

# RF Quadrature Signals

Mathematical Model

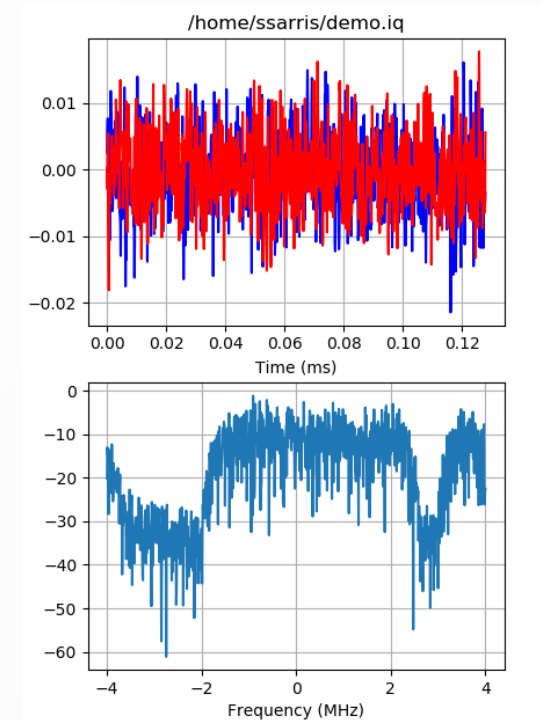
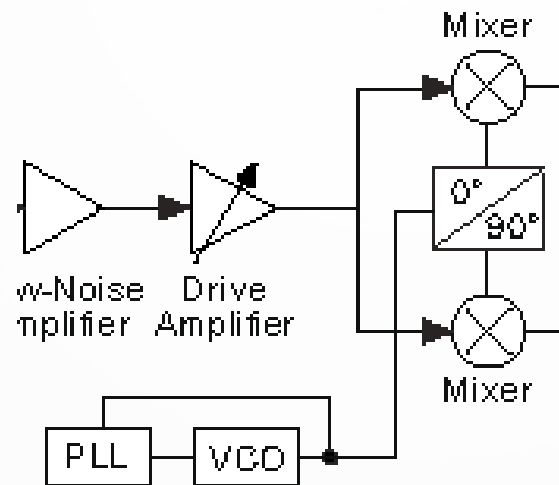
Hardware Implementation of Model

