



College of Engineering
School of Aeronautics and Astronautics

AAE 36401 Lab
Control Systems Lab

Lab 2 Report
The Control of Gantry

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Introduction

Objective

This experiment setup has a cart with a mass of M on top of a track that spans one dimensionally. This cart has a pendulum attached to the cart that swings to apply perturbations on the cart. The objective of this lab is to control the cart so that the cart moves to halt the swinging pendulum. This resembles a functionality of a gantry or crane that are used in many applications.

Method

In the first part of the experiment, the natural frequency of the pendulum was measured experimentally by swinging the pendulum with a small angle while holding the cart still. The sinusoidal oscillations plotted in the scope will give the experimental natural frequency. Meanwhile, the theoretical natural frequency will be computed using the equation of motion (EOM). The EOM will be linearized (assuming that the initial angle of the pendulum is small) and manipulated to give the theoretical formula of the time derivative of the pendulum's angle.

In the second part of the experiment, four gain values K_1 , K_2 , K_3 , and K_4 were input to the feedback system to control the cart to terminate the swing of the pendulum with desired response parameters. The gains were generated using the pole placement method and fed to the Simulink model to simulate the results.

The pole placement method is possible when for a state space system of

$$\dot{x} = Ax + Bv$$

Is defined where the matrices $\{A, B\}$ are controllable. Thereby, a state feedback vector K exists, and $A-BK$ has the eigenvalues of $(\lambda_j)_1^n$ which correspond to the poles of the feedback system G_K .

$$G_K(s) = C(sI - (A_BK))^{-1}B = \frac{q(s)}{(s - \lambda_1)(s - \lambda_2) \cdots (s - \lambda_n)}$$

Knowing this, we are able to use the MATLAB command of `place()` to compute the feedback gains

$$K = \text{place}(A, B, [\lambda_1, \lambda_2, \cdots, \lambda_n])$$

The experiment setup is shown in an image in the appendix.

Results

Part (i)

In this part of the experiment, the natural frequency was computed both theoretically and experimentally.

Theoretical Method

Assuming that the cart is fixed onto the track, the equation of motion becomes

$$(I_p + M_p l_p^2) \ddot{\alpha} + M_p l_p g \sin(\alpha) = 0$$

Now, considering the small angle α , we can change $\sin(\alpha) \approx \alpha$. This linearizes the EOM to

$$(I_p + M_p l_p^2) \ddot{\alpha} + M_p l_p g \alpha = 0$$

Thus, the EOM is boiled down to the second order differential equation of an undamped harmonic oscillator

$$\ddot{\alpha} + \omega_p^2 \alpha = 0$$

$$\text{where } \omega_p = \sqrt{\frac{M_p l_p g}{I_p + M_p l_p^2}}$$

Plugging in the constants identified in the “Notations for Variables” in the appendix, we get the following theoretical natural frequency

$$\omega_p = 4.7543 \text{ rad/s}$$

This is equivalent to the period of

$$\tau = \frac{2\pi}{\omega_p} = 1.3216 \text{ s}$$

Experimental Method

From the data obtained in the first part of the experiment, the following graph can be plotted

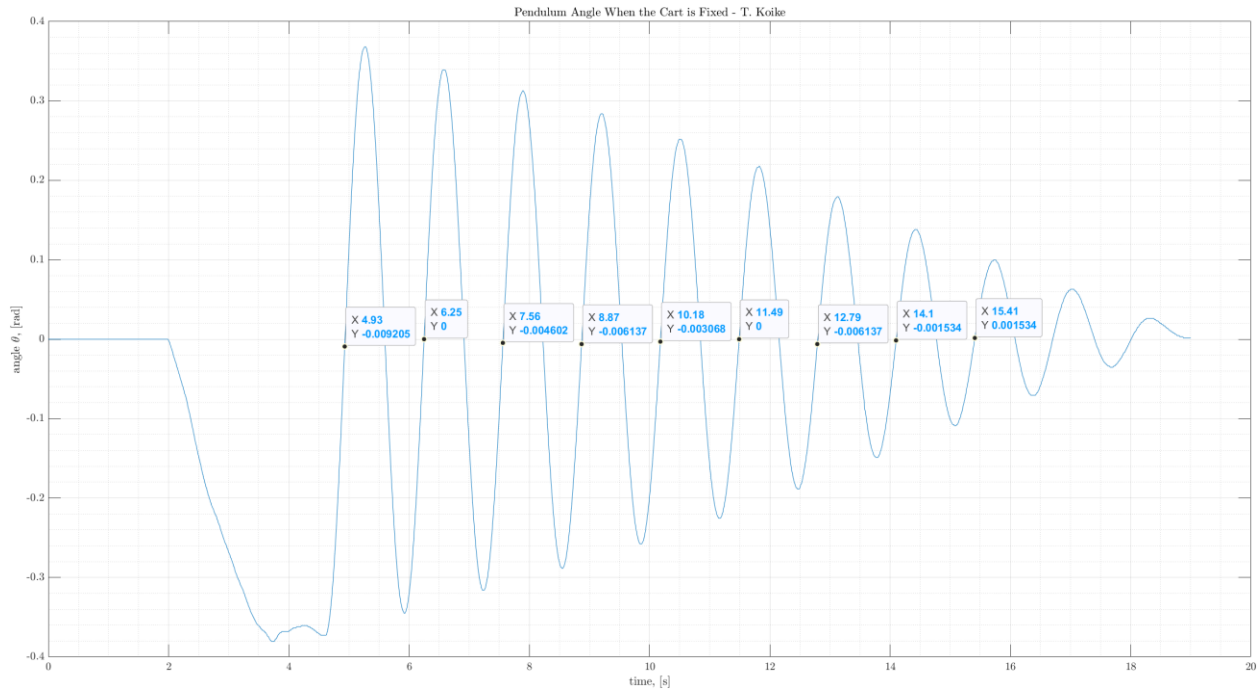


Figure 1: angle oscillation plot from part (i) of the experiment

From this graph, we can see the t-values (or time values) when the angle periodically goes to zero. The values are tabulated below.

