Discrete Time Fourier Transformation (DTFT)

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1. DFT and DTFT

- · DTFT is the Fourier transform of choice for analyzing infinite-length signals and systems
- Useful for conceptual, but not Matlab friendly (infinitely-long vectors)
- ullet We will derive DTFT as the limit of the DFT as the signal length $N o\infty$

$$\omega=rac{2\pi}{N}k$$

The Centered DFT

■ Both x[n] and X[k] can be interpreted as periodic with period N, so we will shift the intervals of interest in time and frequency to be centered around n, k = 0

$$-\frac{N}{2} \le n, k \le \frac{N}{2} - 1$$

■ The modified forward and inverse DFT formulas are

$$X_u[k] = \sum_{n=-N/2}^{N/2-1} x[n] e^{-j\frac{2\pi}{N}kn}, \quad -\frac{N}{2} \le k \le \frac{N}{2} - 1$$

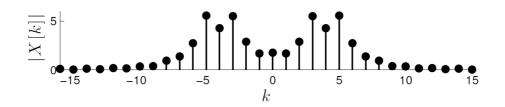
$$x[n] = \frac{1}{N} \sum_{k=-N/2}^{N/2-1} X_u[k] e^{j\frac{2\pi}{N}kn} - \frac{N}{2} \le n \le \frac{N}{2} - 1$$

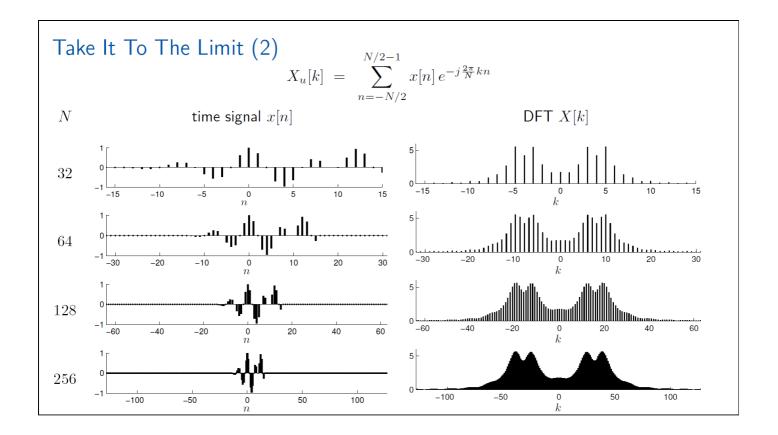
Take It To The Limit (1)

$$X_u[k] = \sum_{n=-N/2}^{N/2-1} x[n] e^{-j\frac{2\pi}{N}kn}, \quad -\frac{N}{2} \le k \le \frac{N}{2} - 1$$

- lacksquare Let the signal length N increase towards ∞ and study what happens to $X_u[k]$
- lacktriangle Key fact: No matter how large N grows, the frequencies of the DFT sinusoids remain in the interval

