value of the voltage gain
in the Pspice simulation is approx $20 \frac{V}{V}$,
which doesn't reconcile to the results derived
here.