Table 1: organization of the periods for each cycle

Cycle	start [s]	end [s]	period [s]
1	4.93	6.25	1.32
2	6.25	7.56	1.31
3	7.56	8.87	1.31
4	8.87	10.18	1.31
5	10.18	11.49	1.31
6	11.49	12.79	1.30
7	12.79	14.10	1.31
8	14.10	15.41	1.31
Average Period			1.31

Thus, the experimental period becomes 1.31 and the corresponding natural frequency is

$$\omega_p = \frac{2\pi}{1.31} = 4.7963 \text{ rad/s}$$

Part (iii)

Pre-Lab Results

The values of the gains, K from the pre-lab are the following

Table 2: gains from the pole placement obtained from the pre-lab

K	Gains
K_1	38.9466
K_2	-33.0734
K_3	14.9005
K_4	2.9887

The corresponding pole values are

Table 3: poles used for the pole placement for the pre-lab

poles
$-3+2.8i$
$-3-2.8i$
-8
-10

The plot of the position x_c is the following

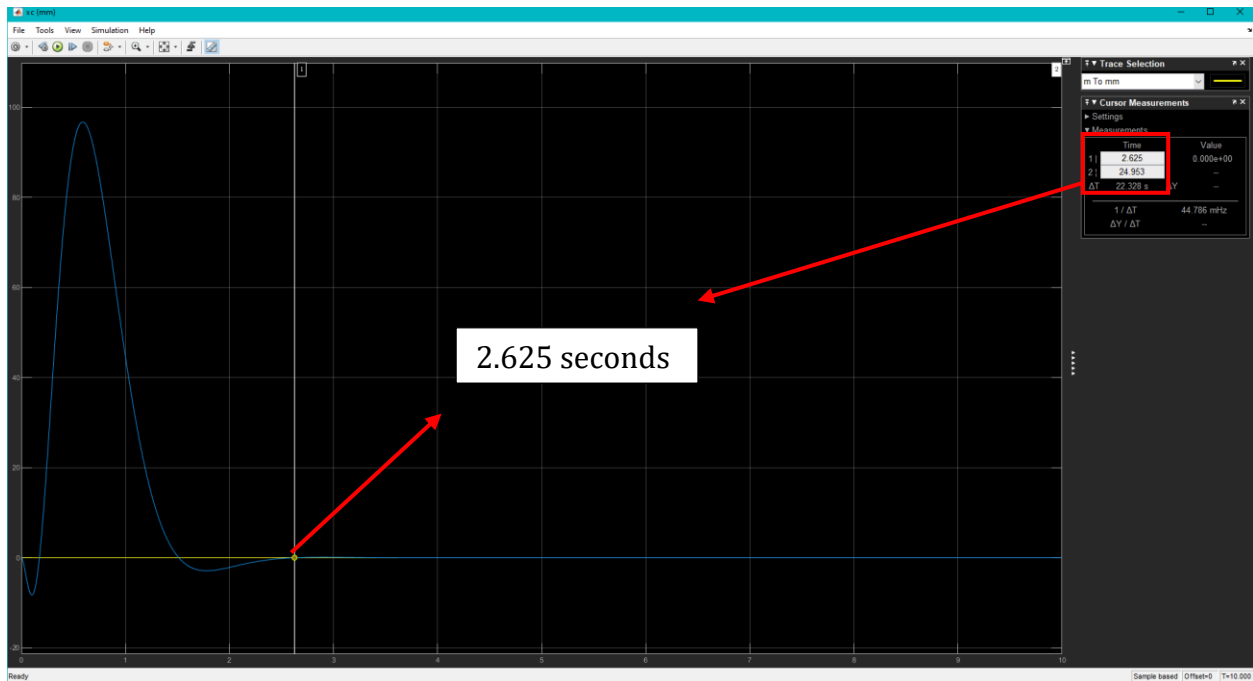


Figure 2: x_c response for the pre-lab

The angle, α response becomes the following,

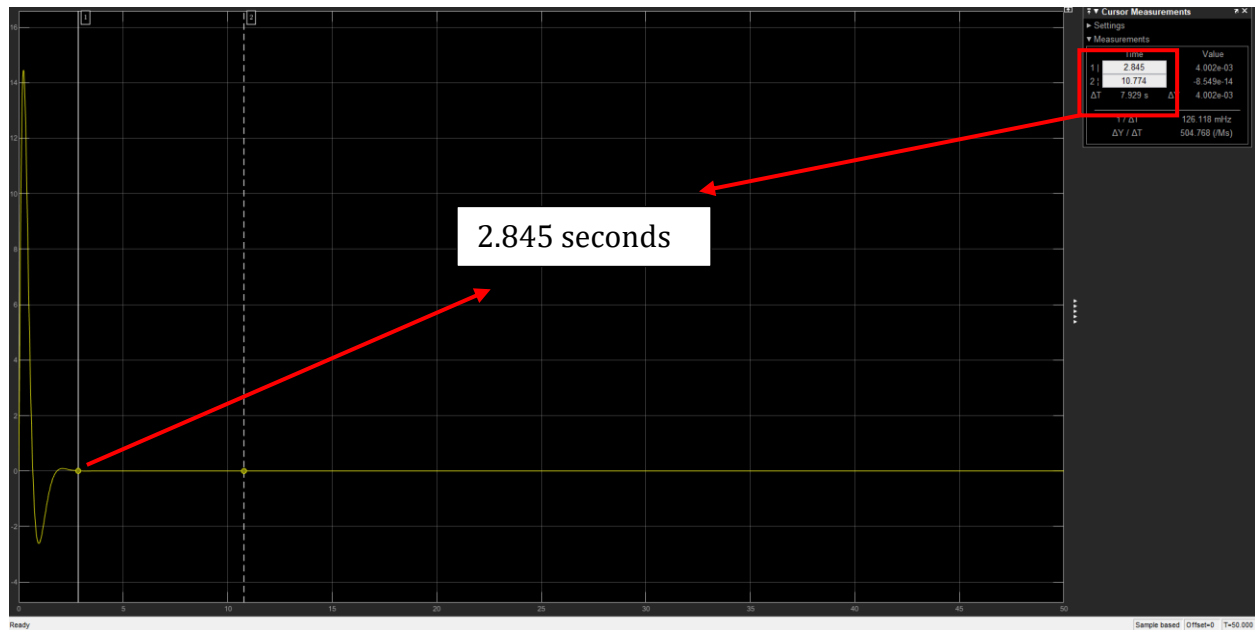


Figure 3: angle response for the pre-lab

This indicates that settling time t_s for when $d\alpha/dt(0) = \pi/2$ is

$$t_s = 2.845 \text{ s}$$

Experiment Results

In the experiment, the exact same gains and poles from the pre-lab were used.

