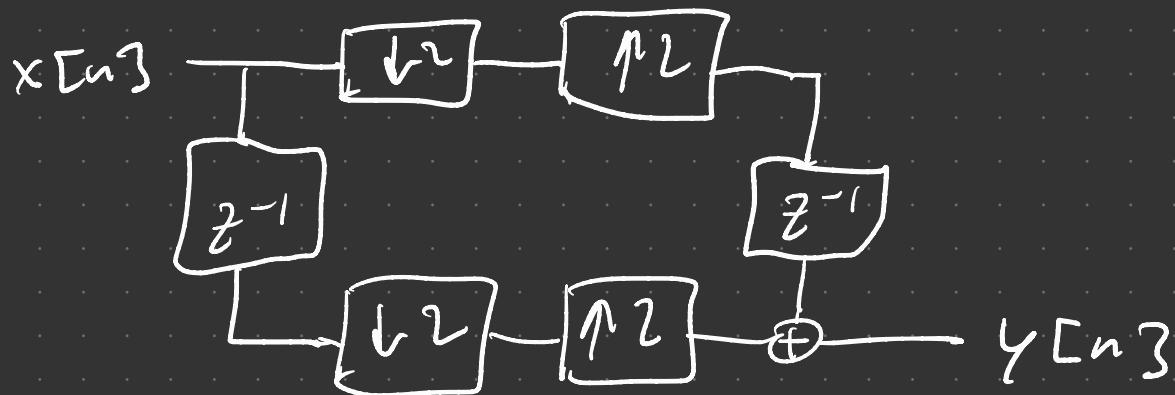


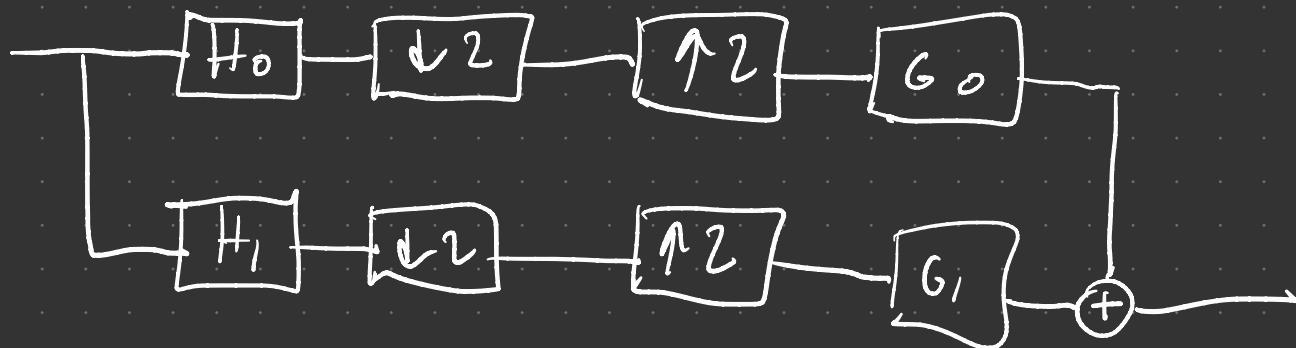
# Polyphase Representation of Filter Banks

Exercise:



Show that  $y[n] = x[n-1]$ .

Consider:



Goal: Write down all of these filters in their respective polyphase representations.

# Analysis Bank:

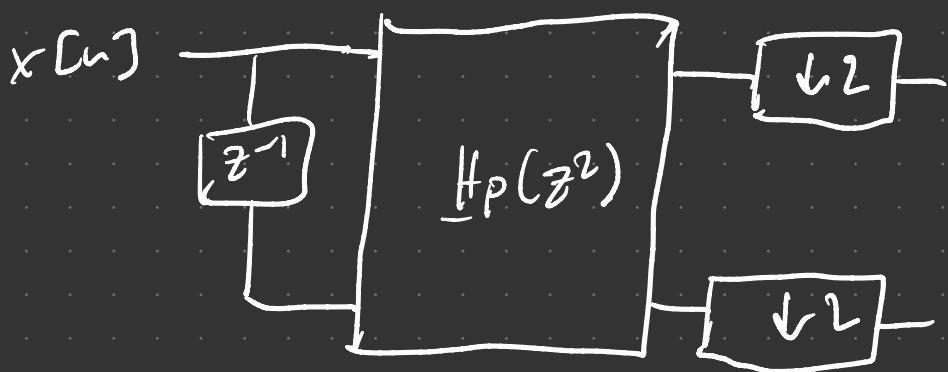
$$\begin{bmatrix} H_0(z) \\ H_1(z) \end{bmatrix} = \begin{bmatrix} H_{0,\text{even}}(z^2) + z^{-1} H_{0,\text{odd}}(z^2) \\ H_{1,\text{even}}(z^2) + z^{-1} H_{1,\text{odd}}(z^2) \end{bmatrix}$$

