



COLLEGE OF ENGINEERING  
SCHOOL OF AERONAUTICAL AND ASTRONAUTICAL ENGINEERING

AAE 666: NONLINEAR DYNAMICS, SYSTEMS, AND CONTROL

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## HW8

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## Exercise 1

Obtain the describing function of the nonlinear function

$$\phi(y) = y^5$$

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### Solution:

Since  $\phi(y)$  is an odd function, and if we subject this to a sinusoidal input we have

$$\phi(a \sin(\theta)) = a^5 \sin^5(\theta).$$

To find  $\sin^5 \theta$  we use the following procedure

$$\begin{aligned} e^{(i\theta)n} &= e^{ni\theta} \\ (\cos(\theta) + i \sin(\theta))^n &= \cos(n\theta) + i \sin(n\theta) \\ (\cos(\theta) + i \sin(\theta))^5 &= \cos(5\theta) + i \sin(5\theta) \\ \cos^5(\theta) + 5i \cos^4(\theta) \sin(\theta) - 10 \cos^3(\theta) \sin^2(\theta) - 10i \cos^2(\theta) \sin^3(\theta) + \dots \\ &\quad 5 \cos(\theta) \sin^4(\theta) + i \sin^5(\theta) = \cos(5\theta) + i \sin(5\theta) \end{aligned}$$

By equating the imaginary terms we have

$$5 \cos^4(\theta) \sin(\theta) - 10 \cos^2(\theta) \sin^3(\theta) + \sin^5(\theta) = \sin(5\theta)$$

Solve this for  $\sin^5(\theta)$

$$\begin{aligned} \sin^5(\theta) &= \sin(5\theta) - 5 \cos^4(\theta) \sin(\theta) + 10 \cos^2(\theta) \sin^3(\theta) \\ \sin^5(\theta) &= \sin(5\theta) + 5 \cos^2(\theta) \sin(\theta) \left( 2 \sin^2(\theta) - \cos^2(\theta) \right) \\ \sin^5(\theta) &= \sin(5\theta) + \frac{5}{2} \sin(2\theta) \cos(\theta) \left( 1 - \cos(2\theta) - \frac{1}{2} - \frac{1}{2} \cos(2\theta) \right) \\ \sin^5(\theta) &= \sin(5\theta) + \frac{5}{2} \sin(2\theta) \cos(\theta) \left( \frac{1}{2} - \frac{3}{2} \cos(2\theta) \right) \\ \sin^5(\theta) &= \sin(5\theta) + \frac{5}{4} \sin(2\theta) \cos(\theta) - \frac{15}{4} \sin(2\theta) \cos(2\theta) \cos(\theta) \\ \sin^5(\theta) &= \sin(5\theta) + \frac{5}{4} \sin(2\theta) \cos(\theta) - \frac{15}{8} \sin(4\theta) \cos(\theta) \\ \sin^5(\theta) &= \sin(5\theta) + \frac{5}{8} \left( \sin(3\theta) + \sin(\theta) \right) - \frac{15}{16} \left( \sin(5\theta) + \sin(3\theta) \right) \\ \sin^5(\theta) &= \frac{5}{8} \sin(\theta) - \frac{5}{16} \sin(3\theta) + \frac{1}{16} \sin(5\theta). \end{aligned}$$

From this we can deduce,

$$b_1(a) = \frac{5a^5}{8}.$$

Hence,

$$N(a) = \frac{b_1(a)}{a} = \frac{5a^4}{8}.$$

## Exercise 2

Determine whether or not the following Duffing system has a periodic solution. Determine the approximate amplitude and period of all periodic solutions.

$$\ddot{y} - \dot{y} + y^3 = 0$$

---

### Solution:

For this system we have

$$\begin{aligned}\ddot{y} - y &= u \\ u &= -\phi(y).\end{aligned}$$

Taking the Laplace transformation of the first equation we have

$$\begin{aligned}s^2 \hat{y} - \hat{y} &= \hat{u} \\ \hat{G}(s) = \frac{\hat{y}}{\hat{u}} &= \frac{1}{s^2 - 1}.\end{aligned}$$

we know for  $\phi(y) = y^3$

$$N(a) = \frac{3a^2}{4}$$

from **Example 152** on p. 188 of the Notes. Therefore, for some pair  $a, \omega > 0$  using condition (15.7) we have

$$\begin{aligned}1 + \frac{1}{(j\omega)^2 - 1} \left( \frac{3a^2}{4} \right) &= 0 \\ 1 - \frac{3a^2}{4\omega^2 + 4} &= 0 \\ \omega &= \frac{\sqrt{3a^2 - 4}}{2}.\end{aligned}$$

