

## LAST LECTURES: LAPLACE T.F.

(= GENERALISED F.T.)

INVERSE L.T.F.

GEOMETRIC INTERPRETATIONS OF POLES  
& ZEROS

THIS LECTURE: PROPERTIES OF L.T.

LTI SYSTEMS & L.T.

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## PROPERTIES OF LAPLACE TRANSFORM (P682)

CONSIDER  $x(t) \xleftrightarrow{\mathcal{L}} X(s)$  WITH  $\text{ROC} = R$

### LINEARITY:

$$a x_1(t) + b x_2(t) \xleftrightarrow{\mathcal{L}} a X_1(s) + b X_2(s)$$

WITH  $\text{ROC} = R_1 \cap R_2$

(at least!)

(ex  $a = -b$  &  $x_1 = x_2$ )

### TIME SHIFT:

$$x(t - t_0) \xleftrightarrow{\mathcal{L}} e^{-st_0} X(s) \quad \text{ROC} = R$$

### CONVOLUTION:

$$x_1(t) * x_2(t) \xleftrightarrow{\mathcal{L}} X_1(s) X_2(s)$$

WITH  $\text{ROC} = R_1 \cap R_2$

(at least!)

### DIFFERENTIATION:

$$\frac{dx(t)}{dt} \xleftrightarrow{\mathcal{L}} s X(s) \quad \text{ROC CONTAINS } R$$

(ROC LARGER IF POLE AT  $s=0$  IS CANCELLED)

## INTEGRATION

$$\int_{-\infty}^t x(\tau) d\tau \quad \xleftrightarrow{Z} \quad \frac{1}{s}$$

ROC CONTAINS

$$R \cap \{ \operatorname{Re}\{s\} > 0 \}$$

## INITIAL VALUE THEOREM

$$x(0+) = \lim_{s \rightarrow \infty} s X(s)$$

## FINAL VALUE THEOREM

$$\lim_{t \rightarrow \infty} x(t) = \lim_{s \rightarrow 0} s X(s)$$

FOR  $x(t) = 0, t < 0$   
& NO SINGULARITIES  
AT  $t = 0$

EXAMPLE:

FIND THE INITIAL VALUE OF ~~SIGNAL~~ THE SIGNAL

$$x(t) = e^{-2t} u(t) + e^{-t} (\cos 3t) u(t)$$

ALSO, FIND ITS L.T. & CONFIRM THE INITIAL VALUE THEOREM.

ANSWER

$$\begin{aligned} x(0^+) &= \lim_{\substack{t \rightarrow 0 \\ \text{from above}}} e^{-2t} + e^{-t} \cos(3t) \\ &= 2 \end{aligned}$$

NOW, TO FIND THE L.T., COULD TRY TO DIRECTLY SOLVE L.T. INTEGRAL, OR, USE LINEAR PROPERTY

$$\begin{aligned} x(t) &= \underbrace{e^{-2t} u(t)}_{\frac{1}{s+2} ; \operatorname{Re}(s) > -2} + \underbrace{\frac{1}{2} e^{-(1-3j)t} u(t)}_{\frac{1}{2} \cdot \frac{1}{s+(1-3j)} ; \operatorname{Re}(s) > -1} + \underbrace{\frac{1}{2} e^{-(1+3j)t} u(t)}_{\frac{1}{2} \cdot \frac{1}{s+(1+3j)} ; \operatorname{Re}(s) > -1} \end{aligned}$$

$$\therefore X(s) = \frac{2s^2 + 5s + 12}{(s^2 + 2s + 10)(s + 2)} \quad ; \operatorname{Re}\{s\} > -1$$

$$\begin{aligned} \text{Now, } \lim_{s \rightarrow \infty} s X(s) &= \lim_{s \rightarrow \infty} \frac{2s^3 + 5s^2 + 12s}{s^3 + 4s^2 + 14s + 20} \\ &= 2 \end{aligned}$$

## ANALYSIS AND CHARACTERIZATION OF LTI SYSTEMS WITH THE LAPLACE TRANSFORM (p 693)

$$Y(s) = H(s) X(s) \quad \text{ie convolution in the time domain.}$$

RECALL THAT IF  $s = j\omega$  THEN IT'S JUST THE FOURIER TRANSFORM!

