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POKHARA UNIVERSITY

Level: Bachelor

Semester: Spring

Year : 2017

Programme: BE

Full Marks: 100

Course: Communication System Engineering

Pass Marks: 45

Time : 3hrs.

Candidates are required to give their answers in their own words as far as practicable.

The figures in the margin indicate full marks.

Attempt all the questions.

1.
 - a) Briefly explain the types of communication channels that play important role in modern telecommunication. 5
 - b) Explain the process of ASK with block diagram of ASK generator. What is the major drawback it goes through? 5
 - c) The bit stream 0010011001111 is to be transmitted using DPSK. Determine the encoded sequence and transmitted phase sequence. 5
2.
 - a) An single tone FM is represented $V(t) = 12 \sin(6 \times 10^8 \pi t + 5 \cos 1250 \pi t)$ Find the carrier and modulating frequencies, the modulation index, and the maximum deviation of the FM wave. Also find the bandwidth of the FM wave. What power will the FM wave dissipate in 16Ω resistor? 8
 - b) Prove that "The total power in the two sidebands of the AM wave with 100% modulation is only one third of the total power in the modulated wave". Also a certain transmitter radiates 9KW with the carrier unmodulated and 10.5KW when the carrier is simultaneously modulated. Calculate the modulation index if another sinewave corresponding to 40% modulation, is transmitted simultaneously and determine the total radiated power. 7
3.
 - a) Briefly explain the generation of DSB-SC signal using balanced modulator to give the general expression of output $v_0 = k \cdot x(t) \cos w_c t$, where $x(t)$ is the baseband signal. 7
 - b) What is the maximum bit rate for a channel having bandwidth of 6.2 kHz and the SNR of 20dB? Also compute the number of level required to transmit at the maximum bit rate. 8

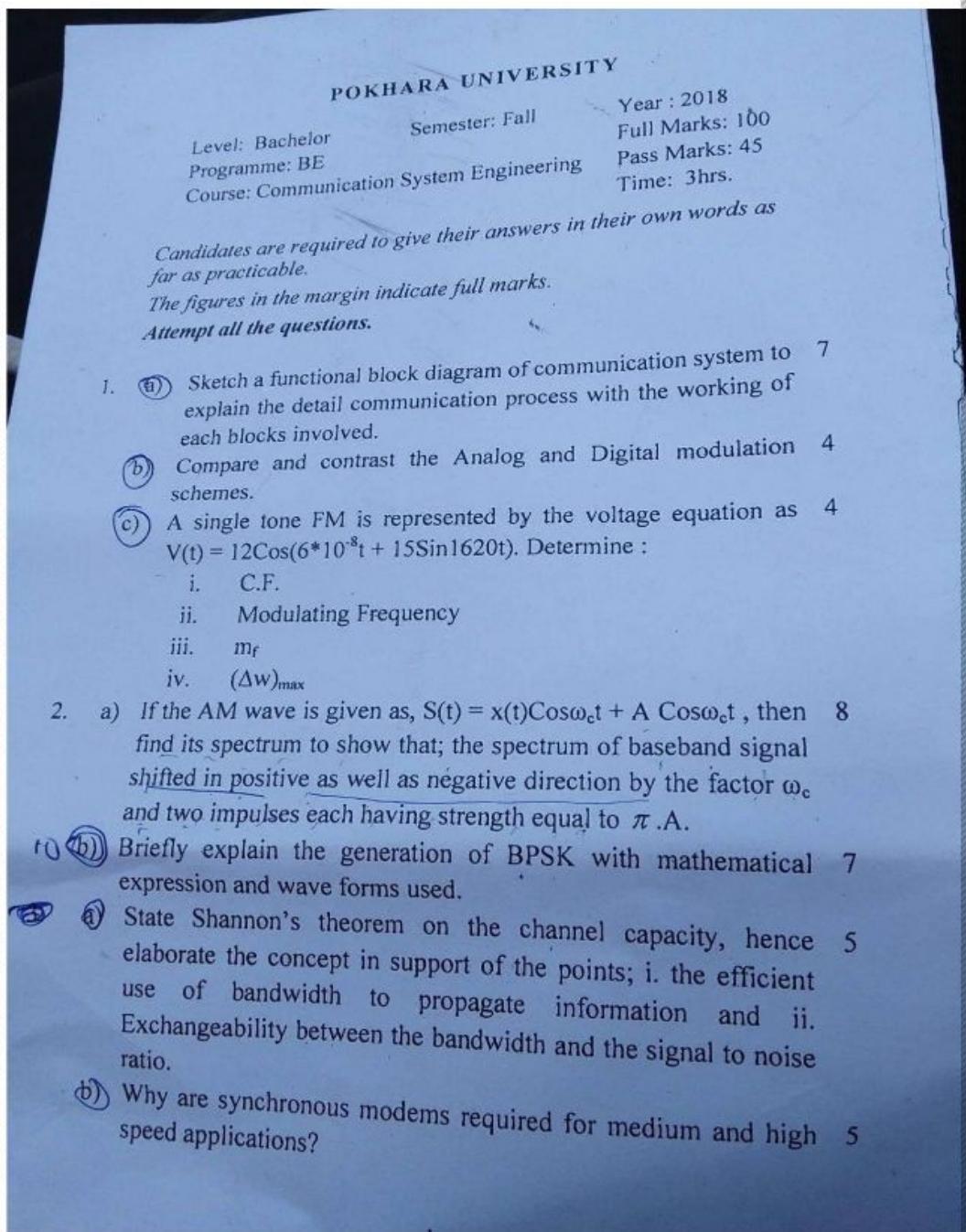
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| 4. | a) For a 12-bit data string of 101100010010, determine the number of Hamming bits required and then display the encoded data stream. | 8 |
| | b) What is the use of Modems in data communication? Contrast and compare synchronous and asynchronous modems. | 7 |
| 5. | a) Private-line circuit are leased on 24-hour basis, so explain the advantages over conventional dial-up circuits. | 5 |
| | b) Explain the role of transponder in satellite system link module, with its block diagram. | 5 |
| | c) Differentiate LEO, MEO and GEO based on the parameters: satellite height, orbital period satellite life, number of hand offs and propagation lost. | 5 |
| 6. | a) Briefly explain the importance of frequency re-use in cellular telephony. Sketch the Hand-off schemes to explain its strategy. | 7 |
| | b) Compare and contrast the SM-SI, MM-SI and multi-mode graded index optical fiber. | 8 |
| 7. | Write short notes on: (Any two) | 2×5 |
| | a) Advantages of FM over AM | |
| | b) OPGW | |
| | c) Kepler's Laws | |

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- Q) What are data communication protocols? Briefly explain with 5 functions of at least two examples.
4. Q) Define Amplitude modulation and demodulation. Briefly 7 explain with block diagram, how can you obtain a DSB-SC signal?
- (b) A given AM broadcast station transmits a total power of 5kW 8 when the carrier is modulated by a sinusoidal signal with a modulation index of 0.75. Compute:
i. The carrier power
ii. The transmission efficiency
iii. The peak amplitude of the carrier assuming the antenna to be represented by a $(50 + j0)$ Ohm load.
5. a) What is cellular system capacity? 'the number of channels assigned to a cell eventually becomes insufficient to support the required number of users' if you agree preset any two common techniques with brief explanation to provide more channels per unit coverage area. 7
- 12 b) Explain Single-mode step index and multi-mode Graded Index 8 types of optical fiber configurations..
6. a) Why are Up-link frequencies greater than down-link 7 frequencies in satellite communication? What is Roundtrip delay of geosynchronous satellite?
- b) What role does a transponder play in satellite link system, 8 briefly explain with the functionalities of each components used.
7. Write short notes on: (Any two) 2×5
- a) Generation of PCM Signal
- b) ISDN
- c) Data communication Hardware

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1. Introduction :

1. Communication process : 20/5/-

The goal of communication is to convey information. The communication process is divided into three basic components. The purpose of a communication system is to transmit message intelligence of a source to a user. Thus, communication may be defined as a process for information exchange. Communication refers to sending and receiving and processing the information by electrical means.

1.1 Elements of communication system : 18/1/-

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graph LR; A[Information source] --> B[Input transducer]; B --> C[Transmitter]; C --> D[Channel]; D --> E[Noise]; D --> F[Receiver]; F --> G[Destination];
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The essential components of a communication system are information source, input transducer, transmitter, communication channel, receiver and destination.

i) Information source :

A communication system serves to communicate a message or information. This message or information originates in the information source. In general, there can be various messages in the form of words, group of words, code, symbols, sound signal, pictures etc. However, out of these messages, only the desired message is selected and conveyed or communicated. i.e., the function of information source is to produce required message which is need to be transmitted.

yours,

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The image shows handwritten notes on a lined notebook page. At the top right, there is a red stamp with the text "Date _____" and "Page _____". Below the stamp, the number "2" is written. The notes are organized into five sections, each with a heading and a descriptive paragraph.

II. Input transducer:
The message from the information source may or may not be electrical form in nature. In case when the message produced by the information source is not electrical in nature, an input transducer is required to convert it into a time varying electrical signal.
For eg: in case of radio broadcasting a microphone converts the information which is in the form of sound waves into corresponding electrical signal.

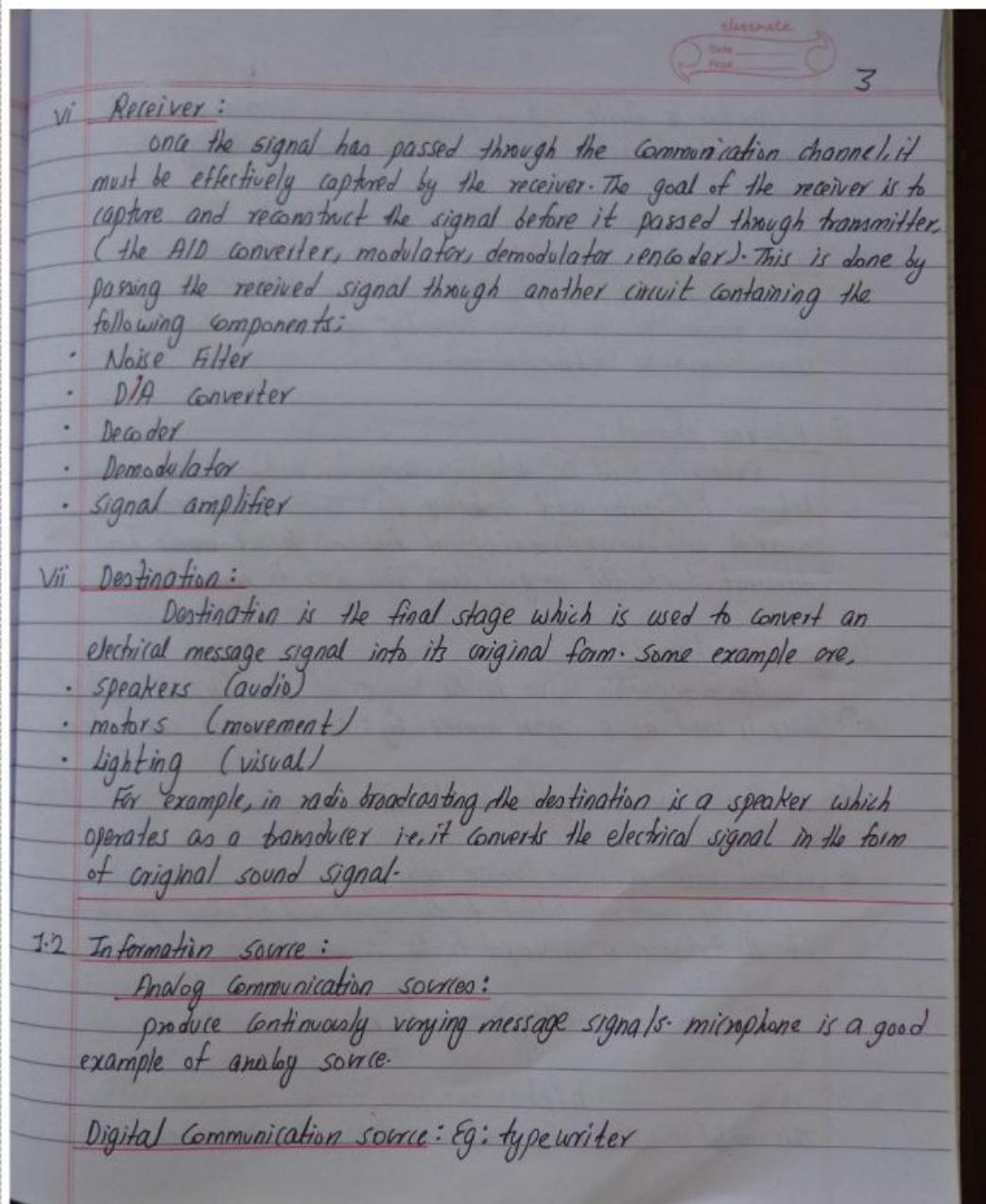
III. Transmitter:
The transmitter processes the incoming information so as to make it suitable for transmission and subsequent reception. Transmitter is a collection of electronic circuit designed to convert the message into a signal suitable for transmission.

IV. Channel:
Channel means the medium through which the message travels from the transmitter to the receiver. It can be said that the function of the channel is to provide a physical connection between the transmitter and the receiver.

V. Noise:
Noise may interfere with signal at any point in the communication system but its impact is most noticeable when it occurs in the channel or at the input of the receiver. When the noise is severe, it can make the information useless. Here, interference and distortion can also occur.

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produce a finite set of possible messages. There are finite number of characters (messages) that can be emitted by this source.

1.3. communication channel : 2016/5, 2017/5

The channel is a physical medium used to pass the signal from transmitter to receiver. Depending upon the medium, the channels can be subdivided into;

(i) wireline channel :

Extremely used in telephony, computer networks, as a link between transmitter and antenna. Eg: Twisted pair cable, co-axial cable, waveguides, optical fiber. Different media have different bandwidth ranging from few KHz to hundred GHz

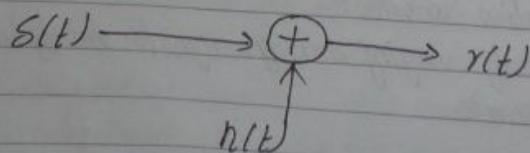
(ii) wireless channel :

Electromagnetic wave in the range of few KHz to hundred GHz is used as a signal carrier. Eg: Air, vacuum, sea water.

A channel can be modeled in the following manners;

(i) channel with additive noise only:

only noise is added to the signal and all other parameters of the channel is assumed to be constant.



This model is suitable for simple wireline communication channel

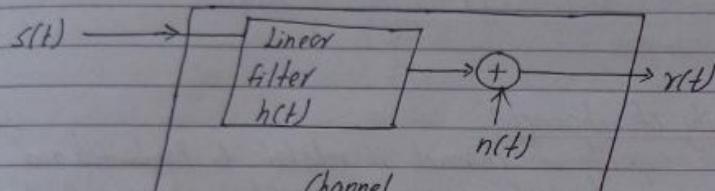
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operating at relatively low frequency.
 $r(t) = s(t) + n(t)$

i) channel with linear time invariant filter :

In most cases, filters are introduced at the inputs and outputs of the receiver and transmitter to limit the bandwidth and to avoid interference from or to other channels. In such cases, the above model is suitable for analysis. Such model is suitable for more complex wireline channel.



$$r(t) = s(t) * h(t) + n(t)$$

$$\therefore r(t) = \int_{-\infty}^{\infty} h(\tau) s(t-\tau) d\tau + n(t)$$

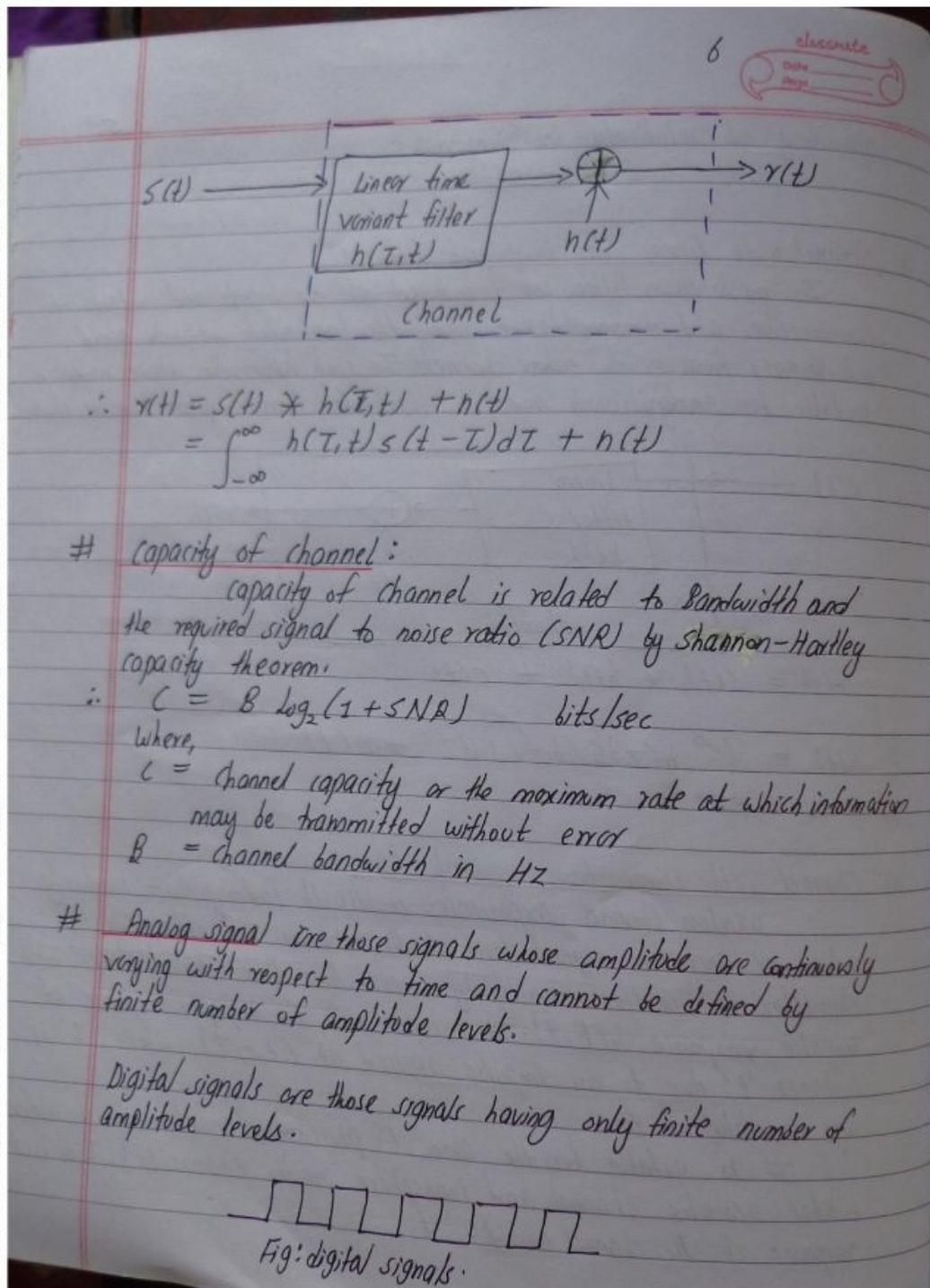
ii) channel with linear time variant filter :

wireless channel experiences multipath fading that randomly reduces or increases signal level at the receiver. The effect can be characterized mathematically by linear time variant filter with impulse response $h(E, t)$. This response is the impulse response at instance ' t ' due to an impulse applied at $(t, -\tau)$, where, ' τ ' is age or elapsed time.

It is suitable for the case of physical channels such as under water acoustic channel and ionosphere radio channels. $h(T, t)$ is the response of the channel at time t .

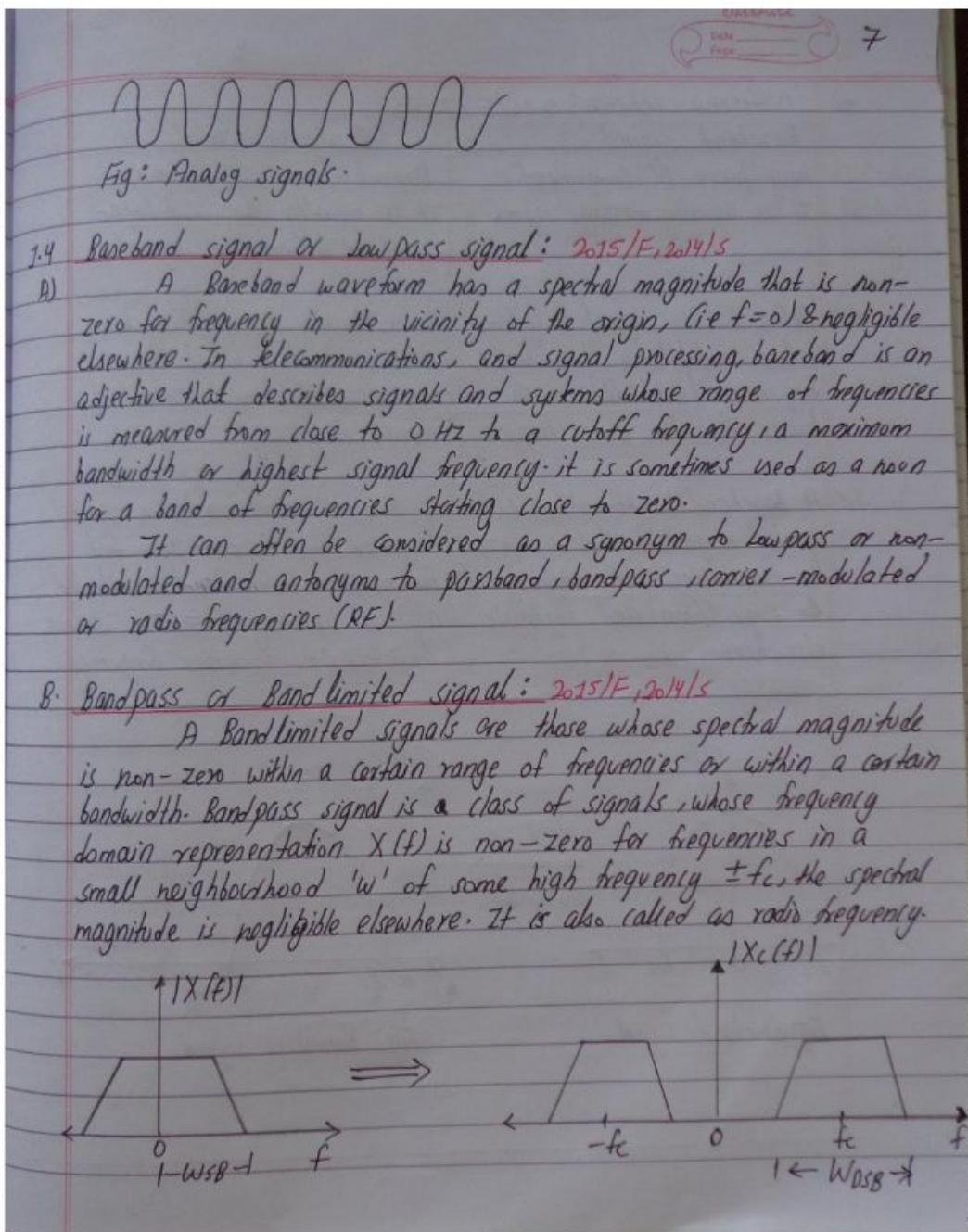
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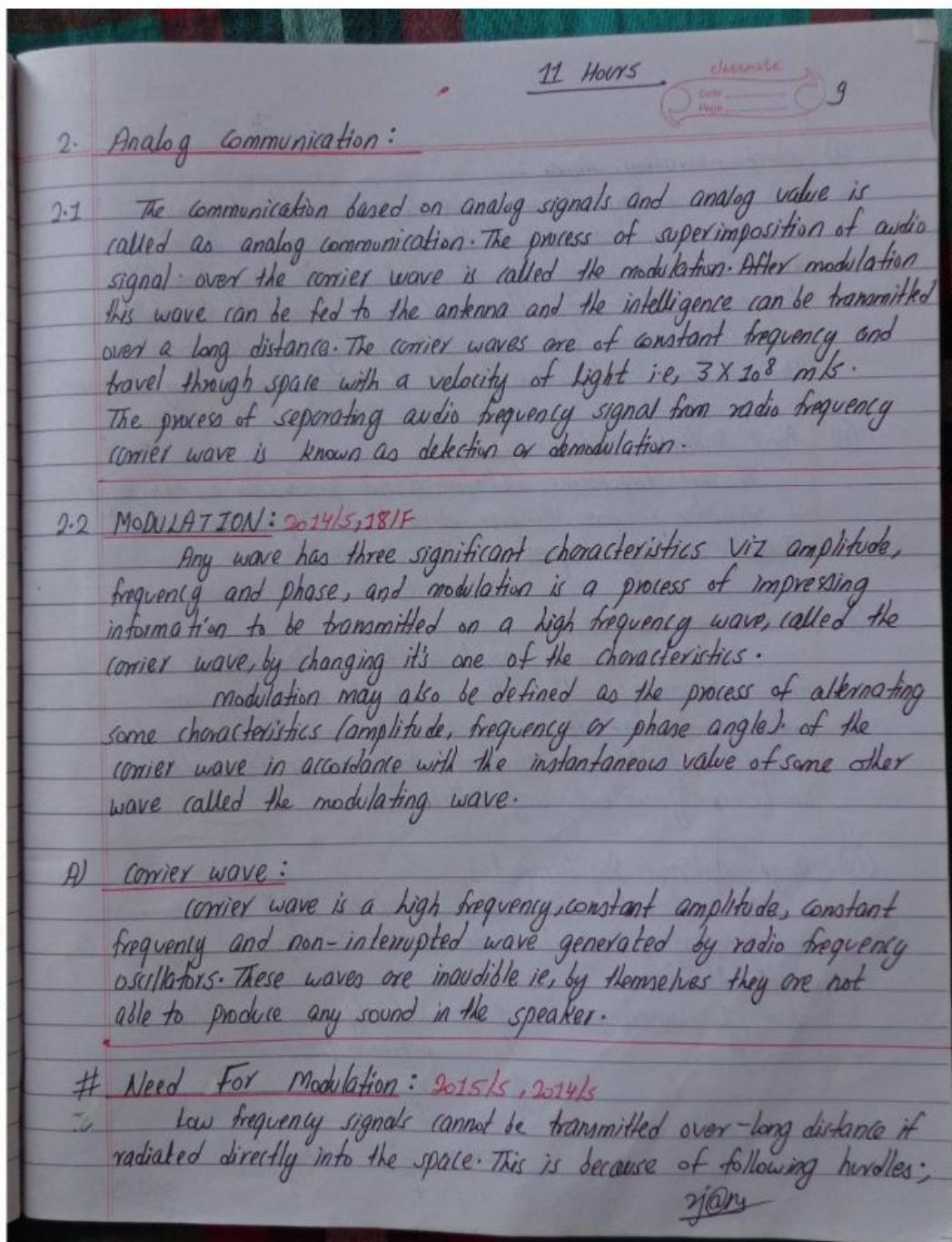
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<p># Difference between: 2015/F, 2014/S</p> <p>Baseband signal</p> <ul style="list-style-type: none">i) may have DC component.ii) It is usually message signal.iii) Refers to the signals & systems before modulation, which have frequencies / bandwidth much lower than the carrier frequency.iv) A baseband waveform has a spectral magnitude that is non-zero for frequencies in the vicinity of the origin ($f=0$) and negligible elsewhere.	<p>Bandpass signal:</p> <p>Does not have DC component.</p> <p>It is usually the modulated signal or transmitted signal.</p> <p>Refers to the signals & systems after modulation, which have frequencies / bandwidth around the carrier frequency.</p> <p>A Bandpass waveform has a spectral magnitude that is non-zero for frequencies in some band concentrated about a frequency $f = \pm f_c$, where $f_c \gg 0$; f_c = carrier frequency.</p>
<p>Fig: Baseband signal</p>	<p>Fig: Bandpass signal</p>

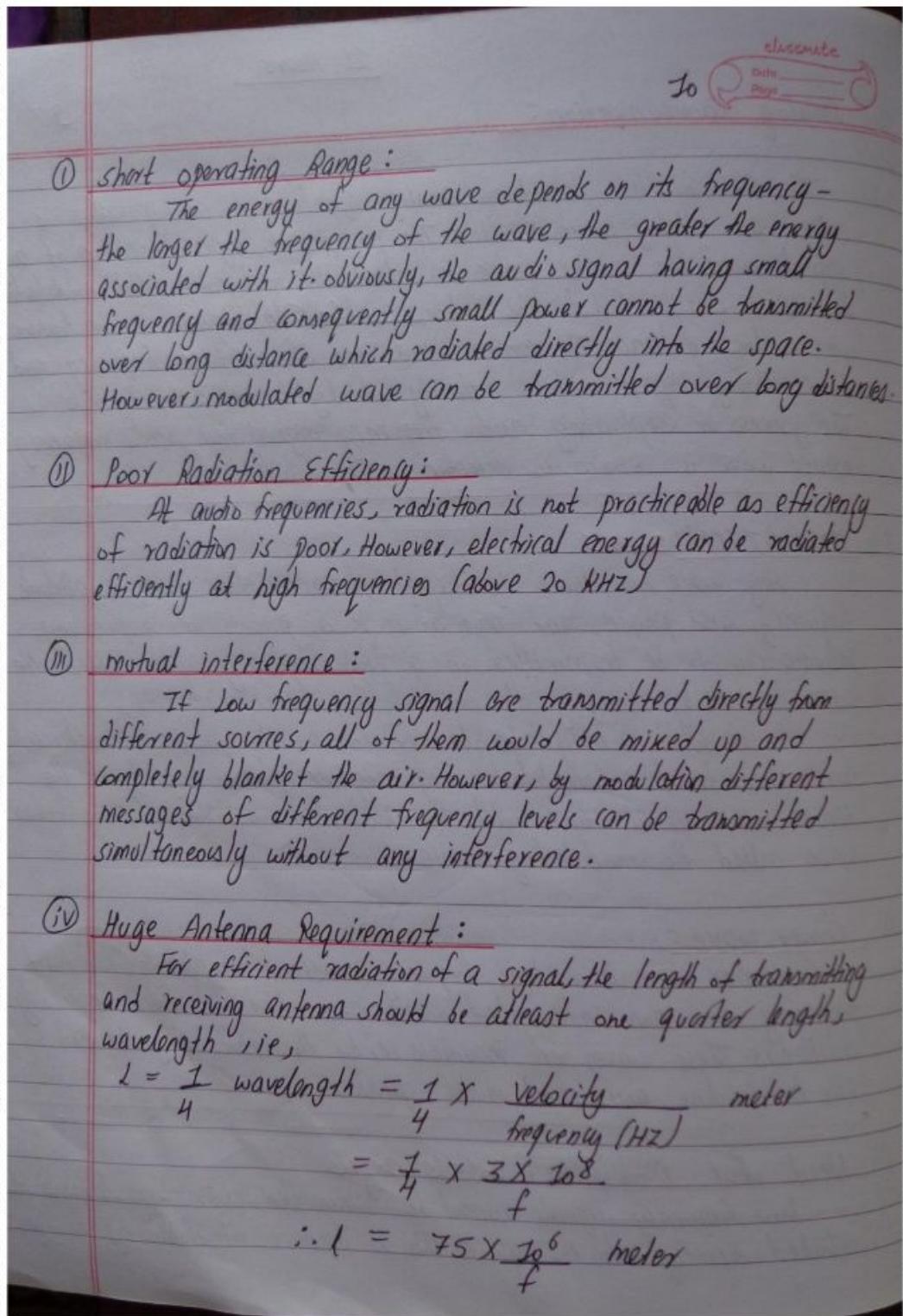
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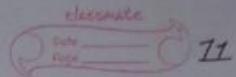
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Thus, for transmitting a signal of frequency 2kHz, an antenna of length 37.5 km will be required, practically impossible. On the other hand, for transmitting a signal of frequency 2 MHz, an antenna of about 37.5 meter would be required which can be easily constructed.

$$\text{Since, } \lambda = \frac{c}{f}$$

c : velocity of light

f = frequency of the signal to be transmitted.

V. Improves Quality of Reception:

With frequency modulation (FM) and the digital communication techniques such as QAM, the effect of noise is reduced to a great extent. This improves quality of reception.

VI. Multiplexing is possible:

Multiplexing is a process in which two or more signals can be transmitted over the same communication channel simultaneously. This is possible only with modulation. The multiplexing allows the same channel to be used by many signals. Hence, many TV channels can use the same frequency range, without getting mixed with each other or different frequency signals can be transmitted at the same time.

B. Types of modulation:

The sinusoidal carrier wave may be represented as,

$$V_c = V_c \sin(\omega_c t + \phi) \\ = V_c \sin(2\pi f_c t + \phi)$$

where,

V_c = maximum value

f_c = frequency

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ϕ = phase relation with some reference of the carrier wave.

In amplitude modulation, the amplitude of the carrier wave is varied in accordance with the modulating signal, keeping the frequency and phase of the carrier wave unchanged.

In frequency modulation, the frequency of the carrier wave is varied in accordance with the modulating signal, keeping the amplitude and phase of the carrier wave unchanged.

In phase modulation, the phase of the carrier wave is varied in accordance with the modulating signal, keeping the amplitude and frequency of the carrier wave unchanged.

However, modulation can also be classified, according to the nature of carrier wave, into continuous wave modulation and pulse modulation. The digital form of pulse modulation is known as pulse-code-modulation (PCM).

c) Amplitude Modulation:

The process of varying amplitude of the high frequency or carrier wave in accordance with the intelligence (code, voice or music) to be transmitted, keeping the frequency & the phase of the carrier wave unchanged is known as amplitude modulation.

Time domain expression and spectrum (Frequency domain representation):

Let, the carrier and modulating voltage waves be represented as,

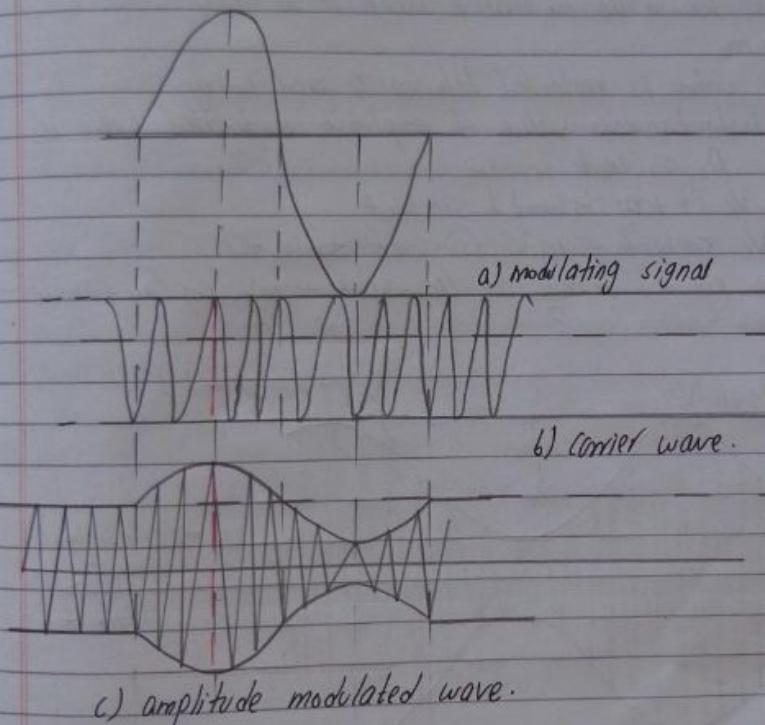
$$v_c = V_c \sin \omega_c t \quad \text{--- (1)}$$

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$$\text{and } v_m = V_m \sin \omega_m t \quad \text{--- (1)}$$

where, v_c , V_c and ω_c are instantaneous value, peak and angular velocity of the carrier and v_m , V_m , ω_m are instantaneous value, maximum value and angular velocity of the modulating signal. phase angle has been ignored in both eqn as it remains unchanged in amplitude modulation process.



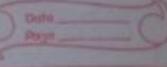
The amplitude of the carrier wave varies at a modulating signal frequency f_m . The amplitude of amplitude modulated wave is given as,

$$A = V_c + V_m$$

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$$\begin{aligned} &= V_c + V_m \sin \omega_m t \\ &= V_c \left[1 + \frac{V_m \sin \omega_m t}{V_c} \right] \\ &= V_c [1 + m \sin \omega_m t] \quad \text{--- (III)} \end{aligned}$$

where,

$m = \text{ratio of peak values of modulating signal \& carrier wave}$
and is known as modulation index = M

$$\therefore M = m$$

It's value is restricted between 0 and unity.

The instantaneous value of amplitude modulated wave is,

$$v = A \sin \omega_c t$$

$$= V_c (1 + m \sin \omega_m t) \sin \omega_c t$$

$$= V_c \sin \omega_c t + m V_c (\sin \omega_m t \sin \omega_c t)$$

$$= V_c \sin \omega_c t + \frac{m V_c}{2} \cos(\omega_c - \omega_m)t - \frac{m V_c}{2} \cos(\omega_c + \omega_m)t \quad \text{--- (IV)}$$

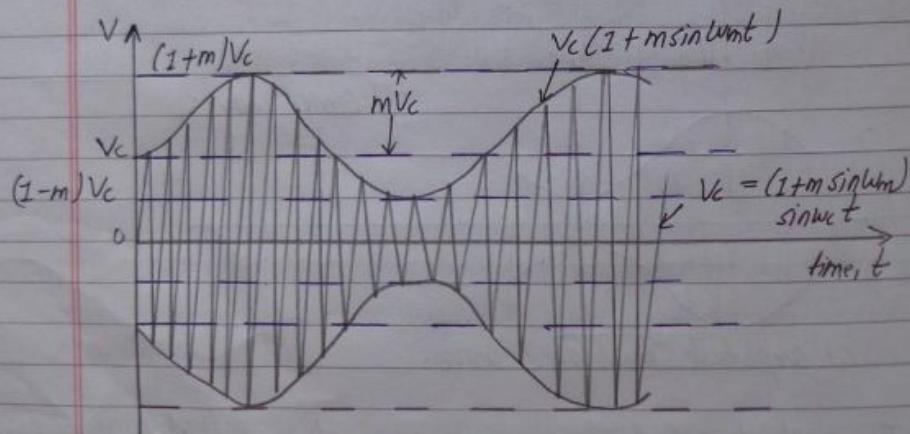


Fig: amplitude modulated sinewave with $m < 1$

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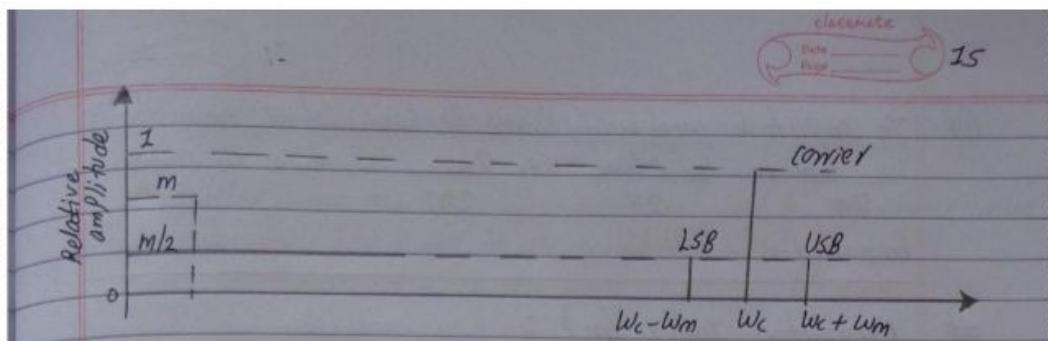


Fig: Frequency spectrum of a sinusoidally modulated wave.

The lower side frequency component $\frac{w_c - w_m}{2\pi}$ is called the lower side

frequency. The upper frequency component $\frac{w_c + w_m}{2\pi}$ is called the upper

side frequency. The amplitude of each side frequency is $\frac{mV_c}{2}$. The

amplitude of each side frequency can never exceed half the carrier amplitude because $m > 1$.

Hence, During the process of amplitude modulation;

- ① the original carrier frequency is not altered but two new frequencies $\frac{w_c + w_m}{2\pi}$ and $\frac{w_c - w_m}{2\pi}$, known as sideband frequencies are produced.

$$\frac{w_c + w_m}{2} = \text{upper sideband (USB)}$$

$$\frac{w_c - w_m}{2} = \text{lower } " \quad (\text{LSB})$$

- ② The carrier voltage component does not transmit any information because the signal frequency $w_m/2\pi$ is contained in sidebands only which is evident from above figure.

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(iii) In amplitude modulated wave, the bandwidth is from $\frac{w_c - \Delta w}{2\pi}$ to $\frac{w_c + \Delta w}{2\pi}$ i.e., $\frac{2\Delta w}{2\pi}$ or twice the signal frequency.

Modulation Index (m):

The extent to which the amplitude of the carrier wave is varied by the modulating signal is called degree of amplitude modulation or modulation index and is denoted by m or M . Hence, the ratio of change in amplitude of carrier wave to the amplitude of normal carrier wave is called the modulation index, m .

i.e.,

$$m = \frac{V_m}{V_c}$$

$\therefore m = \frac{\text{amplitude change of carrier wave}}{\text{amplitude of normal/unmodulated carrier wave}}$

$$\% m = \% M = \frac{V_m}{V_c} \times 100$$

Also,

$$\therefore m = \frac{V_{max} - V_{min}}{V_{max} + V_{min}} = M$$

Where,

V_{max} = maximum value of amplitude of modulated carrier wave.
 V_{min} = minimum " " " " " "

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D. Efficiency (power and Bandwidth):

Modulated wave has more power than that had by the carrier wave before modulation. Total power in the modulated wave is,

$$P_{\text{Total}} = P_{\text{Carrier}} + P_{\text{LSB}} + P_{\text{USB}} \quad \text{--- (I)}$$

When an amplitude modulated wave is impressed upon some resistance (say antenna resistance), R, then

$$P_{\text{Carrier}} = \frac{V_c^2 / \sqrt{2}}{R} = \frac{V_c^2}{2R} \quad \text{--- (II)}$$

Each sideband has peak value of $\frac{U}{2} V_c$ and rms value of $\frac{U \cdot V_c}{\sqrt{2}}$.

Hence, power in each sideband is,

$$P_{\text{LSB}} = P_{\text{USB}}$$

$$= \left(\frac{\frac{U \cdot V_c}{\sqrt{2}}}{2} \right)^2$$

$$= \frac{U^2 V_c^2}{8R}$$

$$= \frac{U^2}{4} \cdot \frac{V_c^2}{2R}$$

$$\therefore P_{\text{LSB}} = \frac{U^2}{4} \cdot P_{\text{Carrier}} \quad \text{--- (III)}$$

We have,

$$\begin{aligned} P_{\text{Total}} &= P_{\text{C}} + P_{\text{LSB}} + P_{\text{USB}} \\ &= \frac{V_c^2}{2R} + \frac{U^2}{4} \cdot \frac{V_c^2}{2R} + \frac{U^2}{4} \cdot \frac{V_c^2}{2R} \end{aligned}$$

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$$= \frac{V_c^2}{2R} \left(1 + \frac{m^2}{2} \right)$$
$$\therefore P_t = P_{carrier} \left(1 + \frac{m^2}{2} \right) \quad \text{--- (iv)}$$

① Maximum power in the amplitude modulated wave (without distortion) will occur for $m=1$,
i.e., $P_{total} = 1.5 P_{carrier}$ when $m=1=N$

② Ratio of P_{SB} and P_{total} is given as,

$$\frac{P_{SB}}{P_t} = \frac{\frac{m^2}{4} P_c + \frac{N^2}{4} P_c}{P_c \left[1 + \frac{N^2}{2} \right]}$$
$$\text{or } \frac{P_{SB}}{P_t} = \frac{m^2/2}{1 + \frac{m^2}{2}} \quad [\because N=m]$$

∴ For $m=1$,

$$\frac{P_{SB}}{P_t} = \frac{1/2}{3/2} = \frac{1}{3}$$

i.e., only one-third of the total power of the modulated wave is contained in the two sidebands and rest of the two-third power lies in the carrier component, which is of no-use.

③ In most applications, carrier is simultaneously modulated by several sinusoidal modulating signals. In such case, the total modulation index is,

$$M_t = \sqrt{M_1^2 + M_2^2 + M_3^2 + \dots}$$

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(iv) If I_c and I_t represents the rms value of unmodulated or carrier current and total modulated current and R is the resistance through which the current flows, then,

$$\frac{P_{total}}{P_{carrier}} = \frac{I_t^2 R}{I_c^2 R} = \left(\frac{I_t}{I_c}\right)^2$$

Also,

$$\text{or, } \frac{P_{total}}{P_{carrier}} = I + \frac{M^2}{2}$$

Hence,

$$\left[\frac{I_t}{I_c}\right]^2 = 1 + \frac{M^2}{2}$$

$$\therefore I_t = I_c \sqrt{1 + \frac{M^2}{2}}$$

Limitations of Amplitude modulation (AM):

i) Low efficiency:

In AM, useful power that lies in the sidebands is quite small, so the efficiency of AM system is low.

ii) Limited operating Range:

Transmitters employing amplitude modulation have small operating range. This is due to low efficiency. Hence, information cannot be transmitted over long distances.

iii) Noisy reception:

In case of AM, the reception is generally noisy. This is because a radio receiver cannot distinguish between the amplitude variations that represent noise and those contain the desired signal.

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N) Poor audio quality:
In order to attain high-fidelity reception, all audio frequencies upto 15 kHz must be reproduced and this necessitates the bandwidth of 30 kHz while the AM broadcasting stations are assigned bandwidth of only 10 kHz to minimize the interference from the adjacent broadcasting stations. Hence, in AM broadcasting stations audio quality is usually poor.

2.3 Numericals :

1. A) A sinusoidal carrier wave of frequency 1MHz & amplitude of 200 V is amplitude modulated by a sinusoidal voltage of frequency 5 kHz producing 50% modulation. Calculate frequency and amplitude of USB and LSB.

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 \Rightarrow Frequency of carrier, $f_c = 1 \text{ MHz}$
" " modulating signal, $f_m = 5 \text{ kHz}$
modulation index, $M = 50\% = 0.5$
lower sideband frequency $= f_c - f_m$
 $= 1000 - 5$
 $= 995 \text{ kHz}$

upper sideband frequency $= f_c + f_m$
 $= 1000 + 5 = 1005 \text{ kHz}$

\therefore Amplitude of each sideband $= M \cdot \frac{V_c}{2}$
 $= 0.5 \times \frac{200}{2}$
 $\therefore A = 25 \text{ V}$

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B) The maximum or crest amplitude of a sinusoidally modulated wave is 1.6 V, its minimum or trough amplitude is 200 mV, and its average (unmodulated) amplitude is 900 mV. The maximum amplitude of modulating signal is 700 mV. Find the modulation index by different relations.

\Rightarrow maximum or crest amplitude, $V_{max} = 1.6 \text{ V}$

minimum or trough " $V_{min} = 200 \text{ mV} = 0.2 \text{ V}$

Average (unmodulated) amplitude,

$$V_c = \frac{V_{max} + V_{min}}{2} = \frac{1.6 \text{ V} + 0.2 \text{ V}}{2} = 0.9 \text{ V}$$

max. amplitude of modulating signal,

$$V_m = 700 \text{ mV} = 0.7 \text{ V}$$

i) modulation index, $M = \frac{V_m}{V_c} = \frac{0.7}{0.9} = 0.777$

ii) $M = \frac{V_{max} - V_{min}}{V_{max} + V_{min}} = \frac{1.6 - 0.2}{1.6 + 0.2} = 0.777$

iii) $M = \frac{V_{max} - V_{min}}{2V_c} = \frac{1.6 - 0.2}{2 \times 0.9} = 0.777$

iv) $M = \frac{V_{max} - V_c}{V_c} = \frac{1.6 - 0.9}{0.9} = 0.777$

v) $M = \frac{V_c - V_{min}}{V_c} = \frac{0.9 - 0.2}{0.9} = 0.777$

The result is same as expected.

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C. The rms antenna current of a radio transmitter is 10 A when unmodulated rising to 12 A, when the carrier is sinusoidally modulated. calculate the modulation index.

soln,

\Rightarrow carrier current, $I_c = 10 \text{ A}$
total modulated current, $I_t = 12 \text{ A}$

modulation index, $M = \sqrt{\left(\frac{I_t}{I_c}\right)^2 - 1} \times 2$

$$= \sqrt{\left(\frac{12}{10}\right)^2 - 1} \times 2$$

since, $\frac{I_t}{I_c} = \sqrt{1 + \frac{m^2}{2}}$

$$\therefore M = 0.938$$

D. Rms value of RF voltage after amplitude modulation to a depth of 60% by a sinusoidal voltage is 60 V. calculate the rms value of modulated voltage when modulated to depth of 75%.

soln,

\Rightarrow For 60% modulation, modulation voltage,
 $V_m(\text{rms}) = 60 \text{ V}$

Also,

$$\frac{P_{\text{total}}}{P_{\text{carrier}}} = \frac{V_t^2/R}{V_c^2/R} = \left(\frac{V_t}{V_c}\right)^2$$

so, we have,

$$\left(\frac{V_t}{V_c}\right)^2 = 1 + \frac{m^2}{2}$$

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A) $V_c = \frac{V_i}{\sqrt{1 + M^2}}$

$= \frac{60}{\sqrt{1 + (0.6)^2}}$

$\therefore V_c = 55.23 \text{ V}$

Rms value of modulated voltage with 75% modulation,

$= V_c \sqrt{1 + \frac{M^2}{2}}$

$= 55.23 \times \sqrt{1 + \frac{(0.75)^2}{2}}$

$\therefore = 62.5 \text{ V}$

E. In an AM system, the modulating signal is sinusoidal with frequency of f_m Hz. If 80% modulation is used, determine ratio of the total sideband power in the modulated signal.

so/11,

$\Rightarrow M = 80\% = 0.8$

sideband power, P_{SB}

total power in modulated wave, P_{Total}

$= \frac{m^2/2}{1 + \frac{m^2}{2}}$

$= \frac{0.8^2/2}{1 + (0.8^2/2)}$

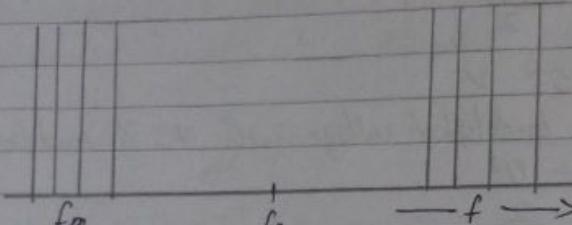
$\therefore \frac{P_{SB}}{P_{Total}} = 0.242$

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2.4) DSB-SC modulation:

In DSB-SC systems, as its name indicates, carrier component is altogether removed resulting in saving of enormous amount of power.



It is known that carrier signal contains two-third of total transmitted power for 100% modulation.

A) Generation of DSBSC : 2015/5, 2017/5, 18/1F

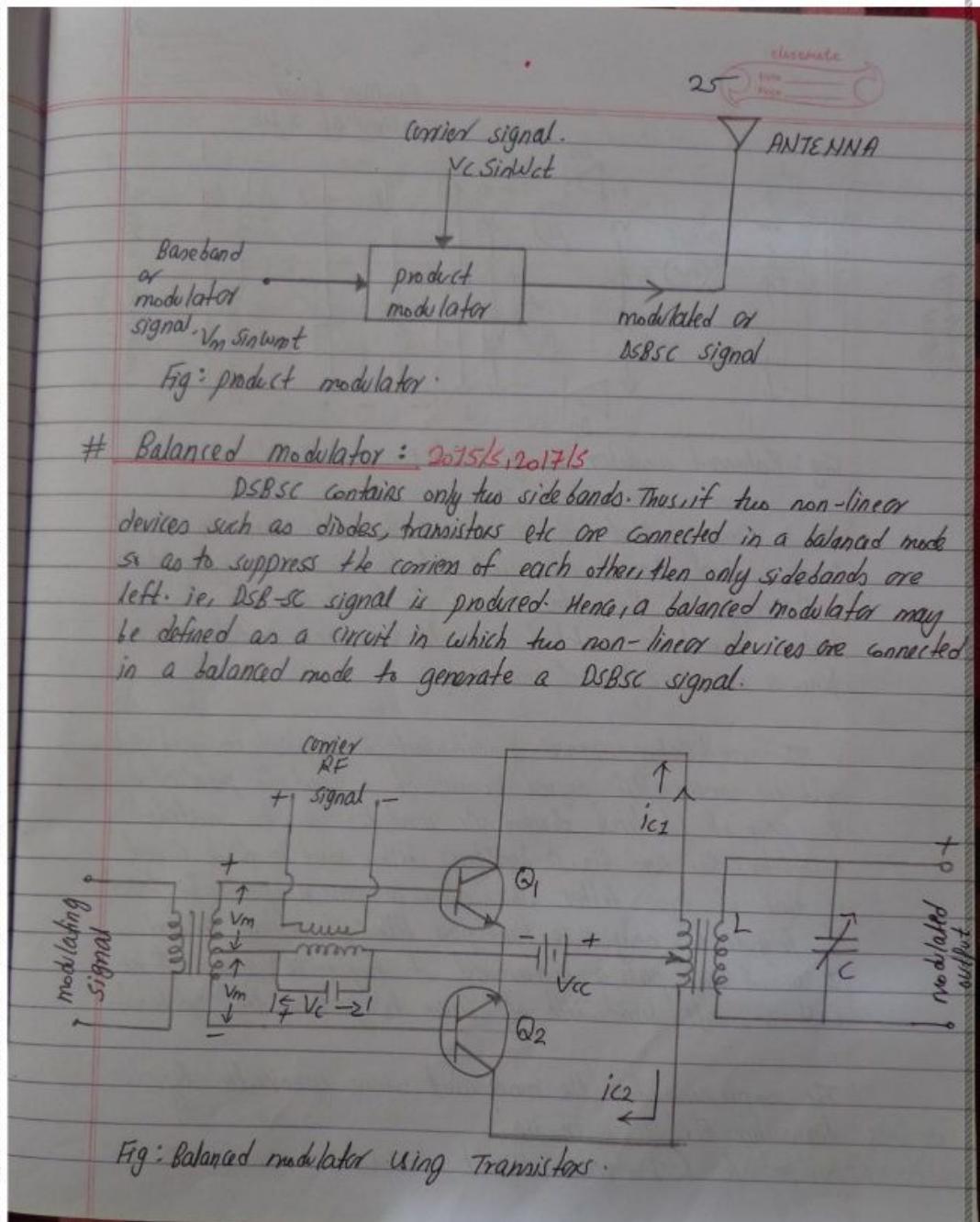
DSBSC can be obtained by simply multiplying the modulating signal with the carrier signal. By simple multiplication of $V_c \sin \omega_c t$ and $V_m \sin \omega_m t$, we have the lower and upper sidebands without carrier is,

$$\begin{aligned} V_c \times V_m &= V_c \sin \omega_c t \times V_m \sin \omega_m t \\ &= \frac{V_c V_m}{2} \times 2 \sin \omega_c t \times \sin \omega_m t \\ &= \frac{V_c V_m}{2} [\cos(\omega_c - \omega_m)t - \cos(\omega_c + \omega_m)t] - \textcircled{1} \end{aligned}$$

It means that a DSBSC signal is obtained by simply multiplying modulating signal ($V_m \sin \omega_m t$) with carrier signal ($V_c \sin \omega_c t$). This is achieved by a product modulator. There are two types of product modulator ; Balanced and Ring modulator.

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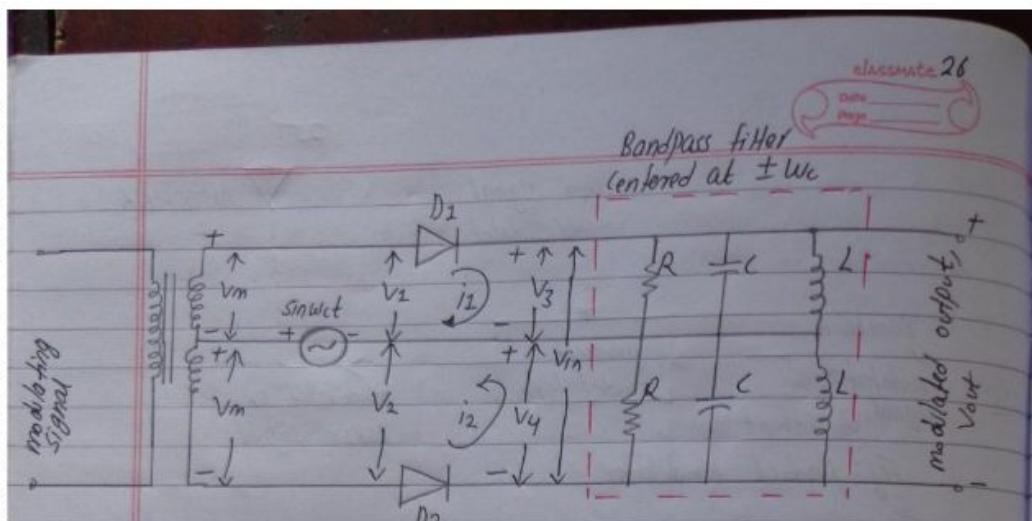


Fig: Balanced modulator using Diode.

It is assumed that both diodes and transistors used in circuits have identical characteristics and that the circuit is symmetrical with respect to the centre tap of the transformer. As the two halves have to be matched or balanced, such modulator is known as a balanced modulator.

The modulating signal $V_m \sin \omega_m t$ is applied in push-pull and the carrier RF signal $V_c \sin \omega_c t$ is put in parallel to a pair of identical diodes D₁ and D₂, or two matched transistors Q₁ and Q₂. A bandpass filter used in diode circuit is that of type filter which allows to pass a particular band of frequencies only. Here, band pass filter is centred around $\pm \omega_c$, it will pass a narrowband of frequencies centred $\pm \omega_c$ with a small bandwidth of $2\omega_m$ to preserve the sidebands.

The expression for the modulated wave generated by transistor Q₁ is given as,
$$V_1 = V_c (1 + m \sin \omega_m t) \sin \omega_c t$$

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$= V_c \sin \omega_c t + m V_c \sin \omega_m t \cdot \sin \omega_c t$

The modulated wave generated by transistor Q2 will be same as generated by Q1 except that it will have a phase difference of 180° (ie, reversed phase), so,

$V_2 = V_c \sin \omega_c t - m V_c \sin \omega_m t \sin \omega_c t$

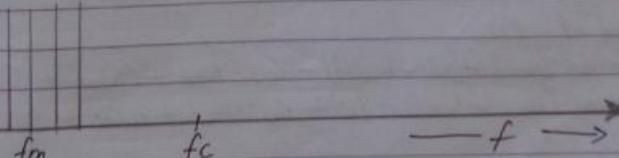
At the output transformer, output voltage, V_{out} , due to push-pull arrangement, is proportional to $(V_1 - V_2)$, so,

modulated output, $V_{out} = V_1 - V_2$
 $= 2 m V_c \sin \omega_m t \cdot \sin \omega_c t$
 $= m V_c [\cos(\omega_c - \omega_m)t - \cos(\omega_c + \omega_m)t]$

Thus, it is proved that the output signal consists of only sidebands and the carrier component is eliminated.

2.5 SSB Modulation: 2016/5

It consists in transmitting only one sideband and suppresses the other sideband and the carrier. It utilizes the fact that the intelligence or messages is contained in each sideband and not in the carrier. Two sidebands being the exact images of each other, carry the same audio intelligence.

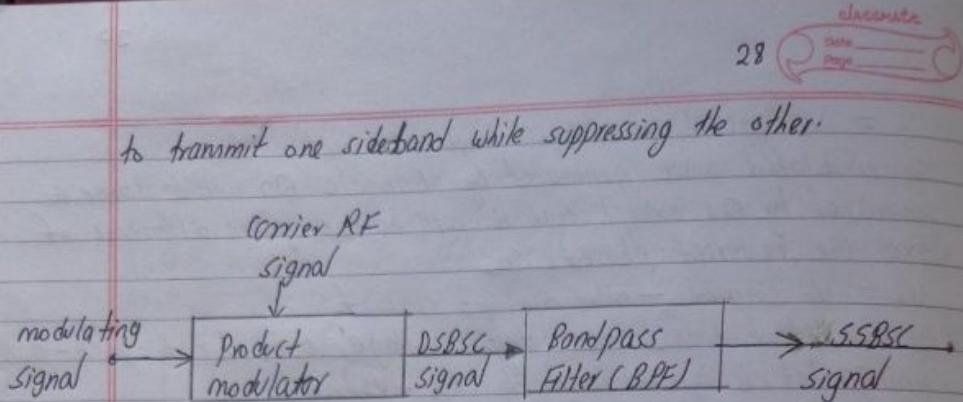


Generation:

SSB can be obtained by passing the output of a carrier suppressor (product modulator) through filter circuits that are selective.

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mathematical analysis :

Let the expression for modulating signal & carrier signal be sinusoid and sinewave respectively,

Now, the balanced modulator M_1 will receive $\sin(\omega_m t)$ and $\sin(\omega_c t + 90^\circ)$ and modulator M_2 will receive $\sin(\omega_m t + 90^\circ)$ and $\sin \omega_c t$.

We know, output of Balanced modulator M₁ will contain sum and difference frequencies.
Hence,

Similarly,

output of M_2 ,

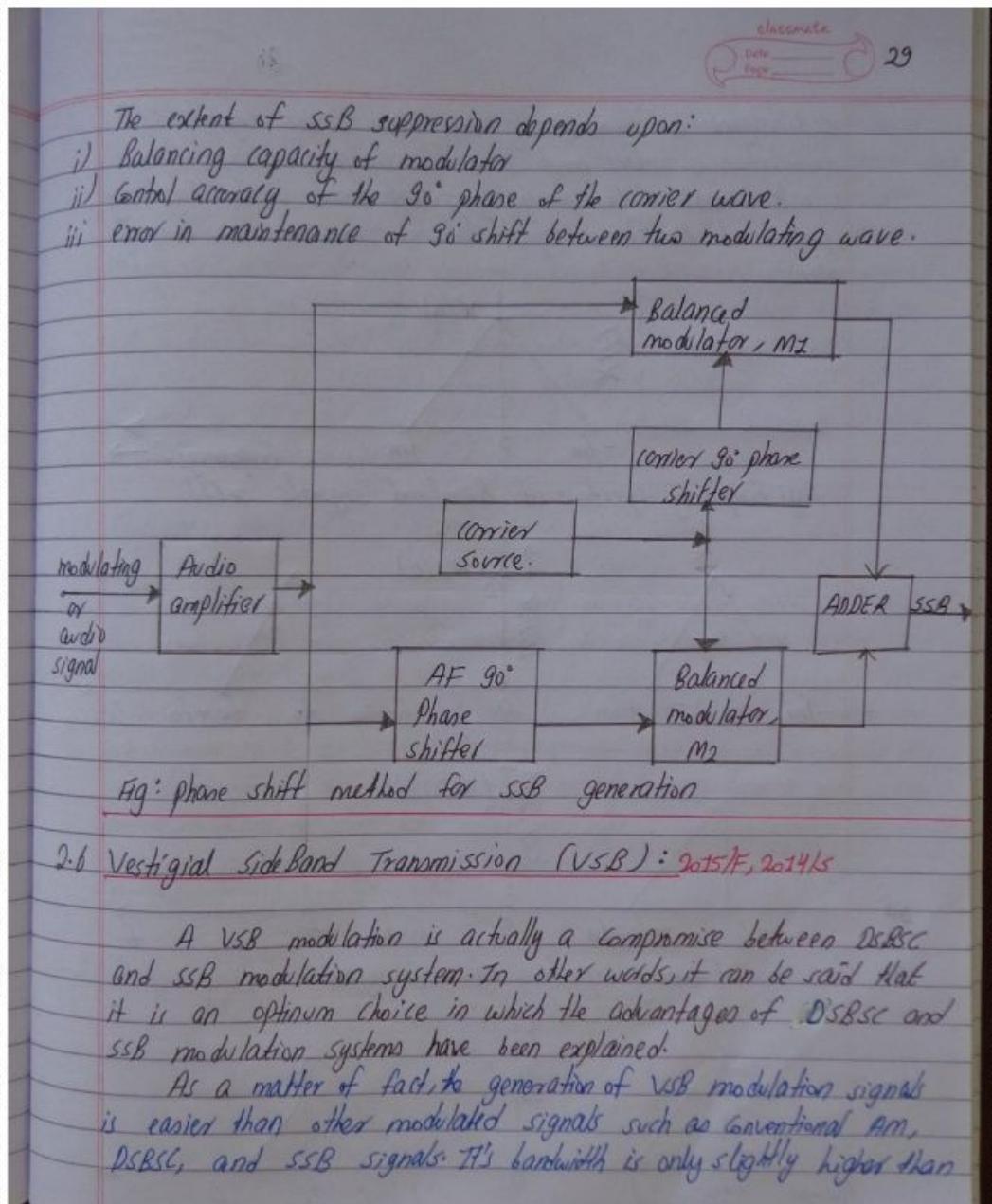
$$V_2 = \cos[\omega_c t + (g_0^\circ - \omega_m t)] - \cos[\omega_c t + (\omega_m t + g_0^\circ)] \\ = \cos(\omega_c t - \omega_m t - g_0^\circ) - \cos(\omega_c t + \omega_m t + g_0^\circ)$$

Thus,

$$\begin{aligned} \therefore V_{out} &= V_1 + V_2 \\ &= 2 \cos(\omega t - \omega_{mt} + \phi_0) \quad \text{--- LSB} \\ &= |V_1 - V_2| \\ &= 2 \cos(\omega t + \omega_{mt} + \phi_0) \quad \text{--- USB} \end{aligned}$$

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SSB signals are considerably less than DSBSC signals. SSB modulation is rather more suited for the transmission of voice signals because of the energy gap that exists in the frequency spectrum of the voice signals between zero and few hundred Hertz.

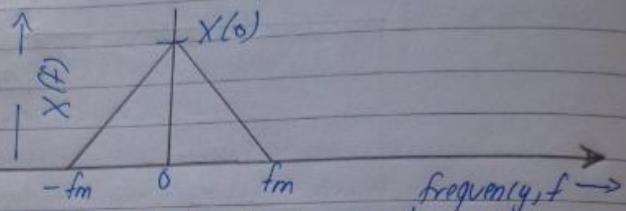


Fig: frequency spectrum of baseband signals $x(t)$

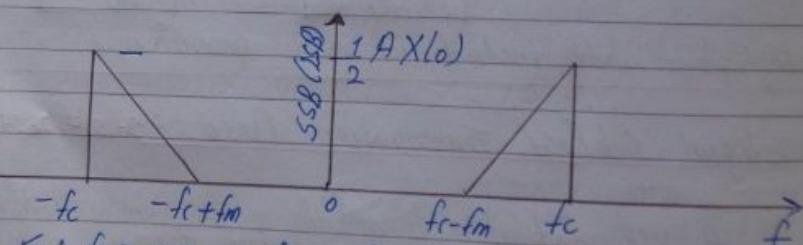
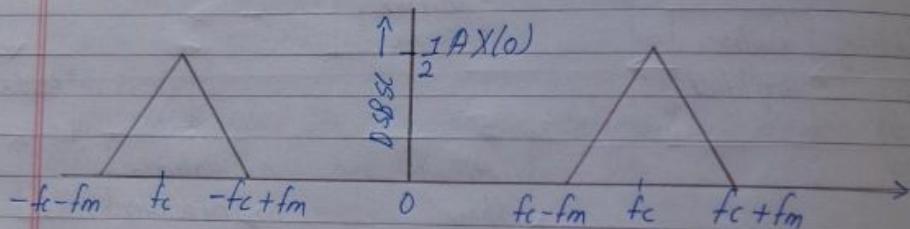
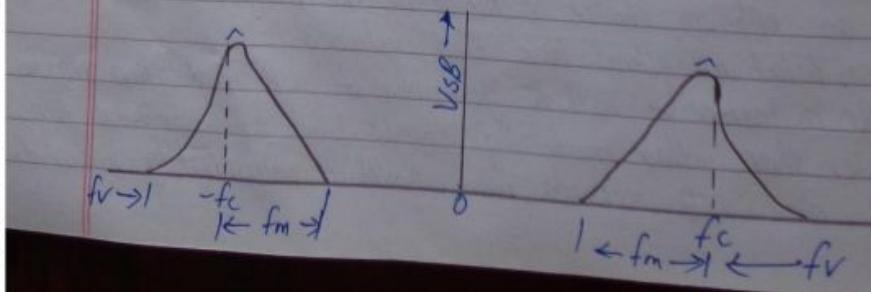


Fig: frequency spectrum of SSB signal.



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When signals contain frequency components at extremely low frequencies, (as in telegraph frequencies) the USB and LSB of the translated signal tend to meet at the carrier frequencies. Under such circumstances, it becomes very difficult to isolate the unavoidable one sideband from the other. Hence, SSB scheme becomes unsuitable for handling such type of signals.

This difficulty has been overcome in a scheme known as VSB modulation. In VSB modulation, instead of rejecting one sideband completely as in SSB modulation scheme, a gradual cutoff of one sideband is allowed. This gradual cut is compensated by a vestige or portion of the other sideband.

27) Frequency Modulation (F.M.):

It is virtually one of the most widely used modes of modulation. From applications in commercial broadcasting, FM radio, TV audio, cordless phone to cellular and mobile communication, FM is indeed both a reliable and important form of modulation. In AM, interference such as static, lightning and manmade noise, cause the amplitude of an RF signal to vary widely. This is because such noises, are predominately amplitude modulated signals in composition. The noise is added and superimposed on the transmitted AM signal carrying the desired intelligence.

27.A Analysis of FM carrier wave:

Let, the carrier and modulating voltage wave be represented as,

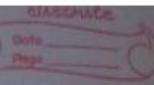
$$V_c = V_c \cos (\omega_c t + \phi) \quad \text{--- (1)}$$
$$V_m = V_m \cos \omega_m t \quad \text{--- (2)}$$

where, V_c , V_m , ω_c and ϕ are the instantaneous value, peak value,

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Angular velocity and the initial phase angle of the carrier and v_m, V_m and ω_m are the instantaneous value, peak value and the angular velocity of the modulating signal.

let,

$$\phi_c(t) = \omega_c t + \phi_0 \quad \text{--- (1)}$$

$\phi_c(t)$ = total instantaneous phase angle of the carrier wave at time t .

$$v_c = V_c \cos \phi_c(t) \quad \text{--- (2)}$$

The instantaneous angular velocity ω_c , defined as the instantaneous rate of increase of instantaneous phase, is related to phase angle, ϕ_c as,

$$\omega_c = \frac{d\phi_c}{dt} \quad \text{--- (3)}$$

In FM, the frequency of the carrier wave varies with time in accordance with the instantaneous value of modulating voltage. Hence, the frequency of the carrier and frequency modulation is given as,

$$\omega = \omega_c + k_f V_m \\ = \omega_c + k_f V_m \cos \omega_m t \quad \text{--- (4)}$$

where,

k_f = constant proportionality ie, frequency conversion factor.

The instantaneous phase of FM wave is obtained by integrating eqⁿ (4),

$$\phi(t) = \int \omega \cdot dt = \int (\omega_c + k_f V_m \cos \omega_m t) dt \\ = \omega_c t + k_f V_m \cdot \frac{1}{\omega_m} \sin \omega_m t + \phi_0$$

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ϕ_i = initial phase.

The initial phase ϕ_i can be neglected since it is insignificant in the modulation process, so,

$$\phi(t) = \omega_c t + k_f \cdot V_m \sin \omega_m t \quad \text{--- (7)}$$

The eqn of FM wave is,

$$V_{fm} = V_c \sin \phi(t)$$

$$= V_c \sin \left[\omega_c t + k_f \cdot \frac{V_m}{\omega_m} \sin \omega_m t \right] \quad \text{--- (8)}$$

The instantaneous frequency of FM wave,

$$f = \frac{\omega}{2\pi} = \frac{\omega_c}{2\pi} + k_f \frac{V_m}{2\pi} \cos \omega_m t \quad \text{--- (9)}$$

$$= f_c + k_f \frac{V_m}{2\pi} \cos \omega_m t \quad \text{--- (10)}$$

From eqn (10),

maximum & minimum frequency are,

$$\therefore f_{max} = f_c + k_f \cdot \frac{V_m}{2\pi} \quad \because \cos 0^\circ = +1$$

$$\therefore f_{min} = f_c - k_f \cdot \frac{V_m}{2\pi} \quad \because \cos 180^\circ = -1$$

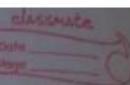
Thus, the frequency deviation, defined as the maximum change in frequency from mean value, f_c is given as,

$$f_d = f_{max} - f_c = f_c - f_{min} = k_f \cdot \frac{V_m}{2\pi} \quad \text{--- (11)}$$

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The total variation in frequency from the minimum to maximum value, i.e., $f_{max} - f_{min}$ is called carrier swing and it is given as,

$$\text{carrier swing (Hz)} = 2 \cdot f_d \\ = k_f \cdot V_m \\ 2\pi$$

The frequency modulation index, mf or lf defined as the ratio of frequency deviation to modulation frequency,

$$mf = lf = \frac{w_d}{w_m} = \frac{k_f \cdot k_m}{w_m} = S$$

Hence, From eqn (8),

$$\therefore V_m = V_c \sin [w_c t + mf \sin w_m t] \\ = V_c [\sin w_c t \cos (mf \sin w_m t) + \cos w_c t \sin (mf \sin w_m t)]$$

unlike amplitude modulation, the frequency modulation index can exceed unity.

2.7. B Narrowband & wideband FM : 2016/5, 2.

I) Narrowband FM:

- modulation index is less than 1.
- maximum modulation frequency is usually 3 kHz & maximum frequency deviation is 75 kHz
- A narrowband FM wave consists of a carrier, an upper side-frequency component & lower side component.
- The modulated narrowband signal differs from the ideal response in two fundamental aspects;

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+ The envelope consists a residual AM, so it varies with time.
+ For sinusoidal modulating wave, the angle $\theta(t)$ contains harmonic distortion in the form of 3rd and higher order harmonics of modulation frequency, f_m .