Hence, we predict that this system has periodic solutions of all amplitudes  $a$  and with approximate periods

$$T = \frac{2\pi}{\omega} = \frac{4\pi}{\sqrt{3a^2 - 4}}.$$

### Exercise 3

Determine whether or not the following Duffing system has a periodic solution. Determine the approximate amplitude and period of all periodic solutions.

$$\ddot{y} + \mu\left(\frac{\dot{y}^3}{3} - y\right) + y = 0$$

---

**Solution:**

For this system we have

$$\begin{aligned}\ddot{y} - \mu\dot{y} + y &= u \\ u &= -\phi(y) \\ \phi(y) &= -\frac{\mu}{3}\dot{y}^3\end{aligned}$$

Taking the Laplace transformation of the first equation we have

$$\begin{aligned}s^2\hat{y} - \mu s\hat{y} + \hat{y} &= \hat{u} \\ \hat{G}(s) = \frac{\hat{y}}{\hat{u}} &= \frac{1}{s^2 - \mu s + 1}.\end{aligned}$$

for  $\phi(y) = -\frac{\mu}{3}\dot{y}^3$  if  $y(t) = a \sin(\omega t)$

$$\begin{aligned}\dot{y}^3 &= \left(a\omega \cos(\omega t)\right)^3 = a^3\omega^3 \cos^3(\omega t) \\ &= a^3\omega^3 \left(\frac{3}{4} \cos(\omega t) + \frac{1}{4} \cos(3\omega t)\right) \\ &\approx \frac{3a^3\omega^3}{4} \cos(\omega t) \\ &\approx \frac{3a^3\omega^3}{4} \sin\left(\omega t + \frac{\pi}{2}\right)\end{aligned}$$

Then,

$$\begin{aligned}b_1(a, \omega) &= \frac{3a^3\omega^3}{4} e^{j\frac{\pi}{2}} \\ &= \frac{3a^3\omega^3}{4} \left(\cos\left(\frac{\pi}{2}\right) + j \sin\left(\frac{\pi}{2}\right)\right) \\ &= \frac{3ja^3\omega^3}{4}\end{aligned}$$

Thus, the describing function becomes

$$N(a, \omega) = \frac{3ja^2\omega^3}{4}$$

from **Example 152** on p. 188 of the Notes. Therefore, for some pair  $a, \omega > 0$  using condition (15.7) we have

$$\begin{aligned} 1 + \frac{1}{(j\omega)^2 - \mu j\omega + 1} \left( \frac{3ja^2\omega^3}{4} \right) &= 0 \\ 4\omega^2 + 4j\mu\omega - 4 &= 3ja^2\omega^3 \\ (4\omega^2 - 4) + j(4\mu\omega - 3a^2\omega^3) &= 0. \end{aligned}$$

From this we get the following values

$$\omega = 1 \quad a = \sqrt{\frac{4\mu}{3}}$$

Hence, we predict that this system has periodic solutions of all amplitudes  $a = \sqrt{\frac{4\mu}{3}}$  and with approximate periods  $T = 2\pi$ .

## Exercise 4

Use the describing function method to predict period solutions to

$$\dot{x}(t) = -x(t) - 2sgm(x(t-h))$$

Illustrate your results with numerical simulations.

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### Solution:

The system can be defined as

$$\begin{aligned}\dot{x}(t) &= -x(t) + u \\ u &= -\phi(y) = -2sgm(y) \\ y &= x(t-h).\end{aligned}$$

From the Notes we know that  $N(a) = \frac{4}{\pi a}$  for  $sgm(x)$ , and therefore,

$$N(a) = \frac{8}{\pi a}.$$

Now we do the Laplace transform of the system

$$\begin{aligned}s\hat{x} &= -\hat{x} + \hat{u} \\ \hat{y} &= e^{-sh}\hat{x} \\ \therefore \hat{G}(s) &= \frac{\hat{y}}{\hat{u}} = \frac{e^{-sh}}{s+1}.\end{aligned}$$

