

COLLEGE OF ENGINEERING
SCHOOL OF AERONAUTICAL AND ASTRONAUTICAL ENGINEERING

AAE 567: Introduction To Applied Stochastic Processes

# Final Exam

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## Table of Contents

1	Problem 1	2
2	Problem 2	5
3	Problem 3	10
4	Problem 4	15

Consider the strictly positive matrix

$$T = \begin{bmatrix} 3 & 2 & 1 \\ 2 & 3 & 2 \\ 1 & 2 & 3 \end{bmatrix}.$$

Using this matrix consider the inner product on  $\mathbb{C}^3$  defined by

$$(x,y)_T = (Tx,y) = y'Tx$$
  $(x \in \mathbb{C}^3 \text{ and } y \in \mathbb{C}^3)$ 

where ' denotes the complex conjugate transpose. The norm of this inner product is given by  $||x||_T = \sqrt{(Tx,x)}$ . Let

$$e_1 = \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}$$
 and  $e_2 = \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix}$  and  $e_3 = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$ .

Solve the following optimization problem

$$\delta = \min\{\|e_1 - \alpha e_2 - \beta e_3\|_T^2 : \alpha \in \mathbb{C} \text{ and } \beta \in \mathbb{C}\}\$$
  
=  $\min\{(T(e_1 - \alpha e_2 - \beta e_3), (e_1 - \alpha e_2 - \beta e_3)) : \alpha \in \mathbb{C} \text{ and } \beta \in \mathbb{C}\}.$ 

In other words, find  $\delta$  and scalars  $\alpha$  and  $\beta$  such that

$$||e_1 - \alpha e_2 - \beta e_3||_T^2 = \delta = \min\{||e_1 - \alpha e_2 - \beta e_3||_T^2 : \alpha \in \mathbb{C} \text{ and } \beta \in \mathbb{C}\}.$$

#### **Solution:**

From the principle of projection theorem we know that

$$e_1 - \alpha e_2 - \beta e_3 \perp e_2, e_3.$$

Then using the inner product we have

$$(e_1 - \alpha e_2 - \beta e_3, e_2)_T = 0$$
  

$$(e_1 - \alpha e_2 - \beta e_3, e_3)_T = 0$$

which becomes

$$(e_1, e_2)_T = \alpha(e_2, e_2)_T + \beta(e_3, e_2)_T$$
  

$$(e_1, e_3)_T = \alpha(e_2, e_3)_T + \beta(e_3, e_3)_T$$

$$\begin{bmatrix} (e_1, e_2)_T \\ (e_1, e_3)_T \end{bmatrix} = \underbrace{\begin{bmatrix} (e_2, e_2)_T & (e_3, e_2)_T \\ (e_2, e_3)_T & (e_3, e_3)_T \end{bmatrix}}_{G} \begin{bmatrix} \alpha \\ \beta \end{bmatrix}.$$

Then we can find  $\alpha$  and  $\beta$  by

$$\begin{bmatrix} \alpha \\ \beta \end{bmatrix} = G^{-1} \begin{bmatrix} (e_1, e_2)_T \\ (e_1, e_3)_T \end{bmatrix}.$$

Then,

$$G_{11} = (e_2, e_2)_T = e2'Te_2$$

$$= \begin{bmatrix} 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} 3 & 2 & 1 \\ 2 & 3 & 2 \\ 1 & 2 & 3 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix}$$

$$= \begin{bmatrix} 2 & 3 & 2 \end{bmatrix} \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix}$$

$$= 3.$$

Similarly,

$$G_{12} = 2$$
  $G_{21} = 2$   $G_{22} = 3$ 

and therefore,

$$G = \begin{bmatrix} 3 & 2 \\ 2 & 3 \end{bmatrix}.$$

Also,

$$(e_1, e_2)_T = e_2' T e_1 = 2$$
  
 $(e_1, e_3)_T = e_3' T e_1 = 1$ 

and now we can compute the coefficients as

$$\begin{bmatrix} \alpha \\ \beta \end{bmatrix} = \begin{bmatrix} 3 & 2 \\ 2 & 3 \end{bmatrix}^{-1} \begin{bmatrix} 2 \\ 1 \end{bmatrix}$$
$$= \begin{bmatrix} 0.8 \\ -0.2 \end{bmatrix}.$$

Thus,

$$\alpha = 0.8$$
  $\beta = -0.2$ .