### CAUSALITY

$$h(t) = 0 \quad \text{FOR } t < 0$$

$\therefore h(t)$  RIGHT SIDED

$\therefore$  ROC MUST BE A RIGHT HALF PLANE  
BUT ONLY HOLDS IN REVERSE IF  $H(s)$  RATIONAL !!

### EXAMPLES:

①  $h(t) = e^{-t} u(t)$

$$\Rightarrow H(s) = \frac{1}{s+1}, \quad \operatorname{Re}\{s\} > -1$$

$\downarrow$   
 RATIONAL  $\Rightarrow$  CAUSAL

EXAMPLE (2):  $h(t) = e^{-|t|}$  NOT CAUSAL

$$H(s) = \frac{-2}{s^2 - 1} ; -1 < \operatorname{Re}(s) < 1$$

(3):  $H(s) = \frac{e^s}{s+1}$  NOT RATIONAL!  
 $\operatorname{Re}(s) > -1$

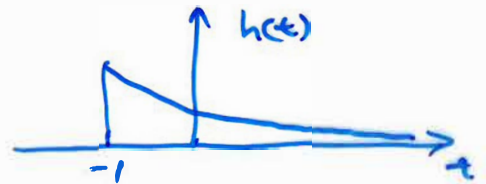
$h(t)$  IS RIGHT SIDED, BUT

SINCE  $e^{-t} u(t) \longleftrightarrow \frac{1}{s+1} ; \operatorname{Re}(s) > -1$

THEN  $e^{-(t+1)} u(t+1) \longleftrightarrow \frac{e^s}{s+1} ; \operatorname{Re}(s) > -1$

$\Rightarrow h(t)$  IS NOT ZERO FOR ALL  $t < 0$

$\therefore$  NOT CAUSAL



## STABILITY:

RECALL: LTI SYSTEM IS STABLE IFF  $\int_{-\infty}^{\infty} |h(t)| dt < \infty$   
 $\Rightarrow$  F.T. CONVERGES

i.e. A STABLE LTI HAS A F.T.

ALSO RECALL: F.T. = L.T. EVALUATED ALONG  $j\omega$ -AXIS

$\therefore$  LTI SYSTEM IS STABLE IFF R.O.C. OF  $H(s)$  INCLUDES  $j\omega$  AXIS.

EXAMPLE: (p 695)

$$H(s) = \frac{s-1}{(s+1)(s-2)}$$

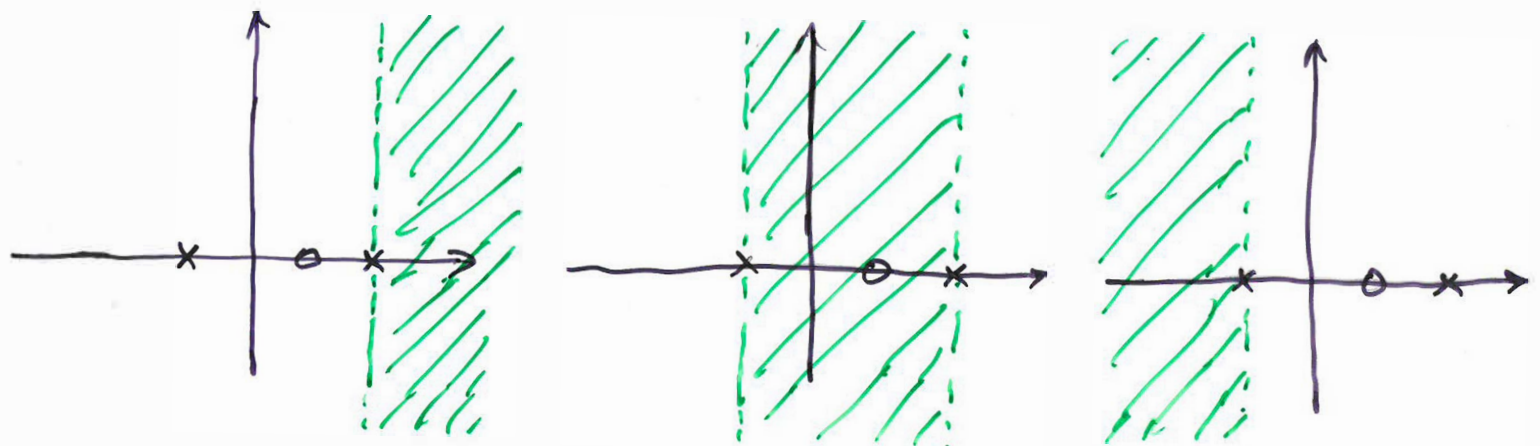
(a) IF YOU ARE TOLD THAT THIS SYSTEM IS STABLE, WHAT IS THE IMPULSE RESPONSE?