Then

$$\begin{aligned}\hat{G}(j\omega) &= \frac{e^{-j\omega h}}{j\omega + 1} = \frac{e^{-j\omega h}(1 - j\omega)}{1 + \omega^2} \\ &= \frac{(1 - j\omega)(\cos(\omega h) - j\sin(\omega h))}{1 + \omega^2} \\ &= \frac{(\cos(\omega h) - \omega \sin(\omega h)) - j(\omega \cos(\omega h) + \sin(\omega h))}{1 + \omega^2}.\end{aligned}$$

From

$$1 + \hat{G}(j\omega)N(a) = 0$$

$$\begin{aligned}1 + \frac{(\cos(\omega h) - \omega \sin(\omega h)) - j(\omega \cos(\omega h) + \sin(\omega h))}{1 + \omega^2} \left( \frac{8}{\pi a} \right) &= 0 \\ (8 \cos(\omega h) - 8\omega \sin(\omega h) + \pi a \omega^2 + \pi a) - 8j(\omega \cos(\omega h) + \sin(\omega h)) &= 0.\end{aligned}$$

Thus, we can numerically compute an  $\omega$  and  $a$  value that satisfies the equation above. That will give us a **period solution for the system**. Let  $h = 1$ , then we can numerically compute the  $\omega$  value from the imaginary part

$$\begin{aligned}\omega \cos(\omega h) + \sin(\omega h) &= 0 \\ \tan(\omega h) &= -\omega \\ \tan(\omega) &= -\omega.\end{aligned}$$

We use MATLAB to numerically solve the smallest  $\omega > 0$  value to be,

$$\omega = 2.0287577.$$

Next we plug the  $h$  and  $\omega$  value into the real part

$$8 \cos(\omega h) - 8\omega \sin(\omega h) + \pi a \omega^2 + \pi a = 0$$

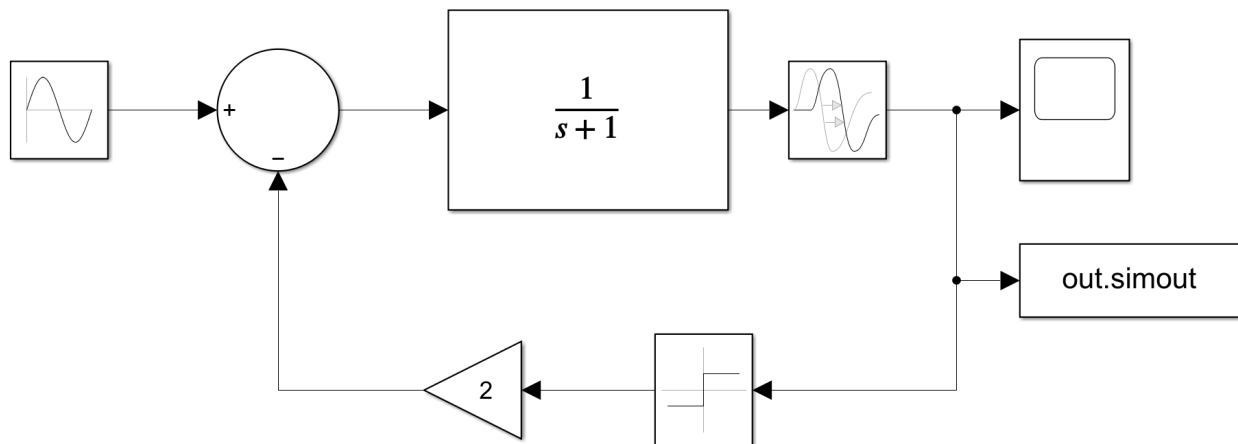
which gives us

$$a = 1.1259$$

This means that we have a sinusoidal input with amplitude of  $a = 0.5629$  and period of

