Bachelor of Engineering Electronic Engineering (HONS)

Headphone Amplifier Design

By

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Chapter 1

Simple transistor circuit

1.1 Transistor basic property

Figure 1.1 shows the basic NPN bipolar junction transistor circuit.

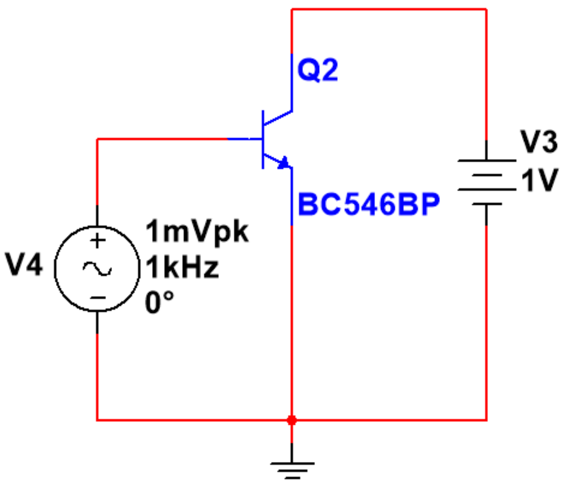


Figure 1.1: Single transistor circuit

We can get transistor operating state from simulation result as Table 1.1. It's obvious that I_C and I_E is proximately 200 times greater than I_B which is the main function of transistor.

Equation 1.1 defines β which is the most important parameter of transistor.

$$\beta = \frac{I_C}{I_B} \tag{1.1}$$

I_B	9.09789μ
I_C	$2.02293m$
I_E	$-2.03003m$

Table 1.1: DC operating point analysis result

1.2 Relationship between Base voltage(V_{be}) and Collector current(I_c)

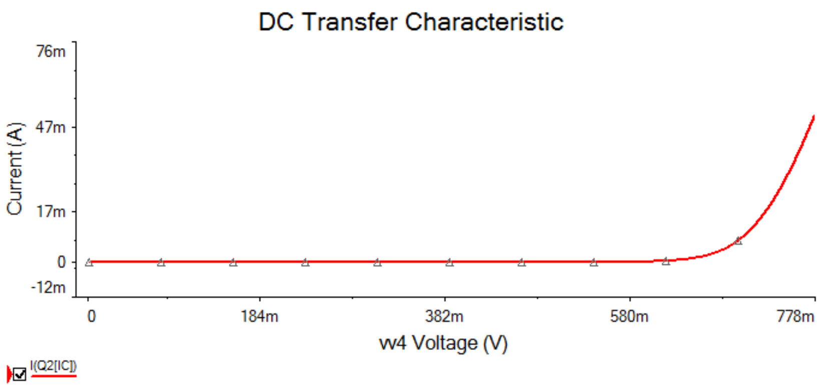


Figure 1.2: V_{be} and I_c curve

After running DC sweep command on $V4$ in circuit of Figure 1.1, We can get the curve of Figure 1.2. This illustrate that when $V_{be} < 0.65V$, I_c is very small and when $V_{be} > 0.65V$, I_c is increase significantly. Therefore, We can simply consider that when $V_{be} > 0.65V$, transistor is on.

1.3 Limit current gain

Generally, we need a method to control the current gain as we want. Figure 1.3 is a simply solution by adding transistor R_C and R_E.

We can derive voltage gain A_V with Equation 1.2. And in circuit in Figure 1.3, A_V is approximate 5 theoretically.

$$A_V \triangleq \frac{V_{out}}{V_{in}} \approx -\frac{R_C}{R_E} \tag{1.2}$$

From simulation result in Figure 1.4, the practical $A_V = \frac{7.6486m}{2m} = 3.8243$ which is close to theoretic value.