Table 4: gains for the experiment

K	Gains
K_1	38.9466
K_2	-33.0734
K_3	14.9005
K_4	2.9887

Table 5: poles for the experiment

poles
-3+2.8i
-3-2.8i
-8
-10

The displacement response was the following

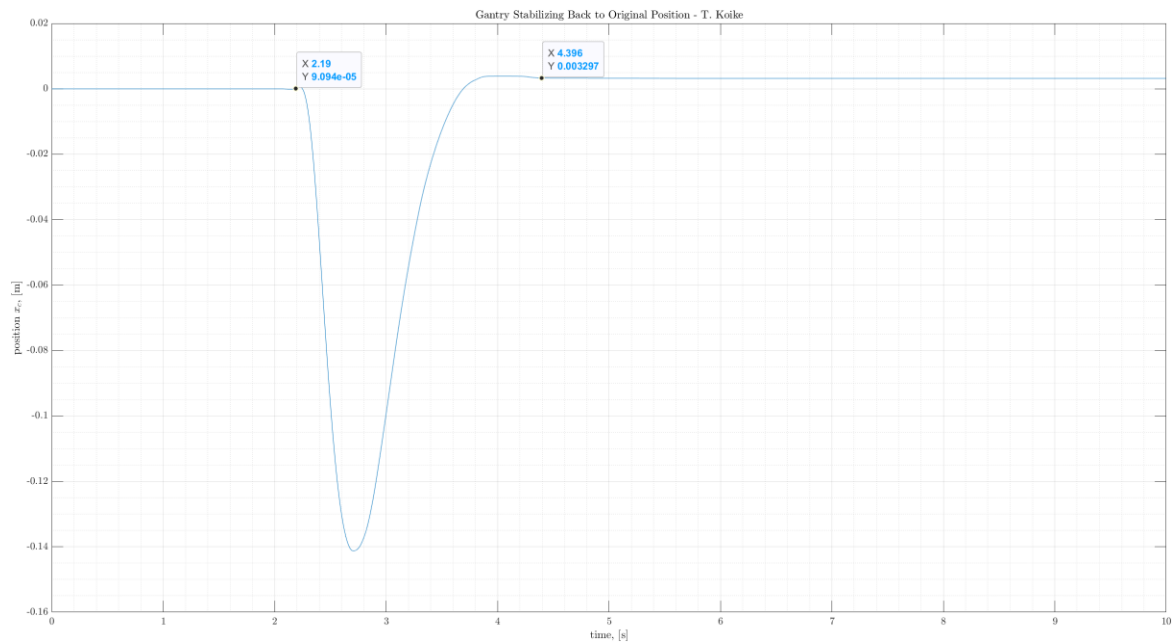


Figure 4: x_c response for the experiment

The angle response is the following

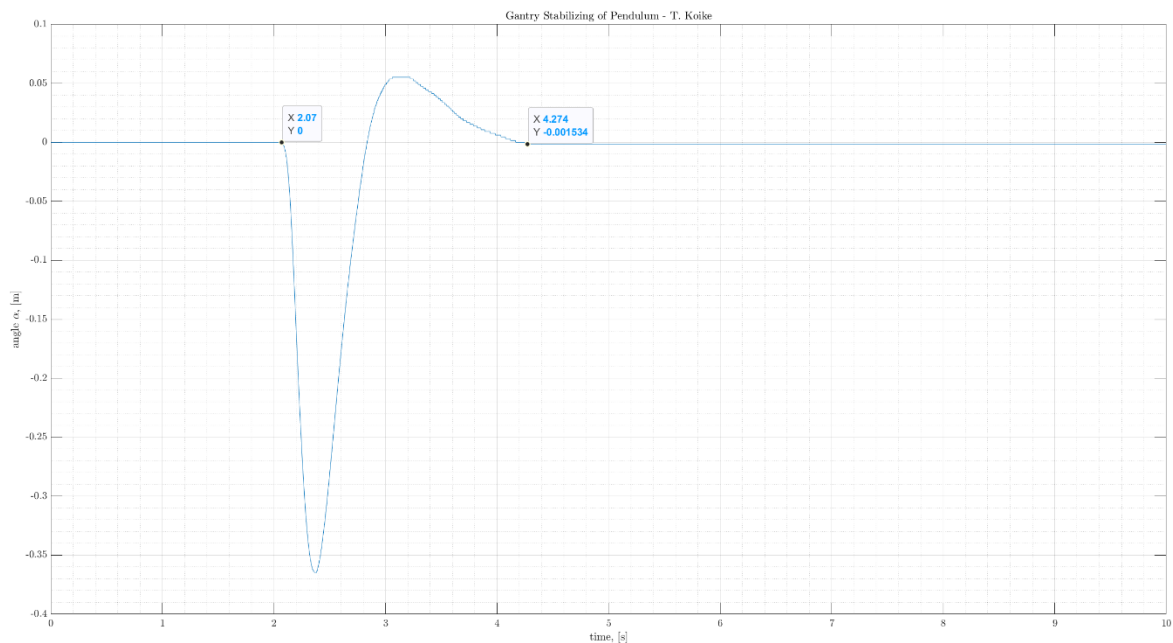


Figure 5: angle response for the experiment

From the two points indicated on the graph we can figure out the settling time, t_s to be

$$t_s = 4.274 - 2.07 = 2.204 \text{ s}$$

Analysis & Discussions

Nonlinear EOMs

The EOM of the cart on the track with a pendulum with force exerted by the servo motor is expressed as the following system equation.