By restricting $\beta \leq 0.3$ radians, the effects of residual AM & harmonic PM are limited to negligible levels. A narrow band signal may be represented by phasor diagram as,

The diagram shows a vector triangle representing the sum of sides frequency phasor. One side of the triangle is labeled 'Resultant' and 'carrier'. The other two sides are labeled 'lower frequency side' and 'upper side frequency', both pointing away from the resultant vector. The angle between the carrier and each side is labeled f_m .

Fig: phasor diagram of narrowband FM

Uses:
mobile communications services such as police wireless, ambulance, taxicabs, short range very high frequency (VHF) ship to shore sources and defence.

II) Wideband / Broadband FM:

- modulation index normally exceed unity.
- modulation frequencies extend from 30 Hz to 15 kHz
- maximum permissible deviation = 75 kHz
- wideband FM system needs large bandwidth typically 25 times that

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- of narrowband FM system.
- used in entertainment broadcasting.
- For large values of β , compared to 1, radian, the FM wave contains a carrier and infinite number of side-frequency components located symmetrically around the carrier.
- The amplitude of the carrier component contained in a wide band FM wave varies with the modulation index, β in accordance with Bessel function $J_0(\beta)$.

C. Transmission Bandwidth of FM signal / *Carson's Rule: 2015/2017/F*

In practice, FM wave is limited to finite number of significant side-frequencies compatible with a specified amount of distortion. Thus, an effective bandwidth is required for the transmission of an FM wave.

Approximate rule for transmission bandwidth of an FM wave generated by a single-tone modulating wave of frequency f_m is,

$$\begin{aligned} B_T &\approx 2Af + 2f_m \\ &= 2Af \left(1 + \frac{1}{B} \right) \end{aligned}$$

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This relation is known as Carson's Rule.

According to international regulations of FM broadcast, maximum frequency deviation, $\Delta f = \pm 75 \text{ kHz}$ allowable bandwidth per channel, $= 200 \text{ kHz}$

In telecommunications, Carson's bandwidth rule defines the approximate bandwidth requirement of communication system components for a carrier signal that is frequency modulated by a continuous or broad spectrum of frequencies rather than a single frequency. Carson's rule doesn't apply well when the modulating signal contains discontinuities, such as square wave. Carson's bandwidth rule is,

$$CBR = 2(\Delta f + f_m)$$

where,

Δf = peak frequency deviation.
 f_m = highest frequency in the modulating signal.

Conditions for application of Carson's rule is only sinusoidal signals.

$$\therefore BT = 2(\Delta f + w) \\ = 2w(D+1)$$

Where,
 w = highest frequency in the modulating signal but non-sinusoidal in nature.

D = deviation ratio of frequency deviation to highest frequency of modulating non-sinusoidal signal.

Carson's Bandwidth rule is often applied to transmitters, antennas, optical resonators, receivers / photodetectors. etc.

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D-I) Consider a carrier waveform $v_c = 10 \cos \omega_c t$ and a modulating message signal $v_m = 3 \cos \omega_m t$ with $f_c = 100 \text{ kHz}$ and $f_m = 4 \text{ kHz}$. Find the modulation index and the channel bandwidth for amplitude and frequency modulation. Assume the sensitivity of the frequency modulator to be 5 kHz per volt.

Sol:

$$\Rightarrow \text{carrier waveform, } v_c = 10 \cos \omega_c t$$

$$\text{modulating message signal, } v_m = 3 \cos \omega_m t$$

$$\text{carrier frequency, } f_c = 100 \text{ kHz}$$

$$\text{modulating "}, f_m = 4 \text{ kHz}$$

① Am signal,

$$U = \frac{v_m}{v_c}$$

$$\therefore U = \frac{3}{10} = 0.3$$

$$\text{channel Bandwidth, } B_W = 2 \times f_m = 2 \times 4 = 8 \text{ kHz.}$$

② FM signal, $m_f = k_f \cdot \frac{v_m}{f_m}$

$$= \frac{5 \times 3}{4}$$

$$= 3.75$$

$$\text{Bandwidth, } B_W = 2\Delta f + 2f_m$$

$$= 2m_f f_m + 2f_m$$

$$= 2 \times 3.75 \times 4 + 2 \times 4$$

$$\therefore B_W = 38 \text{ kHz}$$

II. An FM transmission has a frequency deviation of 18.75 kHz. Calculate percent modulation if it is broadcast ① in 88-108

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MHZ band ① as a portion of a TV broadcast.
So,
 \Rightarrow Frequency deviation, $\Delta f = 18.75 \text{ kHz}$
A maximum frequency deviation of 75 kHz is allowed for commercial broadcast.
i.e., $(\Delta f)_{\max} = 75 \text{ kHz}$
% modulation, $m = \frac{\Delta f}{(\Delta f)_{\max}} \times 100$
 $= \frac{18.75}{75} \times 100$
 $= 25 \%$
A maximum frequency deviation of 25 kHz is allowed for sound portion of the TV broadcast.
i.e., $(\Delta f)_{\max} = 25 \text{ kHz}$
So,
% modulation, $m = \frac{25}{25} \times 100$
 $= 100 \%$

E. Signal to Noise Ratio (SNR):
It is defined as the ratio of signal power to the noise power, often expressed in decibels. A ratio higher than 1:1 (greater than 0 dB) indicates more signal than noise.

$$\text{SNR} = \frac{P_{\text{signal}}}{P_{\text{noise}}}$$

Where, P = average power.
Both signal and noise power must be measured at the same or equivalent points in a system and within same system bandwidth.

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If the variance of the signal and noise are known, and the signal and noise are both zero-mean, SNR can be,

$$SNR = \frac{\sigma^2_{\text{signal}}}{\sigma^2_{\text{noise}}}$$

If the signal and noise are measured across the same impedance, the SNR can be obtained by calculating the square of the amplitude ratio

$$SNR = \frac{P_{\text{signal}}}{P_{\text{noise}}} = \left(\frac{A_{\text{signal}}}{A_{\text{noise}}} \right)^2$$

where,
 $A = \text{root mean square(rms)}$

SNR in decibels:

$$P_{\text{signal}} = 10 \log_{10} (P_{\text{signal}})$$

and,

$$P_{\text{noise}} = 10 \log_{10} (P_{\text{noise}})$$

So,

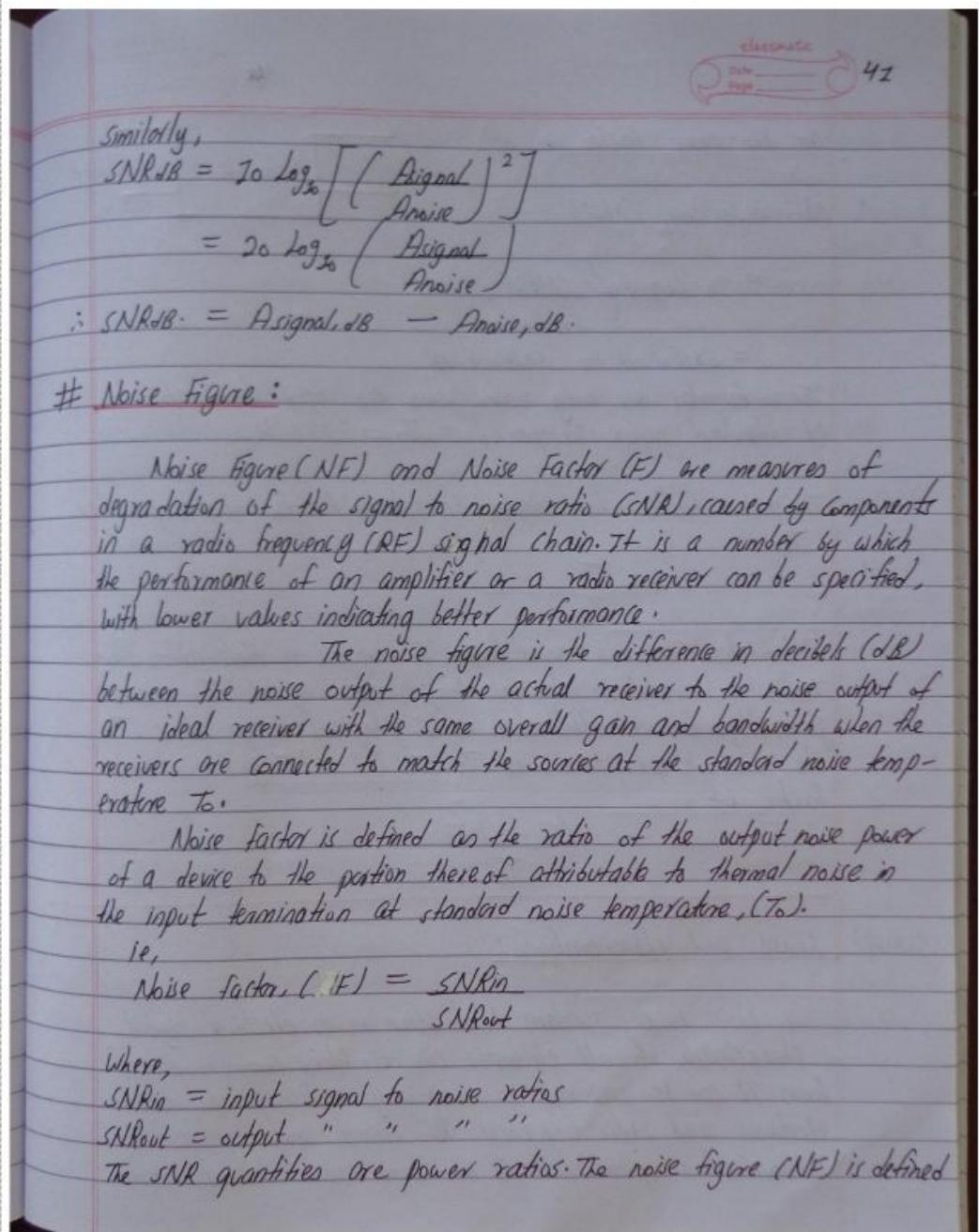
$$SNR_{\text{dB}} = 10 \log_{10} (SNR)$$
$$= 10 \log_{10} \left(\frac{P_{\text{signal}}}{P_{\text{noise}}} \right)$$

Using quotient rule of logarithms,

$$\therefore 10 \log_{10} \left(\frac{P_{\text{signal}}}{P_{\text{noise}}} \right) = 10 \log_{10} (P_{\text{signal}}) - 10 \log_{10} (P_{\text{noise}})$$
$$\therefore SNR_{\text{dB}} = P_{\text{signal, dB}} - P_{\text{noise, dB}}$$

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as the noise factor in dB,
 $\therefore NF = 10 \log_{10} (NF)$
 $= 10 \log_{10} \left(\frac{SNR_{in}}{SNR_{out}} \right)$
 $= SNR_{in, dB} - SNR_{out, dB}$
These formulas are only valid when the input termination is at standard noise Temperature, $T_0 = 290$ K.

The noise factor of a device is related to its noise temperature T_e ;
 $F = 1 + \frac{T_e}{T_0}$

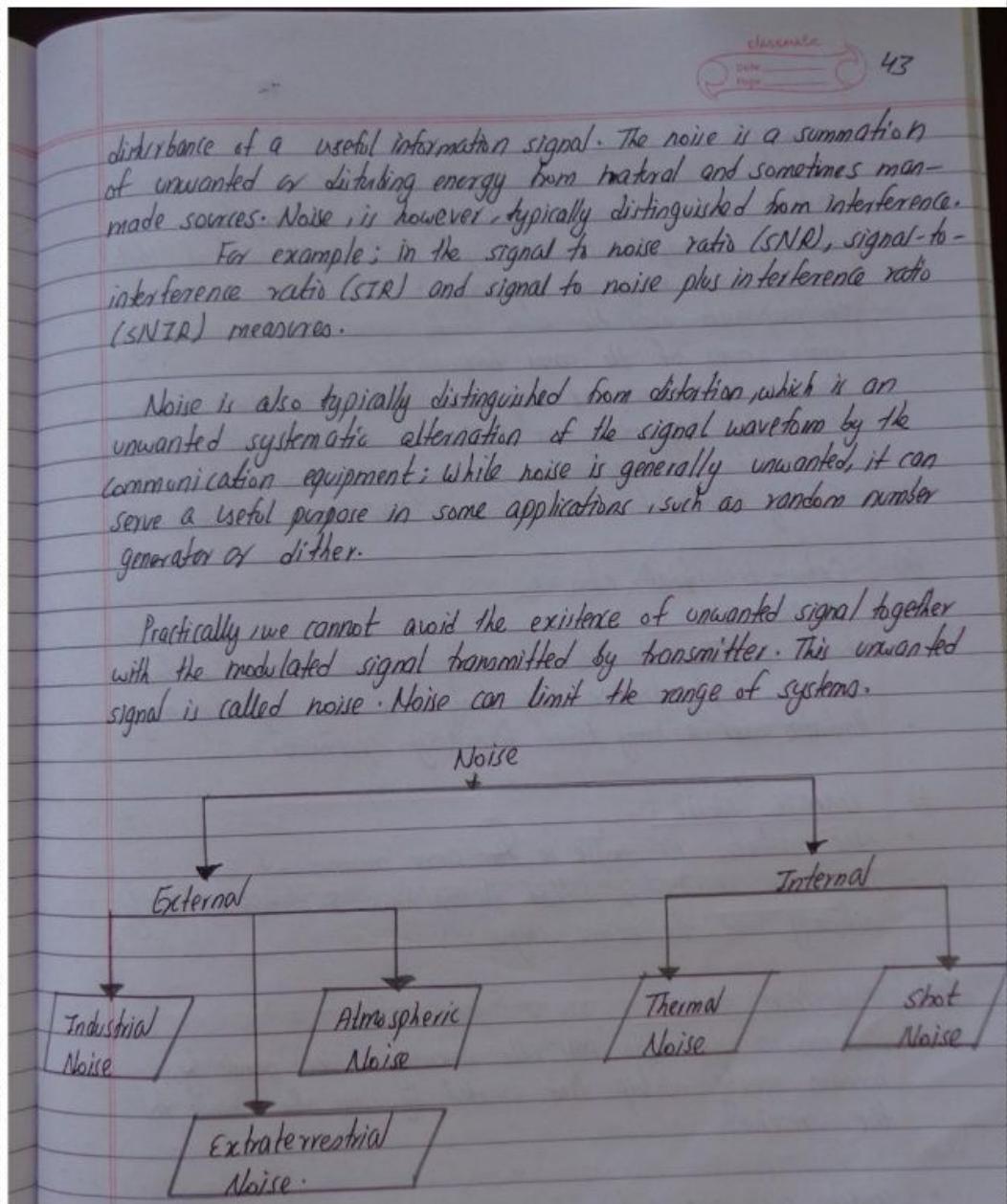
Attenuators have a noise factor F equals to their attenuation ratio L when their physical temperature equals T_0 . more generally, for an attenuator at a physical temperature, T , the noise temperature, is $T_e = (L-1)T$ giving a noise factor of
 $\therefore F = 1 + \frac{(L-1)T}{T_0}$

2.8.A) Noise and classifications:

Noise is a random fluctuation in an electrical signal, a characteristic of all electronic circuits. Noise comes in many forms. It can be generated in many ways and noise can affect electronic and radio frequency, RF circuits and systems. In communication system, noise is an error or undesired random

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The image shows a page from a handwritten notebook. At the top right, there is a red stamp that says "classmate" and "Date _____ Page _____". Below the stamp, the page number "44" is written. The handwriting is in black ink on white paper. The notes are organized into three main sections, each with a circled Roman numeral:

- I** Atmospheric Noise:
 - caused by lightning discharges, in thunderstorms and other natural electric disturbances occurring in the atmosphere.
 - consists of spurious radio signal with components distributed over a wide range of frequencies.
 - It propagate over the earth in the same way as ordinary radio waves of the same frequencies.
 - Became less severe at frequencies above 30 MHZ because,
 - Higher frequencies are limited to line of sight propagation.
 - Nature of mechanism generating this noise is such that very little of it is created in the VHF range & above.
- II** Extra-terrestrial Noise:
 - # SOLAR Noise:
 - Normal condition, there is a constant noise radiation from the sun, simply because large body at a very high frequency.
 - Radiates over a very broad frequency spectrum.
 - # COSMIC Noise:
 - Star radiates RF noise in the same manner of Sun.
 - The noise received is called thermal noise & distributed fairly uniformly over the entire sky.
- III** Industrial Noise:
 - Between 2 to 600 MHZ the intensity noise made by humans easily outstrips that created by any other source to the receiver.
 - Sources such as: automobile, aircraft, electric motors, and

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other heavy machines.

Nature of industrial noise is so variable that it is difficult to analyze.

IV. Shot Noise :

- caused by the random arrival of carriers at the output element of an electronic device.
- sometimes called transistor noise.
- Randomly varying and superimposed onto any signal present.
- The circuit current carriers are not moving continuously steady flow.
- First observed in the anode current of a vacuum-tube amplifier.
- Shot noise also occurs in semiconductors due to liberation of charges carrier.
- For pn junction, the mean square shot noise current is

$$I_h^2 = 2 (I_{dc} + 2 I_o) g_s B \text{ (amp)}^2$$

V. Thermal Noise :

- Is associated with the rapid and random movement of electrons within a conductor due to thermal agitation.
- This type of noise is generated by all resistances (eg: resistor, semiconductor, resistance of a resonant ckt. i.e., real part of impedance, cable etc.)
- Experimentally result by (Johnson) and theoretical studies (by Nyquist) give the mean square noise voltage as,

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$$v_n^2 = 4RKB$$

where,

$$K = \text{Boltzmann's constant} = 1.38 \times 10^{-23} \text{ J/K}$$

T = absolute temperature

B = bandwidth noise

R = resistance

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- Referred as white noise.
- Is a form of additive noise, cannot be eliminated. It increases in intensity with the number of devices in a circuit.
- Thermal noise power is proportional to the product of bandwidth and temperature, mathematically,
Noise power, $N = KTB$
- Noise power can be modeled using voltage equivalent circuit (Thevenin equiv. ckt) or Norton equiv ckt.

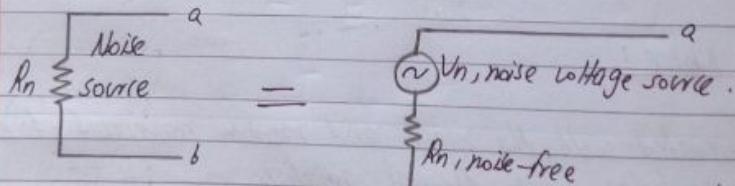


Fig: Noise source ckt

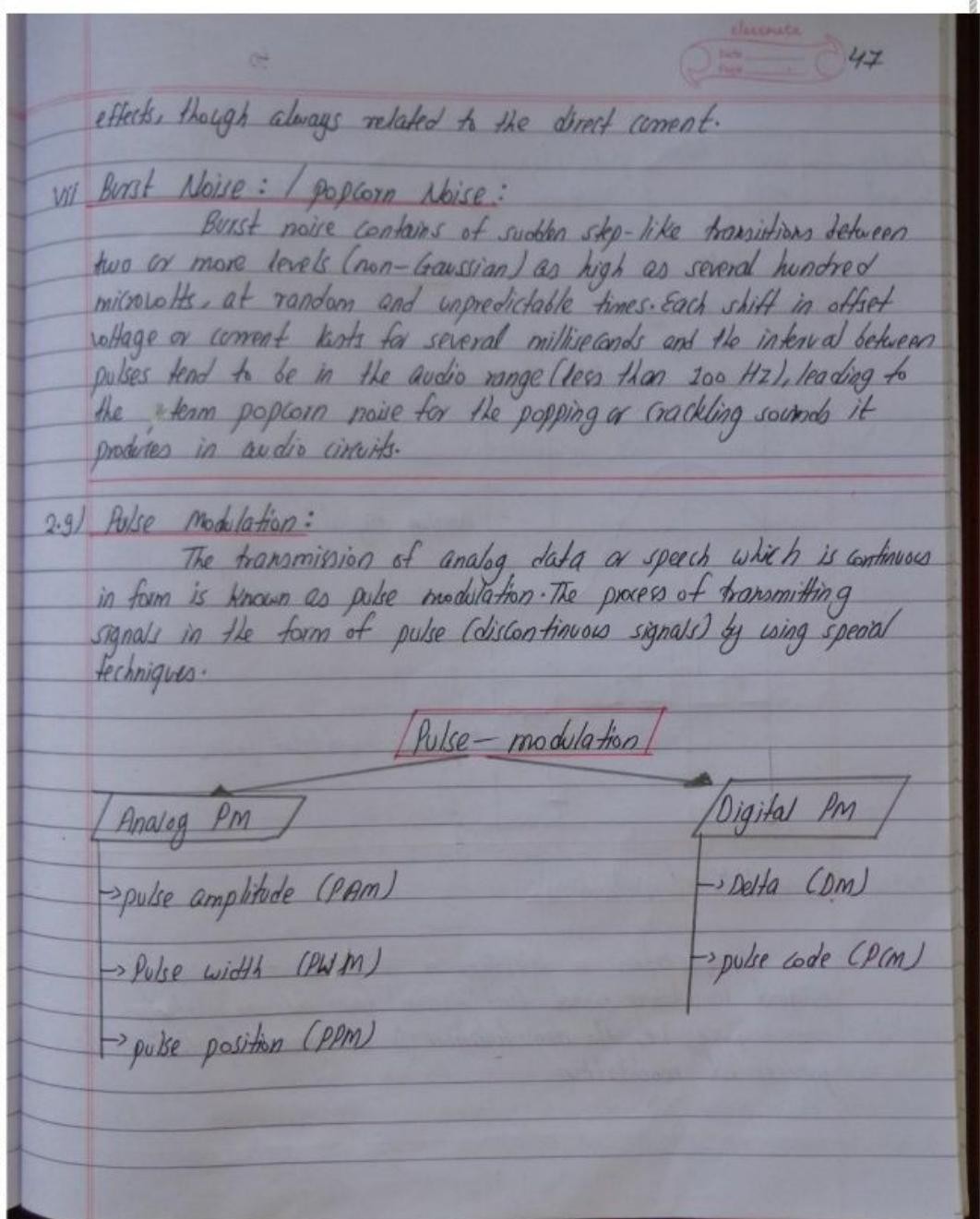
Fig: Thevenin equiv ckt.

VI. Flicker Noise:

Flicker noise, also known as 1/f noise, is a signal or process with a frequency spectrum that falls off steadily into the higher frequencies with a pink spectrum. It occurs in almost all electronic devices, and results from a variety of

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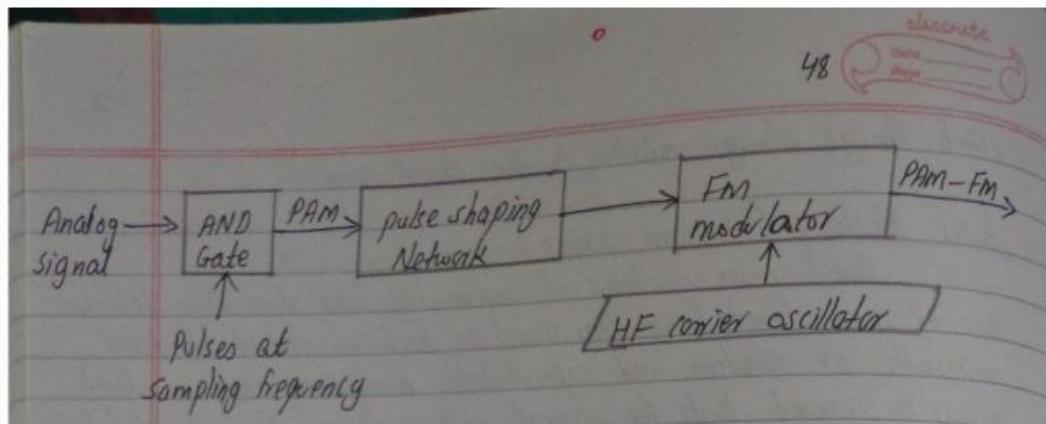
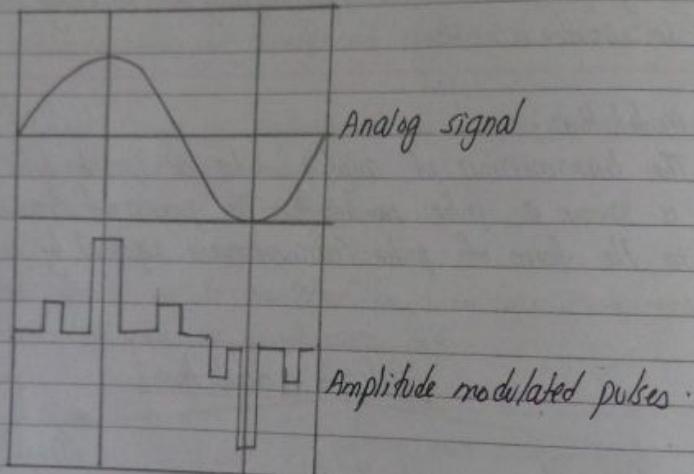


Fig: pulse amplitude modulator.

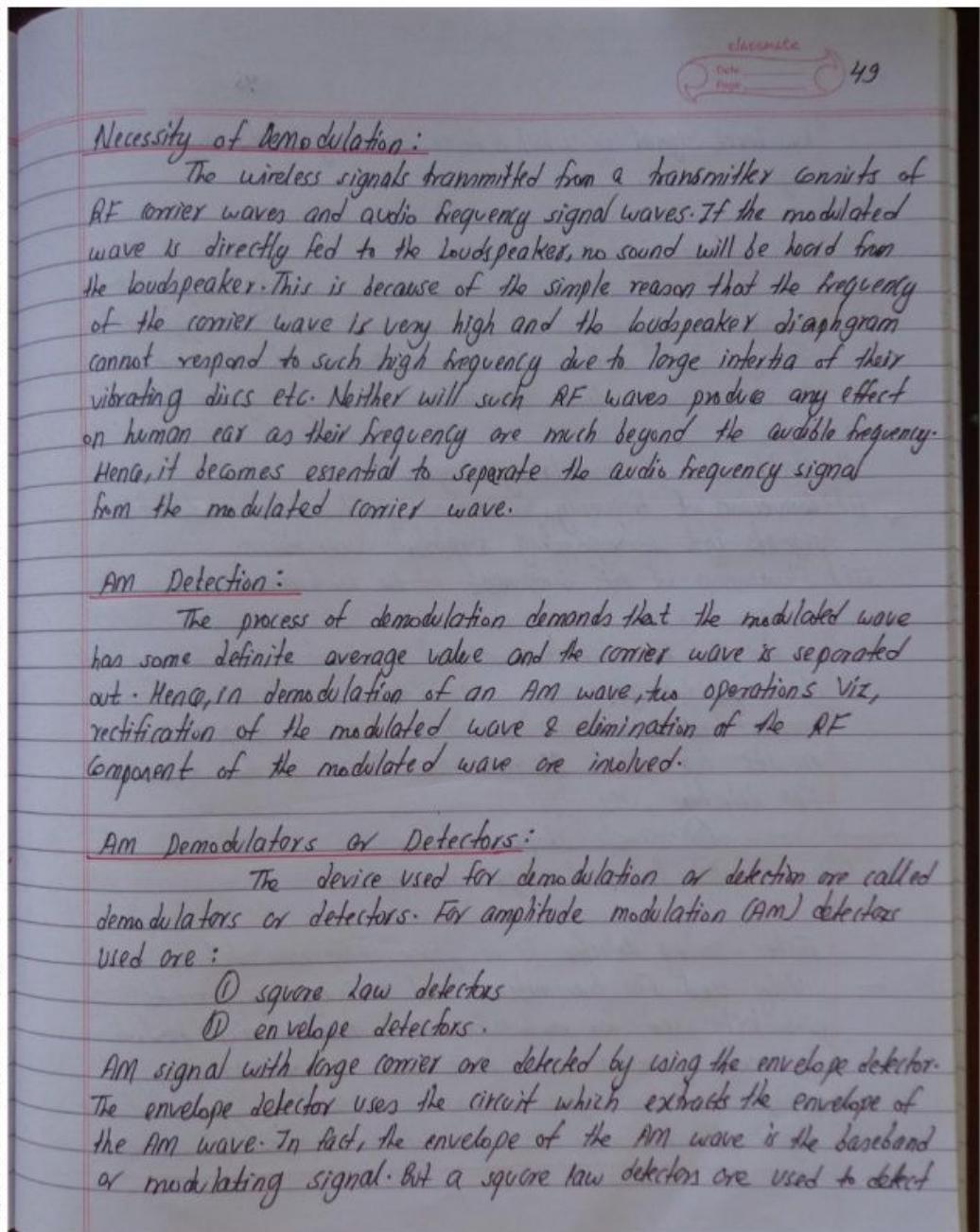


2.9.B Demodulation / Detection: 18/F

Demodulation or detection is a process of recovering the original modulating signal (intelligence) from the modulated carrier wave, i.e., the demodulation is a reverse process of the process of modulation.

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low level signal in which a device operating in the non-linear region.

FM Detection:

For the FM detection, the method usually employed involves the conversion of FM into AM and then applications of conventional method of detection. Thus, demodulation of an FM wave involves three operations which are given to be below :

- i) Rectification of modulating signal
- ii) Conversion of frequency variations produced by modulating signals into corresponding amplitude variations.
- iii) Elimination of AF component of the modulated wave.

For the FM detection, we need a circuit in which magnitude of output voltage varies in accordance with the instantaneous frequency variations in the input voltage. Such circuits are called Discriminators. There are two types of FM detectors, viz

- i) Simple slope detector
- ii) Balanced slope detector .

Foster-Seeley detector is a phase difference detector and is widely used. FM transmitting and receiving equipments particularly used for modulation and demodulation tend to be more complex and hence more costly since, the amplitude of FM signal remains unchanged, the power of FM signal will be same as that of the unmodulated carrier.

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Digital Communications																			
Chapter - 3	To Hovts																		
1) A. Digital Communication:	Data transmission, digital transmission or digital communication is the transfer of data (a digital bit stream or a digitized analog signal) over a point-to-point or point to multipoint communication channel. Data communication traditionally belongs to telecommunications and electrical engineering. Transmitting analog signals digitally allows for greater signal processing capability. The data are represented as an electromagnetic signal such as an electrical voltage, radiowave, microwave or infrared signal.																		
8) Differences:	<table border="1"><thead><tr><th>Digital Communications</th><th>Analog Communications</th></tr></thead><tbody><tr><td>i) Noise immunity</td><td>more noise.</td></tr><tr><td>ii) Greater Bandwidth</td><td>Lesser Bandwidth</td></tr><tr><td>iii) Synchronization problem is relatively difficult.</td><td>- relatively easier.</td></tr><tr><td>iv) Error correction by Coding</td><td>No error correction capability.</td></tr><tr><td>v) can merge different data (voice, audio, video, data) and transmit over a common digital transmission system.</td><td>cannot merge data from different sources.</td></tr><tr><td>vi) Privacy preserved (data encryption)</td><td>No privacy.</td></tr><tr><td>vii) High cost & not portable.</td><td>Low cost and portable.</td></tr><tr><td>viii) - Flexible in implementation.</td><td>Analog Hardware is not flexible.</td></tr></tbody></table>	Digital Communications	Analog Communications	i) Noise immunity	more noise.	ii) Greater Bandwidth	Lesser Bandwidth	iii) Synchronization problem is relatively difficult.	- relatively easier.	iv) Error correction by Coding	No error correction capability.	v) can merge different data (voice, audio, video, data) and transmit over a common digital transmission system.	cannot merge data from different sources.	vi) Privacy preserved (data encryption)	No privacy.	vii) High cost & not portable.	Low cost and portable.	viii) - Flexible in implementation.	Analog Hardware is not flexible.
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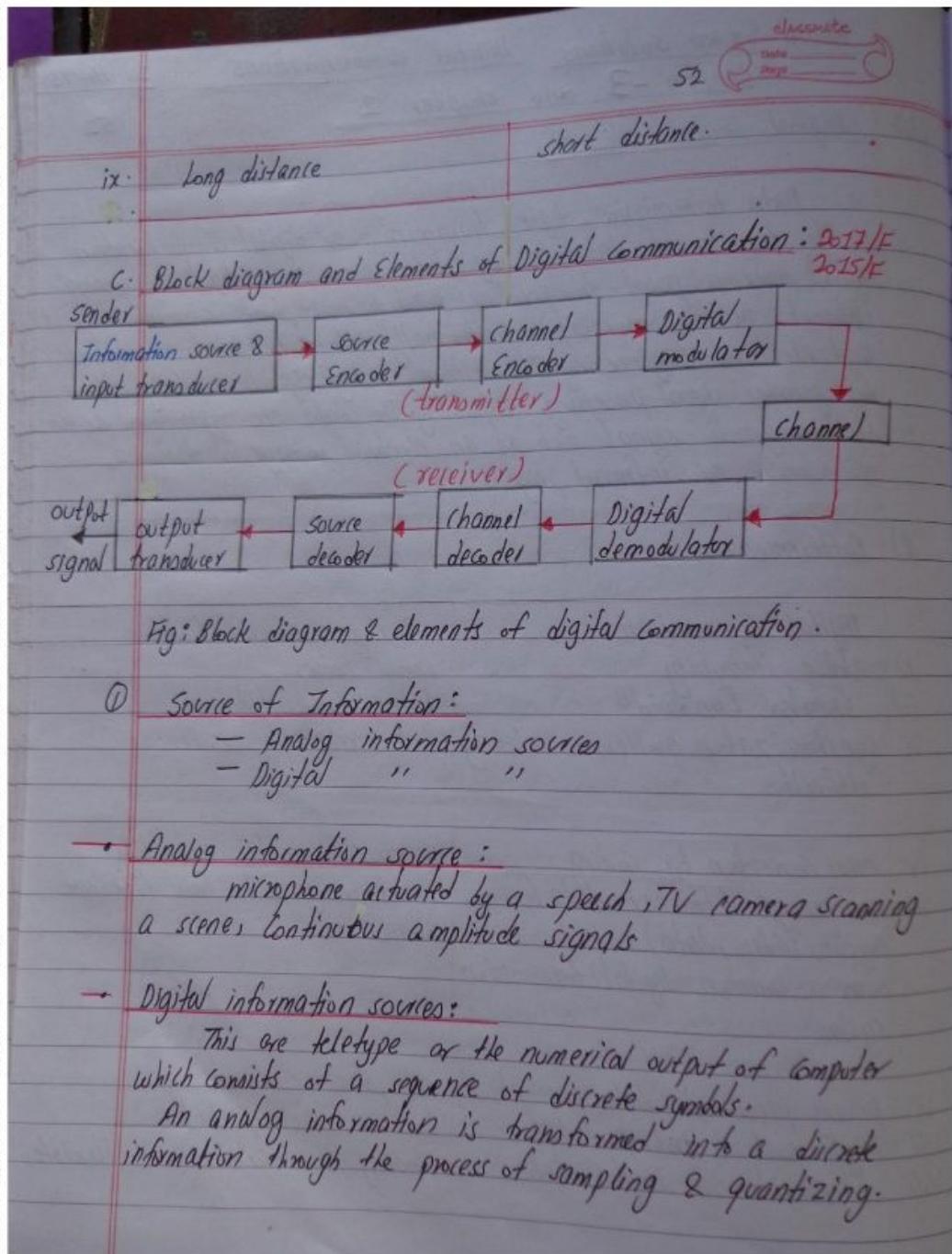


Fig: Block diagram & elements of digital communication.

① Source of Information:

- Analog information sources
- Digital " "

→ Analog information source :

microphone activated by a speech, TV camera scanning a scene, continuous amplitude signals

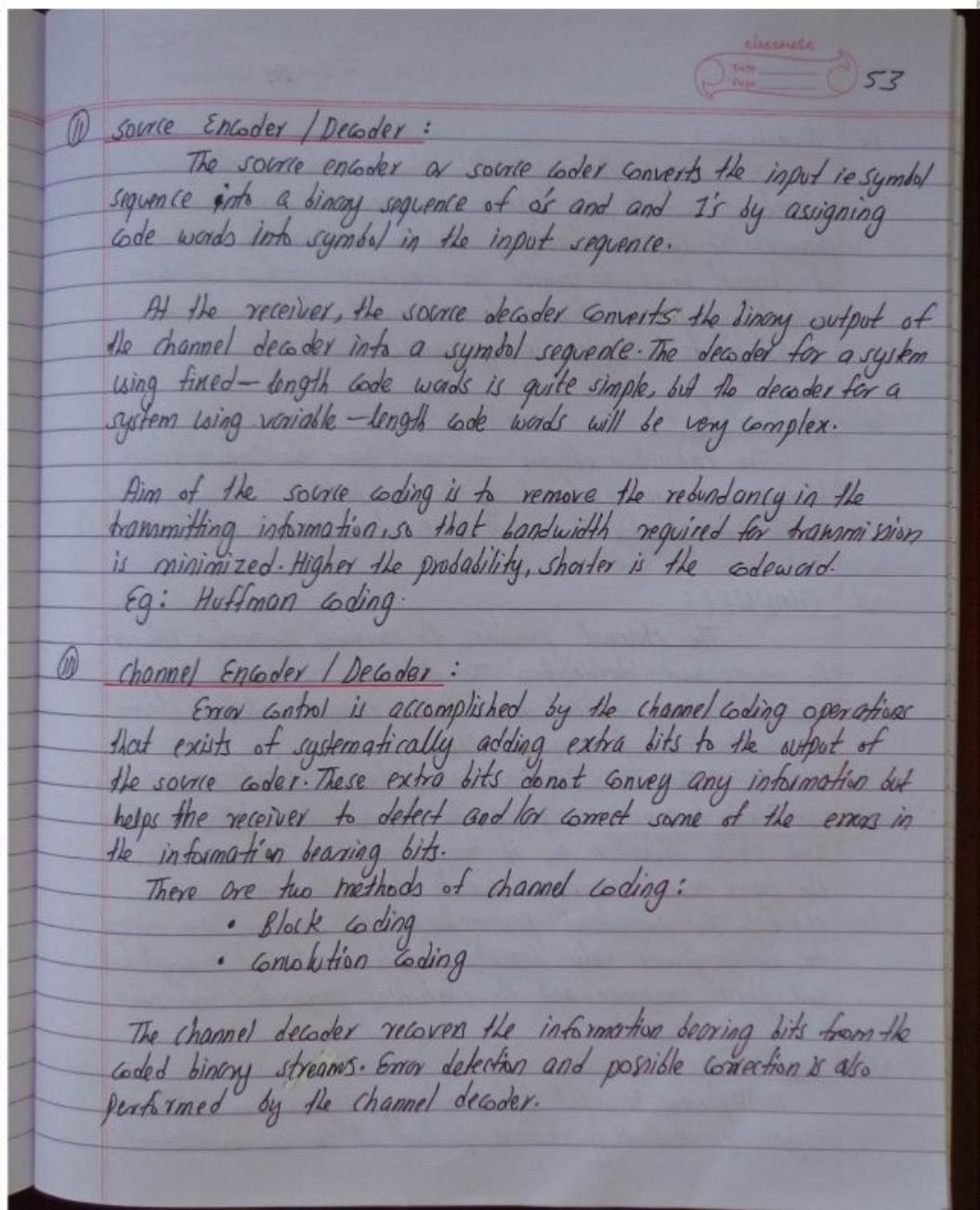
→ Digital information sources:

This are teletype or the numerical output of computer which consists of a sequence of discrete symbols.

An analog information is transformed into a discrete information through the process of sampling & quantizing.

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IV. Modulator:

It converts the input bit stream into an electrical waveform suitable for transmission over the communication channel. It can be effectively used to minimize the effects of channel noise, to match the frequency spectrum of transmitted signal with channel characteristics, to provide the capability to multiplex many signals.

V. Demodulator:

The extraction of the messages from the information bearing waveform produced by the modulation is accomplished by the demodulator. The output of the demodulator is bit stream.

VI. CHANNEL:

The channel provides the electrical connection between the source and destination. The different channels are pair of wires, optical fibre, radio channel, satellite channel or combination of any of these.

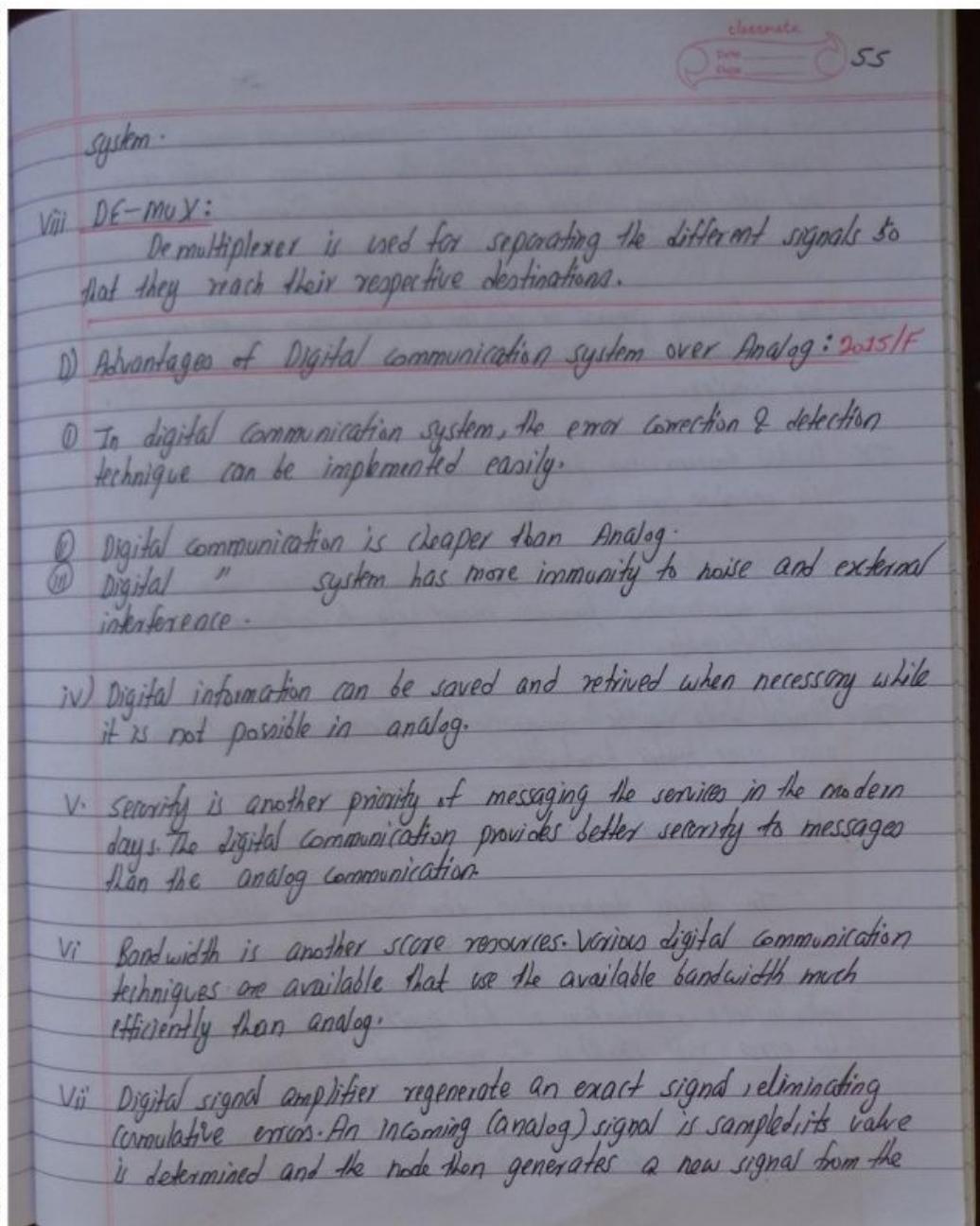
The communication channel have only finite bandwidth, non ideal frequency response, the signal often suffers amplitude and phase distortion as it travels over the channel. Also, the signal power decreases due to the attenuation of the channel. The important parameter of the channel are signal to noise power ratio (SNR), usable bandwidth, amplitude and phase response and the statistical properties of noise.

Vii. MUX:

Multiplexer is used for combining signals from different signals so that they share a portion of the communication

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bit value; the incoming signal is discarded. With analog signal circuits, intermediate nodes amplify the incoming signal, noise and all. Analog circuit requires amplifiers and each amplifier adds distortion and noise to the signal.

VIII. The configuring process of digital communication system is simple as compared to analog communication system, although they are complex.

IX. Digital transmission provides higher maximum transmission rates via medium such as optical fibres.

X. Supports data integrity. Simple to integrate voice, video, & data. Digital transmission provides easier way to integrate different digital formats.

XI. Digital data can be compressed & therefore possible to pass over higher bandwidths.

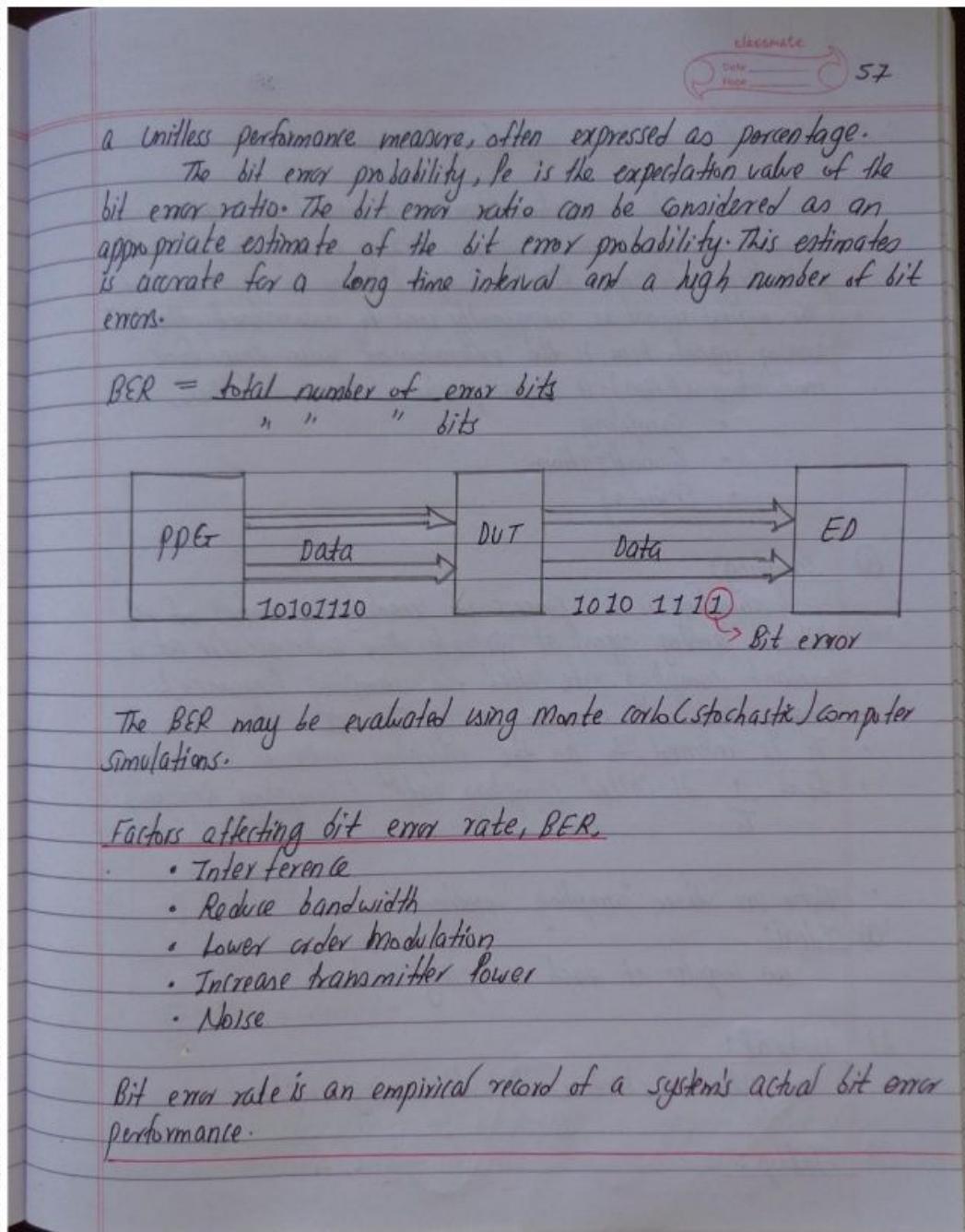
E. Bit error rate :

In digital transmission, the number of bit errors is the number of received bits of a data stream over a communication channel that have been altered due to noise, interference, distortion or bit synchronization errors. The bit error rate (BER) is the number of bit errors per unit time.

The bit error ratio also (BER) is the number of transferred bits during a studied time interval. BER is

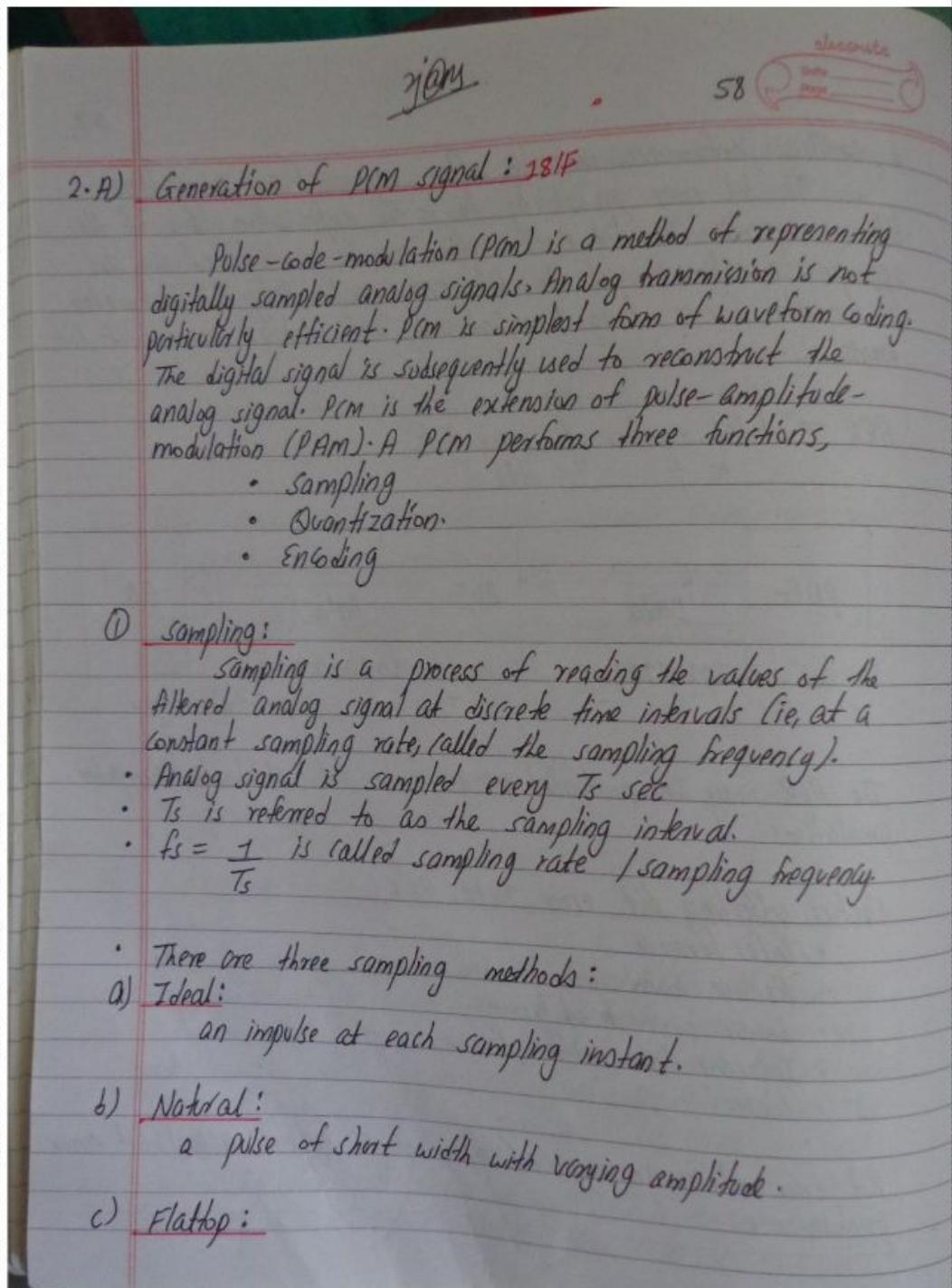
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sample and hold, like natural but with single amplitude value.

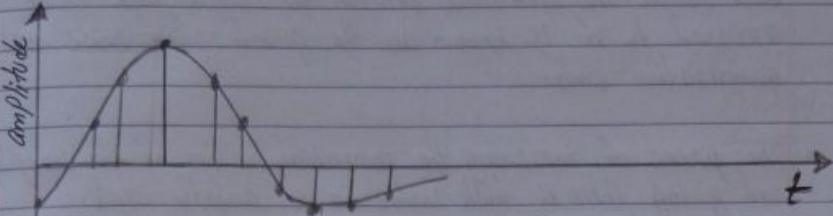


Fig: Ideal sampling.

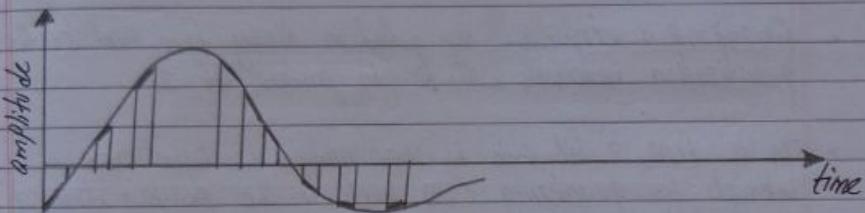


Fig: Natural sampling

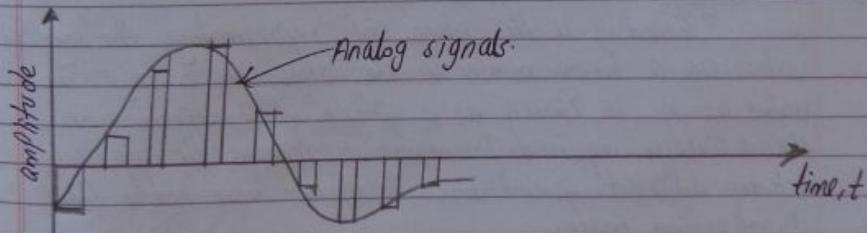


Fig: flat-top sampling.

① Quantizing:

It is the process of assigning a discrete value from a range of possible values to each sample obtained. The number of

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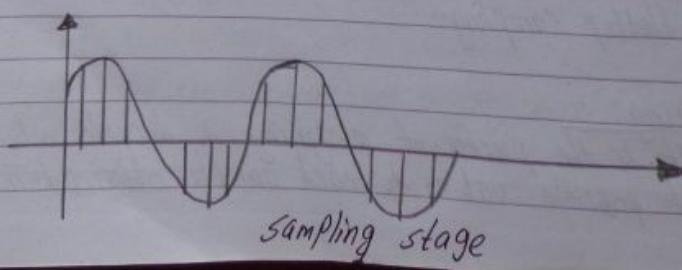
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possible values will depend on the number of bits used to represent each sample. The difference between the sample and value assigned to it is known as the quantization noise or quantization error.

- The process of measuring the numerical values of the samples and giving them a fixed value in a suitable scale.
- Linear quantizing is where the quantizing intervals are of the same size.
- Quantization intervals are coded in binary form, and so the quantization intervals will be in powers of 2.
- In a PCM, 8 bit code is used and so we have 256 intervals for quantizing (128 levels in the positive direction & 128 levels in negative direction).

④ Encoding:

Encoding is the process of representing the sampled values as a binary number in the range of 0 to n. The value of n is chosen as a power of 2, depending on the accuracy required. Increasing n reduces the step size between adjacent quantization levels and hence reduce the quantization noise.



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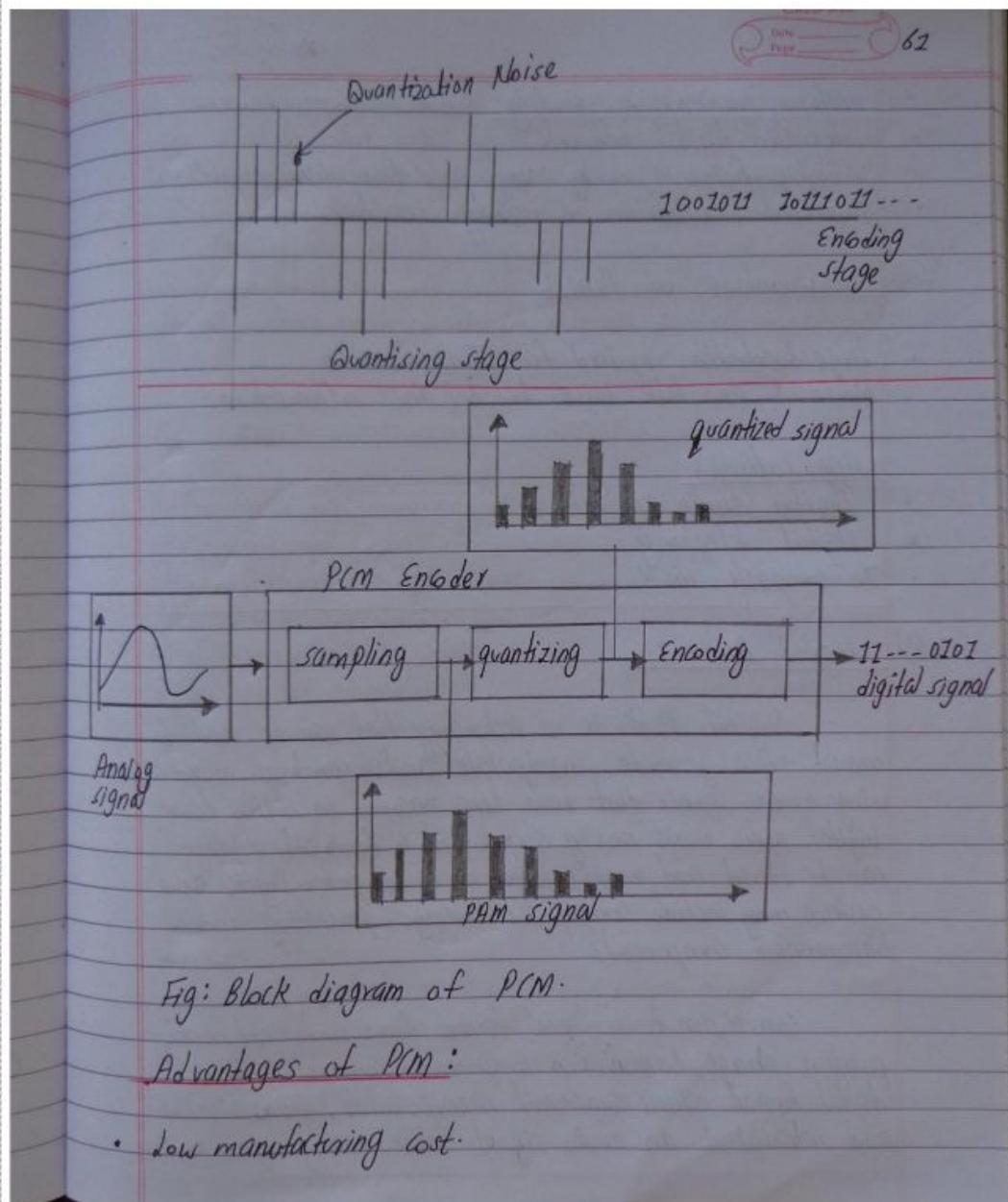


Fig: Block diagram of PCM.

Advantages of PCM :

- Low manufacturing cost.

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• uniform transmission quality
• integrated digital network
• Good performance over very poor transmission paths.
• Increased utilization of existing circuit

Disadvantages of Pcm:

- Large bandwidth required for Transmission
- Noise & crosstalk leaves low but rises attenuation.

Applications:

- Digital audio applications
- Digital telephony
- In compact disk.

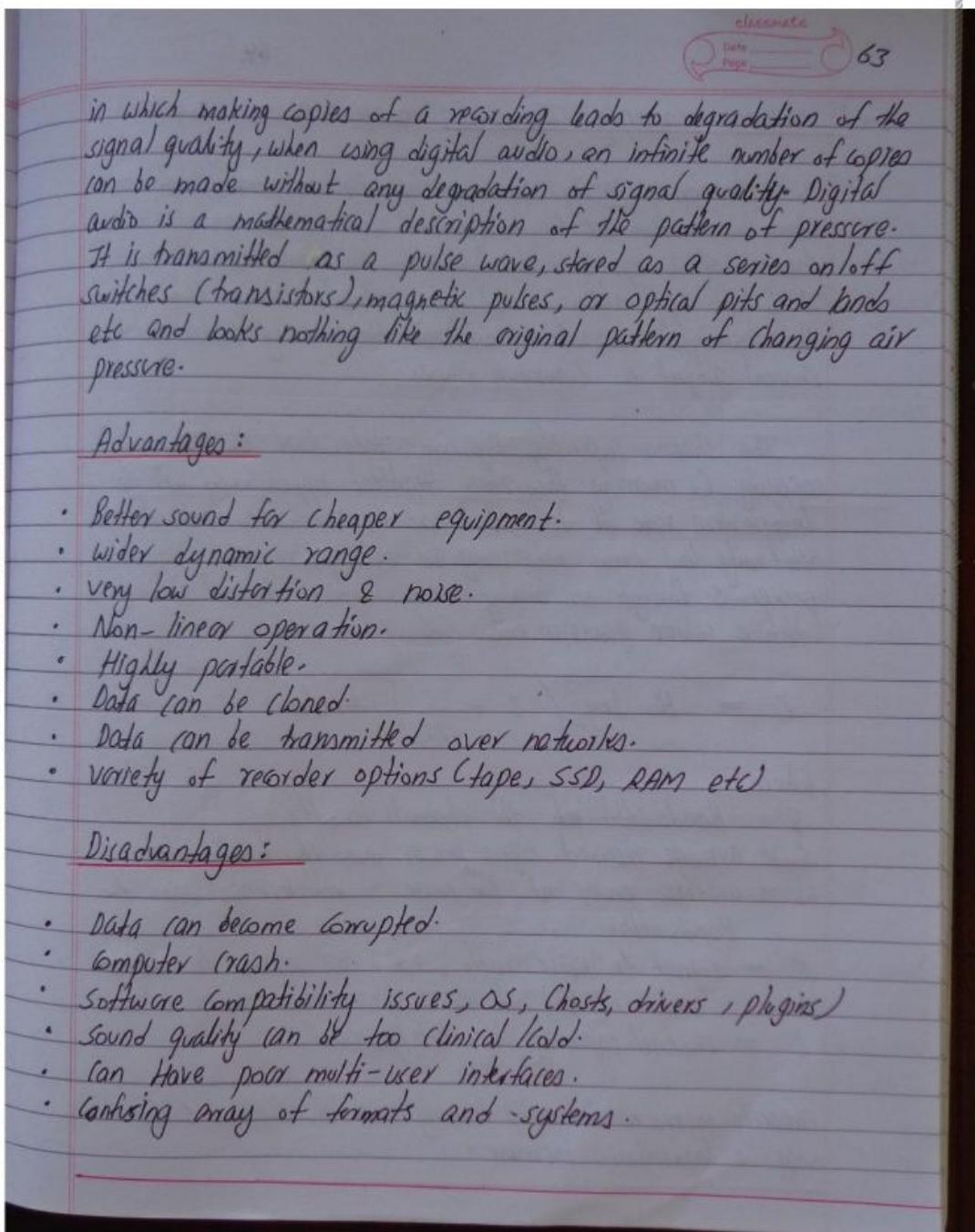
2-B) Digital Audio :

Digital Audio is a technology that can be used to record, store, generate, manipulate and reproduce sound using audio signals that have been encoded in digital form. Digital audio brings analog sounds into a form where they can be stored and manipulated on a computer. Digital audio systems may include compression, storage, processing and transmission components.

Conversion to a digital format allows convenient manipulation, storage, transmission components. Conversion to a digital format allows convenient manipulation, transmission and retrieval of an audio signal. Unlike analog signal audio,

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2.c) Shannon Capacity Theorem: 2014/S, 2015/S, 2016/F

Shannon capacity theorem tells the maximum rate at which information can be transmitted over a communications channel of a specified bandwidth in the presence of noise. It is an application of the noisy channel coding theorem to the archetypal case of a continuous time analog communications channel subject to Gaussian noise.

The Shannon capacity theorem states that the channel capacity C , meaning theoretical tightest upper bound on the information rate of data that can be communicated at an arbitrarily low error rate using an average received signal power S through an analog communication channel subject to additive white Gaussian noise of power N ,

$$C = B \log_2 \left(1 + \frac{S}{N} \right) \text{ bits/sec}$$

where,

B = Bandwidth of the channel in Hz

S = Average received signal power over the bandwidth.

N = average power of the noise & interference over the bandwidth.

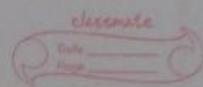
$\frac{S}{N}$ = signal to noise ratio (SNR) or carrier to noise ratio (CNR) of the communication signal to noise.

C = channel capacity in bits/second

Capacity increases (almost) linearly with B , whereas S determines only a logarithmic increase.

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3.A) Bandwidth efficiency:

Spectral efficiency, spectrum efficiency or bandwidth efficiency refers to the information rate that can be transmitted over a given bandwidth in a specific communication systems. It is a measure of how efficiently a limited frequency spectrum is utilized by the physical layer protocols and sometimes by the media access control.

Bandwidth efficiency is defined as the number of bits that can be transmitted within 1 Hz of bandwidth,

$$\therefore \eta = \frac{R_b}{B_T} \text{ bits/sec/Hz}$$

$$\begin{aligned}\therefore \text{BW efficiency} &= \frac{\text{transmitting rate (bps)}}{\text{minimum Bandwidth (Hz)}} \\ &= \frac{\text{bits/second}}{\text{Hertz}} \\ &= \frac{\text{bits}}{\text{cycle.}}\end{aligned}$$

B- ASK : (Amplitude - shift - Keying) : 2015/2017/s

Amplitude-shift keying (ASK) is a form of amplitude modulation that represents digital data as variations in the amplitude of a carrier wave. In an ASK system, the binary symbol 1 is represented by transmitting a fixed-amplitude carrier wave and fixed frequency for a bit duration of T seconds. If the signal value is 1, then the carrier signal will be transmitted, otherwise, a signal value of 0 will be transmitted.

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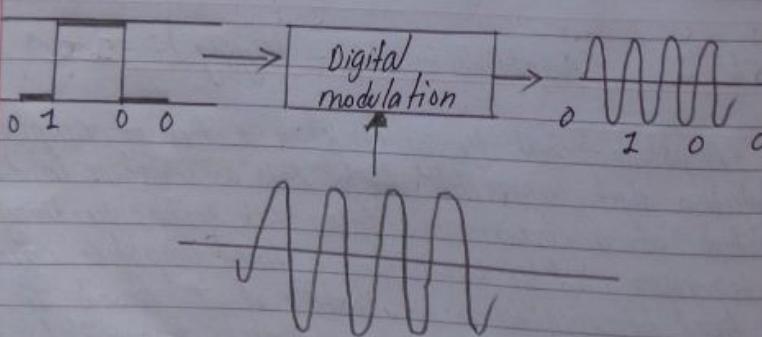
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Like AM, ASK is also linear and sensitive to atmospheric noise, distortions, propagation condition on different routes in PSTN etc. Both ASK modulation and demodulation process are relatively inexpensive. The ASK technique is also commonly used to transmit digital data over optical fiber.

Digital to Analog Conversion -

ASK FSK PSK

In this technique, amplitude of the RF carrier is varied in accordance with baseband digital input signal. The figure depicts operation of ASK modulation. As shown in figure, binary 1 will be represented by carrier signal with some amplitude while binary 0 will be represented by carrier of zero amplitude (i.e. no carrier).



ASK modulation can be represented by,
 $s(t) = A \cos(2\pi f_c t)$ for Binary 1.

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$s(t) = 0$ for Binary 0

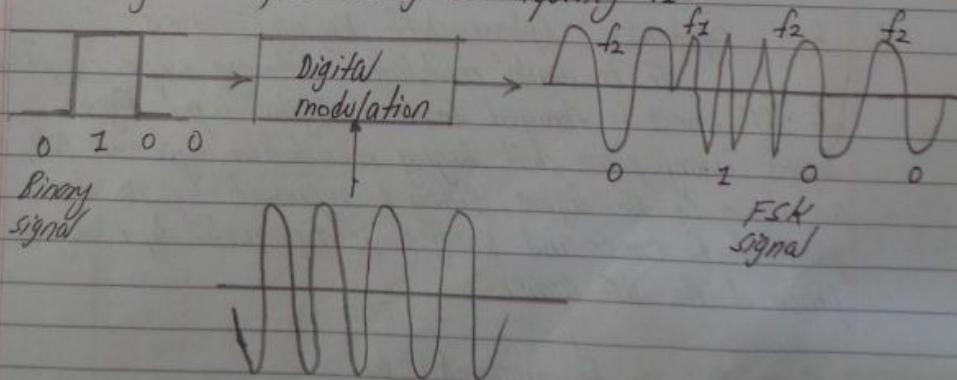
Bandwidth requirement for ASK is,

$$B_W = \frac{2}{T_b} = 2 \times f_B$$

- In ASK, probability of error (P_e) is high and SNR is less.
- It has lowest noise immunity against noise.
- ASK is a bandwidth efficient system but it has lower power efficiency.

C. FSK : (Frequency shift keying): 2015/15

It is also known as digital modulation technique. In this technique, frequency of the AF carrier is varied in accordance with baseband digital input. The figure depicts the FSK modulations. As shown, binary 1 and 0 represented by two different carrier frequencies. Figure depicts that binary 1 is represented by high frequency f_1 and binary 0 is represented by low frequency f_2 .



Binary FSK can be represented by,
 $s(t) = A \cos(2\pi f t)$ for binary 1.

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$$s(t) = A \cos(2\pi f_c t) \text{ for binary 0}$$

$$\text{Bandwidth (BW)} = 2 \times R_b + (f_1 - f_2)$$

- In case of FSK, R_b is low and SNR is high.
- This technique is widely employed in modern designs and development.
- It has increased immunity to noise but requires larger bandwidth compare to other modulation types.

Generation:

carrier: $\cos(2\pi f_c t)$

message: $m(t)$

-90°
phase shift

BPSK output

$A \cos(2\pi f_c t + \delta m(t))$

D. PSK systems: (Phase shift Keying): 2015b

It is digital modulation technique where in phase of RF carrier is changed based on digital input. Figure depicts binary phase shift keying modulation type of PSK, as shown in the figure, Binary 1 represents by 180 degree phase of the carrier and binary 0 is represented by 0 degree phase of the RF carrier.

Binary PSK can be represented by,

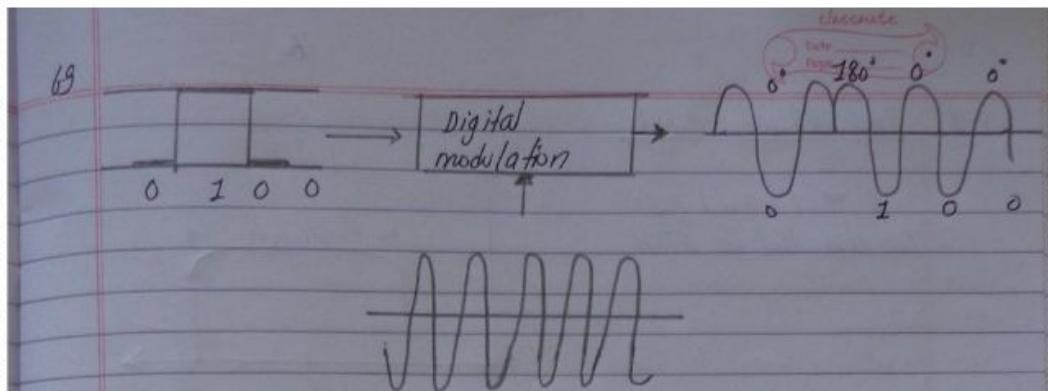
$$s(t) = A \cos(2\pi f_c t) \text{ for binary 1}$$

$$s(t) = A \cos(2\pi f_c t + \pi) \text{ for binary 0}$$

$$\therefore \text{Bandwidth} = 2 \times R_b = 2 \times \text{bit rate}$$

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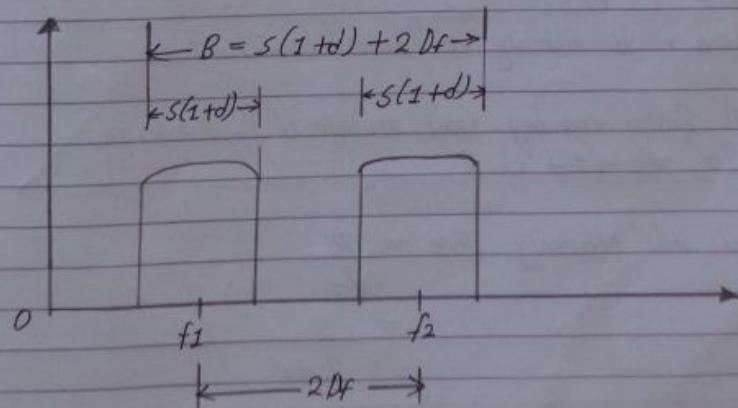
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- PSK modulation is widely used in wireless transmission.
- The variants of basic PSK and ASK modulations are QAM, 64-QAM, 16-QAM and so on.
- It is a power efficient system but it has lower bandwidth efficiency.
- Probability of error is less and SNR is high.

4.A) Bandwidth of FSK:

$$\text{Bandwidth} = (1+d) \times s + 2\Delta f$$



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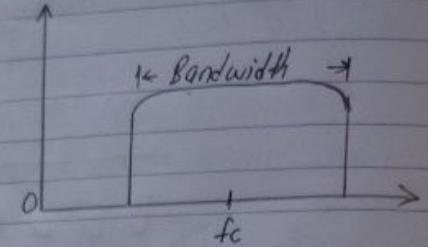
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B) Bandwidth of PSK:

$$B = (1+d) \times S$$

- Less than that for BFSK



C) Bandwidth of QAM:

- quadrature amplitude modulation
- combination of ASK & PSK
- large bandwidth savings.

D) Bandwidth of BPSK:

- B_T directly related to the (signaling, modulation, bandwidth) rate, D.

$\Rightarrow B_T = (1+r) D$.
where, r is filtering coefficient; or $r \ll 1$.