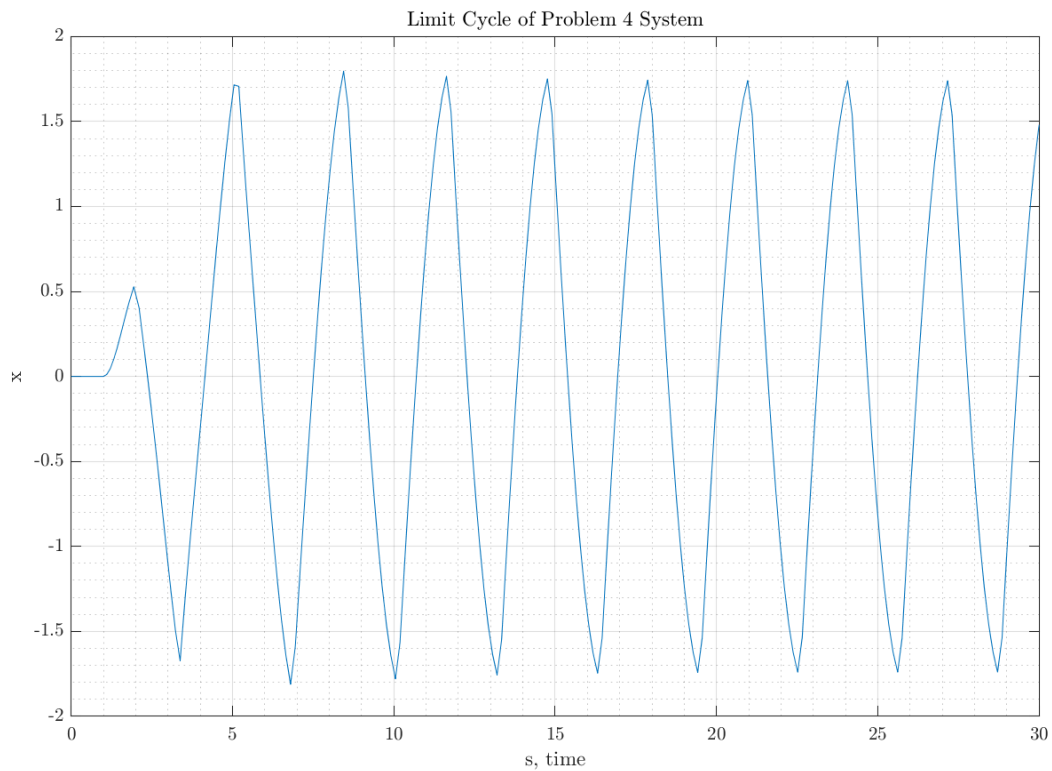
$$T = \frac{2\pi}{\omega} = 3.0971$$

that gives a period solution to the system. Thus, we feed an input of  $a \sin(\omega t)$  to the following system



and we obtain the following simulation.





We can see that the results give us a limit cycle for the system.

MATLAB CODE:

```

1 % AAE 666 HW9 PROBLEM 4 MATLAB CODE
2 % TOMOKI KOIKE
3
4 % Houskeeping commands
5 clear all; close all; clc;
6 %%
7 % Compute omega and a values for the describing function numerically
8 h = 1 % time delay constant
9 tol = 1e-6; % tolerance
10 for omega = 0.1:0.0000001:10
11     delta = abs(tan(omega) + omega);
12     if delta < tol
13         break
14     end
15 end
16 format long;

```

```

17 disp(omega);
18 format;
19
20 % Compute a
21 syms a
22 assume(a, {'positive', 'real'});
23 eqn = 8*cos(omega*h) - 8*omega*sin(omega*h) + pi*a*omega^2 + pi*a == 0;
24 a = solve(eqn, a);
25 format long ;
26 disp(a);
27 format
28 a = double(a);
29 %%
30 % Period
31 T = 2*pi / omega
32 %%
33 % Simulate
34 set(groot, 'defaulttextinterpreter','latex');
35 set(groot, 'defaultAxesTickLabelInterpreter','latex');
36 set(groot, 'defaultLegendInterpreter','latex');
37 out = sim('p4_signum.slx');
38
39 t = out.tout;
40 x = out.simout.signals.values;
41
42 fig = figure("Renderer","painters", "Position",[60 60 900 600]);
43 plot(t, x);
44 grid on; grid minor; box on;
45 xlabel('s, time')
46 ylabel('x')
47 title('Limit Cycle of Problem 4 System')
48 saveas(fig, 'p4_result.png');

```

## Exercise 5

Consider the double integrator

$$\ddot{q} = u$$

subject to a saturating PID controller

$$u = -k_P q - k_D \dot{q} - \text{sat}(\tilde{u}) \quad \text{where} \quad \tilde{u} = k_I \int q dq$$

(a) For  $k_P = 1$  and  $k_D = 2$  determine the largest of  $k_I \leq 0$  for which the closed loop system is asymptotically stable about  $q(t) \equiv 0$ .

(b) For  $k_P = 1$  and  $k_D = 2$ , use the describing function method to determine the smallest value  $k_I \geq 0$  for which the closed loop system has a periodic solution.