Next, we calculate

$$e_1 - \alpha e_2 - \beta e_3 = \begin{bmatrix} 1 \\ -0.8 \\ 0.2 \end{bmatrix}.$$

Hence,

$$\delta = \min\{ (T(e_1 - \alpha e_2 - \beta e_3), (e_1 - \alpha e_2 - \beta e_3)) : \alpha \in \mathbb{C} \text{ and } \beta \in \mathbb{C} \}$$

$$= \begin{bmatrix} 1 & -0.8 & 0.2 \end{bmatrix} \begin{bmatrix} 3 & 2 & 1 \\ 2 & 3 & 2 \\ 1 & 2 & 3 \end{bmatrix} \begin{bmatrix} 1 \\ -0.8 \\ 0.2 \end{bmatrix}$$

and we obtain

$$\delta = 1.6$$
.

Consider the discrete time system

$$\begin{bmatrix} x_1(n+1) \\ x_2(n+1) \end{bmatrix} = \begin{bmatrix} e^{-\frac{n}{50}} & 1 \\ 2 & \cos(\frac{n}{50}) \end{bmatrix} \begin{bmatrix} x_1(n) \\ x_2(n) \end{bmatrix} + \begin{bmatrix} e^{-\frac{n}{50}}\cos(\frac{n}{50}) \\ 1 \end{bmatrix} u(n)$$
$$y(n) = \begin{bmatrix} 1 + e^{-\frac{n}{50}} & 2 + \sin(\frac{n}{50}) \end{bmatrix} \begin{bmatrix} x_1(n) \\ x_2(n) \end{bmatrix} + (1 + \frac{1}{2}\sin(\frac{n}{50}))v(n)$$

where u and v are independent Gaussian white noise processes. The initial conditions x(0) = 0 and  $\hat{x}(0) = 0$ . To generate u and v in MATLAB, set

$$rng(1000);$$
  $u = randn(1, 20);$   
 $rng(2000);$   $v = randn(1, 20);$ 

Let  $\mathcal{M}_n = span\{y(j)\}_0^n$ . Find the following

(i) 
$$P_{\mathcal{M}_{n-1}}x_1(n)$$
 for  $n = 8, 9, 10$ .

(ii) 
$$P_{\mathcal{M}_n} x_2(n)$$
 for  $n = 8, 9, 10$ .

(Note the indices on the state  $x_1(n)$  in Part (i), and  $x_2(n)$  in Part (ii).) Be careful MATLAB does not have a zero index. So for example, in MATLAB

$$A(0) = A\{1\} = \begin{bmatrix} 1 & 1 \\ 2 & 1 \end{bmatrix}$$

$$A(1) = A\{2\} = \begin{bmatrix} e^{-\frac{n}{50}} & 1 \\ 2 & \cos(\frac{n}{50}) \end{bmatrix}$$
etc.

#### **Solution:**

The time varying discrete system matrices are

$$A(n) = \begin{bmatrix} e^{-\frac{n}{50}} & 1\\ 2 & \cos(\frac{n}{50}) \end{bmatrix} \qquad B(n) = \begin{bmatrix} e^{-\frac{n}{50}}\cos(\frac{n}{50})\\ 1 \end{bmatrix}$$
$$C(n) = \begin{bmatrix} 1 + e^{-\frac{n}{50}} & 2 + \sin(\frac{n}{50}) \end{bmatrix} \quad D(n) = 1 + \frac{1}{2}\sin(\frac{n}{50}).$$

