# 4 NIT IMPULSE 2 STEP FUNCTIONS (P36)

$$\delta[n] = \begin{cases} 1 & n=0 \\ 0 & n\neq 0 \end{cases}$$

$$u(n) = \begin{cases} 1 & n > 0 \\ 0 & n < 0 \end{cases}$$

$$u[n] = \sum_{m=-\infty}^{n} \delta[m]$$

$$= \mathcal{\tilde{Z}} \mathcal{S}[n-k]$$

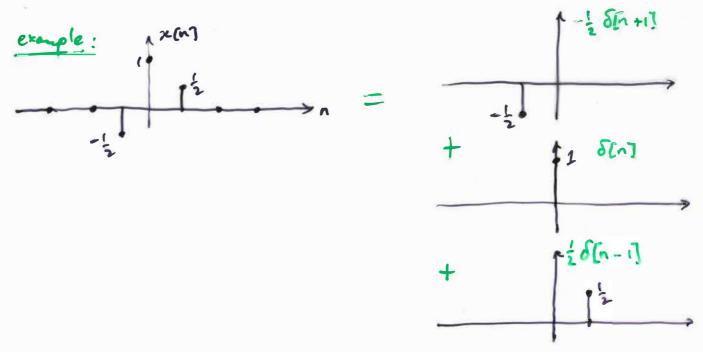
### SAMPLING PROPERTY

$$x(n) \delta(n) = x(0) \delta(n)$$

# LINEAR TIME-INVARIANT SYSTEMS (P74) (L.T. I)

### D.T. SYSTEMS - CONVOLUTION SUM

ANY X[n] CAN BE CONSTRUCTED FROM UNIT IMPULSE



ie. 
$$\infty[n] = -\frac{1}{2} \delta[n+1] + \delta[n] + \frac{1}{2} \delta[n-1]$$

$$= \infty[-1] \delta[n+1] + \infty[-1] \delta[n] + \infty[-1] \delta[n]$$

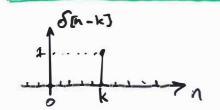
$$= \frac{1}{2} \infty[k] \delta[n-k]$$

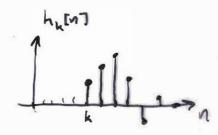
$$x[n] = \sum_{k=-\infty}^{\infty} x[k] \delta[n-k]$$

$$k=-\infty$$
Coefficient
Coefficient
Coefficient

(p77)

### IMPULSE RESPONSE





ALSO, BECAUSE LINEAR

- Recall that in general notation, any x(n) ft can be represented as

$$2[n] = \sum_{k=-\infty}^{\infty} 2[k] \delta[n-k]$$

$$y(n) = \sum_{k=-n}^{n} x(k) h_k(n)$$

Also, since  $T_{1}$   $h_{1}[n] = h_{0}[n-1]$   $h_{2}[n] = h_{0}[n-2]$   $h_{3}[n] = h_{0}[n-3]$ 

$$y(n) = \sum_{k=-\infty}^{\infty} x(k) h(n-k)$$

$$= x(n) \implies h(n)$$

where h(n) = ho[n]

#### PROPORTIOS

- hing fully CHARACTERISES A LTI SYSTEM
- x(n) + h(n) = h(n) \* x(n)
- FINITE IMPULSE RESPONSE (FIR)

$$h[n] = h(0) \delta[n] + h[i] \delta[n-i] + h[i] \delta[n-2] + h[m] \delta[n-m]$$

FOR M FINITE.

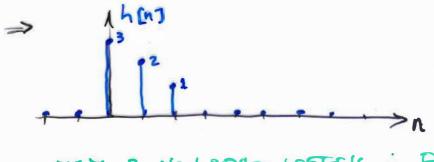
1.5.8

29: AVERTGING FILTER WITH WEIGHTS

y[n] = 3 20[n] + 2 x[n-1] + x[n-2] TEST WITH X(n) = d[n]

=> y(n)= 3 d(n) + 2 d(n-1) + d(n-2)

.... y [-2] =0, y[-1] =0, y[0]=3, y[1]=2, y[2]=1 y[3]=



only 3 NON ZERO LOFFF'S . FIR.

INFINITE IMPULSE RESPONSE (IIR) IF HEN] HAS OD NUMBER OF NON ZERO REMENS

# CALCULATING A CONVOLUTION (>80)

EXAMPLE: SUPPOSE 4 LTI SYSTEM HAS UNIT IMPULSE



WHAT IS THE OUT PUT WHEN THE INPUT IS



ANSWER :