$$\begin{aligned}
 & ((M_c + M_p)I_p + M_c M_p l_p^2 + M_p^2 l_p^2 \sin(\alpha)^2) \ddot{x}_c + (I_p + M_p l_p^2) B_{eq} \dot{x}_c \\
 & = M_p l_p B_p \cos(\alpha) \dot{\alpha} + (M_p^2 l_p^3 + I_p M_p l_p) \sin(\alpha) \dot{\alpha}^2 + (I_p + M_p l_p^2) F_c + M_p^2 l_p^2 g \cos(\alpha) \sin(\alpha); \\
 & ((M_c + M_p)I_p + M_c M_p l_p^2 + M_p^2 l_p^2 \sin(\alpha)^2) \ddot{\alpha} + (M_c + M_p) B_p \dot{\alpha} \\
 & = M_p l_p \cos(\alpha) B_{eq} \dot{x}_c - (M_c + M_p) M_p g l_p \sin(\alpha) - M_p^2 l_p^2 \sin(\alpha) \cos(\alpha) \dot{\alpha}^2 - M_p l_p \cos(\alpha) F_c; \\
 F_c & = \frac{\eta_g K_g \eta_m K_t (v r_{mp} - K_g K_m \dot{x}_c)}{R_m r_{mp}^2}.
 \end{aligned}$$

This system equation is highly nonlinear and complicated with terms such as the voltage of the servo motor, angle of pendulum, force exerted on the cart via the motor, viscous damping of the motor, etc. In order to solve this system analytically, it is essential to linearize the system equations.

Linearized EOMs & Equilibrium Points

Solving these nonlinear equations for the high order differential terms while linearizing them gives the following system equations

$$\begin{aligned}
 \ddot{x}_c & = \frac{-(I_p + M_p l_p^2) B_{eq} \dot{x}_c + M_p l_p B_p \dot{\alpha} + M_p^2 l_p^2 g \alpha + (I_p + M_p l_p^2) F_c}{(M_c + M_p) I_p + M_c M_p l_p^2} \\
 \ddot{\alpha} & = \frac{M_p l_p B_{eq} \dot{x}_c - (M_c + M_p) B_p \dot{\alpha} - (M_c + M_p) M_p g l_p \alpha - M_p l_p F_c}{(M_c + M_p) I_p + M_c M_p l_p^2} \\
 F_c & = \frac{\eta_g K_g \eta_m K_t (v r_{mp} - K_g K_m \dot{x}_c)}{R_m r_{mp}^2}.
 \end{aligned}$$

From the nonlinear EOMs we can find the following relationships

$$\ddot{x}_c = \frac{M_p l_p B_p \cos(\alpha) \dot{\alpha} + (M_p^2 l_p^3 + I_p M_p l_p) \sin(\alpha) \dot{\alpha}^2 + (I_p + M_p l_p^2) F_c + M_p^2 l_p^2 g \cos(\alpha) \sin(\alpha) - (I_p + M_p l_p^2) B_{eq} \dot{x}_c}{(M_c + M_p) I_p + M_c M_p l_p^2 + M_p^2 l_p^2 \sin^2(\alpha)}$$

$$\ddot{\alpha} = \frac{M_p l_p \cos(\alpha) B_{eq} \dot{x}_c - (M_c + M_p) M_p g l_p \sin(\alpha) - M_p^2 l_p^2 \sin(\alpha) \cos(\alpha) \dot{\alpha}^2 - M_p l_p \cos(\alpha) F_c - (M_c + M_p) B_p \dot{\alpha}}{(M_c + M_p) l_p + M_c M_p l_p^2 + M_p^2 l_p^2 \sin^2(\alpha)}$$

$$F_c = \frac{\eta_g K_g \eta_m K_t (v r_{mp} - K_g K_m \dot{x}_c)}{R_m r_{mp}^2}$$

Set the input as the voltage $u = v$, and the output equal to the position of the cart $y = x_c$. Then we model a simple Simulink model.

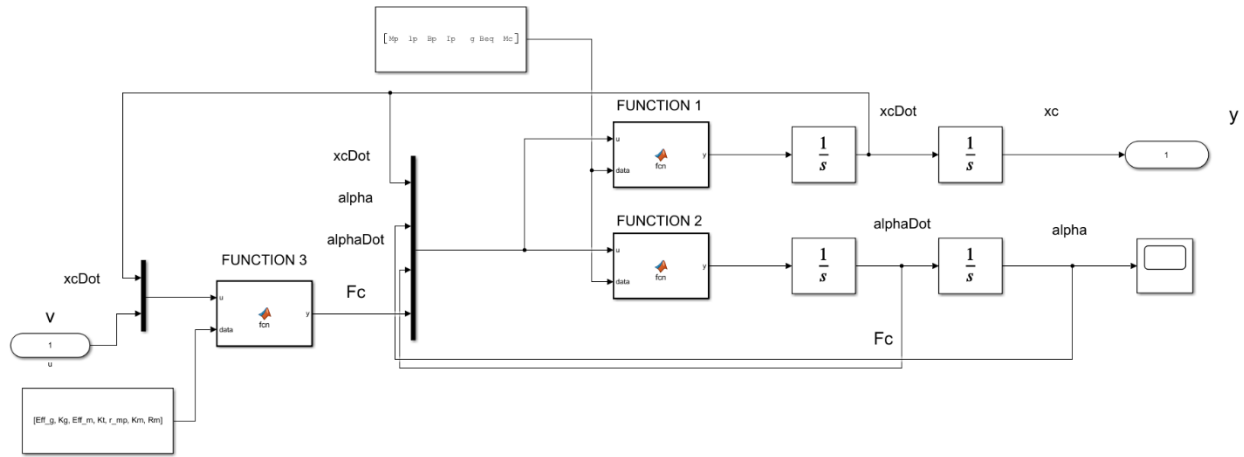


Figure 6: Simulink model originally made to find the equilibrium points and system matrices