### Solution:

(a) For the system we define  $x_1 = q$ ,  $x_2 = \dot{q}$ , and  $x_3 = \int q$ . Then the system can be represented by the following state space

$$\begin{aligned} \dot{x}_1 &= x_2 \\ \dot{x}_2 &= -x_1 - 2x_2 - \text{sat}(k_I x_3) \\ \dot{x}_3 &= x_1 \end{aligned}$$

Now if we assume that  $-1 \leq \tilde{u} \leq 1$ , this can be rewritten to be

$$\begin{aligned} \dot{x}_1 &= x_2 \\ \dot{x}_2 &= -x_1 - 2x_2 - k_I x_3 \\ \dot{x}_3 &= x_1. \end{aligned}$$

The state matrix  $A$  for this linear system becomes

$$A = \begin{bmatrix} 0 & 1 & 0 \\ -1 & -2 & -k_I \\ 1 & 0 & 0 \end{bmatrix}.$$

The eigenvalues for this is

$$\text{eig}(A) = \left[ \begin{array}{c} \frac{0.1111}{\sigma_1} + \sigma_1 - 0.6667 \\ -0.6667 + \frac{-0.0556 - 0.0962i}{\sigma_1} + \sigma_1 (-0.5000 + 0.8660i) \\ -0.6667 + \frac{-0.0556 + 0.0962i}{\sigma_1} + \sigma_1 (-0.5000 - 0.8660i) \end{array} \right]$$

where

$$\sigma_1 = \left( \sqrt{(0.5000 k_I - 0.0370)^2 - 0.0014 - 0.5000 k_I + 0.0370} \right)^{0.3333}$$

The real part for this is

$$\begin{bmatrix} \text{Re}(\text{eig}(A)) = \frac{0.1111 \sigma_4}{|\sigma_5|^{0.3333}} + |\sigma_5|^{0.3333} \sigma_4 - 0.6667 \\ \text{real}(-\sigma_3) + \text{real}(0.8660 \sigma_5^{0.3333} i) - \sigma_1 - \sigma_2 - 0.6667 \\ \text{real}(\sigma_3) + \text{real}(-0.8660 \sigma_5^{0.3333} i) - \sigma_1 - \sigma_2 - 0.6667 \end{bmatrix}$$

where

$$\begin{aligned} \sigma_1 &= \frac{0.0556 \sigma_4}{|\sigma_5|^{0.3333}} \\ \sigma_2 &= 0.5000 |\sigma_5|^{0.3333} \sigma_4 \\ \sigma_3 &= \frac{0.0962 i}{\sigma_5^{0.3333}} \\ \sigma_4 &= \cos(0.3333 \text{ angle}(\sigma_5)) \\ \sigma_5 &= \sqrt{(0.5000 k_I - 0.0370)^2 - 0.0014 - 0.5000 k_I + 0.0370} \end{aligned}$$

We numerically solve the largest value using MATLAB (code provided) by plugging in a value and check if all the eigenvalues have negative real parts, and we figure out that the bounds for  $k_I$  is

$$0 \leq k_I < 2.$$

MATLAB CODE:

```

1 % Houskeeping commands
2 clear all; close all; clc;
3 %%
4 % The system matrix A
5 syms k_I
6 assume(k_I, {'real', 'positive'});
7 A = [0, 1, 0; -1, -2, -k_I; 1, 0, 0];
8 %%
9 ev = eig(A)
10 ev_real = real(ev)
11 %%
12 inc = 0.0001;
13 for ki = 1.999:inc:2.1
14     ev_real_vals = double(subs(ev_real, k_I, ki));
15     if any(ev_real_vals > 0)

```

```

16         break
17     end
18 end
19 format long;
20 disp(ki-inc);
21 format;

```

(b) For this part we use the fact that we set  $U = -\phi(y)$  and  $\phi(y) = -\text{sat}(k_I x_3)$  which means that  $y = k_I x_3$ . Then we can the Laplace transform and get

$$s^2 \hat{q} = -\hat{q} - 2s\hat{q} + \hat{U}$$

$$\hat{y} = \frac{k_I}{s} \hat{q}$$

which gives us

$$\hat{G}(s) = \frac{\hat{y}}{\hat{u}} = \frac{k_I}{s(s^2 + 2s + 1)}.$$

Now,

$$\begin{aligned} \hat{G}(j\omega) &= \frac{k_I}{j\omega(-\omega^2 + 2j\omega + 1)} \\ &= \frac{k_I(-2\omega^2 - j\omega(1 - \omega^2))}{4\omega^4 + (\omega - \omega^3)^2}. \end{aligned}$$