and the system state is

$$x(n) = \begin{bmatrix} x_1(n) \\ x_2(n) \end{bmatrix}.$$

From discrete time Kalman filter we know that if

$$\phi(n) = y(n) - P_{\mathcal{M}_{n-1}}y(n)$$

$$= y(n) - C(n)\hat{x}(n)$$

$$= C(n)\tilde{x}(n) + D(n)v(n)$$

$$\hat{x}(n+1) = P_{\mathcal{M}_n} x(n+1)$$

$$= P_{\mathcal{M}_n} (A(n)x(n) + B(n)u(n))$$

$$= A(n)P_{\mathcal{M}_n} x(n)$$

$$= A(n)\hat{x}(n) + AR_{x(n)\phi(n)} R_{\phi(n)}^{-1} (y(n) - C(n)\hat{x}(n))$$

where

$$R_{x(n)\phi(n)} = Q(n)C(n)^* R_{\phi(n)} = C(n)Q(n)C(n)^* + D(n)D(n)^*.$$

Hence, if 
$$\Delta(n) = A(n)Q(n)C(n)^* \Big(C(n)Q(n)C(n)^* + D(n)D(n)^*\Big)^{-1}$$

$$\hat{x}(n+1) = P_{\mathcal{M}_n} x(n+1)$$
$$= A(n)\hat{x}(n) + \Delta(n) \Big( y(n) - C(n)\hat{x}(n) \Big)$$

and

$$Q(n+1) = A(n)Q(n)A(n)^* + B(n)B(n)^* - \Delta(n)C(n)Q(n)A(n)^*$$

(i) For the first part of the problem, we know from the theory above that we must compute  $\hat{x}_1(8)$ ,  $\hat{x}_1(9)$ , and  $\hat{x}_1(10)$ . We are able to do this with the following MATLAB code.

Listing 1: Problem 2 part (i) MATLAB code

```
% AAE 567 Final Exam Spring 2021 Problem 2 Part (i)
% Tomoki Koike

% Housekeeping commands
clear all; close all; clc;
%%
getA = @(n) [exp(-n/50), 1; 2, cos(n/50)];
```

```
getB = @(n) [exp(-n/50)*cos(n/50); 1];
9 | getC = @(n) [1+exp(-n/50), 2+sin(n/50)];
10 | getD = @(n) 1 + 0.5*sin(n/50);
11
12 % Initialize cells to store matrices
13 A = \{\}; B = \{\}; C = \{\}; D = \{\};
14
15 % Initialize the covariance matrix Q
16 | Q = {}; Q{1} = zeros(2,2);
17
18 % Initialize the x—states and xhat—states
19 |x = \{\}; x\{1\} = zeros(2,1);
20 \mid xhat = \{\}; xhat\{1\} = zeros(2,1);
21
22 \mid \% Initialize the u(n) and v(n) white noise
23 | rng(1000); u = randn(1,20);
24 \mid rng(2000); v = randn(1,20);
25
26 \mid \% Initialize the output states y(n)
27 | y = {};
28
29 | for n = 0:10
30
        A\{n+1\} = getA(n);
31
        B\{n+1\} = getB(n);
32
        C{n+1} = getC(n);
33
        D{n+1} = getD(n);
34
        % Compute x(n+2) which is actually x(n+1)
        x\{n+2\} = A\{n+1\}*x\{n+1\} + B\{n+1\}*u(n+1);
36
        % Compute v(n)
37
        y{n+1} = C{n+1} * x{n+1} + D{n+1}*v(n+1);
38
        [xhat\{n+2\}, Q\{n+2\}] = dkf(A\{n+1\}, B\{n+1\}, C\{n+1\}, D\{n+1\}, ...
39
                                      Q{n+1}, xhat{n+1}, y{n+1});
40
   end
41
   %%
42
   function [Xnew, Qnew] = dkf(Ad, Bd, Cd, Dd, Qd, xhat, y)
43
        % Discrete time Kalman filter
44
        Del = Ad*Od*Cd' * inv(Cd*Od*Cd' + Dd*Dd');
45
        Xnew = Ad*xhat + Del*(y - Cd*xhat);
46
        Qnew = Ad*Qd*Ad' + Bd*Bd' - Del*Cd*Qd*Ad';
47
   end
```

This gives us the following results

$$\hat{x}_1(8) = -31.5388$$
  $\hat{x}_1(9) = -75.9468$   $\hat{x}_1(10) = -174.4413$ 

(ii) For the second part of the problem we modify part (i) so that we obtain

$$P_{\mathcal{M}_n} x(n) = \hat{x}(n) + R_{x(n)\phi(n)} R_{\phi(n)}^{-1} \phi(n)$$
  
=  $\hat{x}(n) + Q(n)C(n)^* \Big( C(n)Q(n)C(n)^* + D(n)D(n)^* \Big)^{-1} \Big( y(n) - C(n)\hat{x}(n) \Big)$ 

and the MATLAB code in Listing 2 will give us these values.

