

Welcome to ECE 251C

Filter Banks and Wavelets
(Multirate Signal Processing)

Course will primarily focus on:

- ① Filters
- ② Filter Banks
- ③ Wavelets

Lectures : Theory

HW : Theory and MATLAB (or python)

Project: Theory OR Applications

Filters / LTI Systems (Review)

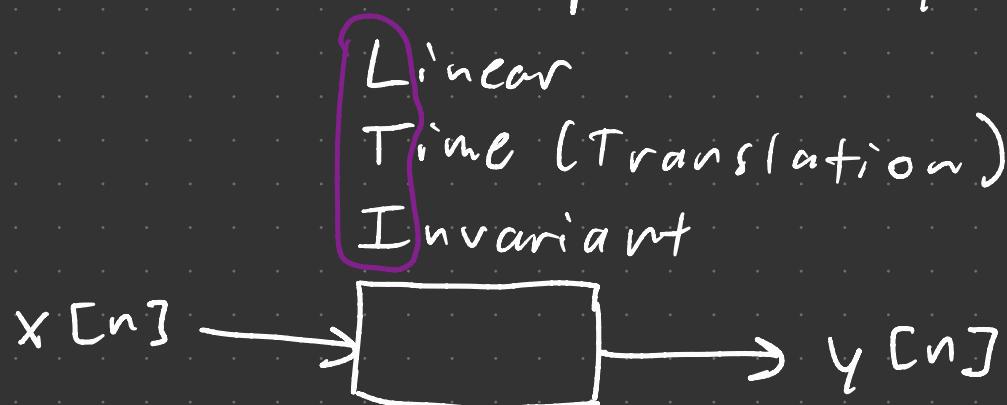
Two kinds of filters:

- ① Finite Impulse Response (FIR)
- ② Infinite Impulse Response (IIR)

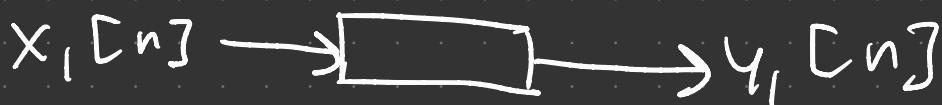
Q: What is the difference?

Q: What is an impulse response?

sampled
signals
 $n \in \mathbb{Z}$



Linear:



$$\alpha x_1[n] + \beta x_2[n] \rightarrow \boxed{\quad} \rightarrow \alpha y_1[n] + \beta y_2[n]$$

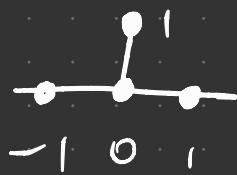
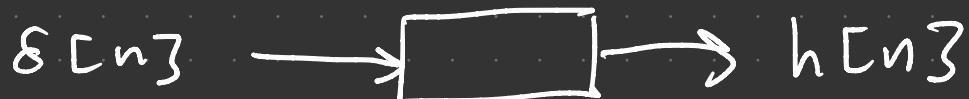
$\alpha, \beta \in \mathbb{R}$

Time / Translation Invariant:

The system does not care when it is turned on.



Impulse Response



"impulse response"

$$x[n] \rightarrow \boxed{\quad} \rightarrow y[n] = (h * x)[n]$$

$$= \sum_{k \in \mathbb{Z}} h[k] x[n-k]$$

FIR: Support of h is finite

IIR: Support of h is infinite

Support = # of nonzero values

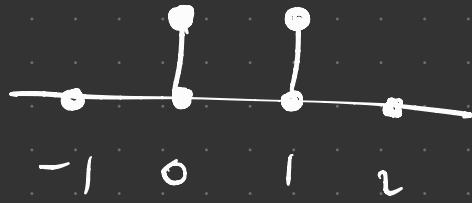
Ex: $y[n] = \frac{1}{2} (x[n] + x[n-1])$

Q: What does this system do?

- A:
- Moving average filter
 - Smoothing
 - Low-pass filter

Impulse response:

$$h[n] = \frac{1}{2} (\delta[n] + \delta[n-1])$$



FIR
2-tap filter

Q: How do we know it is low-pass?