(b) WHAT IF YOU WERE TOLD IT WAS CAUSAL?

ANSWER

$$H(s) = \frac{2/3}{s+1} + \frac{1/3}{s-2}$$

ZERO AT  $s=1$   
POLES AT  $s=-1, 2$





ANSWER (a) STABLE REQUIRES R.O.C. INCLUDES  $j\omega$  AXIS.

$$\Rightarrow H(s) = \frac{2/3}{s+1} + \frac{1/3}{s-2} ; -1 < \operatorname{Re}\{s\} < 2$$

$$\Rightarrow h(t) = \frac{2}{3} e^{-t} u(t) - \frac{1}{3} e^{2t} u(-t)$$

ANSWER (b) CAUSAL REQUIRES R.O.C. IS A RIGHT HALF PLANE

$$\Rightarrow \operatorname{Re}\{s\} > 2$$

$$\Rightarrow h(t) = \frac{2}{3} e^{-t} u(t) + \frac{1}{3} e^{2t} u(t)$$

IMPORTANT OBSERVATION:

A CAUSAL SYSTEM WITH RATIONAL  $H(s)$   
IS STABLE IFF ALL POLES OF  $H(s)$   
HAVE  $\operatorname{Re}\{s\} < 0$

LTI SYSTEMS DESCRIBED BY D.E.s (p 698)

EXAMPLE:

$$\frac{dy(t)}{dt} + 3y(t) = x(t)$$

$$\text{L.T. } \Rightarrow sY(s) + 3Y(s) = X(s)$$

$$\therefore H(s) = \frac{Y(s)}{X(s)} = \frac{1}{s+3}$$

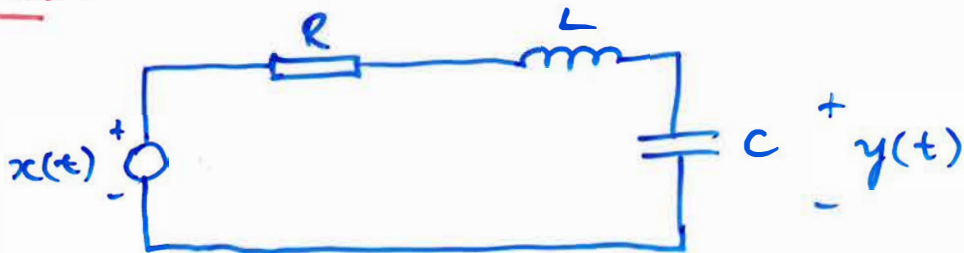
R.O.C. IS NOT  
SPECIFIED BY THE  
DIFF. EQ!!

BUT, IF YOU ALSO KNOW IT IS CAUSAL THEN YOU

KNOW  $\text{Re}\{s\} > -3$  (NOTE: THIS CASE "ALSO STABLE")

$$\Rightarrow h(t) = e^{-3t} u(t)$$

EXAMPLE:



$$x(t) = i(t) R + L \frac{di(t)}{dt} + y(t)$$

$$\& i(t) = C \frac{dy(t)}{dt}$$

$$\Rightarrow x(t) = RC \frac{dy(t)}{dt} + LC \frac{d^2 y(t)}{dt^2} + y(t)$$

$$\Rightarrow H(s) = \frac{1/LC}{s^2 + (R/L)s + (1/LC)}$$

NOW, SINCE  $R, L \& C$  ARE ALL +ve  $\rightarrow$  POLES HAVE <sup>ALL</sup> -ve REAL PTS

$\Rightarrow$  SYSTEM IS ALWAYS STABLE

Q. WHAT IF YOU REMOVE THE RESISTOR?