\Rightarrow with Binary encoding (not multilevel), D = R, so,
 $B_T = (1+r) R$

Bandwidth efficiency, $B_E = \frac{R}{B_T}$
 $= \frac{1}{1+r}$

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Board Exam solutions of chapter 1, 2, 3

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Q1A. What is DSB-SC modulation? Justify with necessary spectrums "DSB-SC" is wasteful for transmission bandwidth than SSB-SC "-But, why SSB not used for broadcasting - Give reasons. 2015/F, 2014/S

⇒ Double-side band suppressed-carrier transmission (DSB-SC) is transmission in which frequencies produced by amplitude modulation (AM) are symmetrically spaced above and below the carrier frequency and the carrier level is reduced to the lowest practical level, ideally being completely suppressed. It is used for radio data system.

In DSBSC system, carrier component is altogether removed resulting in saving of enormous amount of power. DSB-SC is generated by a mixer.

$$s(t) = c(t) \cdot m(t)$$
$$= A_c \cdot \cos(2\pi f_c t) \cdot m(t)$$

By taking Fourier transform on both sides,

$$\therefore S(f) = \frac{A_s}{2} [M(f-f_c) + M(f+f_c)]$$

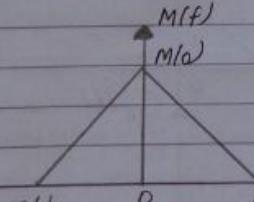


Fig: message signal spectrum

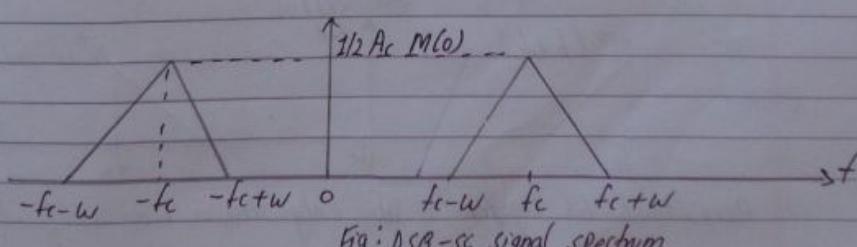
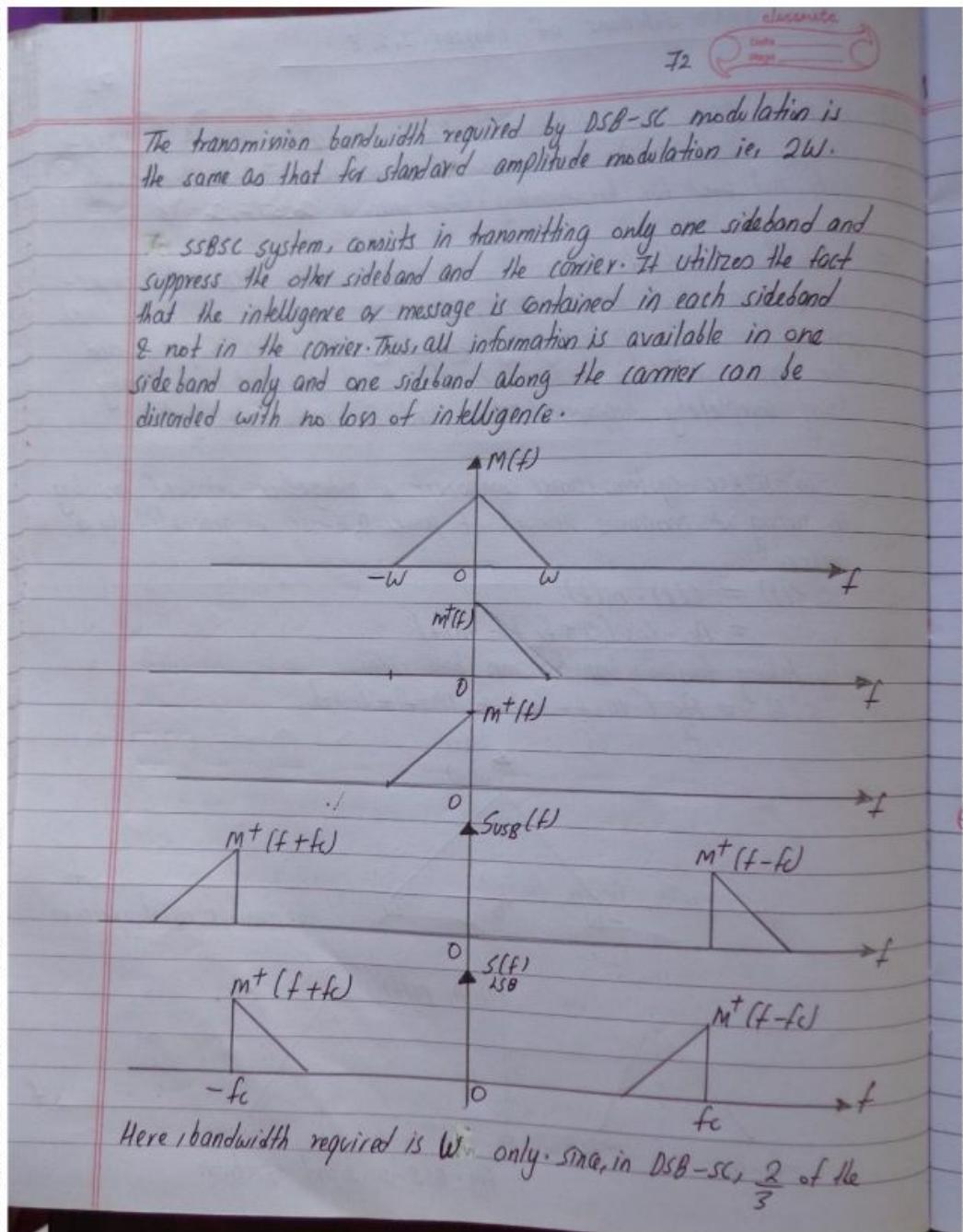


Fig: DSB-SC signal spectrum

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total transmitted power taken up by the carrier. DSB has a wide bandwidth, i.e. info of USB = info of LSB. But, in SSB, half as much as bandwidth required in DSBSC.

$$\text{ie. } B_{WSSB} = \frac{1}{2} \cdot B_{WDSBSC}$$

Hence, DSB-SC is wasteful for transmission.

Reason for not using SSB for Broadcasting:

- ① The generation and reception of SSB signal is complicated. & therefore, limited to radio telephony.
- ② The SSB transmitter and receiver need to have an excellent frequency stability. A slight change in frequency will hamper the quality of transmitted and received signal. Hence, it is used for speech transmission.
However, DSBSC transmission system is widely used in broadcasting because of its relative simplicity of its modulating equipment.

B) Define baseband transmission? Why is modulation used in communication system? *2017/F*

⑤ \Rightarrow Baseband transmission is a transmission in which digital signals are sent through direct current (DC) pulses applied to the wire. It supports half duplexing.
Example: Ethernet is an example of a Baseband system found on many LANs.

Modulation Necessity in Telecommunication are :

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c) Perform the performance comparison between FM & AM system. 2016/5, 2017/1	AM
⑧ \Rightarrow FM	AM can be received as analog.
① FM can be received as digital.	The amplitude of carrier remains constant.
② The amplitude of carrier remains constant.	The amplitude of carrier changes with modulation.
③ The frequency changes with the modulation.	The carrier frequency remains constant with modulation.
④ The value of modulation index (M) can be more than 1.	The value of modulation factor cannot be more than 1 for distortionless Am signal.
V. FM receivers are immune to noise. Am receivers are not immune to noise.	
VI. All transmitted power is useful. carrier power and one sideband power is useless.	
VII. FM broadcasts operate in the upper VHF and UHF frequency ranges. at which there happens to be less noise.	Am broadcast operate in the MF and HF ranges where there exists more noise.
VIII. FM is taken in high frequency & its stereo.	Am is taken in low frequency and not stereo.
IX. The amplitude & phase remains constant.	The frequency & phase remain constant.

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x. Frequency range : in a higher AM ranges from 535 to 175 kHz spectrum from 88 to 108 MHz or upto 1200 bits/sec 1200 to 2400 bits/sec.

xi. Bandwidth required is twice the sum of the modulating signal frequency and the frequency deviation. Bandwidth required is twice the highest modulating frequency.

xii. Zero crossing in modulated signal is not equidistant. Equidistant.

xiii. Good for local radio because of the smaller range. Not ideal because the range is too big.

D) Define amplitude modulation. Derive the expression for AM in frequency domain and show that transmission bandwidth for an AM wave is exactly twice the message signal bandwidth.

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2016/5

⇒ The process of varying amplitude of the high frequency or carrier wave in accordance with the intelligence (code, voice, music) to be transmitted, keeping the frequency and phase of the carrier wave unchanged, is known as amplitude modulation (AM).

Consider a sinusoidal carrier wave, $c(t)$,

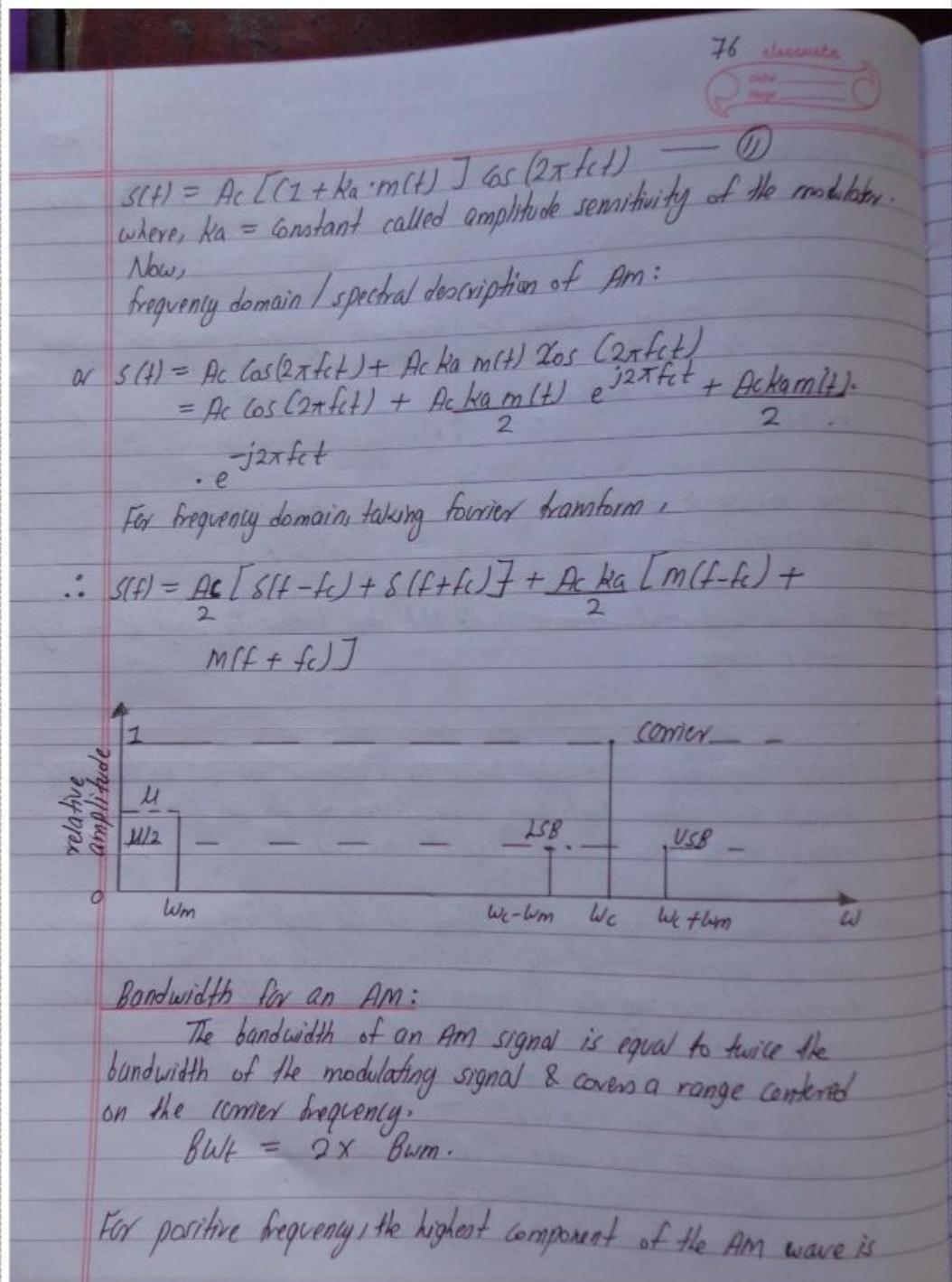
$$c(t) = A_c \cos(2\pi f_c t) \quad \text{--- (1)}$$

where, A_c = peak value of carrier amplitude
 f_c = carrier frequency

The standard form of amplitude modulation (AM) wave is,

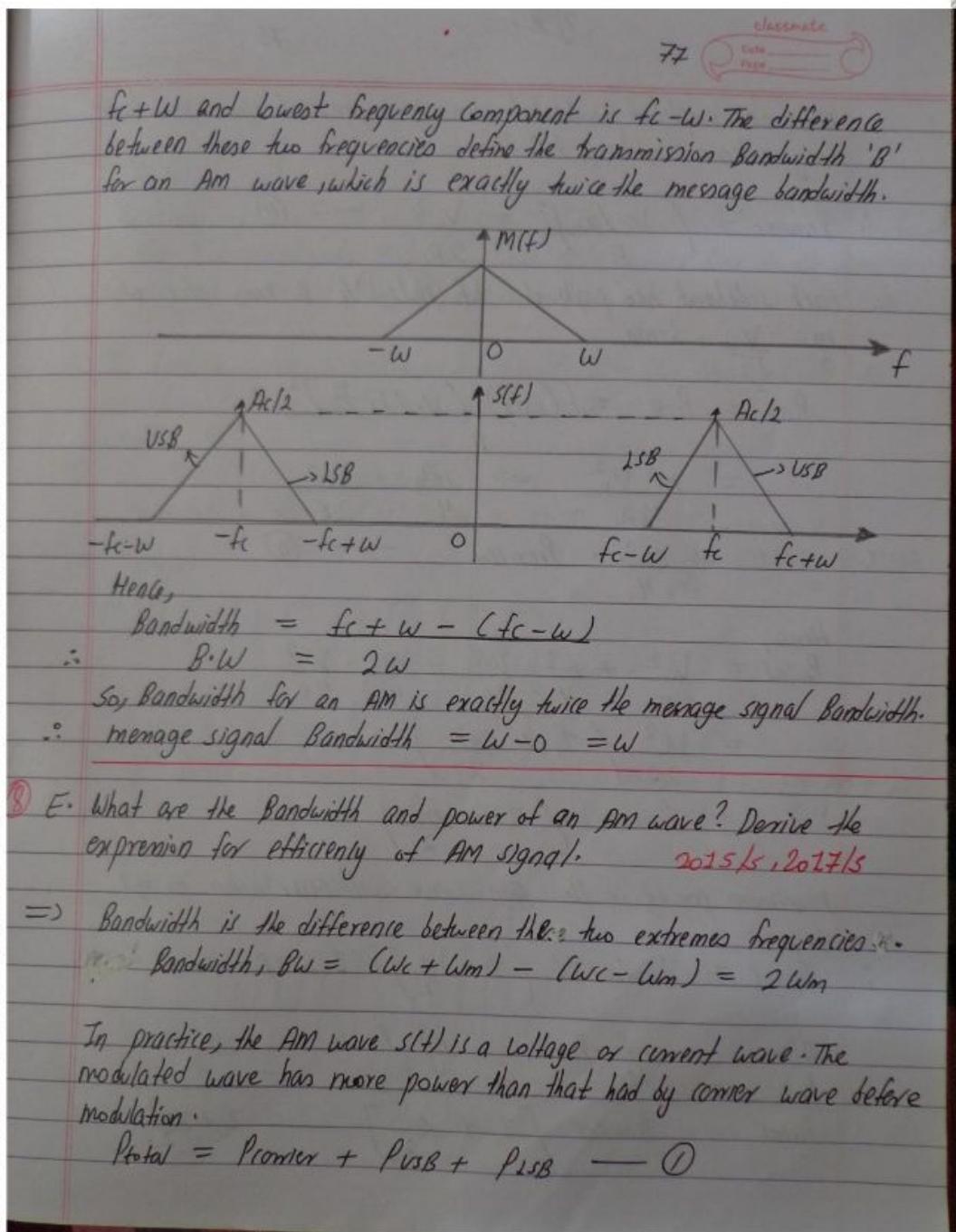
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When an amplitude modulated wave is impressed upon resistance
(say antenna resistance) R , Hen.

$$\therefore P_{\text{carrier}} = \left(\frac{V_c / \sqrt{2}}{R} \right)^2 = \frac{V_c^2}{2R} \quad \text{--- (1)}$$

Each sideband has peak value of $(m/2)V_c$ & rms value of
 $\frac{m}{2} \cdot \frac{V_c}{\sqrt{2}}$, hence,

$$P_{\text{USB}} = P_{\text{LSB}} = \left[(m/2) \left(V_c / \sqrt{2} \right) \right]^2$$

$$= \frac{m^2 V_c^2}{8R} \Rightarrow \frac{m^2}{4} \cdot \frac{V_c^2}{2R}$$

$$\therefore = \frac{m^2}{4} \cdot P_{\text{carrier}} \quad \text{--- (2)}$$

Hence,

$$P_{\text{total}} = \frac{V_c^2}{2R} + \frac{m^2 \cdot V_c^2}{4} + \frac{m^2 \cdot V_c^2}{2R}$$

$$= \frac{V_c^2}{2R} \left(1 + \frac{m^2}{2} \right)$$

$$= P_{\text{carrier}} \left(1 + \frac{m^2}{2} \right)$$

Maximum power in the AM wave will occur when $m = 1$,
i.e.,

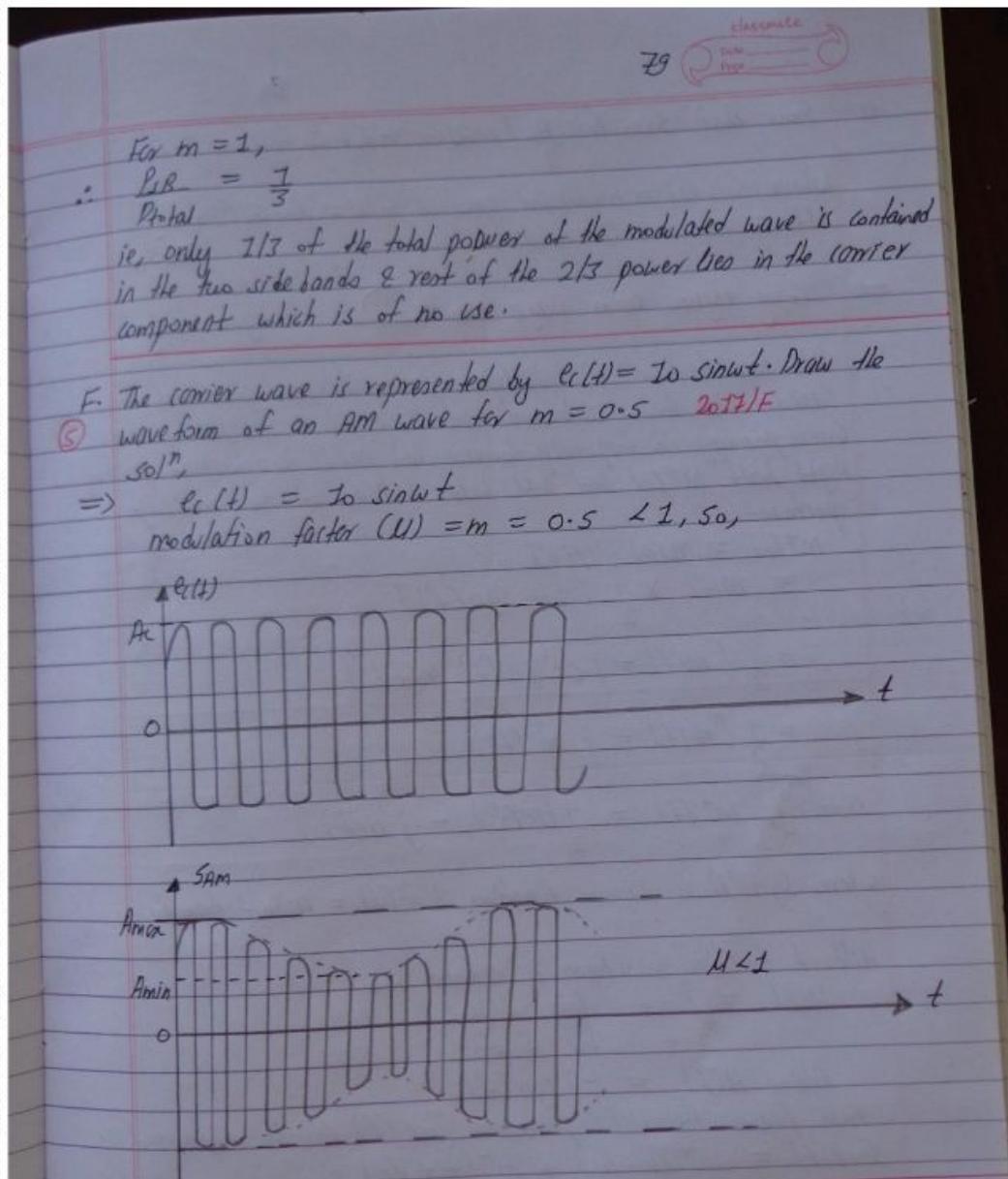
$$\therefore P_{\text{total}} = P_{\text{carrier}} \times \left(1 + \frac{1}{2} \right) = 1.5 \text{ power}$$

Also,

$$\frac{P_{\text{SB}}}{P_{\text{total}}} = \frac{\frac{m^2}{4} \cdot P_{\text{carrier}}}{P_{\text{carrier}} \left[1 + \frac{m^2}{2} \right]} = \frac{\frac{m^2}{4} P_{\text{carrier}}}{1 + \frac{m^2}{2}} = \frac{\frac{m^2}{4}}{1 + \frac{m^2}{2}}$$

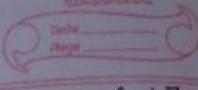
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G. Prove that $S_{SSB}(t) = \frac{Ac}{2} [m(t) \cos 2\pi f_c t + \hat{m}(t) \sin 2\pi f_c t]$

where, $S_{SSB}(t)$ = modulated signal & $m(t) = Am \cos(2\pi f_m t)$
 $\hat{m}(t) = Ac \cos(2\pi f_c t)$ 2015/16

\Rightarrow Copy figure from page No: 72 ; Figure only.

Let, S_{SSB} be an SSB modulated wave with upper sideband & $S_{SSB}(t)$ with lower sideband and $S_{USB}(f)$ & $S_{LSB}(f)$ be the corresponding in frequency domain. Let, $M(f)$ be message spectrum. Right half $M^+(f)$ and left half $M^-(f)$ of the message spectrum.

$$\begin{aligned} M^+(f) &= M(f) \cdot U(f) \\ &= M(f) \cdot \chi \frac{1}{2} [1 + \operatorname{sgn}(f)] \\ &= \frac{1}{2} [M(f) + j(-j \operatorname{sgn}(f) \cdot M(f))] \\ &= \frac{1}{2} [M(f) + j \hat{M}(f)] \end{aligned}$$

Similarly, $M^-(f) = \frac{1}{2} [\bar{M}(f) - j \hat{M}(f)]$

Where, $\operatorname{sgn}(\cdot)$ is signum function & $\hat{M}(f)$ = Hilbert transform

Hilbert function transform is defined as,

$$H(f) = \begin{cases} -f & \text{for } f > 0 \\ f & \text{for } f \leq 0 \end{cases}$$

Also, $H(f) = -j \operatorname{sgn}(f)$

From figure, the USB spectrum $S_{USB}(f)$ can be expressed as,
 $S_{USB}(f) = M^+(f-f_c) + M^-(f+f_c)$

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$$\begin{aligned} &= \frac{1}{2} [m(f-f_c) + j\hat{m}(f-f_c)] + \frac{1}{2} [m(f+f_c) - j\hat{m}(f+f_c)] \\ &= \frac{1}{2} [m(f-f_c) + m(f-f_c) - \frac{1}{2j} [\hat{m}(f-f_c) + \hat{m}(f+f_c)]] \end{aligned}$$

using frequency shifting property, & taking inverse Fourier transforms,

$$S_{USB}(t) = m(t) \cos(\omega_c t) - \hat{m}(t) \sin(\omega_c t)$$

This is the required time-domain expression for upper sideband.

Similarly, for lower sideband S_{LSB} ,

$$S_{LSB} = m(t) \cos(\omega_c t) + \hat{m}(t) \sin(\omega_c t)$$

In general,

$$\therefore S_{SSB}(t) = m(t) \cos(\omega_c t) \pm \hat{m}(t) \sin(\omega_c t)$$

H. An audio signal given by $15 \sin 2\pi (2000 t)$ amplitude modulates a sinusoidal carrier wave $60 \sin 2\pi (100000 t)$, determine,

- (i) modulation index
- (ii) % modulation 2015 LS, 2015/F
- (iii) Frequency of signal & carrier
- (iv) Frequency of USB & LSB of modulated wave.
Soln:

\Rightarrow carrier wave, $c(t) = 60 \sin(2\pi \times 100000 t)$
message " , $m(t) = 15 \sin(2\pi \times 2000 t)$
Comparing with,
 $m(t) = Am \sin(2\pi f_m t + \phi)$
 $= 15 \sin(4\pi \times 2000 t + \phi)$

$\therefore f_m = 2 \text{ KHz}$; $Am = 15$

and,

$$\begin{aligned} c(t) &= Ac \sin(2\pi f_c t + \phi) \\ &= 60 \sin(2\pi \times 100000 t + \phi) \end{aligned}$$

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$$\therefore f_c = 100 \text{ KHz} ; A_c = 60$$

① modulation index, $M = m = |k_a \cdot m(t)|_{\max}$
letting, $k_a = 0.02$, (Not given),
So,

$$M = 0.02 \times 15 = 0.3$$

② % modulation = $0.3 \times 100 = 30 \%$

③ Frequency of carrier = 100 KHz
" " message = 2 KHz

④ Frequencies of USB = $f_c + f_m = (2 + 100) \text{ KHz}$
= 102 KHz
" " LSB = $f_c - f_m = (100 - 2) \text{ KHz}$
= 98 KHz

Q. calculate the % power saving for a DSB-SC signal for the
⑤ % modulation of ① 100% ② 50% 2017/F

Soln,

$$\Rightarrow \% M = 100 \\ M = 1$$

$$\% \text{ power saving} = \frac{P_c}{P_t} \times 100 \%$$

① $= \frac{P_c}{P_c / (1 + \frac{M^2}{2})} = \frac{1}{(1 + \frac{1^2}{2})} = 66.67 \%$

② $M = 0.5$

$$\% \text{ Power saving} = \frac{1}{(1 + \frac{0.5^2}{2})} = 88.89 \%$$

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J. A certain AM transmitter radiates 9 kW with the carrier unmodulated and 10.125 kW when the carrier is simultaneously modulated, find

① modulation index

2016/F

② If another sine wave corresponding to 60% modulation is transmitted simultaneously, determine total radiated power.

2017/S

Sol:

\Rightarrow power of carrier wave, $P_c = 9 \text{ kW}$

total power when modulated, $P_t = 10.125 \text{ kW}$

$$\text{① modulation index } (M) = \sqrt{2 \left(\frac{P_t}{P_c} - 1 \right)}$$

$$= \sqrt{2 \times \left(\frac{10.125}{9} - 1 \right)}$$

$$= \sqrt{0.25}$$

$$\therefore M = 0.50$$

② $\% M_2 = 60\% \Rightarrow M_2 = 0.60$

Total radiated power, $P_t = ?$

$$M_1 = 0.50$$

$$P_c = 9 \text{ kW}$$

The total modulation index by two sinusoidal signal is

$$M_t = \sqrt{M_1^2 + M_2^2} = \sqrt{0.6^2 + 0.5^2} = 0.7810$$

we have,

$$P_t = P_c \left(1 + \frac{M_t^2}{2} \right) = 9 \times \left(1 + \frac{0.7810^2}{2} \right)$$

$$\therefore P_t = 11.745 \text{ kW}$$

K. The total power content of an AM signal is 900 watt. Determine the

power being transmitted at the carrier frequency & at each of the sidebands when % modulation is 75%.

2014/S, 2015/F

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So/11,

$$\Rightarrow \text{Total power } (P_t) = 900 \text{ watt.}$$

% modulation (M) = 75 %

$$M = 0.75$$

we have,

$$P_t = P_c + P_{SSB}$$

$$= P_c + P_c \times \frac{M^2}{2}$$

$$\therefore P_t = P_c (1 + \frac{M^2}{2})$$

$$\therefore P_c = \frac{900}{(1 + \frac{0.75^2}{2})}$$

$$\therefore P_c = 702.44 \text{ watts.}$$

Also,

$$P_{SSB} = P_{USB} = M^2 \times P_c$$

4

$$= (0.75)^2 \times 702.44$$

4

$$\therefore P_{SSB} = P_{USB} = 98.78 \text{ watt}$$

1) A system has bandwidth of 44 KHZ & signal to noise ratio of 28 dB at input to receiver, calculate

① information carrying capacity

② capacity of channel if its Bandwidth is double & transmitted signal power remains constant.

2015/K

So/11,

$$\Rightarrow \text{Bandwidth } (B.W) = 44 \text{ KHZ}$$

$$\text{Signal to noise ratio (SNR)} = 28 \text{ dB.}$$

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Also,
 $C = 0.332 \times \text{Bandwidth} \times \text{SNR in dB}$

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we know,

(1) carrying capacity (C) = $B \log_2 (1 + \text{SNR})$
Here,
 $\text{SNR in dB} = 28 \text{ dB} = 10 \log_{10} (\text{SNR})$

& $\text{SNR} = 10^{2.8} = 10^{28/10}$
 $\therefore \text{SNR} = 630.957$

So,
 $c = B \log_2 (1 + \text{SNR}) = 44 \times 10^3 \log_2 (1 + 630.957)$
 $\therefore \text{carrying capacity } (C) = 409362.033 \text{ bps}$
 $= 0.4094 \text{ Mbps}$

(2) Bandwidth is doubled ($B = 2 \times 44 = 88 \text{ kHz}$),
since, power transmitted remains constant, $P = E_s R_b$,
 $c = B \log_2 (1 + \text{SNR})$
 $= 88 \times 10^3 \log_2 (1 + \frac{\text{signal power}}{\text{Noise power}})$
 $= 88 \times 10^3 \log_2 \left(1 + \frac{630.957}{2} \right)$
 $= 730924.8041 \text{ bps}$
 $= 0.7309 \text{ Mbps}$.

M. A single tone FM is represented by the voltage eq" as $V(t) = 25 \cos (8 \times 10^{-8} t + 10 \sin 1520 t)$. Determine:
(i) C.F 2027 B/S
(ii) modulating frequency 2017/S
(iii) mf 2018/F
(iv) $(\Delta w)_{\text{max}}$.
V. What power will this FM wave dissipate in 2Ω resistor.
Soln:
 $\Rightarrow V(t) = 25 \cos (8 \times 10^{-8} t + 10 \sin 1520 t)$

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Comparing with $s(t) = A_c [\cos(2\pi f_c t) + \beta \sin(2\pi f_m t)]$

$\therefore A_c = 15 \text{ V}$

(i) $f_c = 4 \times 10^{-8} \text{ Hz}$

(ii) $f_m = \frac{1520}{2} = 760 \text{ Hz}$

$\therefore \beta = 10$

(iii) Hence, $SF = \beta f_m = 10 \times 760 = 7600 \text{ Hz}$

Using Carson's rule,

(iv) $B.W = 2(\beta + 1) f_m$
= $2(1 + 10) \times 760$
= 16720 Hz

v) power dissipated across resistor, $P = \frac{A_c^2}{2R}$
= $\frac{15^2}{2 \times 10}$
 $\therefore P = 11.25 \text{ watt}$

N) Compute the maximum bit rate for a channel having bandwidth of 3.1 kHz and the SNR of 40 dB. Also calculate the no. of levels required to transmit at the maximum bit rate.

④ $2^{16}/s$

\Rightarrow channel capacity (C) = ?

Bandwidth (B) = 3.1 kHz

SNR = 40 dB

we have,

$C = B \log_2 (1 + \text{SNR})$

Also, $C = 0.322 \times B \times \text{SNR}$
= $0.322 \times 3.1 \times 40$

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$$\therefore C = 41.168 \text{ kbps}$$

Now,

No. of levels Required (L) = ?

We have,

Using Nyquist formula,

$$C = 2 \times B \times \log_2(L)$$

$$\text{or } 41.168 \times 1000 = 2 \times 3.1 \times 1000 \times \log_2(L)$$

$$\text{or } \log_2(L) = 6.64$$

$$\text{or } L = 2^{6.64} = 6.64 \text{ bits/baud. } \underline{\underline{27}}$$

$$\therefore L = 6.64 \approx 7$$

Hence, No. of levels required to transmit at maximum bit rate is ~~7~~ ≈ 7 .

Q. Verify that the transmission efficiency of single-tone modulating signal is $P_t = P_c (1 + \frac{M^2}{2})$ ~~2016/5~~

⇒ Consider a modulating wave $m(t)$ that consists of a single tone or frequency component, i.e.

$$m(t) = A_m \cos(2\pi f_m t) \quad \text{--- (1)}$$

where, A_m is peak amplitude of modulating wave.

f_m = frequency of modulating wave.

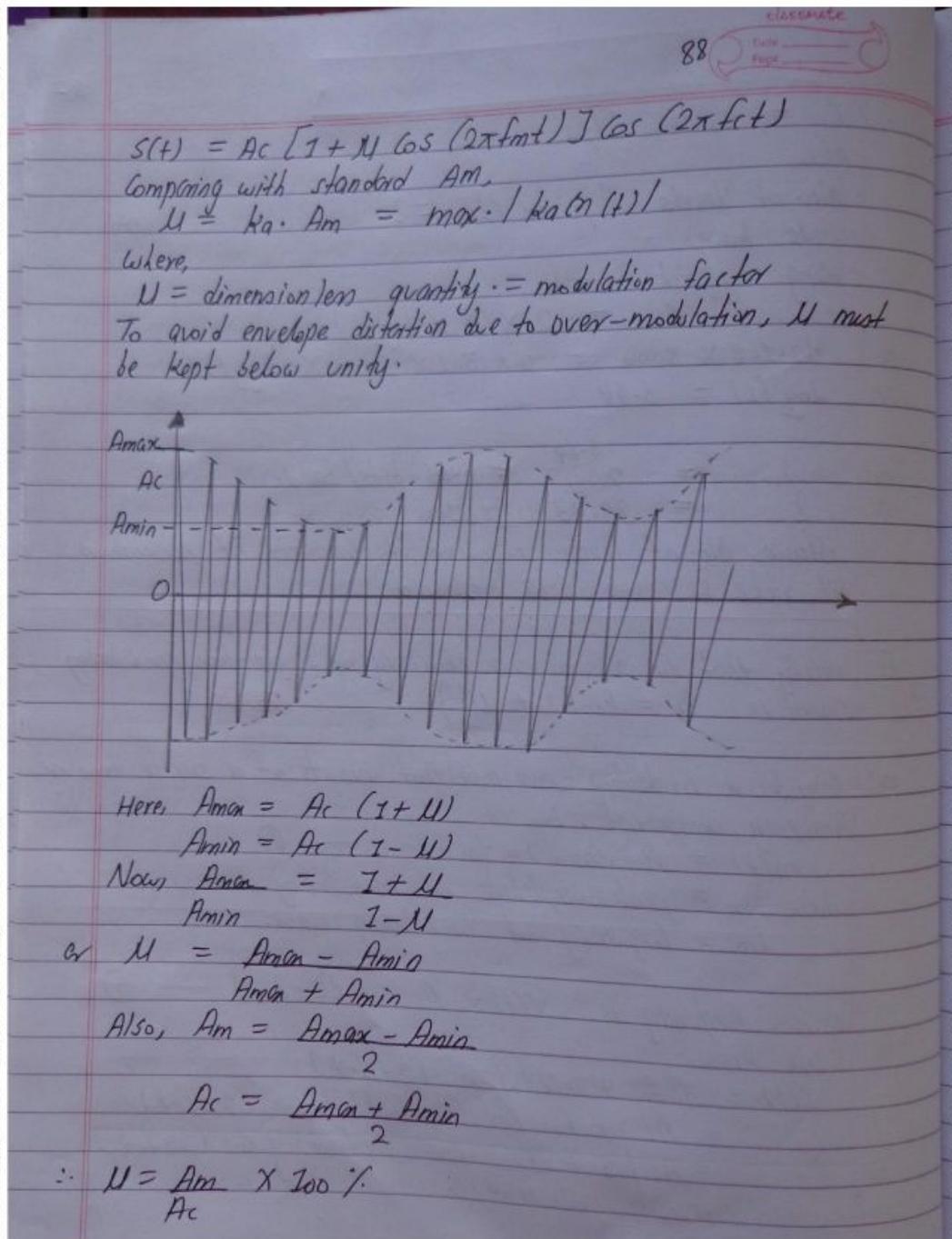
CARRIER frequency is, $c(t) = A_c \cos(2\pi f_c t) \quad \text{--- (2)}$

We know,

$$\begin{aligned} s(t) &= [A_c + m(t)] \cos(2\pi f_c t) \\ &= [A_c + A_m \cos(2\pi f_m t)] \cos(2\pi f_c t) \\ &= A_c \left[1 + \frac{A_m \cos(2\pi f_m t)}{A_c} \right] \cos(2\pi f_c t) \end{aligned}$$

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Again,
 $s(t) = A_c [1 + M \cos(2\pi f_m t)] \cos(2\pi f_c t)$

Hence,
carrier power, $P_c = \frac{1}{2} A_c^2$

upper side-frequency power, $P_{USB} = \frac{1}{8} M^2 A_c = \frac{1}{4} M^2 P_c$

lower " " " , $P_{LSB} = \frac{1}{8} M^2 A_c = \frac{1}{4} M^2 P_c$

Hence,
Total side band power, $P_{SB} = P_{USB} + P_{LSB} = \frac{1}{4} M^2 A_c = \frac{1}{2} M^2 P_c$

Total power, $P_t = P_c + P_{SB}$
 $= P_c + \frac{1}{2} M^2 P_c$

$\therefore P_t = P_c \left(1 + \frac{1}{2} M^2 \right)$

Q. Briefly explain the concept of exchangeability between bandwidth, signal power and SNR, then finally state shannon's eqn for channel capacity. *2014/5, 181F*

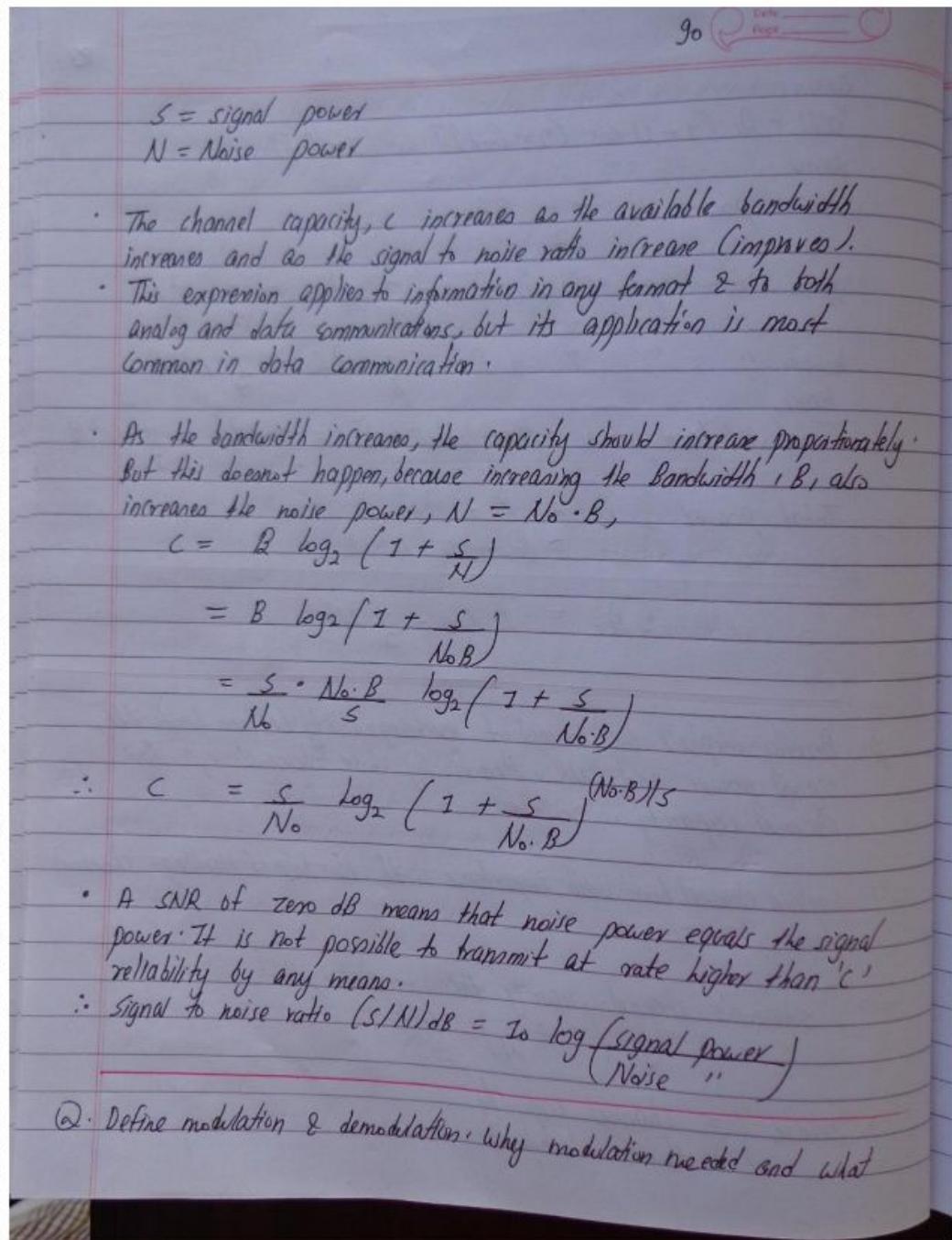
Ans: Explain channel bandwidth importance with the help of Hartley-Shannon law. *2015/5*

\Rightarrow Shannon's channel capacity theorem states that,
 $C = B \log_2 \left(1 + \frac{S}{N} \right)$ bits.

where, C = channel capacity
 B = channel bandwidth in Hz

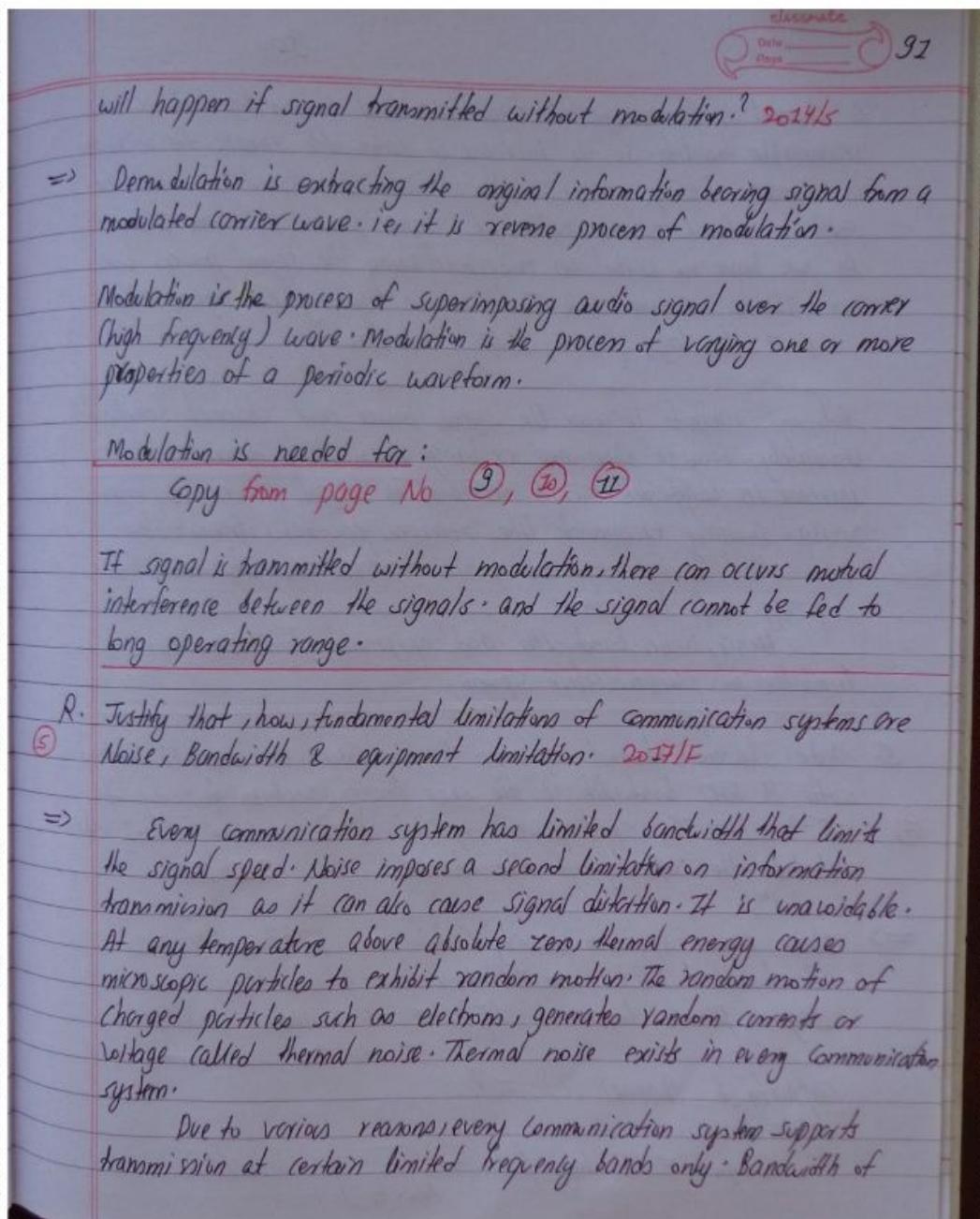
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a channel is the range of frequencies that it can transmit with reasonable fidelity. So, the increase in bandwidth means the increase in number of frequencies. So, the transmission speed also increases.

SNR is defined as the ratio of signal power to noise power. As we have no control on noise, increasing the signal power, we are able to increase SNR. That means reducing the signal effect.

$$C = B \log_2 \left(1 + \frac{S}{N} \right) \text{ bits/sec}$$

But, we cannot increase the signal power and channel bandwidth ultimately. Because they are exchangeable, which describes that increase in bandwidth causes decrease in signal power & vice-versa. Similarly, equipment like receiver, encoder, transmitter, and other devices used for communication have its own noise, frequency and limitations;

Hence, noise, bandwidth and equipment limitations are limitation of communication system.

5. State Shannon's channel capacity theorem. Given: On AWGN channel with 4 kHz bandwidth & the noise power spectrum density of 10^{-12} W/Hz . The signal power required at the receiver is 0.1 mW calculate the capacity of the channel. 2016/F

So, \therefore

\Rightarrow Noise power spectrum density (N_0) = $10^{-12} \text{ W/Hz} \times 2$
Bandwidth (B) = $4 \text{ kHz} = 4000 \text{ Hz}$.
Signal power transmit (\bar{P}) = 0.1 mW

Capacity of channel (C) = ? $= 0.001 \text{ W}$

We have,

$$C_{\text{AWGN}} = B \log \left(1 + \frac{\bar{P}}{N_0 \cdot B} \right) \text{ bits/sec.}$$

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Handwritten notes on Shannon's Channel Capacity Theorem:

Calculation of Channel Capacity:

$$SNR = \frac{P}{N_0 B}$$
$$C_{awgn} = 4000 \log_2 \left(1 + \frac{S}{N_0 B} \right)$$
$$= 4000 \log_2 \left(1 + \frac{4000}{2 \times 10^{-12} \times 4000} \right) = 4000 \log_2 (12500) = 4000 \times 13.6096$$
$$\therefore C_{awgn} = 54438.562 \text{ bps}$$
$$= 54.44 \text{ Kbps}$$

Q) Theoretical limitation of shannon's channel capacity Theorem:

- ① The noise in channel tends to zero, $SNR \rightarrow \infty$, subsequently, $C \rightarrow \infty$, it means that noiseless channel has infinite bandwidth. This kind of channel is called ideal channel.
- ② As the bandwidth of the channel tends to infinite, the channel capacity reaches an upper limit C_{max} . This is because, noise power is proportional to the bandwidth B as the bandwidth is increased, the noise power also increases correspondingly.
$$C = B \log_2 \left(1 + \frac{S}{N_0 B} \right)$$

N_0 = power spectrum density
 S = signal power
i.e.,
$$C = \left(\frac{S \cdot n}{n_0 S} \right) B \log_2 \left(1 + \frac{S}{n_0 B} \right)$$
$$= \frac{B}{n_0} \log_2 \left(1 + \frac{S}{n_0 B} \right)^{(n_0 B)/S}$$

Let, $\gamma = \frac{S}{n_0 B}$ & Considering limit,

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$\lim_{Y \rightarrow 0} (1+Y)^{1/Y} = e$		
$\lim_{S \rightarrow \infty} C = C_{\max} = \frac{S}{n} \log_2 e = 1.44 \left(\frac{S}{n} \right)$		
$\therefore C_{\max} = 1.44 \times \frac{S}{n}$		
Also,		
$\frac{C}{B} = \log_2 (1 + SNR)$		
$\Rightarrow SNR = 2^{(C/B)} - 1$		
<p>Q. Compares BASK, BFSK, BPSK for variables Bandwidth, bit rate and performance.</p> <p>2014/5, 2017/18</p>		
$\Rightarrow BASK$	BFSK	BPSK
i) Bandwidth = 2 fb	4 fb	2 fb
ii) suitable upto 200 bits/sec	upto about 1200 bits/sec	Bitrate is the no. of digital bits sent / second. It has 1 bit / symbol. High bit rate.
iii) BASK performance in presence of noise is poor.	BFSK is relatively simple, low performance form of digital modulation.	It is very robust transmission method but consumes significant bandwidth.
iv) Simple system	moderately complex system	very Complex system.
v) Noise immunity is Low	High	High.
vi) probability of error is high	Low	Low.

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v) Differentiate between coherent and non-coherent digital modulation techniques. Explain DPSK system. 2016/F.

Coherent digital modulation	Non-coherent digital modulation.
① Better performance on AWGN channel, slow flat fading.	Better performance in fast frequency selective fading.
② more complex than non-coherent demodulation & harder to implement.	Less complex than coherent detection and easier to implement but has worse performance.
③ expensive	Relatively less expensive.
iv). In coherent modulation technique process received signal with a local carrier of some frequency & phase.	Here, no requirement of reference wave.
v) It is those technique which employ coherent detection. In coherent detection, the local carrier generated at the receiver is phase locked with the carrier at the Transmitter. Thus, the detection is done by correlating received noisy signal & locally generated carrier. The coherent detection is also known as synchronous detection.	It is those technique in which the detection process doesn't need receiver carrier to be phase locked with transmitter ckt. But, the drawback of such system is that error probability increases.

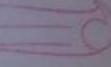
DPSK : 2015/S, 2016/F

A DPSK system may be viewed as the non-coherent version of

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of the PSK. It eliminates the need for coherent reference signal at the receiver by combining two basic operations at the transmitter.

- ① Differential encoding of the input binary wave &
- ② phase shift keying.

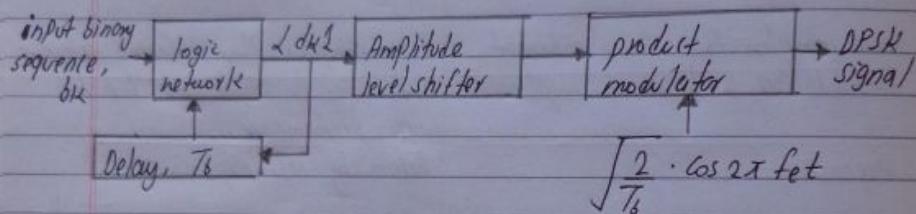


Fig: DPSK Transmitter

To send symbol '1' we phase advance the current signal waveform by 180° & to send symbol '0' we leave the phase of the current signal waveform unchanged.

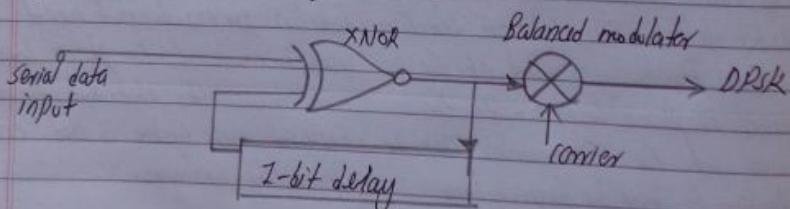


Fig: DPSK modulator.

The differential encoding process at the transmitter input starts with an arbitrary first bit serving as reference & thereafter the differentially encoded sequence d_{k+1} is generated by using the logical eqn,

$$d_k = d_{k-1} \oplus b_k$$

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w) Determine minimum Bandwidth for a 8PSK modulator with carrier frequency of 50 MHz & input bit rate of 500 kbps. 2017/18, 2014/15

(S) So/11, \Rightarrow minimum Bandwidth (B) = ?
carrier frequency (f_c) = 50 MHz
input bit rate (f_b) = 500 kbps
we have,
$$B = 2 f_a$$
$$= 2 \times \frac{f_b}{2}$$
$$= f_b$$
$$\therefore B = 500 \text{ kbps}$$

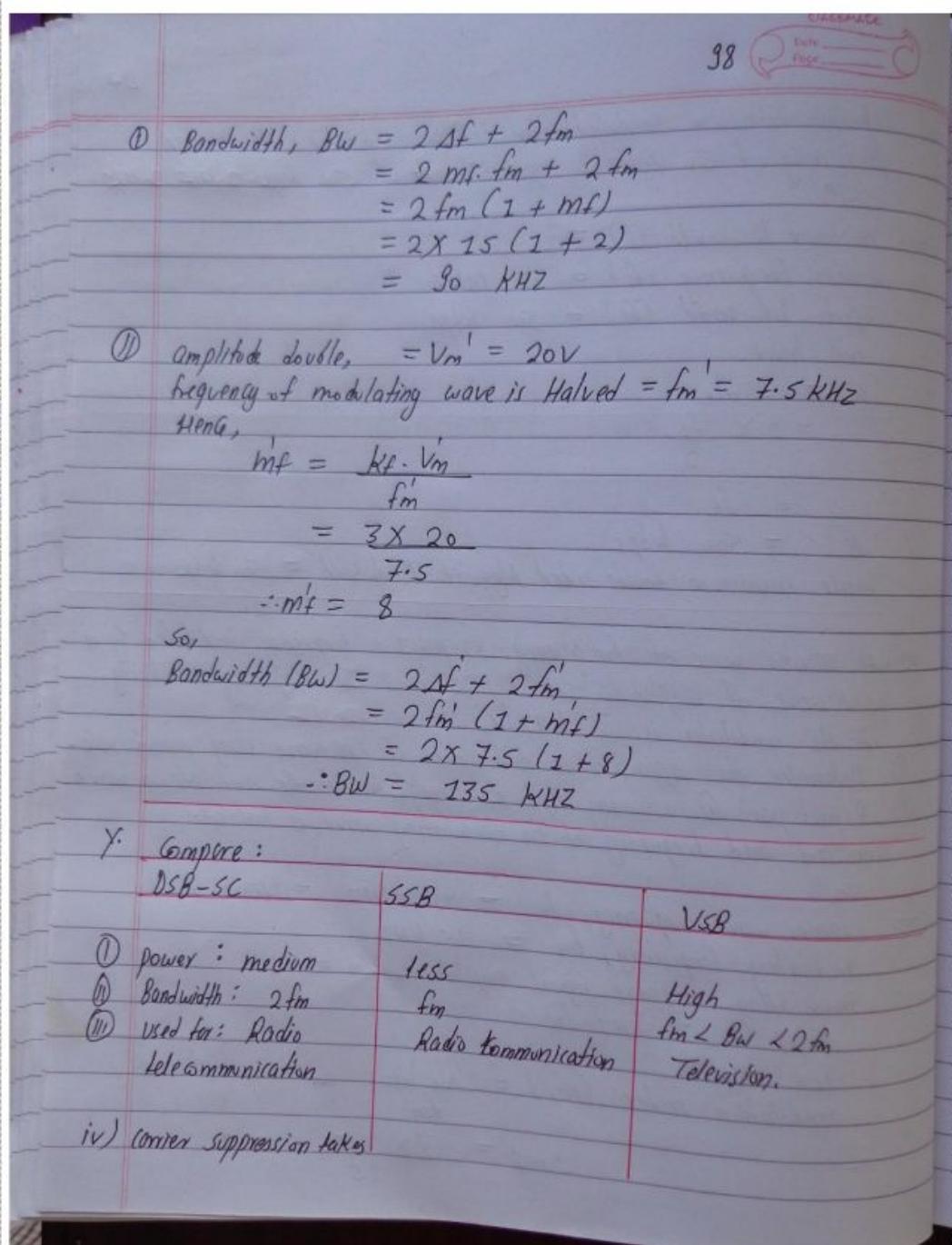
Hence, minimum double sided Nyquist Bandwidth = 500 kbps.

x) A carrier wave of frequency 91 MHz is frequency modulated by a sine wave of amplitude 10V & 15 kHz. The frequency sensitivity of the modulator is 3 kHz/V. 2016/17

(i) Determine the approximate bandwidth of FM wave using Carson's rule.
(ii) Repeat part (i), assuming that the amplitude of the modulation wave doubled and frequency of the modulating wave is halved.
So/11, \Rightarrow carrier wave frequency, $f_c = 91 \text{ MHz} = 91000 \text{ kHz}$
modulating frequency, $f_m = 15 \text{ kHz}$
modulating message signal amplitude, $V_m = 10 \text{ V}$
 $\Delta f = 3 \text{ kHz/V}$
we know,
modulation index (m_f) = $\frac{k_f V_m}{f_m}$
 $\therefore m_f = \frac{3 \times 10}{15} = 2$

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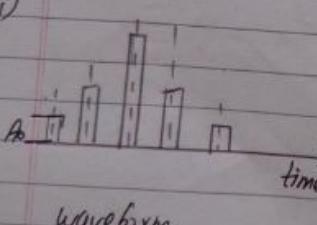
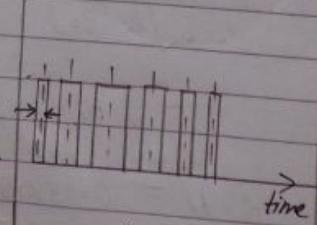
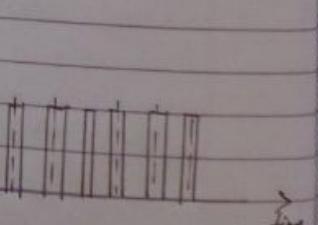
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place completely	- Completely	No suppression.		
v) transmission efficiency: moderate	maximum	moderate.		
vi. sideband suppression: - No	one sideband completely one sideband partially			
vii more complex to modulate & demodulate.	more difficult to modulate & demodulate	easier to implement.		
<hr/>				
2) The rms value of carrier voltage is 80 V. After amplitude modulation by a sinusoidal audio signal, the rms value of the carrier voltage becomes 96 Volts, determine modulation factor. <i>Soln,</i> => carrier voltage (V_c) = 80 V Total modulated voltage (V_T) = 96 V modulation index (M) = $\sqrt{[(\frac{V_T}{V_c})^2 - 1]} \times 2$				
$= \sqrt{[(\frac{96}{80})^2 - 1]} \times 2$ $= \sqrt{0.44} \times 2$				
$\therefore M = 0.93808$ $\% M = 93.808 \%$				

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2) A. Compare:	pulse amplitude modulation (PAM)	pulse width modulation PWM / PDM	pulse position modulation, PPM
i)	Amplitude of the pulse is proportional to amplitude of modulating signal.	width of the pulse is proportional to amplitude of modulating signal.	The relative position of the pulse is proportional to the amplitude of modulating signal.
ii)	Bandwidth of the transmission channel depends on width of the pulse.	- depends upon the rise time of the pulse.	- depends on rising time of the pulse.
iii)	Noise, interference is high. Noise, interference is complex system.	noise, interference is minimum.	minimum.
iv)	Instantaneous power of transmitter varies.	Instantaneous power of transmitter varies.	- Remains constant.
v)	Similar to amplitude modulation.	similar to frequency modulation.	similar to phase modulation.
vi)			

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b) Signal to Quantization Noise Ratio for linear quantization:

We know, that in a PCM system for linear quantization the signal to noise ratio is given as,

$\therefore \frac{S}{N} = \frac{\text{normalized signal power}}{\text{Normalized Noise}}$

But, normalized noise power is $= \frac{1^2}{12}$

Hence,

$\therefore \frac{S}{N} = \frac{\text{normalized signal power}}{(1^2/12)}$

We know that the number of bits 'v' and quantization levels are,

$q = 2^v$.

Also,

Step size, $\Delta = \frac{2 \cdot X_{\max}}{2^v}$

$\therefore \Delta = \frac{2 \cdot X_{\max}}{2^v}$

$\therefore \frac{S}{N} = \frac{\text{normalized signal power}}{\left(\frac{2 \cdot X_{\max}}{2^v}\right)^2 \cdot \frac{1}{12}}$

Let normalized signal power be denoted as ' P ',

then, $\frac{S}{N} = \frac{P}{\frac{4 \cdot X_{\max}^2 \cdot 1}{2^{2v}} \cdot \frac{1}{12}} = \frac{3P}{X_{\max}^2} \times 2^{2v}$

Hence, signal to quantization noise ratio is,

$\therefore \frac{S}{N} = \left(\frac{3P}{X_{\max}^2}\right) \cdot 2^{2v}$

This expression shows that signal to noise power ratio of quantizer increases exponentially with increasing bits per sample.

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let, input $x(t)$ is normalized, i.e.,
 $X_{max} = 1$,
Signal to quantization noise ratio will be,
 $\therefore \frac{S}{N} = 3 \times P \times 2^{2V}$

Also if destination signal power 'P' is normalized, i.e.,
 $P \leq 1$,
Then, signal to noise ratio will be,
 $\frac{S}{N} \leq 3 \times 2^{2V}$,

$\therefore \left(\frac{S}{N} \right) dB = 10 \log_{10} \left(\frac{S}{N} \right) dB \leq 10 \log_{10} [3 \times 2^{2V}] \leq (4.8 + 6V) dB$

c) A compact disc (CD) records audio signals digitally by P.M.-P.A.M.
audio signal's bandwidth to be 15 kHz, if signals are sampled at a rate of 20% above Nyquist rate for practical reasons & the samples are quantized into 65,536 levels, determine bits/sec required to encode the signal and minimum bandwidth required to transmit encoded signal.

soln,

\Rightarrow Given, $f_m = 15 \text{ kHz}$,
 $f_s = 1.2 \times 2 f_m = 2.4 \times 15 \text{ kHz} = 36 \text{ kHz}$
 $q = 65,536$

Signaling rate (r):

$$q = 2^v$$
$$v = \log_2 q$$
$$v = \frac{\log_{10} (65536)}{\log_{10} 2} = 16$$

Now, signaling rate, $r = v \cdot f_s = 16 \times 36 \text{ kHz}$

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= 576 kbit/sec

Q) minimum bandwidth,
 $B.W = \frac{1}{2} (\text{signaling rate}) = \frac{576}{2} = 288 \text{ kHz}$.

D) An analog waveform with bandwidth 15 kHz is to quantized with 200 levels and transmitted via binary PAM signal. Find rate of transmission and bandwidth required. If 10 such signals are to be multiplexed, find the bandwidth required.

Sol:

$\Rightarrow f_m = 15 \text{ kHz}$

$Q = 200$

$f_s = 2 f_m = 2 \times 15 \times 10^3 = 30 \times 10^3 \text{ Hz}$

$Q = 2^N$

No. of bits / sample, $N = \log_2 Q$
 $= \frac{\log_{10} 200}{\log_{10} 2}$
 $\therefore N = 8$

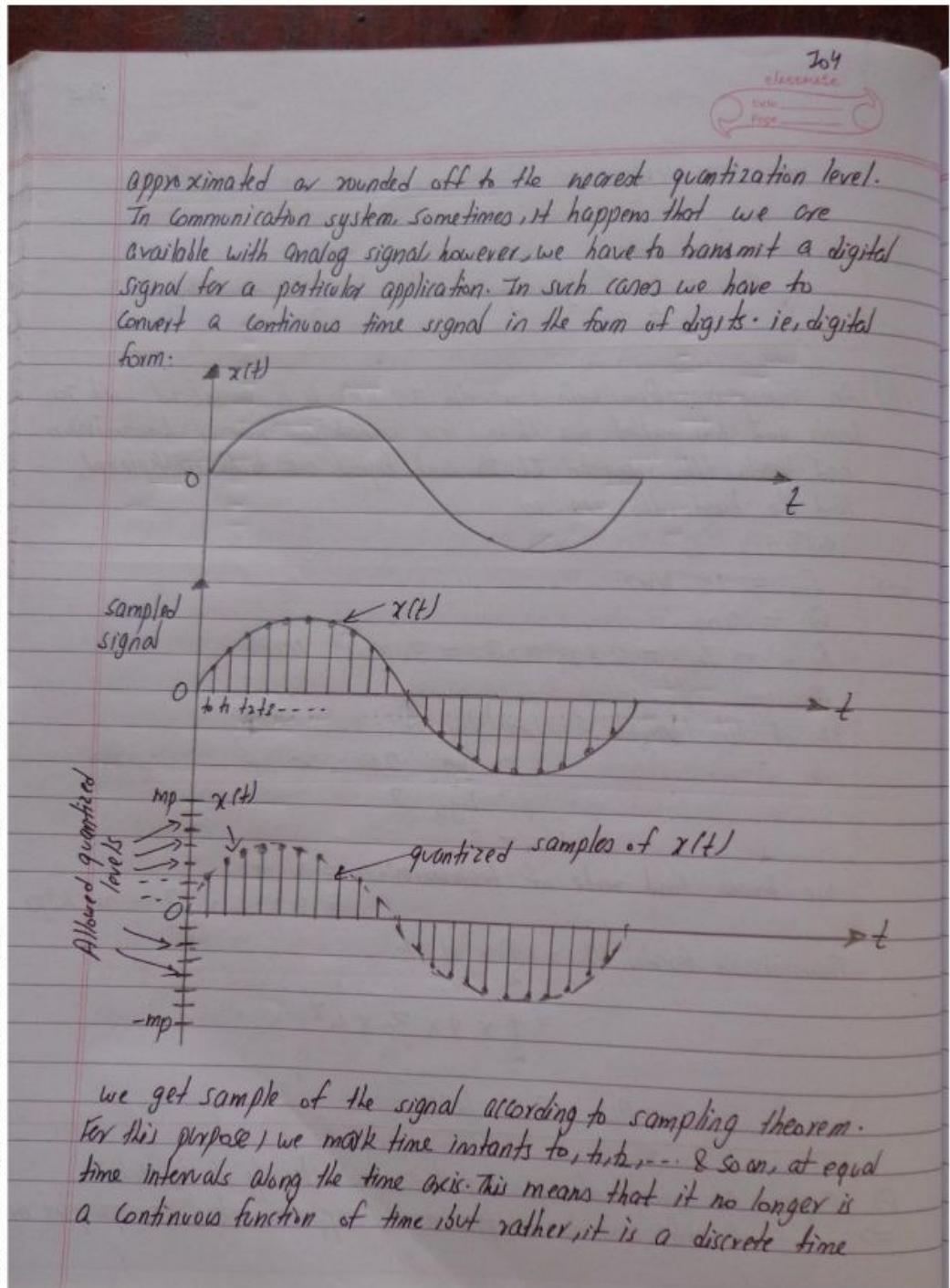
We know that rate of transmission = $N \cdot f_s$
 $= 8 \times 30 \times 10^3 = 240 \text{ kbps}$

Transmission Bandwidth = $\frac{1}{2} \times N \cdot f_s$
 $= \frac{1}{2} \times 8 \times 30 \times 10^3$
 $\therefore B.W = 120 \text{ kHz}$

E) Explain the concept of quantization and classify it in details.
Quantization is a process of approximation. The samples are

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A photograph of handwritten notes on a lined notebook page. The page has a red header 'Elementary' and a red footer 'Date _____'. The date '205' is written in the top right corner. The main text discusses signal quantization, mentioning that while the signal is sampled, it remains analog due to the continuous range of values. It then describes the quantization process where the total amplitude range is divided into L standard levels. The amplitude of the signal $x(t)$ is within the range $(-m_p, m_p)$, partitioned into L intervals of magnitude $\Delta V = \frac{2m_p}{L}$. Each sample is approximated to the nearest quantized level, resulting in digitized information.

Classification of Quantization process :

Quantization :

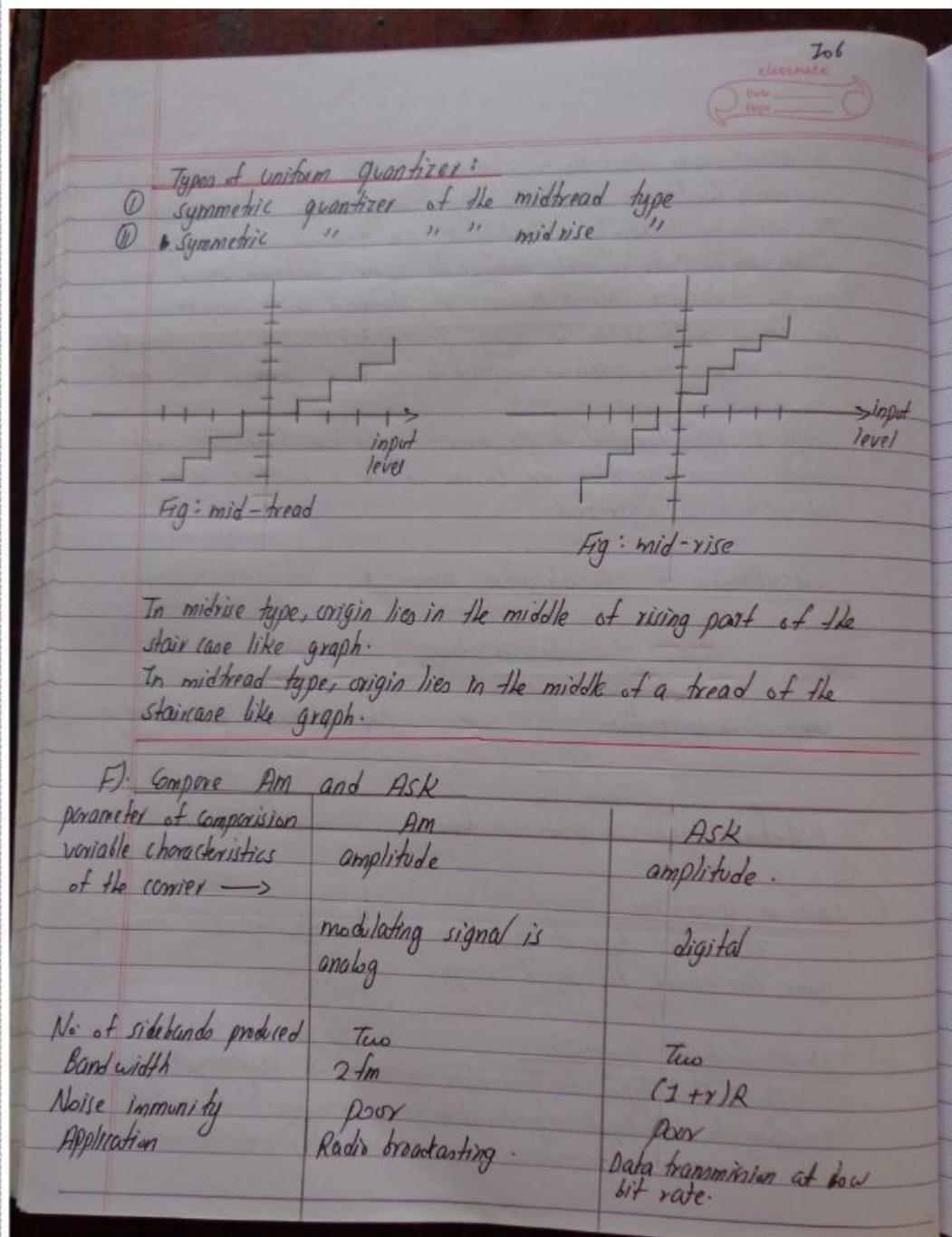
```
graph TD; A[Quantization] --> B[Uniform quantization]; A --> C[Non-uniform quantization]; B --> D[Mid-tread type]; B --> E[Mid-rise type]
```

① Uniform quantizer :
It is that type of quantizer in which the step size remains same throughout the input range.

② Non-uniform Quantizer :
It is that type of quantizer in which the step size varies according to the input signal values.

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G. Compare :	
BPSK	DPSK
i) variable characteristics is phase.	phase.
ii) Bandwidth : f_b	f_b .
iii) probability of error : low	Higher than BPSK
iv) Noise effect is low	" " "
v) Demodulation technique : synchronous	synchronous.
vi) system complexity : lower than DPSK	Higher than BPSK.
vii) Requirement of synchronous carrier is required.	Not required.
H) BPSK	
i) variable characteristics is phase	OPSK
ii) Types of modulation : Two level (binary)	phase
iii) Complexity : complex	Four level
iv) Detection method : Coherent	very complex.
v) Bit rate/baud rate \rightarrow Bit rate = baud rate	Coherent.
vi) Types of representation: A binary bit is represented by one phase state.	Bit rate = $2 \times$ baud rate. A group of two binary bits is represented by one phase state.

I) write short notes on : Regenerative Repeater :

Regenerative repeaters are used at regularly spaced intervals along a digital transmission line to detect the incoming digital signal & regenerate new clean pulses for further transmission along the line.

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This process periodically eliminates and thereby combats, the accumulation of noise and signal distortion along the transmission path. If the pulses are transmitted at a rate of R_b pulses per second, we require the periodic timing information, - the clock signal at R_b Hz - to sample the incoming pulses at a repeater. This timing information can be extracted from the received signal itself if the line code is chosen properly.