$$= \begin{bmatrix} H_{0,\text{even}}(z^2) & H_{0,\text{odd}}(z^2) \\ H_{1,\text{even}}(z^2) & H_{1,\text{odd}}(z^2) \end{bmatrix} \begin{bmatrix} 1 \\ z^{-1} \end{bmatrix}$$

Poly phase matrix

$$\underline{H}_P(z^2)$$

The analysis bank is equivalent to:

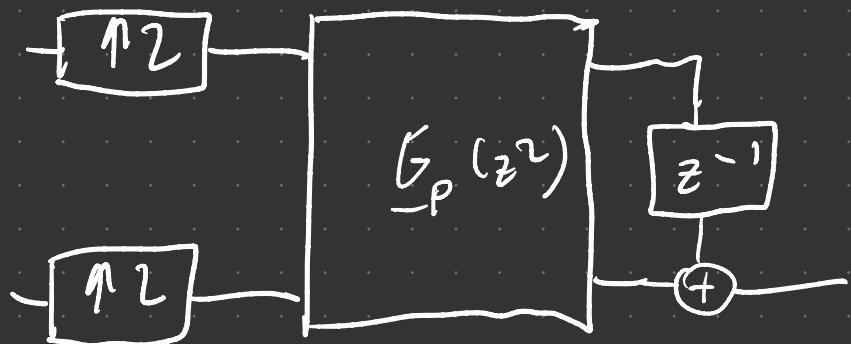


## Synthesis Bank:

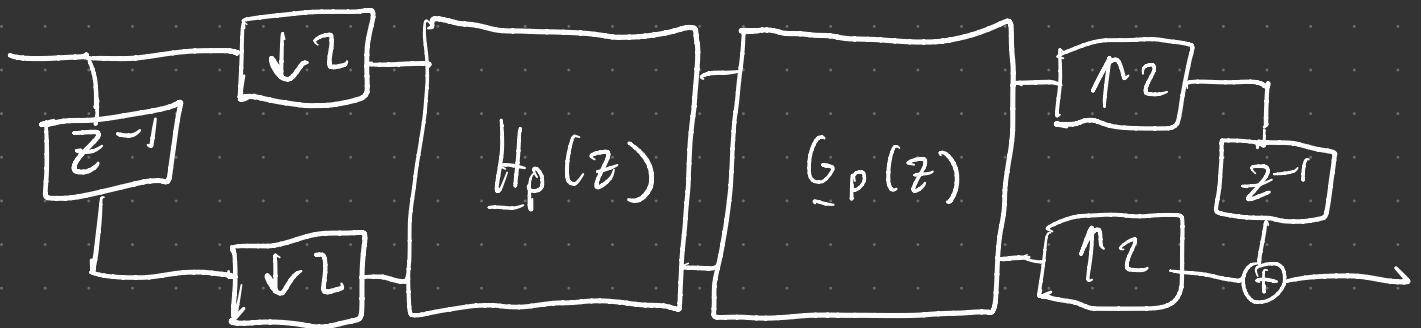
$$\begin{bmatrix} G_0(z) & G_1(z) \end{bmatrix} = \begin{bmatrix} z^{-1} & 1 \end{bmatrix} \begin{bmatrix} G_{0,\text{odd}}(z^2) & G_{1,\text{odd}}(z^2) \\ G_{0,\text{even}}(z^2) & G_{1,\text{even}}(z^2) \end{bmatrix}$$

polyphase matrix

The synthesis bank is equivalent to:



By applying the multirate identities, we find the equivalent system



polyphase representation of  
the two-channel filter bank.