Digital Data Representation

Spiro Sarris. 07 / 2018



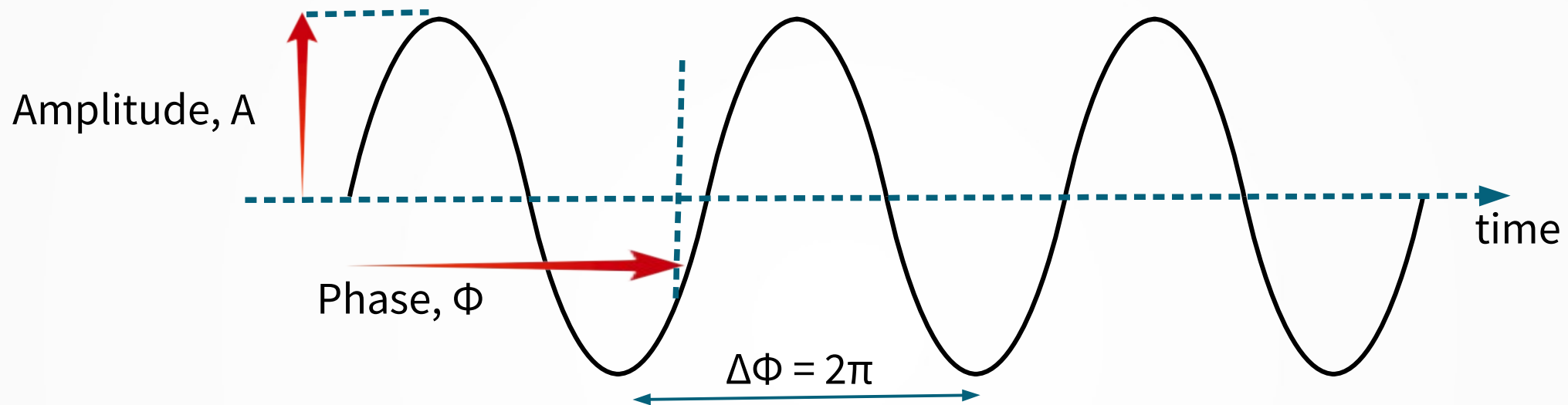
$$A \cdot \cos(\omega \cdot t + \phi)$$



# Outline

1. Amplitude, phase. Why is it interesting?
2. Euler's formula
3. Mathematical model of signal
4. Receiver hardware implementation of model
5. Analog to digital conversion
6. Convenience in software
7. Tradeoffs between I/Q sampling and real-value sampling

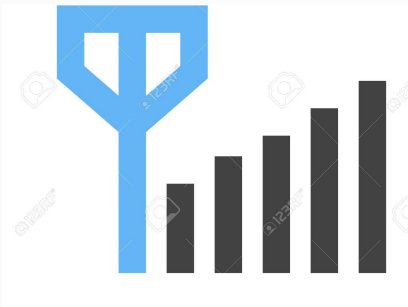
# 1. Amplitude, Phase



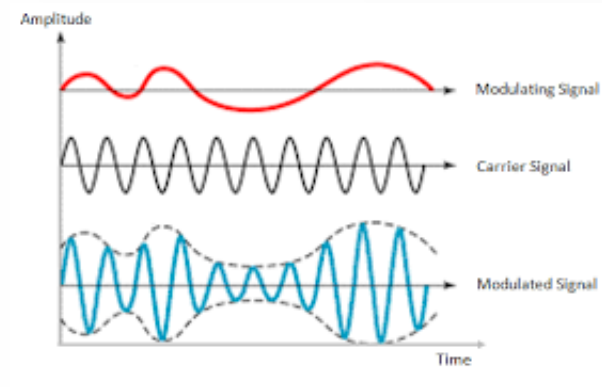
- $\Delta\Phi = 2\pi$  radians for 1 cycle in time and for 1 wavelength ( $\lambda$ ) [m] in space
- Angular frequency ( $\omega$ ) [radians/sec] is the rate of change of phase vs. time  $\omega = \frac{d\phi}{dt}$
- Temporal frequency ( $f$ ) [Hz] is the number of cycles per second  $f = \frac{\omega}{2\pi}$

# 1. Amplitude

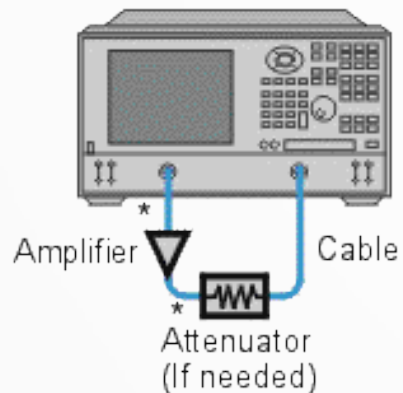
RSSI – Received Signal Strength Indicator