In digital systems, the 'repeater' detects and regenerates a clean signal (noise free) along the transmission channel. If an error occurs in the detector, this error is propagated forward. Regenerative repeaters are used on wired line, fiber & wireless transmission systems. It consists of demodulators / detectors and transmitters.

Advantages:

- improvement in overall error rate performance.
- Ease of amplification at Baseband.

J) A TV signal having a Bandwidth of 4.2 MHz transmitted using PAM system. Given that no. of quantization level is 512, find,

- I) Code word length
- II) Transmission bandwidth
- III) Final bit rate
- IV) Output signal to quantization noise ratio.

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Given, Bandwidth = $4.2 \text{ MHz} = 4.2 \times 10^6 \text{ Hz}$
No. of quantization level, $L = 512$,

- ① Code word length (b) = ?
we know,

$$L = 2^b$$
$$\therefore b = \log_2(L) = \log_2(512) = 9 \text{ length.}$$

② Bandwidth, $\geq b \cdot f_m = 9 \times 4.2 \times 10^6 \geq 37.8 \text{ MHz}$

③ Final bit rate = $r = b \cdot f_m = 9 \times 8.0$
 $= 2 \times b \cdot f_m$
 $= 37.8 \times 2$
 $= 75.6 \times 10^6 \text{ bps}$

④ SQNR $\leq (4.8 + 6.6) \leq 4.8 + 6 \times 9 \leq 58.8 \text{ dB}$

k. Various PAM Formats for line codes:

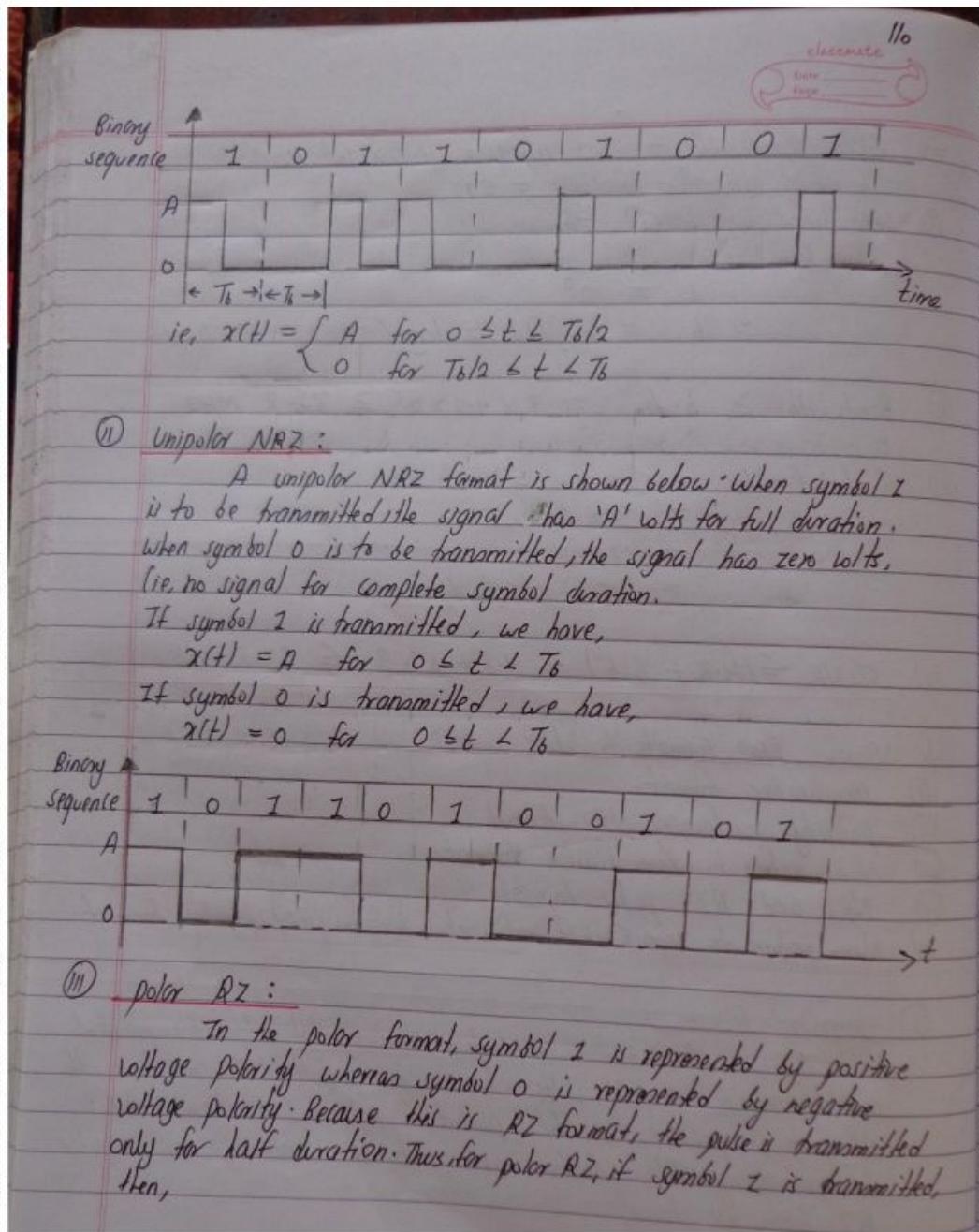
- ① Manchester format
- ② Polarity quaternary NRZ format
- ③ Non-return to zero bipolar format.
- ④ NRZ and RZ polar format.
- ⑤ Non-return to zero (NRZ) and return to zero (RZ) unipolar format.

① Unipolar RZ :

The waveform has zero value when symbol '0' is transmitted and waveform has ' A ' volts when '1' is transmitted. In RZ form, the ' A ' volts is present for $T_{b/2}$ period if symbol '1' is transmitted & for remaining $T_{b/2}$ waveform returns to zero value, i.e., for unipolar RZ form, we have,

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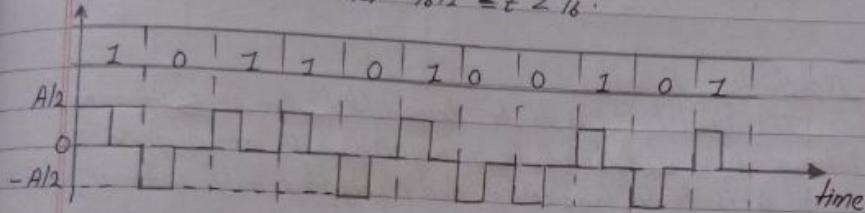
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$$x(t) = \begin{cases} A/2 & \text{for } 0 \leq t < T_{b/2} \\ 0 & \text{for } T_{b/2} \leq t < T_b \end{cases}$$

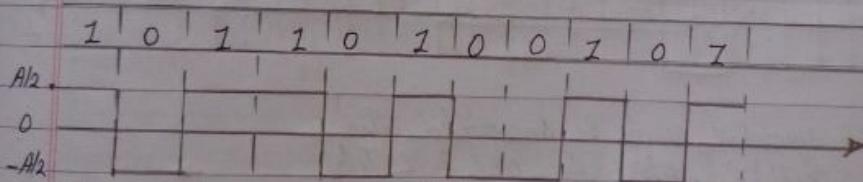
If symbol 0 is transmitted, then,

$$x(t) = \begin{cases} -A/2 & \text{for } 0 \leq t < T_{b/2} \\ 0 & \text{for } T_{b/2} \leq t < T_b \end{cases}$$



iv) Polar NRZ :

In polar NRZ format, symbol 1 is represented by positive polarity whereas symbol 0 is represented by negative polarity. These polarities are maintained over the complete pulse duration.



If symbol 1 is transmitted, then,

$$x(t) = +A/2 \text{ for } 0 \leq t < T_b$$

If symbol 0 is transmitted, then,

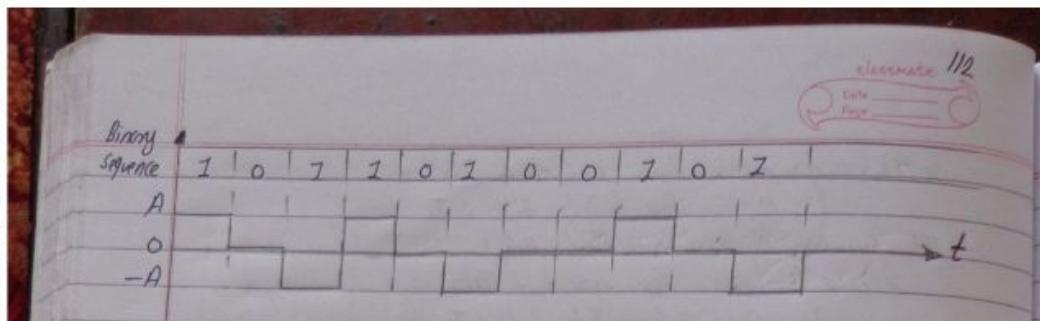
$$x(t) = -A/2 \text{ for } 0 \leq t < T_b.$$

v) Bipolar NRZ [Alternate mark inversion] AMI :

In this format, the successive 1's are represented by pulses with alternate polarity and 0's are represented by no pulses.

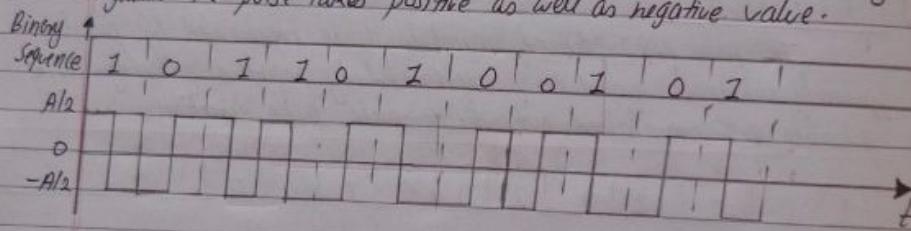
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v) Manchester Format:

If symbol '1' is to be transmitted, then a positive half interval pulse is followed by a negative half interval pulse. If symbol 0 is to be transmitted, then a negative half interval pulse is followed by a positive half interval pulse. Hence, for any symbol the pulse takes positive as well as negative value.



If symbol 1 is to be transmitted, then,

$$x(t) = \begin{cases} A/2 & \text{for } 0 \leq t < T_b/2 \\ -A/2 & \text{for } T_b/2 \leq t < T_b \end{cases}$$

If symbol 0 is to be transmitted, then,

$$x(t) = \begin{cases} -A/2 & \text{for } 0 \leq t < T_b/2 \\ A/2 & \text{for } T_b/2 \leq t < T_b \end{cases}$$

i) Draw the following data formats for the bit stream 1100110

ii) polar NRZ

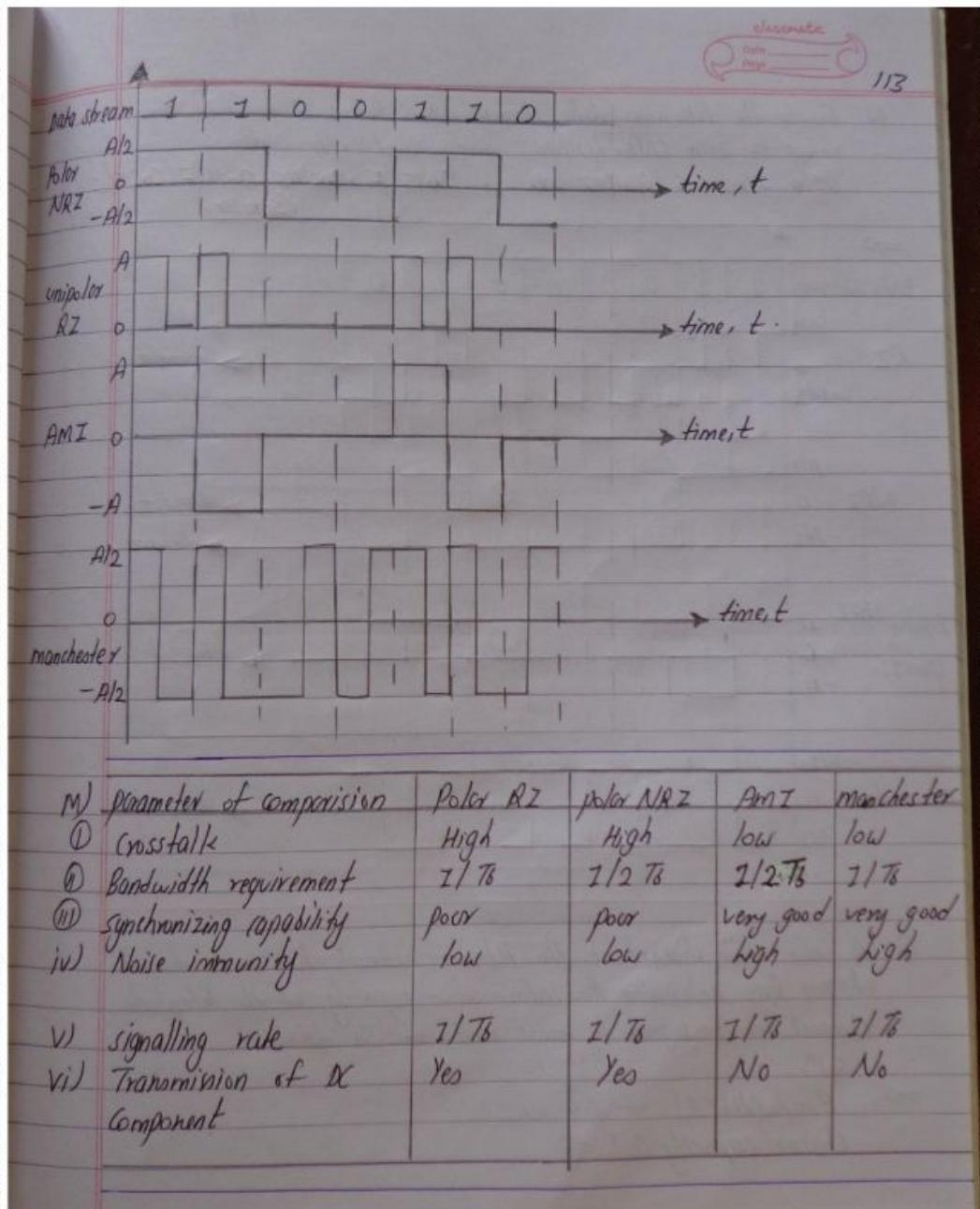
iv) Manchester

iii) Unipolar NRZ

iii) AMI

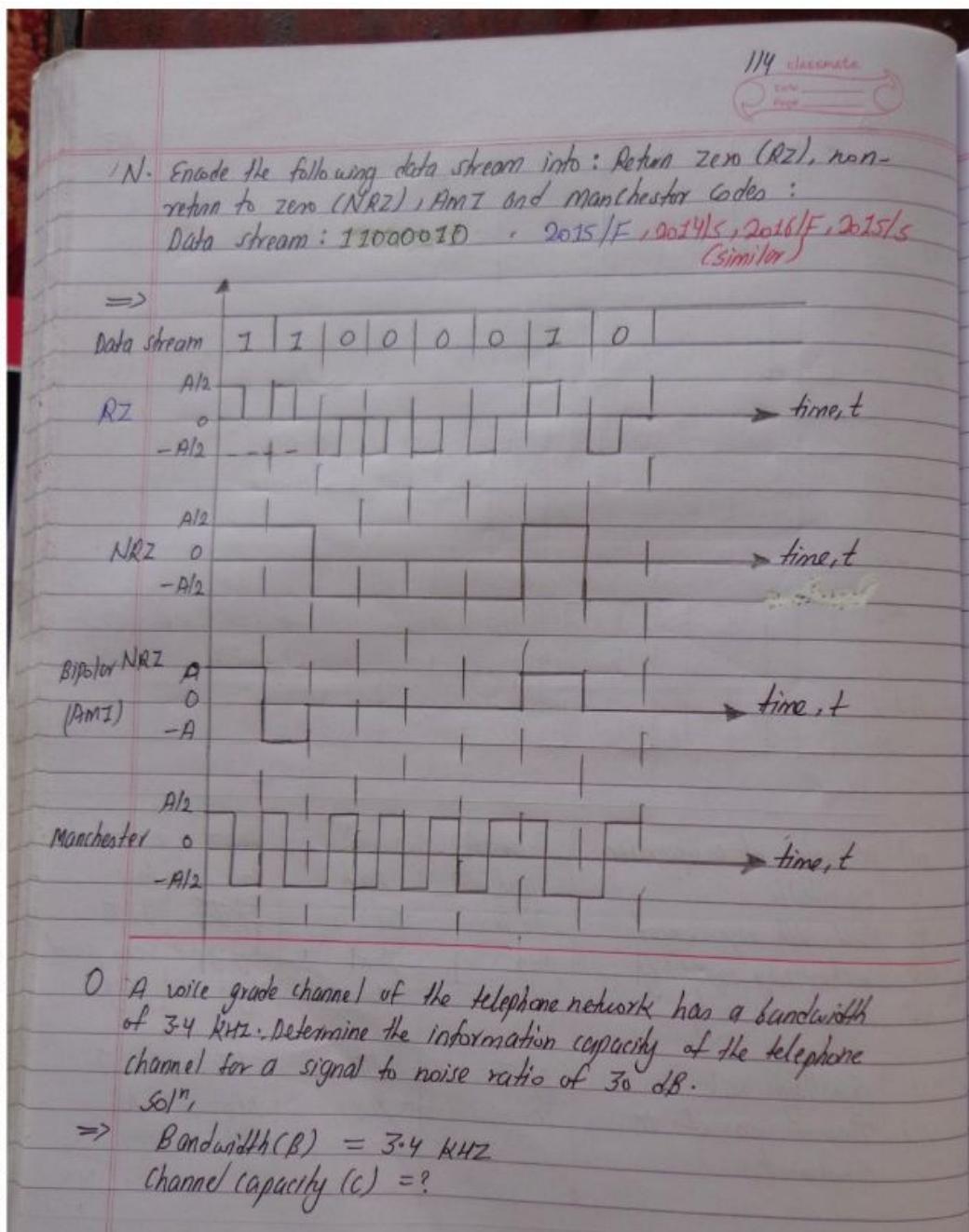
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O A voice grade channel of the telephone network has a bandwidth of 3.4 kHz. Determine the information capacity of the telephone channel for a signal to noise ratio of 30 dB.

Soln,

$$\Rightarrow \text{Bandwidth}(B) = 3.4 \text{ kHz}$$
$$\text{channel capacity } (C) = ?$$

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or $\left[\frac{S}{N} \right]_{dB} = 10 \log_{10} \left[\frac{S}{N} \right]$

or $30 = 10 \log_{10} \left[\frac{S}{N} \right]$

or $\left[\frac{S}{N} \right] = \text{antilog } (3) = 1000$
Now,
 $C = B \log_2 \left[1 + \frac{S}{N} \right]$
 $= 3.4 \times 10^3 \log_2 [1 + 1000]$
 $= 33888.58 \text{ bits/sec}$
 $\therefore C = 33.89 \text{ kbps}$

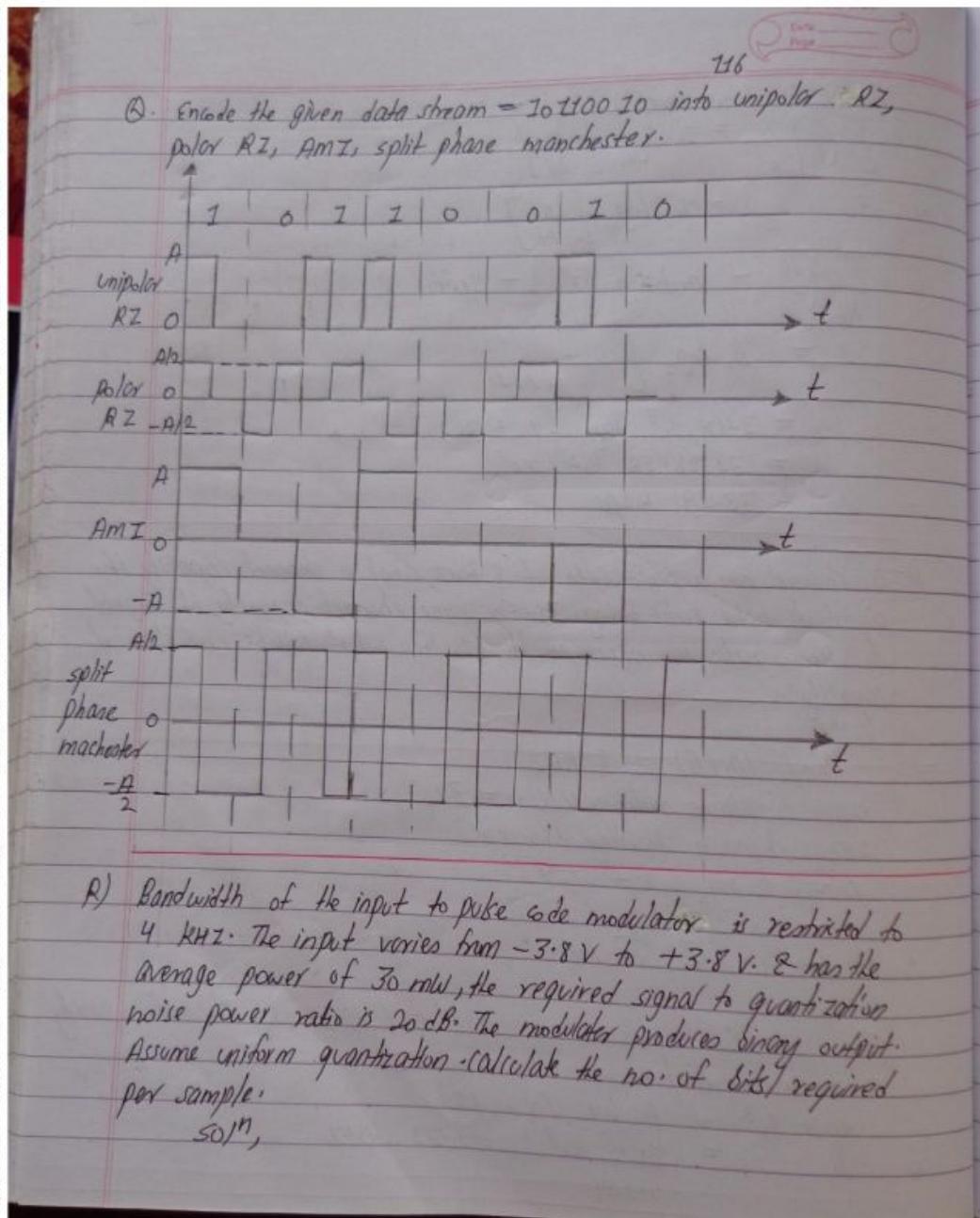
P: A channel has a bandwidth of 8 kHz, what is channel capacity if signal to noise ratio being 31. For same channel capacity, if signal to noise ratio is increased to 61, then, what will be new channel capacity?
Soln:
 $\Rightarrow \text{Bandwidth } (B) = 8 \text{ kHz}$
 $\text{signal to Noise ratio } (S/N) = 31$
Using shannon Hartley theorem,
 $C = B \log_2 \left[1 + S/N \right]$
 $= 8 \times 10^3 \log_2 [1 + 31]$
 $\therefore C = 40 \times 10^3 \text{ bits/sec}$

The new Bandwidth when $S/N = 61$, assuming channel capacity C to be constant.

or $B = \frac{40 \times 10^3}{5.954} = 6.71 \text{ kHz}$

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R) Bandwidth of the input to pulse code modulator is restricted to 4 kHz. The input varies from -3.8 V to $+3.8\text{ V}$. It has the average power of 30 mW , the required signal to quantization noise power ratio is 20 dB . The modulator produces binary output. Assume uniform quantization. calculate the no. of bits required per sample.
Soln,

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Soln:

$\Rightarrow \text{Given, } \left(\frac{s}{N_q}\right)_{dB} = 20 \text{ dB}$

or $I_0 \log \left(\frac{s}{N_q}\right)_0 = 20 \text{ dB}$

or $\left(\frac{s}{N_q}\right)_0 = \text{antilog} \left(\frac{20}{I_0}\right) = I_0^2$

$\therefore (s/N_q)_0 = 100$

Quantization step size, $A = \frac{2A}{L}$
where,

$L = 2^n$, n = number of binary digits,

Average quantizing power:

$N_q = (q_e)^2 = \frac{A^2}{I_0^2} = \frac{A^2}{3I^2}$

or $\left(\frac{s}{N_q}\right)_0 = \frac{\text{Average signal power}}{\text{Average quantizing power}}$

or $100 = \frac{30 \times 10^{-3}}{A^2 / 3I^2}$

or $L = \sqrt{\frac{30 \times 10^{-3}}{3 \times 100 \times 3.8^2}} = 126.67$

or $2^n = 128 \Rightarrow 2^n = 2^7$

$\therefore n = 7$

Hence, No. of bits required/sample = 7.

5) Derive an expression for signal to quantization noise ratio for a PCM system which employs linear (i.e. uniform) quantization technique.
Given that input to the PCM system is a sinusoidal signal.
 \Rightarrow Let us assume that the modulating signal is a sinusoidal signal

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(voltage), having peak amplitude equal to A_m . Also, let the signal cover the complete excursion of representation levels. Then, power of this signal is,

$$P = \frac{V^2}{R}$$

$$V = \text{rms value, i.e. } V = \left[\frac{A_m}{\sqrt{2}} \right]^2$$

$$\text{we have, } P = \frac{A_m^2}{2} \cdot \frac{1}{R}$$

In case, when $R=1$, the power P is normalized, i.e.,
Normalized power, $P = \frac{A_m^2}{2}$ with $R=1$,

we know, signal to quantization noise ratio is,

$$\frac{S}{N} = \frac{3P}{x_{\max}^2} \times 2^{2V}$$

$$\text{or, } P = \frac{A_m^2}{2}$$

$$\text{and, } x_{\max} = A_m$$

So,

$$\frac{S}{N} = 3 \times \frac{A_m^2/2}{A_m^2} \times 2^{2V}$$

$$\text{or } \frac{S}{N} = \frac{3}{2} \times 2^{2V}$$

$$\text{or, } \frac{S}{N} = 1.5 \times 2^{2V}$$

Expressing in dB,

$$\left(\frac{S}{N} \right) \text{dB} = 10 \log_{10} \left(\frac{S}{N} \right) = 10 \log_{10} (1.5 \times 2^{2V})$$

$$= 10 \log_{10} (1.5) + 10 \log_{10} (2^{2V})$$

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$\alpha \left(\frac{S}{N}\right) dB = 2.76 + 2V \times 20 \times 0.3$ Date _____
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Here,
 $\therefore \left(\frac{S}{N}\right) dB \text{ in } \mu\text{m} = \left(\frac{S}{N}\right) dB = 2.8 + 6V \text{ for sinusoidal signal.}$

T) The signal having bandwidth equal to 3.5 kHz is sampled, quantized & coded by a PCM system. The coded signal is then transmitted over a transmission channel of supporting a transmission rate of 50 kbps. Find the maximum signal to noise ratio that can be obtained by this system. The input signal has peak to peak value of 4 volts & rms voltage value of 0.2 V.

So,

\Rightarrow maximum frequency of the signal, $f_m = 3.5 \text{ kHz}$
sampling " ", $f_s \geq 2f_m \geq 2 \times 3.5 \geq 7 \text{ kHz}$
we know, signaling rate is,
 $r \geq V \cdot f_s$

$50 \times 10^3 \geq r \times 7 \times 10^3$

$\therefore V \leq 7.142 \text{ bits} \cong 8 \text{ bits.}$

The rms value of signal is 0.2 V, Normalized signal power is,
Normalized signal power, $P = \frac{(0.2)^2}{2}$

i.e., $P = 0.04 \text{ W}$

maximum signal to noise ratio is,

$$\frac{S}{N} = \frac{3P \cdot 2^{2V}}{2^{2mca}}$$

$\alpha \left(\frac{S}{N}\right) = 3 \times 0.04 \times \frac{2^{2 \times 8}}{4} = 1996.08 \cong 33 \text{ dB}$

Here, $\left(\frac{S}{N}\right)_{max} = 33 \text{ dB.}$

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v) Probability of Error:

It defines average probability of error that can occur in a communication system. Let, P_e be the probability of occurrence of '0' and $P(1)$ be the probability of occurrence of '1'. Then, the error is of two types.

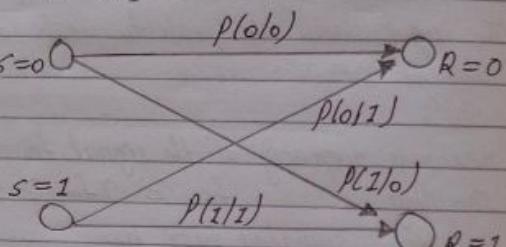
$P(1|0)$: i.e., transmitted as '1' but received as '0'

$P(0|1)$: i.e., " " "0" " " "1"

The total probability of error, P_e is,

$$P_e = P(1|0) \cdot P(0) + P(0|1) \cdot P(1)$$

It is measured on the basis of following functions.



Error function:

$$\text{erf}(v) = \frac{2}{\sqrt{\pi}} \int_0^v \exp(-z^2) dz$$

Complementary error function,

$$\text{erfc}(v) = \frac{2}{\sqrt{\pi}} \int_v^\infty \exp(-z^2) dz$$

Q-function, $\text{Q}(v) = \frac{1}{\sqrt{2\pi}} \int_v^\infty \exp(-x^2/2) dx$

- v) Binary data is transmitted over a microwave link at a rate of 10^6 bits/sec and the PSD of noise at the receiver input is

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10^{-10} watts/Hz. Find the average carrier power required to maintain an average probability of error $P_e \leq 10^{-4}$ for coherent binary FSK. What will be the required channel Bandwidth?

Soln,

\Rightarrow Bit rate = 10^6 bits/sec
 $\frac{N_0}{2} = 10^{-10}$ w/Hz

or $N_0 = 2 \times 10^{-10}$ w/Hz

Also,
 $P_e \leq 10^{-4}$,
 $P_s = ?$

we know, error probability of FSK with coherent detection is,

$$P_e = \frac{1}{2} \operatorname{erfc} \left[\frac{0.6 E_b}{N_0} \right]^{1/2}$$

But, $E_b = P_s T_b$

$$\text{or } P_e = \frac{1}{2} \operatorname{erfc} \left[0.6 P_s \frac{T_b}{N_0} \right]^{1/2}$$

But, $T_b = \frac{1}{\text{bitrate}}$

$$\text{or } 10^{-4} = \frac{1}{2} \operatorname{erfc} \left[\frac{0.6 P_s}{2 \times 10^{-10} \times 10^6} \right]^{1/2}$$
$$\text{or } 2 \times 10^{-4} = \operatorname{erfc} [3000 P_s]^{0.5}$$

But,

$$1 - \operatorname{erfc}(W) = \operatorname{erf}(W)$$
$$\text{or } 1 - 2 \times 10^{-4} = 1 - \operatorname{erfc} [3000 P_s]^{1/2}$$
$$\text{or } 0.9988 = \operatorname{erf} [3000 P_s]^{0.5}$$

we have,

$$0.9988 = \operatorname{erf} [2.8]$$

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or $3000 P_s = 2.8$

Hence,
 $\therefore P_s = \frac{2.8}{3000} = 0.933 \text{ mW}$

w) Binary data is transmitted at a rate of 10^6 bits/second over a channel having a bandwidth of 3 MHz. Assume that the noise PSD at the receiver is $N_0/2 = 10^{-20} \text{ watt/Hz}$. Find the average carrier power required at the receiver input for coherent PSK and DSK signalling schemes to maintain a probability of error $P_e = 10^{-4}$.

so/17,

\Rightarrow ① PSK system.

We know that the average error probability of FSK system,

$$P_e = \frac{1}{2} \operatorname{erfc} \sqrt{\frac{E}{N_0}}$$

But, $P_e = 10^{-4}$ (Given)

Hence,

$$10^{-4} = \frac{1}{2} \operatorname{erfc} \sqrt{\frac{E}{N_0}}$$

or $2 \times 10^{-4} = \frac{1}{2} \operatorname{erfc} \sqrt{\frac{E}{N_0}}$

or, $1 - 2 \times 10^{-4} = 1 - \operatorname{erfc} \sqrt{\frac{E}{N_0}} \quad [\because \operatorname{erf}(v) = 1 - \operatorname{erfc}(v)]$

or, $0.9998 = \operatorname{erfc} \sqrt{\frac{E}{N_0}}$

We have, $\operatorname{erf}(2.6) \approx 0.9998$

Thus, $\sqrt{\frac{E}{N_0}} = 2.6$

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W $\frac{E}{N_0} = 6.76$

Therefore, $E = 6.76 N_0$

But, $N_0 = 2 \times 10^{-10}$

Therefore, $E = 6.76 \times 2 \times 10^{-10}$
= $1.352 \times 10^{-9} \text{ J}$

But, $E = PT$

and, $T = \frac{1}{\text{bit rate}} = \frac{1}{10^6}$

Hence, $P = \frac{E}{T} = 1.352 \times 10^{-9} \times 10^6 = 1.352 \text{ mW}$.

① DSK system:
 $P_e = \frac{1}{2} e^{-(E_b/N_0)}$

or $10^{-4} = \frac{1}{2} e^{-(E_b/N_0)}$

Thus,
 $\frac{-N_0}{N_0} = -8.5171$

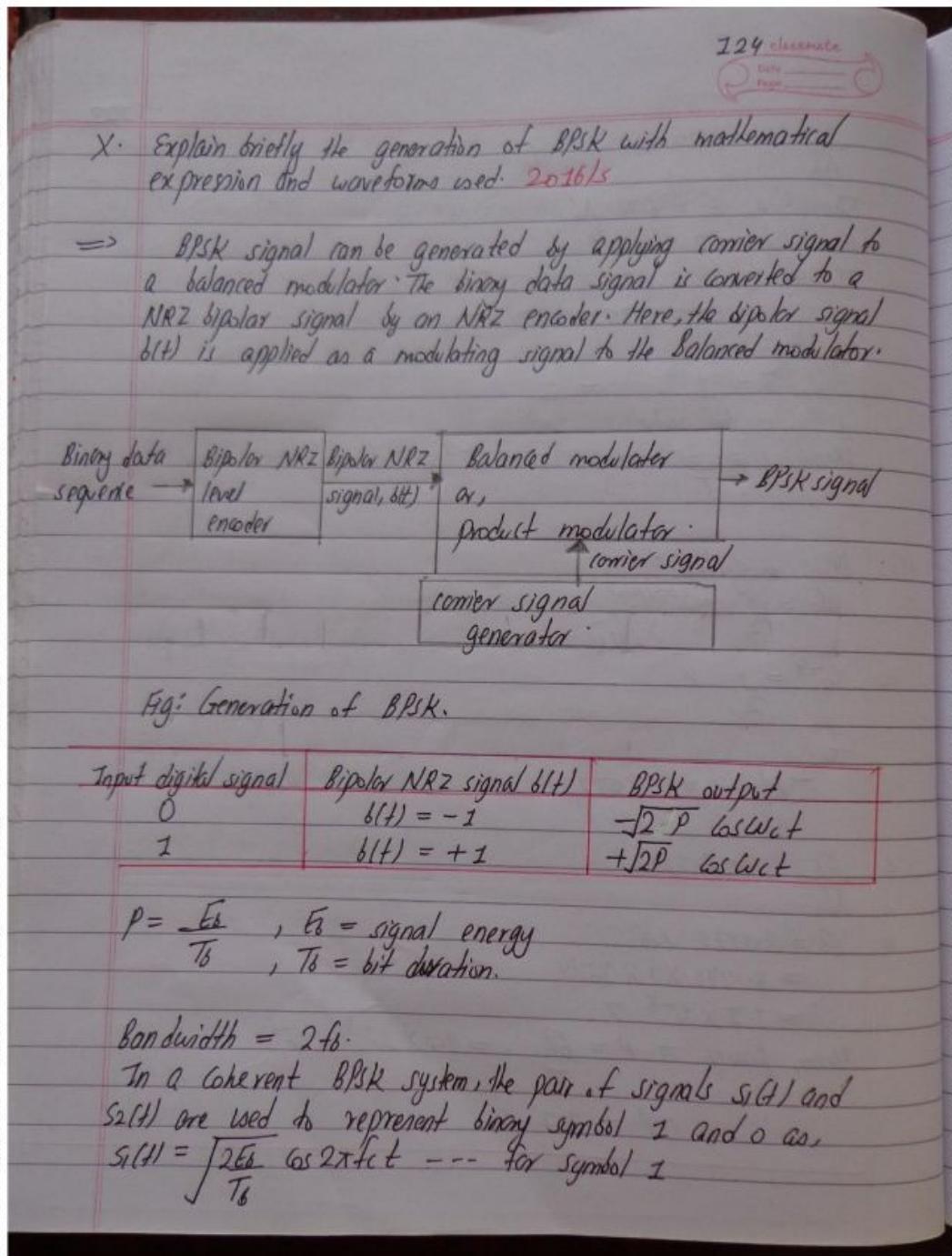
or $\frac{E_b}{N_0} = 8.5171$

or $E_b = 8.5171 N_0$
= $8.5171 \times 2 \times 10^{-10}$
= $1.7 \times 10^{-9} \text{ J}$

Hence, power = $P = \frac{E_b}{T_b} = 1.7 \times 10^{-9} \times 10^6$
 $\therefore P = 1.7 \text{ mW}$

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$$S_2(t) = \int \frac{2E_s}{T_b} \cos(2\pi fct + \pi) = -\int \frac{2E_s}{T_b} \cos 2\pi fct \text{ for '0'}$$

In case of BPSK, there is only one basic function of unit energy which is,

$$\phi_1(t) = \int \frac{2}{T_b} \cos 2\pi fct, \quad 0 \leq t \leq T_b$$

Hence,

$$S_1(t) = \sqrt{E_s} \phi_1(t) \quad 0 \leq t \leq T_b \text{ for symbol } 1.$$

$$S_2(t) = -\sqrt{E_s} \phi_1(t) \quad 0 \leq t \leq T_b \text{ for "0".}$$

The phase changes occurs in the corner only at zero crossing. The BPSK waveform has full cycles of sinusoidal corner.

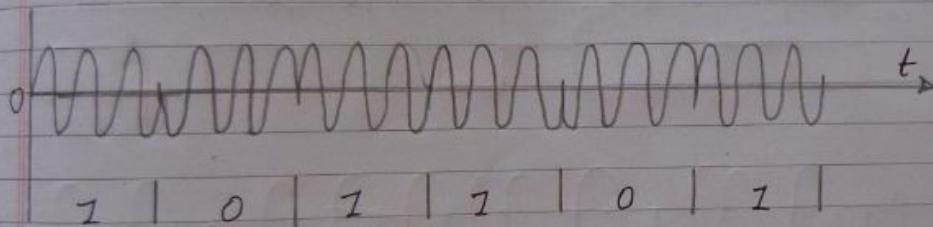


Fig: BPSK waveform

- Q) Explain DPSK system. Encode binary sequence 00110100010 into differentially encoded sequence & draw DPSK waveform. *2015/16*

=> A DPSK system can be viewed as the non coherent version of the PSK. To send symbol 1 we phase advance the current signal waveform by 180° and to send symbol '0' we leave the phase of the current signal waveform unchanged.
dk is generated by using the logical eqn,

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$d_k = d_{k-1} \oplus b_k$
where, b_k = input binary digit at time kT_b
 d_{k-1} = previous value of differentially encoded digit.

input binary sequence	0	0	1	1	0	1	0	0	0	1	0
d_{k-1}	0	0	0	1	0	0	1	1	1	1	0
d_k	0	0	1	0	1	0	1	0	0	1	0

For symbol '0' to phase shift by 180° and symbol '1' to leave the phase, we take,

$$d_k = \overline{d_{k-1} \oplus b_k}$$

0 0 1 1 0 1 0 0 0 1 0

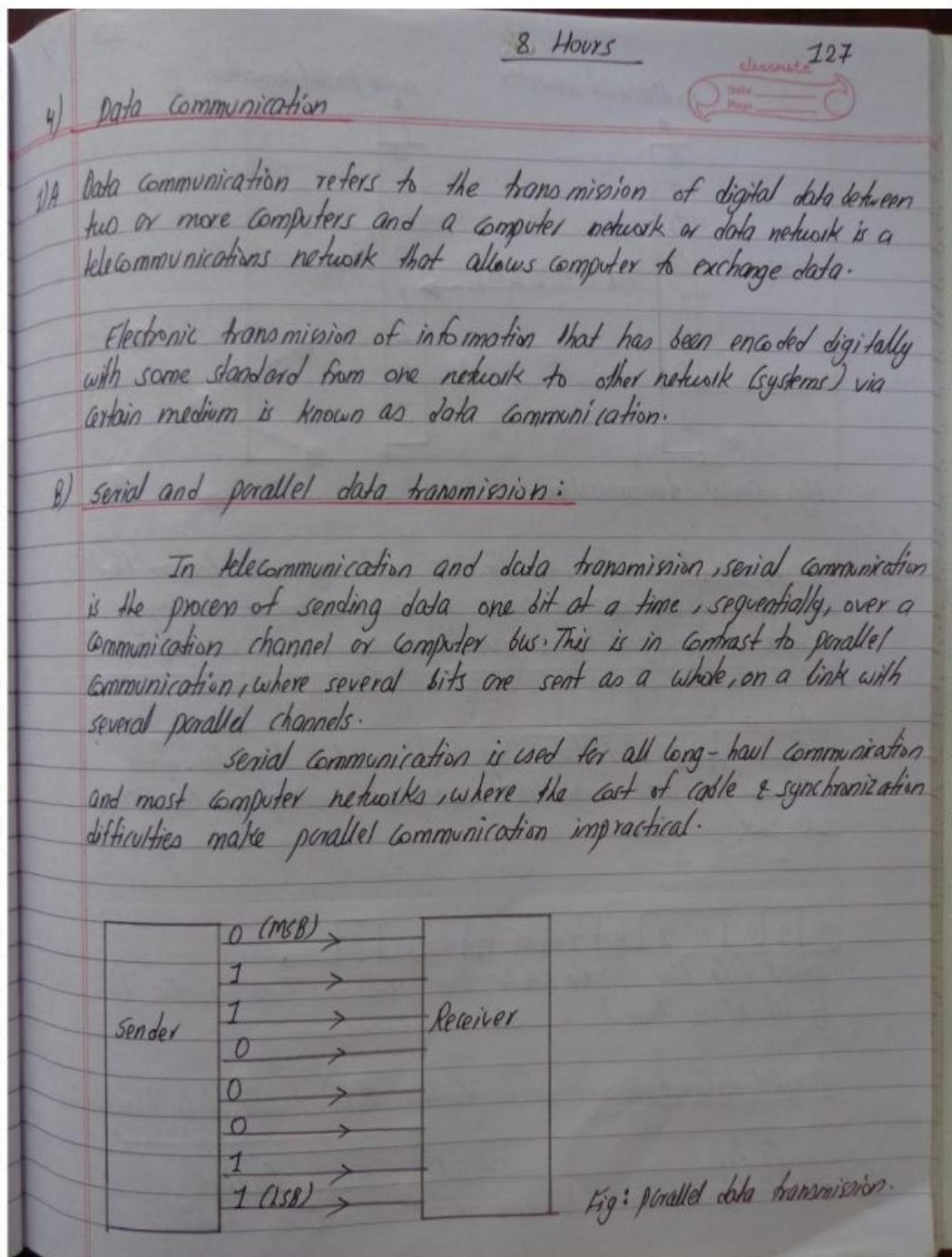
↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ ↓ →

o ——————
A sinusoidal waveform starting at zero.

Fig: DPSK waveform.

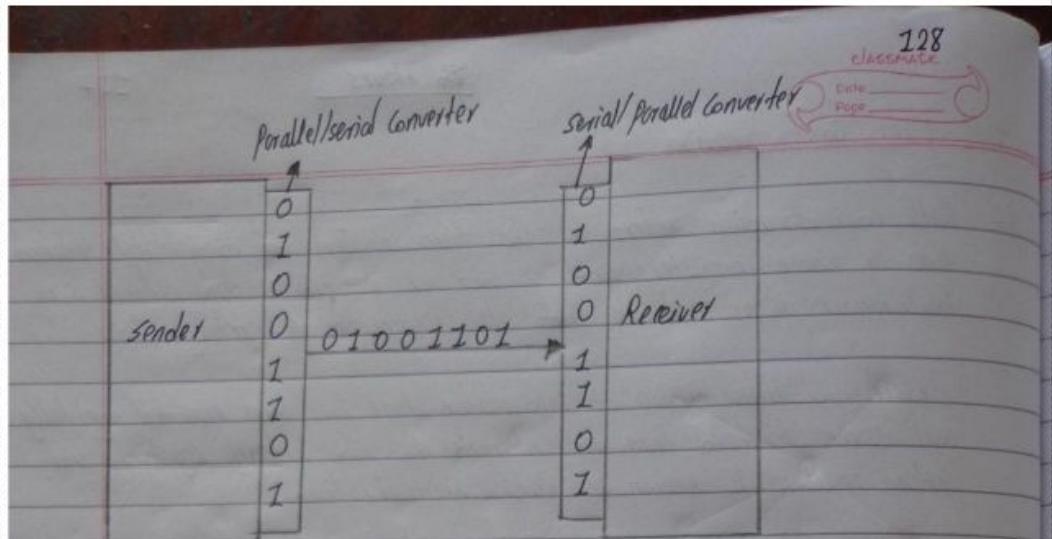
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In parallel transmission, all the bits of a byte are transmitted simultaneously on separate wires. It is a method of conveying multiple binary digits (bits) simultaneously.

0 →
1 →
1 →
0 →
parallel data lines (multi parallel lines for data transmission, bits are sent parallel in the same time.)

0 | 1 | 0 | 1 | 1 | 0 | 1 | 0 | 0 | 1 | 0 | 1 | 0 | 1
serial data line (single data line used in transmission bit by bit at a time.)

Parallel data transmission is faster than serial data transmission.

Factor	Serial	Parallel
① No. of bits transmitted	one -bit	n bits.

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at one clock.		
① No. of lines required to transmit n bits	one line	n lines.
② speed of data transfer	slow	Fast
iv) cost of transmission	low as one line is needed.	High as n lines are needed.
v) Data rate	slower.	Faster
vi) cable	use less no. of wires.	use large no. of wires.
vii) cable length	use long shield cables protected from EMI	can't use lengthy cables EMI limits data rate
viii) Communication error	only one bit is communicated at a time hence, less chance of capacitive effects of error.	Bits get corrupted due to between cable wires.
ix. Applications	Long distance communication	short distance communication like computer to printer.
x. Communication modes	can be Full duplex.	can be Half duplex.
<u>G) Data Communication Topologies :</u>		
Network topologies may be physical or logical with respect their functionality. In general, physical topology relates to a core network whereas logical topology relates to basic network. It describes the layout or appearance of a network. Eg: star, bus, ring, mesh and Hybrid.		

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Types :

① Star Topology :
Each computer is connected to a central device (Hub) by a separate cable.

Advantages :

- Easy to add new stations
- Easy to monitor and troubleshoot.
- Can accommodate different wiring.

Disadvantages :

- Failure of Hub cripples attached stations.
- More cable required and more expensive.

② Bus - Network Topology :

A Bus network is a network topology in which nodes are directly connected to a common linear or branched half duplex link called a Bus.

Advantages :

- i) Easy to connect a computer or peripheral to a linear Bus.
- ii) Requires less cable length than a star topology.

Disadvantages :

- i) Entire network shuts down if there is a break in the main

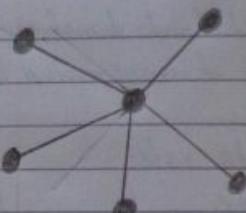


Fig : Star network

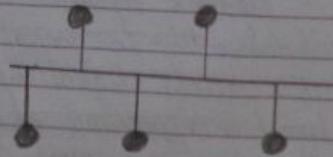


Fig : Bus

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cable.

- ii) Terminators are required at both ends of the backbone cable.
- iii) Difficult to identify the problem if the entire network shuts down.
- iv) Not meant to be used as a stand-alone solution in a large building.

① Ring network Topology :

If it is a network topology in which each node connects to exactly two other nodes, forming a single continuous pathways for signals through each node. Data travels from node to node, with each node along the way handling every packet.

Rings are normally implemented using twisted pair or fiber-optic cable.

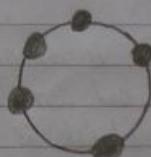


Fig: Ring network

Advantages :

- All stations have equal access.
- Growth of system has minimal impact on performance.

Disadvantages :

- most expensive topology
- Failure of one computer may impact others.
- Complex.

② Tree topology :

A tree topology combines characteristics of linear bus and star topologies. It allows the expansion of an existing network and enables

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to configure a network to meet their needs.

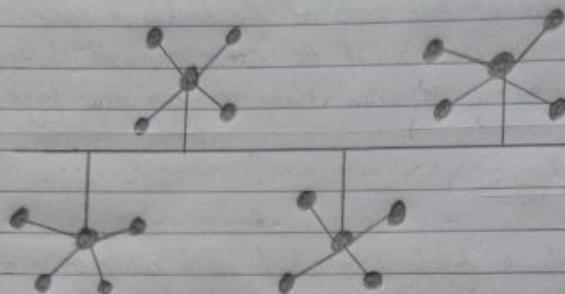


Fig: Tree topology.

Advantages:

- If one segment is damaged, other segments are not affected.
- error detection and correction is easy.
- expansion of network is possible and easy.

Disadvantages:

- scalability of the network depends on the type of cable used.
- maintenance becomes difficult.
- It relies heavily on main Bus cable if it breaks whole network is crippled.

v) Hybrid Topology :

It is an integration of two or more different topologies to form a resultant topology which has many advantages of all constituent basic topologies rather than having characteristics of one specific topology.

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Advantages:

- Reliable
- Scalable
- Flexible
- Effective

Disadvantages:

- Complexity of design.
- Costly Hub
- Costly infrastructure

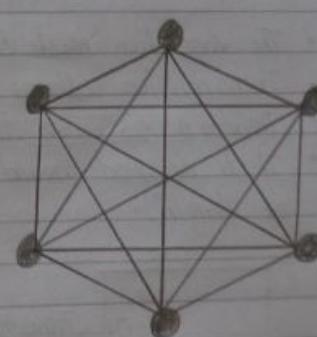
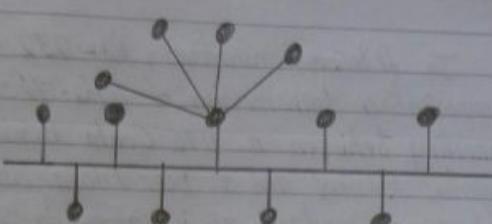
vi mesh Topology:
Each and every node of the network is interconnected.

Advantages:

- It has multiple lines, so if one route is blocked then other routes can be used for communication.
- performance is not affected with heavy load of data transmission.
- ensures data privacy or security
- expansion and modification can be done without disrupting other nodes.

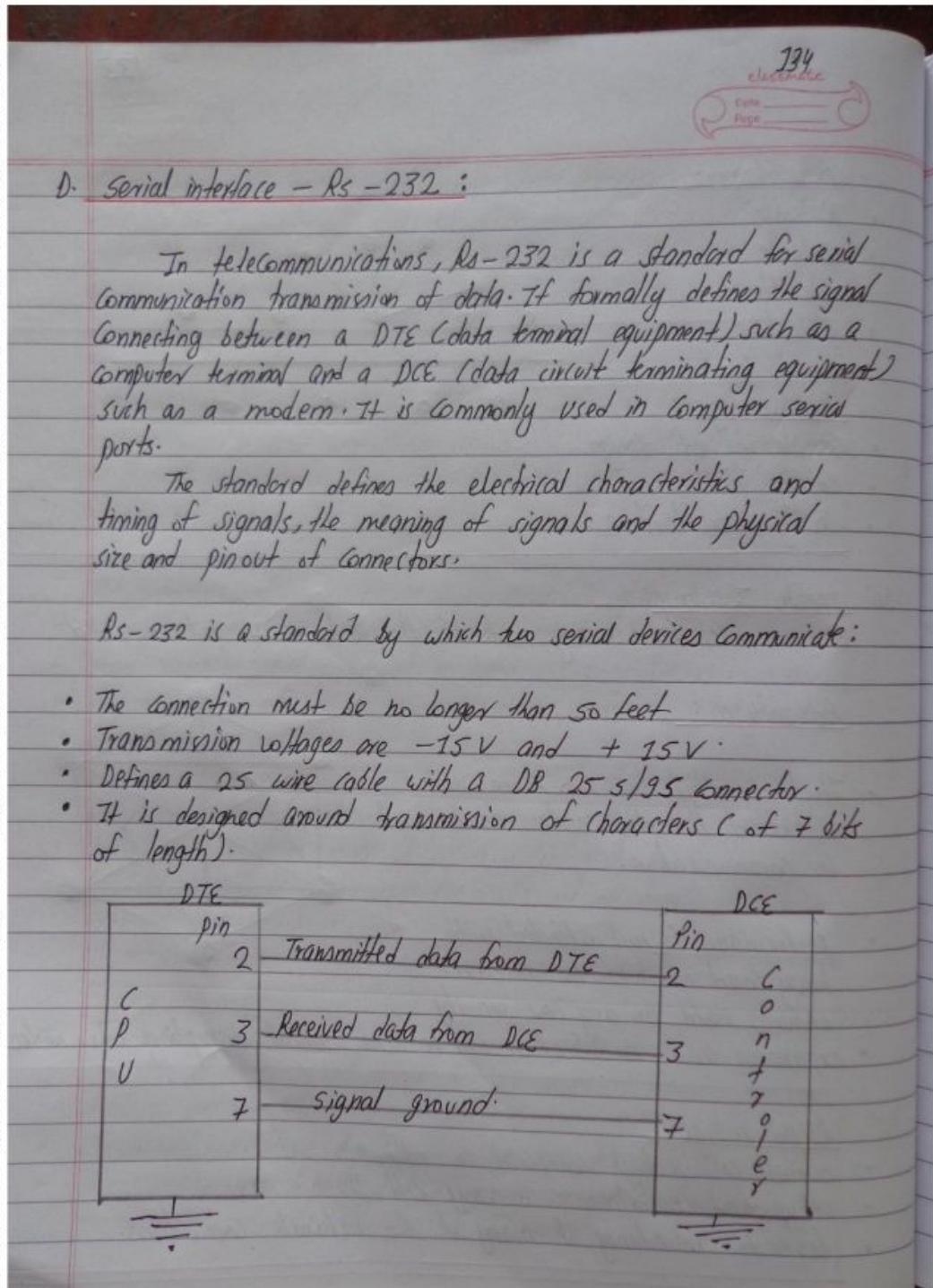
Disadvantages:

- overall cost is high compared to other.
- setup and maintenance is very difficult.
- leads to redundancy of many of the network connections.



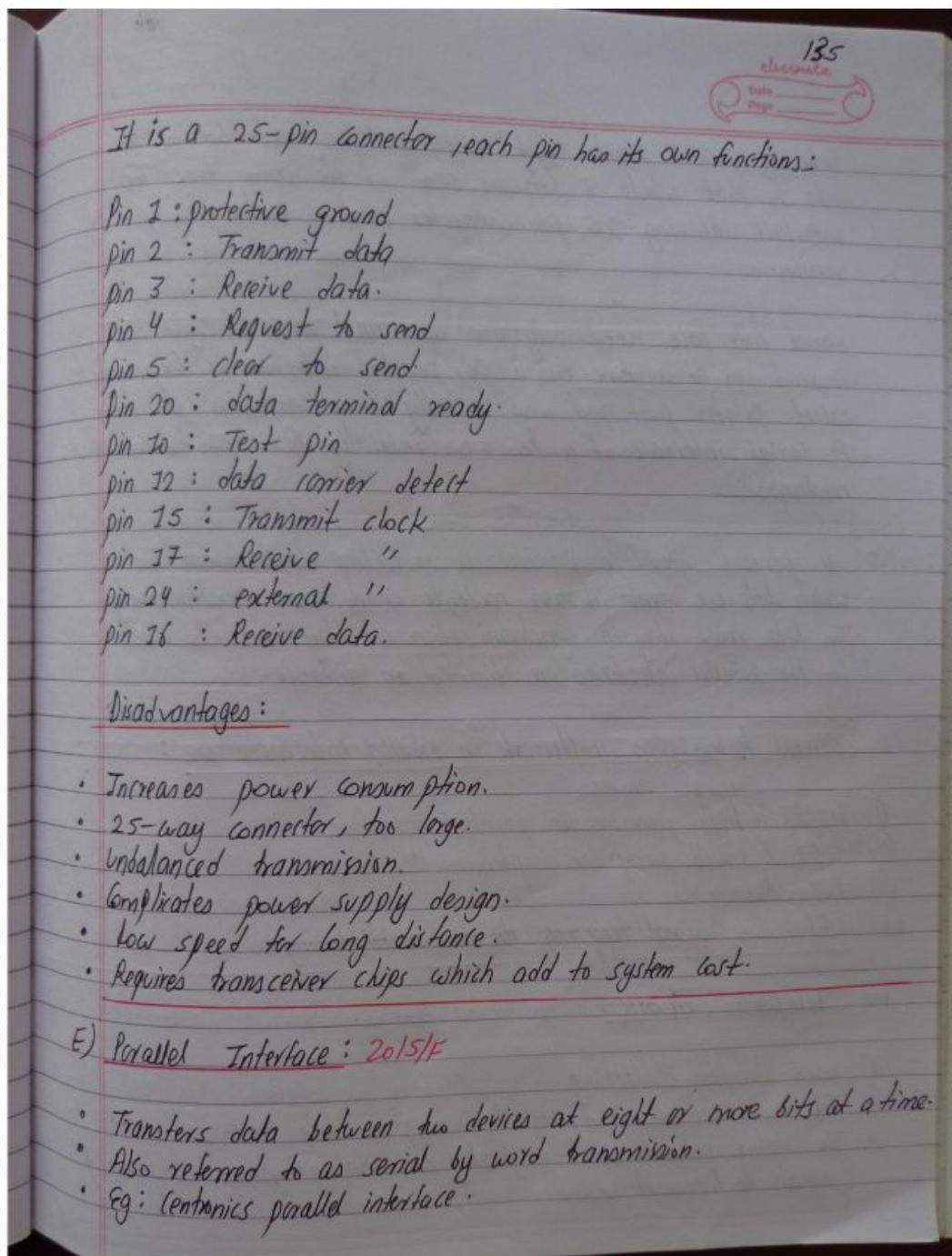
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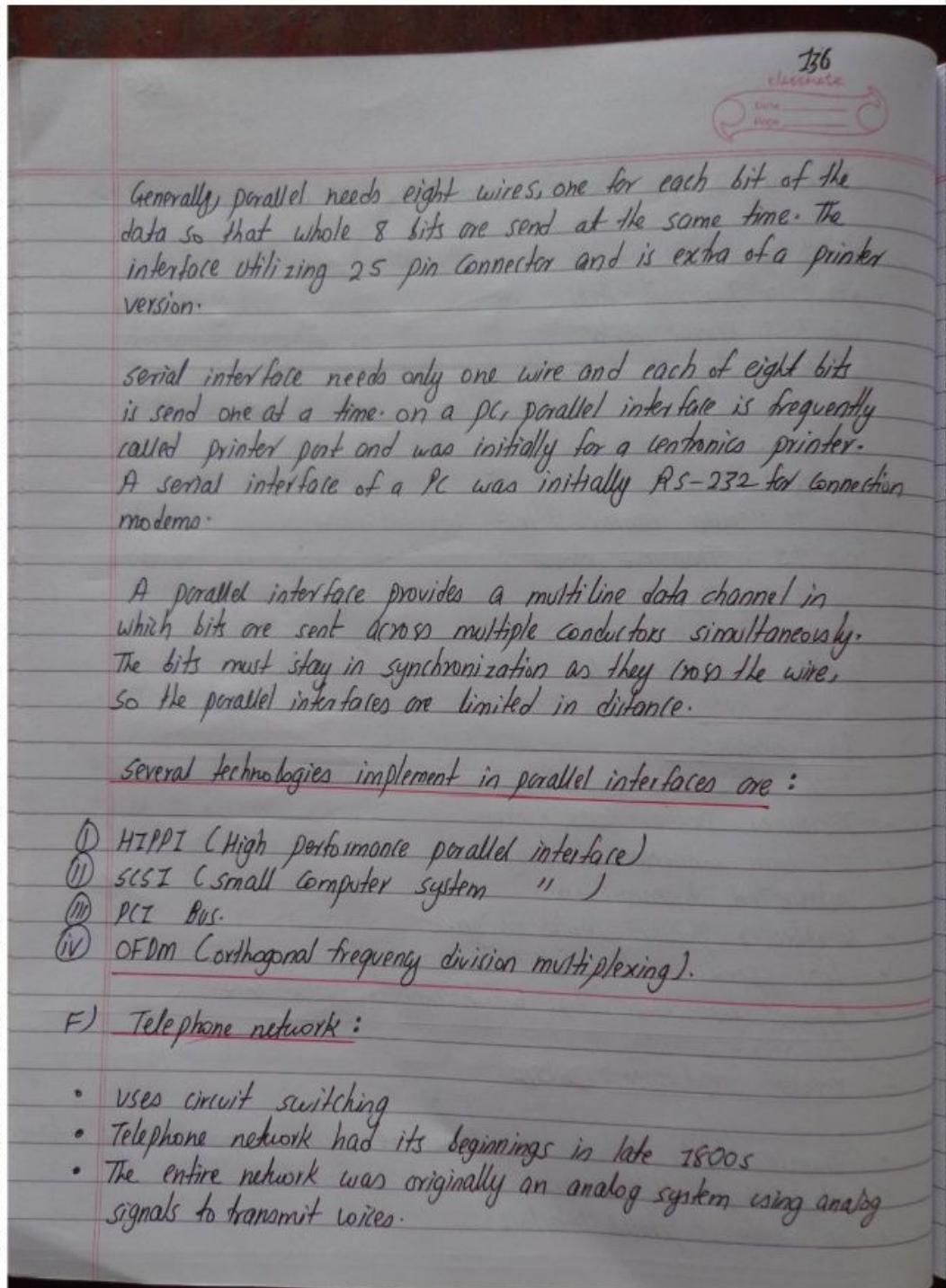
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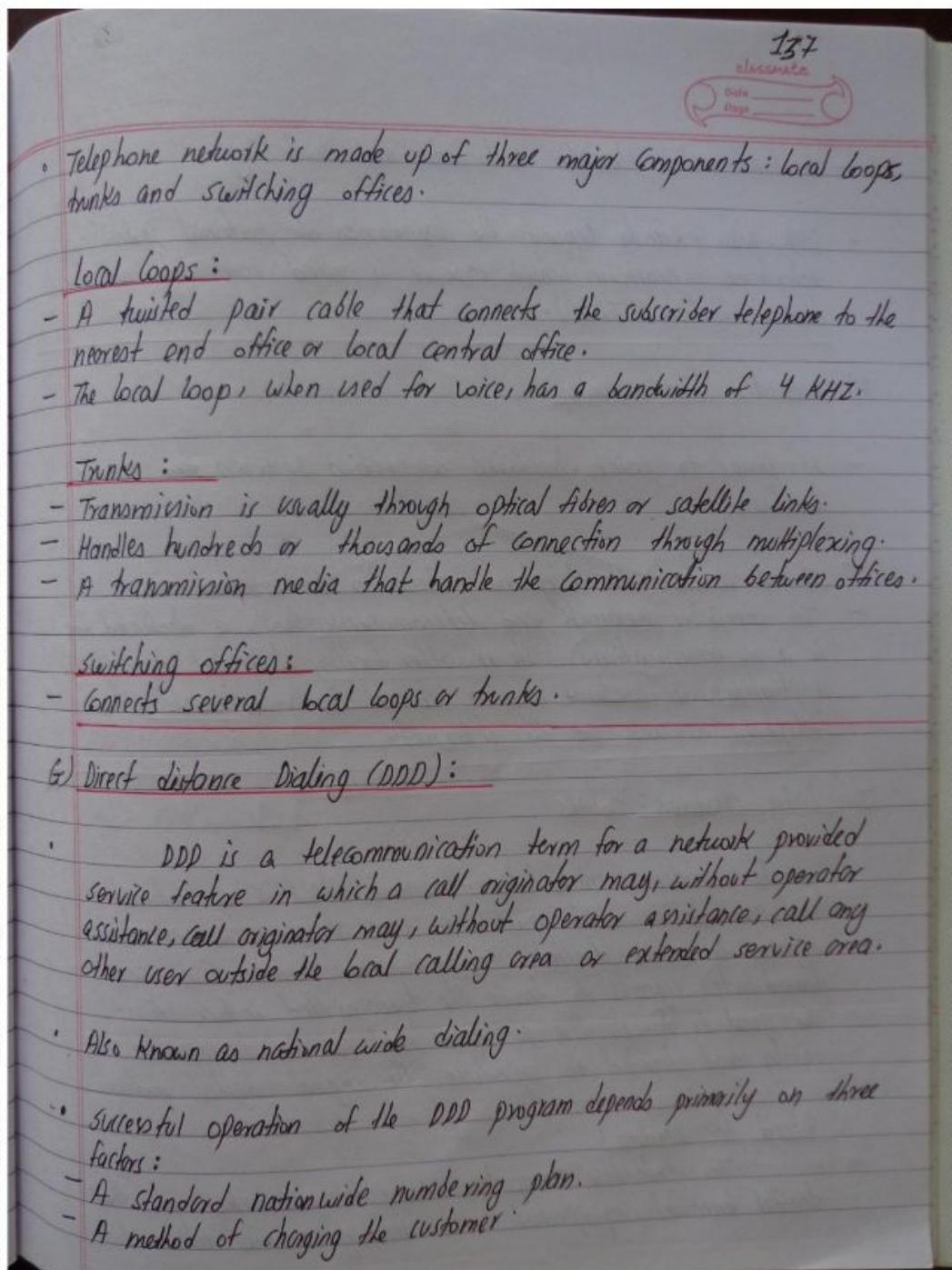
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H) Dedicated Line service :

- A fundamental plan for automatic toll switching.
- DDD also extends beyond the boundaries of national public telephone network, in which case it is called international direct distance dialing (IDDD).

I. Data Modems: 2014/5, 2017/5, 18/1F

A modem (modulator - demodulator) is a network hardware device that modulates one or more carrier wave signals to encode digital information for transmission and demodulates signals to decode the transmitted information. The goal is to produce a signal that can be transmitted easily and decoded to reproduce the original digital data.

Need for modem:

To interface computers, computer networks and other digital terminal equipment with analog communication lines

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and radio channel.

Types of modems :

- (1) Half Duplex
- (2) Full Duplex
- (3) 2 wire modem
- (4) Asynchronous and synchronous modems.

Synchronous modems : 2017/5

- Clocking information is recovered at the receiver
- use PSK or QAM modulation technique.
- used for mostly medium and high speed applications.

Asynchronous modems : 2017/5

- No clocking information is sent.
- Mostly use ASK / FSK
- Restricted to use for low speed applications.
eg: Bell systems 202 T/5 modems uses FSK
 - 202 T-full duplex, four wire operation
 - 202 S-half " , " " "

J. ISDN : 2015/F, 2016/F, 2014/K, 18/F

Integrated service digital Network (ISDN) is a set of communication standards for simultaneous digital transmission of voice, video, data and other network services over the traditional circuits of the public switched telephone network.

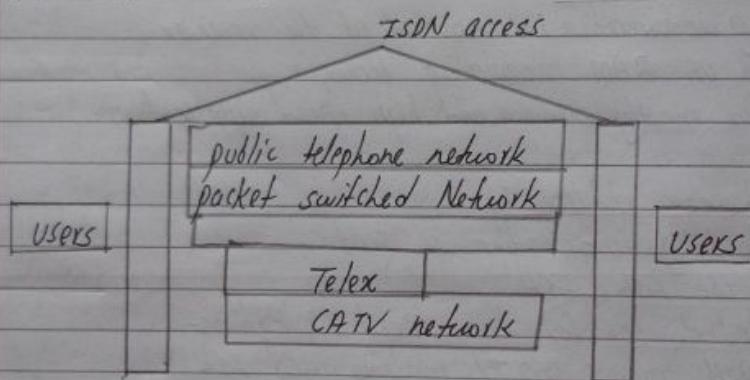
ISDN is an all digital communications line that allows for the transmission of voice, data, video and graphics at very high

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speeds, over standard communication lines.

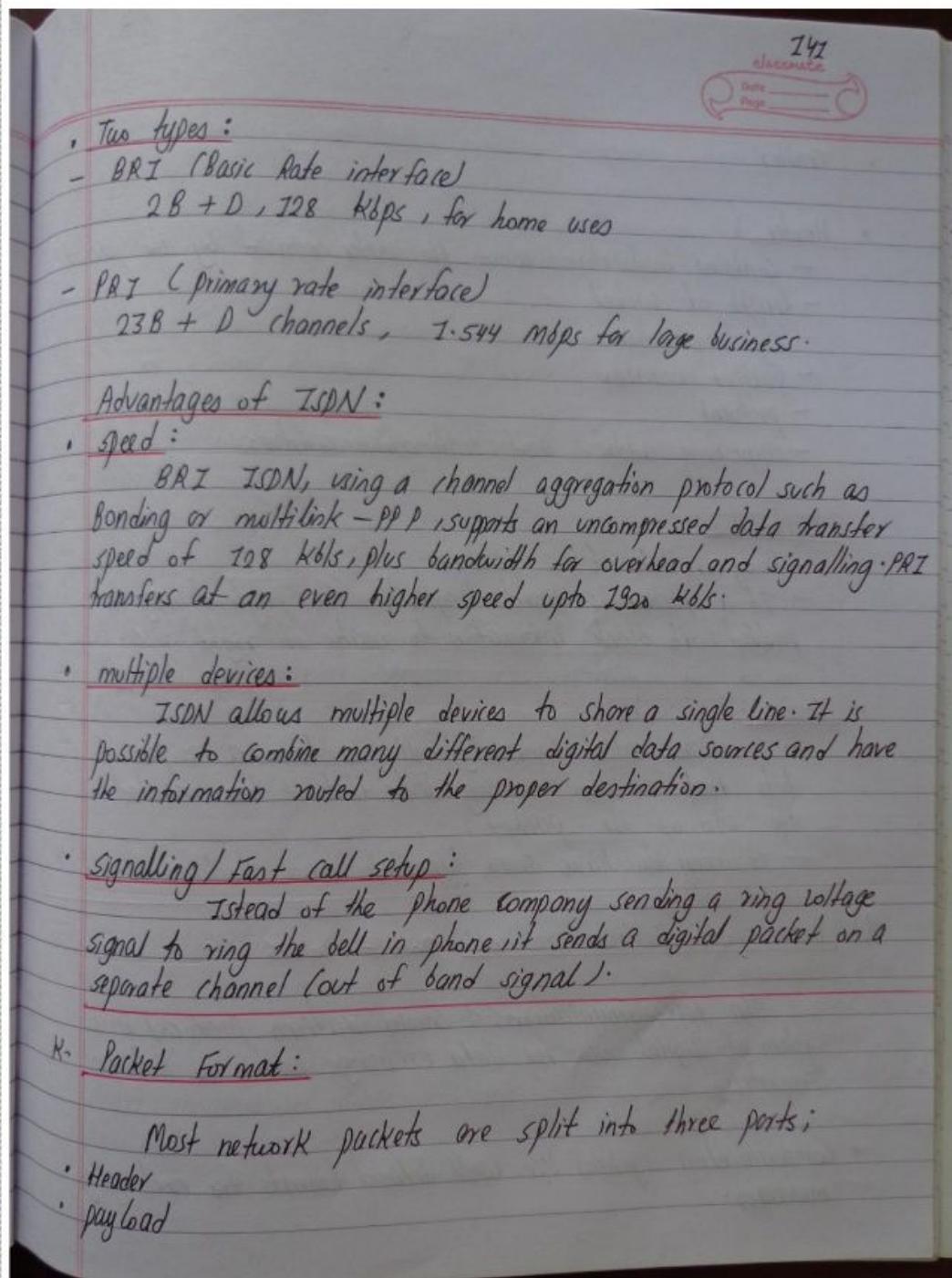
- ISDN provides a single, common interface with which to access digital communications services that are required by varying devices, while remaining transparent to the user.
- ISDN is not restricted to public telephone network alone; it may be transmitted via packet switched networks, telex, CATV network etc.



- It offers circuit switched connections (for either voice or data) and packet-switched connections (for data) in increments of 64 kilobits/s.
- With ISDN, voice and data are carried by bearer channels (8 channel) occupying a bandwidth of 64 kbit/s. Some switches limit 8 channels to a capacity of 56 kbit/s.
- A data channel (D channel) handles signalling at 16 kbit/s or 64 kbit/s, depending on the service type.

