

Channel Shaping Using Reconfigurable Intelligent Surfaces: From Diagonal to Beyond

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I. ASSUMPTION

All proposals in this paper based on assumption of *asymmetric* passive Beyond-Diagonal (BD) Reconfigurable Intelligent Surface (RIS), i.e., symmetry constraint $\Theta_g = \Theta_g^T$ is relaxed. This is feasible when asymmetric passive components (e.g., ring hybrids and branch-line hybrids) [1] are available. This assumption was also made in Hongyu's papers [2], [3]. For quadratic problems, the proposed algorithms may be extended to symmetric BD RIS by replacing singular value decomposition with Takagi factorization [4].

II. POINT-TO-POINT MIMO

A. Channel Power Maximization

Consider a BD RIS with N_s elements, which is divided into G groups of equal L elements.

$$\max_{\Theta} \left\| \mathbf{H}^D + \sum_g \mathbf{H}_g^B \Theta_g \mathbf{H}_g^F \right\|_F^2 \quad (1a)$$

$$\text{s.t.} \quad \Theta_g^H \Theta_g = \mathbf{I}_L, \quad \forall g \in \mathcal{G} \triangleq \{1, \dots, G\} \quad (1b)$$

For *symmetric* BD-RIS, the problem has been solved in

- Matteo's paper [5]: SISO and equivalent¹;
- Ignacio's paper [6]: SISO and directless MISO/SIMO.

Remark 1. *The difficulty of (1) is that the RIS needs to balance the additive (direct-indirect) and multiplicative (backward-forward) eigenspace alignment. Interestingly, it has the same form as the weighted orthogonal Procrustes problem [7]:*

$$\min_{\Theta} \left\| \mathbf{C} - \mathbf{A} \Theta \mathbf{B} \right\|_F^2 \quad (2a)$$

$$\text{s.t.} \quad \Theta^H \Theta = \mathbf{I} \quad (2b)$$

There exists no trivial solution to (2). One lossy transformation, by moving Θ to one side [8], formulates a standard orthogonal Procrustes problem:

$$\min_{\Theta} \left\| \mathbf{A}^\dagger \mathbf{C} - \Theta \mathbf{B} \right\|_F^2 \quad (3a)$$

$$\text{s.t.} \quad \Theta^H \Theta = \mathbf{I} \quad (3b)$$

(3) has a global optimal solution $\Theta^* = \mathbf{U} \mathbf{V}^H$, where \mathbf{U} and \mathbf{V} are left and right singular matrix of $\mathbf{A}^\dagger \mathbf{C} \mathbf{B}^H$ [9]. This low-complexity solution will be compared with the one proposed later.

Inspired by [10], we propose an iterative algorithm to solve (1). The idea is to successively approximate the quadratic

objective with a sequence of affine functions and solve the resulting subproblems in closed form.

Proposition 1. *Start from any $\Theta^{(0)}$, the sequence*

$$\Theta_g^{(r+1)} = \mathbf{U}_g^{(r)} \mathbf{V}_g^{(r)}, \quad \forall g \in \mathcal{G} \quad (4)$$

converges to a stationary point of (1), where $\mathbf{U}_g^{(r)}$ and $\mathbf{V}_g^{(r)}$ are left and right singular matrix of

$$\begin{aligned} \mathbf{M}_g^{(r)} = & \mathbf{H}_g^{B^H} \mathbf{H}^D \mathbf{H}_g^{F^H} + \sum_{g' < g} \mathbf{H}_{g'}^{B^H} \mathbf{H}_{g'}^B \Theta_{g'}^{(r+1)} \mathbf{H}_{g'}^F \mathbf{H}_{g'}^{F^H} \\ & + \sum_{g' \geq g} \mathbf{H}_{g'}^{B^H} \mathbf{H}_{g'}^B \Theta_{g'}^{(r)} \mathbf{H}_{g'}^F \mathbf{H}_{g'}^{F^H}. \end{aligned} \quad (5)$$

Proof. TODO □

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¹Single-stream MIMO with given precoder and combiner.