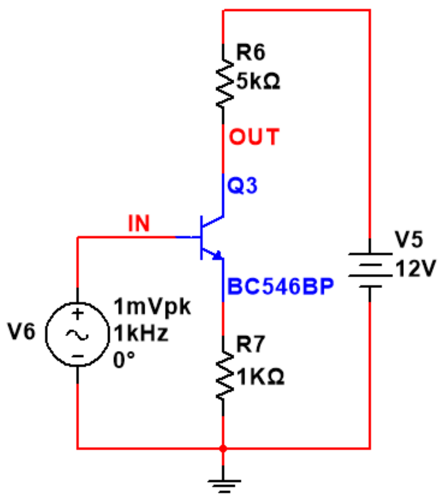
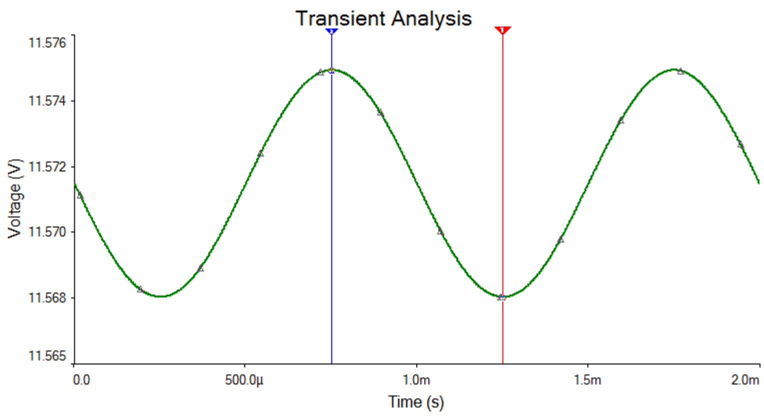


Figure 1.3: Basic transistor circuit with R_c and R_e



	V(in)	V(out)
x1	1.2508m	1.2508m
y1	658.7000m	11.5676
x2	749.1749µ	749.1749µ
y2	656.7000m	11.5752
dx	-501.6502µ	-501.6502µ
dy	-1.9999m	7.6486m
dy/dx	3.9867	-15.2469
1/dx	-1.9934k	-1.9934k

Figure 1.4: Output of the circuit in Figure 1.3

1.4 Add voltage divider

As we know, we need make sure $V_{be} > 0.65V$ for transistor operating correctly. But in practical application, it's hard to keep input signal always meeting this requirement. So we can add capacitor and voltage divider solve this problem like Figure 1.5. In which, capacitor block the original DC voltage of input signal and voltage divider add the DC voltage which we require to signal. Finally, we use another capacitor for outputting pure AC signal form our circuit.

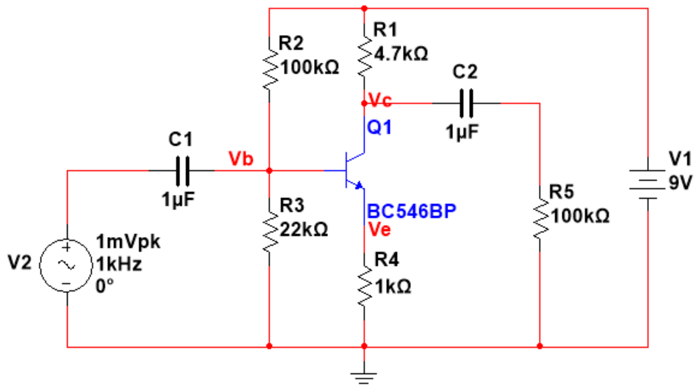


Figure 1.5: Add voltage divider and capacitors

Chapter 2

Negative Feedback

2.1 Simple Negative Feedback system

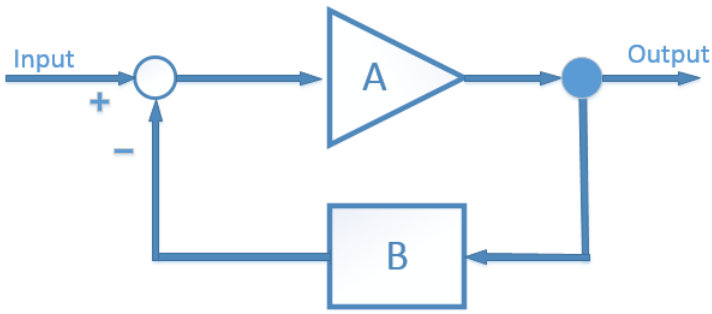


Figure 2.1: simple negative feedback system

Figure 2.1 show a simple negative feedback system in which A is ideal amplifier and B is feedback network.