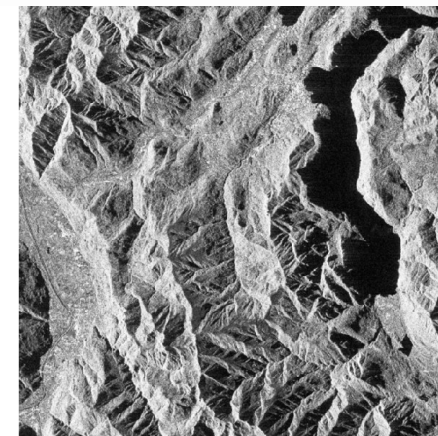
Communication Modulation



Component Gain / Loss

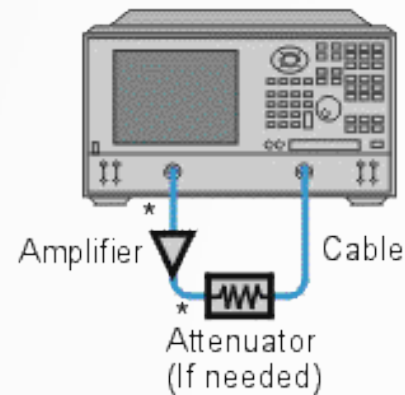


Synthetic Aperture Radar Image

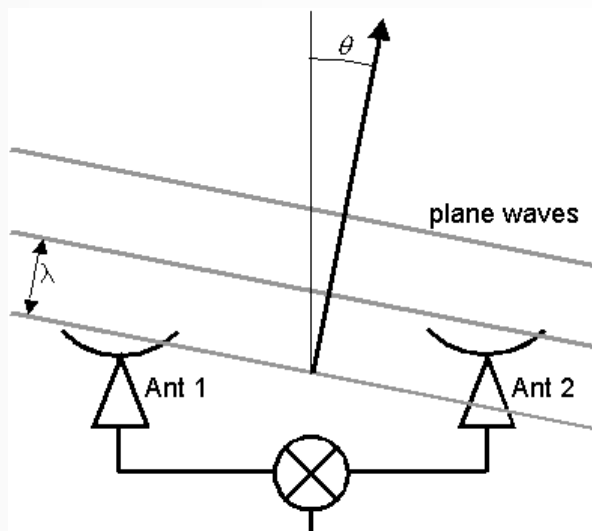


# 1. Phase

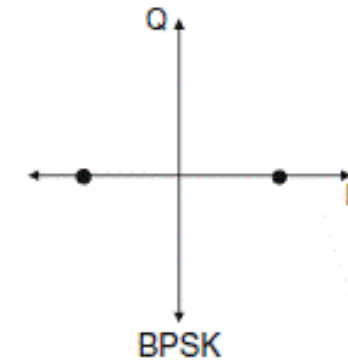
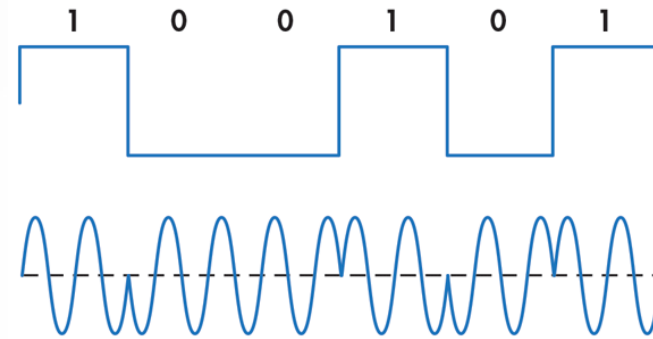
Length / Distance Measurement