Obs: Perfect - reconstruction is guaranteed by:

$$G_p(z) H_p(z) = z^{-k} I \quad \text{identity matrix}$$

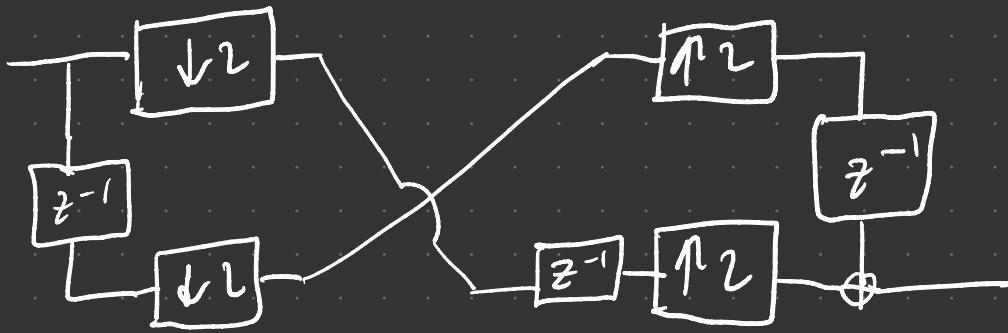
Exercise: Verify that this is true and determine L.

Q: Are there other conditions that guarantee PR?

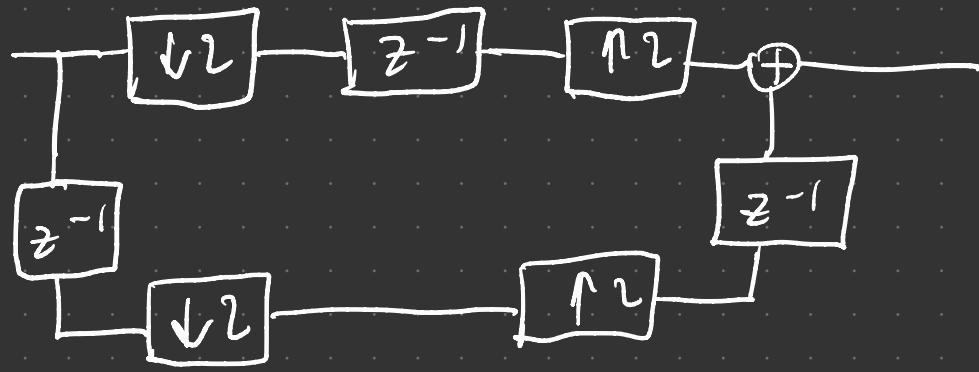
A: If  $G_p(z) H_p(z) = z^{-K} \begin{bmatrix} 0 & 1 \\ z^{-1} & 0 \end{bmatrix}$ , then PR is guaranteed.

First we will ignore  $z^{-K}$ .

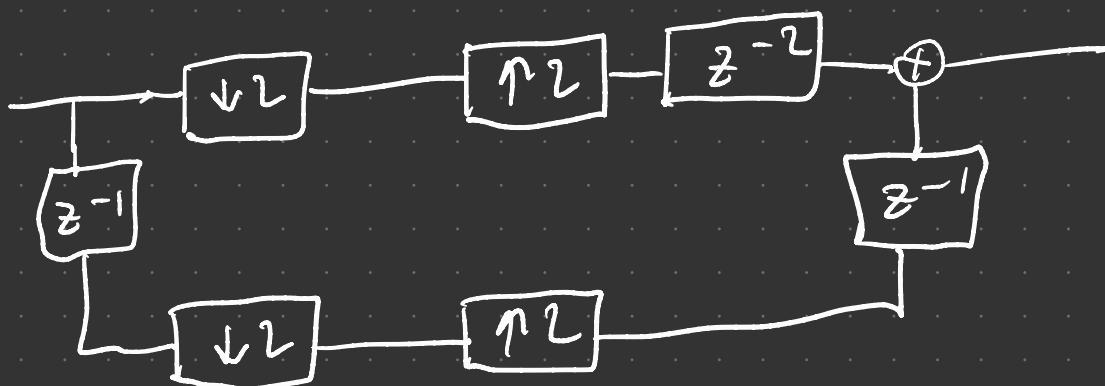
We have the system:



=

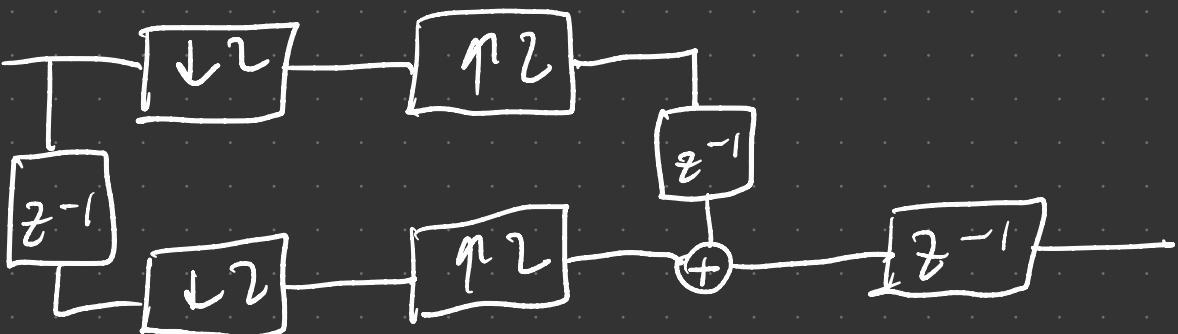


=



multirate  
identity

=





Therefore, if  $G_p(z) \underline{H}_p(z) = z^{-K} \begin{bmatrix} 0 & 1 \\ z^{-1} & 0 \end{bmatrix}$ ,  
then the system is PR  
Exercise: Determine  $L$ .

Exercise: Suppose that  $H_1(z) = H_0(-z)$ , where  $H_0$  is FIR. Find all PR systems.

Soln: Use the polyphase representation.

$$H_0(z) = E_0(z^2) + z^{-1} E_1(z^2)$$

even      odd  
 polyphase      polyphase

$$H_1(z) = H_0(-z) = E_0(z^2) - z^{-1} E_1(z^2)$$

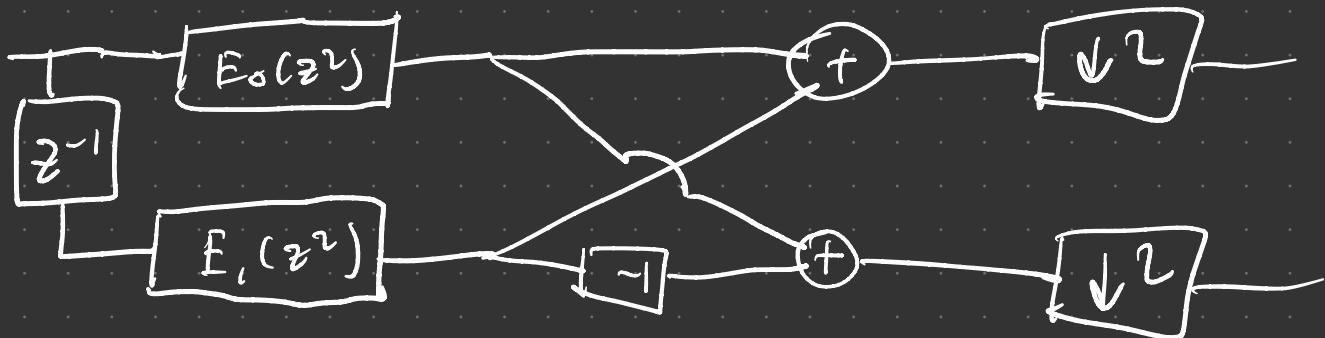
$$\begin{bmatrix} H_0(z) \\ H_1(z) \end{bmatrix} = \begin{bmatrix} E_0(z^2) & E_1(z^2) \\ E_0(z^2) & -E_1(z^2) \end{bmatrix} \begin{bmatrix} 1 \\ z^{-1} \end{bmatrix}$$