The embedded MATLAB blocks in the model have the following functions

FUNCTION 1:

```
function y = fcn(u, data)
% FUNCTION 1
Mp = data(1); lp = data(2); Bp = data(3); Ip = data(4); g = data(5);
Beq = data(6); Mc = data(7);
xcdot = u(1); a = u(2); adot = u(3); Fc = u(4);

den = (Mc + Mp)*Ip + Mc*Mp*lp^2 + Mp^2*lp^2*sin(a)^2;
num1 = Mp*lp*Bp*cos(a)*adot;
num2 = (Mp^2*lp^3 + Ip*Mp*lp)*sin(a)*adot^2;
num3 = (Ip + Mp*lp^2)*Fc;
num4 = Mp^2*lp^2*g*cos(a)*sin(a);
num5 = -(Ip + Mp*lp^2)*Beq*xcdot;

y = (num1 + num2 + num3 + num4 + num5) / den;
end
```

FUNCTION 2:

```
function y = fcn(u, data)
% FUNCTION 2
Mp = data(1); lp = data(2); Bp = data(3); Ip = data(4); g = data(5);
Beq = data(6); Mc = data(7);
xcdot = u(1); a = u(2); adot = u(3); Fc = u(4);

den = (Mc + Mp)*Ip + Mc*Mp*lp^2 + Mp^2*lp^2*sin(a)^2;
num1 = Mp*lp*cos(a)*Beq*xcdot;
num2 = -(Mc + Mp)*Mp*g*lp*sin(a);
num3 = -Mp^2*lp^2*sin(a)*cos(a)*adot^2;
num4 = -Mp*lp*cos(a)*Fc;
num5 = -(Mc + Mp)*Bp*adot;

y = (num1 + num2 + num3 + num4 + num5) / den;
end
```

FUNCTION 3:

```
function y = fcn(u, data)
% FUNCTION 3
Eff_g = data(1); Kg = data(2);
Eff_m = data(3); Kt = data(4); r_mp = data(5);
Km = data(6); Rm = data(7);
xcdot = u(1); v = u(2);

den = Rm*r_mp^2;
num = Eff_g*Kg*Eff_m*Kt * (v*r_mp - Kg*Km*xcdot);

y = num / den;
end
```

With this model we use the `trim('sys', X0)` (code is in the appendix) to find the equilibrium points of the system variables. (*X0 is equal to where $\dot{\alpha}(0) = \pi/2$ rad/s and everything else is 0). The raw MATLAB results are the following.

Table 6: raw output of the equilibrium points from MATLAB

$\mathbf{x_e} = 4 \times 1$ $10^{-23} \times$ 0.0000 0 0.0009 0.1654	$\mathbf{dx_e} = 4 \times 1$ $10^{-23} \times$ 0.0009 0.1654 0.0065 -0.0271
$\mathbf{u_e} = 1.0340\text{e-}25$	$\mathbf{y_e} = 1.1638\text{e-}38$

The results are organized to be.

Table 7: organized equilibrium points

Variable	Equilibrium Point
x_c^e	~ 0
\dot{x}_c^e	~ 0
α^e	~ 0
$\dot{\alpha}^e$	~ 0
F_c^e	~ 0
v^e	~ 0

State Variables, State Vector, System Matrices, and Transfer Function

Defining the state variables as,

$$\mathbf{x} = \begin{pmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{pmatrix} = \begin{pmatrix} x_c \\ \alpha \\ \dot{x}_c \\ \dot{\alpha} \end{pmatrix}$$

The state space of this system is

$$\begin{pmatrix} \dot{\mathbf{x}} \\ \mathbf{y} \end{pmatrix} = \begin{pmatrix} \mathbf{A}\mathbf{x} + \mathbf{B}\mathbf{v} \\ \mathbf{C}\mathbf{x} + \mathbf{D}\mathbf{v} \end{pmatrix}$$

Here,

$$\mathbf{y} = x_c$$

Then, we use these equilibrium points to compute the A, B, C, and D matrices of the linearized model based on those equilibrium points. The raw output of A, B, C, and D are the following

Table 8: raw output of the sysmte matrices from MATLAB

$\mathbf{A} = 4 \times 4$ <pre> 0 0 1.0000 0 0 0 0 1.0000 0 1.5216 -11.6513 0.0049 0 -26.1093 26.8458 -0.0841 </pre>	$\mathbf{B} = 4 \times 1$ <pre> 0 0 1.5304 -3.5261 </pre>
$\mathbf{C} = 1 \times 4$ <pre> 1 0 0 0 </pre>	$\mathbf{D} = 0$

These values agree with the values on the provided lab manual. Now that we have verified the system matrices, use the `ss('sys')` command is used to create state space system. Next, convert the state space system to a transfer function using the `tf('sys')` command and obtain the following,

```
sys_tf =
      1.53 s^2 + 0.1114 s + 34.59
      -----
      s^4 + 11.74 s^3 + 26.96 s^2 + 263.4 s
Continuous-time transfer function.
```

This also agrees with the transfer function provided by the lab manual. (*The full code is in the appendix).

Part (i)

For this part of the experiment we got somewhat close values for the theoretical and experimental natural frequencies.

$$\begin{aligned}\omega_{p,th} &= 4.7543 \text{ rad/s} \\ \omega_{p,exp} &= 4.7963 \text{ rad/s}\end{aligned}$$

The percent error of this is

$$\%error = \frac{|4.7963 - 4.7543|}{4.7543} \times 100 = 88.34\%$$

12% deviation is slightly above the bounds of an acceptable experimental result. We will look into why there is a difference between the two in the next conclusion section.

Part (iii)

For this part of the experiment, there was only one combination of gains, K used for the experiment. This is because the response for the gains obtained in the pre-lab was worked out very well for the experiment. Considering the Simulink models being different for the pre-lab and actual experiment, this was fortunate. The settling time for the experiment was faster by 0.641 seconds and showed a very smooth control of the swinging pendulum.