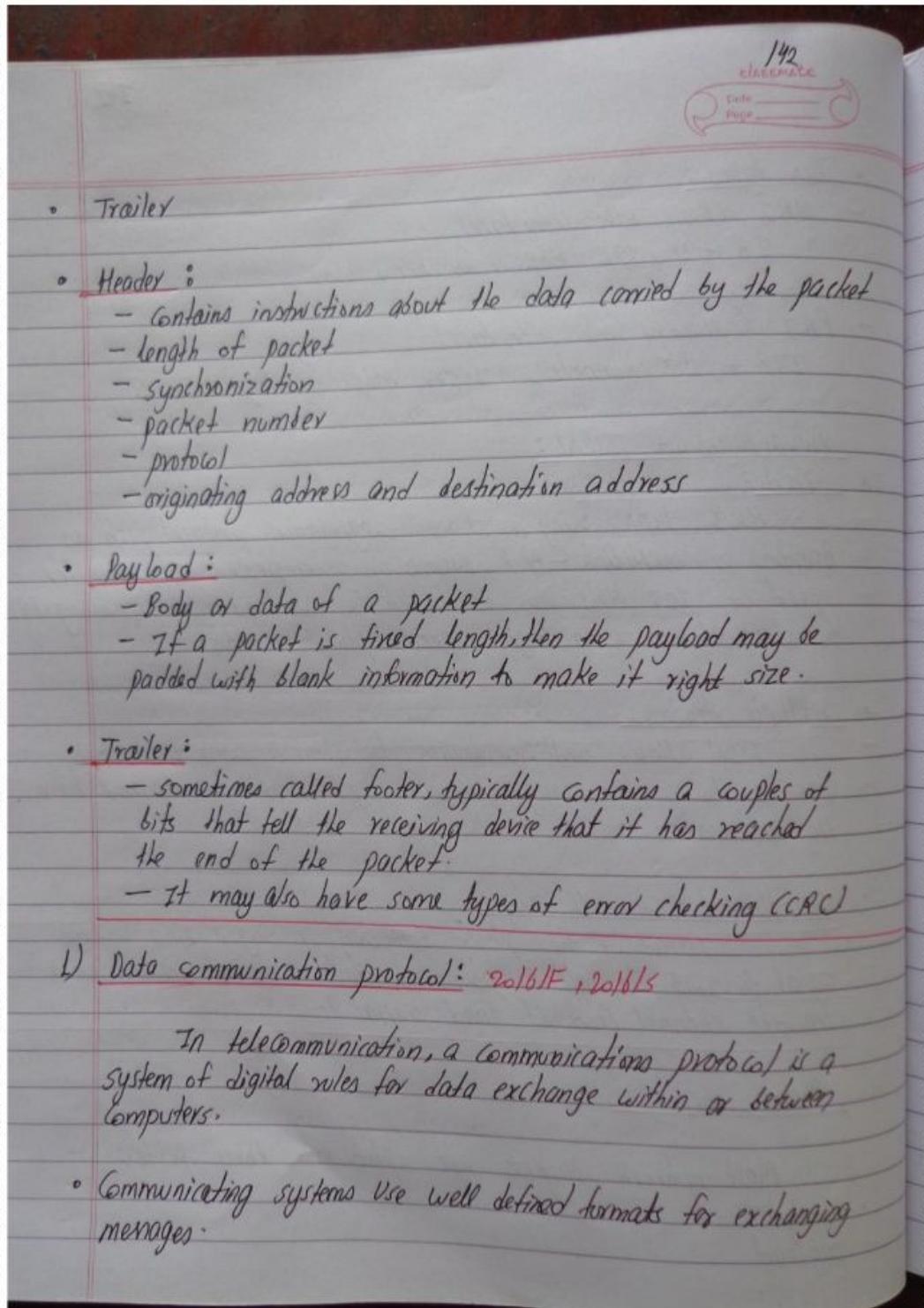
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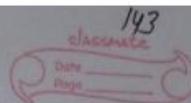
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- Some network protocols :
- open systems interconnections (OSI)
 - File transfer protocol (FTP)
 - Transmission control protocol / internet protocol (TCP/IP)



Basic requirements of protocols :

In general much of the following should be addressed.

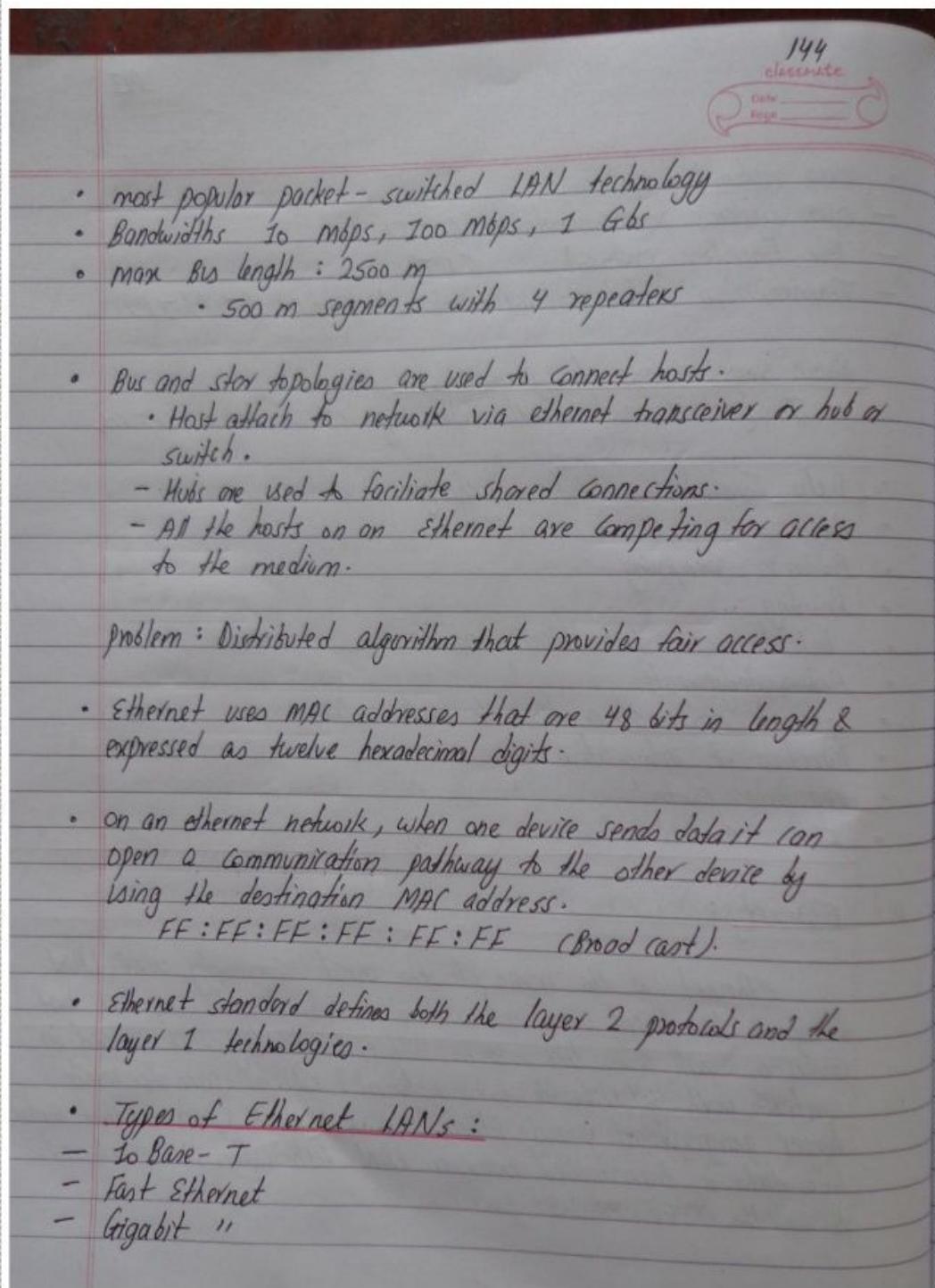
- Data formats for data exchange.
- Address " " " "
- Address mapping
- Routing
- Detection of transmission errors.
- Acknowledgements
- loss of information timeouts and retries
- direction of information flow
- sequence control
- Flow control

M) Ethernet :

Ethernet is the name of the most commonly used LAN today. A LAN (local area network) is a network of computers that covers a small area like a room, office, a building. It is used in contrast with WAN (wide area network) which spans for much larger geographical areas. Ethernet is a network protocol that controls how data is transmitted over a LAN. Technically, it is referred to as the IEEE 802.3 protocol.

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- 10 Gbps ethernet
- wireless ethernet
- Ethernet uses baseband signalling.

N. Error detection and correction : 2017/F, 2015/S, 2015/F, 2016/F, 2014/S

Networks must be able to transfer data from one device to another with complete accuracy. While the transmission data can be corrupted, for reliable communication errors must be detected and corrected.

Types of errors :

- single bit error :
only one bit in the data unit has changed.
- Burst error :
two or more bits in the data unit has changed.

Fig: Single bit error.

Received: 0 1 0 0 1 Sent: 0 0 0 0 1

0 changed to 1

Fig: Burst error.

Received: 0 1 0 1 0 0 Sent: 0 0 0 1 1 0

Bits corrupted by burst error

The diagram illustrates two types of errors in data transmission. In the first part, a single bit error is shown where the third bit of the received sequence (0 1 0 0 1) is corrupted from 0 to 1, while all other bits remain the same as the sent sequence (0 0 0 0 1). In the second part, a burst error is shown where the fourth and fifth bits of the received sequence (0 1 0 1 0 0) are corrupted from 0 0 to 1 1, while the first three and last one bits remain the same as the sent sequence (0 0 0 1 1 0).

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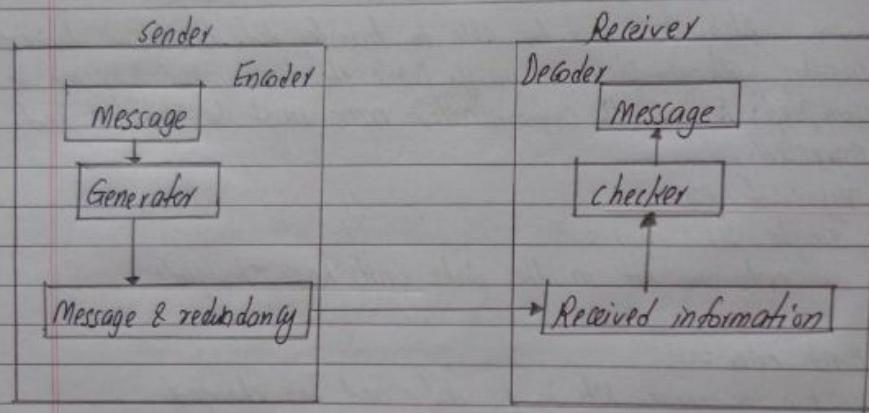
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Error detection :

Error detecting code is to include only enough redundancy to allow the receiver to deduce that an error occurred, but not which errors and have it request a re-transmission.

Error detection uses the concept of redundancy, which means adding extra bits for detecting error at the destination.



Error detection methods :

- ① parity checking
- ② check sum error detection
- ③ cyclic redundancy check (CRC)

For error detection and correction, it is essential to add some check bits to a block of data bits. These check bits are also known as redundant bits because they do not carry any useful information.

Before correcting the error introduced in data bits, it is essential to first detect them.

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Q) parity checking :

In this method, an additional bit called as parity bit is added to each data word. This additional bit is so chosen that the resultant word will have either an even parity or odd parity. If even parity is to be used then a parity bit is added to make the total number of 1 bits even. Similarly, for odd parity the total number of 1 bits is made odd by adding the parity bit.

P | data word → even parity as total number of 1 bits in the data word is 4 ie, even. hence, parity bit P = 0

P	0	1	0	0	1	0	1
---	---	---	---	---	---	---	---

P | data word → even parity. Total number of 1 bits in the data word is 3 originally. Add a parity bit P of 1 to make the total number of 1 bits to be 4 ie, even.

P	1	0	0	1	0	0	1
---	---	---	---	---	---	---	---

error detection :

The parity checking at the receiver can detect the presence of an error if the parity of the received signal is different from the expected parity.

Transmitted code:	P message bits	Parity	Receiver's decision
0 1 0 0 1 0 1	even	correct word	
Received code with one error	0 0 0 0 1 0 1	odd	correct word
Received code with three errors	0 0 1 0 0 0 1	odd	correct word

The receiver detects the presence of error if the number of errors is odd.

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The image shows a page from a handwritten notebook. At the top right, there is a red stamp with the number '148' and the word 'classmate'. Below the stamp, there is a small drawing of a person's head with a speech bubble containing the text 'Disha Patel'. The main text on the page is handwritten in black ink. It starts with a statement about the parity of errors, followed by a section on the failure of parity checking methods, and then a list of limitations. There is also a section on checksum error detection with a detailed explanation.

of error is odd i.e. 1, 3, 5, 7--

Failure of parity checking methods:
If the number of errors introduced in the transmitted code is two or any even numbers then the parity of the received code word will not change. It will remain same i.e. even and the receiver will fail to detect the presence of errors.

Limitation of parity checking:

- (i) It is not suitable for detection of multiple errors (2, 4, 6 etc)
- (ii) It cannot reveal the location of erroneous bit. It cannot correct the error either.

(ii) checksum Error detection:

The errors usually occur in bursts. The parity check method is not useful in detecting the errors under such conditions. The checksum error detection method can be used successfully in detecting such errors. A checksum is transmitted along with every block of data bytes. The carries of the MSB are ignored while finding out the checksum byte.

The advantage of this method over the simple parity checking method is that the data bits are mixed up due to the 8-bit addition. Hence, checksum represents the overall data block. In checksum, hence, there is 255 to 1 chance of detecting random errors.

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⑩ cyclic Redundancy check (CRC):

- CRC error detection method treats packet of data to be transmitted as a large polynomial.
- Transmitter:
Using polynomial arithmetic divides polynomial by a given generating polynomial.
- Quotient is discarded.
Remainder is attached to the end of message.
- message with the remainder is transmitted to the receiver.
- Receiver divides the message and remainder by some generating polynomial.
- If a remainder not equal to zero results → error during transmission.
- " " " of zero results → no error during transmission.

divisor) dividend
generator polynomial quotient → not used
 |
 Remainder → dividend: Content of frame
 |
 fixed size used for checksum

The CRC is calculated by performing a modulo 2 division of the data by a generator polynomial and recording the remainder after division.

Eg: A codeword is received as 1100 1001 01011. Check whether there are

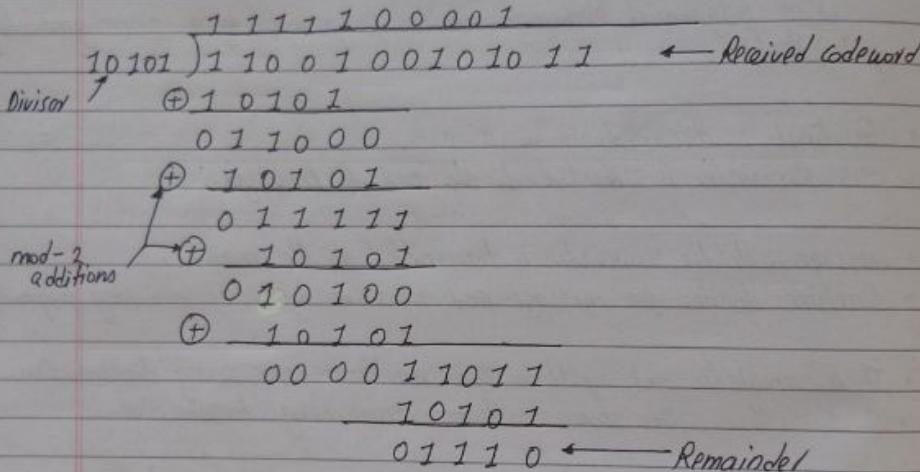
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error in the received codeword, if the divisor is 10101. (The divisor corresponds to the generator polynomial).

\Rightarrow Data word = 1100100101011
Divisor = 10101



The non-zero remainder shows that there are errors in the received codeword.

CRC cannot detect all types of errors. The probability of error detection and the types of detectable errors depend on the choice of divisor.

Hamming Codes :

Hamming codes are linear block codes. The family (n, k) Hamming codes for $q \geq 3$ is defined by,

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- Code word length: $n = 2^q - 1$
- no. of message bits: $K = 2^q - 1 - q$
- no. of parity bits: $(n-K) = q$
where $q \geq 3$, ie, minimum number of parity bit is 3
- minimum distance, $d_{min} = 3$

Code rate or code efficiency = $\frac{K}{n} = \frac{2^q - q - 1}{2^q - 1} = 1 - \frac{q}{2^q - 1}$

If $q >> 1$, then code rate, $r \approx 1$

Code word length, $n = 2^q - 1$ →
 $K = 2^q - 1 - q$ $q = (n-K)$

Message bits → Parity bits →

Fig: Code word structure of Hamming code.

Error detection and correction capabilities of Hamming codes:

- Number of errors that can be detected per word = 2
minimum distance, $d_{min} = 3$
- no. of errors that can be corrected per word = 1
since, $d_{min} \geq (2t+1)$, therefore, $3 \geq (2t+1)$ or $t \leq 1$
Thus, with $d_{min} = 3$, it is possible to detect upto 2 errors and it is possible to correct upto only 1 error.

Q: What are the advantages and disadvantages of ISDN? What benefits does it offer to user, network provider & manufacturer? 2016/F, 2014/k

=> Advantages:

i) The basic advantage of ISDN is to facilitate the user with multiple

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digital channels. These channels can operate concurrently through the same one copper wire pair.

- ii) ISDN takes only two second to launch a connection while other modem takes 30 to 60 seconds for establishment.
- iii) The digital signals broadcasting transversely the telephone lines.
- iv) It provides high data rate because of digital scheme which is 56 kbps.
- v) ISDN network lines are able to switch manifold devices on the single line such as phones, computers, cash registers and many other devices. These all devices can work together and directly be connected to a single line.

Disadvantages:

- i) very costly than the other typical telephone system.
- ii) ISDN requires specialized digital devices.

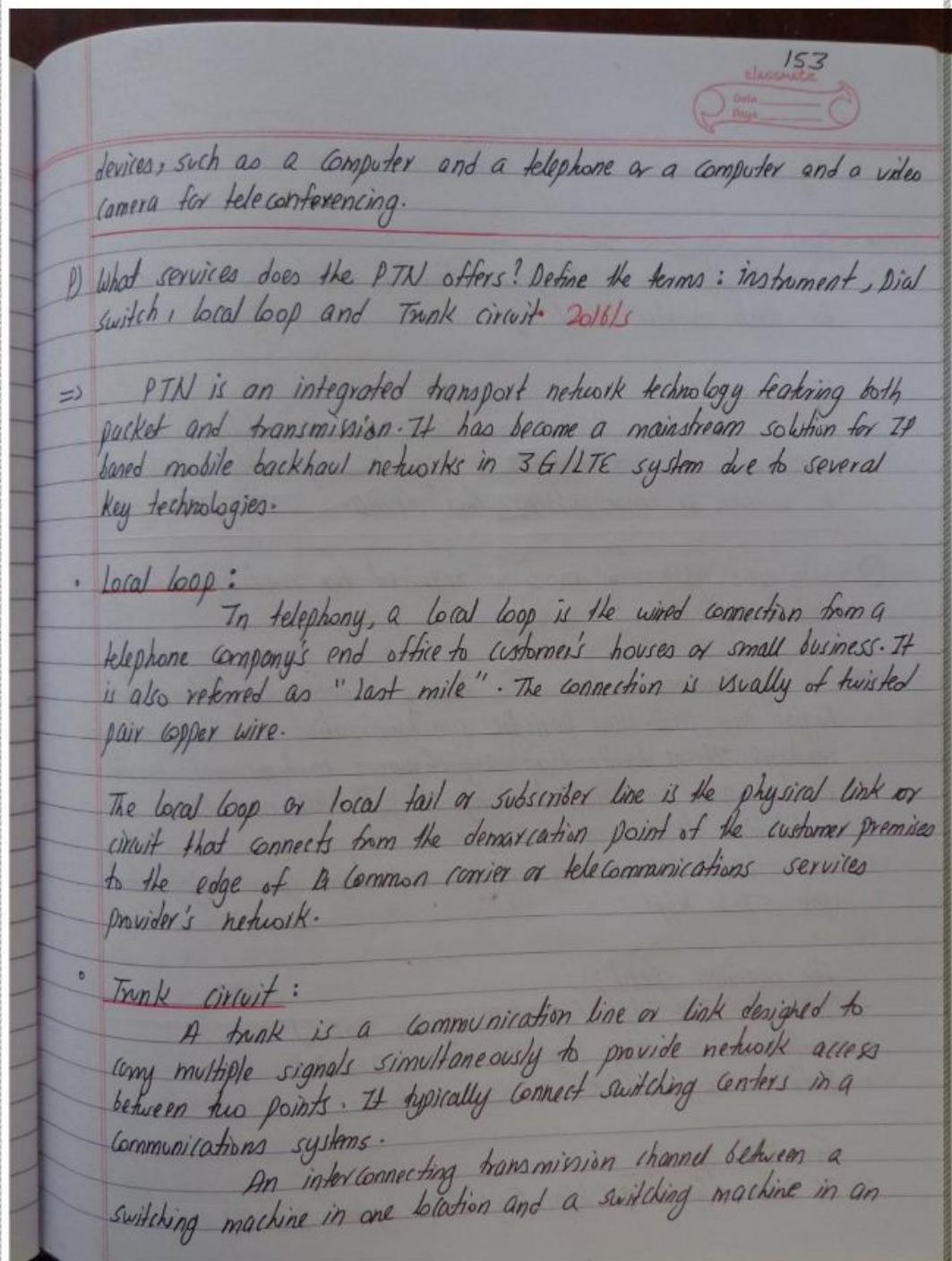
Typical ISDN applications:

- * Internet access and on-line service
- * Telecommuting
- * Remote office routing
- * Video-Conferencing
- * Disaster Recovery
- * PC-to-PC screen sharing and collaboration.

What makes ISDN unique is that each B channel is a separate communication circuit. That means that just one ISDN line can support simultaneous two-way communication for two

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Adjacent node.

- Dial switch :
A disc on a telephone that is rotated a fixed distance for each number called.
- Instrument :
Devices that communicate, denotes, detects, indicates, measures, observes, records or signals a quantity or phenomenon or controls or manipulates other device.

Q Why are synchronous modems required for medium and high speed applications? *2016 LS, 181F*

⇒ Synchronous modems use timing to determine where data begins and ends. Uses periodic synchronization bits to synchronize modems. It is faster than asynchronous modems and provides function such as error checking. It uses QPSK and QAM modulation techniques.

- Used for mostly medium and high speed applications upto 57.6 kbps

For medium speed :

- QPSK for 2.4 kbps (eg: Bell system 201 C)
- 8 - PSK for 4.8 kbps (eg: " " 208 A)
- Both are duplex (full), 4 wire systems.

For High speed :

- 16 QAM - for a 9.6 kbps (eg: Bell system 209 A)

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• Full duplex, four wire transmission.

Since the synchronous modems are used for medium and high speed modems, these modems contain the following additional hardware:

- clock recovery
- equalizers
- scramblers and descrambler circuit.

Q) What do you mean by packet switching? Explain the packet data transmission with an example. of ethernet packet format. 2015/16

⇒ Packet switching is a method of grouping data transmitted over a digital network into packets which are composed of a header and a payload.

Packet switching is the transmission method used for most computer network because the data transported by these networks is fundamentally bursty in character and can tolerate latency (due to lost or dropped packets). The transmission bandwidth needed varies greatly in time, from relatively low traffic because of background services such as name resolution services, to periods of high bandwidth usage during activities such as file transfer. The Internet is the prime example of a packet switched network based on the TCP/IP protocol suite.

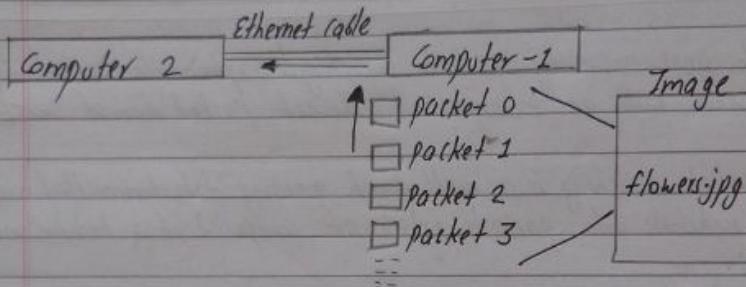
Packets - Data from Here to There :

- Eg: To send an image file between ethernet connected computers
- This is 'one hop' LAN case
- eg: 50 KB image.jpg
- 50,000 bytes

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- How to send the image.jpg on the wire?
- use packets
- Divide bytes of image.jpg into packets
- say each packet is 1500 bytes.
- Then image.jpg divides into about 32 packets.
- Network transmits one packet at a time.



For transmission, the 50 KB of the image is divided into packets. The packet is the natural unit of transmission within networking.

By sending a large file in several small chunks over a network, packet switching minimizes the impacts of data transmission errors. Statistical multiplexing is used to enable devices to share these circuits.

5) Explain serial and parallel interface in data communication.

2015/F

=> serial interface

① Data is transmitted bit after bit in a single line.

parallel interface.

Data is transmitted simultaneously through group of lines (Bus).

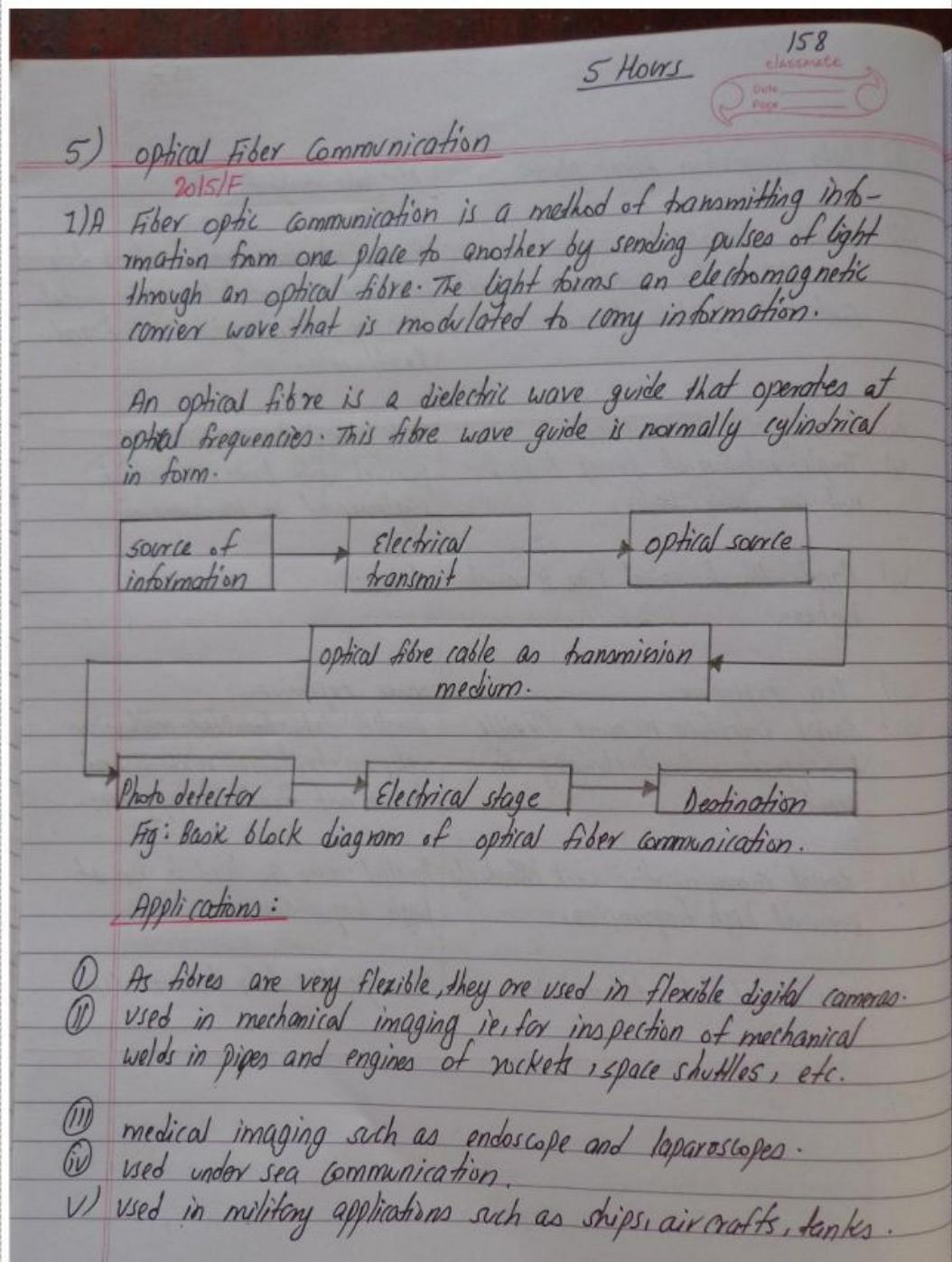
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① Data congestion takes place.	No data congestion.
⑩ serial interface consists of an I ₂ C bus, SPI bus or synchronous serial control and data lines.	Parallel interface consists of 8 data pins and 3 control lines. The control lines are enable, register select and Read/write.
iv) Low speed transmission.	High speed transmission.
v) Implementation of serial links is not an easy task.	parallel data links are easily implemented in Hardware.
vi) Bandwidth of serial wire is much higher.	lower.
vii) less expensive	more expensive.
viii) serial interface is more flexible to upgrade without changing the Hardware.	parallel data transfer mechanism rely on hardware resources and hence not flexible to upgrades.
ix. serial communication work effectively even at high frequencies.	parallel buses are hard to run at high frequencies.
x. uses serial data transmission.	uses parallel data transmission.

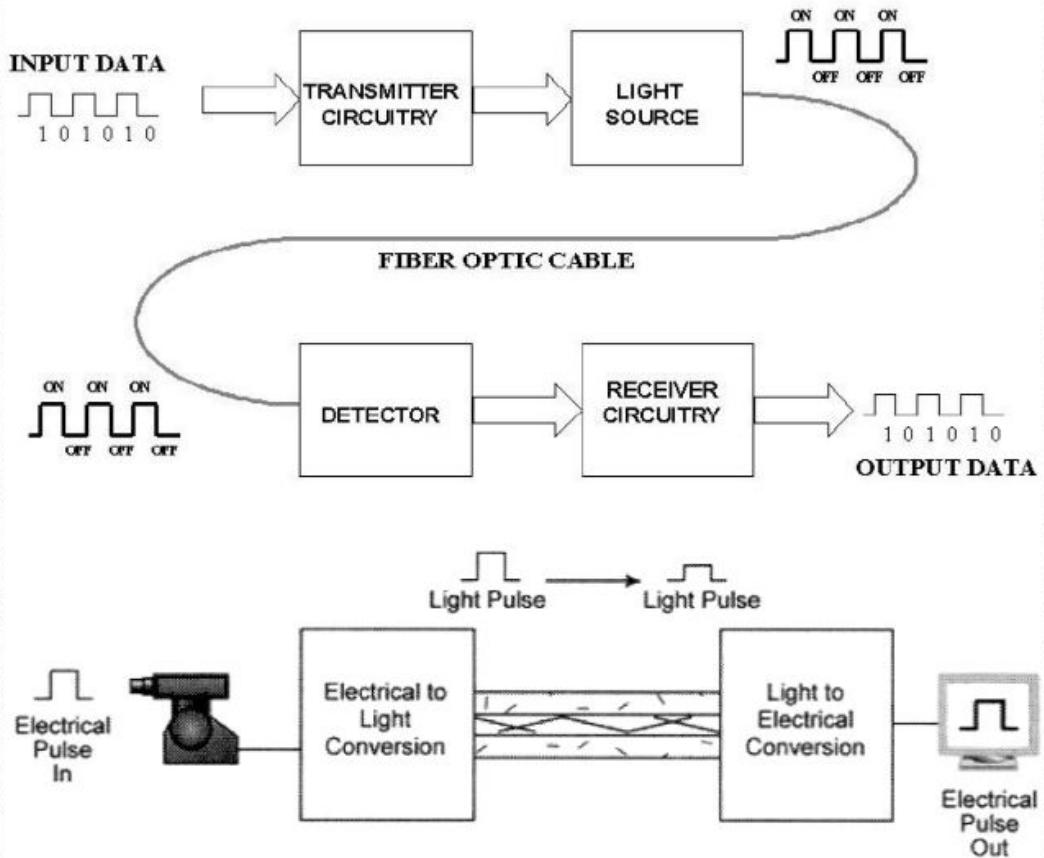
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Fiber optic communication system

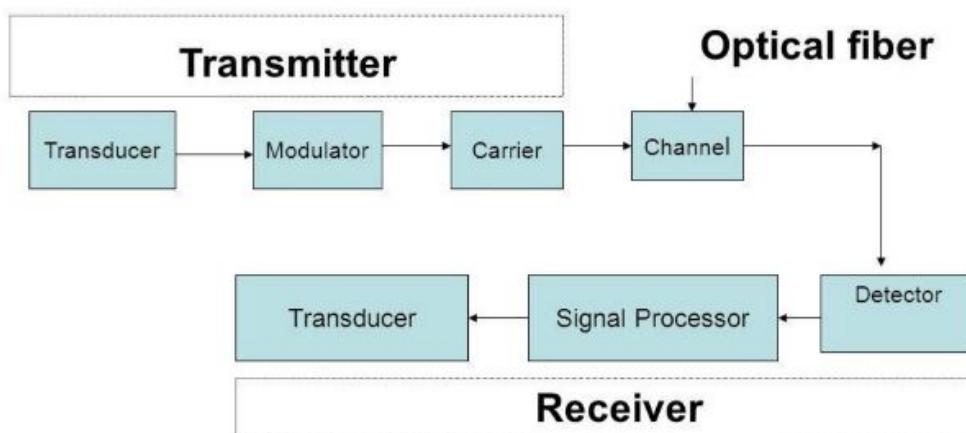
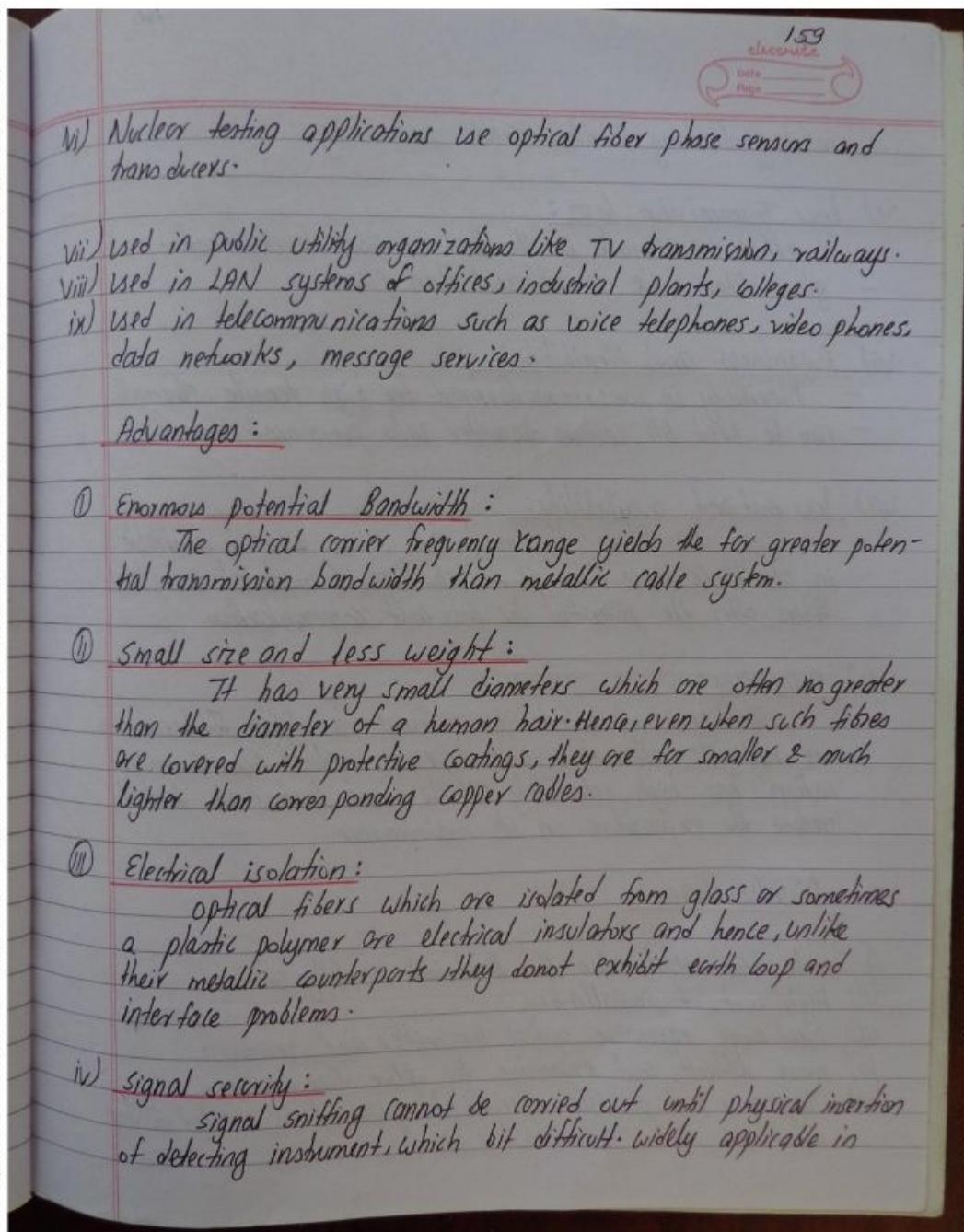


Fig : Block diagram of communication system

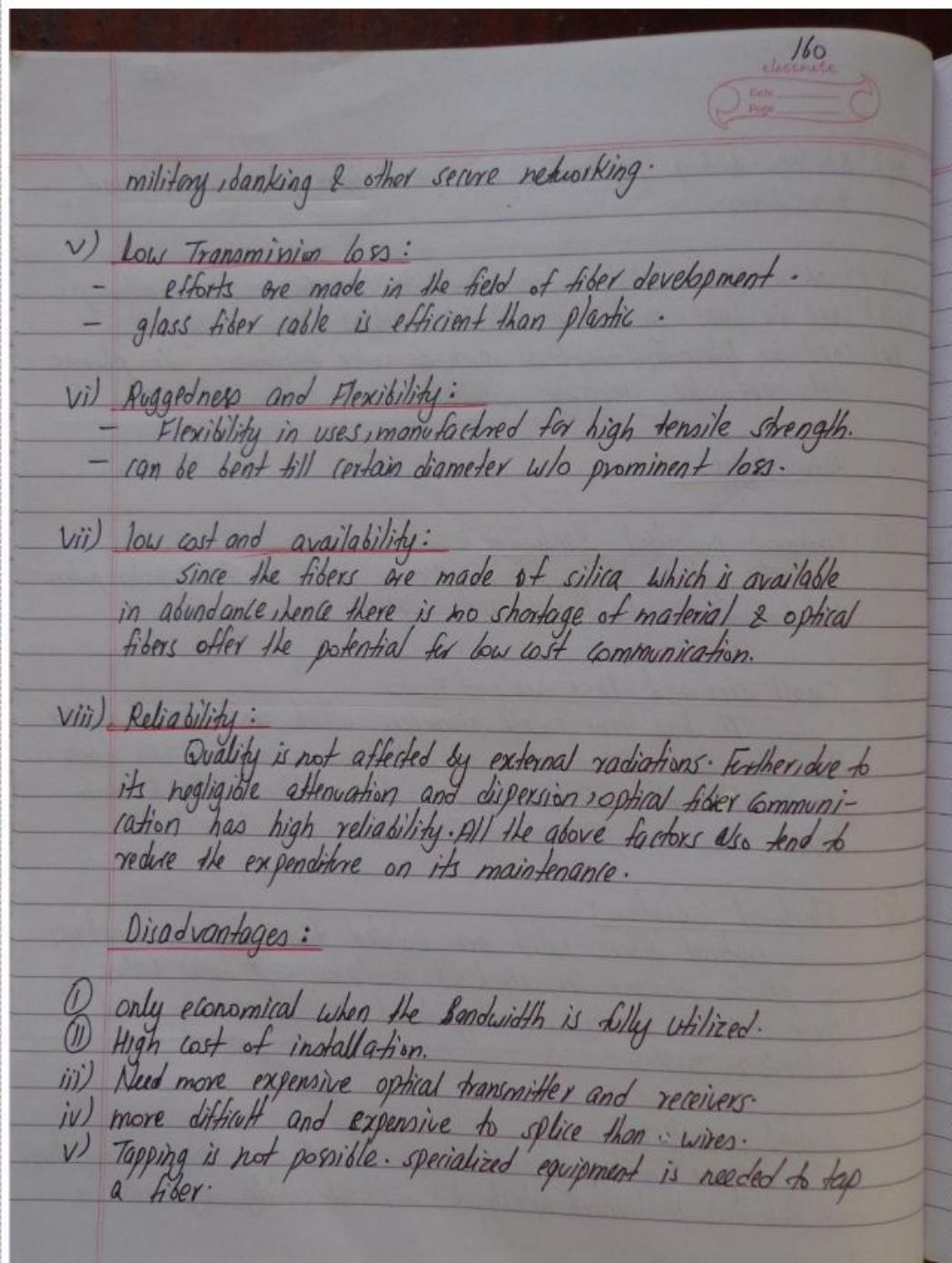
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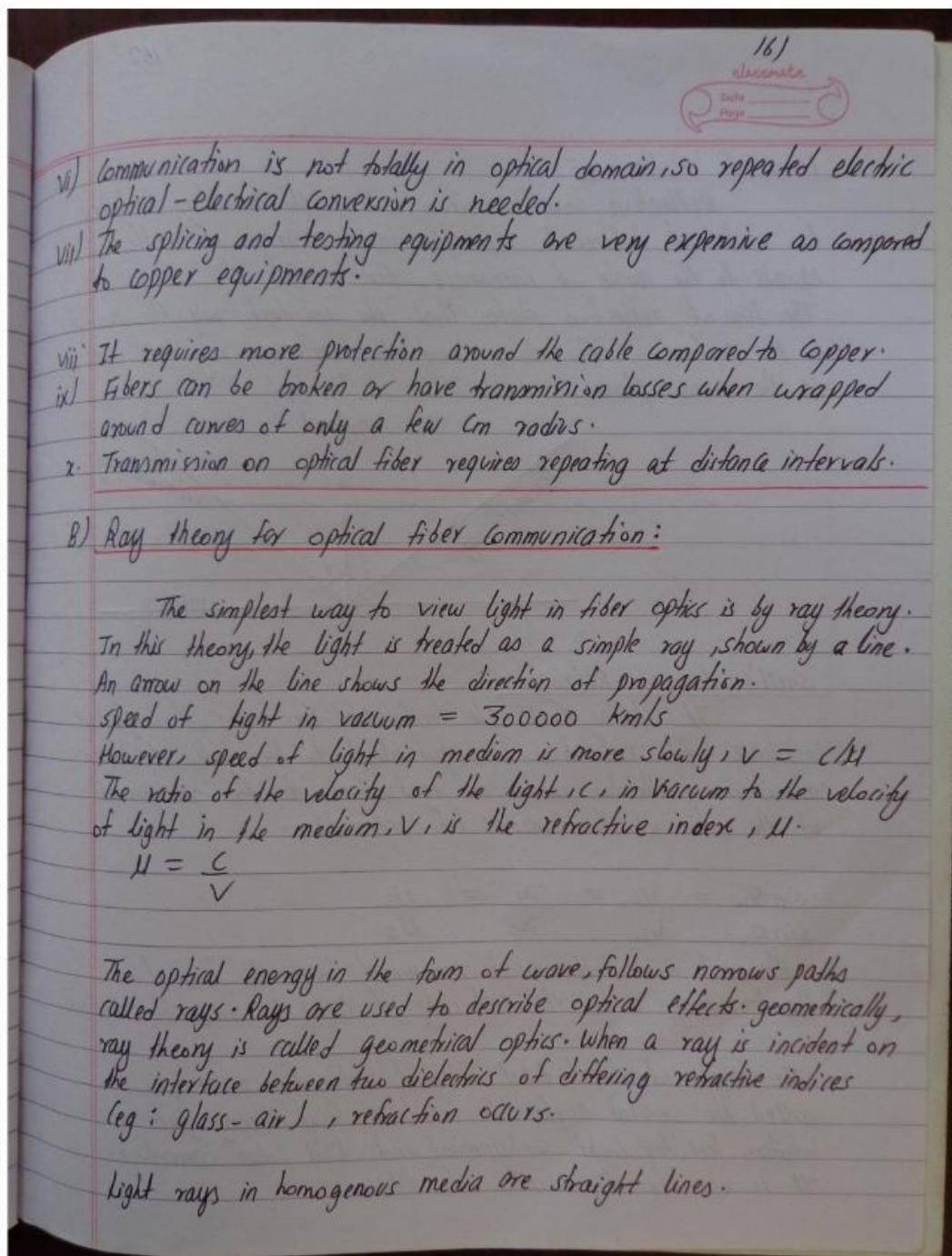
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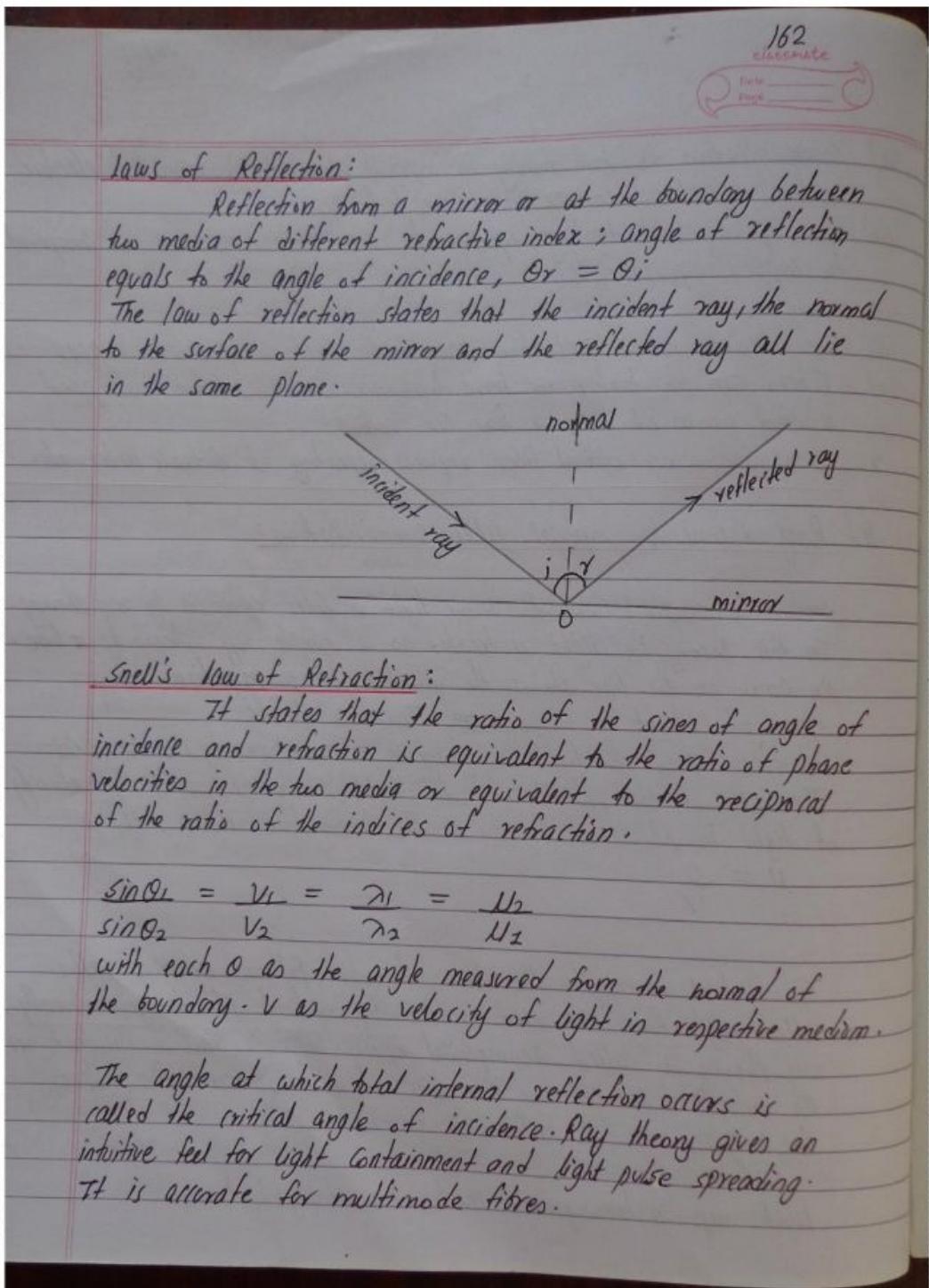
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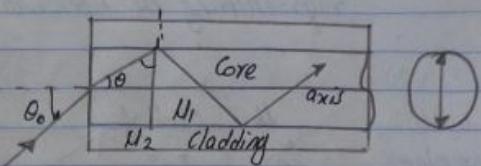
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For the meridional ray,

Snell's law at the fiber enter,
 $\mu_0 \sin \theta_0 = \mu_{\text{core}} \sin \theta_i$



If the ray is refracted from the core to the cladding, then according to Snell's law,

$$\sin \theta_i = \frac{\mu_{\text{cladding}}}{\mu_{\text{core}}} \cdot \sin \theta_r$$

$$\text{if } \sin \theta_i > \frac{\mu_{\text{cladding}}}{\mu_{\text{core}}} -$$

then, $\sin \theta_r \leq 1$, there is no tunneling from core to cladding.
since, $\theta + \theta_i = 90^\circ$; we have,

$$\begin{aligned} \sin \theta_0 &= \mu_{\text{core}} \cdot \sin \theta_i \\ &= \mu_{\text{core}} \cdot \cos \theta_r \\ &= \mu_{\text{core}} \cdot \sqrt{1 - \sin^2 \theta_i} = \sqrt{\mu_{\text{core}}^2 - \mu_{\text{cladding}}^2} \end{aligned}$$

Hence,

Total internal reflection occurs if,

$$\therefore \sin \theta_0 \leq \sqrt{\mu_{\text{core}}^2 - \mu_{\text{cladding}}^2} = \mu_{\text{core}} \sqrt{1 - \left(\frac{\mu_{\text{cladding}}}{\mu_{\text{core}}} \right)^2}$$

c) Numerical Aperture and acceptance angle : 20.15%

The ray which passes through axis of the fiber core is called meridional ray.

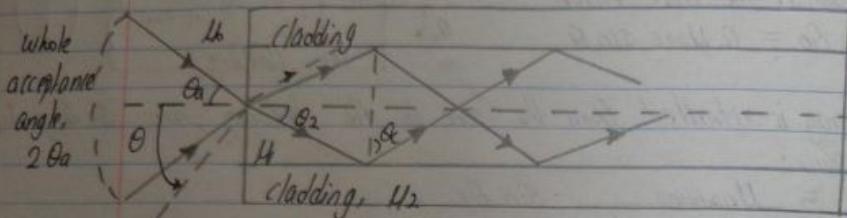
Acceptance angle, θ_a is the maximum angle over which light

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rays entering the fiber will be guided along its core.



The sine of that acceptable angle is called the numerical aperture and it is essentially determined by the refractive index contrast between core and cladding of the fiber, assuming that incident beam comes from air to vacuum.

$$\text{Acceptance angle, } \theta_a = \sin^{-1} (\sqrt{n_1^2 - n_2^2})$$

n_1 = refractive index of glass fiber core

n_2 = " " " quartz fiber cladding.

Numerical aperture (NA) is used to describe the light gathering or light collecting ability of an optical fiber. In optics, the numerical aperture (NA) of an optical system is a dimensionless number that characterizes the range of angles over which the system can accept or emit light.

$$\therefore NA = \sin \theta_a = (\sqrt{n_1^2 - n_2^2})$$

\therefore Full acceptance angle = $2 \theta_a$.

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D. Fiber Types, properties and Applications :

Based on the refractive index profile, there are two types of fibers :

- step index fibers
- Graded " "

Based on the modes, there are two types of fibers

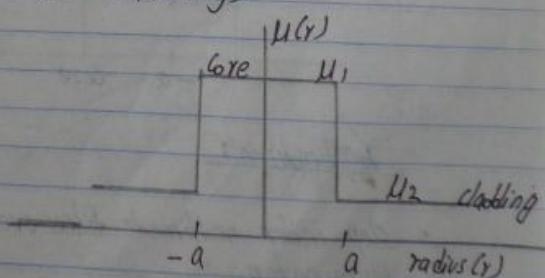
- single mode fibers
- Multimode fibers

① Step index fibers : 2017/18

This is called so because the refractive index of the fibre steps up as we move from the cladding to the core and this type of fibre allows single mode to propagate at a time due to very small diameter of its core.

Cladding \perp Core.
i.e,

$$N(r) = \begin{cases} n_1 & \text{for } r \leq a \quad (\text{Core}) \\ n_2 & \text{for } r \geq a \quad (\text{cladding}) \end{cases}$$



② Graded Index Fibers :

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Graded index fibers do not have a constant refractive index in the core but a decreasing core index $\mu(r)$ with radial distance from a maximum value of μ_1 at the axis to a constant value μ_2 beyond the core radius 'a' in the cladding.

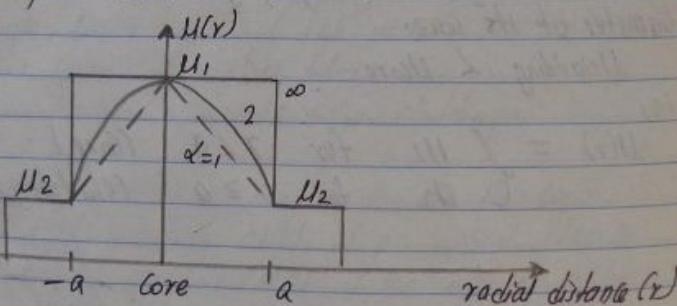
The index variation (r) :

$$\mu(r) = \begin{cases} \mu_1 (1 - 2\Delta (r/a)^2)^{1/2} & r \leq a \text{ (core)} \\ \mu_1 (1 - 2\Delta)^{1/2} = \mu_2 & r \geq a \text{ (cladding)} \end{cases}$$

where,

Δ = relative refractive index difference.

α = profile parameter which gives characteristics refractive index profile of the fiber core.



Applications :

- Step index multimode fibers are mostly used for imaging and illumination.
- Graded index multimode fibers are used for data communication

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yours

and network carrying signals for typically no more than a couple of km.

(iii) single mode Fibers : *2014/5, 2017/1s*

- carries light pulses along single path. only the lowest order mode (fundamental mode) can propagate in the fiber and all higher order modes are under cut off condition (non-propagating).

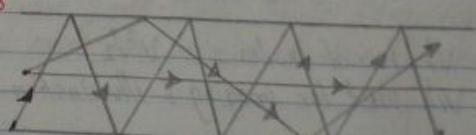
Source → → → → → → → →

Fig: single mode.

- only one path is available
- Core diameter is small , 8 - 10 μm
- No dispersion
- Higher Bandwidth
- Fabrication is costly and difficult.
- Used for long haul communication.
- supports one mode of propagation.
- Optical source - LASER

(iv) multimode Fiber: *2014/5, 2017/1s*

- more than one path is available.
- High dispersion.
- Lower Bandwidth.
- Core diameter is higher.
- Fabrication is less difficult and not costly.



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	<i>yours</i>	
	• used for short distance communication.	
E)	Step index Fiber <small>2017/18</small>	Graded index fiber
i)	The refractive index of the core is uniform throughout & undergoes an abrupt change at the core cladding boundary.	The refractive index of the core is made to vary gradually such that it is maximum at the center of the core.
ii)	Diameter of the core is about 50 - 200 μm in the case of multimode fiber & 10 μm in case of single mode fiber.	Diameter of the core is about 50 μm in the case of multimode fiber.
iii)	path of light propagation is zig-zag in manner.	path of light is helical in manner.
iv)	Attenuation is more for multimode step index fiber but for single mode it is very less.	Attenuation is less.
v)	$\text{no. of modes} = V^2/2$	$\text{no. of modes} = V^2/4$
vi)	used for imaging & illumination.	used for data communication & networks over moderate distance.
vii)	modal dispersion is found.	solves the problem of modal dispersion.

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F. Single mode Fiber

- i) core radius is small.
- ii) supports one mode of propagation.
- iii) optical source - LASER.
- iv) Launching of optical power into fiber is difficult as the core radius is small.

v) Application: submarine cable system.

- vi) cheaper
- vii) Higher bandwidth.
- viii) Do not exhibit dispersion.
- ix) High cost connector.

Multimode Fiber

- core radius is large.
- supports hundreds of modes.
- optical source - LED.
- The launching of optical power into fiber is easier as the core radius is large.

Application: Telephone links.

- costly.
- limited bandwidth.
- limited by modal dispersion.
- low cost connectors.

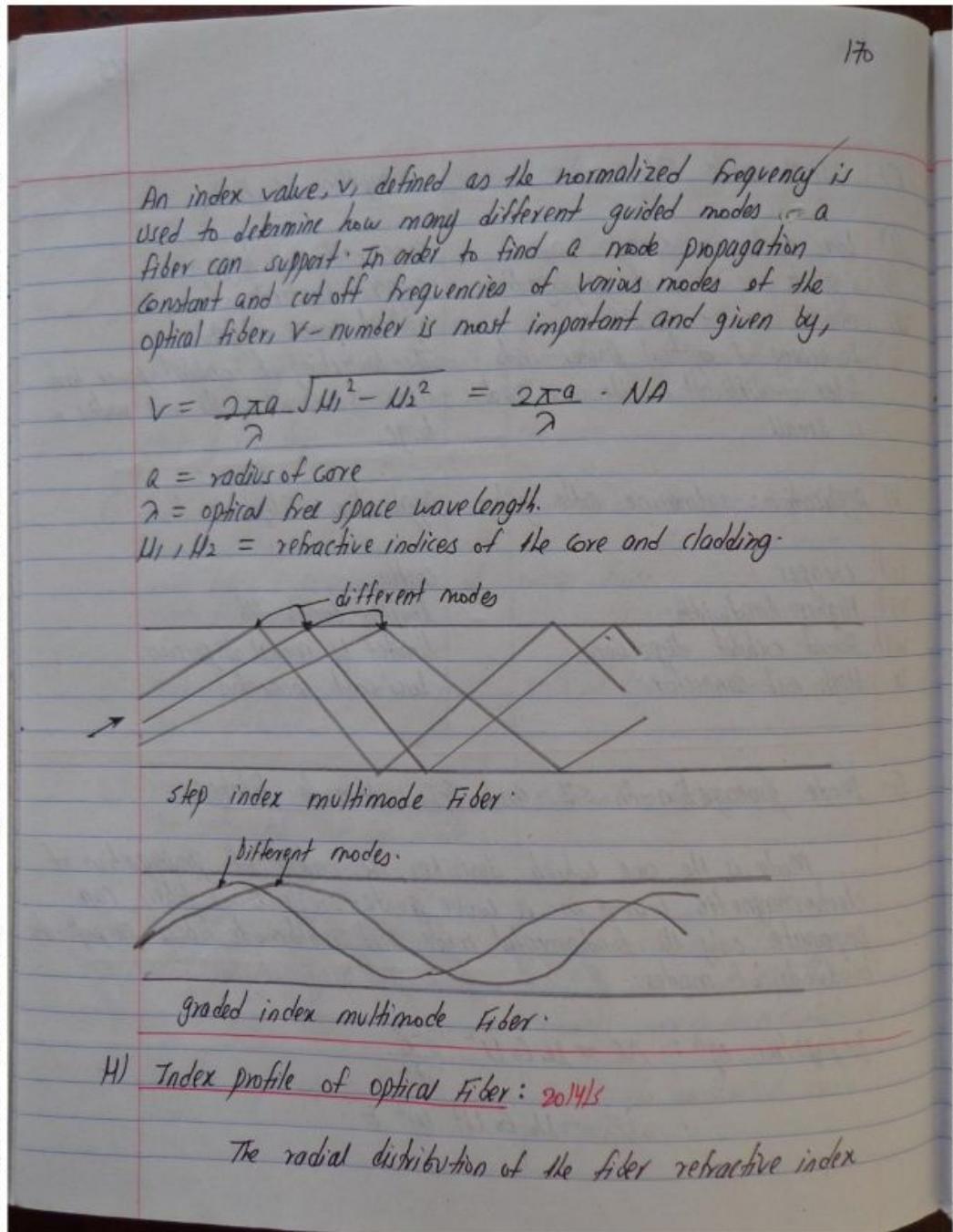
G. Mode propagation in SI and GI multimode fibers:

Mode is the one which describes the nature of propagation of electromagnetic waves in a wave guide. Single mode fibers can propagate only the fundamental mode and multimode fibers can propagate hundreds of modes.

$$\text{Propagation eqn: } \Delta E = \mu_0 \epsilon_0 N^2 \frac{\partial^2 E}{\partial t^2}$$
$$= -\mu_0 \epsilon_0 N^2 \omega^2 E$$

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is called the index profile. In the case of a slab waveguide, the transverse refractive index is called the index profile. The index profile determines guiding properties of the fiber or slab waveguide. In general, the core region has a higher index than the cladding region. However, the index profile can have regions where the index is lower than the cladding value. Modern fiber or slab waveguide designs are based on index profile that assure proper operation within a range of wavelengths.

Constant index profile, $n(x) = \text{constant}$

Linear " ", $n(x) = n(0) + \frac{\pi \cdot b(w)}{w} - n(0)$

Parabolic index profile :

$$n(x) = [n(w) - n(0)] \cdot \left(\frac{x}{w}\right)^2 + n(0)$$

where, $n(0), n(w)$ is the refractive index at $x=0$ and $x=w$.

I. Attenuation, Dispersion, Bend loss in optical Fiber: 2015ks

Attenuation is defined as the ratio of optical output power to the input power in the fiber of length L.

$$\alpha = 10 \log_{10} \left(\frac{P_o}{P_i} \right)$$

P_i = input power

P_o = output "

α = attenuation constant

Attenuation means loss of light energy as the light pulse travels from one end of the cable to the other. Also known as signal or

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power loss. It decides the number of repeaters required between transmitter and receiver. Attenuation is directly proportional to the length of the cable. Attenuation in optical fiber take place due to elements like couplers, splices, connector and fiber itself. Modern fiber material is very pure, but there is still some attenuation.

Bending losses: Radiative loss:

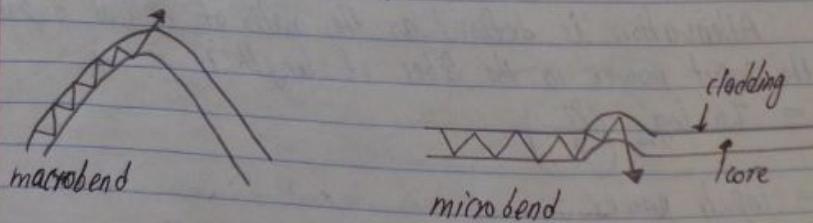
The loss which exists when an optical fiber undergoes bending is called bending losses. There are two types of bending:

① macroscopic bending:

Bending in which complete fiber undergoes bends which causes certain modes not to be reflected and hence, causes loss to the cladding.

② Microscopic Bending:

Either the core or cladding undergoes slight bends at its surfaces. It cause light to be reflected at angles when there is no further reflection.



Microbending is a much more critical feature and can be a major cause of cable attenuation. Macro bending losses are called large curvature radiation losses.

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Dispersion :

As a pulse travels down a fiber, dispersion causes pulse spreading. This limits the distance and the bit rate of data on an optical fiber. As an optical signal travels along the fiber, it becomes increasingly distorted. This distortion is a sequence of intermodal and intramodal dispersion.

Intermodal dispersion :

Pulse broadening due to intermodal dispersion results from the propagation delay difference between modes within a multimode fiber.

Intramodal dispersion / chromatic dispersion :

It is the pulse spreading that occurs within a single mode.

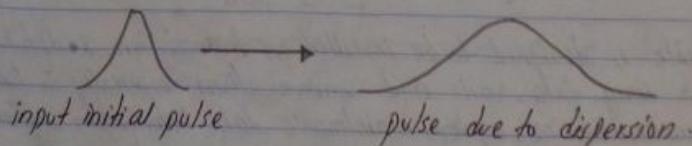
- material dispersion
- waveguide "

Material dispersion / spectral dispersion :

Wavelength based effect caused by glass of which fiber is made.

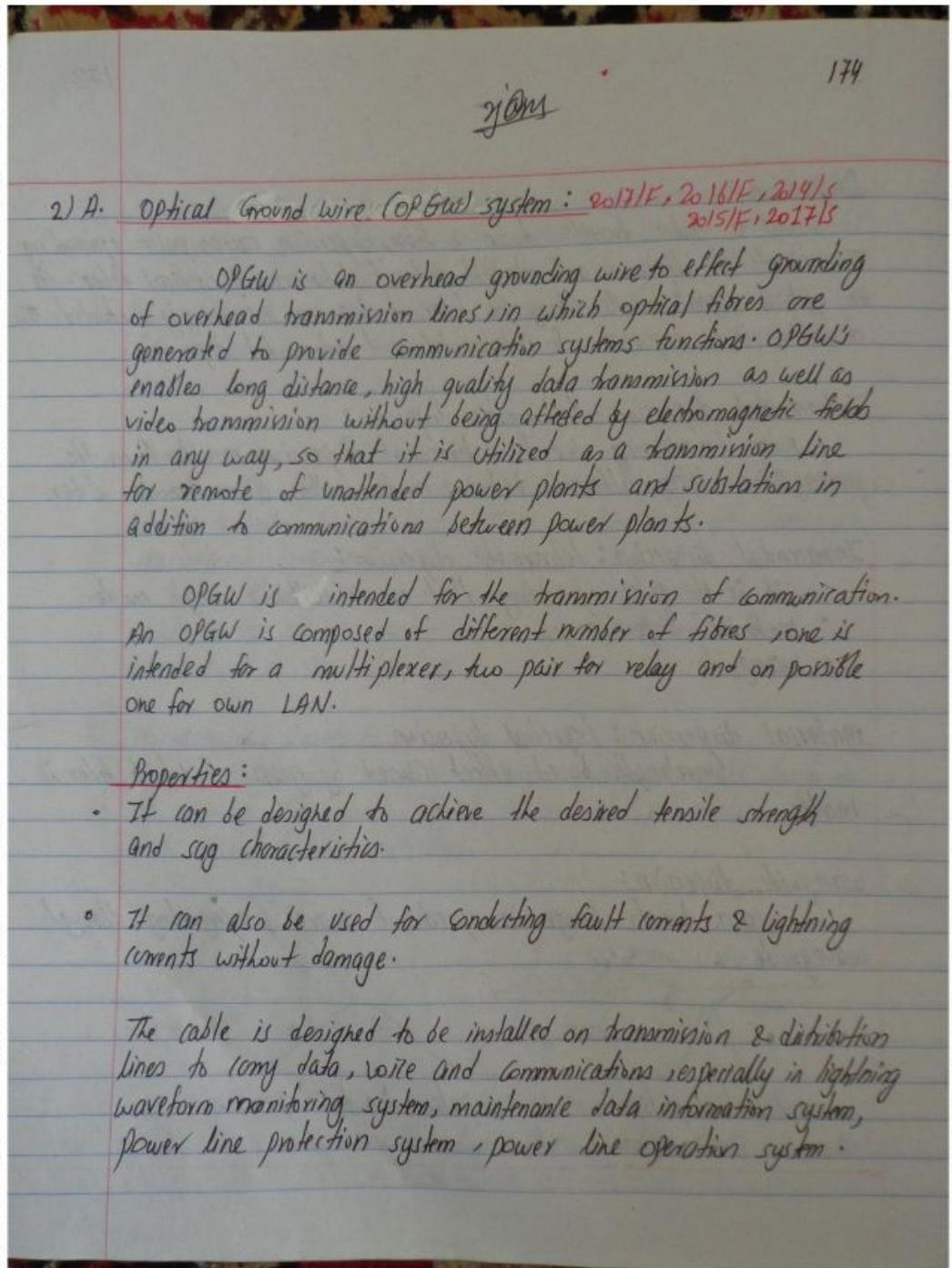
Waveguide dispersion :

Occurs due to change in speed of wave propagating through waveguide.



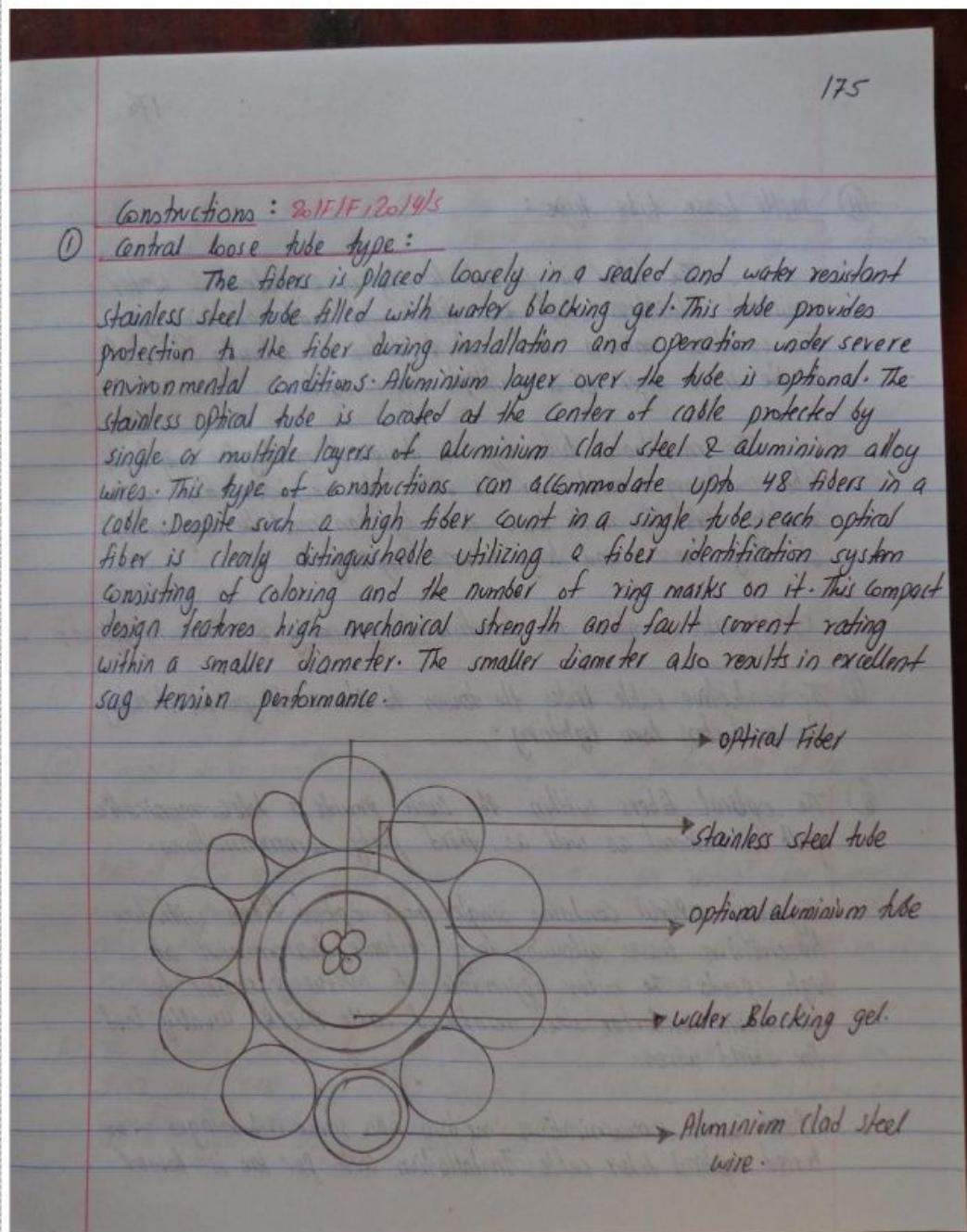
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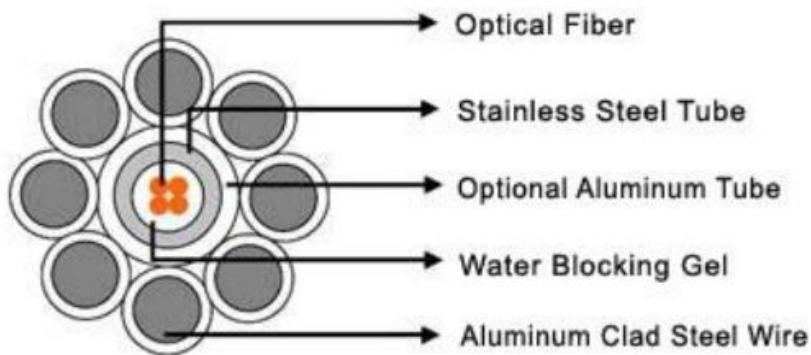
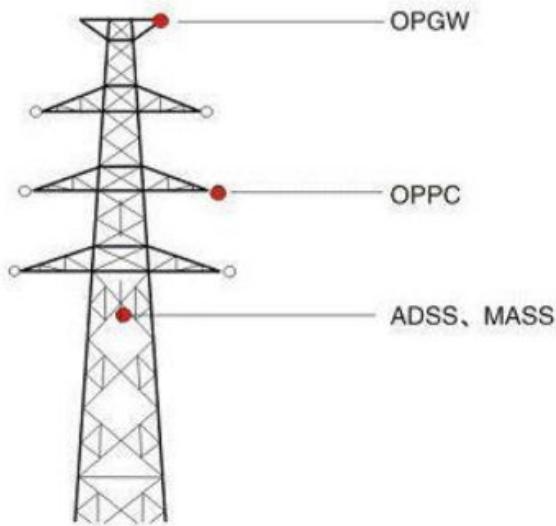
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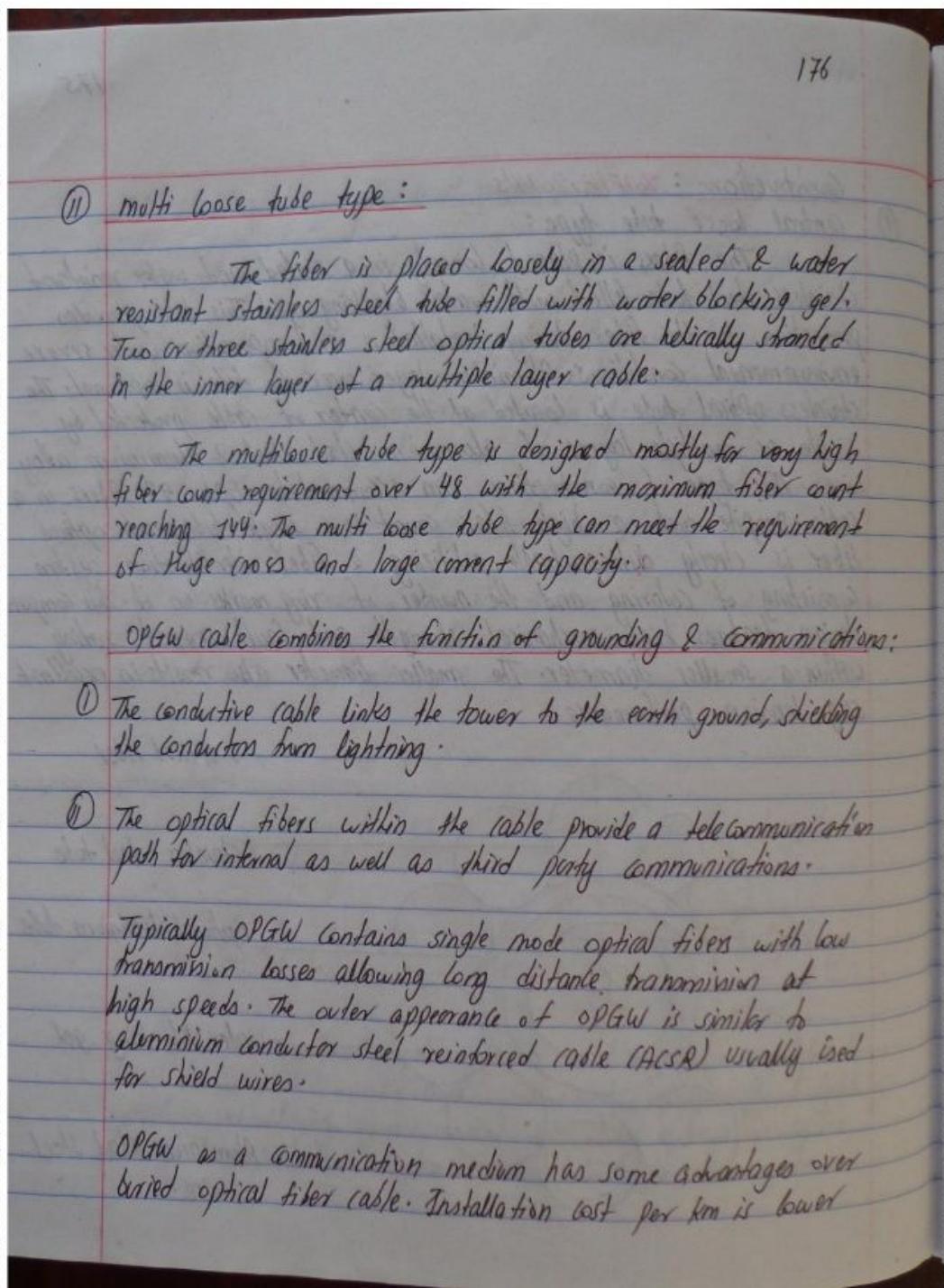


Central Loose Tube Type



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than a buried cable. Effectively, the optical circuits are protected from accidental contact by the high voltage cables below. (and by the elevation of the OPGW from ground).

