

Project Overview

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June 2024

1 Condition of Oscillation

The Condition of oscillation is given by the Backhausen criteria which states that the loop gain of the positive feedback network must be 1 for sustained oscillations.

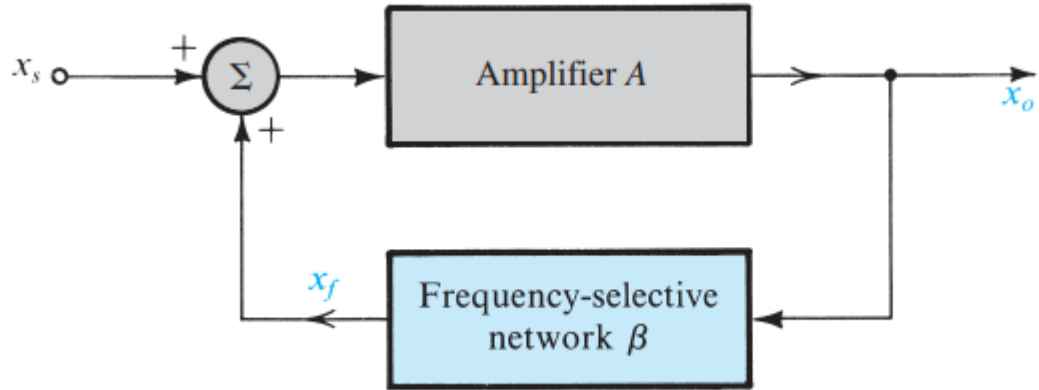


Figure 1: Block diagram of a Positive feedback network

$$Transfer\ function = x_o/x_s = \frac{A\beta}{1 - A\beta} \quad (1)$$

Let

$$D(s) = 1 - A\beta$$

So for the system to be unstable

$$A\beta \geq 1$$

$$Therefore, D(s) \leq 0$$

2 Design Overview

2.1 Amplifier

I have used the common emitter configured BJT (Bipolar Junction Transistor) as the amplifier. I am going to use the BJT 2N2222 NPN transistor manufactured by NXP Semiconductors for this Simulations. As due to heating effect if used in

$$A\beta = 1$$

condition the amplitude will decay with time. So it has to be operated during the condition

$$A\beta > 1$$

As we set in unstable mode the oscillations will rise so need to limit the amplitude,

Fortunately BJT has inherent amplitude stabilization as h_{fe} decreases for larger i_c .

h_{FE}	DC current gain	$I_C = 0.1 \text{ mA}; V_{CE} = 10 \text{ V}$	35	–	
		$I_C = 1 \text{ mA}; V_{CE} = 10 \text{ V}$	50	–	
		$I_C = 10 \text{ mA}; V_{CE} = 10 \text{ V}$	75	–	
		$I_C = 150 \text{ mA}; V_{CE} = 1 \text{ V}; \text{note 1}$	50	–	
		$I_C = 150 \text{ mA}; V_{CE} = 10 \text{ V}; \text{note 1}$	100	300	
h_{FE}	DC current gain 2N2222A	$I_C = 10 \text{ mA}; V_{CE} = 10 \text{ V}; T_{amb} = -55^\circ \text{C}$	35	–	
h_{FE}	DC current gain	$I_C = 500 \text{ mA}; V_{CE} = 10 \text{ V}; \text{note 1}$			
	2N2222		30	–	
	2N2222A		40	–	

Figure 2: NPN 2N2222 Datasheet Important Data

From the datasheet , the h_{FE} of the transistor 2N2222 increases upto $i_c = 150 \text{ mA}$, after that it decreases upto 30 at $i_c = 500\text{mA}$ for a constant V_{ce} . Here increase of i_b results in increase of i_c which results in decrease of h_{FE} value that implies decrease of gain.

The value of h_{FE} can be measured at DC operating point by dividing the value of i_c and i_b at that operating point.

Whereas decrease in i_c results in inc of h_{fe} which increases the gain.

This stabilizes the amplitude to a particular limit.

The amplitude limiter characteristics can be realized by operating the amplifier at its desired DC operating Point and passing a ramp signal as input. The output curve can be measured from the output terminal.