METURO 1: APPLY 
$$y[n] = \sum_{k=-\infty}^{\infty} x(k)^{2} h[n-k]$$
 COLLITION

 $y[-17] = x(0)^{2} h[-1-0] + x[1]^{2} h[-1-1]$ 
 $= \frac{1}{2}$   $0 + 2$   $0 = 0$ 
 $y[0] = x[0)^{2} h[0-0] + x[1]^{2} h[0-1]$ 
 $= \frac{1}{2}$   $0 = \frac{1}{2}$ 
 $y[1] = x[0]^{2} h[1-0] + x[1]^{2} h[1-1]$ 
 $= \frac{1}{2}$   $1 + 2$   $1 = 2\frac{1}{2}$ 
 $y[2] = x[0]^{2} h[2-0] + x[1]^{2} h[2-1]$ 
 $= \frac{1}{2}$   $1 + 2$   $1 = 2\frac{1}{2}$ 
 $y[3] = x[0]^{2} h[3-0] + x[1]^{2} h[3-1] = 2$ 
 $y[4] = x[0]^{2} h[4-0] + x[1]^{2} h[4-1] = 0$ 
 $= x[1]^{2} h[4-1] = x[1]^{2} h[4-1] = 0$ 

### METHOD 2: GRAPHICAL

THINK OF X[N] AS TWO IMPULSES SUMMED

\[
\frac{1}{2} \lambda[\text{N}] \lambda 2 \lambda[\text{N}-1]
\]

SINCE LTI SYSTEM, OUTPUT WILL BE THE SUM OF 2 RESPONSES

2 (S BELAUSE XCh) IS ARRE UP OF 2

DELTA FUNCTIONS IN

THIS EX AMPLE.

NOTE THE SLIDING EFFECT

EXAMPLE: WHAT IS THE IMPULSE RESPONSE OF THIS 11R SYSTEM:

ALSO , DRAW A BLOCK DIAGRAM OF THE SIGNAL FLOW

ANSWER: CHOOSE X(n) = S(n)

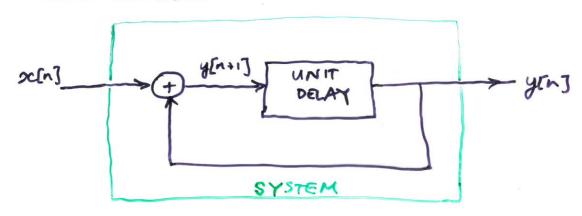
$$\frac{1}{2} y^{2}(0) = y^{2}(0) + x^{2}(0) = 0$$

$$y^{2}(0) = y^{2}(0) + x^{2}(0) = 1$$

$$y^{2}(2) = y^{2}(1) + x^{2}(1) = 1$$



BLOCK DIAGRAM



THIS IS AN INTEGRATOR !!

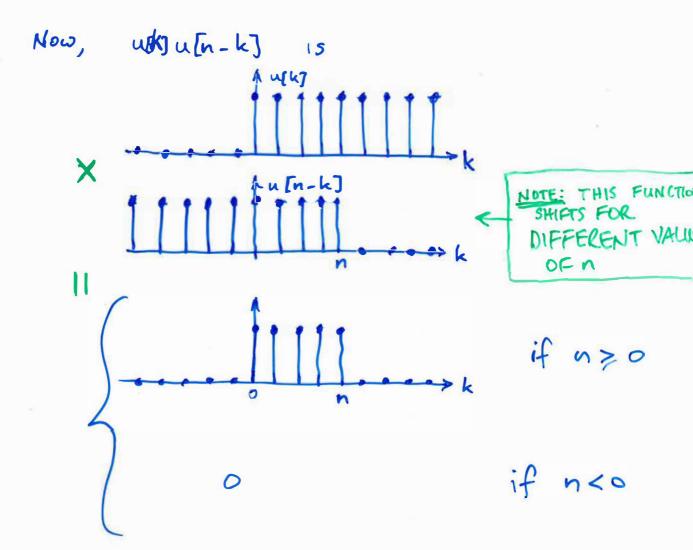
#### CONVOLUTION GXAMPLE

COMPUTE hin] \* Dein] FOR Dein] = 
$$\alpha^n u[n]$$

AND hin] =  $\beta^n u[n]$ 
 $(\alpha \neq \beta)$ 

ANSWER: 
$$h(n) \neq M(n) = \sum_{k=-\infty}^{\infty} c(k) h(n-k)$$

$$= \sum_{k=-\infty}^{\infty} d^k u(k) p^{n-k} u(n-k)$$



$$\frac{1}{16} h = \sum_{k=0}^{n} \frac{1}{2} \left(\frac{\alpha}{\beta}\right)^{k}$$

$$= \frac{1}{16} \frac{1}{2} \left(\frac{\alpha}{\beta}\right)^{k}$$

$$= \frac{1}{16} \frac{1}{16} \left(\frac{\alpha}{\beta}\right)^{k}$$

$$= \frac{1}{16} \frac{1}{16} \left(\frac{\alpha}{\beta}\right)^{k}$$

$$= \frac{1}{16} \frac{1}{16}$$

## PROPERTIES & EXAMPLES OF DIFT

TIME SHIFT  $\infty[n-n_0] \xrightarrow{\mathcal{F}} e^{-jwn_0} \times (e^{jw})$ FRER SHIFF  $e^{jw_0} \cap \infty[n] \xrightarrow{\mathcal{F}} \times (e^{j(w-w_0)})$ 

RECALL THAT IN D.T. ONCE YOU HAVE SHIFTED BY

JUT IN FREQ, YOU ARE BACK TO THE START.