Listing 2: Problem 2 part (ii) MATLAB code

```
% AAE 567 Final Exam Spring 2021 Problem 2 Part (ii)
   % Tomoki Koike
 3
 4
   % Housekeeping commands
   clear all; close all; clc;
 6
   getA = @(n) [exp(-n/50), 1; 2, cos(n/50)];
   getB = @(n) [exp(-n/50)*cos(n/50); 1];
   getC = @(n) [1+exp(-n/50), 2+sin(n/50)];
   getD = @(n) 1 + 0.5*sin(n/50);
11
12 % Initialize cells to store matrices
13 A = \{\}; B = \{\}; C = \{\}; D = \{\};
14
15
   % Initialize the covariance matrix Q
   Q = \{\}; Q\{1\} = zeros(2,2);
17
18
   % Initialize the x—states and xhat—states
   x = \{\}; x\{1\} = zeros(2,1);
   xhat = {}; xhat{}1} = zeros(2,1);
20
21
22
   % Initialize the u(n) and v(n) white noise
23
   rng(1000); u = randn(1,20);
24
   rng(2000); v = randn(1,20);
25
26 % Initialize the output states y(n)
27
   y = \{\};
28
29 % Initialize the value we are looking for
30 | res = {};
```

```
31
32
   for n = 0:10
33
        A{n+1} = getA(n);
34
        B\{n+1\} = getB(n);
35
        C{n+1} = getC(n);
36
        D{n+1} = getD(n);
37
        % Compute x(n+2) which is actually x(n+1)
38
        x\{n+2\} = A\{n+1\}*x\{n+1\} + B\{n+1\}*u(n+1);
39
        % Compute y(n)
40
        y{n+1} = C{n+1} * x{n+1} + D{n+1}*v(n+1);
41
        [xhat\{n+2\}, Q\{n+2\}, res\{n+1\}] = dkf(A\{n+1\}, B\{n+1\}, C\{n+1\}, D\{n+1\}, ...
42
                                                  Q{n+1}, xhat{n+1}, y{n+1});
43
   end
44
   %%
45
   function [Xnew, Qnew, XX] = dkf(Ad, Bd, Cd, Dd, Qd, xhat, y)
46
        % Discrete time Kalman filter
47
        Del = Ad*Qd*Cd' * inv(Cd*Qd*Cd' + Dd*Dd');
        Xnew = Ad*xhat + Del*(y - Cd*xhat);
48
49
        Qnew = Ad*Qd*Ad' + Bd*Bd' - Del*Cd*Qd*Ad';
50
        % What we are looking for
51
52
        XX = xhat + Qd*Cd' * inv(Cd*Qd*Cd' + Dd*Dd')*(y - Cd*xhat);
53
   end
```

This gives us the following results

$$P_{\mathcal{M}_n} x_2(8) = -48.0371$$
  $P_{\mathcal{M}_n} x_2(9) = -111.7608$   $P_{\mathcal{M}_n} x_2(10) = -260.9925$ 

Let x be a mean zero, variance one, Gaussian random variable, and  $\{v_n\}_0^{\infty}$  be a mean zero, variance one, Gaussian white noise process, which is independent of x. Consider the discrete time random process  $y_n$  defined by

$$y_n = x + v_n$$

where  $n \geq 0$  is a positive integer.

(i) Find best estimate for x given  $\{y_j\}_{j=0}^{n-1}$ , that is, find

$$\hat{x}_n = E(x|y_0, y_1, \cdots, y_{n-1}) = P_{\mathcal{M}_{n-1}}x$$

where  $\mathcal{M}_{n-1} = span\{y_j\}_{j=0}^{n-1}$ .