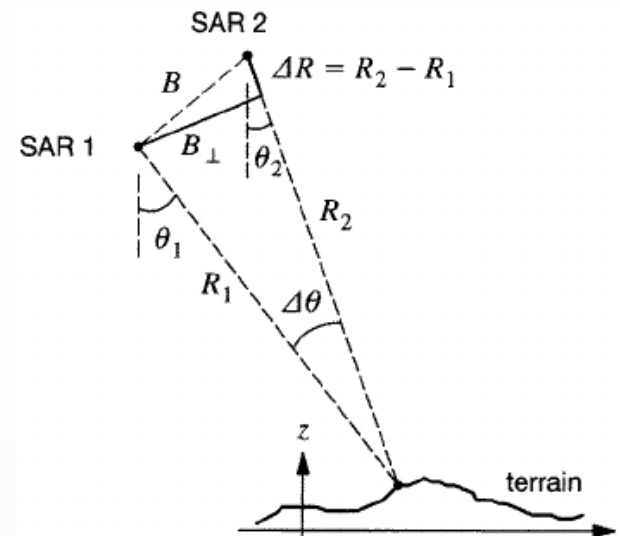
Radio Interferometer



Data Communication Modulation



Synthetic Aperture Radar Interferometer



# 2. Euler's Formula

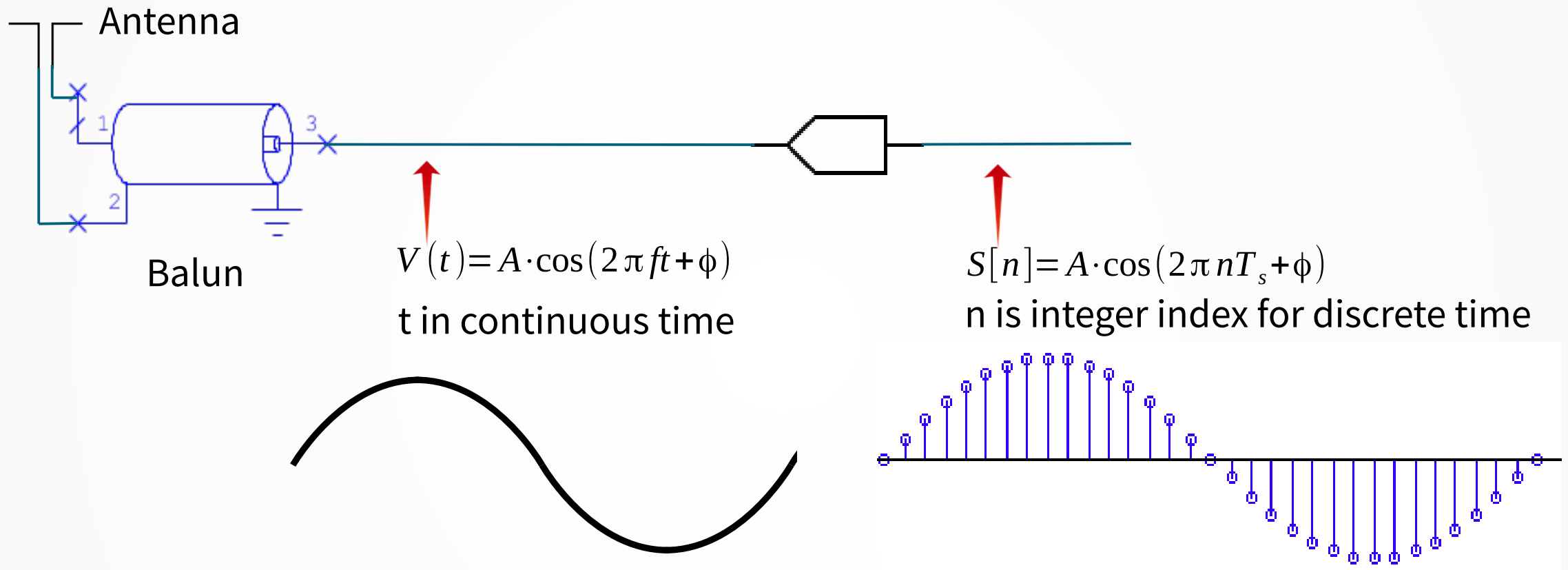


$$j = \sqrt{-1}$$

$$A \cdot e^{j \cdot \phi} = A \cdot [\cos(\phi) + j \cdot \sin(\phi)]$$

- Year 1748 - Mathematician Leonhard Euler published that periodic functions such as cosine and sine can be represented as complex exponentials of instantaneous phase (phasors)
- This model simplifies math related to amplitude and phase

# 3. Mathematical Model of Signal

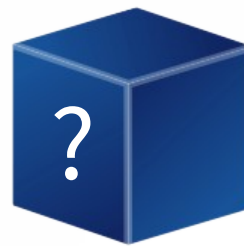
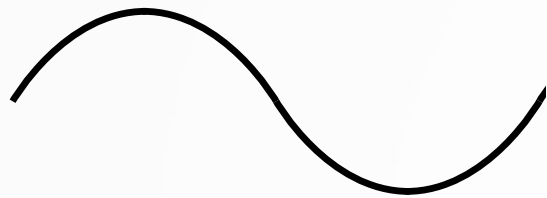


At any point in space, we can measure real-valued cosine. Convenient, but .. not ideal for DSP software that requires quadrature sampled signal.