 $e^{j(w+2\pi i)n}$  =  $e^{jwn}$  =  $e^{jwn}$  =  $e^{jwn}$ ALWAYS = 1 BECAUSE IN

DT. n is AN INTEGRA

T, 3TC, 5TC, 7TC, .... REPRESENT HIGH FREELY

$$x(at) \stackrel{7}{\rightleftharpoons} \frac{1}{|a|} \times (\frac{jw}{a})$$

IN D.T. IF 
$$x_{(k)}[n] = \int x_{(n)}[n]$$

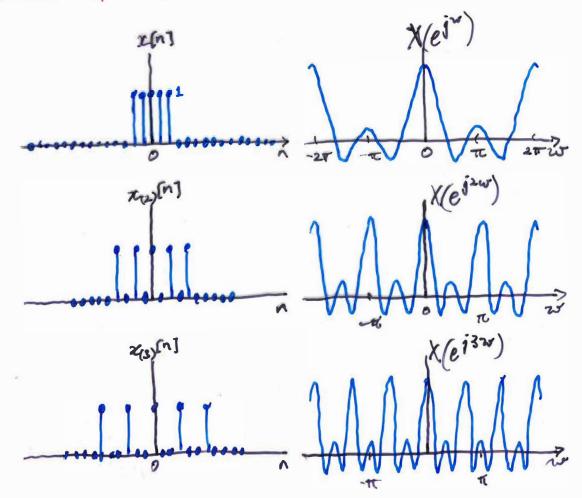
t n alot a

IF A IS A MULT OF K

MULTIPLE OF K

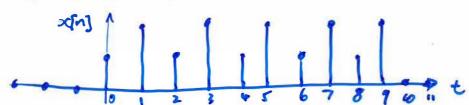
$$x_{(h)}(n) \stackrel{7}{\longleftrightarrow} \chi(e^{jkw})$$

Example: ( > 378)



EXAMPLE (TIME SHIFT, TIME EXPANSION & LINEARITY)

FIND THE D.T. F.T. OF



ANSWEL: YOU COULD APPLY THE DIFT FORMULA X(ejw) = É acni e-1wn

AND THEN SPEND TIME MONIPULATING ALL THE TERRAS

1.18 %.

OR

REACUSE THAT

IN THE PREVIOUS EXAMPLE & IS A SHIPTED VERSION OF THE DECAT LAST LECTURE

SMFT 
$$\Rightarrow Y(e^{jw}) = e^{-j2w} \left( \frac{\sin(5w/2)}{\sin(w/2)} \right)$$

EX PANSION

$$\Rightarrow \chi(e^{jw}) = e^{-jw} \left(1 + 2e^{-jw}\right) \frac{\sin(5w)}{\sin(w)}$$

$$x_i[n] x_i[n] \stackrel{\mathcal{F}}{\Longleftrightarrow} \frac{1}{2n} \int \chi_i(e^{j\theta}) \chi_i(e^{j(w-\theta)}) d\theta$$

periodic convolution

### DIFFERENCE EQUATIONS

$$y(n) - \frac{3}{4}y(n-1) + \frac{1}{8}y(n-2) = 2 x(n)$$
  
FIND h[n]

### ANSWER:

BY INSPECTION

$$H(e^{jw}) = \frac{2}{1 - \frac{3}{4}e^{-jw} + \frac{1}{8}(e^{-jw})^{2}}$$

$$= \frac{A}{1 - \frac{1}{2}e^{-jw}} - \frac{B}{1 - \frac{1}{4}e^{-jw}}$$

:. 
$$h(n) = 4(\frac{1}{2})^n u(n) - 2(\frac{1}{4})^n u(n)$$

$$= 2(2(\frac{1}{2})^n - (\frac{1}{4})^n) u(n)$$

Summary of Fourier Series and Transform

Time	istoriii	Frequency
1	CTFS	
<u> </u>	$x(t) = \sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t} ,$	
CT, periodic $W_0 = \frac{2\pi}{T}$ with period T	$a_k = \frac{1}{T} \int_{\Gamma} x(t) e^{-jk\omega_0 t} dt$	Discrete, aperiodic
	DTFS	· · · · · · · · · · · · · · · · · · ·
111 111 N	$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk\omega_0 n}$	
DT, periodic $\omega_0 = \frac{2\pi}{N}$	$a_k = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-jk\omega_0 n}$	Discrete, periodic with period N
	CTET	
	CTFT	
$\overline{}$		<b>~</b>
CT, aperiodic	$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega$	Cont, aperiodic
	$X(j\omega) = \int_{-\infty}^{+\infty} x(t)e^{-j\omega t} dt$	2nak
$\frac{M}{-T}$		1111 t
CT, periodic with		CT impulsive; a periodic
	DTFT	-27 0 27 w
BT, aperiodic	$x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega$	Cont, periodic with period 27
	+60	
-N 0 N n	$X(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} x[n]e^{-j\omega n}$	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1
DT, periodic with  period N, Wo = N		et, impulsive, poriodic with period 271