(ii) Find the error  $\sigma_n$  in your estimate, that is

$$\sigma_n^2 = E(x - \hat{x}_n)^2$$

#### **Solution:**

(i) We are given that

$$Ex = Ev_n = 0 Ex^2 = Ev_n^2 = 1.$$

If we let

$$g = \begin{bmatrix} 1 \\ y_0 \\ y_1 \\ \vdots \\ y_{n-1} \end{bmatrix}$$

then,

$$\hat{x}_n = R_{xg} R_g^{-1} g.$$

Now,

$$R_{xg} = Ex \begin{bmatrix} 1 & y_0 & y_1 & \cdots & y_{n-1} \end{bmatrix}$$
$$= \begin{bmatrix} Ex & Exy_0 & Exy_1 & \cdots & Exy_{n-1} \end{bmatrix}.$$

Since,

$$Exy_m = Ex(x + v_m) = Ex^2 + ExEv_m = 1.$$

Thus,

$$R_{xg} = \begin{bmatrix} 0 & 1 & 1 & \cdots & 1 \end{bmatrix}_{1 \times (n+1)}.$$

Then we consider,

$$Ey_n y_m = E(x + v_n)(x + v_m)$$

$$= Ex^2 + Ev_n Ex + Ev_m Ex + 0$$

and here

$$Ev_n v_m = \begin{cases} 1 & \text{if} & n = m \\ 0 & \text{if} & n \neq m \end{cases}.$$

Therefore,

$$Ey_n y_m = \begin{cases} 2 & \text{if} \quad n = m \\ 1 & \text{if} \quad n \neq m \end{cases}.$$

Then,

$$R_g = Eg = \begin{bmatrix} 1 \\ y_0 \\ y_1 \\ \vdots \\ y_{n-1} \end{bmatrix} \begin{bmatrix} 1 & y_0 & y_1 & \cdots & y_{n-1} \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 0 & 0 & \cdots & 0 \\ 0 & 2 & 1 & \cdots & 1 \\ 0 & 1 & 2 & \ddots & \vdots \\ \vdots & \vdots & \ddots & \ddots & 1 \\ 0 & 1 & \cdots & 1 & 2 \end{bmatrix}_{(n+1)\times(n+1)}.$$

Listing 3: Problem 3 MATLAB code

```
% AAE 567 Final Exam Spring 2021 Problem 3
2
   % Tomoki Koike
3
4
   % Housekeeping commands
5
   clear all; close all; clc;
6
   %%
   % Problem 3
   Rq = [];
9
   sz = 4;
10
   for i = 1:sz
        for j = 1:sz
11
12
            if (i==1) && (j==1)
13
                Rg(i,j) = 1;
14
            elseif (i==1) || (j==1)
15
                Rg(i,j) = 0;
16
            elseif i==j
17
                Rg(i,j) = 2;
18
            else
19
                Rg(i,j) = 1;
20
            end
21
        end
22
   end
23
   Rxg = ones(1,sz);
   Rxg(1) = 0;
24
25
   inv(Rg)
26
   coefs = Rxg * inv(Rg)
```

Now running the MATLAB code shown in Listing 3, we can numerically deduce that

$$R_g^{-1} = \begin{bmatrix} 1 & 0 & 0 & \cdots & 0 \\ 0 & \frac{n}{n+1} & -\frac{1}{n+1} & \cdots & -\frac{1}{n+1} \\ 0 & -\frac{1}{n+1} & \frac{n}{n+1} & \ddots & \vdots \\ \vdots & \vdots & \ddots & \ddots & -\frac{1}{n+1} \\ 0 & -\frac{1}{n+1} & \cdots & -\frac{1}{n+1} & \frac{n}{n+1} \end{bmatrix}_{(n+1)\times(n+1)}$$