# 3. Mathematical Model of Signal

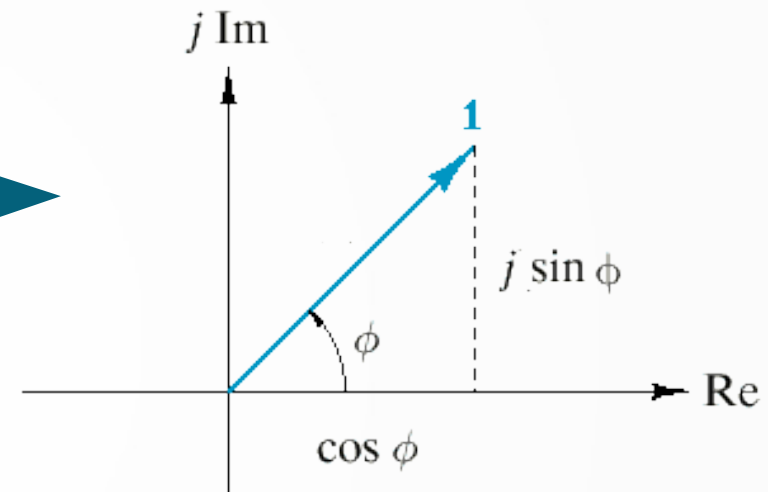
Real-value cosine



Receiver



Euler's model of complex phasor



$$V(t) = A \cdot \cos(2\pi ft + \phi)$$



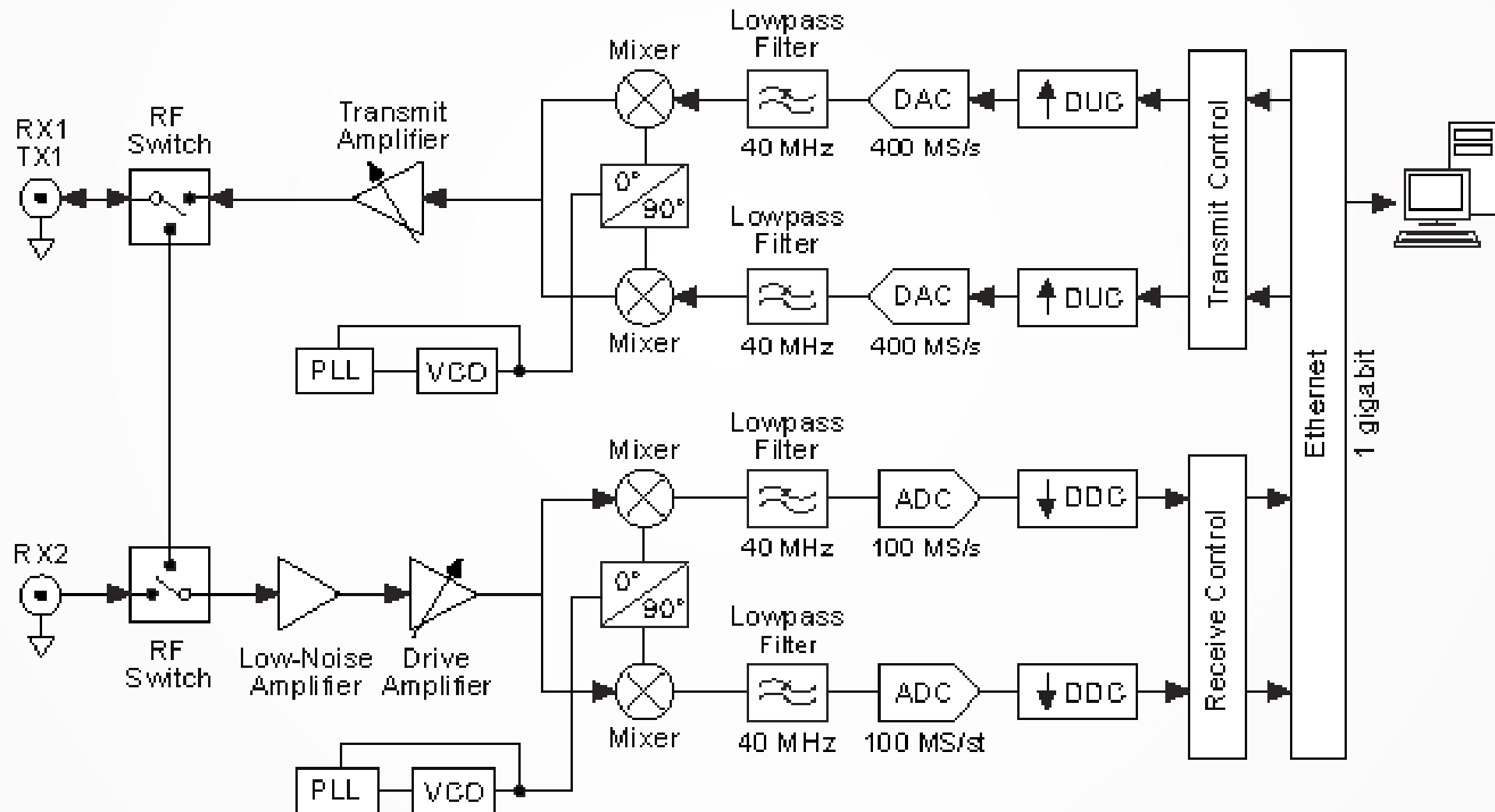
$$A \cdot e^{j\phi} = A \cdot [\cos(\phi) + j \cdot \sin(\phi)]$$