Applications : 2014/s, 2017/F

- ① OPGW is a dual functioning cable performing the duties of a ground wire and also providing a path for the transmission of voice, video or data signals. The fibers are protected from environmental conditions like lightning, short circuit, loading to ensure reliability and longevity.

Properties :

- ① High load, long span capability
② Unique design has maximum allowable tension to control fiber strain.
③ Compact design results in excellent sag tension performance of the cable.

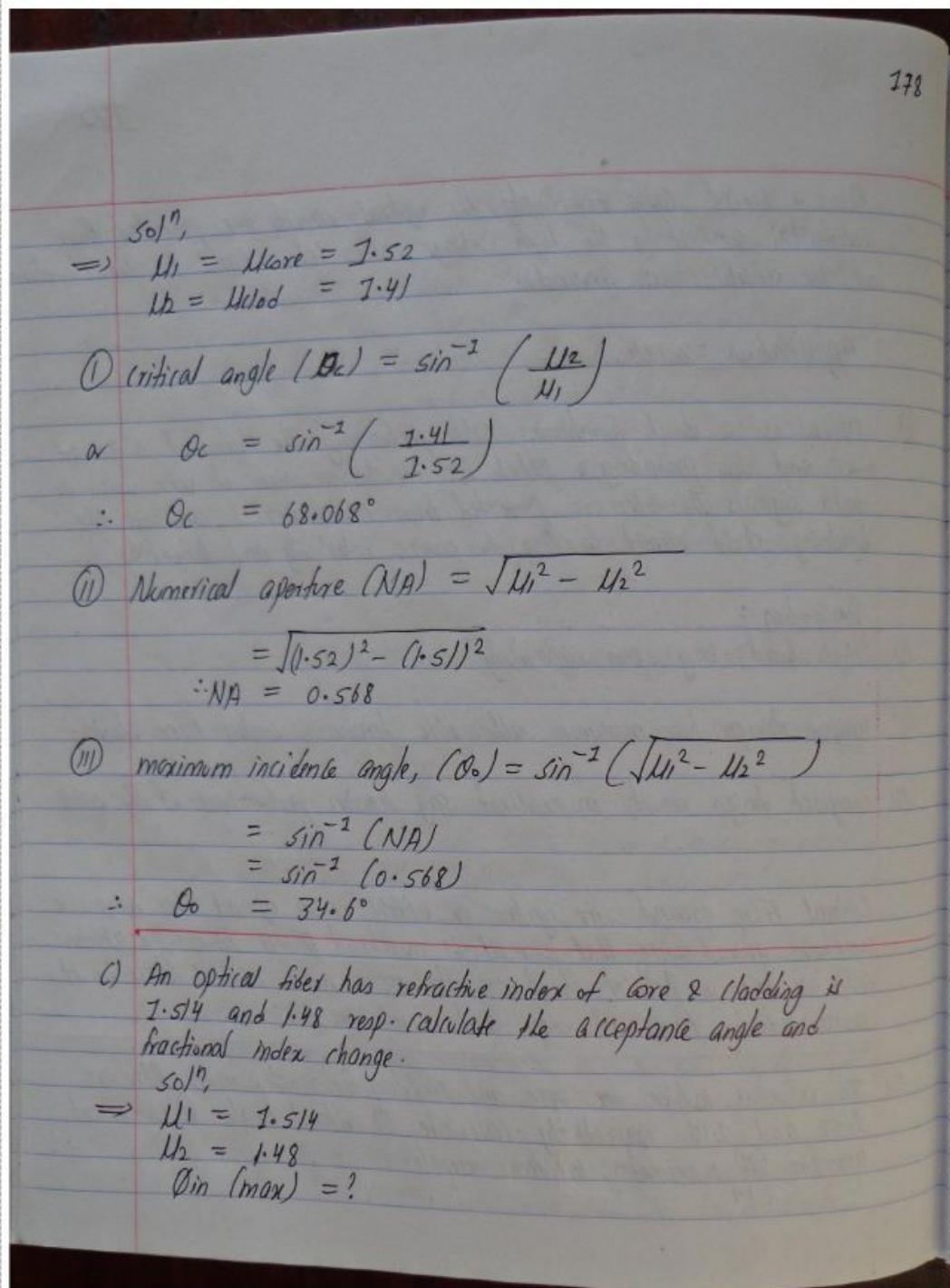
Optical Fibre ground wire system or OPGW is an optical fiber composite overhead ground wire that is used in overhead power lines. The OPGW cable contains a tubular structure with one or more optical fibers in it, surrounded by layers of steel and aluminium wire.

- B) The refractive indices for core and cladding for a step index fibre are 1.52 and 1.41 respectively - calculate ① critical angle ② numerical aperture ③ maximum incidence angle.

sol:

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$\Delta = ?$

we have,

① acceptance angle, $\phi_{in(max)} = \sin^{-1}(n_1^2 - n_2^2)^{1/2}$

$$= \sin^{-1}(1.514^2 - 1.48^2)^{1/2}$$
$$= \sin^{-1}(0.316)$$
$$\therefore \phi_{in(max)} = 18^\circ 42'$$

② $\Delta = \frac{n_1^2 - n_2^2}{2n_1^2}$

$$= \frac{(1.514)^2 - (1.48)^2}{2 \times (1.514)^2}$$
$$\therefore \Delta = 0.02220$$

D. Consider a multimode fiber that has a core refractive index of 1.48 and a core cladding index difference of 3 percent ($\Delta = 0.03$). find,

① Numerical Aperture.

② acceptance angle 2016/F

③ critical angle 50° ,

\Rightarrow Core refractive index (n_1) = 1.48
cladding " " (n_2) = ?

Core-cladding index difference (Δ) = 0.003

We know,

① $\Delta = \frac{n_1^2 - n_2^2}{2n_1^2} \approx \frac{n_1 - n_2}{n_1} = \frac{\Delta n}{n_1}$

or $0.03 = \frac{1.48^2 - n_2^2}{2 \times 1.48^2}$

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$$\text{or } 0.13142 = 2.1840 - \mu_2^2$$

$$\therefore \mu_2 = 1.4349$$

$$\begin{aligned} \text{① critical angle, } (\theta_c) &= \sin^{-1} \left(\frac{\mu_2}{\mu_1} \right) \\ &= \sin^{-1} \left(\frac{1.434}{1.48} \right) \\ &= 75.677^\circ \end{aligned}$$

② acceptance angle :

$$\theta_{\text{accept}} = \sin^{-1} (\mu_1^2 - \mu_2^2)^{1/2}$$

$$= \sin^{-1} (1.48^2 - 1.434^2)^{1/2}$$

$$= \sin^{-1} (0.36612)$$

$$\therefore \theta_{\text{accept}} = 21.4765^\circ$$

③ Numerical aperture (NA) = $\sin \theta_{\text{in}} = \mu_1^2 - \mu_2^2$

$$\therefore \text{NA} = 0.36612$$

E. Draw the simplified fiber optic communication link and briefly explain the working of each components used in order to copy the information to the destination. 2016ks

⇒ Copy Block diagram from page No : 158

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Main basic elements of fiber optic communications system are,

- (i) Compact light source
- (ii) Photo detector
- (iii) low loss optical fiber.

i) Compact light source:

Depending on the applications like local area networks and the long haul communication systems, the light source requirement vary. The requirement of the sources include power, speed, spectral line width, noise, ruggedness, cost and so on. Two components are used as light sources: Light emitting diodes and LASER diodes.

For longer distances and high data rate transmission laser diodes are preferred due to high power, high speed and narrower spectral line width characteristics.

ii) Low loss optical fiber:

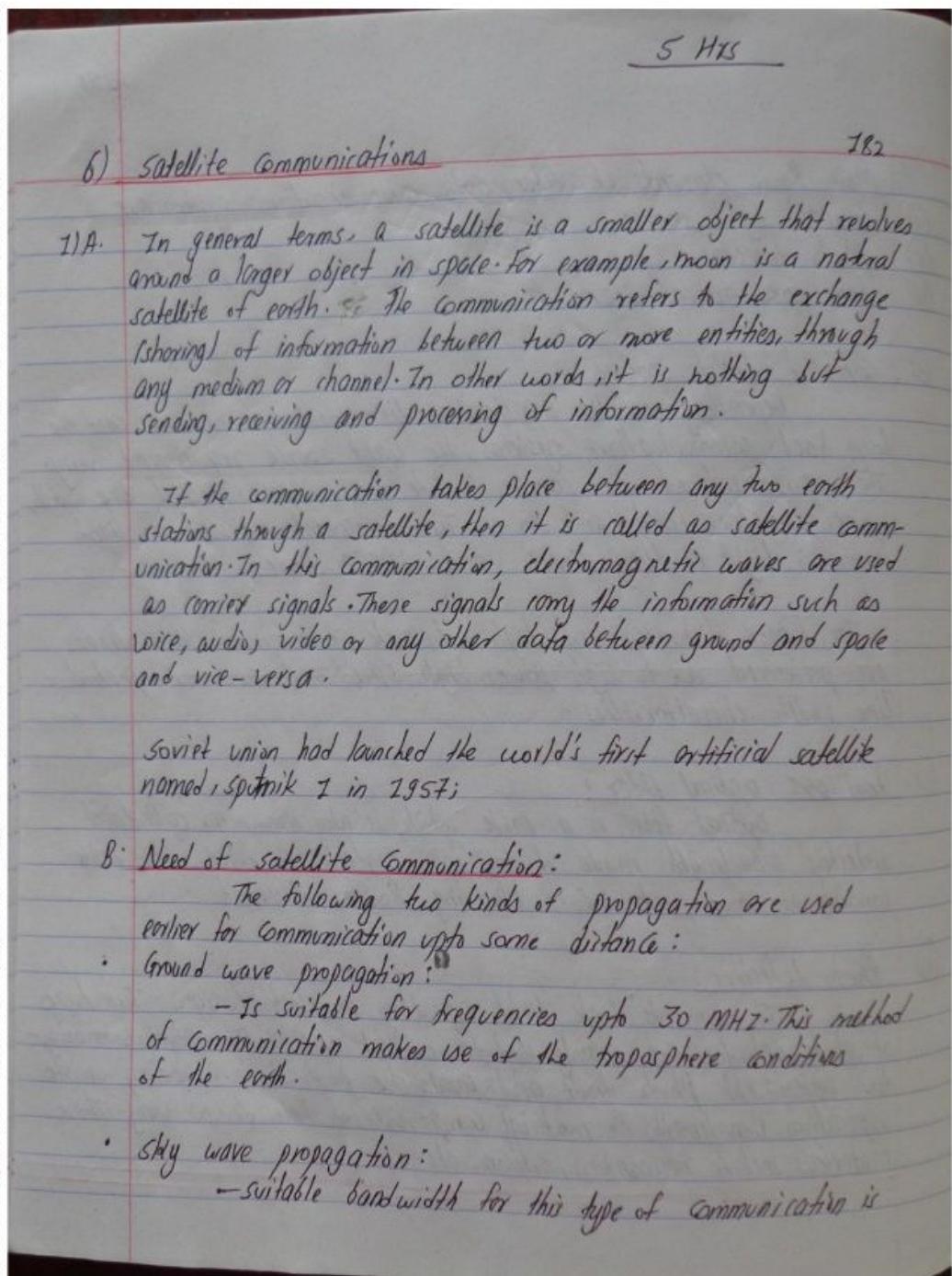
Optical fiber is a cable which is also known as cylindrical dielectric waveguide made of low loss material. A fiber optic cable consists of four parts: Core, cladding, Buffer, Jacket.

iii) Photo detectors:

It converts back light signal to an electrical signal. Two types of photo detectors are mainly used for optical receiver in optical communication system: PN Photo diode and avalanche photo diode. Depending on the applications wavelength, the material composition of these devices vary. These materials include germanium, silicon etc.

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broadly between 30-40 MHz and it makes use of the ionosphere properties of the earth.

The maximum hop or the station distance is limited to 1500 km only in both ground wave propagation and sky wave propagation, satellite communication overcomes this limitation. In this method, satellites provide communication for long distances, which is well beyond the line of sight.

Since the satellite locate at certain height above earth, the communication takes place between any two earth stations easily via satellite. So, it overcomes the limitation of communication between two earth stations due to earth's curvature.

C. How a satellite works ??

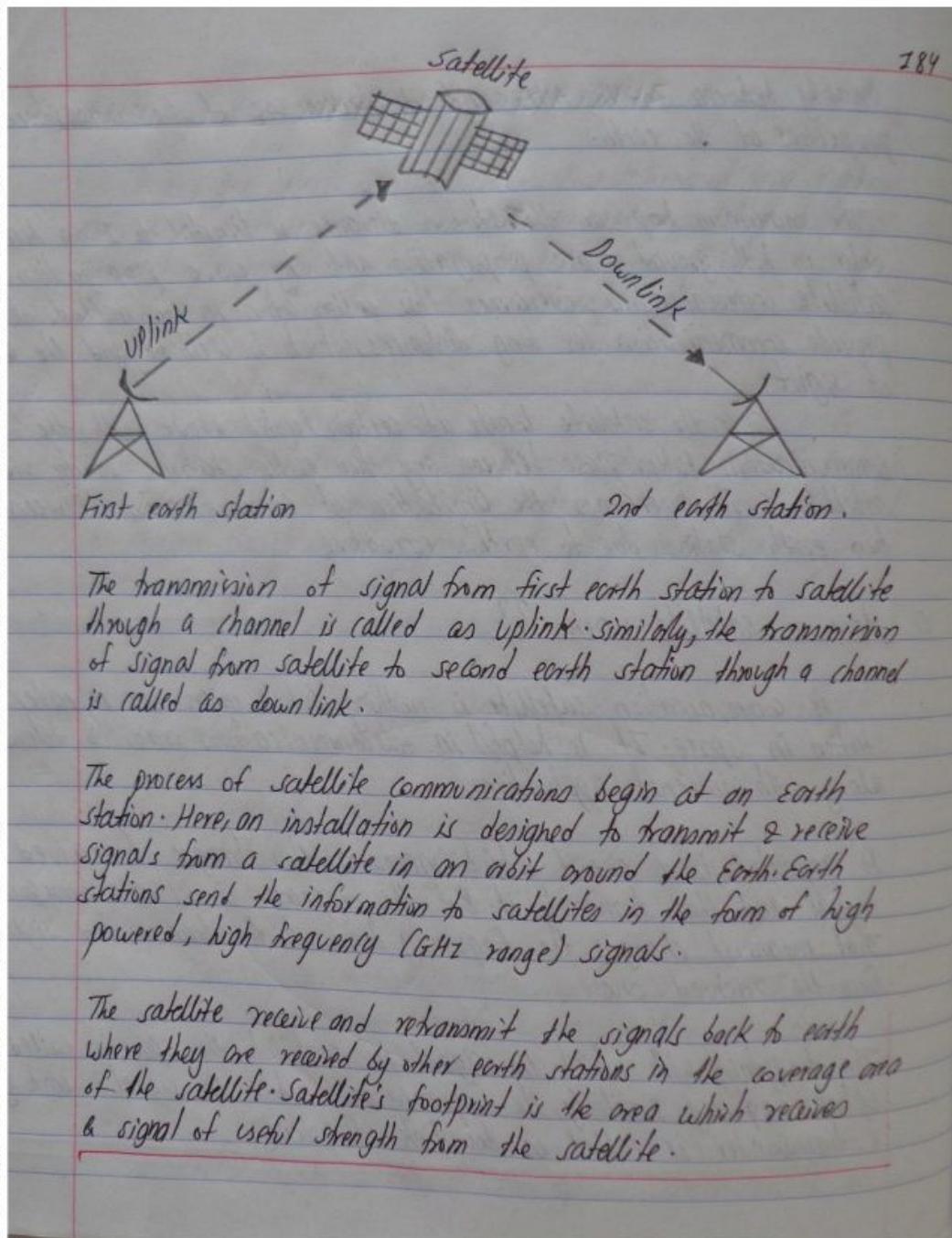
A communication satellite is nothing but a microwave repeater station in space. It is helpful in telecommunications, radio & television along with internet applications.

A repeater is a circuit which increases the strength of received signal and then transmits it. But this repeater works as a transponder. That means, it changes the frequency band of the transmitted signal from the received one.

The frequency with which the signal is sent into the space is called as uplink frequency. Similarly, the frequency with which the signal is sent by a transponder is called as downlink frequency.

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D. Advantages:

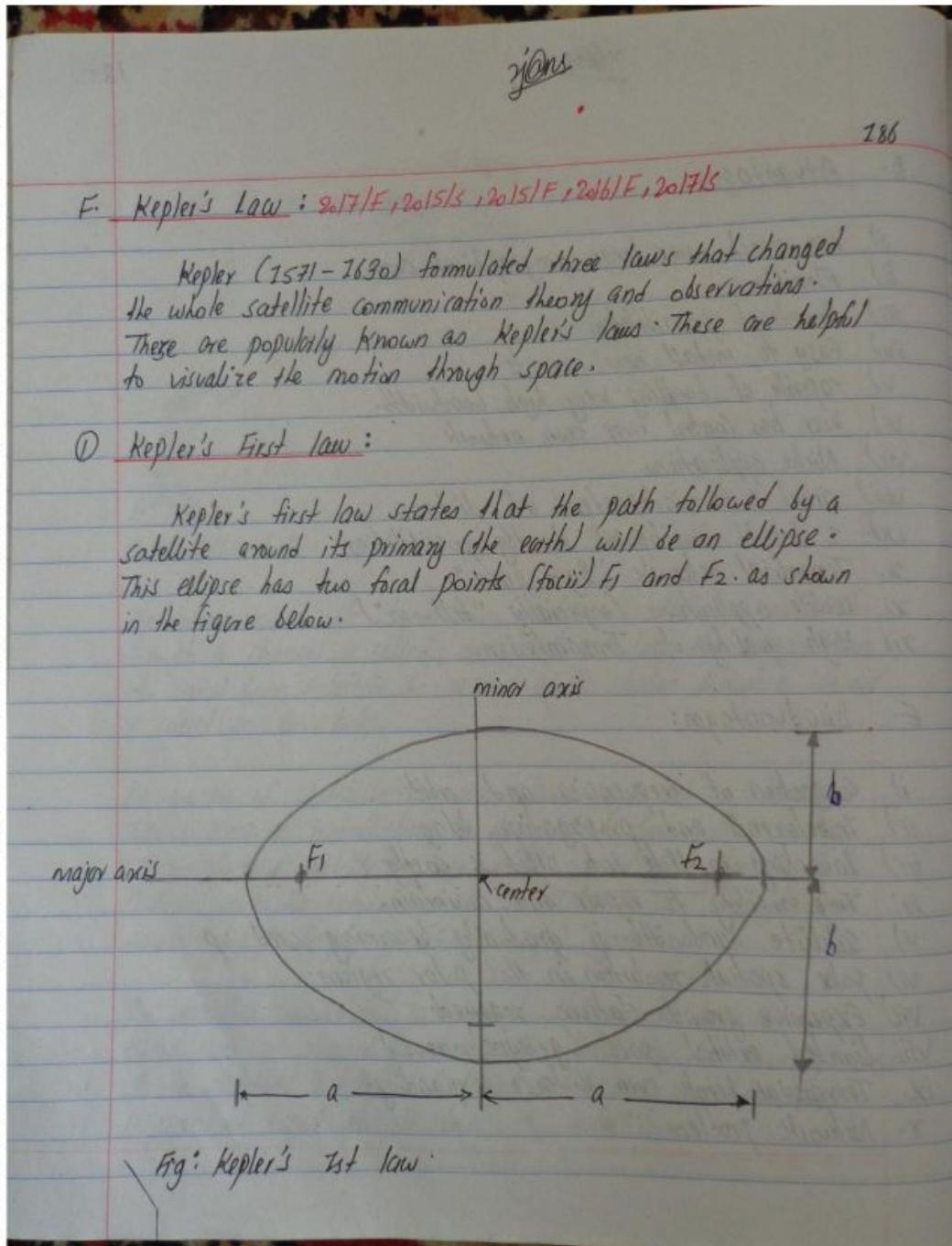
- i) can reach over large geographical area.
- ii) Flexible
- iii) Broadcast possibilities.
- iv) Easy to install new circuits.
- v) capable of handling very high bandwidth.
- vi) User has control over own network.
- vii) Niche applications.
- viii) circuit costs independent of distance.
- ix). 1 for N multipoint standby possibilities.
- x. Terrestrial network "By-pass".
- xi) Mobile applications (especially "fill-in").
- xii) High quality of Transmission.

E. Disadvantages:

- i) Congestion of frequencies and orbits.
- ii) Interference and propagation delay.
- iii) launching satellite into orbit is costly
- iv) Impossibility to repair and maintain.
- v) Satellite bandwidth is gradually becoming used up.
- vi) poor spatial resolution in the polar regions.
- vii) Expensive ground stations required
- viii) Limited orbital space (geosynchronous)
- ix. Terrestrial break even distance expanding.
- x. Network problem.

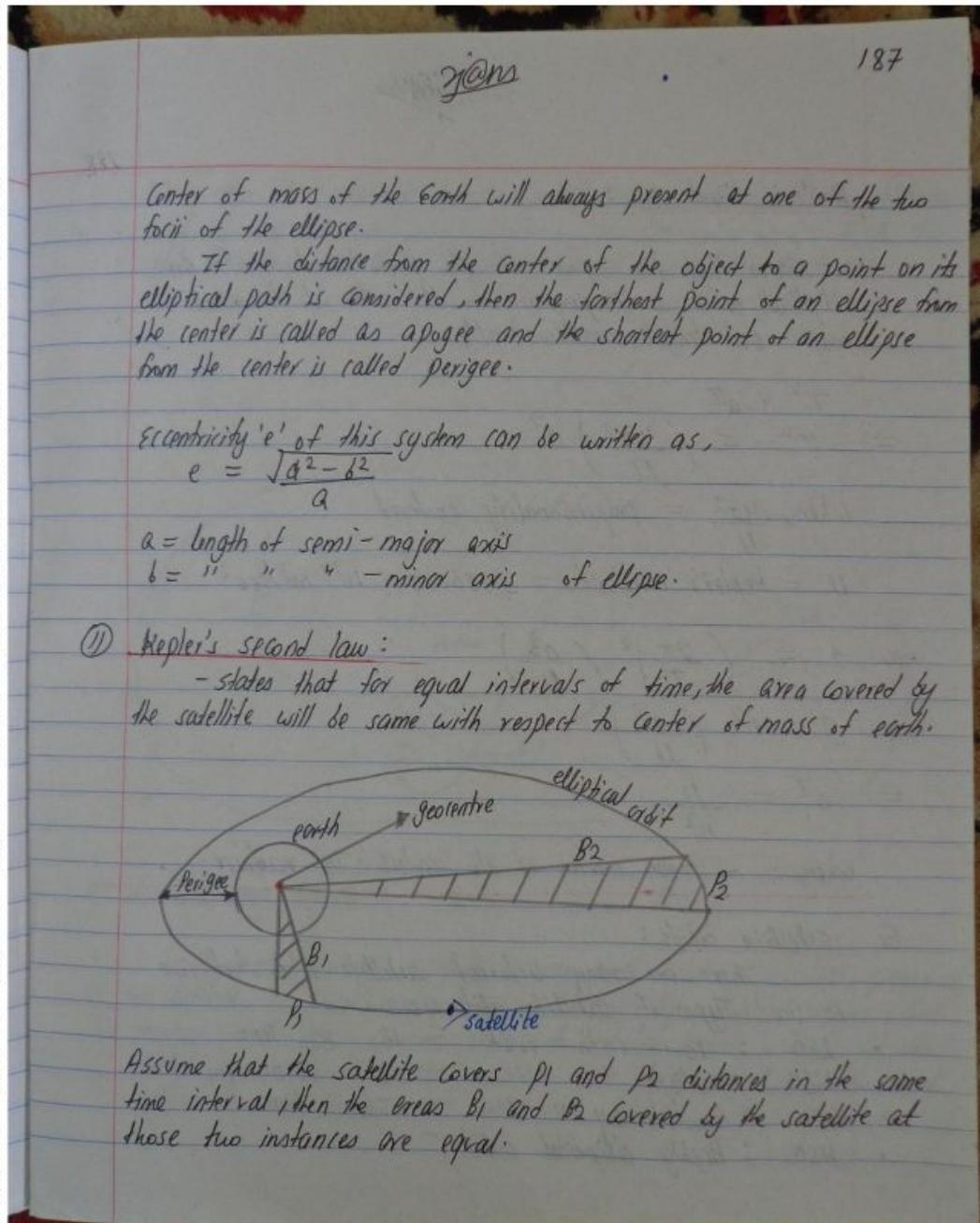
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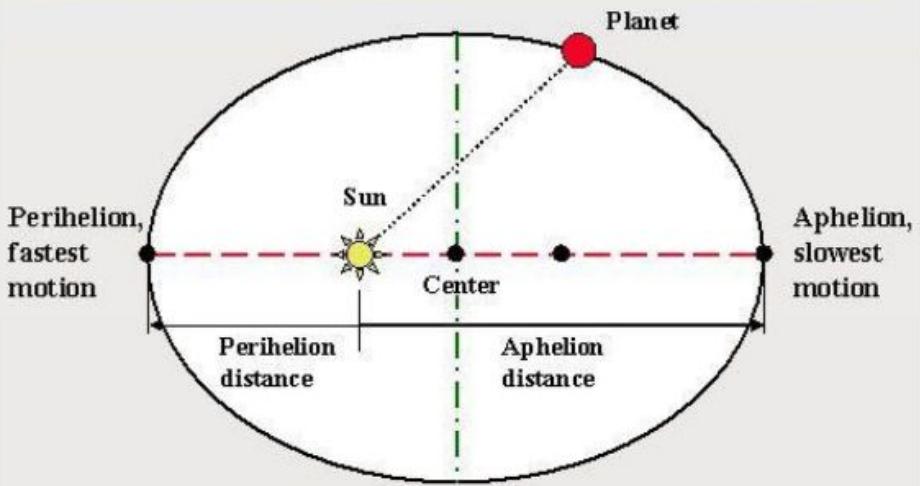
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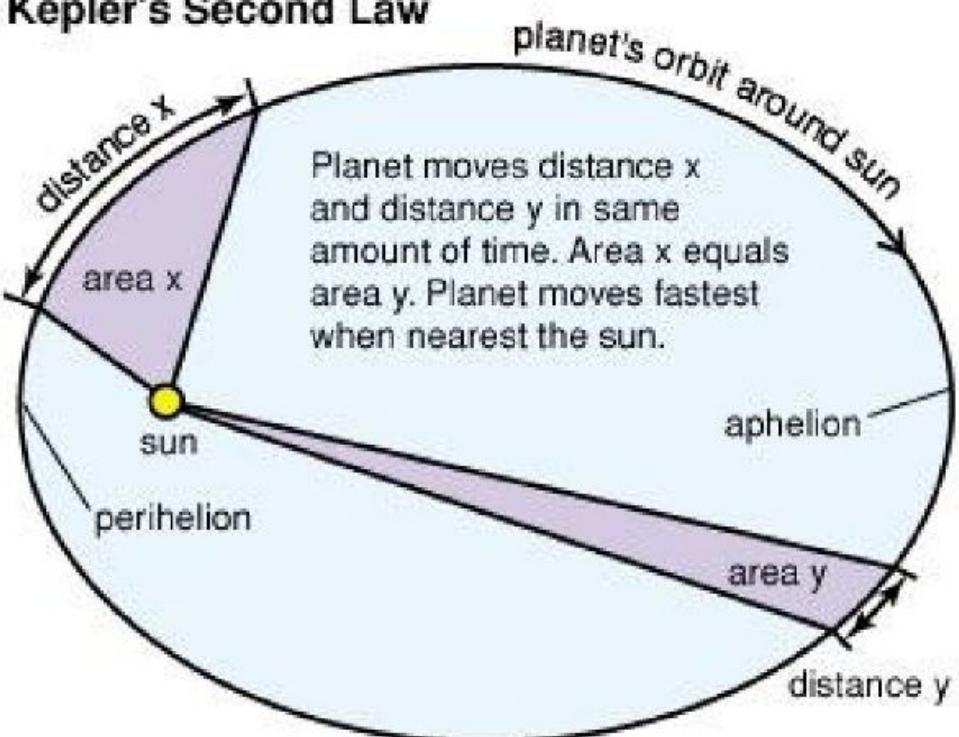
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Kepler's 1st Law



Kepler's Second Law



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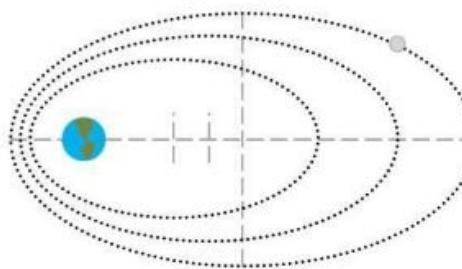
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Kepler's 3rd Law

Kepler's Third Law is a very powerful statement relating orbital period to orbital distance.

$$\frac{T_1^2}{T_2^2} = \frac{R_1^3}{R_2^3}$$



The full proof of this for elliptical orbits is a little in depth but for circular orbits, it is pretty straight forward:

Starting with the force of gravity causing circular motion:

$$\frac{GmM_E}{R^2} = m \frac{v^2}{R}$$

The amount of time for one period, T can be determined by:

$$2\pi R = v T \quad \frac{2\pi R}{T} = v$$

Combining everything and simplifying:

$$\frac{GM_E}{R} = \left(\frac{2\pi R}{T}\right)^2$$

$$\frac{GM_E}{4\pi^2} = \frac{R^3}{T^2}$$

Kepler and the Physics of Planetary Motion

- Laws of Planetary Motion
 - Law 1 - Law of Ellipses
 - Law 2 - Law of Equal Areas
 - Law 3 - Harmonic Law ($r^3/T^2 = C$)
- Kepler's laws provide a concise and simple description of the motions of the planets

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yours

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③ Kepler's third law:

- states that the square of the periodic time of an elliptical orbit is proportional to the cube of its semi-major axis length. mathematically,

$$\Rightarrow T^2 \propto a^3$$

where, $\frac{4\pi^2}{\mu}$ = proportionality constant.

μ = kepler's constant = $3.986005 \times 10^{14} \text{ m}^3/\text{sec}^2$

$$\text{or } T^2 = \left(\frac{2\pi}{n}\right)^2 \left(\frac{a^3}{\mu}\right)$$

$$\text{or } T^2 = n^2 \left(\frac{a^3}{\mu}\right)$$

$$\Rightarrow a^3 = \frac{\mu}{n^2}$$

where, n = mean motion of the satellite in rad/sec.

G. satellite orbits:

There are many different satellite orbits that can be used. Types of satellite orbits are :

- LEO : low earth orbit - 100 - 1500 Km
- MEO : medium " " - 5000 - 10000 Km
- GEO : Geostationary " " - 36000 Km.
- HEO : Highly elliptical orbit. -

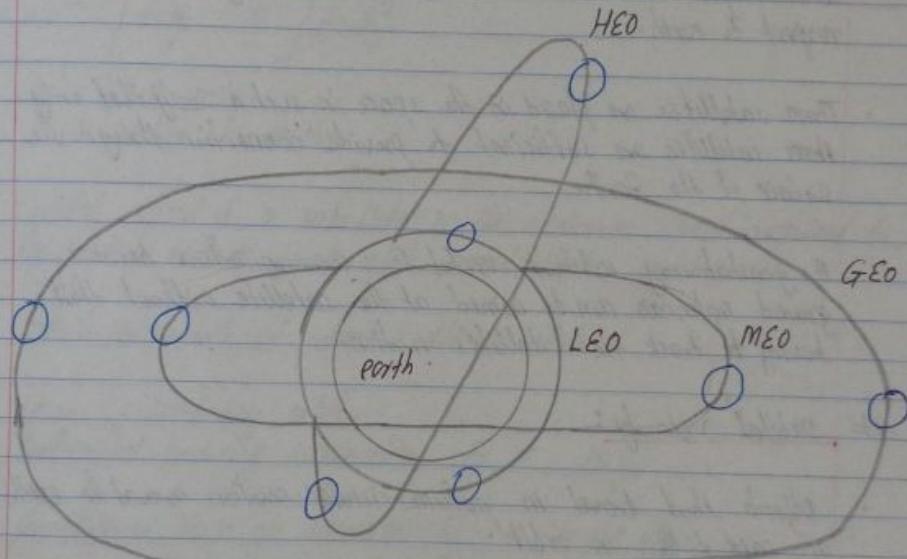
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The ones that receive the most attention are the geostationary orbit used as they are stationary above a particular point on the Earth. The orbit that is chosen for a satellite depends upon its application.

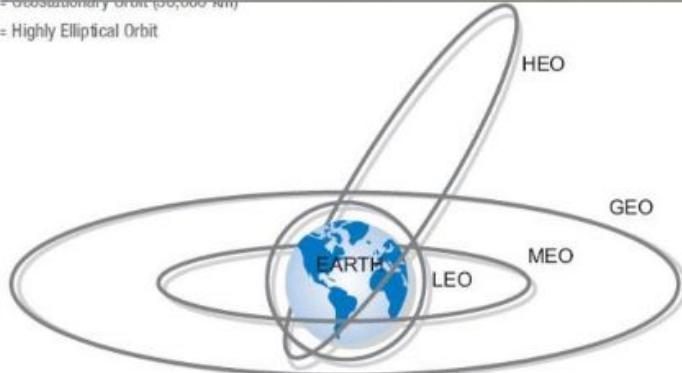


H. Geostationary satellite :

A geostationary satellite is an earth-orbiting satellite, placed at an altitude of approximately 35800 km directly over the equator, that revolves in the same direction the earth rotates (west to east). At this altitude, one orbit takes 24 hours, the same length of time as the earth requires to rotate once on its axis.

- Geostationary Orbit (36,000 Km)

- Highly Elliptical Orbit



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Geostationary or geosynchronous earth orbit (GEO): 2015/1

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- A satellite in a geostationary orbit appears to be stationary with respect to the earth. GEO satellites are synchronous with respect to earth.
- These satellites are placed in the space in such a way that only three satellites are sufficient to provide connection through the surface of the Earth.
- A geostationary orbit is useful for communications because ground antennas can be aimed at the satellite without them having to track the satellite's motion.

orbital velocity:

- Objects that travel in uniform circular motion around the earth are said to be "in orbit".
- velocity of this orbit depends on the distance from the object to the center of the earth.
- velocity has to be just right, so that the distance to the center of the Earth is always the same.

$$v = \sqrt{\frac{G \cdot M_e}{r}}$$

v = velocity of an object (m/s).

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r = distance from the objects to the center of the Earth.
 G = universal gravitational constant = $6.674 \times 10^{-11} \text{ Nm}^2/\text{kg}$.
 M_E = mass of Earth = $6 \times 10^{24} \text{ kg}$

or $v = \sqrt{\frac{GM}{r}} = \sqrt{g \cdot \frac{R^2}{r}}$

$g = 9.8 \text{ m/s}$
 R_E = radius of earth

- orbital speed of a satellite is the minimum speed required to put the satellite into a given orbit around earth.
- It is independent of mass of the satellite.
- For a geostationary orbit, centripetal producing acceleration, $F_c = \frac{mv^2}{r}$

Gravitational Force on satellite, $F_g = \frac{G m \cdot M_E}{r^2}$

or, $\frac{mv^2}{r} = \frac{G \cdot m \cdot M_E}{r^2}$

$\therefore v = \left(\sqrt{\frac{GM_E}{r}} \right) \text{ m/s}$

Round trip time delay of GSS : 2014s, 2015/F, 2015s, 18/F

- Round trip time delay also known as latency is the time taken for information to pass from the origin to the destination & potentially

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for the response to return.

- Round trip time delay in satellites is very much more obvious because of the very long distances that the signals must travel into space and back.
- The distance travelled depends on the location of the satellite and thus its orbit.
- Round trip time delay: $\geq 2 \times \frac{\text{distance}}{\text{velocity of light}}$
- calculating a transmission's RTT is advantageous because it allow users and operators to identify how long a transmission will take to complete and how fast a network can operate.

Characteristics of a Geostationary satellite orbit:

- Eccentricity (e) = 0
- inclination of the orbital plane (i) = 0°
- period (T) = 23 hour 56 min 4 sec.
- semi major axis (a) = 42184 km
- satellite altitude (R) = 35786 km
- " velocity (v) = 3075 m/s

Advantages: 2015/F

- No ground station tracking required.
- No inter satellite handoff, permanently in view.

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- 3 satellites give full earth coverage.
- Almost no Doppler shift, yields reduced complexity receivers.
- High covering area.
- No problem with frequency changes.
- one ground segment is enough for the satellite monitoring.

Disadvantages: 2015/F

- 35786 km orbits imply long transmission latencies
- Weak received signal.
- poor coverage at high latitudes (> 77 degrees).
- Large free space loss.
- Transmission delay of the order 250 ms

I) Why are uplink frequencies greater than downlink frequencies in satellite communication? 2014/LS, 28/F

⇒ The signals have to cross the atmosphere which presents a great deal of attenuation. The higher the frequency, the more is the signal loss and more power is needed for reliable transmission.

A satellite is a light weight device which cannot support high-power transmitters on it. So it transmits at a low frequency (higher the frequency, higher is the transmitter power to accommodate losses) as compared to the stationary earth station which can afford to use very high power transmitters. This is compensated by using highly sensitive receiver circuits on the earth station which is in the line of sight of the satellite.

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J) Footprint : 2015/16, 2017/18

- The area of the Earth covered by the microwave radiation from a satellite dish (transponder) is called the satellite footprint.
- The size of the footprint depends on the location of satellite in its orbit, the shape and size of beam produced by its transponder and the distance from the earth.
- The footprint can be seen as:
 - The area in which a broadcast signal from a particular satellite can be received.
 - The area of the surface of the earth within a satellite's transponder (transmitter or sensor) field of view.
- The signal power at the center of the footprint is maximum. The power decreases as we move from the footprint center. The boundary of the footprint is the location where the power reaches a predefined threshold.
- Satellite beam their signals in a straight path to the earth. The satellite focus these microwaves signals onto the specified portion of the earth's surface to most effectively use the limited power of their transponders. These focused signals create unique beam pattern called footprints.

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Types of Footprints:

- Global Beam footprint
- Zone " "
- Hemispheric " "

- Geo satellites have no coverage on the poles.

K. Transponder: 2015/5, 2016/5, 2017/5, 18/5

- In a communication satellite, a transponder gathers signals over a range of uplink frequencies and retransmits them on a different set of downlink frequencies to receivers on Earth.
- Transmitter responder.
- Known as satellite repeater.
- A transponder is typically composed of:
 - An input band limiting device (input band pass filter)
 - an " low noise amplifier.
 - Frequency translator
 - An output band pass filter
 - A power amplifier.
- Transponder performs the functions of both transmitter and receiver (responder) in a satellite. Hence, the word transponder is obtained by the combining few letters of two words, Transmitter & Responder.

Transponder performs mainly two functions. Those are amplifying the signals.

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The received input signal and translates the frequency of it.
In general, different frequency values are chosen for both uplink
and down link in order to avoid the interference between the
transmitted and received signals.

Two types of Transponder :

- Re-generative Transponder
- Transistor (amplifying) Transponder.

i) Regenerative Transponder :

- A regenerative transponder is a complete transceiver including a demodulator and modulator.
- The uplink signal is down-converter to IF and then demodulated to baseband.
- These pulses are then remodulated into a downlink, carrier up-converted and amplified before retransmission to the receiving earth station.
- The up and down links are decoupled and hence the interference and noise of two links do not add.

ii) Amplifying Transponder :

- It consists of three main components:
 - Low noise amplifier
 - Frequency converter
 - High power amplifier (HPA) with relevant filtering.

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I) Satellite system up link and downlink :			
- In these transducers, the uplink noise and interference are amplified at the transponder and transmitted to the receiving earth station. Hence, the uplink and downlink noise and interference are added at the earth station receiver.			
1) satellite system up link and downlink :			
- Power is the limitation at the satellite and power is directly proportional to frequency.			
- At earth station we can generate as much power as we need, but this is not possible at satellite.			
- For less attenuation, and better signal to noise ratio, lower frequency is more suitable for down link and higher frequency is commonly used for uplink.			
- RF amplifiers are most efficient at lower frequencies.			
- So, uplink frequencies are higher than downlink frequencies.			
M. Perform the brief comparison between satellites orbit (LEO, MEO, GEO) based on the points: 2016/5, 2017/5 Height, RTT, Technology used, Footprint, launching cost.			
⇒ Features	LEO	MEO	GEO
① Height in KM	100-3000	3000-36000	36000
② RTT	20-25 ms	110-130 ms	250-270 ms

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III) Transmitter power	small	medium	large.
IV) Antenna size	small	medium	large.
v) satellite cost	minimum	medium	maximum
vi) Satellite life	3-7 years	10-15	10-15
vii) Trunking capabilities	low	medium	High.
viii) Application	internet, data, voice, paging	24 GPs	satellite TV
ix. orbital period	90 min	6 hrs	24 hrs.
x. propagation loss	Least	High	Highest
xi. No. of satellites	40-80	8-20	3

N. What are the different satellite orbits that can be used? Explain each orbits in brief. 2016/F

⇒ make figure from page (189) and explain above answer.

O) Satellite system parameters :

The basic carrier to noise (C/N) relationship in a system establishes the transmission performance of RF portion of the system.

- The downlink thermal carrier to noise ratio is
 $C/N = C - 10 \log (KTB)$

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$C = \text{Received power in dBW}$
 $B = \text{noise bandwidth in Hz}$
 $T = \text{absolute temperature of receiving system in } ^\circ\text{K}$
 $K = \text{Boltzmann's constant} = 1.38 \times 10^{-23} \text{ W/K/Hz}$

Link equation:

$H(C/N) = EIRP - L + G - 10 \log (KTB)$
where,
 $EIRP = \text{equivalent isotropically radiated power (dBW)}$
 $L = \text{Transmission loss}$
 $G = \text{gain of received antenna.}$

$EIRP = 10 \log P_T + G_T$
. $P_T = \text{antennal input power}$
. $G_T = \text{Transmit. antenna gain}$

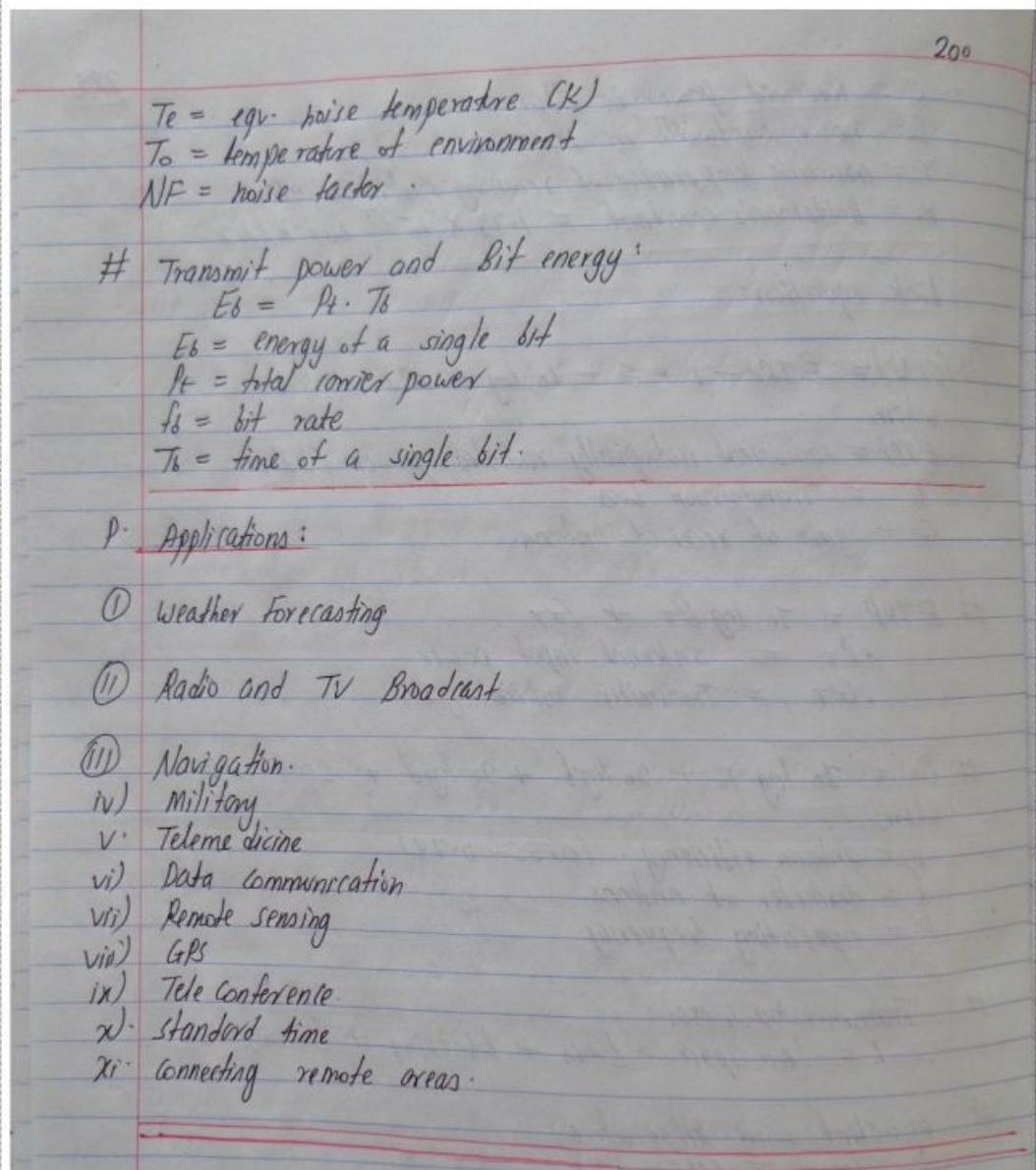
$G = 10 \log n + 20 \log f + 20 \log d + 20.4 \text{ dB}$
where,
 $n = \text{antenna efficiency: (0.55 - 0.78)}$
 $d = \text{diameter of antenna}$
 $f = \text{operating frequency.}$

Transmission losses:
 $L = L_{free space} + L_{rain} + L_{tracking} + L_{atmospheric}.$

equivalent noise temperature:
 $T_e = T_0 (N/F - 1)$

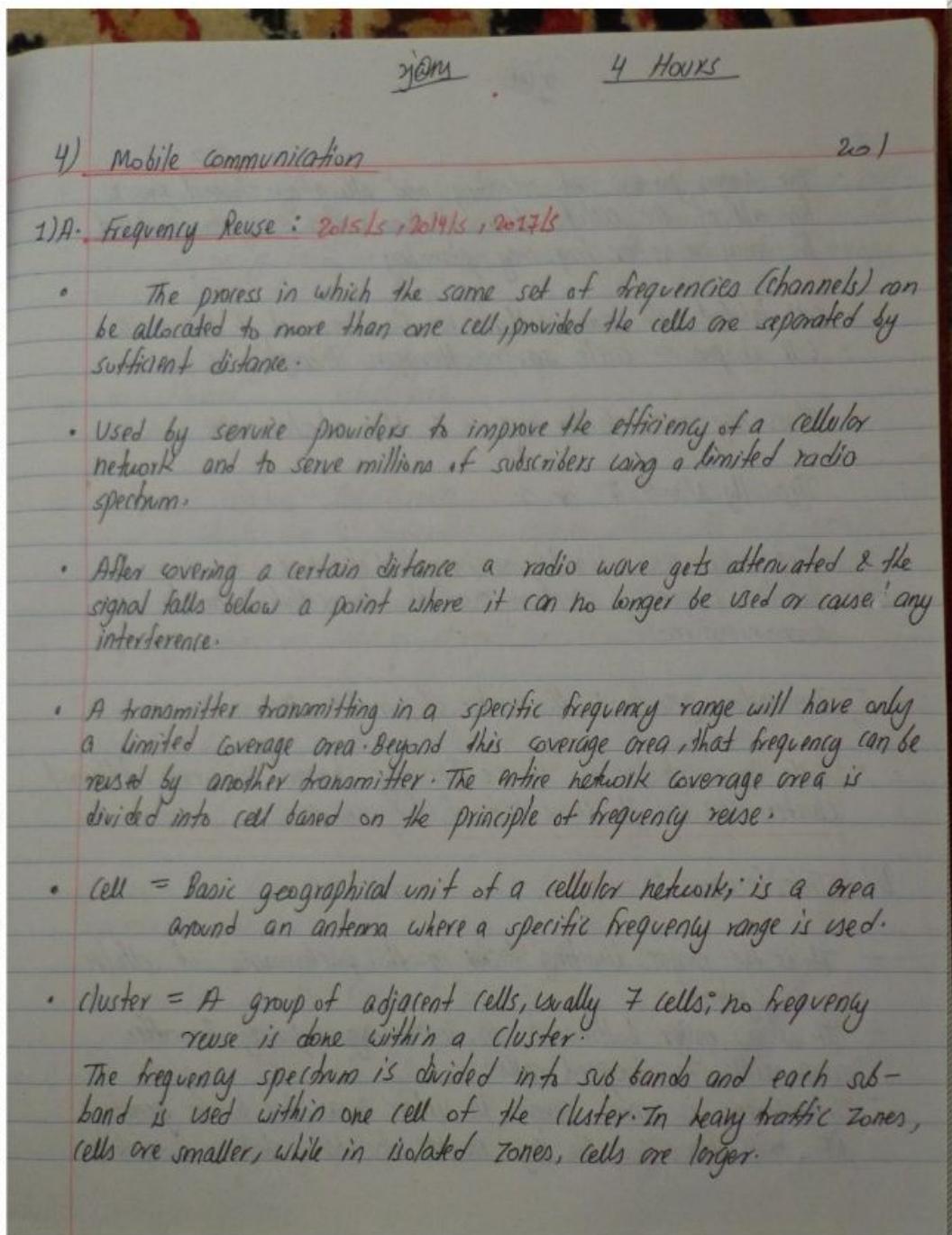
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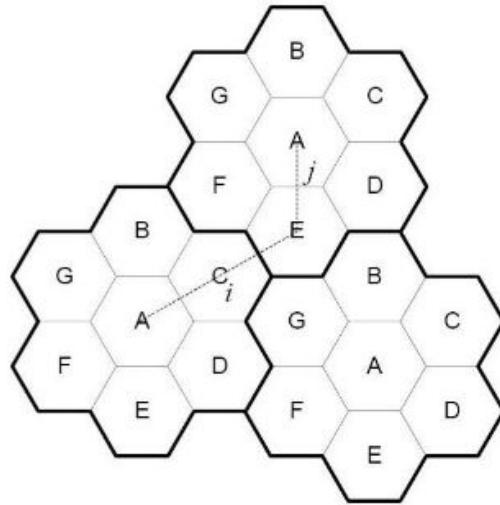
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Frequency Reuse ($N=7$, $i=2$, $j=1$)



- The Frequency Reuse factor is given as $1/N$, each cell is assigned $1/N$ of total channels.
- Lines joining a cell and each of its neighbor are separated by multiple of 60° , certain cluster sizes and cell layout possible
- Geometry of hexagon is such that no of cells per cluster i.e N , can only have values which satisfy the equation

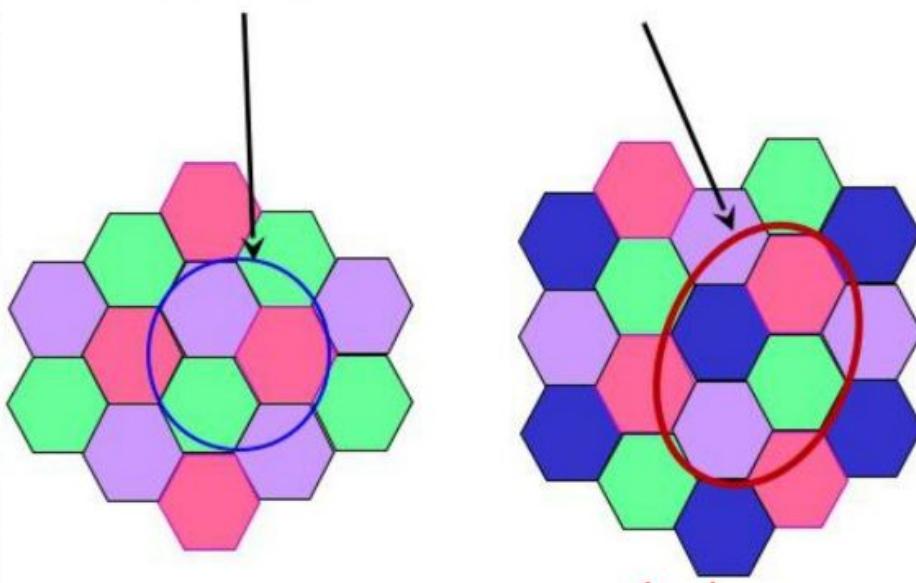
$$N = i^2 + ij + j^2$$

N , the cluster size is typically 4, 7 or 12.

In GSM normally $N = 7$ is used.

- i and j are integers, for $i=3$ and $j=2$ $N=19$.

Reuse factor = 3 Reuse factor = 4
3 cell Cluster 4 cell Cluster



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- The design process of selecting and allocating channel groups for all of the cellular base stations within a system is called frequency reuse or frequency planning.
- cell labeled with same letters use the same set of frequencies.
- cell shapes = circle, square, Hexagon, Triangle.
- Frequency reuse factor = $\frac{1}{N}$; N = cluster size
Typically, $N = 4, 7$ or 12 .
- Frequency reuse has become essential due to :
 - Limited frequency spectrum available for cellular mobile communications.
 - Tremendous growth in the number of mobile users.
- Cells in a cluster use unique frequency channels. However, different clusters can use the same set of frequencies.

B) Interference : 2015/1s

- It is the major limiting factor in the performance of cellular radio systems.
- It is a major bottleneck in increasing capacity & often responsible for dropped calls.
- Interference is more severe in urban areas, due to the greater RF noise floor and the large number of base stations & mobiles.

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- Sources of interference includes another mobile in the same cell/a call in progress in a neighboring cell, other base stations operating in the same frequency band or any noncellular systems which leaks energy into the cellular frequency band.

Two major types of interference :

- ① co-channel interference.
- ② Adjacent channel "

③ Adjacent channel Interference :

It is due to imperfect receiver filters, allowing nearby frequencies to leak into pass band. It can be minimized by careful filtering and channel assignments. channels are assigned such that the frequency separations between channels are maximized.

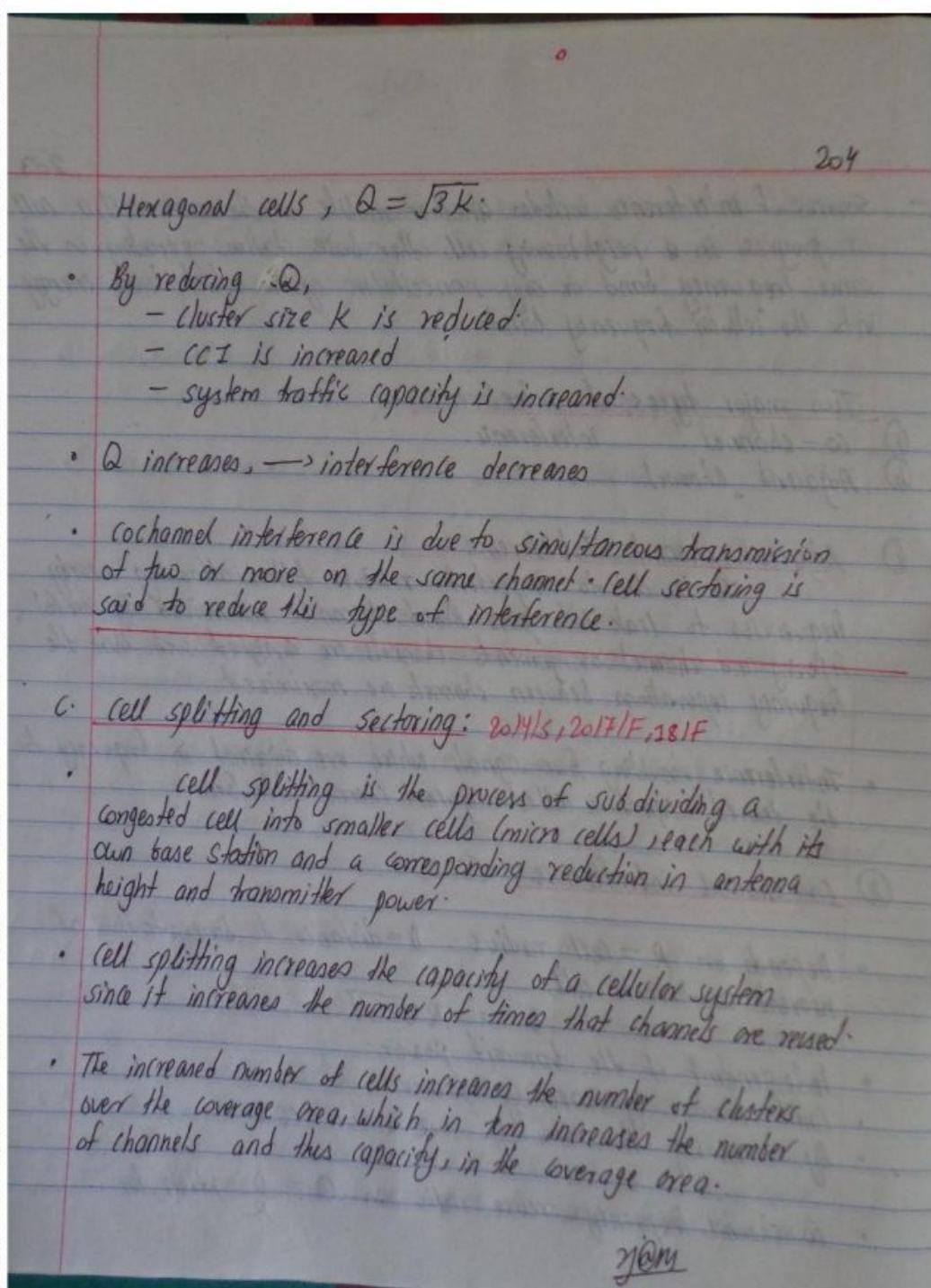
- Interference resulting from signals which are adjacent in frequency to the desired signal is called adjacent channel interference.

④ co-channel interference : (CCI)

- Depends on R = cell radius, D = distance to base station of nearest co-channel cell.
- $$D = R \sqrt{3} N$$
- Independent of the transmit power.
- Can't remove by increasing power.
- By increasing the ratio $\frac{D}{R}$ we reduce CCI
- Co-channel frequency reuse ratio as $Q = \frac{D}{R}$, then, for

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Limitations :

- Handoffs are more frequent.
- Channel assignments become difficult.
- All cells are not split simultaneously so special care have to be taken for proper allocation of problem.

To overcome some limitations like co-channel interference, cell sectoring is done.

- Sectoring is another way to increase the capacity.
- In this method, directional antennas are used to improve the signal to interference ratio so that the number of cells in a cluster can be reduced to increase the capacity.
- By using directional antennas, a given cell will receive interference and transmit with only a fraction of available co-channel cells.
- The further improvement in SIR is achieved by downtilting the sector antennas.

Advantages :

- ① Better S/I ratio
- ② Reduces interference.
- ③ Reduces cluster size, more freedom in assigning channels.

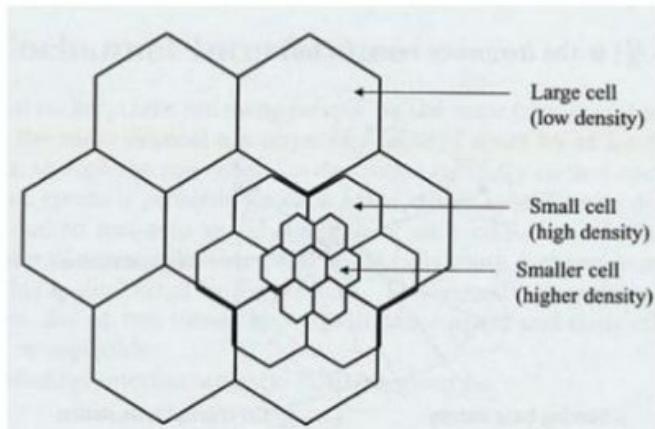
Limitations :

- i) Loss of Traffic.

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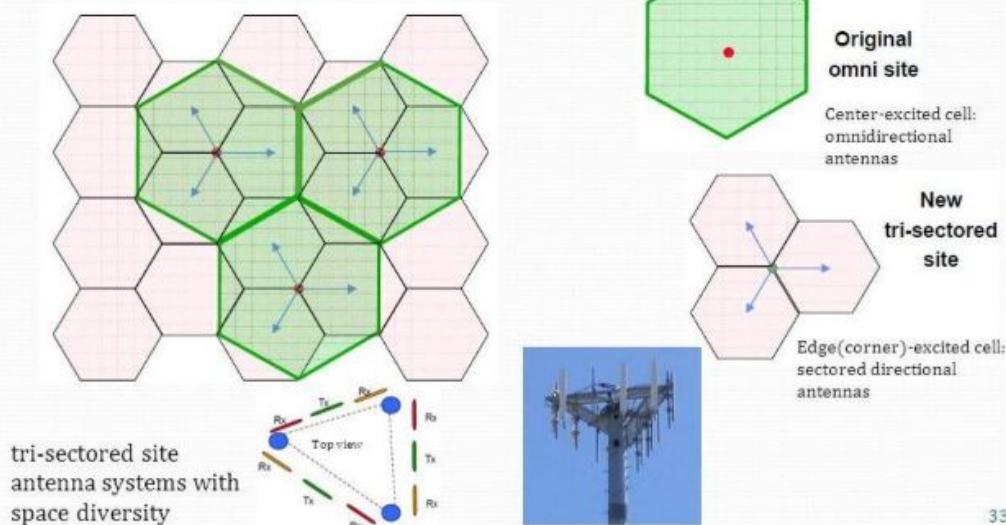
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Cell Splitting



Cell Splitting Techniques

- To increase capacity, split each cell into 3 using sectored antennas



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- (I) increased number of Handoffs.
- (II) Decrease in Trunk efficiency.
- (IV) Increased number of antennas per base station.

D) Basic cellular system: 2014/5

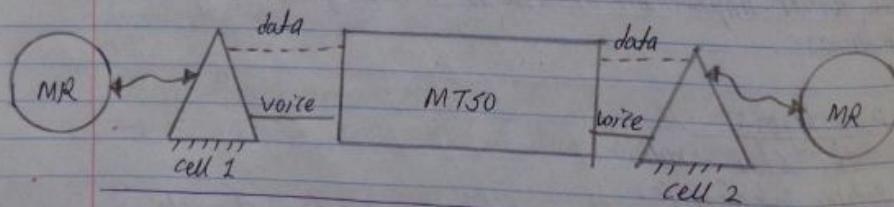
A basic cellular system consists of three parts:

- i) Mobile units.
- ii) cell units/site

Also known as Base Transceiver station. It contains a control unit, radio cabinet, antennas, data terminal and power plant.

iii) MSC:

Mobile switching center, also known as mobile telecommunication switching office (MTSO). It is central coordinating element for all cell sites and contains the cellular processor and a cellular switch.



E) Setting up a call process:

- When powered on, the phone does not have a frequency / time

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slot assigned to it yet, so it scans for the control channel of the BTS and picks the strongest signal.

- Then it sends a message (including its identification number) to the BTS to indicate its presence.
- The BTS sends an acknowledgement message back to the cell phone.
- The phone then registers with the BTS and informs the BTS of its exact location.
- After the phone is registered to the BTS, the BTS assigns a channel to the phone and the phone is ready to receive or make calls.

F. Making a call process : 2015/F

- The subscriber dials the receiver's number and sends it to the BTS.
- The BTS sends to its BSC the ID, location and number of the caller and also the number of the receiver.
- The BSC forwards this information to its MSC.
- The MSC routes the call to the receiver's MSC which is then sent to the receiver's BSC and then to its BTS.
- The communication with the receiver's cell phone is established.

G. Receiving a call process:

- When the receiver's phone is in an idle state it listens for the control channel of its BTS.

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- If there is an incoming call the BSC and BTS sends a message to the cells in the area where the receiver's phone is located.
- The phone monitors its message and compares the number from the message with its own.
- If the numbers matches the cell phone sends an acknowledgement to the BTS.
- After authentication, the communication is established between the caller and the Receiver.

H). Hand-OFF : /Handover : 2017/5

Hand off is the process of changing the channel associated with current connection while a call is in progress. Hand off is often initiated either by crossing a cell boundary or by a deterioration in quality of signal in current channel.

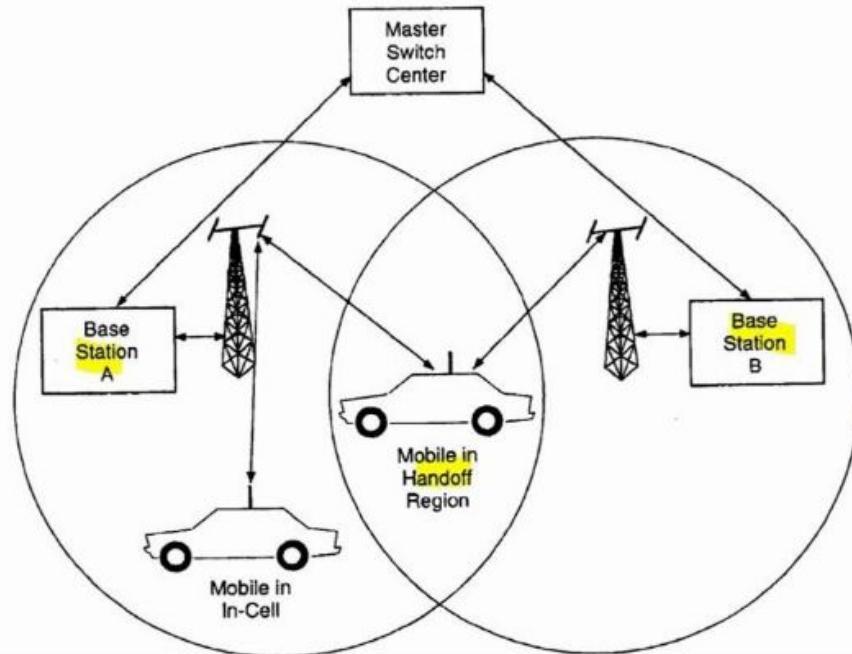
- Normally induced by the quality of the ongoing communication channel parameters: Received signal strength (RSS), signal-to-noise ratio (SNR) and Bit error rate (BER).
- Handoffs are triggered either by the BS or the mobile station itself.
- It is similar to an initial call request.

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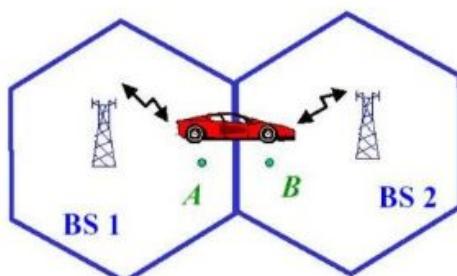
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System Model for the Cellular Handoff



- Handoff!!
 - The process of transferring a mobile user from one channel or base station to another.



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- The handoff has the priority over a new call to avoid call cut-off in the mid conversation.
- In reality, a fraction of total channels can be reserved for handoff requests in each cell.
- The handoff must be successful as infrequent as possible and unnoticeable to the user.
- Types of Handoff:

① Hard Handoff:

- mobile user is panned between disjoint towers that assign different frequency or adapt different air interface technology.

② Soft Handoff:

- Mobile user communicates to two towers simultaneously and the signal is treated as a multipath signal.
- make-before-break
- Generally used in CDMA systems.
- The signal of the best of all connected channels is utilized.

Advantages of soft handoff:

- Reduces number of call drops.
- increase the overall capacity.
- mobile transmit power is reduced.
- voice quality near the cell boundaries are improved.

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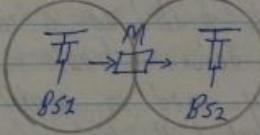
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- Handoff must be performed:
 - successfully
 - infrequently
 - imperceptible
- Types of Handoff

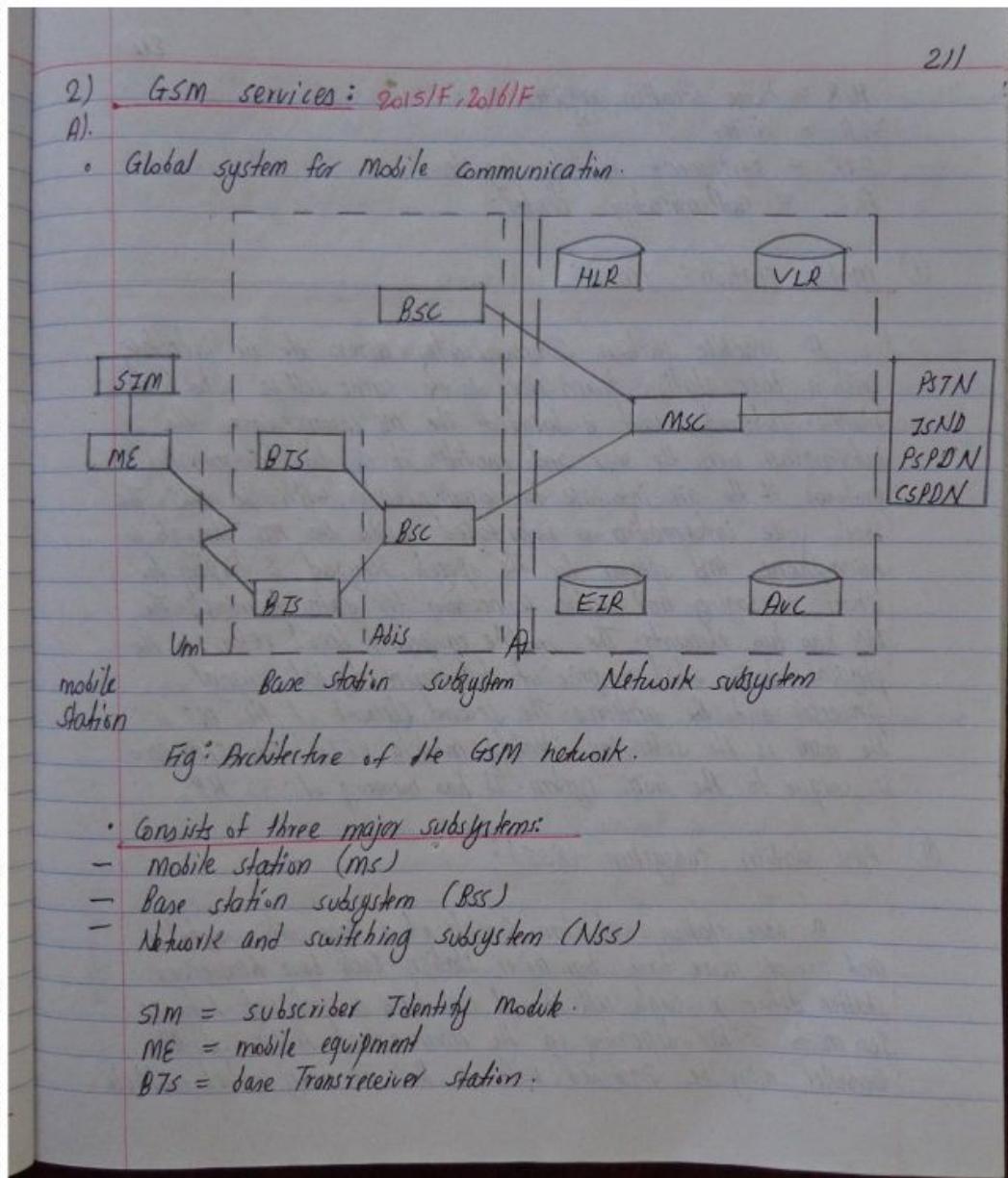
		classification	
		Intra cell	Inter cell
		soft	Hard
Horizontal			
vertical		Downward soft	upward Hard

- Soft handover = make-before-break
- Hard handover = Break-before-make.
- Δ = Handoff threshold.
- Δ too large:
 - Too many handoffs
 - Burden for MSC
- Δ too small:
 - More call losses
 - before call is lost
 - To complete Handoff.
 - Insufficient time.