$$-\pi \le \omega_k = \frac{2\pi}{N}k < \pi$$





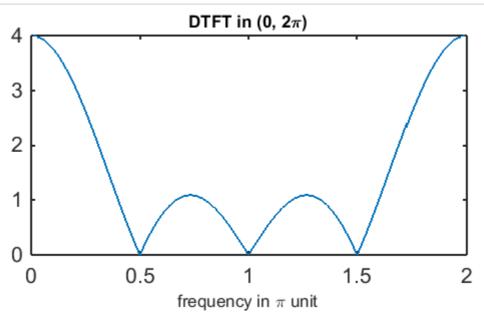
```
In [11]: %plot -s 560,300

% dtft from definition
n = 0:3;
x = [1 1 1 1];

N = 200;

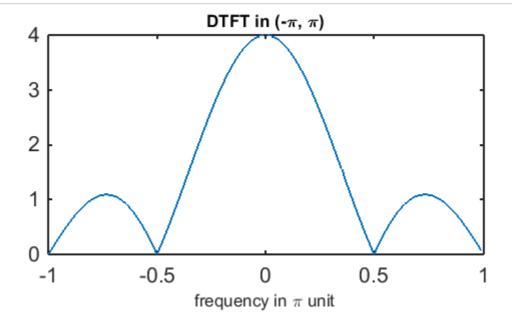
w2 = [0:N]*2*pi/N; w = w2(1:end-1);
Xdtft = sin(2*w)./sin(w/2).*exp(-1j*3*w/2);

plot(w/pi,abs(Xdtft))
xlabel('frequency in \pi unit','fontsize',8),
title('DTFT in (0, 2\pi)','fontsize',8)
```



```
In [8]: k = [0:N/2-1 -N/2:-1];
w = k*2*pi/N;
ws = fftshift(w);
Xdtfts = fftshift(Xdtft);

plot(ws/pi,abs(Xdtfts))
xlabel('frequency in \pi unit','fontsize',8),
title('DTFT in (-\pi, \pi)','fontsize',8)
```



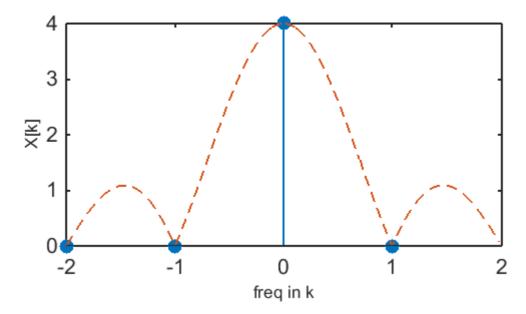
Out[8]:

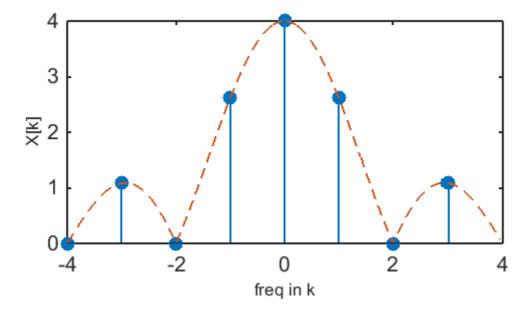
```
In [14]: x = [1,1,1,1];
N = length(x);

k = [0:N/2-1 -N/2:-1];
ks = fftshift(k);

X = dft(x,N);
Xs = fftshift(X);

stem(ks,abs(Xs),'filled'),
xlabel('freq in k','fontsize',8), ylabel('X[k]','fontsize',8),
xlim([-N/2,N/2]), hold on
plot(ws*N/(2*pi),abs(Xdtfts),'--'), hold off
```





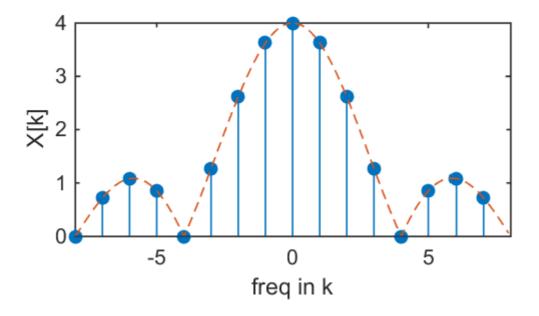
Out[15]:

```
In [5]: x = [1,1,1,1,zeros(1,12)];
N = length(x);

k = [0:N/2-1 -N/2:-1];
ks = fftshift(k);

X = dft(x,N);
Xs = fftshift(X);

stem(ks,abs(Xs),'filled'),
xlabel('freq in k'), ylabel('X[k]'), xlim([-N/2,N/2]), hold on
plot(ws*N/(2*pi),abs(Xdffts),'--'), hold off
```



Out[5]:

```
In [2]: % DTFT using the output of FFT (or DFT)

x = [1,1,1,1,zeros(1,2^6-4)];
N = length(x);

k = [0:N/2-1 -N/2:-1];
ks = fftshift(k);

X = dft(x,N);
Xs = fftshift(X);

stem(ks,abs(Xs),'filled','markersize',3)
xlabel('freq in k'), ylabel('X[k]'), xlim([-N/2,N/2]), hold on plot(ws*N/(2*pi),abs(Xdtfts),'--'), hold off
```

Out[2]: Undefined function or variable 'ws'.