How can we translate real-value signal model (easy to understand in the physical world) to quadrature I/Q signal model (easy to process using DSP software)?

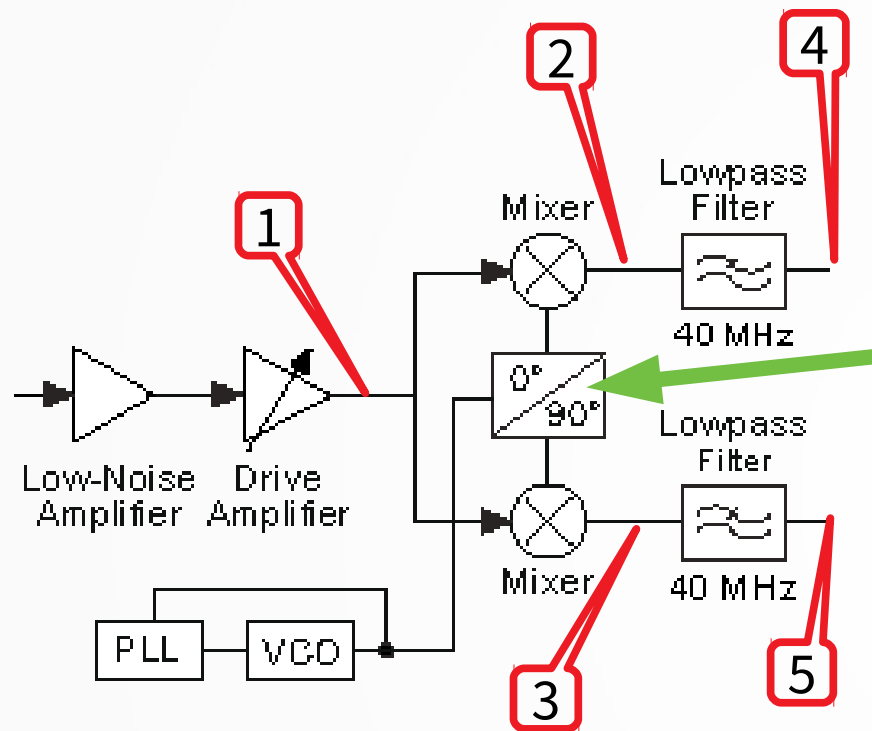


# 4. Hardware Implementation of Model

- Option 1- Direct Conversion Receiver (example from Ettus USRP N210 + WBX)



# 4. Direct Conversion Receiver



$$f_{IF} = f_{RF} - f_{LO} \rightarrow$$

$$V_1(t) = A \cdot \cos(2\pi f_{RF}t + \phi_0)$$

Use trig identity - product to sum

$$V_2(t) = A \cdot \cos(2\pi f_{RF}t + \phi_0) \cdot \cos(2\pi f_{LO}t) \dots$$

$$A \cdot \cos[2\pi(f_{RF} + f_{LO})t + \phi] + A \cdot \cos[2\pi(f_{RF} - f_{LO})t + \phi]$$

Lowpass filter removes

$$V_3(t) = A \cdot \cos(2\pi f_{RF}t + \phi_0) \cdot \cos(2\pi f_{LO}t - \frac{\pi}{2}) \dots$$

$$A \cdot \cos(2\pi f_{RF}t + \phi) \cdot \sin(2\pi f_{LO}t) \dots$$

$$A \cdot \cos[2\pi(f_{RF} + f_{LO})t + \phi] + A \cdot \sin[2\pi(f_{RF} - f_{LO})t + \phi]$$