From the **Notes** we know that the describing function for a saturation function becomes

$$N(a) = \begin{cases} 1 & \text{if } 0 \leq a \leq 1 \\ \frac{2}{\pi} \left[ \arcsin\left(\frac{1}{a}\right) + \frac{\sqrt{a^2 - 1}}{a^2} \right] & \text{if } 1 < a \end{cases}$$

If  $0 \leq a \leq 1$  then  $\omega = 0$  and is not adequate for our purpose, and therefore

$$1 + \hat{G}(j\omega)N(a) = 0$$

$$1 + \frac{k_I(-2\omega^2 - j\omega(1 - \omega^2))}{4\omega^4 + (\omega - \omega^3)^2} N(a) = 0.$$

We separate this into the real and imaginary parts.

$$\begin{aligned} \text{Real} &:= 4\omega^4 + (\omega - \omega^3)^2 - 2k_I\omega^2 N(a) \\ \text{Imag} &:= -jk_I\omega(1 - \omega^2)N(a) \end{aligned}$$

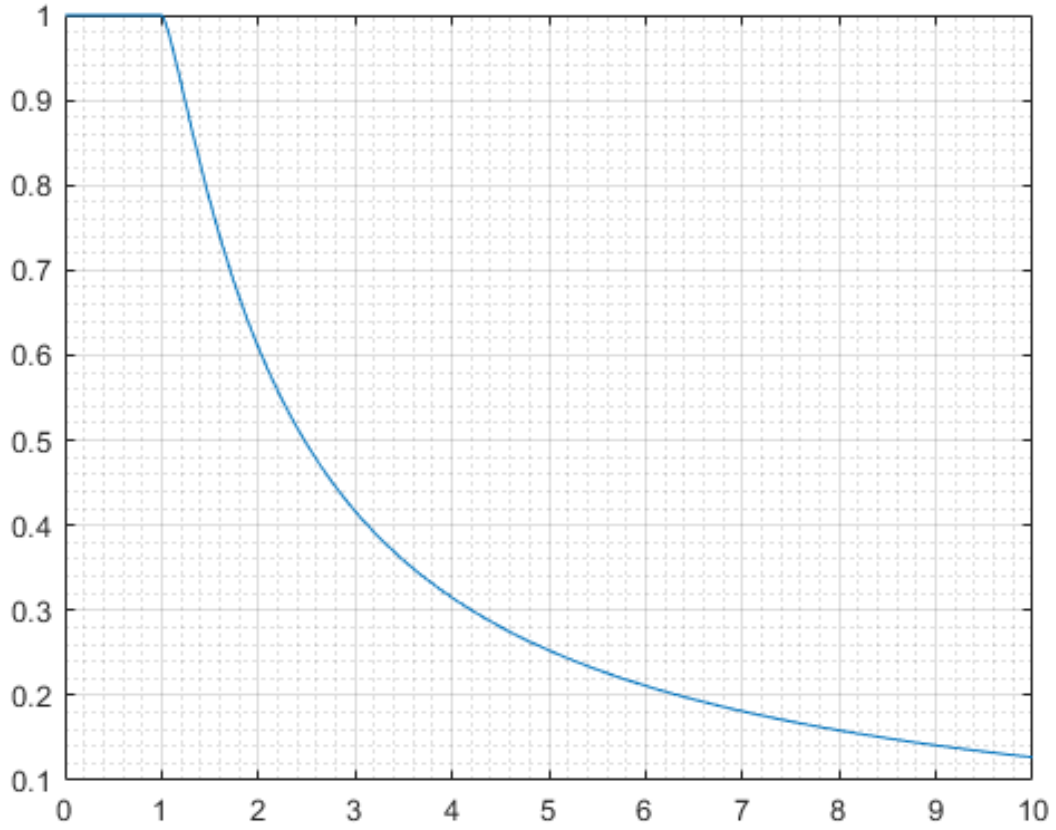
We solve this so that the imaginary and real parts are equal to 0. From the imaginary part we have

$$\omega = 1$$

since  $N(a)$  at  $0 \leq a \leq 1$  is a positive function. Then we solve the real part as

$$4 - 2k_I N(a) = 0.$$

Since,  $N(a)$  behaves as the following



the smallest  $k_I \geq 0$  value will be when  $N(a) = 1$  where  $0 \leq a \leq 1$ , and therefore,

$$k_I = 2.$$