THE LAPLACE TRANSFORM [p 654]

· GENERALIZATION OF FOURIER TRANSFORM

. USES est BASIS FUNCTIONS

$$(s = jw \Rightarrow F.\tau.)$$

· CAN IN CLUDE UNSTABLE SIGNALS

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

H(S) IS CONLED THE SYSTEM FUNCTION

IF h(t) IS THE IMPULSE OF A LTI SYSTEM, THEN

SET 
$$S = \sigma + jw$$

$$X(S) = X(\sigma + jw) = \int_{-\infty}^{\infty} (x(k) e^{-\sigma t}) e^{-jwt} dt$$

$$= F.T. \text{ of } x(k) e^{-\sigma t}$$

$$EXAMPLE: (p 656) \quad x(t) = e^{-at} \text{ in } (t) \quad , \text{ a REAL}$$

$$FIRST \text{ RECALL } X(jw) = jw + a \quad , \text{ a > 0}$$

$$X(S) = \int_{-\infty}^{\infty} e^{-at} \text{ if } e^{-st} dt = \int_{0}^{\infty} e^{-(a+s)t} dt$$

$$= \frac{1}{S+a} \quad \text{if } \text{ Re}\{a+s\} > 0$$

$$\text{if } \text{ Re}\{s\} > -a$$

\* ROC: REGION OF CONVERGENCE

: VALUES OF S FOR WHICH THE I CONVERGES

EXAMPLE: (p658)

$$x(t) = 3e^{-2t}u(t) - 2e^{-t}u(t)$$

WHAT IS THE ROC. ? WHAT IS THE L.T.?

ANSWER:

$$e^{-t}$$
  $u(t)$   $\longleftrightarrow$   $\frac{1}{s+1}$   $Re{s}$  > -1
$$e^{-2t}$$
  $u(t)$   $\longleftrightarrow$   $\frac{1}{s+2}$   $Re{s}$  > -2

$$\therefore x(t) \iff X(s) = \frac{3}{s+2} - \frac{2}{s+1}$$

$$= \frac{s-1}{s^2 + 3s + 2}, \quad Re\{s\} > -1$$

$$= \frac{8}{s^2 + 3s + 2}, \quad Roc. C.$$

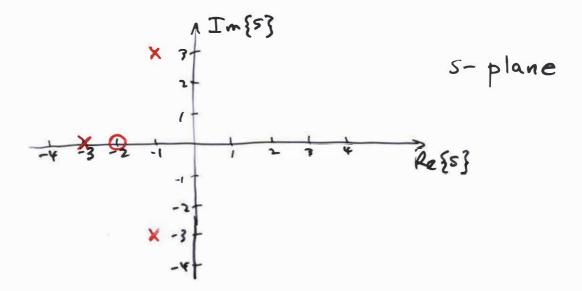
# \* POLES & ZEROS

A POLE IS LOCATED WHERE THE T/F = 00

A ZERO IS LOCATED WHERE THE T/F = 0

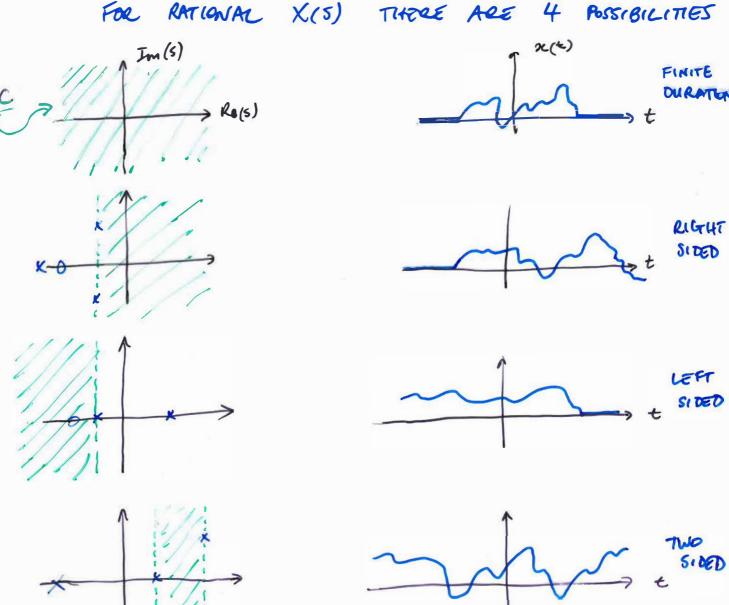
EXAMPLE: 
$$\chi(s) = \frac{10s+20}{(s^2+2s+10)(s+3)} = \frac{10(s+2)}{[(s+1)^2+3^2](s+3)}$$