A: Look at the frequency response

$$\begin{aligned} e^{j\omega n} &\xrightarrow{\text{Block}} \sum_{k \in \mathbb{Z}} h[k] e^{j\omega(n-k)} \\ &= e^{j\omega n} \left(\sum_{k \in \mathbb{Z}} h[k] e^{-j\omega k} \right) \\ &= \underbrace{e^{j\omega n}}_{\text{DTFT / frequency response}} \underbrace{H(e^{j\omega})}_{\text{frequency response}} \end{aligned}$$

Obs: Pure frequencies (complex exponentials) are eigen functions of LTI systems.

Convolution Theorem:

$$y[n] = (h * x)[n]$$

$$Y(e^{j\omega}) = H(e^{j\omega}) X(e^{j\omega})$$

$$Y(\omega) = H(\omega) X(\omega)$$

Signal-processing notation
reduced / math notation

Z - Transform

$$X(z) = \sum_{n \in \mathbb{Z}} x[n] z^{-n}, \quad z \in \mathbb{C}$$

$$X(e^{j\omega}) = X(z) \Big|_{z=e^{j\omega}}$$

Ex: $h[n] = \frac{1}{2} (\delta[n] + \delta[n-1])$

$$H(e^{j\omega}) = \frac{1}{2} + \frac{1}{2} e^{-j\omega}$$

$$= e^{-j\frac{\omega}{2}} \left[\frac{e^{j\frac{\omega}{2}} + e^{-j\frac{\omega}{2}}}{2} \right]$$

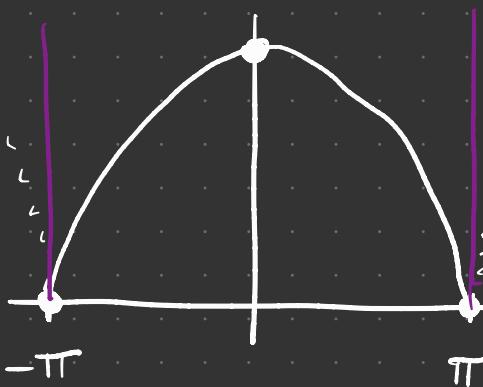
$$= e^{-j\frac{\omega}{2}} \cos\left(\frac{\omega}{2}\right)$$

phase

magnitude

$$|H(e^{j\omega})|$$

$$\hookrightarrow = e^{j\phi(\omega)} \Rightarrow \phi(\omega) = -\frac{\omega}{2}$$



low-pass

(attenuating high frequencies)

phase
response

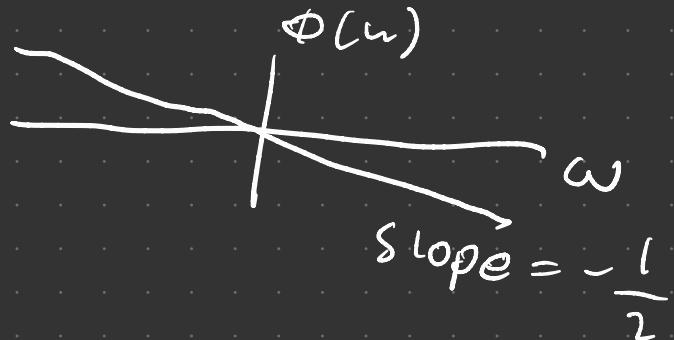
Linear-Phase Systems

Defn: Given $H(e^{j\omega}) = e^{j\phi(\omega)} |H(e^{j\omega})|$,
 the system is called linear-phase
 if $\phi(\omega)$ is linear.

Ex: $\phi(\omega) = -\frac{\omega}{2}$



$\frac{1}{2} = \text{center of symmetry}$



Symmetric / Anti-Symmetric filters provide
 a complete characterization of linear-
 phase systems.

Type I: odd length / symmetric

Type II: even length / symmetric

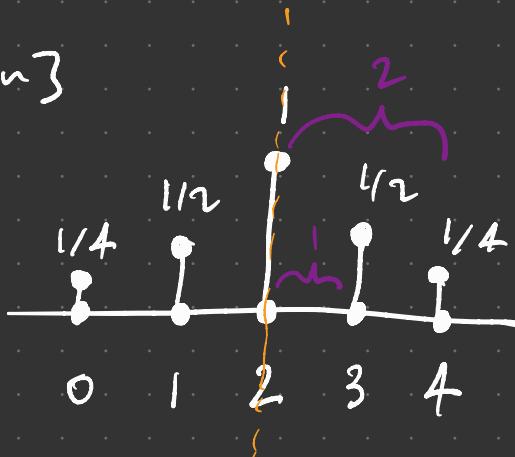
Type III: odd length / anti-Symmetric

Type IV: even length / anti-Symmetric

Type I: center of symmetry $L \in \mathbb{Z}$

Ex:

$$h[n]$$



$$L=2 \in \mathbb{Z}$$

$$H(e^{j\omega}) = e^{-j2\omega} \left[1 + 2\left(\frac{1}{2}\right) \cos(\omega) + 2\left(\frac{1}{4}\right) \cos(2\omega) \right]$$

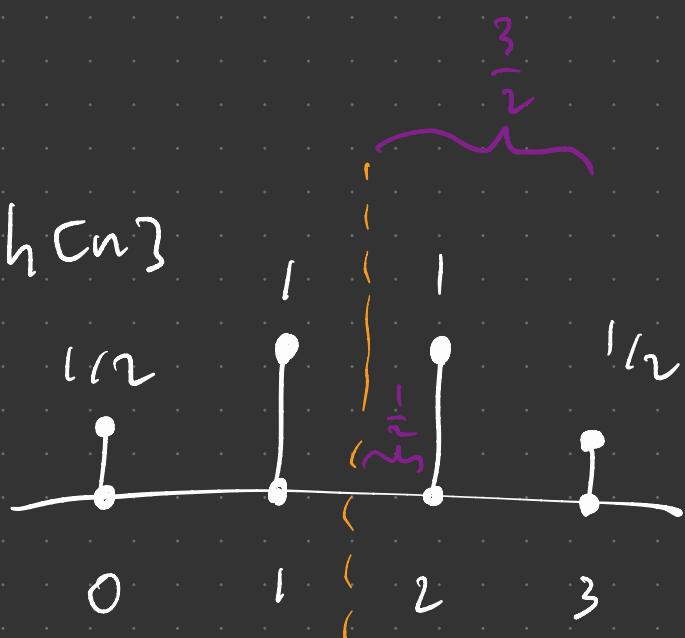
General:

$$H(e^{j\omega}) = e^{-jL\omega} \sum_{n=0}^L a[n] \cos(n\omega)$$

Exercise: Express $a[n]$ in terms of $h[n]$

Type II: Center of symmetry $L - \frac{1}{2} \in \mathbb{Z}$

Ex:



$$L = \frac{3}{2} \Rightarrow L - \frac{1}{2} = 1 \in \mathbb{Z}$$

$$H(e^{j\omega}) = e^{-j\frac{3}{2}\omega} \left[2(1) \cos\left(\frac{\omega}{2}\right) + 2\left(\frac{1}{2}\right) \cos\left(\frac{3\omega}{2}\right) \right]$$

General:

$$H(e^{j\omega}) = e^{-jL\omega} \sum_{n=0}^{L-\frac{1}{2}} b[n] \cos\left((n+\frac{1}{2})\omega\right)$$

Exercise: Express $b[n]$ in terms of $h[n]$

Q: What makes a filter symmetric?

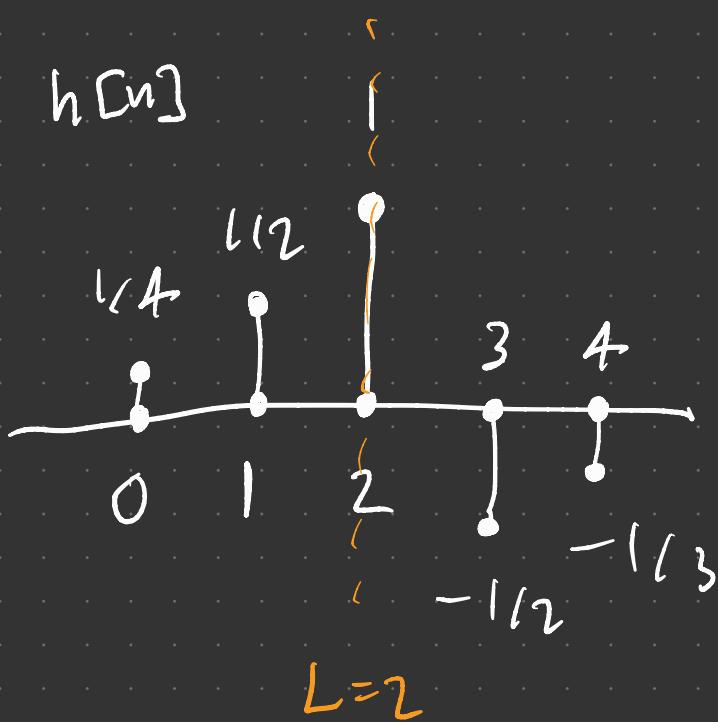
A: $h[n] = h[2L - n]$

Q: What makes a filter antisymmetric?