Then,

$$R_{xg}R_g^{-1} = \begin{bmatrix} 0 & 1 & 1 & \cdots & 1 \end{bmatrix} \begin{bmatrix} 1 & 0 & 0 & \cdots & 0 \\ 0 & \frac{n}{n+1} & -\frac{1}{n+1} & \cdots & -\frac{1}{n+1} \\ 0 & -\frac{1}{n+1} & \frac{n}{n+1} & \ddots & \vdots \\ \vdots & \vdots & \ddots & \ddots & -\frac{1}{n+1} \\ 0 & -\frac{1}{n+1} & \cdots & -\frac{1}{n+1} & \frac{n}{n+1} \end{bmatrix}$$

$$= \begin{bmatrix} 0 & h & h & \cdots & h \end{bmatrix}$$

where

$$h = \frac{n}{n+1} + \left(-\frac{1}{n+1}\right)(n-1)$$

$$= \frac{n}{n+1} - \frac{n-1}{n+1}$$

$$= \frac{1}{n+1}.$$

Therefore,

$$\hat{x}_n = \begin{bmatrix} 0 & \frac{1}{n+1} & \frac{1}{n+1} & \cdots & \frac{1}{n+1} \end{bmatrix} \begin{bmatrix} 1 \\ y_0 \\ y_1 \\ \vdots \\ y_{n-1} \end{bmatrix}$$

and

$$\hat{x}_n = \frac{1}{n+1}(y_0 + y_1 + \dots + y_{n-1}).$$

(ii) Next we will find the error.

$$E\hat{x}_n^2 = \frac{1}{(n+1)^2} E(y_0 + y_1 + \dots + y_{n-1})^2$$

$$= \frac{1}{(n+1)^2} \Big[ \sum_{i=0}^{n-1} Ey_i^2 + \sum_{i=0, i>j}^{n-1} \sum_{j=0}^{n-1} 2Ey_i y_j \Big]$$

$$= \frac{1}{(n+1)^2} \Big[ \sum_{i=0}^{n-1} 2 + 2\binom{n}{2} \Big]$$

$$= \frac{2}{(n+1)^2} \Big( n + \frac{n!}{(n-2)!2!} \Big)$$

$$= \frac{2}{(n+1)^2} \Big( n + \frac{n(n-1)}{2} \Big)$$

$$= \frac{1}{(n+1)^2} \Big[ n(n+1) \Big]$$

$$= \frac{n}{n+1}.$$

Finally,

$$E(x-\hat{x})^2 = Ex^2 - E\hat{x}^2 = 1 - \frac{n}{n+1} = \frac{1}{n+1}.$$

Hence,

$$\sigma_n^2 = \frac{1}{n+1}.$$

Consider the unstable state space system given by

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & -0.2 & 3 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & -0.5 & 30 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \\ 0 \\ 5 \end{bmatrix} w + \begin{bmatrix} 0 & 0 \\ 0.25 & 0 \\ 0 & 0 \\ 0 & 0.1 \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

$$y_1 = x_1 + 0.25v_1$$
$$y_2 = x_3 + 0.05v_2$$

where  $u_1$ ,  $u_2$ ,  $v_1$  and  $v_2$  are all independent white noise processes. Moreover, w is the input. Assume that all the initial conditions are zero. Design a feedback controller  $w = -K\hat{x}$  based on the steady state Kalman filter such that  $|x_1(t)| \le 1$  and  $|x_3(t)| \le 0.35$ . Your state feedback gain  $K = \begin{bmatrix} k_1 & k_2 & k_3 & k_4 \end{bmatrix}$  must statisfy  $|k_j| \le 25$  for j = 1, 2, 3, 4. Simulate you controller in SIMULINK for 30 seconds. Hand the graphs from your SIMULINK program for:

- (i) The state  $x_1$  and its estimate  $\hat{x}_1$  on the same graph.
- (ii) The state  $x_2$  and its estimate  $\hat{x}_2$  on the same graph.
- (iii) The state  $x_3$  and its estimate  $\hat{x}_3$  on the same graph.
- (iv) The state  $x_4$  and its estimate  $\hat{x}_4$  on the same graph.
- (v) Hand in your gain K.

On the bank limited white noise generators, set the seed for  $u_1$ ,  $u_2$ ,  $v_1$ , and  $v_2$  respectively to, 22341, 23342, 23343, and 23344.