However, when observing figure 4 it is notable that there is a slight steady state error for the returning position of the cart. This steady state error was not visible in the pre-lab. The discussion of this will be elucidated in the next section.

State Feedback and Pole Placement

With the state feedback implemented on the closed loop system, we are able to force the state of the system to go to zero. This is why we can see how the system controls the pendulum quickly and smoothly.

When observing the poles in tables, it is evident that all the poles are located in the left-hand side of the complex plane. This indicates that the feedback system is stable and remains consistent with how the system is controllable. If any of the poles are located in the right-hand side of the complex plane, the system will become unstable and the results will become erroneous.

Conclusion & Recommendation

Main Points

For the first part of the experiment, the natural frequencies computed using the theoretical derivation from the EOM and the data from the actual swinging of the pendulum agreed with a 12% deviation. Ideally, it would be better to have a percent error of under 5%-10% but the results are not significantly bad.

For the second part of the experiment, the gantry system was tested using the gains obtained in the pre-lab and the results were very good. The settling time for the pendulum was more than 0.5 seconds quicker than the pre-lab. However, there was a slight error, which is the steady state error for the position of the cart. This is obvious that this error was produced when the angular momentum of the pendulum was offset by the slight momentum of the cart to stop the pendulum from swinging.

Theoretical/Experimental Limitations

In the first part of the experiment where the experimental natural frequency of the pendulum was measured, a probable cause for the discrepancy between the theoretical and experimental values is how the experimenter held the cart down on the track to keep it

still. The shaking of the hand could have been a factor of human error that changed the behavior of the pendulum. Meanwhile, the voltage input could also have been a potential cause of error.

From the results, it is reasonable to deduce the limitations of the cart having a steady state error of its position. As discussed, there has to a slight displacement in order to stop the pendulum from swinging. The tradeoff of reducing the settling time of the pendulum and smoothness of its transition is made by this displacement. Another limitation is the maximum possible values of the gains, K . Due to the providable maximum voltage input the gains cannot exceed a number of 200. The system does not necessarily need such high gains but still is a constraint set on the system.

Moreover, theoretically the system must be fed gains that have stable poles. This is crucial to the system and must be satisfied at all times.

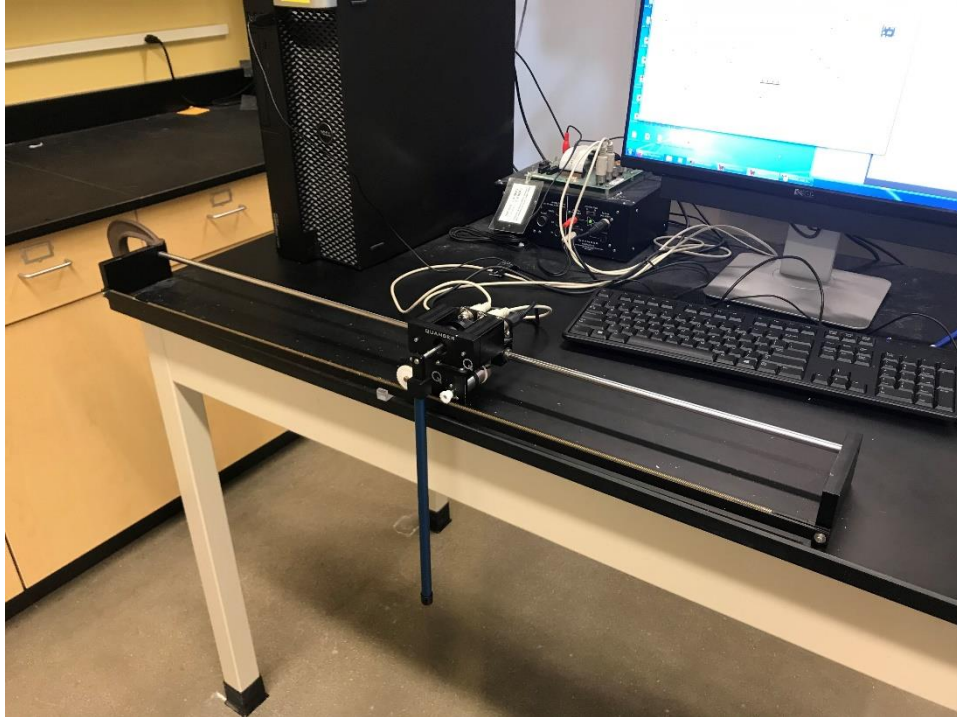
Lessons Learned & Suggestions for Improvement

For the part (i) of the experiment, it would have been better to clip the cart onto the track firmly than to hold it by hand. This will solve the issue of potential human errors.

From this experiment, the most valuable lesson learned was the determination of the gains from the pole placement and how it improved the systems performance by a significant degree. This preaches students to always consider the best method of a closed loop feedback system for future projects.

Appendix

Experiment Setup



Notations for Variables

Symbol	Description	Value	Unit
R_m	motor armature resistance	2.6	Ω
L_m	motor armature inductance	0.18	mH
K_t	motor torque constant	0.00767	$N.m/A$
η_m	motor efficiency	100%	%
K_m	back-electromotive-force(EMF) constant	0.00767	$V.s/rad$
J_m	rotor moment of inertia	3.9×10^{-7}	$kg.m^2$
K_g	planetary gearbox ratio	3.71	
η_g	planetary gearbox efficiency	100%	%
M_{c2}	cart mass	0.57	kg
M_w	cart weight mass	0.37	kg
M_c	total cart weight mass including motor inertia	1.0731	kg
B_{eq}	viscous damping at motor pinion	5.4000	$N.s/m$
L_t	track length	0.990	m
T_c	cart travel	0.814	m
P_r	rack pitch	1.664×10^{-3}	$m/tooth$
r_{mp}	motor pinion radius	6.35×10^{-3}	m
N_{mp}	motor pinion number of teeth	24	
r_{pp}	position pinion radius	0.01482975	m
N_{pp}	position pinion number of teeth	56	
KEP	cart encoder resolution	2.275×10^{-5}	$m/count$
M_p	long pendulum mass with T-fitting	0.230	kg
M_{pm}	medium pendulum mass with T-fitting	0.127	kg
L_p	long pendulum length from pivot to tip	0.6413	m
L_{pm}	medium pendulum length from pivot to tip	0.3365	m
l_p	long pendulum length: pivot to center of mass	0.3302	m
l_{pm}	medium pendulum length: pivot to center of mass	0.1778	m
J_p	long pendulum moment of inertia \odot center of mass	7.88×10^{-3}	$kg.m^2$
J_{pm}	medium pendulum moment of inertia \odot center of mass	1.20×10^{-3}	$kg.m^2$
B_p	viscous damping at pendulum axis	0.0024	$N.m.s/rad$
g	gravitational constant	9.81	m/s^2
v	voltage of servo motor	variable	V