A: $h[n] = -h[2L - n]$

(Non)

Ex:



$h[n]$ is not
antisymmetric!

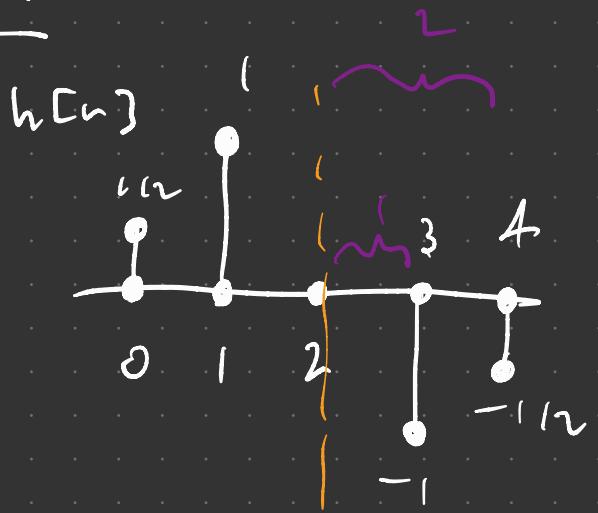
Obs: $h[n] = -h[2L - n]$

$$\Rightarrow h[L] = -h[L]$$

$$\Rightarrow h[L] = 0$$

Type #: Center of antisymm. $L \in \mathbb{Z}$

Ex:



$$L=2 \in \mathbb{Z}$$

$$H(e^{j\omega}) = e^{-j2\omega} \left[2(1)j \sin(\omega) + 2\left(\frac{1}{2}\right)j \sin(2\omega) \right]$$

$$j = e^{j\frac{\pi}{2}}$$

$$= e^{j(-2\omega + \frac{\pi}{2})} \left[2 \sin(\omega) + \sin(2\omega) \right]$$

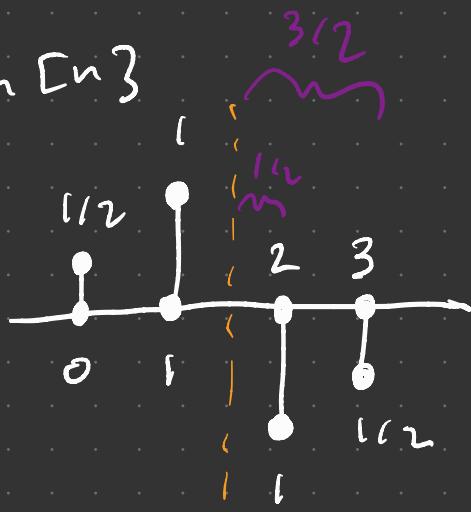
General:

$$H(e^{j\omega}) = e^{j(-L\omega + \frac{\pi}{2})} \sum_{n=0}^L c[n] \sin(n\omega)$$

Exercise: Express $c[n]$ in terms of $h[n]$

Type IV: Center of antisymm. $L - \frac{1}{2} \in \mathbb{Z}$

Ex:



$$L = \frac{3}{2} \Rightarrow L - \frac{1}{2} = 1 \in \mathbb{Z}$$

$$H(e^{j\omega}) = e^{j(-\frac{3}{2}\omega + \frac{\pi}{2})} \left[2(1) \sin\left(\frac{\omega}{2}\right) + 2\binom{1}{2} \sin\left(\frac{3\omega}{2}\right) \right]$$

General:

$$H(e^{j\omega}) = e^{j(-L\omega_f \frac{\pi}{2})} \sum_{n=0}^{L-\frac{1}{2}} d[n] \sin\left(\omega n + \frac{1}{2}j\omega\right)$$

Exercise: Express $d[n]$ in terms of $h[n]$.