90 degree phase shift

Lowpass filter removes

$$V_4(t) = A \cdot \cos[2\pi(f_{RF} - f_{LO})t + \phi]$$

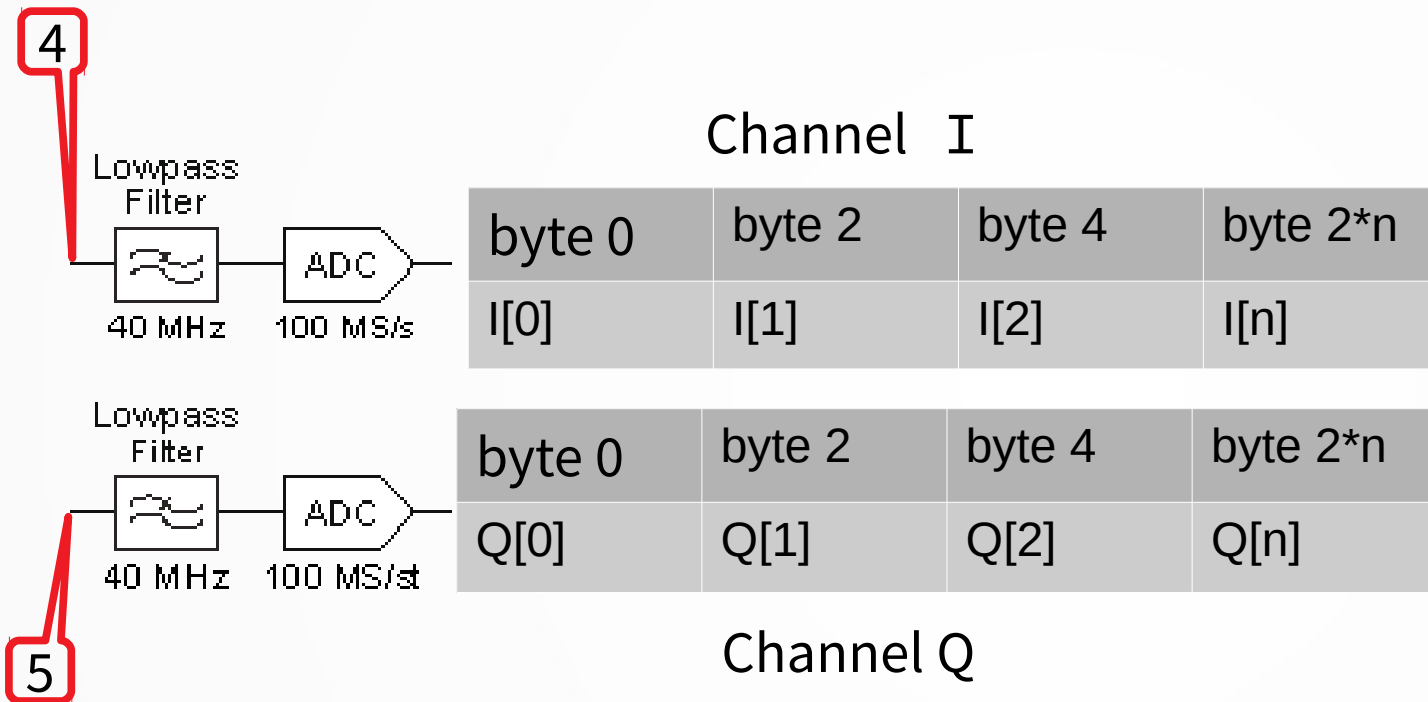
$$V_5(t) = A \cdot \sin[2\pi(f_{RF} - f_{LO})t + \phi]$$

$$V_4(t) = A \cdot \cos[2\pi f_{IF}t + \phi]$$

$$V_5(t) = A \cdot \sin[2\pi f_{IF}t + \phi]$$

# 5. Analog to Digital Conversion

- 2 ADCs with same sample clock and same length of cable / PCB trace



$$S_I[n] = A \cdot \cos(2\pi f_{IF} nT_s + \phi)$$

$$S_Q[n] = A \cdot \sin(2\pi f_{IF} nT_s + \phi)$$

Translate?