$$\frac{V_{out}}{V_{in}} = B$$

2.2 Implement Using Op-amp

In Figure 2.2, an Op-amp 741 is used to implement the negative feedback circuit in Figure 2.1.

741 is part A while R1 and R2 form feedback network.

$$\frac{V_{out}}{V_{in}} = B = \frac{R1 + R2}{R2} = \frac{10K + 1K}{1K} = 11$$

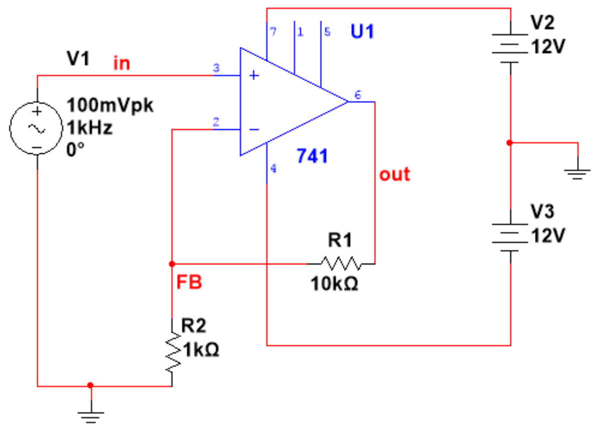
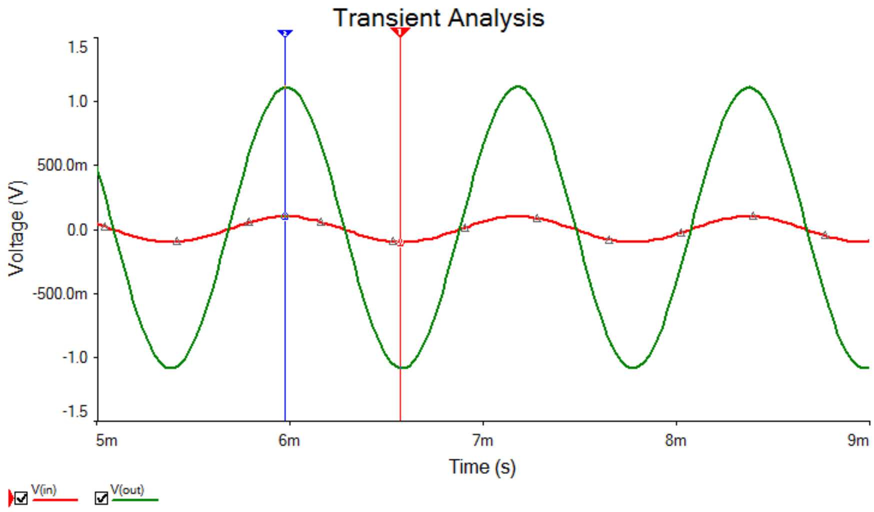


Figure 2.2: Implement negative feedback circuit with Op-amp



	V(in)	V(out)
x1	6.7433m	6.7433m
y1	-99.9112m	-1.0862
x2	6.2453m	6.2453m
y2	99.9562m	1.1110
dx	-498.0000µ	-498.0000µ
dy	199.8674m	2.1972
dy/dx	-401.3402	-4.4121k
1/dx	-2.0080k	-2.0080k

Figure 2.3: Op-amp feedback simulation result curve

From simulation result in Figure 2.3,

$$B = \frac{\Delta V_{out}}{\Delta V_{in}} = \frac{2.1972}{199.8674m} = 10.99328$$

Obviously, the simulation result is very close to our estimation.