$\underbrace{\quad}_{\underline{H}_p(z^2)}$

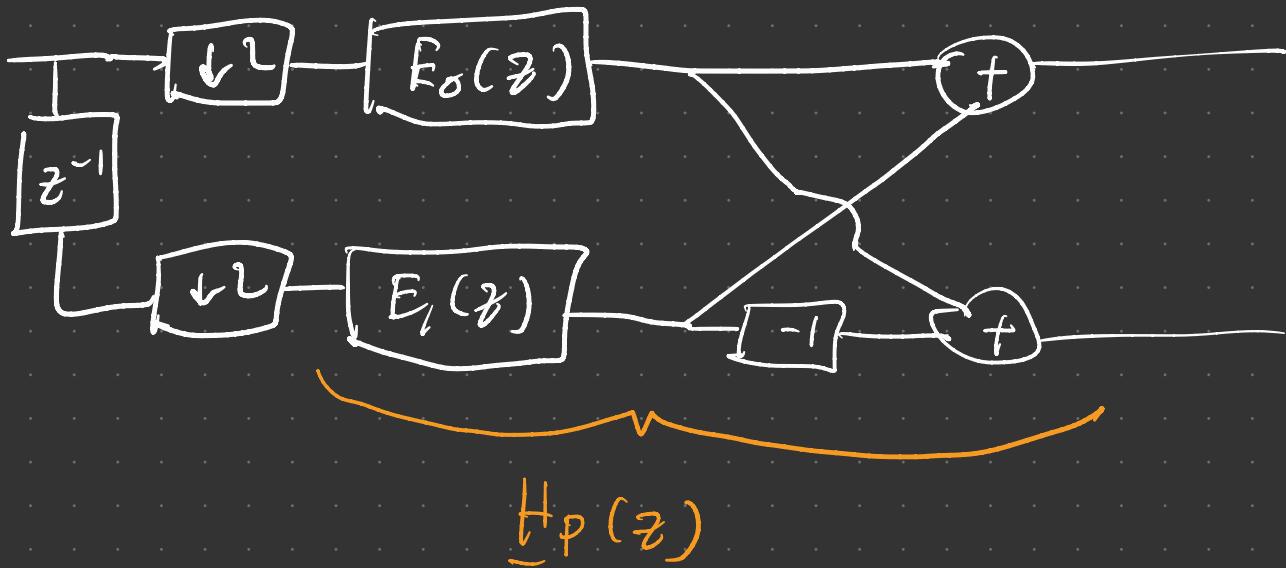
$$= \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} \begin{bmatrix} E_0(z^2) & 0 \\ 0 & E_1(z^2) \end{bmatrix} \begin{bmatrix} 1 \\ z^{-1} \end{bmatrix}$$