#### **Solution:**

The system matrices are defined to be

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & -0.2 & 3 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & -0.5 & 30 & 0 \end{bmatrix} \qquad B_0 = \begin{bmatrix} 0 & 0 \\ 0.25 & 0 \\ 0 & 0 \\ 0 & 0.1 \end{bmatrix} \qquad B_1 = \begin{bmatrix} 0 \\ 1 \\ 0 \\ 5 \end{bmatrix}$$

$$C = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 \end{bmatrix} \qquad D = \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix}.$$

Now in MATLAB using the lqr() command we obtain the following gain matrix K

$$K = \begin{bmatrix} -3.1623 & -7.3031 & 24.8433 & 6.9672 \end{bmatrix}$$
.

This gain satisfies the condition of having magnitudes smaller than 25. Then since the system is observable and controllable we can use the are() to find a positive definite matrix P which is a solution to an algebraic Ricatti equation.

$$P = \begin{bmatrix} 0.0792 & 0.0505 & 0.0013 & 0.0025 \\ 0.0505 & 0.0771 & 0.0133 & 0.0668 \\ 0.0013 & 0.0133 & 0.0273 & 0.1490 \\ 0.0025 & 0.0668 & 0.1490 & 0.8152 \end{bmatrix}.$$

Then the observer gain L becomes  $L = PC^*(DD^*)^{-1}$  which is as follows.

$$L = \begin{bmatrix} 1.2671 & 0.5179 \\ 0.8081 & 5.3104 \\ 0.0207 & 10.9191 \\ 0.0400 & 59.6184 \end{bmatrix}$$

Below we can see that the negative real eigenvalues indicate a stable system with the controller.

$$eig(A - B_1K) \bigcup eig(A - LC) = \begin{bmatrix} -23.9934 \\ -0.8102 \\ -1.4647 + 0.5430 \, \mathrm{i} \\ -1.4647 - 0.5430 \, \mathrm{i} \\ -0.7126 + 0.7022 \, \mathrm{i} \\ -0.7126 - 0.7022 \, \mathrm{i} \\ -5.4805 + 0.1759 \, \mathrm{i} \\ -5.4805 - 0.1759 \, \mathrm{i} \end{bmatrix}$$

Next we define new system matrices for the states and the estimations. The matrices for the actual states become

$$A_s = A$$
  $B_s = [B_0, B_1]$   $C_s = eye(4)$   $D_s = zeros(size(C_s, 1), size(B_s, 2))$ 

and for the estimation

$$A_{kf} = A - LC$$
  $B_{kf} = [L, B_1]$   $C_{kf} = eye(4)$   $D_{kf} = zeros(size(C_{kf}, 1), size(B_{kf}, 2)).$ 

Based on this we create the following SIMULINK model in Fig 1 for the simulation

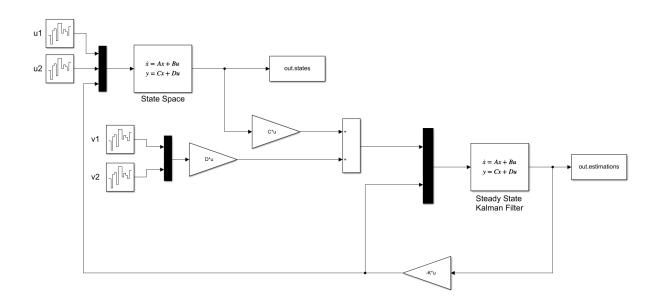


Figure 1: Simulink model for problem 4

Then we run the following MATLAB code.