 M = mobile

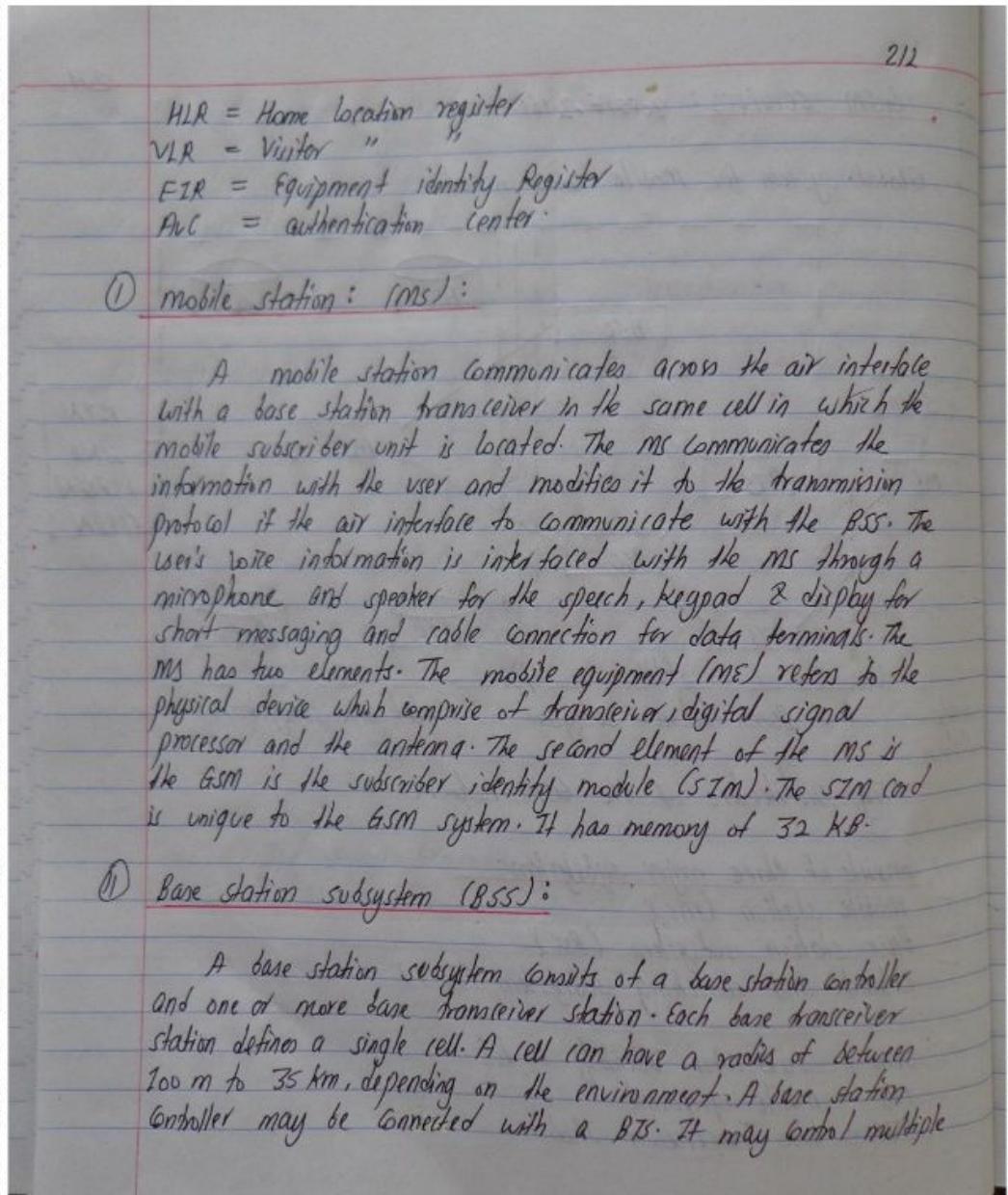
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BTS units and hence multiple cells. There are two main architectural elements in the BSS - the Base Transceiver subsystem (BTS) and Base station controller (BSC). The interface that connects a BTS to a BSC is called the A-bis interface.

(iii) Network and switching subsystem (NSS):

The NSS is responsible for the network operation. It provides the link between the cellular network and the public switched telecommunications Network (PSTN or ISDN or data networks). The NSS controls handoffs between cells in different BSS, authenticates users and validates their accounts, and includes functions for enabling worldwide roaming of mobile subscribers. In particular switching subsystem consists of,

- MSC • AUC
- HLR • EIR
- VLR • IWF

The NSS has one hardware, mobile switching center and four software database elements : HLR, VLR, AUC and EIR.

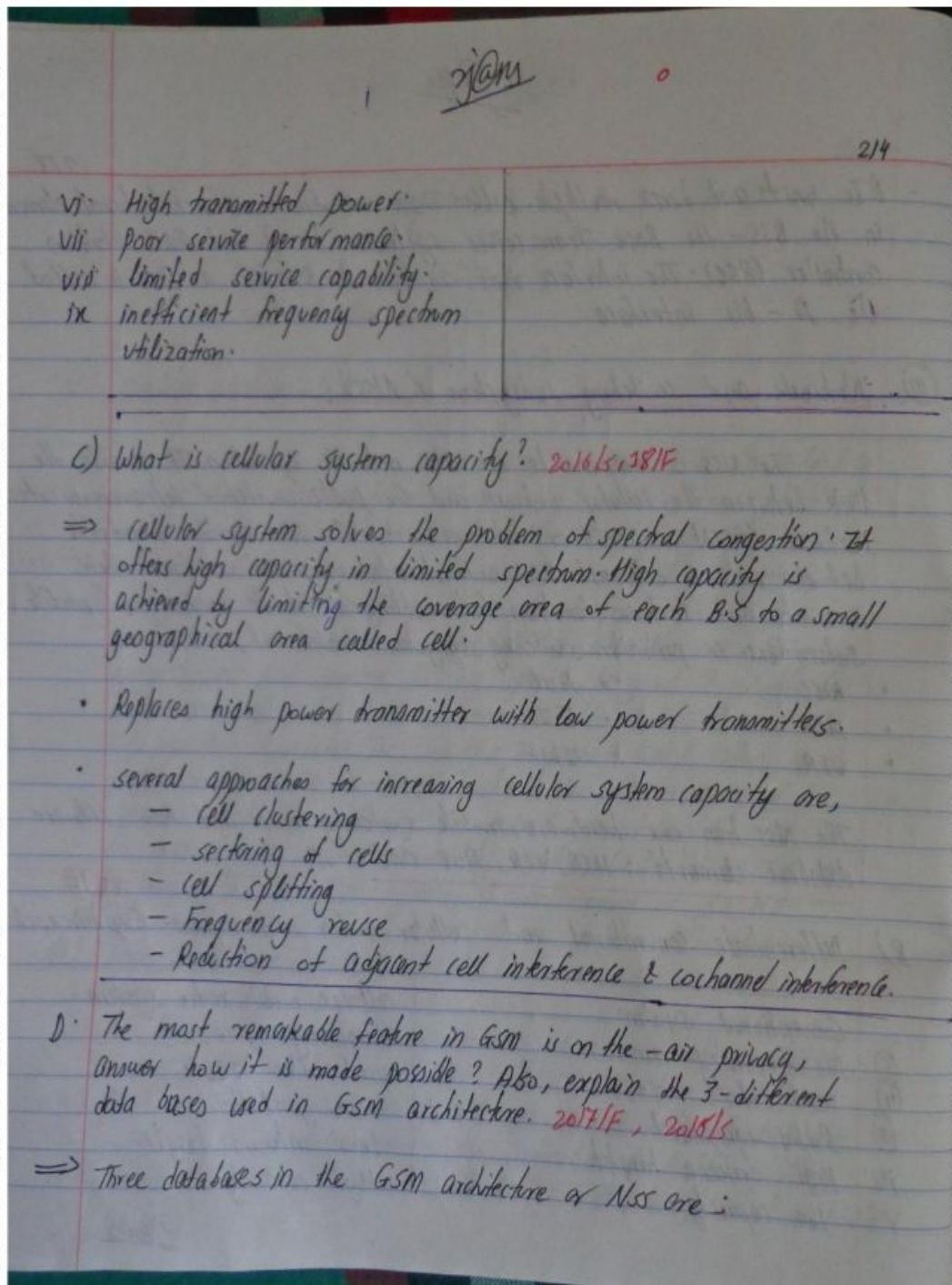
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8) Differentiate conventional and cellular mobile radio system.

Conventional system	cellular mobile radio system.
i) No frequency reuse	Frequency reuse.
ii) used before 1980's	used after 1980's
iii) Bulky equipment.	Hand portable.
iv) High antenna height.	Low antenna height.
v) Low capacity	High capacity.

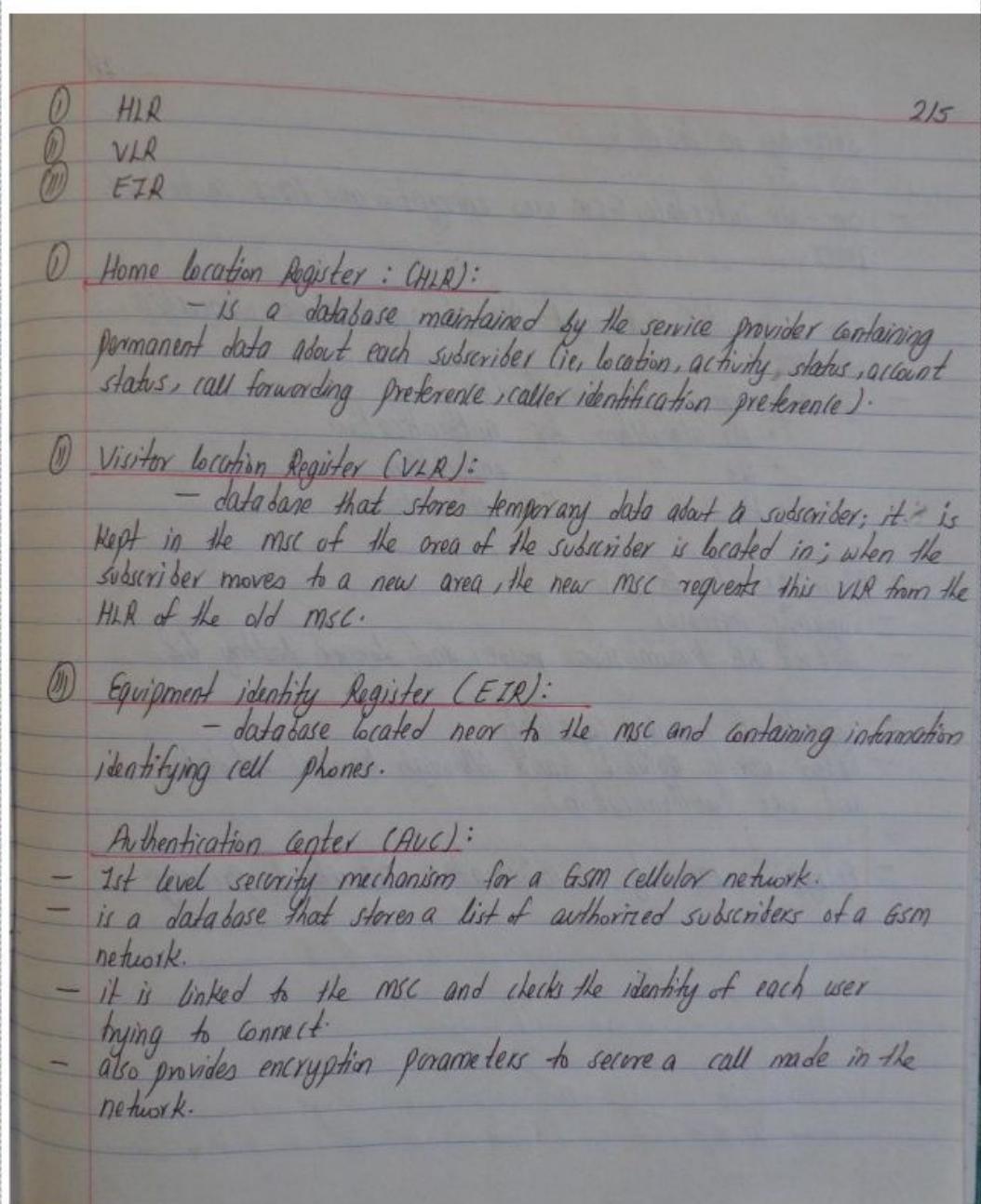
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Security in GSM :

- on-air interface, GSM uses encryption and TMSI instead of IMSI.
- SIM is provided 4-8 digit PIN to validate the ownership of SIM.
- 3-algorithms are specified:
 - A3 algorithm for authentication
 - A5 " " encryption.
 - A8 " " generation.
- Advantages of GSM :
- capacity increases
- reduced RF transmission power and longer battery life
- international roaming capability
- Better security against fraud (through terminal validation and user authentication)
- Encryption capability for information security & privacy.

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New + Probable Questions solutions:

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1) A. Back-off loss :

High power amplifiers used in Earth station transmitters and the travelling wave tubes typically used in the satellite transponders are non-linear devices; their gain (output power versus input power) is dependent on input signal level. It can be seen that as the input power is reduced by 4 dB, the output power is reduced by only 1 dB. There is an obvious power compression.

To reduce the amount of intermodulation distortion caused by the non-linear amplification of the HPA, the input power must be reduced (backed off) by several dB. This allows the HPA to operate in a more linear region. The amount the output level is backed off from rated level is equivalent to a loss and is appropriately called Back-off loss (L_{bo}).

Thus,

$$EIRP = Pt - L_{bo} - L_{tf} + At$$

where,

At = transmit antenna gain (dB)

L_{tf} = total branching and feeder loss in dB

L_{bo} = Back-off losses of HPA in dB

Pt = actual power output of the transmitter in dBW

Hence,

$$L_{bo} = EIRP + L_{tf} - At - Pt$$

To operate as efficiently as possible, a power amplifier should be operated as close as possible to saturation.

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The saturated output power is designated (P_o) or simply P_t . The output power of a typical satellite earth station transmitter is much higher than the output power from a terrestrial microwave power amplifier.

Most modern satellite systems use either phase shift keying (PSK) or quadrature amplitude modulation (QAM) rather than conventional, frequency modulation (FM).

B. Noise - Temperature :

With terrestrial microwave systems, the noise introduced in a receiver or a component within a receiver was commonly specified by the parameter noise figure. In satellite communication systems, it is often necessary to differentiate or measure noise in increments as small as a tenth or a hundredth of a decibel. Noise figure, in its standard form, is inadequate for such precise calculations. Hence, it is common to use environmental temperature (T) and equivalent noise temperature (T_e) when evaluating the performance of a satellite system.

Total noise power is,

$$N = kTB \quad \text{--- (1)}$$

$$\text{or } T = \frac{N}{kB} \quad \text{--- (2)}$$

where, N = total power (watt)

k = Boltzmann's constant (J/K)

B = Bandwidth (Hz)

T = temperature of the environment (K)

For,

$$F = 1 + \frac{T_e}{T} \quad \text{--- (3)}$$

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where, T_e = equivalent noise temperature (K)

F = noise factor (unitless)

T = temperature of the environment (K).

Hence,

$$T_e = T(F - 1)$$

Typically, equivalent noise temperatures of the receivers used in satellite transponders are about 1000 K. For Earth station receivers, T_e values are between 20 K and 1000 K.

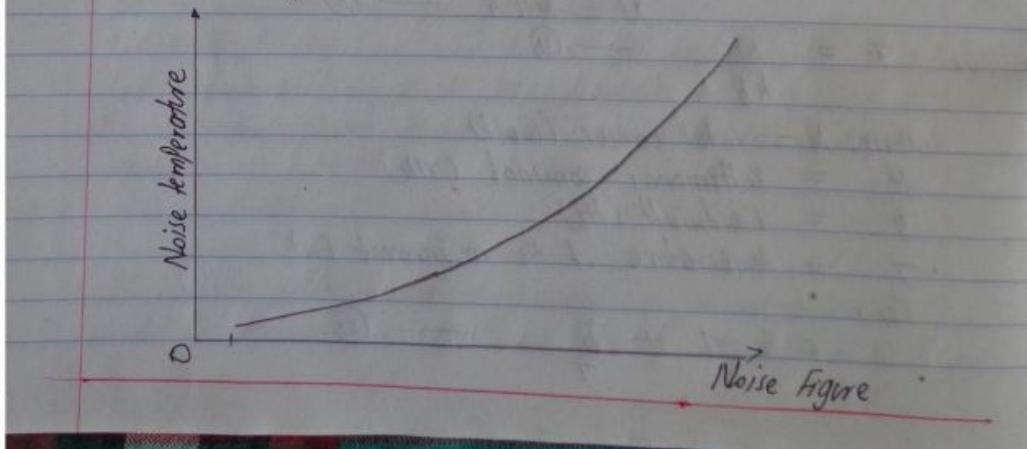
In dB,

$$T_e (\text{dBK}) = 10 \log T_e$$

For an equivalent noise temperature of 100 K,

$$\begin{aligned} T_e &= 10 \log 100 \\ &= 20 \text{ dBK} \end{aligned}$$

Equivalent noise temperature is a hypothetical value that can be calculated but cannot be measured. It is often used rather than noise figure because it is a more accurate method of expressing the noise contributed by a device or a receiver when evaluating its performance.



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C. Noise Density:

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Noise density (N_b) is the power normalized to a 1-Hz bandwidth or the noise power present in a 1-Hz bandwidth. Mathematically,

$$N_b = \frac{N}{B}$$

$$= \frac{k T_e \cdot B}{B}$$

$$= k \cdot T_e$$

where,

N_b = noise density (watt/Hz).

$$1 \text{ W/Hz} = \frac{1 \text{ joule/sec}}{1 \text{ cycle/sec}} = \frac{1 \text{ J}}{1 \text{ cycle}}$$

N = total noise power (watt)

B = Bandwidth (Hz)

k = Boltzmann's constant (J/K)

T_e = equivalent noise temperature (K)

Expressing as a log with 1 W/Hz as the reference,

$$\begin{aligned} N_b (\text{dBW/Hz}) &= I_0 \log N - I_0 \log B \\ &= I_0 \log k + I_0 \log T_e \end{aligned}$$

D. Carrier to Noise Density Ratio:

C/N_b is the average wideband carrier power to noise density ratio. The wideband carrier power is the combined power of the carrier and its associated sidebands. The noise density is the thermal noise

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Present in a normalized 1-Hz Bandwidth. The carrier-to-noise density may also be written as a function of noise temperature. Mathematically,

$$\frac{C}{N_0} = \frac{C}{kT_e}$$

Expressing in dB,

$$\frac{C}{N_0} (\text{dB}) = (C_{\text{dBW}}) - N_0 (\text{dBW})$$

E- Energy Bit-to-Noise density Ratio:

E_b/N_0 is one of the most important and most often used parameters when evaluating a digital radio system. The E_b/N_0 ratio is a convenient way to compare digital systems that use different transmission rates, modulation schemes or encoding techniques.

$$\frac{E_b}{N_0} = \frac{C/f_b}{N/B} = \frac{C_B}{N f_b} = \text{①}$$

E_b/N_0 is a convenient term used for digital system calculations and performance comparisons, but in the real world it is more convenient to measure the wideband carrier power-to-noise density ratio and convert it to E_b/N_0 .

Thus,

$$\frac{E_b}{N_0} = \frac{C}{N} \times \frac{B}{f_b}$$

The E_b/N_0 ratio is called product of the carrier-to-noise

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noise ratio (C/N) and the noise band-width-to-bit rate ratio (B/No).

$$\therefore \frac{E_b}{N_0} (\text{dB}) = \frac{C}{N} (\text{dB}) + \frac{B}{f_b} (\text{dB})$$

The energy per bit (E_b) will remain constant as long as the total wide band carrier power (C) and the transmission rate (B ps) remains unchanged. Also, the noise density (N_0) will remain constant as long as the noise temperature remains constant.

For a given carrier power, bit rate, and noise temperature, the E_b/N_0 ratio will remain constant regardless of the encoding technique, modulation scheme or Bandwidth.

The minimum E_b/N_0 equals the minimum C/N when the receiver noise bandwidth equals the bit rate which for BPSK also equals the minimum Nyquist bandwidth.

F. Gain to equivalent Noise temperature Ratio:

Gain to equivalent noise temperature ratio (G/T_e) is a figure of merit used to represent the quality of a satellite or earth station receiver. The G/T_e ratio is the ratio of the receiver antenna gain (G) to the equivalent system noise temperature (T_e) of the receiver. G/T_e is expressed as,

$$\frac{G}{T_e} = G - 10 \log (T_s)$$

Where,

G = receive antenna gain (dB)

T_s = operating or system temperature (K)

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$$\text{And } T_a + T_r = T_e$$

where,

T_a = antenna temperature

T_r = receiver effective input noise temperature.

Antenna gain is a unitless value whereas temperature has the unit of degree Kelvin.

G. For an equivalent noise bandwidth of 10 MHz , and a total noise power of 0.0276 pW , calculate the noise density & equivalent noise temperature.

Soln,

\Rightarrow we know,

$$N_b = \frac{N}{B} = \frac{276 \times 10^{-16}}{10 \times 10^6} \text{ W} = 276 \times 10^{-23} \text{ W/Hz}$$

$$N_b = 10 \log(276 \times 10^{-23}) = -205.6 \text{ dBW/Hz.}$$

$$\begin{aligned} \text{or } N_b &= 10 \log(276 \times 10^{-16}) - 10 \log 10 \text{ MHz} \\ &= -135.6 \text{ dBW} - 70 \text{ dBW} \\ &= -205.6 \text{ dBW} \end{aligned}$$

Also,

$$\begin{aligned} T_e &= \frac{N_b}{K} \\ &= \frac{276 \times 10^{-23}}{1.38 \times 10^{-23}} \text{ J/cycle} \\ &= 200 \text{ K/cycle.} \end{aligned}$$

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$$\begin{aligned}\log T_e &= T_0 \log 200 = 23 \text{ dBk} \\ &= N_0 - T_0 \log k \\ &= N_0 - T_0 \log (1.38 \times 10^{-23}) \\ &= -205.6 \text{ dBW} - (-228.6 \text{ dBWk}) \\ &= 23 \text{ dBk.}\end{aligned}$$

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H) Effective Isotropic Radiated Power (EIRP):

The EIRP is the transmit power of a hypothetical antenna radiating equally in all directions (like a light bulb) so as to have the same power flux density over the coverage area as the actual antenna.

The power flux density of the actual antenna is,

$$\begin{aligned}\phi &= \frac{P}{S} = \frac{n^* P_{in}}{\Omega_A \cdot d^2} \\ &= n^* \times \frac{4\pi}{\Omega_A} \times \frac{P_{in}}{4\pi d^2} \\ &= G_t \cdot \frac{P_{in}}{4\pi d^2}\end{aligned}$$

where, n^* = Antenna power loss efficiency

$P = n^* \cdot P_{in}$ = transmitted power.

S = total coverage area at distance d .

Ω_A = antenna beam solid angle.

G_t = transmit gain.

By definition of EIRP,

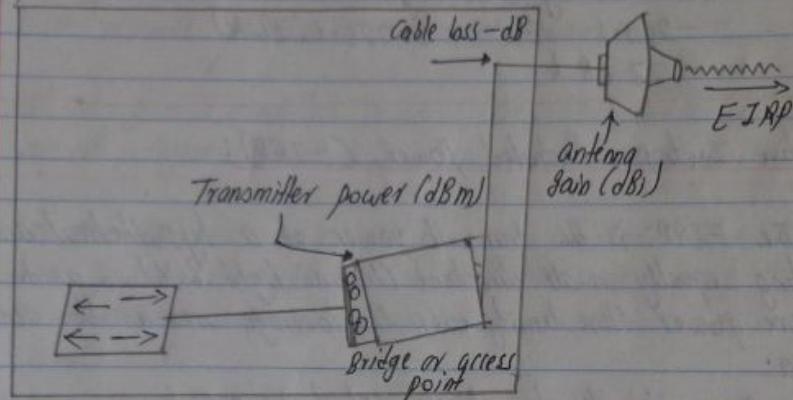
$$\phi = \frac{EIRP}{4\pi d^2}$$

Hence, $EIRP = G_t \cdot P_{in}$.

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The EIRP is the product of the antenna transmit gain and the power applied to the input terminal of the antenna.



$$\text{Also, } \text{EIRP} = P_r \cdot A_t$$

where,

P_r = power radiated power from antenna (W)

A_t = transmit antenna gain (W/W or unitless)

Expressing in log,

$$\therefore \text{EIRP (dBW)} = P_r (\text{dBW}) + A_t (\text{dBW})$$

In respect to the transmitter output,

$$P_r = P_t + L_{bo} - L_{tf}$$

P_t = actual power output of transmitter (dBW)

L_{bo} = backoff losses of HPA (dB)

L_{tf} = total branching and feeder loss (dB)

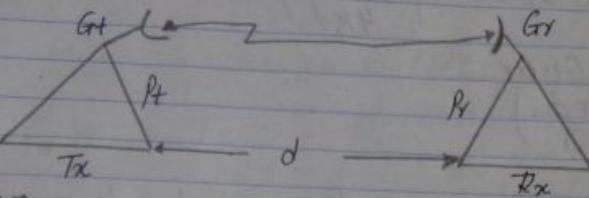
A_t = transmit antenna gain (dB)

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Proof:

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Here,

$$Pr = |Pfd| \cdot Ad \quad \text{--- (I)}$$

where, $Pfd = \frac{Pt \times Gt}{4\pi d^2}$

Ad = aperture of the receiving antenna.
So,

$$Pr = \frac{Pt \cdot Gt}{4\pi d^2} \cdot Ad \quad \text{--- (II)}$$

Also,

$$G = \frac{n \cdot 4\pi A}{\lambda^2}$$

$$\text{or } A = \frac{G \cdot \lambda^2}{n \cdot 4\pi} \quad \text{--- (III)}$$

$$\begin{aligned} \text{or } Ad &= \frac{n \cdot G \cdot \lambda^2}{4\pi} \quad \text{where } Ad = nA \\ &= \frac{G \cdot \lambda^2}{4\pi} \end{aligned}$$

and,

$$\begin{aligned} Pr &= \frac{Pt \cdot Gt}{4\pi d^2} \cdot Ad. \quad \text{From (I)} \\ &= \frac{Pt \cdot Gt}{4\pi d^2} \cdot \frac{G \cdot \lambda^2}{4\pi} \end{aligned}$$

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$$= P_t \cdot G_t \cdot G_r \cdot \left(\frac{\lambda}{4\pi d} \right)^2$$

$$= \frac{P_t \cdot G_t \cdot G_r}{\left(\frac{4\pi d}{\lambda} \right)^2}$$

where,

$L_p = \left(\frac{4\pi d}{\lambda} \right)^2$ = path loss/free space loss ie, energy spread out as a EM wave travel away from

$$\therefore P_r = \frac{P_t \cdot G_t \cdot G_r}{L_p} \quad \text{Transmitting wave.}$$

$$\therefore P_r = P_t \cdot G_t \cdot G_r \left(\frac{\lambda}{4\pi d} \right)^2$$

$$= \frac{\text{EIRP}}{\left(\frac{4\pi d}{\lambda} \right)^2} \cdot G_r$$

$$\therefore P_r = \frac{\text{EIRP}}{L_p} \cdot G_r$$

Expressing in dB,

$$\therefore P_r (\text{dB}) = 10 \log (\text{EIRP}) + 10 \log (G_r) - 10 \log (L_p).$$

$$\therefore P_r (\text{dB}) = (\text{EIRP})_{\text{dB}} + (G_r)_{\text{dB}} - (L_p)_{\text{dB}}.$$

G_t = transmitter gain

G_r = receiver gain.

P_r = power radiated from antenna.

P_t = transmitted power.

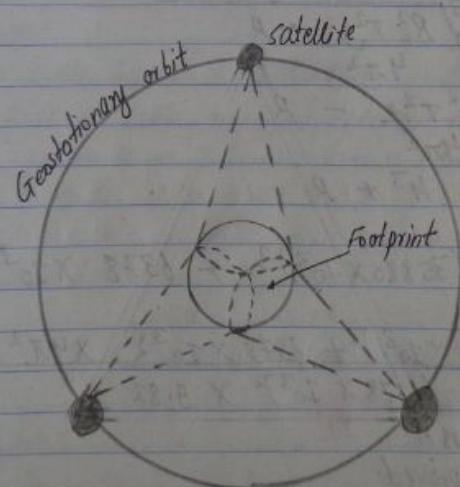
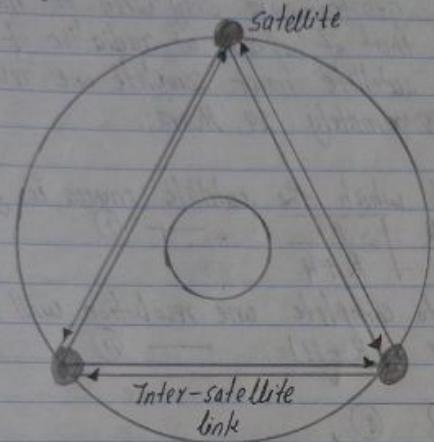
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I. Figure of footprint:

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J. Prove that a satellite orbiting the earth at a height of 35880 Km from the surface of the earth will be geostationary.
Soln,

→ A satellite is geostationary only when it moves with a velocity equal to that of speed of rotation of earth about its axis i.e. the satellite must complete one revolution around the earth in approximately 24 Hours.

velocity v with which the satellite moves is given by,

$$v = R \sqrt{\frac{g}{R+H}} \quad \text{--- (1)}$$

Also, Time, T to complete one revolution will be,

$$T = 2\pi \frac{(R+H)}{v} \quad \text{--- (2)}$$

From (1) & (2),

$$\text{Height, } H = \sqrt[3]{\frac{R^2 T^2 g}{4\pi^2}} - R$$

$$\text{or } H^3 = \frac{R^2 T^2 g}{4\pi^2} - R$$

$$\text{or } \frac{R^2 T^2 g}{4\pi^2} = H^3 + R$$

$$= (35880 \times 10^3)^3 + 6378 \times 10^3$$

$$\text{or } T = \left[\frac{(35880 \times 10^3)^3 + 6378 \times 10^3}{(6378 \times 10^3)^2 \times 9.81} \right]^{1/2}$$

$$\therefore T = 24 \text{ hours.}$$

Hence, proved.

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R. Determine the optical power received in dB, dBm and watts for a 50 km optical fiber link with the following design parameters: LED output = 45 mw, eight, 5 km sections of optical cable each with a loss of 0.2 dB/km, four cable to cable connectors with loss of 1 dB each, two splices with loss 0.5 dB each, light source to fiber interface loss 1.5 dB, fiber to light detector loss of 1.6 dB and no losses due to cable bends. 2017/F, 2016/S

Soln:

\Rightarrow LED \rightarrow source, converted to dBm

$$P_{out} = 10 \log \left(\frac{45 \text{ mw}}{1 \text{ mw}} \right) = 16.5321 \text{ dBm.}$$

$$\text{Total cable loss} = 50 \text{ km} \times 0.2 \text{ dB/km}$$

$$= 10 \text{ dB}$$

$$\text{Total connector loss} = 4 \text{ connectors} \times 1 \text{ dB/connector}$$

$$= 4 \text{ dB}$$

Hence,

$$\therefore \text{total loss} = \text{cable loss} + \text{connector loss} + \text{light source} - \text{to cable loss} + \text{cable to light detector loss} + \text{splices loss}$$

$$= 10 + 4 + 1.5 + 1.6 + 2 \times 0.5$$

$$\therefore P_t = 18.10 \text{ dB}$$

$$\therefore \text{Received power (P_r)} = P_{out} - P_t = \text{dBm} - \text{dB}$$

$$= 16.532 - 18.10$$

$$= -1.5680 \text{ dBm}$$

Now,

$$10 \log (P_r) = -1.5680$$

$$\text{or, } P_r = \frac{\text{antilog} (-1.5680)}{10}$$

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$$\therefore P_r = 0.70 \text{ mWatt} \cdot = 0.00070 \text{ watt}$$

$$dBm = 10 \log \left(\frac{P}{1 \text{ mW}} \right)$$

$\therefore P$ in watts

$$\begin{aligned} P_r &= P_t - \text{losses} \leftarrow \text{link budget analysis of optical fiber} \\ &= P_t (dBm) - A (dB) \end{aligned}$$

$$\begin{aligned} \therefore P_r \text{ in dB} &= -30 + dBm \\ &= -30 - 1.5680 \\ &= -31.5680 \text{ dB} \end{aligned}$$

Note:

$$dB = dBm - 30$$

$$\text{watt} = 10^{\frac{dBm - 30}{20}}$$

$$\text{mW} = 10^{\frac{dBm}{20}}$$

1. For a single mode optical cable with 0.25 dB/km loss, determine the optical power 100 km from a 0.1 mW source.

Soln,

$$\Rightarrow \text{we know, } P = P_t \times 10^{\frac{-\alpha \times L}{20}}$$

$$P_t = 0.1 \text{ mW}$$

$$\alpha = 0.25 \text{ dB/km}$$

$$L = 100 \text{ km}$$

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$$\therefore P = 0.1 \times 10 \stackrel{-0.25 \times 100}{\cancel{10}} \\ = 0.316 \text{ mW}$$

Now,

$$\text{Power in dBm} = 10 \log \left(\frac{P_{\text{out}}}{1 \text{ mW}} \right) \\ = 10 \log \left(\frac{0.316 \times 10^{-3}}{0.001} \right) \\ = -35 \text{ dBm.}$$

M. A rule of bandwidth for FM signal is sometimes used as,

$$B.W = (2mf + 1) \text{ fm},$$

Find the fraction of the signal power that is included in that frequency band. Assume $mf = 1$ and 10 .

Soln,

\Rightarrow The bandwidth for FM signal can be calculated on the basis of 98% power requirement given by Carson's rule as,

$$B.W = 2 \Delta f \left(1 + \frac{1}{mf} \right) = 2 (mf + 1) \text{ fm} \quad \text{--- (1)}$$

The fraction of signal power included in the frequency band B.W.

$$P = \frac{B}{B.W} \times \left(\frac{98}{100} \right) \quad \text{--- (2)}$$

For $mf = 1$,

$$P = \frac{B}{B.W} \times \frac{98}{100} = \frac{B}{B.W} \times 0.98$$

$$\text{or } P = \frac{(2mf + 1) \text{ fm}}{2(mf + 1) \text{ fm}} \times 0.98$$

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$$= \frac{2 \times 1 + 1}{2(1+1)} \times 0.98 \\ \therefore p = 73.5 \%$$

For $mf = 10$,

$$p = \frac{(2mf + 1)fm}{2(mf + 1)fm} \times \left(\frac{98}{100}\right) \\ = \frac{(2 \times 10 + 1)}{2(10+1)} \times 0.98 \\ \therefore p = 97.6 \%$$

N. Determine the permissible range in maximum modulation index for,

- i) Commercial FM which has 30 Hz to 15 kHz modulating frequencies.
- ii) Narrow band FM system which allows maximum deviation of 10 kHz and 100 Hz to 3 kHz modulating frequencies.

Soln,

=> maximum deviation in commercial FM is,

$$\Delta f = 75 \text{ kHz}$$

modulation index in FM is,

$$mf = \frac{\Delta f}{fm}$$

modulation index for commercial FM at $fm = 30 \text{ Hz}$ is,

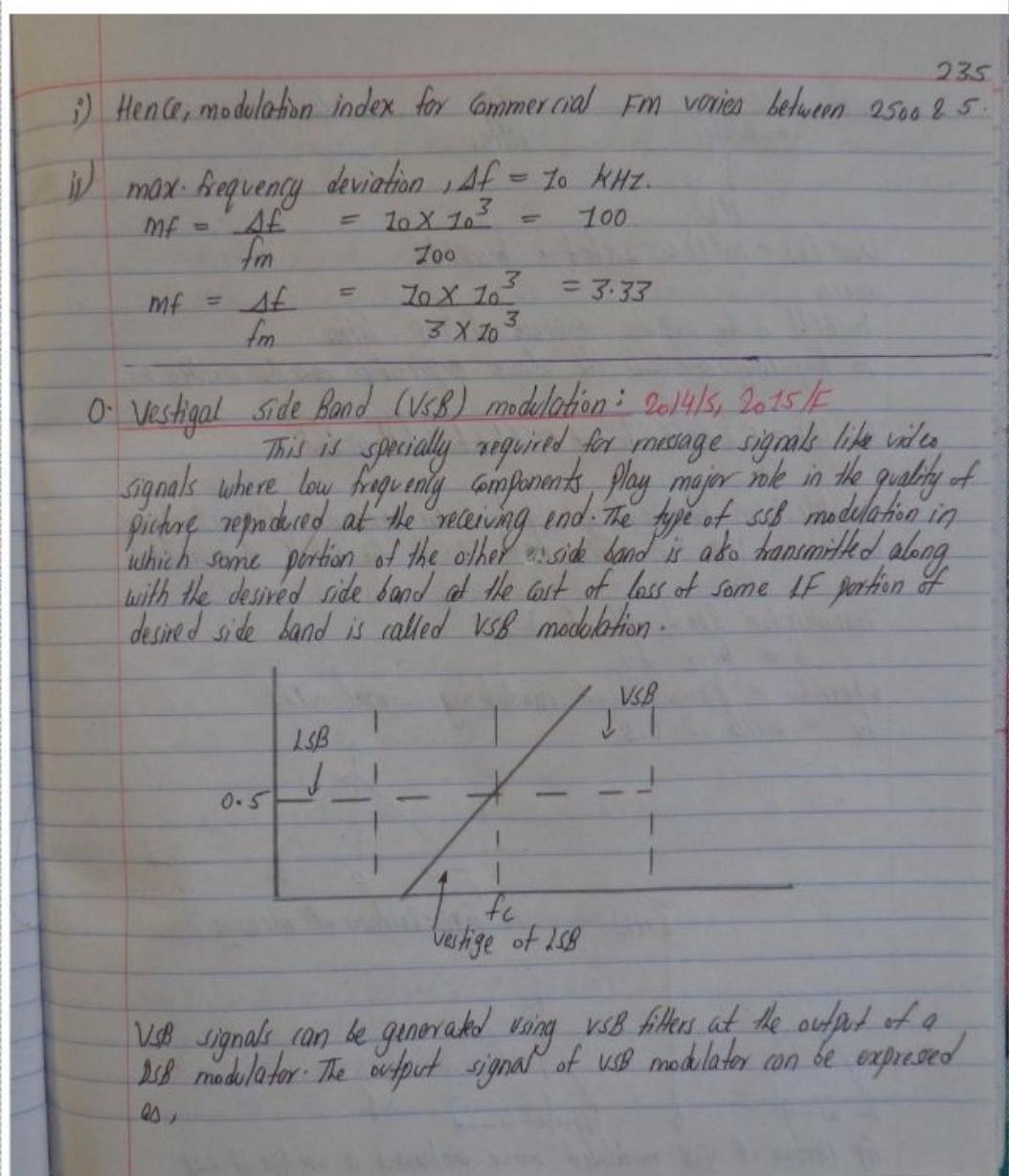
$$mf = \frac{\Delta f}{fm} = \frac{75 \times 10^3}{30} = 2500$$

modulation index for commercial FM at $fm = 15 \text{ kHz}$ is,

$$mf = \frac{\Delta f}{fm} = \frac{75 \times 10^3}{15 \times 10^3} = 5$$

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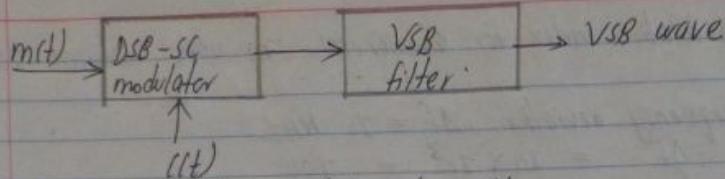
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$$U_{VSB}(t) = m(t) \cos 2\pi f_c t + h_{VSB}(t)$$

where,

$h_{VSB}(t)$ is the impulse response of VSB filter.

In frequency domain, the above expression can be written as,

$$U_{VSB}(f) = \frac{1}{2} [m(f - f_c) + m(f + f_c)] H_{VSB}(f).$$

The television broadcasting signal is an example of VSB signal in which VSB, full carrier and 30% of the LSB is transmitted.

Transmission Bandwidth of VSB :

$$\delta = w + f_v$$

where, w = bandwidth of modulating signal $m(t)$

f_v = width of VSB.

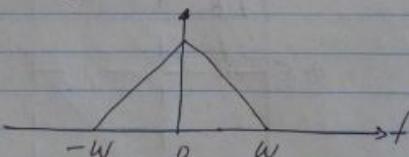


Fig: spectrum of message signal

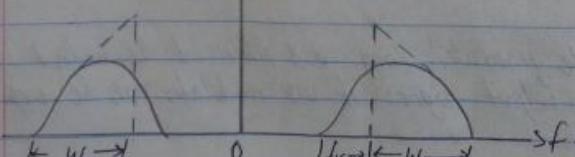


Fig: spectrum of VSB modulated wave containing a vestige of LSB.

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D- Frequency modulation :

Frequency modulation (FM) is that form of angle modulation in which the instantaneous frequency $f(t)$ varied linearly with message signal $m(t)$.

$$f(t) = f_c + k_f m(t)$$
$$\theta(t) = 2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau$$
$$\therefore s(t) = A_c \cos [2\pi f_c t + 2\pi k_f \int_0^t m(\tau) d\tau]$$

Time domain :

$$s_{fm} = A_c \cos [2\pi f_c t + 2\pi k_f \int_0^t A_m \cos 2\pi f_m \tau d\tau]$$
$$= A_c \cos [2\pi f_c t + 2\pi k_f \cdot \frac{A_m}{2\pi f_m} \sin 2\pi f_m t]$$
$$= A_c \cos [2\pi f_c t + k_f \cdot \frac{A_m}{f_m} \cdot \sin 2\pi f_m t]$$
$$= A_c \cos [2\pi f_c t + \frac{\Delta f}{f_m} \cdot \sin 2\pi f_m t]$$

where,

$\Delta f = k_f \cdot A_m$ = frequency deviation .

$\therefore s_{fm}(t) = A_c \cos [2\pi f_c t + 2\pi \Delta f \sin 2\pi f_m t]$

where,

$M = \frac{\Delta f}{f_m}$ = modulation index.

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spectrum analysis of Fm :

The FM wave for sinusoidal modulation is given by,

$$s(t) = A_c \cos [2\pi f_c t + \beta \sin 2\pi f_m t]$$

In exponential form,

$$s(t) = \operatorname{Re} [A_c \exp (j2\pi f_c t + j\beta \sin 2\pi f_m t)]$$

$$= \operatorname{Re} [A_c \exp (j2\pi f_c t) \exp (j\beta \sin 2\pi f_m t)]$$

Here,

$\exp (j\beta \sin 2\pi f_m t)$ can be expanded in Fourier series,

$$\exp (j\beta \sin 2\pi f_m t) = \sum_{n=-\infty}^{\infty} c_n \exp (j2\pi f_m n t)$$

where,

$$c_n = \frac{1}{T_m} \int_{-T_m/2}^{T_m/2} \exp (j\beta \sin 2\pi f_m t) \exp (-j2\pi f_m n t) dt$$

$$= f_m \int_{-1/2f_m}^{1/2f_m} \exp (j\beta \sin 2\pi f_m t) \exp (-j2\pi f_m n t) dt$$

$$= f_m \int_{-1/2f_m}^{1/2f_m} \exp (j\beta \sin 2\pi f_m t - j2\pi f_m n t) dt$$

Assume,

$$x = 2\pi f_m t \text{ for } t = -\frac{T_m}{2}, x = -\pi \text{ and}$$

$$t = \frac{T_m}{2}, x = \pi$$

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$$\therefore dt = \frac{dx}{2\pi f_m}$$
$$C_n = \frac{f_m}{2\pi f_m} \int_{-\pi}^{\pi} \exp [J\beta \sin x - J_n x] dx$$
$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} \exp [J(\beta \sin x - nx)] dx$$
$$= J_n (\beta)$$
$$= \text{Bessel function}$$

where,

$J_n (\beta)$ is n^{th} order Bessel function with argument β .

$$\exp (J\beta \sin 2\pi f_m t) = \sum_{n=-\infty}^{\infty} J_n (\beta) \cdot \exp (J2\pi f_m n t)$$

we know,

$$S_m(t) = \operatorname{Re} \left[A_c \sum_{n=-\infty}^{\infty} J_n (\beta) \exp (J2\pi f_m n t) \exp (J2\pi f_c t) \right]$$
$$= A_c \sum_{n=-\infty}^{\infty} J_n (\beta) \cos [2\pi (f_c + n f_m) t]$$

The discrete spectrum of $S(t)$ is obtained by taking Fourier transform

$$\therefore S(f) = \frac{A_c}{2} \sum_{n=-\infty}^{\infty} J_n (\beta) [\delta[f - (f_c + n f_m)] + \delta[f + (f_c + n f_m)]]$$
$$\therefore J_n (\beta) = \begin{cases} J_n (\beta) & ; n \text{ even} \\ -J_n (\beta) & ; n \text{ odd} \end{cases}$$

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Q. Narrowband FM:

If $B \ll 1$, radian, the FM wave consists of a carrier, an upper side frequency component, and a lower side frequency component, such FM is known as the narrow band FM.

Let, $B \leq 0.3$ rad, then from the plots of Bessel function,

$$J_0(B) \approx 1, J_1(B) \approx B/2$$

$$J_n(B) \approx 0, n > 1,$$

Then, from eqn,

or $s(t) = A_c \sum_{n=-\infty}^{\infty} J_n(B) \cos[2\pi(f_c + n f_m)t]$ become,

$$\text{or } s(t) \approx \frac{A_c}{2} [s(f_c - f_m) + s(f_c + f_m)] + \frac{B A_c}{4} [s(f_c - f_m) t$$

$$\therefore s(f_c + f_c + f_m) t] - \frac{B A_c}{4} [s(f_c - f_c + f_m) + s(f_c + f_c - f_m) t]$$

Since,

$$s(t) \approx A_c \cos(2\pi f_c t) + \frac{B A_c}{2} \cos[2\pi(f_c + f_m)t] - \frac{B A_c}{2}$$

$$\cos[2\pi(f_c - f_m)t] [\because J_2(B) = -J_1(B)]$$

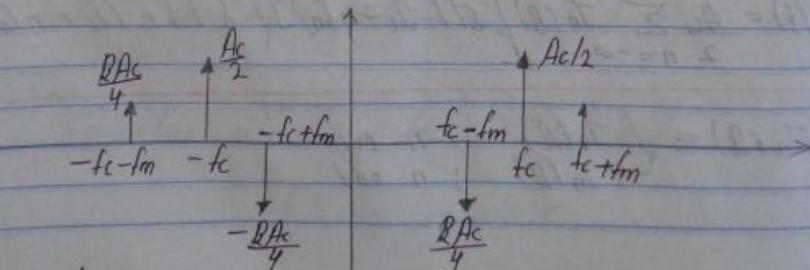
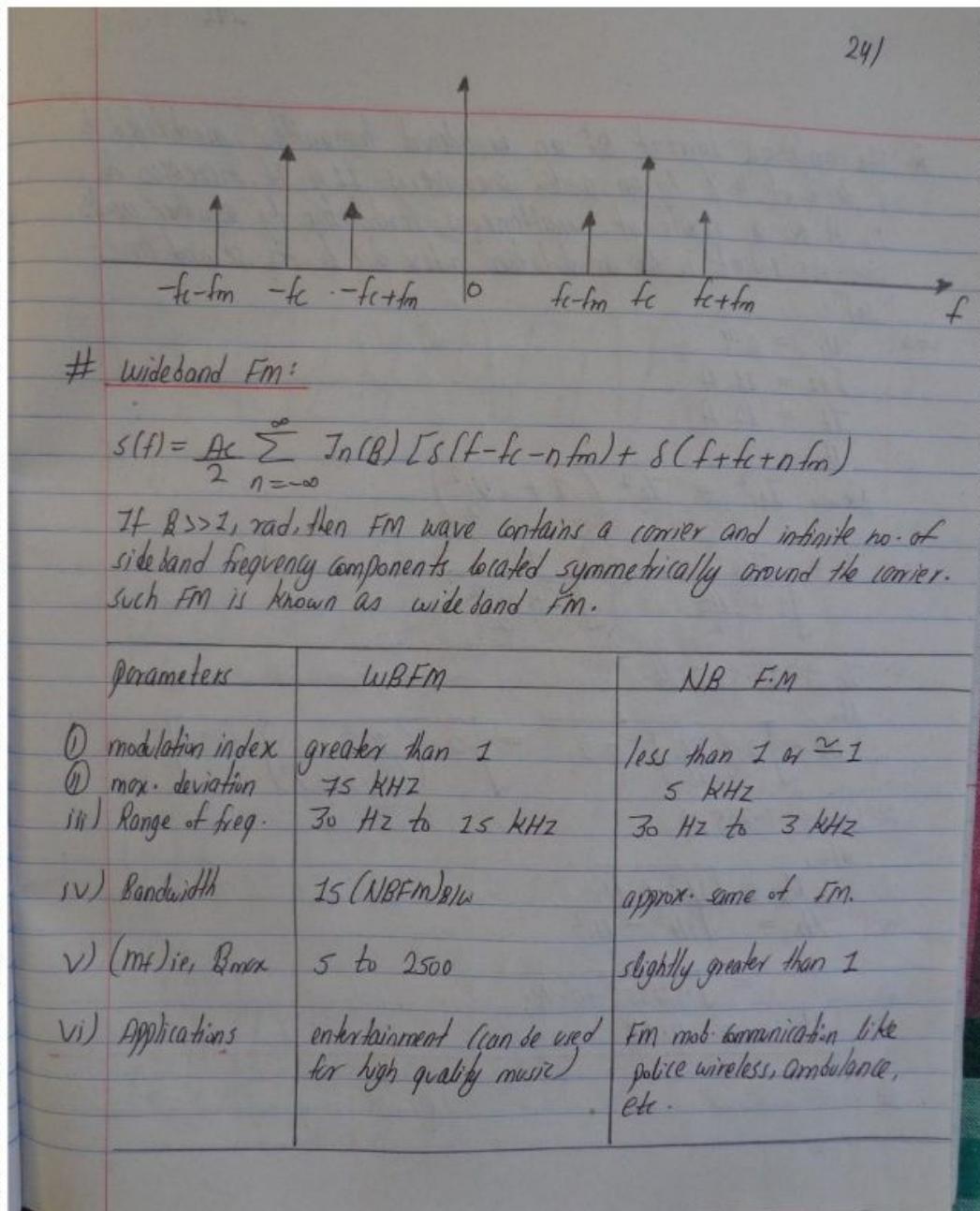


Fig: Spectrum

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R. The antenna current of an broadcast transmitter modulated to a depth of 40% by an audio sinewave, is 11 A. It increases to 12 A as a result of simultaneous modulation by another audio sinewave. What is the modulation index due to its second case.

Soln,

$$\Rightarrow M_1 = 0.4$$

$$I_{t1} = 11 \text{ A}$$

$$I_t = 12 \text{ A}$$

$$M_2 = ?$$

$$\text{Now, } I_{tr}^2 = I_c^2 \left(1 + \frac{M_1^2}{2} \right)$$

$$\text{or } I_c = \frac{I_{t1}}{\sqrt{1 + \frac{M_1^2}{2}}} = \frac{11}{\sqrt{1 + 0.4^2}}$$

$$\therefore I_c = 10.585 \text{ A.}$$

$$\text{Now, } M_t = \sqrt{2 \left(\frac{I_t}{I_c} \right)^2 - 1} = \sqrt{2 \left(\frac{12^2}{(10.585)^2} \right) - 1}$$

$$\therefore M_t = 0.75528$$

also,

$$M_t = \sqrt{M_1^2 + M_2^2}$$

$$\text{or } M_2 = \sqrt{M_t^2 - M_1^2}$$

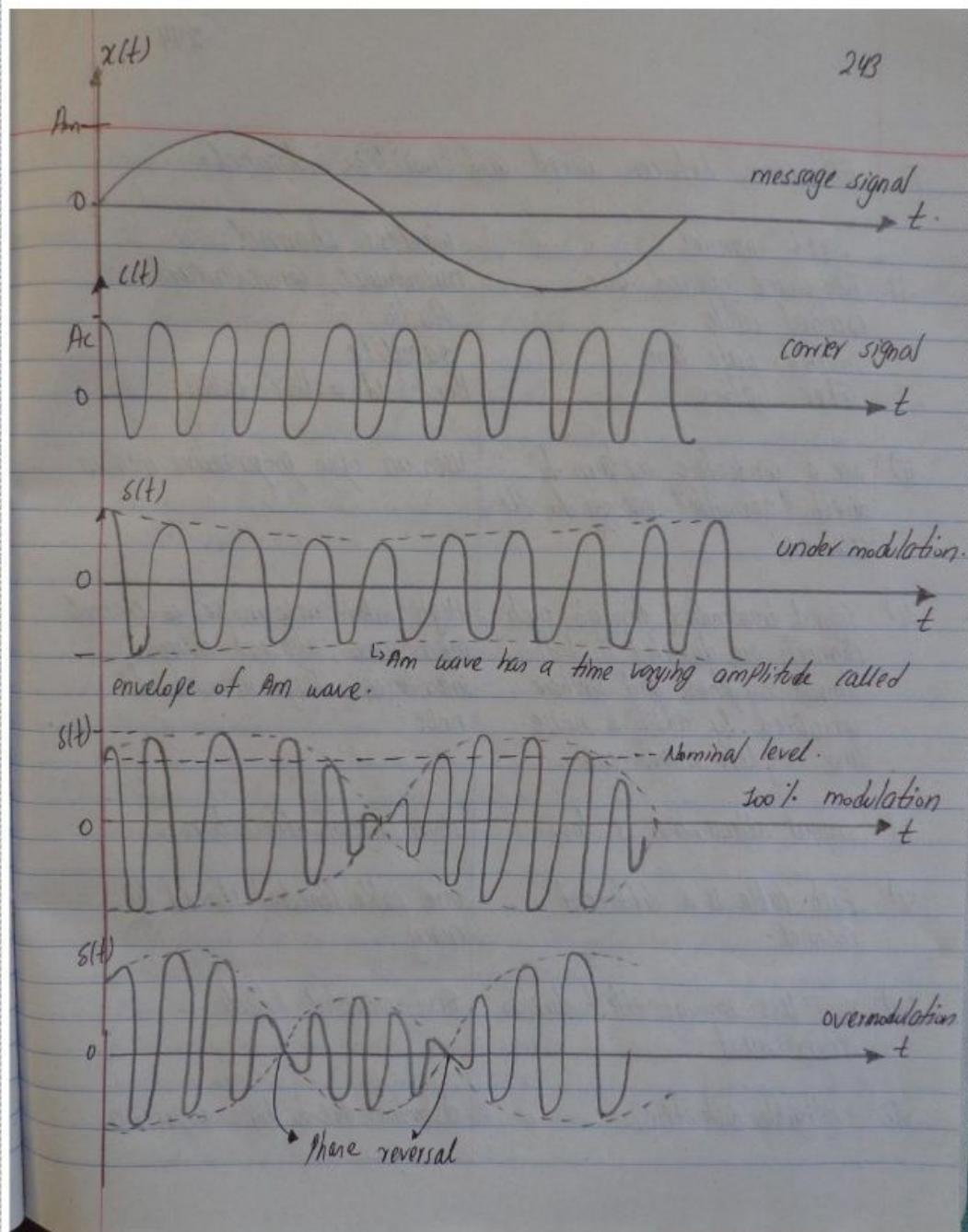
$$= \sqrt{0.755^2 - 0.4^2}$$

$$\therefore M_2 = 0.64$$

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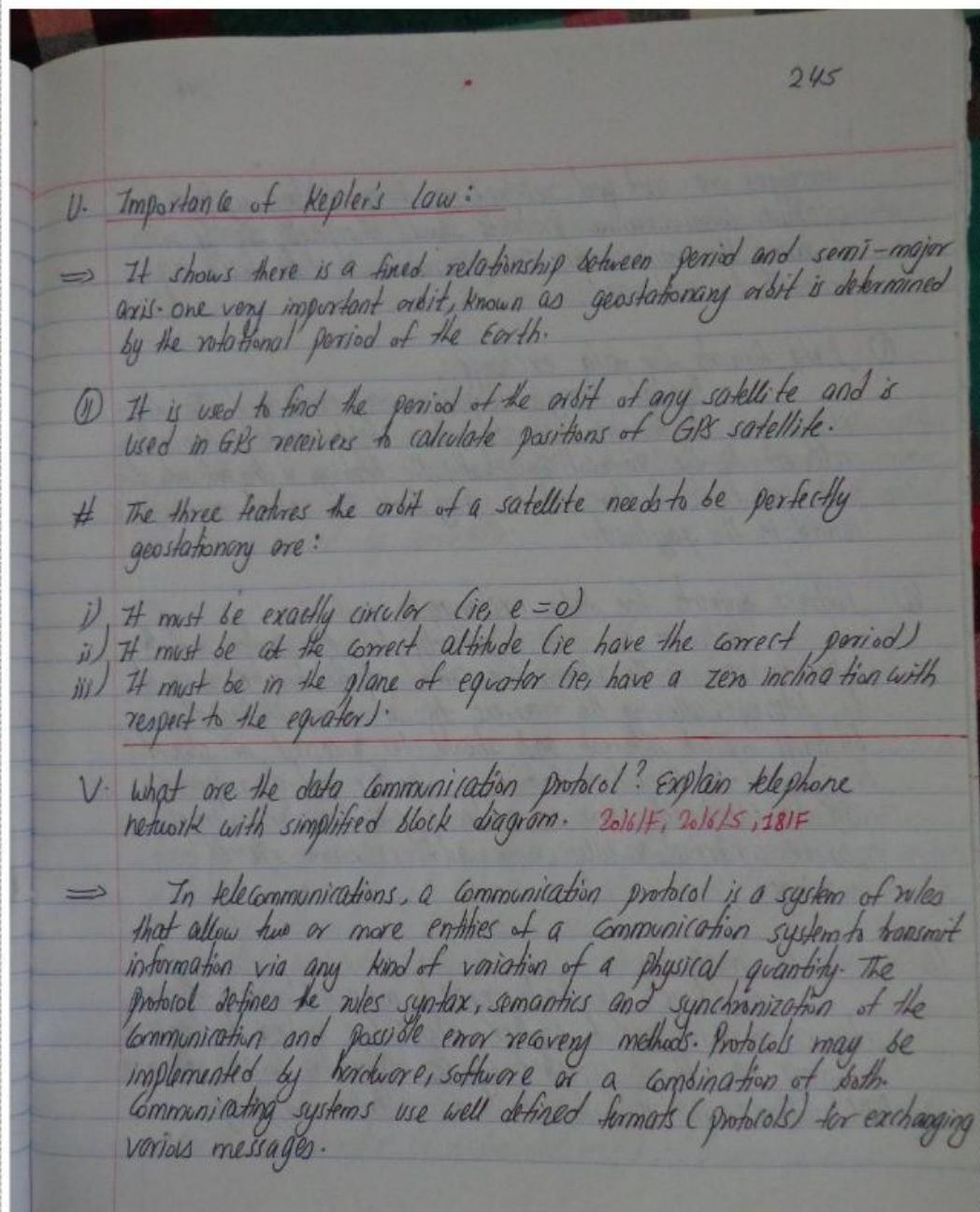
2016/15, 2017/18

T. Difference between wired and wireless channel.

wired channel	wireless channel
i) Two-wire opened line coaxial cable Twisted pair line Fiber optics	microwave communication Radio satellite. Hand held walkie-talkies
ii) use a conductive medium to direct transmitted energy to the receiver.	uses an open propagation medium.
iii) signal transmitted through such channels are distorted in both amplitude, phase and further corrupted by additive noise. and very less interference.	High interference, noise, co-channel interference, adjacent channel interference. Large amount of noise.
iv) signal attenuation is low.	High signal attenuation.
v) Each cable is a different channel.	one cable (media) shared by many
vi) very less amount effect due to environment.	Environmental effects.
vii) negligible distortion.	distortion due to time dispersion.

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messages are sent and received on communicating systems to establish communications. Protocols should therefore specify rules governing the transmission. In general, much of the following should be addressed.

I Data formats for data exchange:

Digital message bitstrings are exchanged. The bitstrings are divided in fields and each field carries information relevant to the protocol. Conceptually the bitstring is divided into two parts called the header and the payload. The actual message is carried in the payload.

II Address formats for data exchange:

Addresses are used to identify both the sender and the intended receiver. The address are carried in the header area of the bitstrings, allowing the receivers to determine whether the bitstrings are of interest and should be processed or should be ignored. A connection between a sender and a receiver can be identified using an address pair (sender address, receiver address). Usually some addresses have special meanings.

III Address mapping:

Sometimes protocols need to map addresses of one scheme on addresses of another scheme. For instance to translate a logical IP address specified by the application to an ethernet MAC address. This is referred to as address mapping.

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iv) Routing:

When systems are not directly connected, intermediary systems along the route to the intended receiver need to forward messages on behalf of the sender. On the internet, the networks are connected using routers.

v) Detection of Transmission errors:

Error detection is necessary on networks where data corruption is possible. In a common approach, CRCs of the data area are added to the end of packets, making it possible for the receiver to detect differences caused by corruption.

vi) Acknowledgements:

Acknowledgement of correct reception of packets is required for connection-oriented communications. It is sent from receivers back to their respective senders.

vii) Loss of information - timeout and retries:

Packets may be lost on the network or be delayed in transit. To cope with this, under some protocols, a sender may expect an acknowledgement of correct reception from the receiver within a certain amount of time. On timeouts, the sender must assume the packet was not received and retransmit it. Exceeding the retry limit is considered an error.

viii) Direction of information flow:

Direction needs to be addressed if transmissions can only occur in one direction at a time as on half-duplex-links or

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from one sender at time as on a shared medium. This is known as media access control. Arrangements have to be made to accommodate the case of collision or contention where two parties respectively simultaneously transmit.

viii) Sequence Control:

If long bitstreams are divided in pieces and then sent on the network individually, the pieces may get lost or delayed or on some types of network, they take different routes to their destination. As a result, pieces may arrive out of sequence. Retransmission can result in duplicate pieces.

ix) Flow Control:

Flow control is needed when the sender transmits faster than the receiver or intermediate network equipment can process the transmissions. Flow control can be implemented by messaging from the receiver to sender.

- W. A multimode step index fiber with a core diameter of 80 μm and a relative index difference of 1.5 %. is operating at a wavelength of 0.85 μm . If the core refractive index is 1.48, estimate the normalized frequency for the fiber and number of guided modes.

soln,

$$\Rightarrow \text{multimode step index fiber, } = 80 \mu\text{m} = 2a \\ \text{core radius, } = 40 \mu\text{m} = a \\ \text{Relative index difference, } \Delta = 1.5 \% = 0.015 \\ \text{wavelength, } \lambda = 0.85 \mu\text{m}$$

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Core refractive index, $n_1 = 1.48$

Normalized frequency, $v = ?$

number of modes, $m = ?$

We know,

Numerical aperture, $NA = n_1 (2A)^{1/2} = 0.2563$

Normalized frequency is.

$$v = \frac{2\pi A}{\lambda} \cdot NA$$

$$v = \frac{2\pi \times 40}{0.85} \times 0.2563$$

$$\therefore v = 75.78$$

$$\therefore \text{Number of modes, } m = \frac{v^2}{2}$$

$$= \frac{(75.78)^2}{2}$$

$$\therefore m = 2871.50$$

X Prove that: $S_{SSB}(t) = \frac{Ac}{2} [m(t)\cos 2\pi f_ct + \hat{m}(t)\sin 2\pi f_ct]$

where,

$S_{SSB}(t)$ is the SSB modulated signal and $c(t) = Ac \cos 2\pi f_c(t)$

So,

\Rightarrow , Let,

$m(t) = Am \cos 2\pi f_m t \rightarrow \text{monochromatic signal}$

We have,

$$S_{SSB-SC}(t) = c(t) \cdot m(t)$$

$$= Ac \cos 2\pi f_c t \cdot Am \cos 2\pi f_m t$$

$$= \frac{AcAm}{2} [\cos 2\pi(f_c + f_m)t + \cos 2\pi(f_c - f_m)t]$$

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$$S_{USB}(t) = \frac{A_c A_m}{2} \cos 2\pi (f_c + f_m)t$$

$$= \frac{A_c A_m}{2} [\cos 2\pi f_c t + \cos 2\pi f_m t - \sin 2\pi f_c t - \sin 2\pi f_m t]$$

$$= \frac{A_c}{2} [m(t) \cos 2\pi f_c t - \hat{m}(t) \sin 2\pi f_c t]$$

Thus,

$$\therefore S_{USB}(t) = \frac{A_c}{2} m(t) \cos 2\pi f_c t - \frac{A_c}{2} \hat{m}(t) \sin 2\pi f_c t$$

Similarly,

$$S_{LSB}(t) = \frac{A_c A_m}{2} \cos 2\pi (f_c - f_m)t$$

$$= \frac{A_c}{2} m(t) \cos 2\pi f_c t + \frac{A_c}{2} \hat{m}(t) \sin 2\pi f_c t$$

Hence,

$$\therefore S_{SSB}(t) = \frac{A_c}{2} [m(t) \cos 2\pi f_c t \mp \hat{m}(t) \sin 2\pi f_c t]$$

-ve sign for USB
+ve " " LSB.

Y Cell sectoring and cell splitting : Figure :

Copy theory from page 204, 18/F

Sectoring:

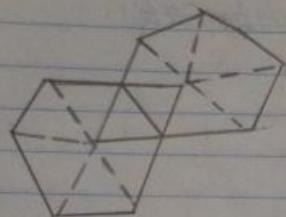
Use directional antennas to further control the interference and frequency reuse of channels.

Example: omni, 120°, 60°, 90°,

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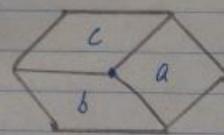
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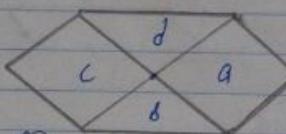


① Omni

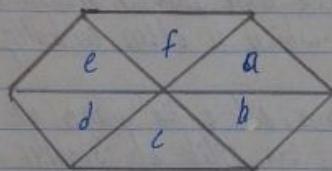
Fig: cell sectoring



② 120° sector



③ 60° sector



④ 90° sector

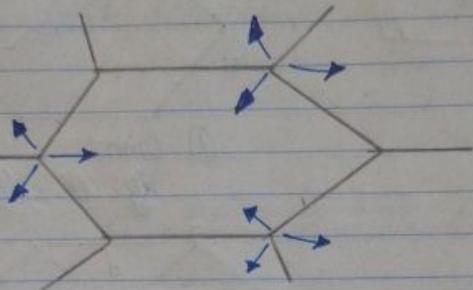
- Sectorization consists of dividing an omnidirectional (360°) view from the cell site into non-overlapping slices called sectors. When combined, sectors provides the same coverage but they are considered to be separate cells.
- Also, considered as one of easy and inexpensive capacity increasing solution. The base station can either be located at,
 - center of original (large) cell , or
 - corners of " " "
- Sectorization is less expensive than cell splitting, as it does not

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require the acquisition of new base station sites.



cell splitting!

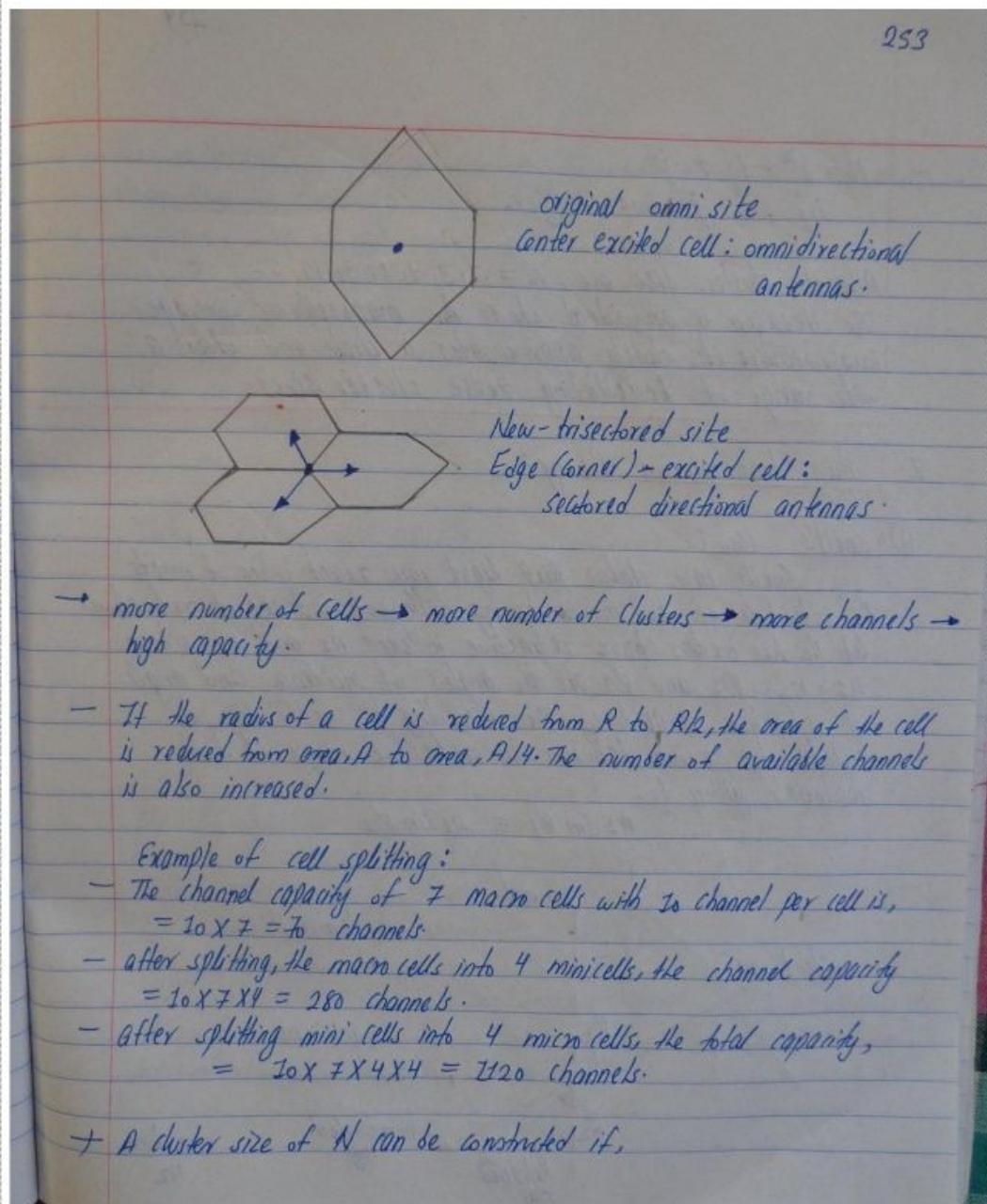
The unit area of RF coverage for cellular network is called a cell. In each cell, a base station transmits from a fixed cell site location, which is often centrally located in the cell.

In base stations, where the usage of cellular network is high, these cells are split into smaller cells. The radio frequencies are realigned, and transmission power is reduced. A new cell site must be constructed when a cell is split. Splitting the cells into smaller ones also lead to a new solution called cell sectoring.

- smaller cell size, small transmitting power and reduces cochannel interference.
- To increase capacity, split each cell into 3 using sectorized antennas.

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- $N = i^2 + ij + j^2$
 i, j are positive integer.

Allowable clusters size are, $N = 1, 3, 4, 7, 9, 12, \dots$

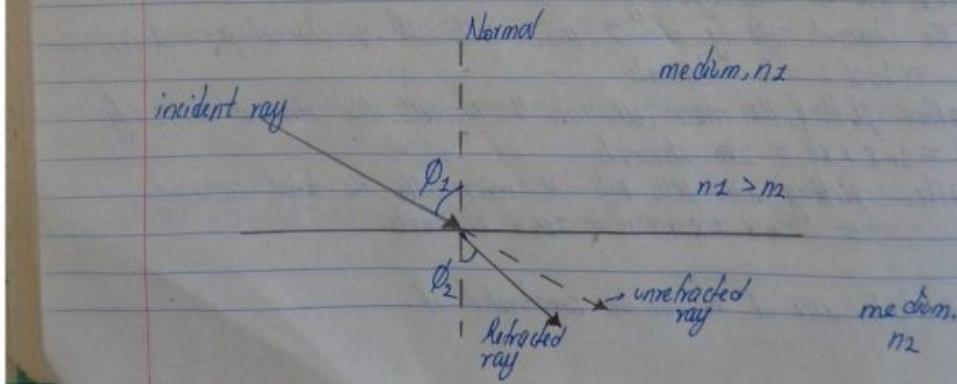
The Hexagon is an ideal choice for macrocellular coverage areas, because it closely approximates a circle and offers a wide range of tessellating reuse cluster sizes.