2.3 Implement Using transistor

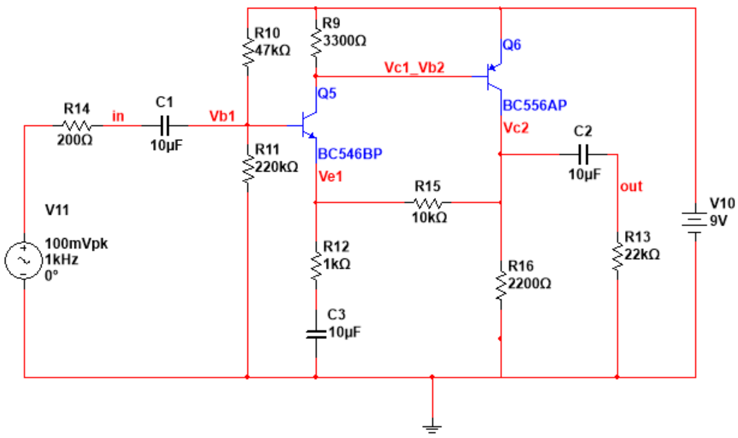
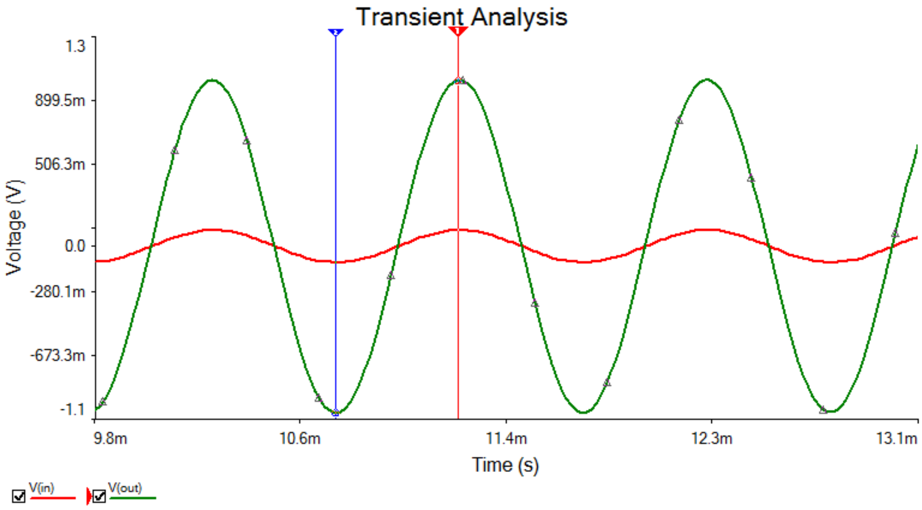


Figure 2.4: Implement negative feedback circuit with transistor

In Figure 2.4 circuit, Op-amp replaced by circuit in Figure 1.5. R12 and R15 make up feedback network which $B = \frac{R15+R12}{R12} = 11$. As we see in Figure 2.3, the output is reverse to input. Therefore, we add another transistor Q6 to eliminate the phase difference of signal.

Apparently, there's no phase difference between input and output signal. The voltage gain of circuit in Figure 2.4 is $Gain = \frac{\Delta V_{out}}{\Delta V_{in}} = \frac{2.0534}{198.8221m} = 10.3707$. It's also very close to theory result.