Listing 4: Problem 4 MATLAB code

```
% AAE 567 Final Exam Spring 2021 Problem 4
 2
    % Tomoki Koike
 3
    % Housekeeping commands
 4
 5
    clear all; close all; clc;
 6
    % Define system matrices
    A = [0 \ 1 \ 0 \ 0; \ 0 \ -0.2 \ 3 \ 0; \ 0 \ 0 \ 0 \ 1; \ 0 \ -0.5 \ 30 \ 0];
    B0 = [0 \ 0; \ 0.25 \ 0; \ 0 \ 0; \ 0 \ 0.1];
10 \mid B1 = [0; 1; 0; 5];
    C = [1 \ 0 \ 0 \ 0; \ 0 \ 0 \ 1 \ 0];
11
    D = [0.25 \ 0; \ 0 \ 0.05];
12
13
14
    % Obtain the K gain with LQR
15
    K = lqr(A, B1, diag([1,2,4,2]), diag([0.1]));
16
    sym(K)
17
18 % Find L gain
19 rank(obsv(A,C))
20 rank(ctrb(A,B0))
21 P = are(A', C'*inv(D*D')*C, B0*B0');
```

```
22 | sym(P)
23 L = P*C'*inv(D*D');
24 sym(L)
25
26 % Check the eigenvalues
27 \mid sym([eig(A-B1*K); eig(A-L*C)])
28
29 % System matrices for the actual states
30 | As = A;
31 | Bs = [B0, B1];
32 | Cs = eye(4);
33 |Ds = zeros(size(Cs,1), size(Bs,2));
34
35 |% System matrices for the steady state Kalman filter states
36 \mid Akf = A - L*C;
37 | Bkf = [L, B1];
38 \mid Ckf = eye(4);
39 Dkf = zeros(size(Ckf,1), size(Bkf,2));
40 %%
41 % Plotting results
42 | set(groot, 'defaulttextinterpreter', 'latex');
43 | set(groot, 'defaultAxesTickLabelInterpreter', 'latex');
44 | set(groot, 'defaultLegendInterpreter', 'latex');
45
46 % Simulate
47 | simout = sim('final_p4_ssKF.slx');
48
49 % Data rendering
50 | xs = simout.states.signals.values;
51 | xkf = simout.estimations.signals.values;
52 \mid t = simout.tout;
53
54 % Plot
55 | fig = figure("Renderer", "painters", "Position", [60 60 900 1000]);
56
        subplot(4,1,1)
57
        plot(t, xs(:,1))
58
        grid on; grid minor; box on; hold on;
59
        plot(t, xkf(:,1))
60
        hold off;
61
        ylabel('$x_1$')
62
        legend('states', 'estimate')
63
64
        subplot(4,1,2)
65
        plot(t, xs(:,2))
66
        grid on; grid minor; box on; hold on;
```

```
plot(t, xkf(:,2))
67
68
       hold off;
69
       ylabel('$x_2$')
70
71
       subplot(4,1,3)
72
       plot(t, xs(:,3))
73
       grid on; grid minor; box on; hold on;
74
       plot(t, xkf(:,3))
75
       hold off;
       ylabel('$x_3$')
76
77
78
       subplot(4,1,4)
79
       plot(t, xs(:,4))
       grid on; grid minor; box on; hold on;
80
81
       plot(t, xkf(:,4))
82
       hold off;
       ylabel('$x_4$')
83
       xlabel('time [sec]')
84
  saveas(fig, 'final_p4_plot.png')
85
```

And this yields the results on the next page.

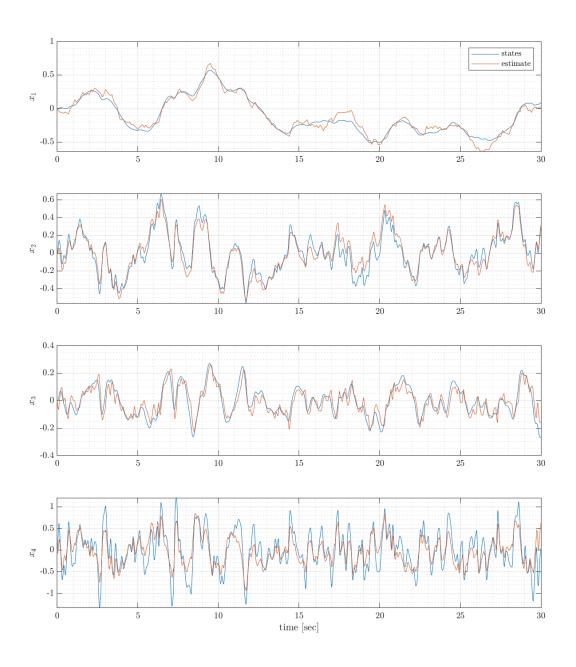


Figure 2: States and Kalman filter estimations for all 4 states