2. Ray transmission theory:

① Snell's law:

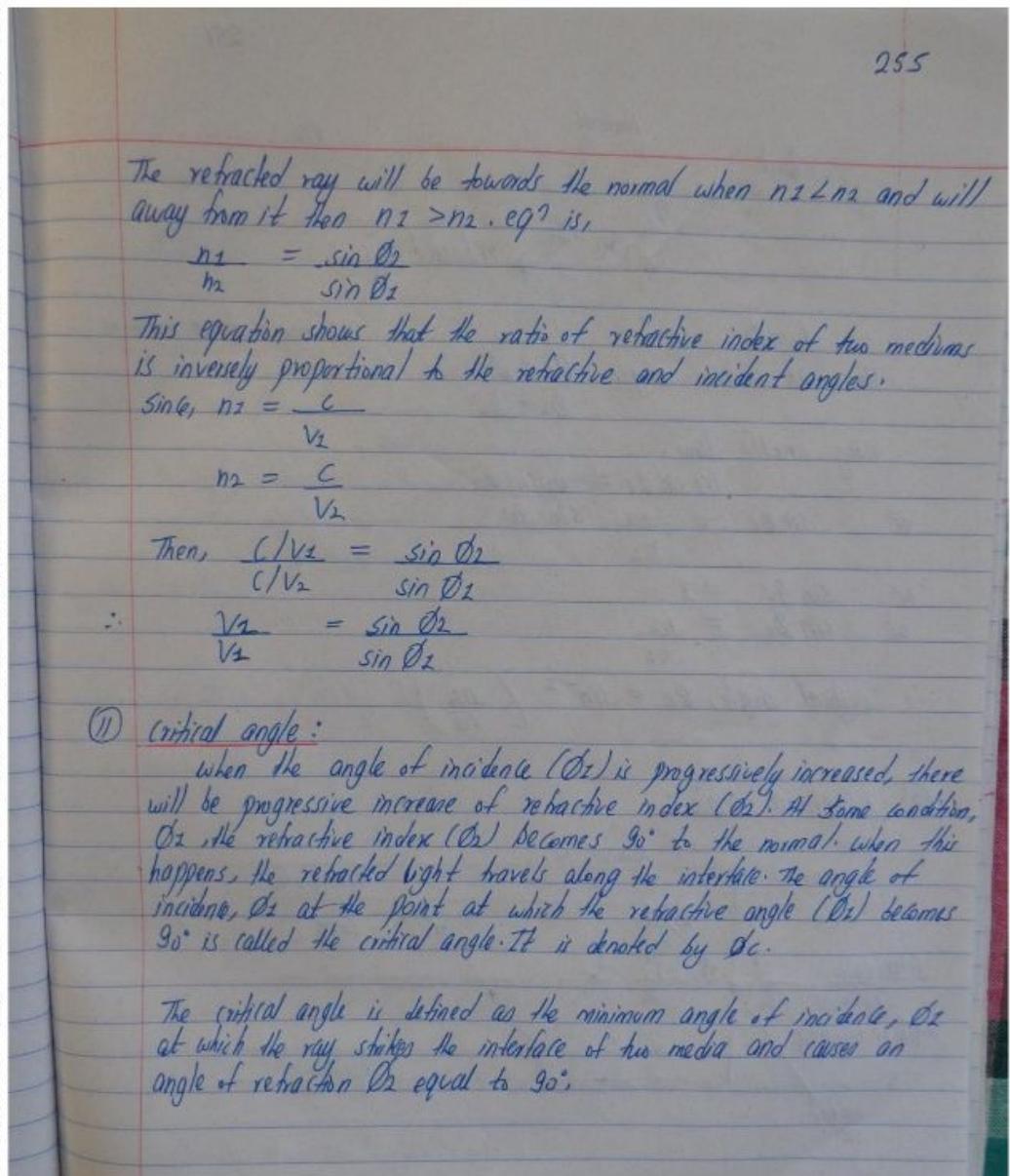
Snell's law states how light rays react when it meets the interface of two media having different indexes of refraction. Let the two medias have refractive indexes n_1 and n_2 where $n_1 > n_2$. θ_1 and θ_2 are the angles of incidence and angle of refraction respectively, Then according to snell's law, a relationship exists between the refractive index of both materials given by.

$$n_2 \sin \theta_1 = n_1 \sin \theta_2$$



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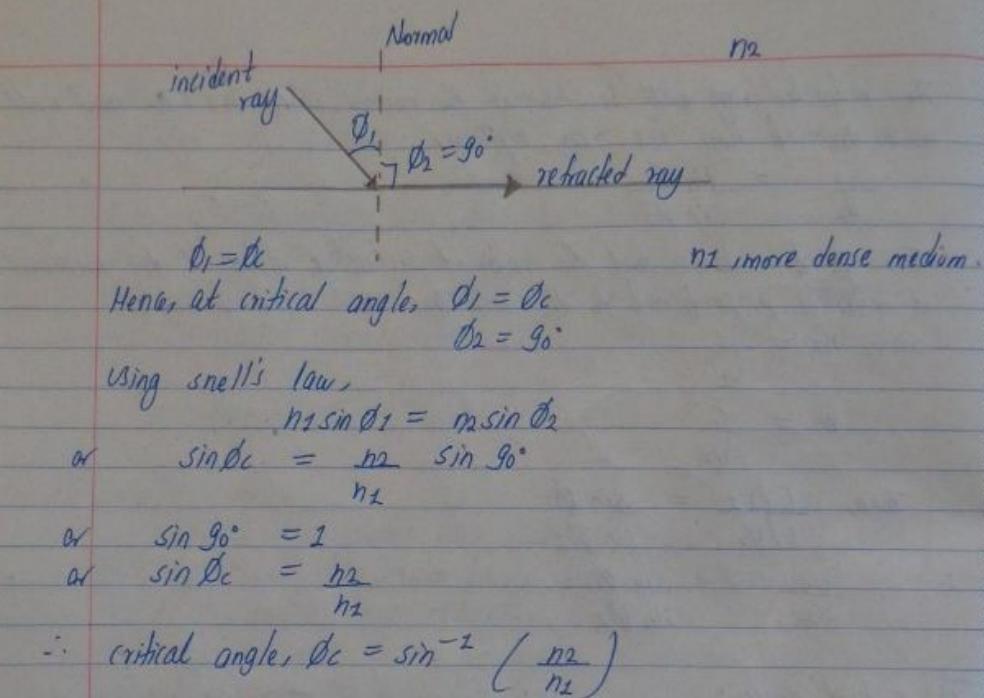
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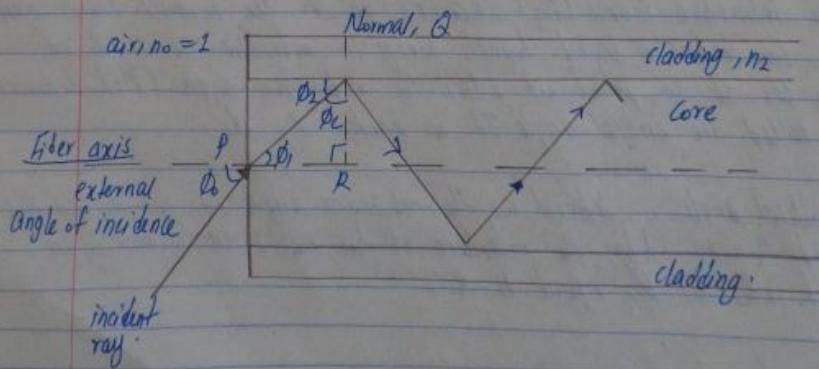
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⑩ Acceptance angle :



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Applying snell's law to external incidence angle,

$$n_0 \sin \theta_0 = n_2 \sin \theta_2$$

$$\text{But, } \theta_2 = (90^\circ - \phi_c)$$

$$\sin \theta_2 = \sin (90^\circ - \phi_c) = \cos \phi_c$$

Hence,

$$n_0 \sin \theta_0 = n_2 \cos \phi_c$$

$$\text{or } \sin \theta_0 = \frac{n_2 \cdot \cos \phi_c}{n_0}$$

Applying pythagoras theorem to $\triangle PQR$,

$$\cos \phi_c = \sqrt{n_1^2 - h_2^2} / h_2$$

$$\text{or } \sin \theta_0 = \frac{n_2}{h_0} \left[\sqrt{n_1^2 - \frac{n_2^2}{n_1}} \right]$$

$$\therefore \sin \theta_0 = \frac{\sqrt{n_1^2 - h_2^2}}{h_0}$$

Hence,

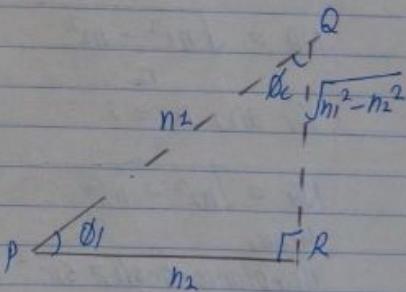
$$\phi_c = \sin^{-1} \frac{\sqrt{n_1^2 - h_2^2}}{n_0}$$

The maximum value of external incidence angle for which light will propagate in the fiber is.

$$\therefore \phi_{\max} = \sin^{-1} \frac{\sqrt{n_1^2 - h_2^2}}{n_0}$$

When the light ray enters the fiber from air as medium, $n_0 = 1$,

$$\therefore \phi_{\max} = \sin^{-1} \sqrt{n_1^2 - h_2^2}$$



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The angle θ_0 is called an acceptance angle and θ_{\max} defines the maximum angle in which the light ray may incident on fiber to propagate down the fiber.

(III) Numerical Aperture:

The numerical aperture (NA) of a fiber is a figure of merit which represents its light gathering capability. Larger the numerical aperture, the greater the amount of light accepted by fiber. The acceptance angle also determines how much light is able to enter the fiber and hence there is relation between the numerical aperture and the cone of acceptance.

$$NA = \sin \theta_{\max}$$

$$NA = \sqrt{n_1^2 - n_2^2}$$

For air, $n_0 = 1$

$$NA = \sqrt{n_1^2 - n_2^2}$$

Hence,
acceptance angle = $\sin^{-1}(NA)$.

NA is not a function of fiber dimension.

The index difference (Δ) and the numerical aperture (NA) are related to the core and cladding indices.

$$\Delta = \frac{n_1 - n_2}{n_2}$$

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or $\Delta = \frac{NA^2}{2n_1^2}$

Also, $NA = \sqrt{n_1^2 - m^2}$

$NA = \sqrt{(n_1^2 - m^2)}$

$\therefore NA = n_1 (2\Delta)^{1/2}$

Ques: A continuous 72 km long optical fiber link has a loss of 1.5 dB/km.

(i) what is the minimum optical power level that must be launched into the fiber to maintain an optical power level of 0.3 mW at the receiving end?

(ii) what is the required input power if the fiber has a loss of 2.5 dB/km.

Soln,

$\Rightarrow z = 72 \text{ km}$

$\alpha = 1.5 \text{ dB/km}$

$P(0) = 0.3 \text{ mW}$

(i) Attenuation in optical fiber;

$$\alpha = 10 \times \frac{1}{2} \log \left(\frac{P(0)}{P(z)} \right)$$

or $1.5 = 10 \times \frac{1}{2} \log \left(\frac{0.3 \text{ mW}}{P(z)} \right)$

or $\log \left(\frac{0.3 \text{ mW}}{P(z)} \right) = \frac{1.5}{10} = 0.15$

or $\frac{0.3 \text{ mW}}{P(z)} = 10^{0.15} = 1.80$

or $P(z) = \frac{0.3 \text{ mW}}{1.80} = 0.166 \text{ mW}$

$\therefore P(z) = 1.66 \times 10^{-4} \text{ watt}$

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optical power output = $4.76 \times 10^{-9} \text{ W}$

① input power = ? = $P_{(0)}$
when, $\alpha = 2.5 \text{ dB/km}$,
$$\alpha = 10 \times \frac{1}{2} \times \log \left(\frac{P_{(0)}}{P_{(2)}} \right)$$

or $2.5 = 10 \times \frac{1}{2} \log \left(\frac{P_{(0)}}{4.76 \times 10^{-9}} \right)$
or $\log \left(\frac{P_{(0)}}{4.76 \times 10^{-9}} \right) = \frac{2.5}{0.833} = 3$
or $\frac{P_{(0)}}{4.76 \times 10^{-9}} = 10^3 = 1000$
or $P_{(0)} = 4.76 \text{ mW}$
Hence,
input power, $P_{(0)} = 4.76 \text{ mW}$.

2- A. Round trip time delay: 2014/3, 2015/F, 2015/S, 181F

Copy from [191] pages:

Round trip time, RTT, also known as round trip delay time; is the time taken for a signal to be sent from a transmitter to a receiver plus the time it takes to verify that the signal has been received; hence, RTT is the time taken by a signal to be sent from one location to another and back again. RTT always relates to telecommunications, but may refer to the internet or radar systems, satellite communications.

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How round trip time works ?

Round trip time is dependent on a number of factors, including the data transfer rate of the source, the nature of the transmission medium, the distance between the source and the destination, the number of nodes between the source and the destination, the amount of traffic or bandwidth on the network that is being used, the number of other requests being handled by the receiver or nodes along the transmission path, the processing capabilities of the source, receiver and nodes and the presence of interference.

Advantages:

Calculating a transmitter's RTT is advantageous because it allows user and operator to identify how long a transmission will take to complete and how fast a network can operate. RTT in both internet and satellite transmission is particularly important because it allows network providers to establish data speeds and allocate bandwidth to subscribers accordingly.

$$\text{Round trip delay} \geq 2 \times \frac{\text{distance}}{\text{velocity of light}}$$

LEO, MEO, GEO satellites, assuming the speed of light is 3×10^8 m/s, if the maximum acceptable delay for a voice system is 30 ms, which of these satellite systems would be acceptable for a two way voice communication?

- LEO \sim 13.3 ms
- MEO \sim 733 ms
- GEO \sim 239 ms

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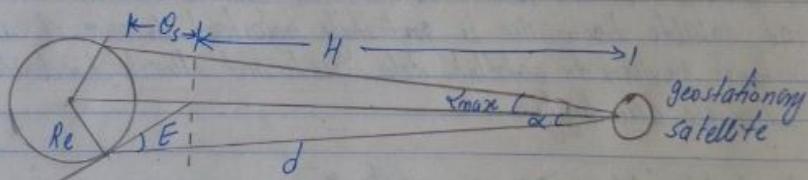
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- only LEO and MEO satellites below 4500 km have delay less than 30 ms.

$$\therefore T = \frac{d}{c} = 2 \times \frac{(35,786 \text{ km})}{3 \times 10^5 \text{ km/s}} = 238 \text{ ms}$$

This includes the time delay within the earth station (uplink + downlink) and satellite equipment. It takes more than a quarter of a second for an electromagnetic wave to travel from an Earth station to a satellite and back where earthstation is located on earth directly below satellite.

Round trip time delay of Geostationary satellite :



We know,

$$R_E = 6378 \text{ km}$$

$$H = 35,786 \text{ km}$$

Also,

$$d^2 = R_E^2 + (R_E + H)^2 - 2(R_E + H)\sin\left[E + \sin^{-1}\frac{R_E \cdot \cos E}{R_E + H}\right]$$

Here,

d = slant range

R_E = radius of earth

H = altitude of satellite = L

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E = angle of elevation
 θ_s = central angle.
 α = coverage angle.
 $E_{\min} = 5^\circ$

So,
∴ maximum slant range (d) = $6378^2 + (6378 + 35786)^2 - 2 \cdot (6378 + 35786) \sin \left[5 + \sin^{-1} \left(\frac{6378}{6378 + 35786} \right) \times \cos 5^\circ \right]$

∴ $d = 41,127.8 \text{ km}$

satellite round trip time delay. = $2 \cdot d / c$

c = velocity of light = $3 \times 10^5 \text{ km/s}$
Thus,

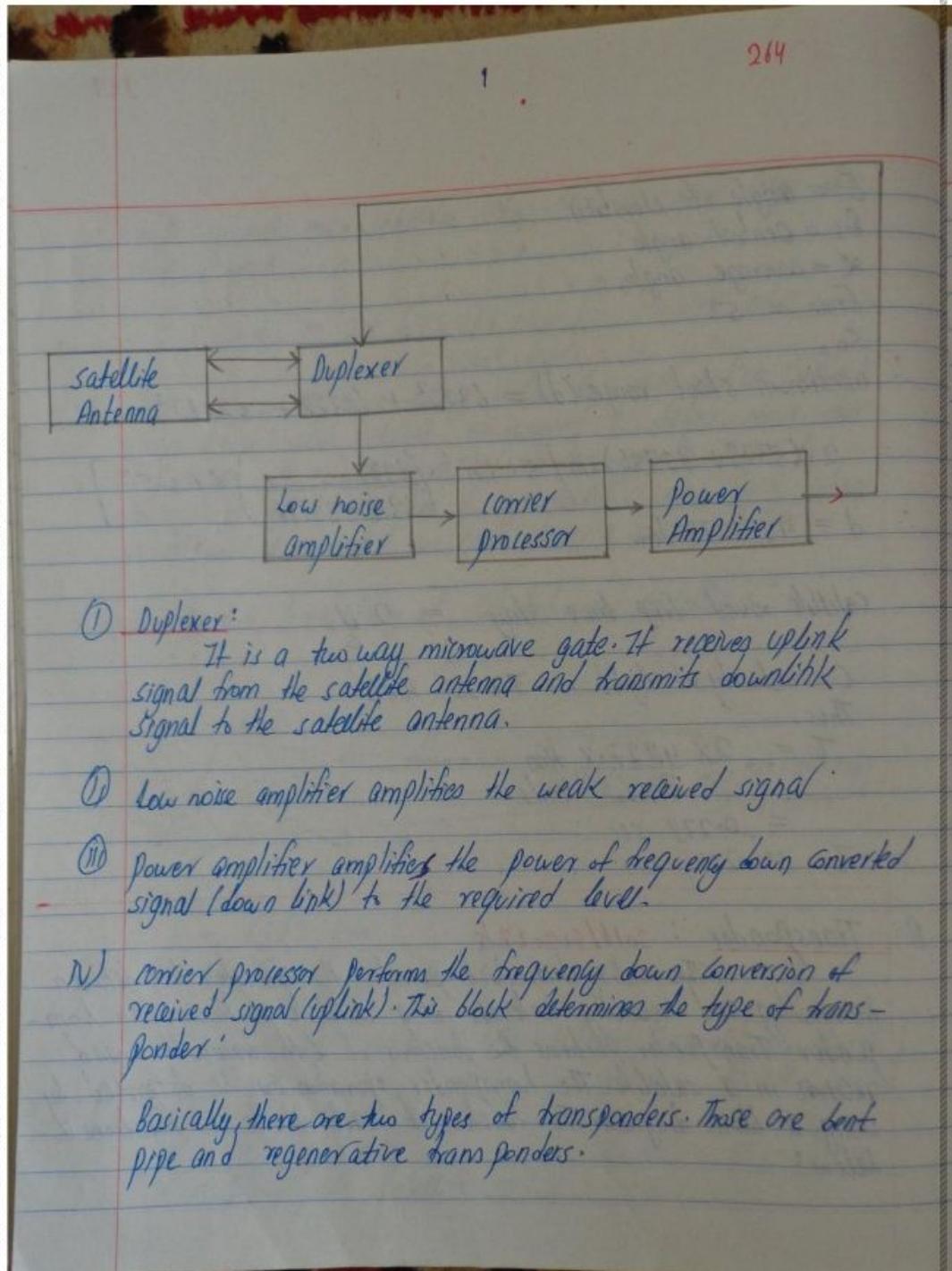
$T = \frac{2 \times 41127.8 \text{ km}}{3 \times 10^5 \text{ km/s}}$
= 0.274 sec
∴ $T_r = 274 \text{ ms}$

B. Transponder : 2016/5/20 17:15, 181F

The subsystem, which provides the connecting link between transmitting and receiving antennas of a satellite is known as transponder. Transponder performs the functions of both transmitter and receiver in a satellite. The transponder operation can be described by using a block diagram. The function of each block is mentioned below:

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① Bent pipe Transponders:

It receives microwave frequency signal. It converts the frequency of input signal to RF signal frequency & then amplifies it. It is also called as repeater and conventional transponder. It is suitable for both analog and digital signals.

② Regenerative Transponders:

It performs the function of bent pipe transponders. i.e., frequency translation and amplification. In addition to these two functions, it also performs the demodulation of RF carrier to baseband regeneration of signals and modulation. It is also called as processing transponder. It is suitable for only digital signals. Its main advantages is improvement in SNR and have more flexibility in implementation.

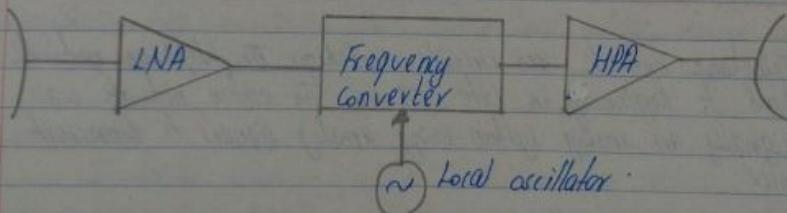


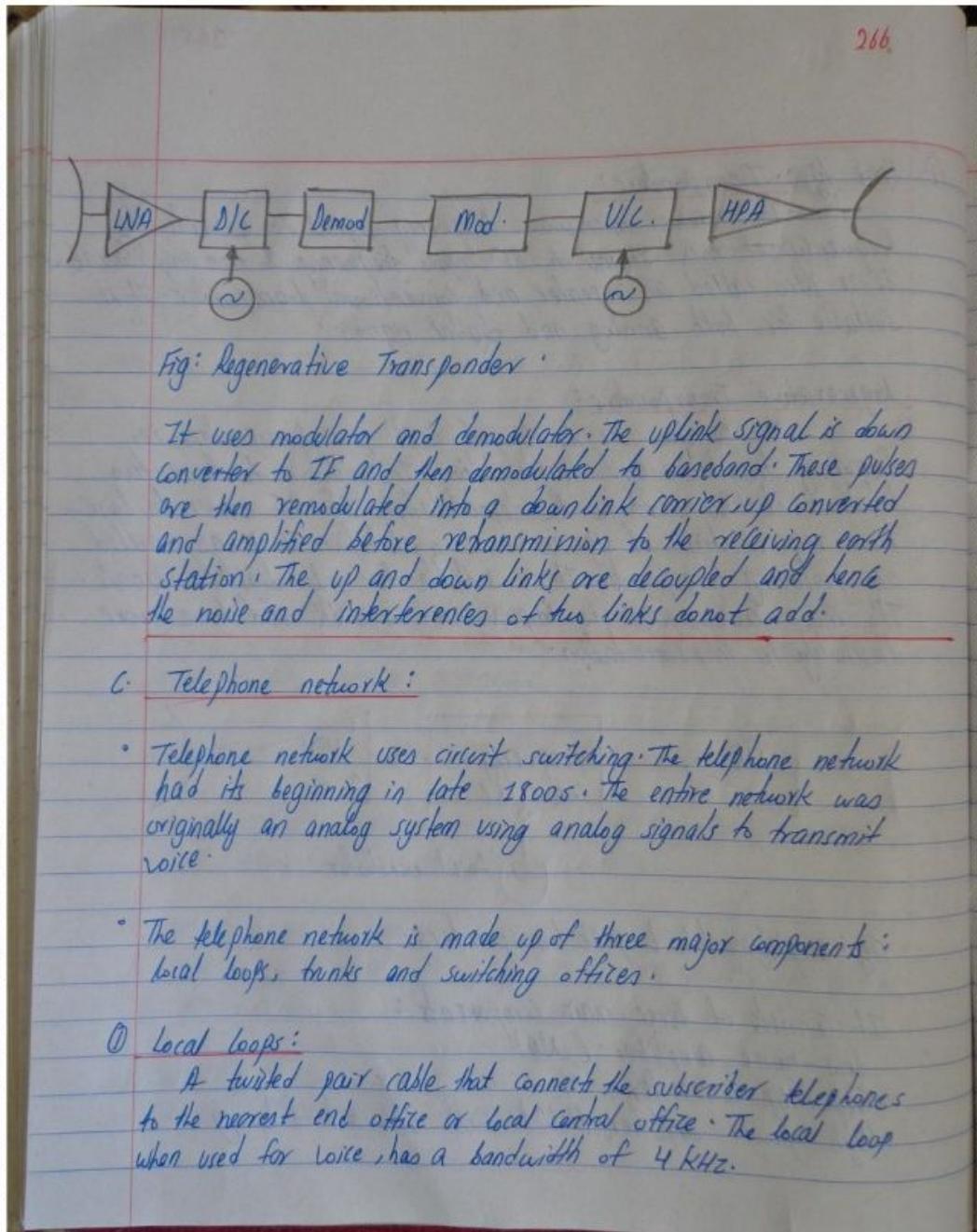
Fig: Amplifying / bent pipe transponders.

It consists of three main components :

- Low-noise amplifier (LNA)
- Frequency converter
- High power amplifier (HPA) with relevant filtering

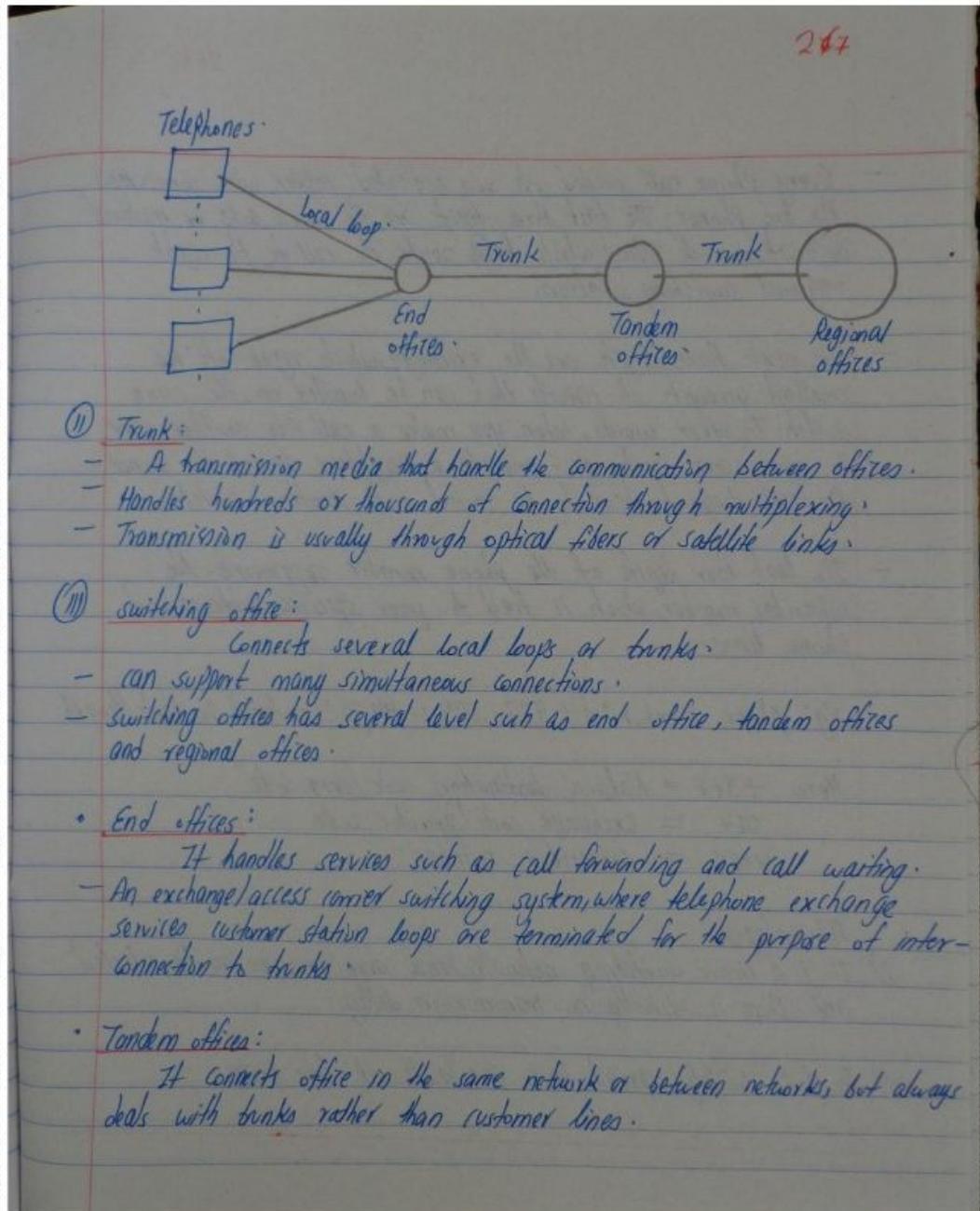
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- Every phone call needed its own dedicated copper wire connecting the two phones. The first three digits are the area code or national destination code (Ndc) which helps route the call to the right regional switching station.
- The next three digits are the exchange, which represents the smallest amount of circuits that can be handled on the same switch. In other words, when you make a call to another user in your same exchange -- maybe a neighbour around the corner -- the call does not have to be routed onto another switch.
- The last four digits of the phone number represents the subscriber number, which is tied to your specific address and phone lines.

Eg: phone number: [+ 977 - 014 - 4450] [- fixed 10 digits/length]

Here, + 977 = National destination code/area code

014 = exchange code/provider code

4450 = subscriber number.

Advantages :

- i) It is a circuit switching network, hence any receiver can be selected and there is virtually no transmission delay.
- ii) As it is widely spread, hence available at low price.

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D. Error correction technique : 2017/F, 2015/S, 2015/F, 2016/F, 2014/S.

Copy from page 151, and

In the digital world, error correction can be done in two ways :

- Backward error correction :

When the receiver detects an error in the data received, it requests back the sender to retransmit the data unit.

- Forward error correction :

When the receiver detects some error in the data received, it executes error correcting code, which helps it to autorecover and to correct some kinds of errors.

The first one, backward error correction is simple and can only be efficiently used where retransmitting is not expensive. For example, fiber optics. But in case of wireless transmission, retransmitting may cost too much. In the latter case, forward error correction is used.

① Hamming Code :

It is a single bit error correction method using redundant bits. In this method, redundant bits are included with the original data. Now the bits are arranged such that different incorrect bits produce different error results and the corrupt bit can be identified. Once the bit is identified, the receiver can reverse its value and correct the error.

Hamming code can be applied to any length of data unit and uses the relationships between the data and the redundancy bits.

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$$2^n \geq m+n+1$$

$m = \text{no. of bits in data character}$
 $n = " \text{ " hamming bit.}$

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Algorithm :

- Parity bits are positions at the power of two (2^r)
 - Rest of the positions is filled by original data.
 - Each parity bit will take care of its bits in the code.
 - Final code will send to the receiver.

∴ Hamming codes = data bits + parity bits.
To transmit an 8-byte using Hamming codes, 12 bits are required:

 - Bits at positions, 1, 2, 4, 8 are used as parity bits.
 - " " other positions - are used to store the data to send.

Bit position

12	11	10	9	8	7	6	5	4	3	2	1
D8	D7	D6	D5	P4	D4	D3	D2	P3	D1	P2	P1

↓ ↓ ↓

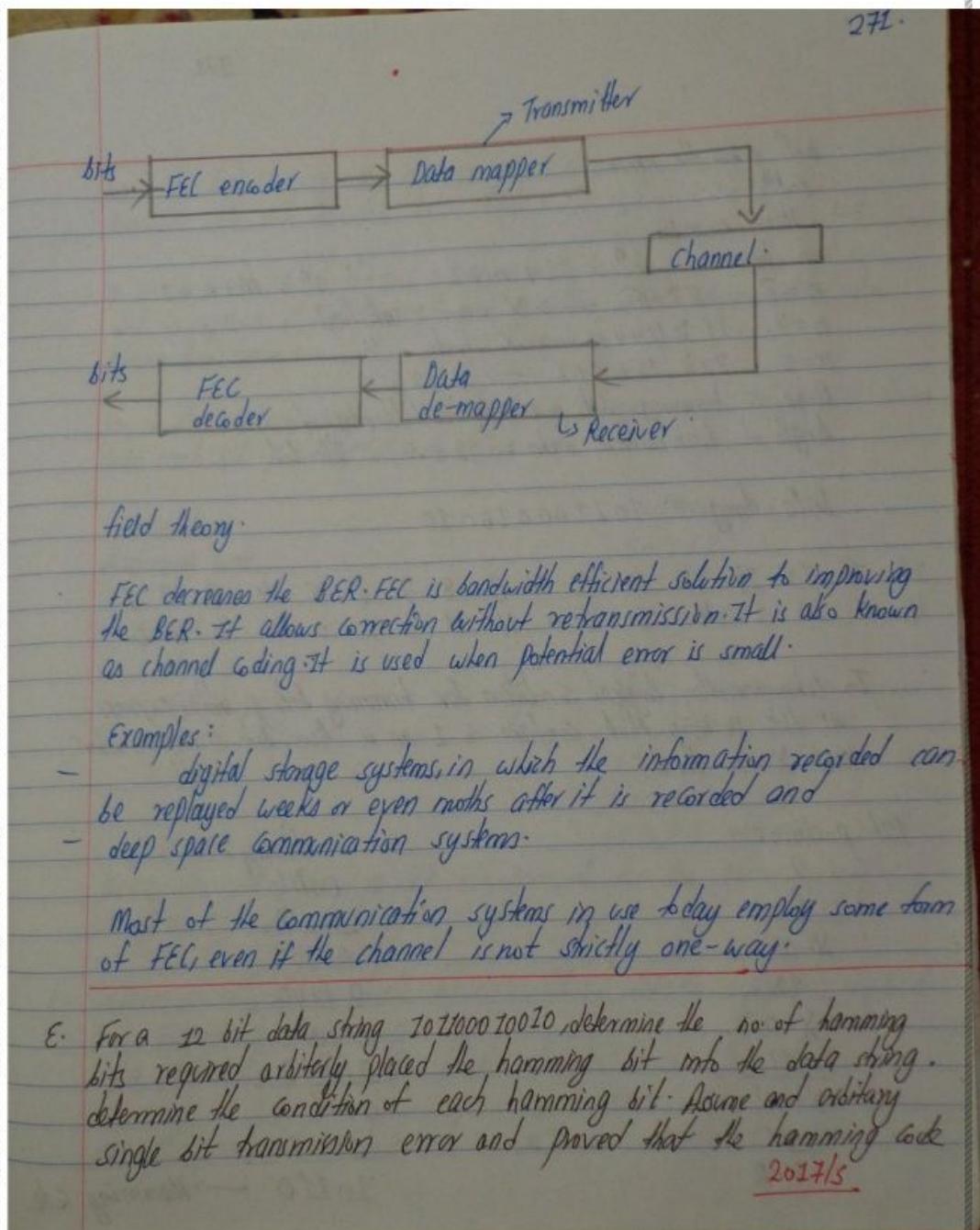
error correcting bits.

① Forward error correction (FEC):

In a one-way communication system, the transmission or recording is strictly in one direction, from transmitter to receiver. Error control strategy must be FEC; i.e., they employ error correcting codes that automatically correct errors detected at the receiver. FEC is bandwidth efficient solution to increasing the BER. No need for retransmission. Based on highly complex mathematical field known as sets and finite

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field theory.

FEC decreases the BER. FEC is bandwidth efficient solution to improving the BER. It allows correction without retransmission. It is also known as channel coding. It is used when potential error is small.

Examples:

- digital storage systems, in which the information recorded can be replayed weeks or even months after it is recorded and
- deep space communication systems.

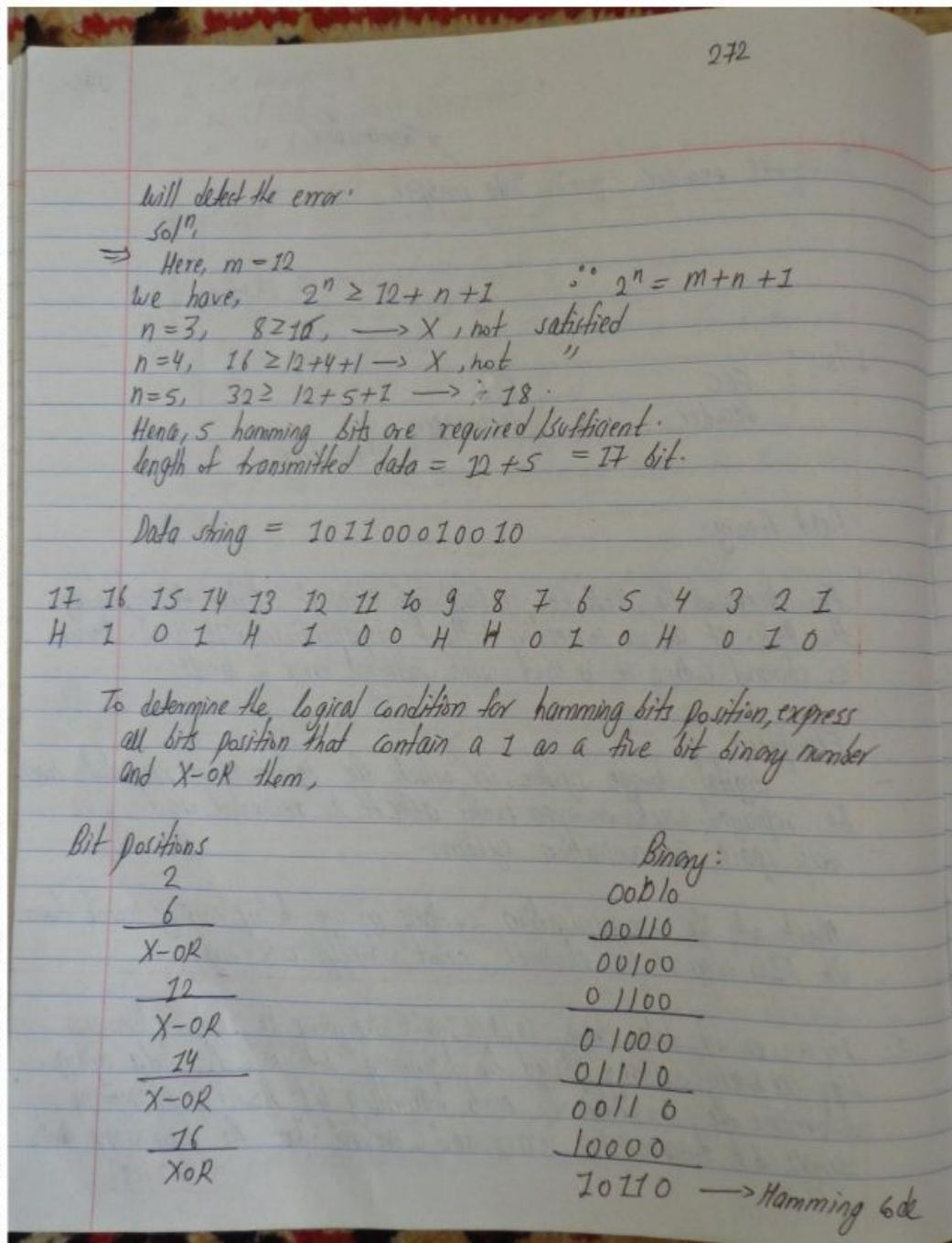
Most of the communication systems in use today employ some form of FEC, even if the channel is not strictly one-way.

E. For a 12 bit data string 101100010010, determine the no. of hamming bits required arbitrarily placed the hamming bit into the data string. determine the condition of each hamming bit. Assume an arbitrary single bit transmission error and proved that the hamming code

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$b_{17} = 1, b_{13} = 0, b_9 = 1, b_6 = 1, b_4 = 0$
Hence, the 77 bit encoded data string is,

1 1 0 1 0 1 0 0 1 1 0 1 0 0 0 1 0
let us suppose that the 14th bit position of the transmitted data string has error.

So in the receiver section extract that hamming bit and X-OR them with each bit positions that contains 1.

Bit position	Hamming code	Binary
7		1 0 1 1 0
6		0 0 0 1 0
5	X-OR	1 0 1 0 0
4		0 0 1 0 0
3	X-OR	1 0 0 0 0
2		0 1 1 0 0
1	X-OR	1 1 1 1 0
0		1 0 0 0 0
	X-OR	0 1 1 1 0 = 14

Hence, the bit position 14 has error, to fix the error, simply complement the 14th bit position.

F. Determine the block check sequence (BCS) for the following data & CRC generating polynomial.
data, $G(x) = x^7 + x^5 + x^4 + x^2 + x^1 + x^0$ or
 10110111 .

$CRC P(x) = x^5 + x^4 + x^2 + x^0$ or 110011 .

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$$\begin{aligned}
 & \Rightarrow \text{Now, 1st } G(x) \text{ is multiplied by the msb-bit.} \\
 & = x^5(x^7 + x^5 + x^4 + x^2 + x^1 + x^0) \\
 & = x^{12} + x^{10} + x^9 + x^7 + x^6 + x^5 \text{ or} \\
 & 101101110000
 \end{aligned}$$

Divide the result by $P(x)$.

110011) 10110111000000 (11010111
 ④ 110011
 0111101
 ④ 110011
 00111010
 ④ 110011
 00100100
 ④ 110011
 0101110
 ④ 110011
 0111010
 ④ 110011
 01001
 CR → 01001

So, the CRC is appended to the data to give following data stream.

At the receiver the transmitted data are divided by $R(x)$,

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$$B(s) = G(x) - CRC$$
$$\begin{array}{r} 110011) 1011011101001 (1101011 \\ \underline{-110011} \\ .111101 \\ -110011 \\ .111010 \\ -110011 \\ .100110 \\ -110011 \\ .101010 \\ -110011 \\ .110011 \\ -110011 \\ 000000 \rightarrow \text{Remainder} = 0 \end{array}$$

Since, remainder = 0

No error occurred.

BSC = 1011011101001.

G. Private line circuit are leased on 24-hour basis, so explain the advantages over conventional dial up circuits. 2017/5

=> In wired telephony, a private line or tie-line is a service that involves dedicated circuits, private switching arrangements and/or pre-defined transmission paths, whether virtual or physical which provide communications between specific locations.

Advantages of private line:

- High security as one dedicated link is provided.
- Instant data transmission.

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- Always available as it is one's own link and can be used when pleased.
- Symmetric : which means one can upload and download at the same speed.
- Flexible Connection : Leased line can be used for VPN access, phone calls and internet traffic all in one.
- Reliable :
- Private and secure .
- Unlimited usage .
- Transparent Costs
- Wide coverage area.

Disadvantages of leased line & dedicated line :

- expensive
- more number of CPE ports are required.
- Network complexity increases as network grows.
- number of links required grows exponentially if full mesh connectivity is required and network expands.

H. Advantages of FM over AM:

- Less sensitivity to noise . Noise amplitude variations that are eliminated in the receiver.

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- more efficient use of power.
- Improved system fidelity.
- operate in High Frequency range.
- In FM interfering signals on the same frequency are rejected. This is known as the capture effect.
- FM signal has constant amplitude and there is no need to use linear amplifiers to increase power levels. This increases transmitter efficiency.
- Bandwidth is large hence wide channel is required.
- The amplitude of an FM wave remains constant. This provides the systems designers an opportunity to remove the noise from the received signal. This is done in FM receivers by employing an amplitude limiter circuit so that the noise above the limiting amplitude is suppressed.

Disadvantages of FM over AM :

- The equipment of FM system is more complex than AM systems.
- Increased Bandwidth.
- excessive use of spectrum
- Area covered by FM is limited to line of sight area but AM coverage area is large.

I. The bit stream 00100 1100 1111 is to be transmitted using DPSK. Determine the encoded sequence & transmitted phase sequence. 2017k

=> Data stream = 00100 1100 1111

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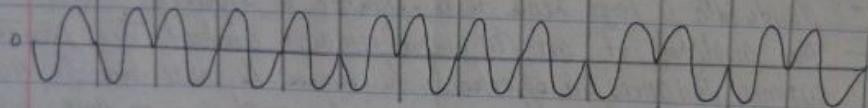
msg, information:	0	0	1	0	0	1	1	0	0	1	1	1	1	1
$d_k = 0$	1	0	0	1	1	1	1	0	1	0	1	0	1	1
$d_{k-1} = 1$	0	0	0	1	1	1	0	1	1	1	0	1	0	0
o/p	0	0	1	0	0	1	1	0	0	1	1	1	1	1

m_k = message / Signal.

Hence,

$m_k = o/p$.

o/p 0 0 1 0 0 1 1 0 0 1 1 1 1 1



For symbol 0 to phase shift by 180° and symbol 1 to leave the phase.

$$\therefore o/p = d_k \oplus d_{k-1}$$

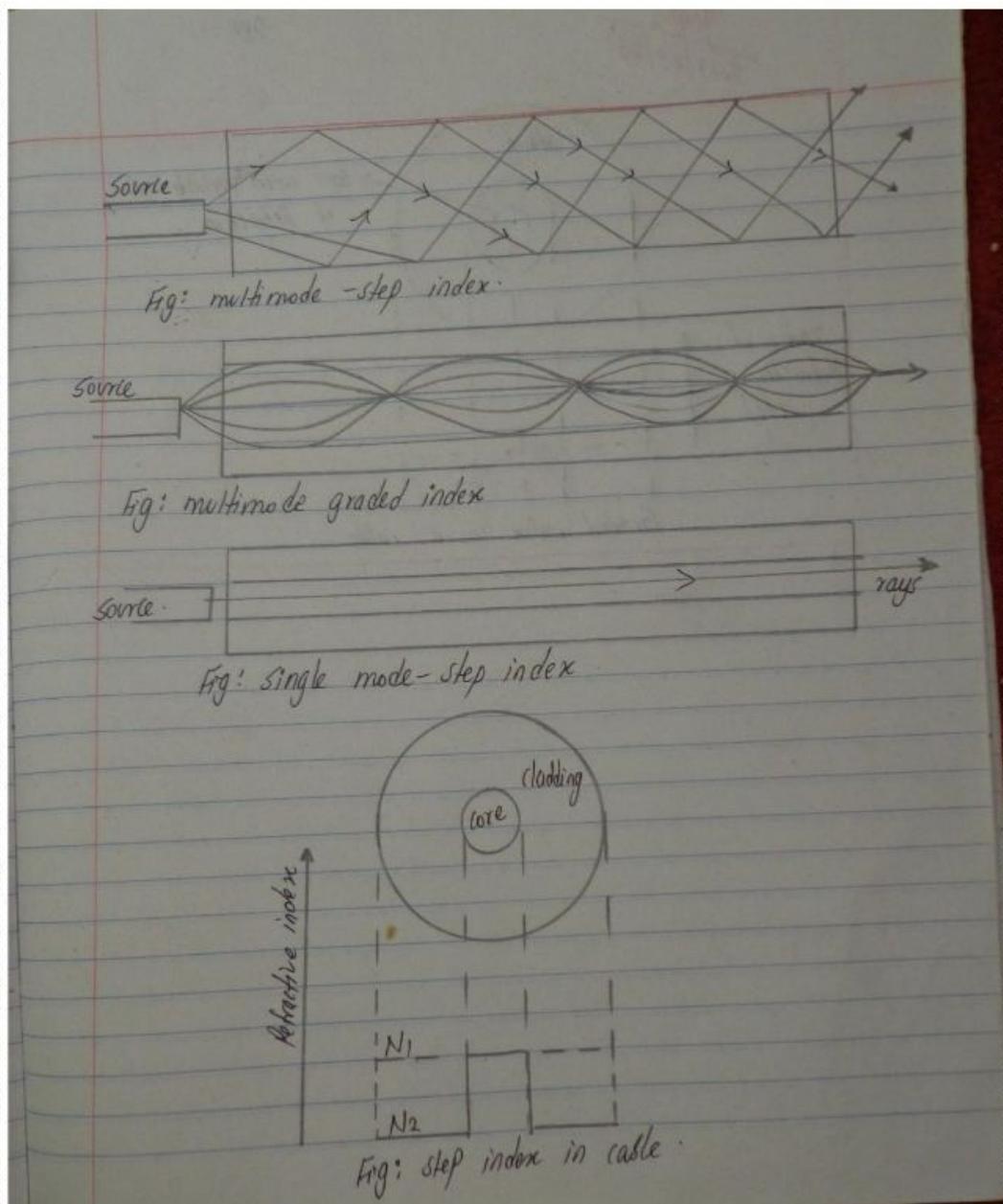
d_k	d_{k-1}	$o/p = d_k \oplus d_{k-1}$
0	0	0
0	1	1
1	0	1
1	1	0

J. single mode - step index, multi-mode step index and
multi-mode graded index optical fiber: 18/F, 2017-18.

\Rightarrow

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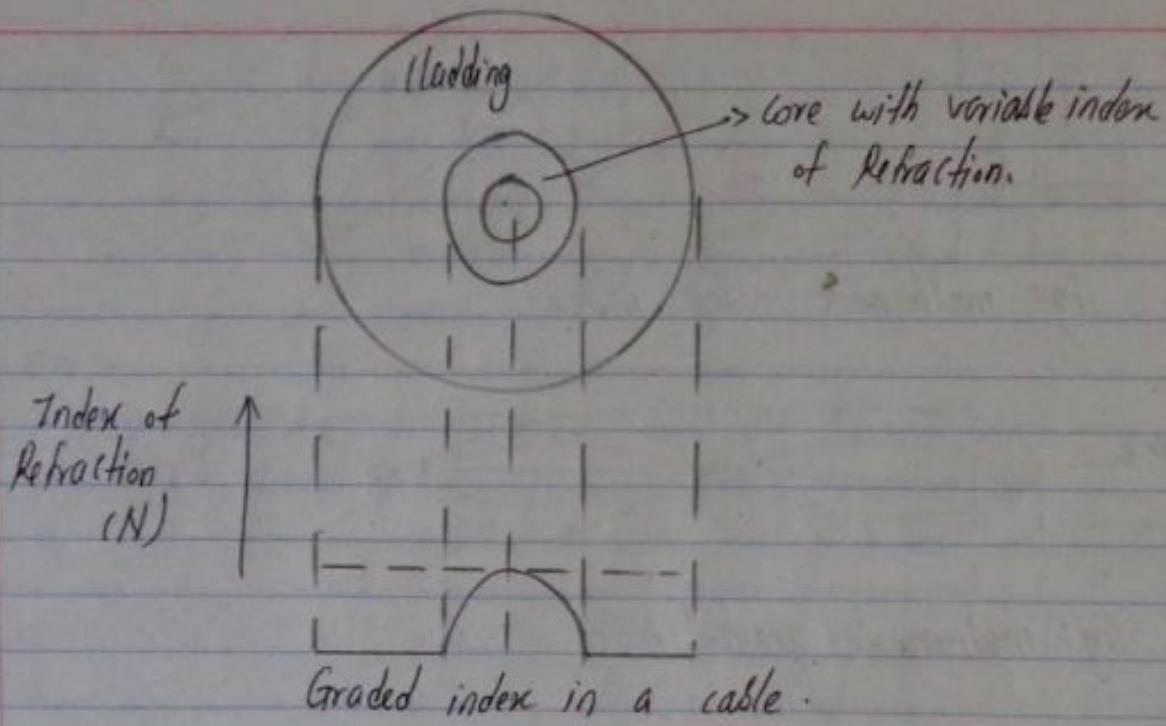
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K) Compare and contrast Analog and digital modulation schemes.

18/F

=>

Analog modulation

Digital modulation.

- i) Less bandwidth
- ii) more accurate

Large bandwidth.
less accurate due to quantization error that can't be avoided or corrected.

- iii) Low level of security

High level of security due to occurrence of encryption and authentication.

- iv) Less QoS

High QoS

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V) Low noise immunity vi) more difficult to design than digital.	High noise immunity easily designed using software.
Vii) Input is continuous signal.	Input is time sequence of symbols or pulses.

I) A given AM broadcast station transmits a total power of 5 kW when the carrier is modulated by a sinusoidal signal with a modulation index of 0.75. Compute,

- carrier power
- transmission efficiency
- the peak amplitude of the carrier assuming the antenna to be represented by a $(50 + j0)$ ohm load.

Soln,
 $\Rightarrow P_t = 5 \text{ kW}$
 $P_c = ?$
 $M = 0.75$

We know,

$$M = \sqrt{2 \times \left[\left(\frac{P_t}{P_c} \right) - 1 \right]}$$

or $M^2 = 2 \times \left[\left(\frac{P_t}{P_c} \right) - 1 \right]$

or $(0.75)^2 = 2 \times \left(\frac{5}{P_c} - 1 \right)$

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$$\text{Also, } P_t = P_c (1 + M^2/2)$$

$$\text{or } S = P_c (1 + 0.75^2/2)$$

$$\text{or } P_c = \frac{S}{1.281}$$

$$\therefore P_c = 3.9024 \text{ kW}$$

carrier power, $P_c = 3.91 \text{ kW}$

$$\text{i) Transmission efficiency, } n = \frac{P_c}{P_t} \times 100 \\ = \frac{3.91}{S} \times 100 \% \\ = 78 \%$$

$$\text{iii) } R = (50 + j0) - r = 50 - r$$

$$P_c = \frac{A_c^2}{2R}$$

$$\text{or } 3.91 \text{ kW} = \frac{A_c^2}{2 \times 50}$$

$$\text{or } A_c^2 = 3910 \times 2 \times 50 \\ = 21.53 \times 625.299 \text{ W}$$

$$\therefore A_c = 0.6260 \text{ kW}$$

M) Data Communication Hardware : 18/F

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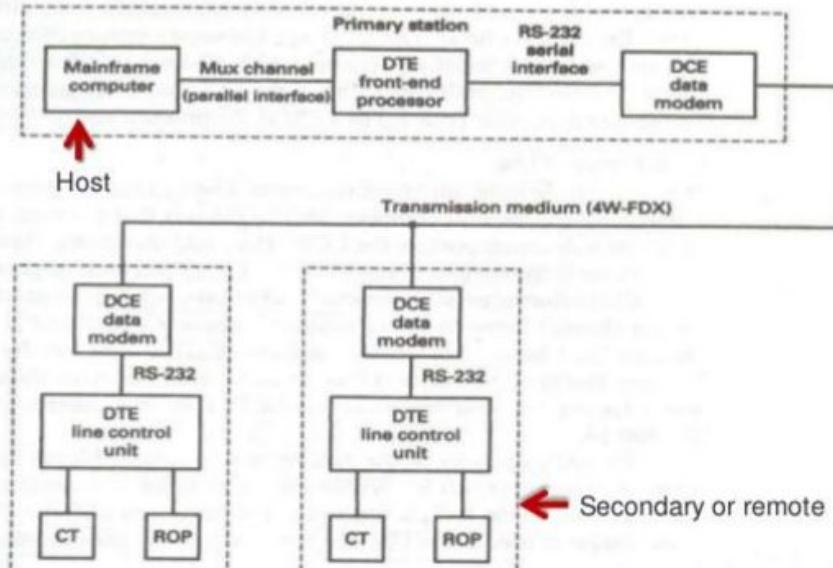
Data Communications Hardware

All endpoints in a data communication network have three fundamental components namely:

- **Data Terminal Equipment (DTE).** Any binary digital device that generates, transmits, receives, and/or interprets data messages. (i.e.: LCU, STACO, ACIA, UART, USRT)
- **Data Communication Equipment (DCE).** These are equipment that interfaces data terminal equipment to the transmission channel. (i.e.: MODEM)
- **Serial Interface.** A communication line that ensures the orderly flow of data between a DTE and DCE. (i.e.: USB, RS 232)

Data Communication Hardware

- A multipoint data communication circuit block diagram



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Analog Communication

Tutorial 2_Solutions

1. A 107.6 MHz carrier signal is frequency modulated by a 7 kHz sine wave. The resultant FM signal has a frequency deviation of 50 kHz. Determine the following:
 - a. the carrier swing of the FM signal,
 - b. the highest and the lowest frequencies attained by the modulated signal, and
 - c. the modulation index of the FM wave.

Solution:

Carrier frequency, $f_c = 107.6 \text{ MHz}$

Modulating frequency, $f_m = 7 \text{ kHz}$

Frequency deviation, $\Delta f = 50 \text{ kHz}$

- (a) Carrier swing = $2\Delta f = 100 \text{ kHz}$
- (b) Highest frequency = $f_c + \Delta f = 107.6 + 0.05 = 107.65 \text{ MHz}$
Lowest frequency = $f_c - \Delta f = 107.6 - 0.05 = 107.55 \text{ MHz}$
- (c) Modulation index, $\beta = \frac{\Delta f}{f_m} = \frac{50}{7} = 7.413$

2. Determine the frequency deviation and carrier swing for a frequency-modulated signal which has a carrier frequency of 105 MHz and whose upper frequency is 105.007 MHz when modulated by a particular wave. Find the lowest frequency reached by the FM wave.

Solution:

Carrier frequency, $f_c = 105 \text{ MHz}$

Upper frequency = 105.007 MHz

Frequency deviation, $\Delta f = 105.007 - 105 = 7 \text{ kHz}$

Carrier swing = $2\Delta f = 14 \text{ kHz}$

Lowest frequency = $f_c - \Delta f = 105 - 0.007 = 104.993 \text{ MHz}$

3. Determine the bandwidth of a narrow-band FM signal which is generated by a 4 kHz audio signal modulating a 105 MHz carrier.

Solution:

Frequency of audio signal, $f_m = 4 \text{ kHz}$

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Carrier frequency, $f_c = 105$ MHz

Bandwidth of narrow-band FM signal = $2f_m = 8$ kHz

4. The maximum deviation allowed in the FM broadcast system is 75 kHz. If the modulating signal is a single-tone sinusoid of 8 kHz, determine the bandwidth of FM signal. What will be the bandwidth when modulating signal amplitude is doubled?

Solution:

Frequency deviation, $\Delta f = 75$ kHz

Modulating frequency, $f_m = 8$ kHz

Using Carson's rule,

$$\text{bandwidth of FM signal, } B = 2(\Delta f + f_m) = 2(75 + 8) = 166 \text{ kHz}$$

Since $\Delta f = k_f A_m$, the new frequency deviation after doubling the modulating amplitude is $\Delta f' = 2\Delta f = 150$ kHz and the bandwidth becomes

$$B' = 2(\Delta f' + f_m) = 2(150 + 8) = 316 \text{ kHz}$$

5. In an FM system, the modulating frequency $f_m = 1$ kHz, the modulating amplitude $A_m = 2$ volt and the frequency deviation is 6 kHz. If the modulating voltage is raised to 4 volt then what will be the new deviation? If the modulating voltage is further increased to 8 volts and modulating frequency is reduced to 500 Hz, what will be the deviation?

Solution:

Modulating frequency, $f_m = 1$ kHz

Modulating amplitude, $A_m = 2$ V

Frequency deviation, $\Delta f = 6$ kHz

$$\text{Modulation index, } \beta = \frac{\Delta f}{f_m} = \frac{6}{1} = 6$$

Case I

Modulating amplitude, $A'_m = 4$ V

Since frequency deviation, $\Delta f = k_f A_m$, the new frequency deviation for the increased modulating voltage is

$$\Delta f' = 2\Delta f = 12 \text{ kHz}$$

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Case II

Modulating amplitude, $A_m'' = 8 \text{ V}$

Modulating frequency, $f_m'' = 500 \text{ Hz}$

Frequency deviation, $\Delta f'' = 2\Delta f' = 24 \text{ kHz}$

6. The modulating frequency and voltage in an FM system are respectively 400 Hz and 2.4 volts, and the modulation index is 60. Calculate the maximum deviation. What will be the modulation index when modulating frequency is reduced to 250 Hz and the modulating voltage is simultaneously raised to 3.2 volts.

Solution:

Modulating frequency, $f_m = 400 \text{ Hz}$

Modulating amplitude, $A_m = 2.4 \text{ V}$

Modulation index, $\beta = 60$

Maximum deviation, $\Delta f = \beta f_m = 60 * 400 = 24 \text{ kHz}$

Frequency sensitivity, $k_f = \frac{\Delta f}{A_m} = \frac{24}{2.4} = 10 \text{ kHz/V}$

New modulating frequency, $f_m' = 250 \text{ Hz}$

New modulating amplitude, $A_m' = 3.2 \text{ V}$

New frequency deviation, $\Delta f' = k_f A_m' = 10 * 3.2 = 32 \text{ kHz}$

New modulation index, $\beta' = \frac{\Delta f'}{f_m'} = \frac{32}{0.25} = 128$

7. What is the modulation index of an FM signal having a carrier swing of 100 kHz when the modulating signal has a frequency of 8 kHz?

Solution:

Carrier swing, $2\Delta f = 100 \text{ kHz}$

Frequency deviation, $\Delta f = \frac{100}{2} = 50 \text{ kHz}$

Modulating frequency, $f_m = 8 \text{ kHz}$

Modulation index, $\beta = \frac{\Delta f}{f_m} = \frac{50}{8} = 6.25$

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8. Given an angle-modulated signal $s(t) = 10 \cos(w_c t + 3 \sin(w_m t))$. Assume this to be PM and $f_m = 1$ kHz. Calculate the modulation index and find the bandwidth when (i) f_m is double and (ii) f_m is half.

Solution:

Angle-modulated wave, $s(t) = 10 \cos[w_c t + 3 \sin(w_m t)]$

Modulating frequency, $f_m = 1$ kHz

Comparing the given wave with the single-tone PM wave

$$s(t) = A_c \cos[w_c t + \beta_p \sin(w_m t)]$$

with message signal, $m(t) = A_m \sin(w_m t)$, we find the modulation index to be

$$\beta_p = 3.$$

(i) $f'_m = 2f_m = 2$ kHz

The change in modulating frequency does not change the modulation index in phase modulation. So, the modulation index,

$$\beta'_p = 3.$$

Frequency deviation, $\Delta f' = \beta'_p f'_m = 3 * 2 = 6$ kHz

Bandwidth, $B' = 2(\Delta f' + f_m') = 2(6 + 2) = 16$ kHz

(ii) $f''_m = \frac{f_m}{2} = 0.5$ kHz

Modulation index, $\beta''_p = 3$

Frequency deviation, $\Delta f'' = \beta''_p f''_m = 3 * 0.5 = 1.5$ kHz

Bandwidth, $B'' = 2(\Delta f'' + f_m'') = 2(1.5 + 0.5) = 4$ kHz

9. An Armstrong FM modulator is required in order to transmit an audio signal of bandwidth 50 Hz to 15 kHz. The narrowband phase modulator used for this purpose utilized a crystal controlled oscillator to provide a carrier frequency $f_{c1} = 0.2$ MHz. The output of the narrowband phase modulator is multiplied by n_1 by a multiplier and passed to a mixer with a local oscillator frequency $f_{c2} = 10.925$ MHz. The frequency deviation $\Delta f = 75$ kHz and carrier frequency $f_c = 100$ MHz is obtained by multiplying the mixer output with n_2 using another multiplier. Find n_1 and n_2 . Assume that NBFM produces deviation of 25 Hz for the lowest baseband signal.

Solution:

NBFM frequency deviation, $\Delta f_1 = 25$ Hz

NB phase modulator carrier frequency, $f_{c1} = 0.2$ MHz

Order of first multiplier = n_1

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LO frequency of frequency converter, $f_{c2} = 10.925$ MHz

Order of second multiplier = n_2

Required carrier frequency, $f_c = 100$ MHz

Required frequency deviation, $\Delta f = 75$ kHz

At output of first frequency multiplier,

Frequency deviation = $n_1 \Delta f_1$

Carrier frequency = $n_1 f_{c1}$

At output of mixer,

Frequency deviation = $n_1 \Delta f_1$

Carrier frequency = $f_{c2} - n_1 f_{c1}$

We can have relationships

$$n_1 n_2 \Delta f_1 = 75$$

$$\text{or, } n_1 n_2 = 75000 / 25 = 3000 \quad (1)$$

$$\text{and, } f_{c2} - n_1 f_{c1} = f_c / n_2$$

$$\text{or, } 10.925 - n_1 * 0.2 = 100 / n_2$$

$$\text{or, } 10.925 * n_2 - 0.2 * n_1 n_2 = 100 \quad (2)$$

Solving equations (1) and (2), we get

$$n_1 \approx 47 \text{ and } n_2 \approx 64$$

10. A carrier wave of frequency 1 GHz and amplitude 3 volts is frequency modulated by a sinusoidal modulating signal of frequency 500 Hz and of peak amplitude 1 volt. The frequency deviation is 1 kHz. The level of the modulating waveform is changed to 5 volts peak and the modulating frequency is changed to 2 kHz. Obtain the expression for the new frequency modulated waveform.

Solution:

Carrier signal: $f_c = 1$ GHz, $A_c = 3$ V

Message signal: $f_m = 500$ Hz, $A_m = 1$ V

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Frequency deviation, $\Delta f = 1 \text{ kHz}$

Frequency sensitivity, $k_f = \frac{\Delta f}{A_m} = 1 \text{ kHz/V}$

New message signal: $f'_m = 2 \text{ kHz}$, $A'_m = 5 \text{ V}$

New FM signal: $\Delta f' = k_f A'_m = 1 * 5 = 5 \text{ kHz}$, $\beta' = \frac{\Delta f'}{f'_m} = \frac{5}{2} = 2.5$

The new FM signal,

$$s(t) = 3 \cos[2\pi 10^9 t + 2.5 \sin(4000\pi t)]$$

11. In an FM system, the audio frequency is 1 kHz and audio voltage is 2 volts. The deviation is 4 kHz. If the AF voltage is now increased to 8 volts and its frequency is dropped to 500 Hz, find the modulation index in each case and the corresponding bandwidth using Carson's rule.

Solution:

Case I

Message signal: $f_m = 1 \text{ kHz}$, $A_m = 2 \text{ V}$

Frequency deviation, $\Delta f = 4 \text{ kHz}$

Frequency sensitivity, $k_f = \frac{\Delta f}{A_m} = \frac{4}{2} = 2 \text{ kHz/V}$

Modulation index, $\beta = \frac{\Delta f}{f_m} = \frac{4}{1} = 4$

Bandwidth, $B = 2(\Delta f + f_m) = 2(4 + 1) = 10 \text{ kHz}$

Case II

Message signal: $A'_m = 8 \text{ V}$, $f'_m = 500 \text{ Hz}$

Frequency deviation, $\Delta f' = k_f A'_m = 2 * 8 = 16$

Modulation index, $\beta' = \frac{\Delta f'}{f'_m} = \frac{16}{0.5} = 32$

Bandwidth, $B' = 2(\Delta f' + f'_m) = 2(16 + 0.5) = 33 \text{ kHz}$

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POKHARA UNIVERSITY

Level: Bachelor

Semester – Fall

Year : 2015

Programme: BE

Full Marks: 100

Course: Communication System

Time: 3 Hours

*Candidates are required to give their answers in their own words as far as practicable.
The figure in the margin indicates full marks.*

Attempt all the questions

- | | | |
|----|---|-----|
| 1 | a) What is DSB-SC modulation? Justify with necessary spectrums. “DSB-SC is wasteful of transmission bandwidth than SSB-SC”. But why SSB not used for broadcasting give reasons. | 8 |
| b) | The total power content of an AM signal is 900 watt. Determine the power being transmitted at the carrier frequency and at each of the sidebands when the percentage modulation is 75%. | 7 |
| 2 | a) An audio signal given by $15\sin 2\pi(2000t)$ amplitude modulates a sinusoidal carrier wave $60\sin 2\pi(100000t)$. Determine:
i. Modulation index and % modulation.
ii. Frequency of signal and carrier.
iii. Frequency of USB and LSB of modulated wave. | 7 |
| b) | Explain serial and parallel interface in data communication. | 8 |
| 3 | a) Encode the following data stream into; Return zero (RZ), non-return to zero (NRZ), AMI, and Manchester codes, Data Stream: 0011101. | 8 |
| b) | State Shannon’s channel capacity theorem. A system has bandwidth of 44 KHz and signal to noise ratio of 28 dB at input to the receiver. Calculate
i. Its information carrying capacity.
ii. The capacity of channel if its bandwidth is double while the transmitted signal power remains constant. | 7 |
| 4 | a) Draw a generic block diagram of a digital communication system and explain its elements. Also list the merits of digital over analog communication system. | 8 |
| b) | What do you understand by pulse modulation? Explain the various pulse modulation techniques. | 7 |
| 5 | a) Why is it necessary to develop and implement procedures for error control in communications circuits? Briefly explain error detection and error correction mechanism with examples. | 7 |
| b) | State and explain Kepler’s laws. List out the advantage and application of Geostationary Satellite. What is round trip delay? | 8 |
| 6 | a) How call is made in wireless communication. Explain with GSM system architecture block diagram. | 8 |
| b) | What is optical fiber communication? Explain properties and implementation of OPGW. | 7 |
| 7 | Write short notes on: (any two)
a) ISDN
b) Baseband and Passband Signals
c) Vestigial Sideband transmission | 5+5 |

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POKHARA UNIVERSITY

Level: Bachelor

Semester: Spring

Year : 2014

Programme: BE

Full Marks : 100

Course: Communication System

Pass Marks : 45

Time : 3hrs.

Candidates are required to give their answers in their own words as far as practicable.

The figures in the margin indicate full marks.

Attempt all the questions.

1. a) What is SSB-SC Modulation? Justify with necessary spectrums 2+3+3
"DSB-SC is wasteful of transmission bandwidth than SSB-SC". But why SSB not used for broadcasting give reasons.
1. b) The total power content of an AM signal is 1000 watt. Determine the power being transmitted at the carrier frequency and at each of the sidebands when the percentage modulation is 100%. 7
2. a) What is modulation and demodulation? Why modulation needed and what will happen if signal transmitted without modulation? 3+5
2. b) Briefly explain the concept of exchangeability between bandwidth, signal power and SNR, then finally state Shannon's equation for channel capacity. $C = B \log_2 (1 + S/N)$ 7
3. a) Encode the following data stream into; Return zero (RZ), non-return to zero (NRZ), AMI, and Manchester codes. Data Stream: 11000010. 8
3. b) Compare BASK, BFSK and BPSK for the variables Bandwidth, Performance and bit rate. Determine minimum Bandwidth for a BPSK modulator with carrier frequency of 50 MHz and an input bit rate of 500kbps. 3+4
4. a) Why is it necessary to develop and implement procedures for error control in communications circuit? Briefly explain error detection and error correction mechanism with examples. 1+6
4. b) Explain ISDN system. What benefits does it offer to user, network provider and Manufacturer? 5+3
5. a) What is the basic concept of cellular mobile radio and frequency 4+4

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- reuse? What is the primary purpose of cell splitting and sectoring, explain?
- b) Why are Up-link frequencies greater than down-link frequencies in satellite communication? What is Roundtrip delay of geosynchronous satellite? 5+2
6. a) What is index profile of optical fiber? Briefly explain the single mode step-index and Multi-mode Step-index configuration of optical fiber. 2+6
b) What is OPGW system? How it is constructed and what are its applications? 7
7. Write short notes on: (Any two) 2×5
a) Baseband and Passband Signals
b) VSB
c) Data Modems

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PO KHARA UNIVERSITY

Level: Bachelor

Semester: Spring

Year : 2015

Programme: BE

Full Marks: 100

Course: Communication System Engineering

Pass Marks: 45

Time : 3 hrs.

Candidates are required to give their answers in their own words as far as practicable.

The figures in the margin indicate full marks.

Attempt all the questions.

1.
 - a) Why modulation needed? Explain in brief the different blocks of a typical communication system with neat diagram. 8
 - b) Explain the importance of channel bandwidth with the help of Hartley-Shannon law. A system has a bandwidth of 4 KHz and signal to noise ratio of 28 dB at input to the receiver. Calculate;
 - i. Its information carrying capacity.
 - ii. The capacity of channel if its band width is double while the transmitted signal power remains constant.7
2.
 - a) Explain DSB-SC AM generation by balanced modulator method. 8

Prove that $S_{ssb}(t) = \frac{A_c}{2} [m(t) \cos 2\pi f_c t \mp \hat{m}(t) \sin 2\pi f_c t]$ where
 $S_{ssb}(t)$ is SSB modulated signal and $m(t) = A_m \cos 2\pi f_m t$,
 $c(t) = A_c \cos 2\pi f_c t$

 - b) Explain DPSK system. Encode binary sequence 00110100010 into differentially encoded sequence and draw DPSK waveform. 7
3.
 - a) An audio signal given by $15 \sin 2\pi (2000 t)$ amplitude modulates a sinusoidal carrier wave $60 \sin 2\pi (100000 t)$. Determine
 - i. Modulation index
 - ii. % modulation
 - iii. Frequency of signal and carrier.
 - iv. Frequency of USB and LSB of modulated wave.7
 - b) What are the bandwidth and the power of AM Wave? Derive the expression for efficiency of AM signal. 8
4.
 - a) Describe the ASK, FSK and PSK technique with neat block diagram and wave forms. 8

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|--|--------------|
| 1. b) Why is it necessary to develop and implement procedures for error control in communication circuit? Briefly explain error detection and correction techniques with examples. | 7 |
| 5. a) What do you mean by packet switching? Explain the packet data transmission with an example of Ethernet Packet Format. | 7 |
| b) Why GEO satellites are called geosynchronous? Briefly explain Transponder, Footprint and Round trip-delay of GSS. | 8 |
| 6. a) What is frequency reuse? Explain the two most predominant forms of interferences in cellular telephone systems. | 8 |
| b) What is optical fibre ground wire system? Explain Numerical aperture and acceptance angle of optical fiber. | 7 |
| 7. Write short notes on: (Any two) | 2×5 |
| a) Carson's Rule | |
| b) Kepler's Laws. | |
| c) Attenuation, dispersion and bend loss in optical fiber. | |

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POKHARA UNIVERSITY

Level: Bachelor
Programme: BE
Course: Communication System Engineering

Semester: Fall

Year : 2016
Full Marks: 100
Pass Marks: 45
Time : 3hrs.

Candidates are required to give their answers in their own words as far as practicable.

The figures in the margin indicate full marks.

Attempt all the questions.

1. a) Define Amplitude Modulation. Derive the expression for AM in frequency domain and show that transmission bandwidth for an AM wave is exactly twice the message bandwidth. 8
- b) A certain AM transmitter radiates 9 KW with the carrier unmodulated and 10.125 KW when the carrier is simultaneously modulated.
 - i. Calculate the modulation index.
 - ii. If another sine wave corresponding to 60% modulation is transmitted simultaneously, determine the total radiated power.7
2. a) A carrier wave of frequency 91 MHz is frequency modulated by a sine wave of amplitude 10 V and 15 KHz. The frequency sensitivity of the modulator is 3 KHz/V.
 - i. Determine the approximate bandwidth of FM wave using Carson's Rule.
 - ii. Repeat part i., assuming that the amplitude of the modulation wave is doubled and frequency of the modulating wave is halved.8
- b)* Briefly differentiate coherent and non-coherent digital modulation techniques. Explain DPSK system. 7
3. a) Encode the data stream 110000000101 into; Return to zero (RZ), non-return to zero (NRZ), AMI, and B8ZS. Data Stream: 0011101. 8
- b) State Shannon's channel capacity theorem. Given an AWGN channel with 4 KHz bandwidth and the noise power spectrum density of 10^{-12} W/Hz. The signal power required at the receiver is 0.1 mW. Calculate the capacity of the channel. 7

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|----|--|--------------|
| 4. | a) Why is it necessary to develop and implement procedures for error control in communications circuits? Briefly explain error detection and error correction mechanism with examples. | 8 |
| b) | What do you mean by Protocol in communication? What are the basic requirements of Protocols? | 7 |
| 5. | a) Explain ISDN system. What benefits does it offer to user, network provider and manufacturer? | 7 |
| b) | What are the different satellite orbits that can be used? Explain each orbits in brief. | 8 |
| 6. | a) Differentiate conventional and cellular mobile radio system. Briefly explain GSM network architecture. | 8 |
| b) | Consider a multimode fiber that has a core refractive index of 1.48 and a core cladding index difference of 3 percent ($\Delta=0.03$). Find:
i. The Numerical Aperture
ii. The acceptance angle
iii. The critical angle | 7 |
| 7. | Write short notes on: (Any two) | 2×5 |
| a) | Block Diagram of Communication System | |
| b) | Kepler's Laws | |
| c) | OPGW | |

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