MATLAB Code

Data Manipulation (loading & plotting)

```
% MATLAB CODE TO FIND THE NATURAL FREQUENCY FROM EXPERIMENTALLY
% TOMOKI KOIKE
clear all; close all; clc;
fdir = 'C:\Users\Tomo\Desktop\studies\2020-Fall\AAE364lab\matlab\lab2\outputs';
set(groot, 'defaulttextinterpreter','latex');
set(groot, 'defaultAxesTickLabelInterpreter','latex');
set(groot, 'defaultLegendInterpreter','latex');

% Load the experimental data
load("lab2-part1-theta-tomoki.mat");

% Plotting the angles
tspan = part1_theta.time;
thetaS = part1_theta.signals.values;

fig1 = figure(1);
plot(tspan, thetaS)
title('Pendulum Angle When the Cart is Fixed - T. Koike')
xlabel('time, [s]')
ylabel('angle $\theta$, [rad]')
grid on; grid minor; box on;
```

```
% MATLAB CODE FOR PART (III) OF THE EXPERIMENT
% TOMOKI KOIKE
clear all; close all; clc;
fdir = 'C:\Users\Tomo\Desktop\studies\2020-Fall\AAE364lab\matlab\lab2\outputs';
set(groot, 'defaulttextinterpreter','latex');
set(groot, 'defaultAxesTickLabelInterpreter','latex');
set(groot, 'defaultLegendInterpreter','latex');

load("lab2-part2-displacemnet-tomoki.mat");
load("lab2-part2-theta-tomoki.mat");

% Get the data
tspan = part2_displacement.time;
xc = part2_displacement.signals.values;
alpha = part2_theta.signals.values;

% Plot
fig1 = figure(1);
plot(tspan, xc)
title('Gantry Stabilizing Back to Original Position - T. Koike')
xlabel('time, [s]')
ylabel('position $x_c$, [m]')
grid on; grid minor; box on;

fig1 = figure(1);
```

```
plot(tspan, alpha)
title('Gantry Stabilizing of Pendulum - T. Koike')
xlabel('time, [s]')
ylabel('angle $\alpha$, [m]')
grid on; grid minor; box on;
```