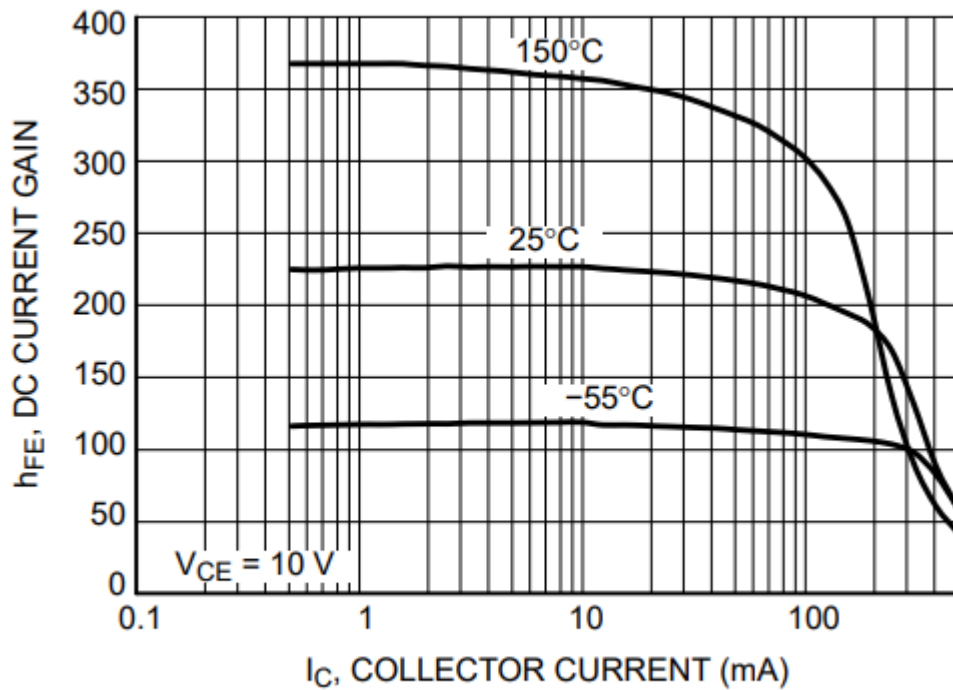


Figure 1. DC Current Gain

Figure 3: NPN 2N2222A Datasheet plot of h_{FE} vs i_c

2.2 Selection of β network circuit

As the Common Emitter Amplifier is an inverting amplifier which means that along with amplification it also shifts the phase by 180 degrees, I had to choose a β network such that it also creates a phase difference of 180 degrees so that the Loop gain is in phase of one another.

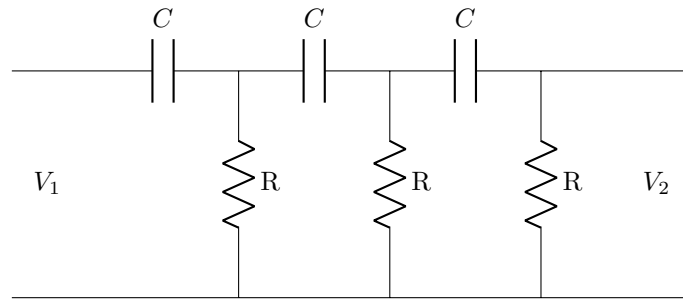


Figure 4: RC phase shift network Circuit

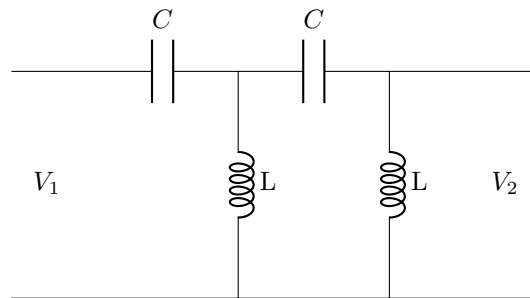


Figure 5: LC network Circuit

- For low frequency applications typically ranging from 10 Hz to 1 KHz the cascaded RC network is used as it provides the phase difference of 180 degrees at a particular frequency.
- For high frequency applications , all the way upto some Megahertz the LC network is used. It provides phase difference of 180 degrees at some particular frequency.