	V(in)	V(out)
x1	11.2431m	11.2431m
y1	99.3884m	1.0254
x2	10.7451m	10.7451m
y2	-99.4337m	-1.0281
dx	-498.0000µ	-498.0000µ
dy	-198.8221m	-2.0534
dy/dx	399.2412	4.1234k
1/dx	-2.0080k	-2.0080k

Figure 2.5: transistor feedback circuit simulation result

Chapter 3

Current Source

3.1 Single current source circuit

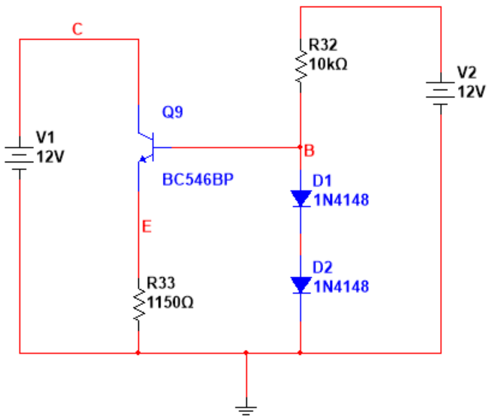


Figure 3.1: single current source circuit

Generally, we need a constant current source in circuit and the most classic one shows in Figure 3.1.

As we know the forward voltage cross a diode is about 0.65V which is approximately equal to $V_{be} of transistor$.

$$V_B = 2 \times V_{diode} = V_{be} + V_{R33}$$

Therefore:

$$V_{R33} = V_{diode} = 0.65V$$

$$I_e = \frac{V_{R33}}{R_{33}} = \frac{0.65V}{1150\Omega} = 565.217\mu A$$

DC Operating Point Analysis

	Variable	Operating point value
1	@qq9[ic]	503.27881 u
2	@qq9[ie]	-505.07050 u

Figure 3.2: single current source circuit simulation result

From Figure 3.2, we can see simulation result is close to the value we calculated. This simple circuit are able to supply constant current.

3.2 Use current source to replace output resistor

Now we can use current source to replace the R_{16} in circuit of Figure 2.4. Current source can supply stable current output. Till this step, we have finished the voltage amplifier part circuit but current of output is still enough to drive a headphone.

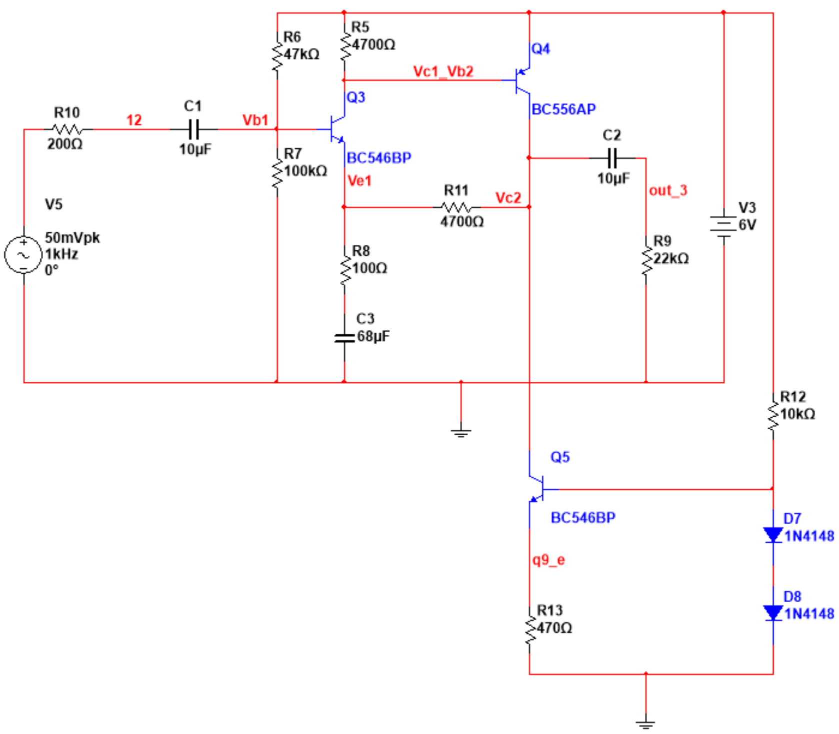


Figure 3.3: the circuit after adding current source

Chapter 4

Output Stage