Data Analysis

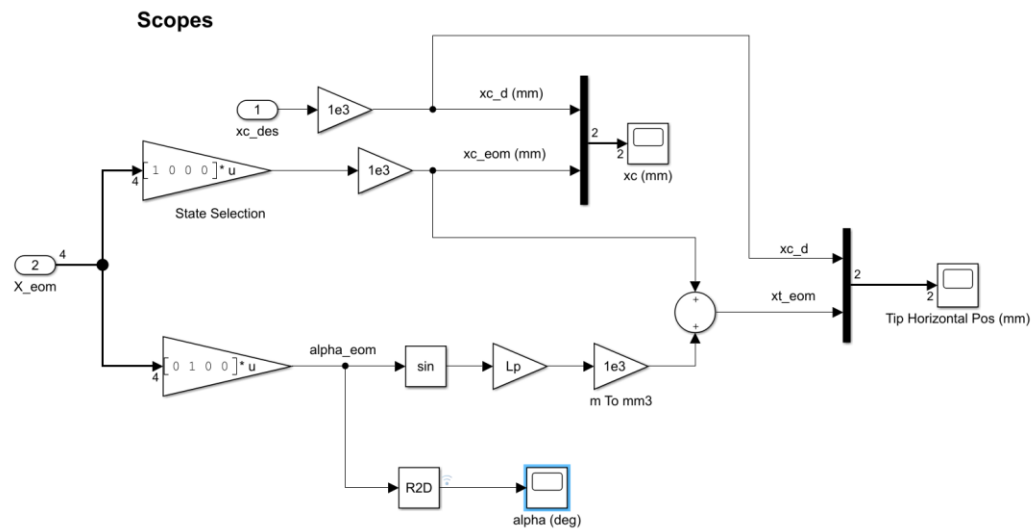
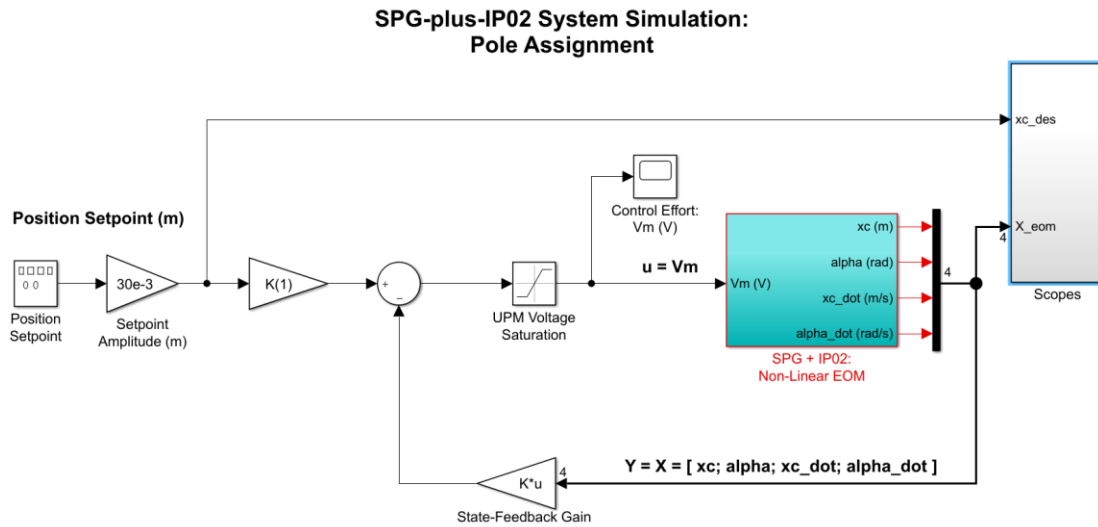
```
% EXPERIMENT DATA ANALYSIS MATLAB CODE
% TOMOKI KOIKE
clear all; close all; clc;

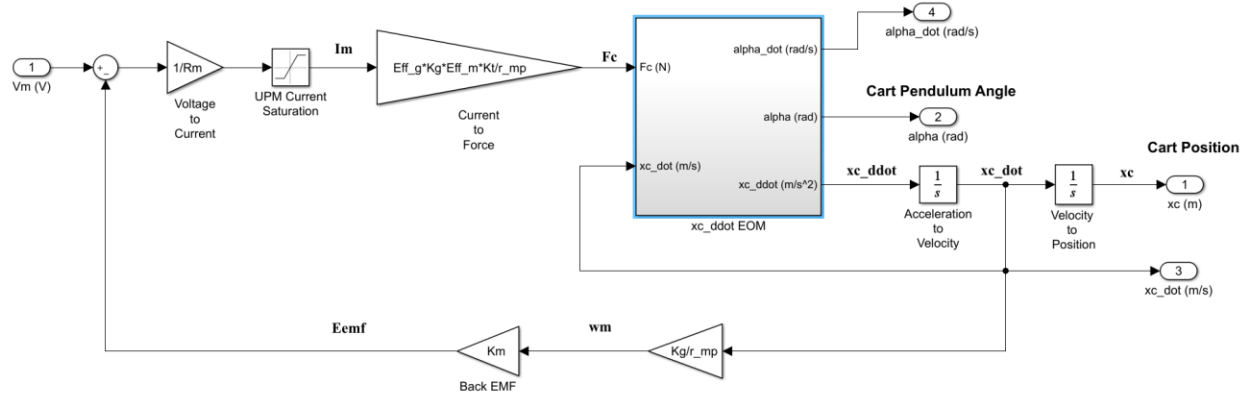
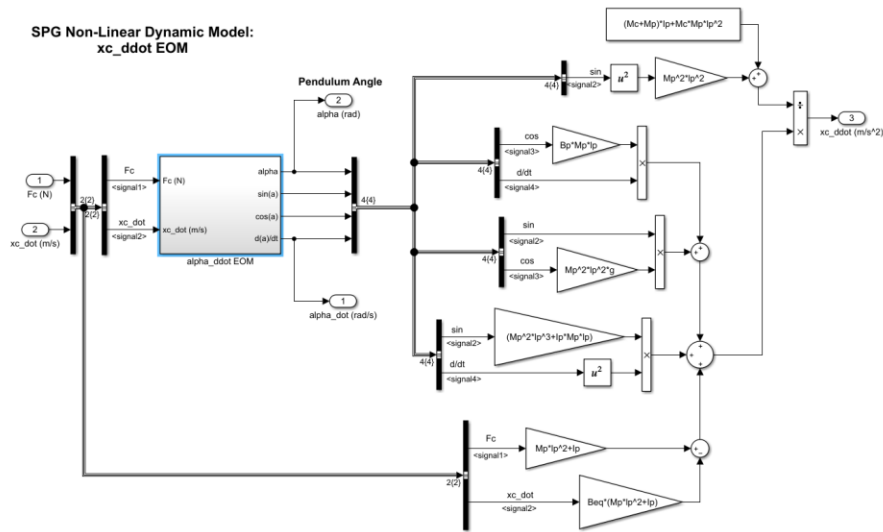
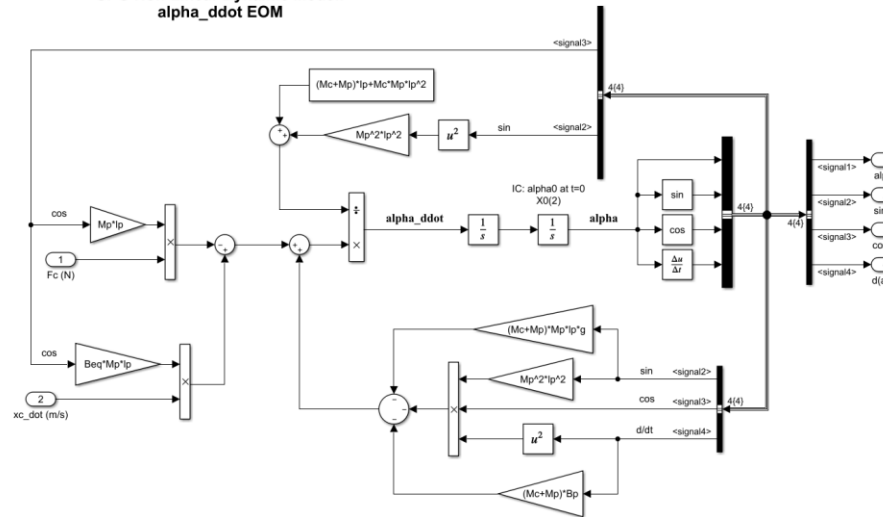
% Get the equilibrium points of the system
[xe, ue, ye, dxe] = trim('linearized_EOM')
[A, B, C, D] = linmod('linearized_EOM', xe, ue)

sys_ss = ss(A,B,C,D)
sys_tf = tf(sys_ss)
```

Simulink Models

Prelab Models



**SPG-plus-IP02 System:
Non-Linear Model**

**SPG Non-Linear Dynamic Model:
xc_ddot EOM**

**SPG Non-Linear Dynamic Model:
 α_ddot EOM**


Original Models