How to combine I and Q channel data into complex exponential (phasor) in software?

$$S_{iq}[n] = S_I[n] + j \cdot S_Q[n] = e^{j(2\pi f_{IF} nT_s + \phi)}$$

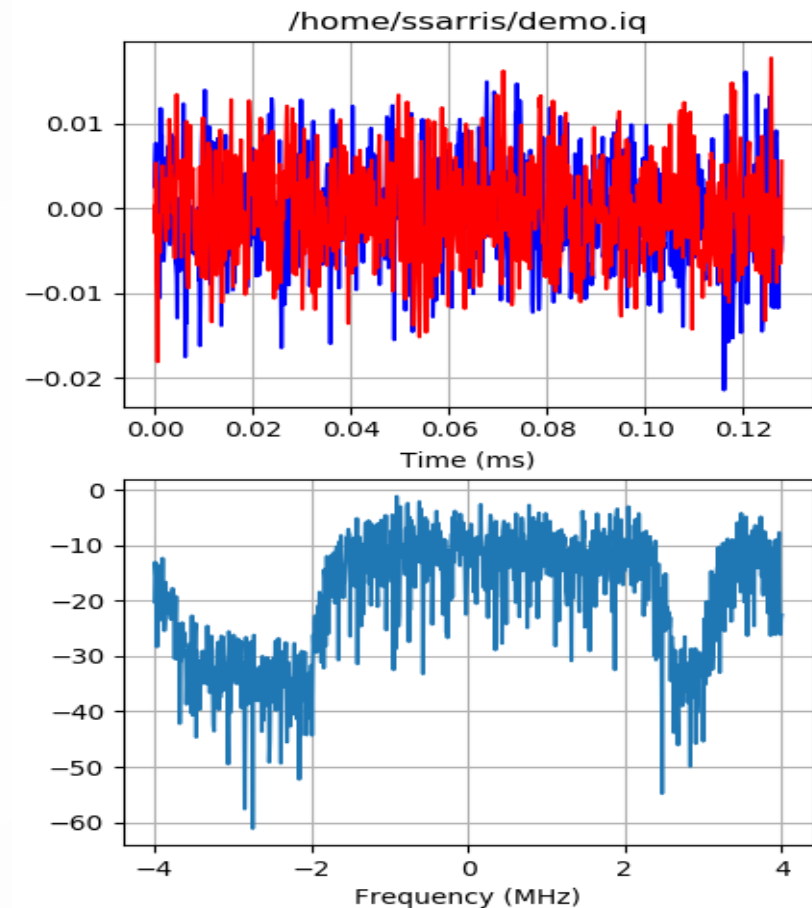
# 6. Convenience in Software

- Programming languages include support for complex/imaginary numbers and exponential function.
- Many libraries of software are available to process data in this format

Language	$\sqrt{-1}$	$e^x$
Python	1j	<code>numpy.exp(x)</code>
Matlab	J, j, I, i	<code>exp(x)</code>
C	<complex.h> I	<complex.h> <code>cexp(x)</code>
Octave	J, j, I, i	<code>exp(x)</code>

# 6. Convenience in Software

```
• import numpy as np
• import matplotlib
• matplotlib.use("QT4Agg")
• import matplotlib.pyplot as plt
•
• fftsize = 1024
• nsamples = 1024
• fs = 8e6
• tt = np.arange(0,nsamples)/fs
• ff = np.linspace(-fs/2,fs/2,fftsize)
• ftest = 100
•
• fp = open('/home/ssarris/demo.iq','rb')
• fp.seek(0,0)
• dt = np.dtype([('i',np.float32),
•               ('q',np.float32)])
• data = np.fromfile(fp,dtype=dt,count=nsamples)
• sigt = data['i']+1j*data['q']
• sigf = np.fft.fft(sigt,fftsize)
```



$$S_{iq}[n] = S_I[n] + j \cdot S_Q[n] = e^{(j2\pi f_{IF} n T_s + \phi)}$$

# 7. Tradeoff: Real-value vs I/Q sampling

- TODO – graphic of spectrum.
- TODO – graphic of streams of bytes. Overall samples and total data rate is the same.