$$\Rightarrow$$
 ZERO AT  $S = -2$   
POLE AT  $S = -1 = j3$ ,  $S = -3$ 



## REGIONS OF CONVORGENCE (p 662)

FOR RATIONAL X(5) THERE ARE 4 POSSIBILITIES



NOTE: NO POLES IN THE ROC. ROC. EXTENDS TO 00 OR TO A POLE

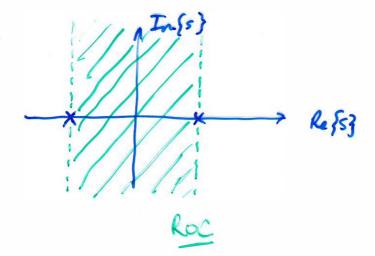
$$= e^{-bt} u(t) + e^{bt} u(-t)$$

$$= e^{-bt} u(t) + e^{-bt} u(-t)$$

$$\frac{1}{s+b}$$
 , Re{s} >-b  $-\frac{1}{s-b}$  , Ref

$$6. \times (6) = \frac{1}{5+5} - \frac{1}{5-5}$$

$$= -\frac{2b}{5^2-b^2}$$
,  $-b < Re[5] < b$ 



# INVERSÉ LAPLACE TRANSFORM (p670)

RECALL THAT 
$$\chi(s) = F.T.$$
 OF  $x(t) e^{-\sigma t}$ 

$$\therefore x(t) e^{-\sigma t} = \int_{-\infty}^{\infty} X(\sigma + jw) e^{jwt} dw$$

$$\Rightarrow \chi(t) = \frac{1}{2\pi} \int_{-a}^{a} \chi(s) e^{st} dw$$

$$\chi(S) = \frac{1}{(S+1)(S+2)}$$
,  $Ra\{S\} > -1$   
=  $\frac{A}{S+1} + \frac{B}{S+2} \Rightarrow A = 1, B = -1$ 

NOW , RECALL THAT

:. 
$$x(t) = e^{-t} u(t) - e^{-2t} u(t)$$

EXAMPLE:

$$X(5) = \frac{1}{(5+1)(5+2)}$$
, Re  $\{5\}$   $(5+2)$ 

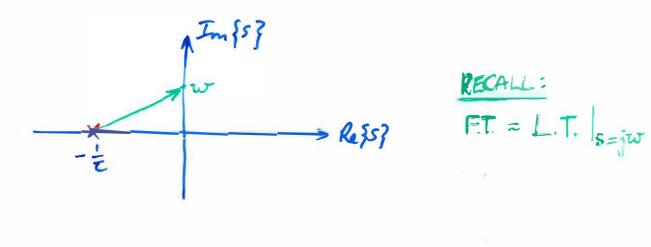
$$x(t) = -e^{-t}u(-t) + e^{-2t}u(-t)$$

=) Roc's must 85 Re{s} < -1 & Re{s} > -2

:. 
$$x(t) = -e^{-t} u(-t) - e^{-2t} u(t)$$

# GEOMETRIC EVALUATION OF X(S) - OR X(JW)

$$h(t) = \frac{1}{\tau} e^{-t/\tau} u(t)$$



### OBSERVATIONS:

1) FOR 
$$w = \frac{1}{2}$$
,  $\left| H(j\frac{1}{2}) \right| = \frac{1}{\sqrt{2}}$   
FOR  $w = 0$ ,  $\left| H(o) \right| = 1$ 

@ AS POLE IS MOVED FURTHER TO THE LEFT, THE SYSTEM CAN RESPOND FASTER (BIGGER B.W.)

CONCLUSIONS: POLES & ZEROS ARE EXCELLENT DESIGN