2-point  
DFT

## Analysis Banks



By the multirate identities

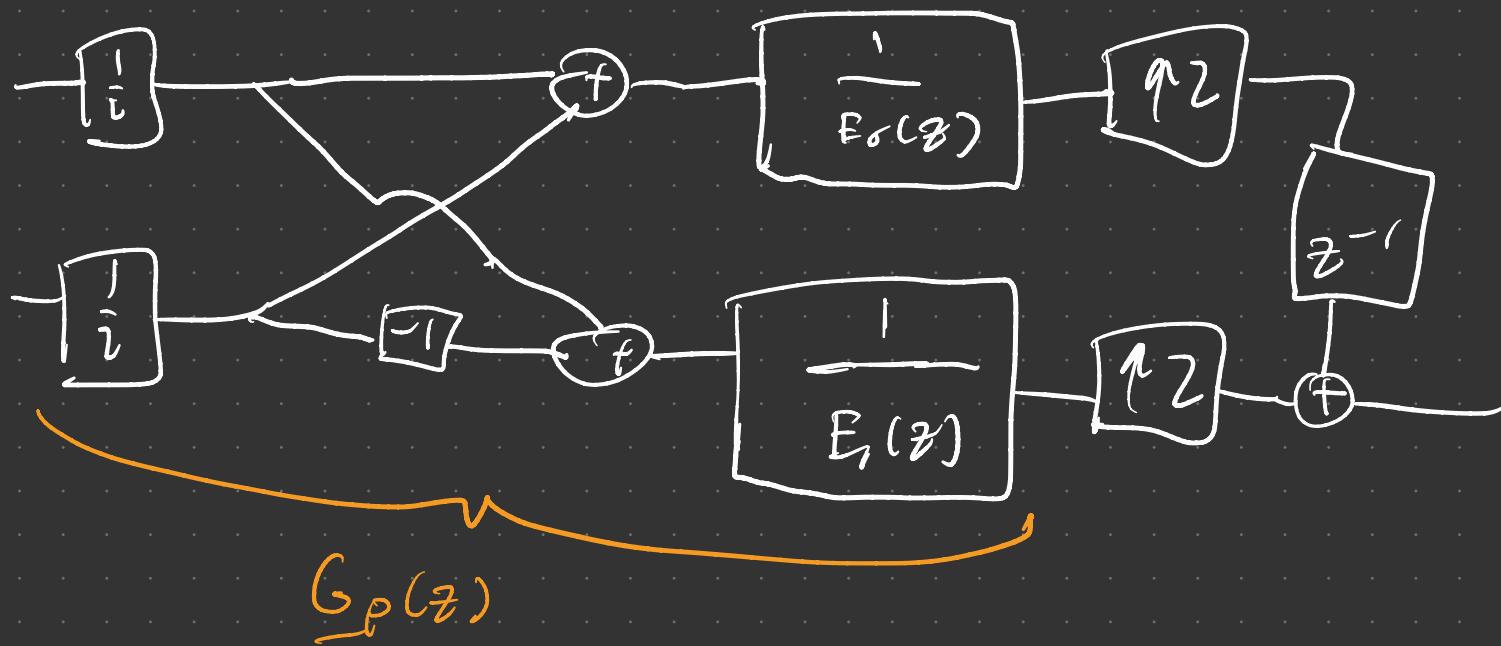


The problem has been reduced to inventing  $H_p(z)$ ,  
i.e., finding  $F_p(z)$ .

Synthesis Bank:

Observe that

$$\begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} \cdot \frac{1}{2} \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$



$$G_p(z) = \frac{1}{2} \begin{bmatrix} \frac{1}{E_o(z)} & 0 \\ 0 & \frac{1}{E_i(z)} \end{bmatrix} \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

$$\begin{bmatrix} G_o(z) & G_I(z) \end{bmatrix} = \left[ z^{-1} \quad 1 \right] \cdot \frac{1}{2} \begin{bmatrix} \overbrace{\frac{1}{E_0(z^2)}}^0 & 0 \\ 0 & \overbrace{\frac{1}{E_I(z^2)}}^1 \end{bmatrix} \begin{bmatrix} 1 & \\ & -1 \end{bmatrix}$$

$$= \left[ z^{-1} \quad 1 \right] \cdot \frac{1}{2} \begin{bmatrix} \overbrace{\frac{1}{E_0(z^2)}}^0 & \overbrace{\frac{1}{E_0(z^2)}}^1 \\ \overbrace{\frac{1}{E_I(z^2)}}^1 & -\overbrace{\frac{1}{E_I(z^2)}}^0 \end{bmatrix}$$

$$= \frac{1}{2} \left[ \overbrace{\frac{1}{E_I(z^2)}}^1 + z^{-1} \overbrace{\frac{1}{E_0(z^2)}}^1 - \overbrace{\frac{1}{E_I(z^2)}}^1 + z^{-1} \overbrace{\frac{1}{E_0(z^2)}}^1 \right]$$

Remark: So far, we have not assumed anything about the filters,

Q: What if we want FIR filters?

A: Polyphase components must be delays.

$$\text{i.e.) } E_0(z) = a z^{-l_0}, \quad E_1(z) = b z^{-l_1}$$

The general FIR solution takes the form:

$$\begin{aligned} H_0(z) &= E_0(z^2) + z^{-1} E_1(z^2) \\ &= az^{-2l_0} + bz^{-(2l_1+1)} \end{aligned}$$

Obs: For Haar wavelets,  $l_0 = 0, l_1 = 0,$   
 $a = \frac{1}{\sqrt{2}}, \quad b = \frac{1}{\sqrt{2}}$

- For IIR solutions, there are infinite possibilities.
- For causal & stable IIR solution,  
 $E_0$  &  $E_1$  must be min-phase.

(Zeros of  $E_0$  &  $E_1$  are inside unit circle)

$\Rightarrow$  poles of  $\frac{1}{E_0}$  &  $\frac{1}{E_1}$  are inside  
unit circle )