Discrete Time Fourier Transform (Forward)

■ As $N \to \infty$, the forward DFT converges to a function of the **continuous frequency variable** ω that we will call the **forward discrete time Fourier transform** (DTFT)

$$\sum_{n=-N/2}^{N/2-1} x[n] e^{-j\frac{2\pi}{N}kn} \longrightarrow \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} = X(\omega), \qquad -\pi \le \omega < \pi$$

■ Recall: Inner product for infinite-length signals

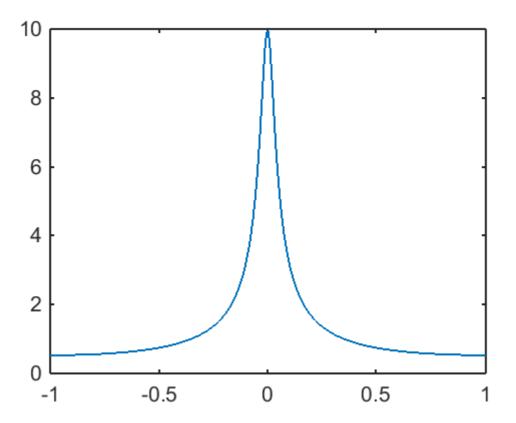
$$\langle x, y \rangle = \sum_{n=-\infty}^{\infty} x[n] y[n]^*$$

■ Analysis interpretation: The value of the DTFT $X(\omega)$ at frequency ω measures the similarity of the infinite-length signal x[n] to the infinite-length sinusoid $e^{j\omega n}$

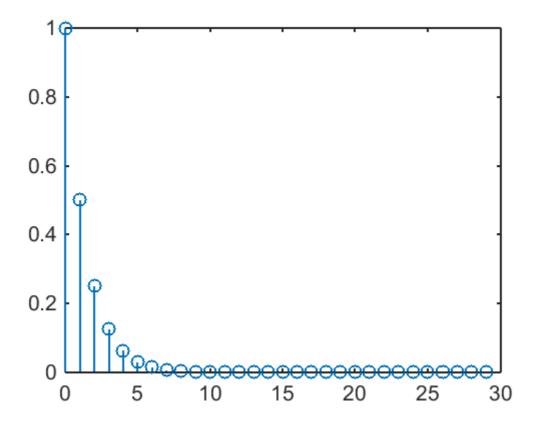
1.1. Analytic form of DTFT (exact transformation)

$$x[n] = {(0.5)}^n u[n] \qquad \leftrightarrow \qquad X(e^{j\omega}) = rac{e^{j\omega}}{e^{j\omega} - 0.5}$$

```
In [4]: w = linspace(-pi,pi,2^8);
X = exp(1j*w)./(exp(1j*w) - 0.9);
plot(w/pi,abs(X))
```

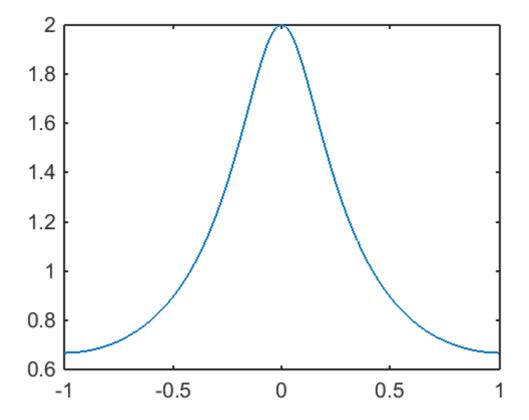


Out[4]:



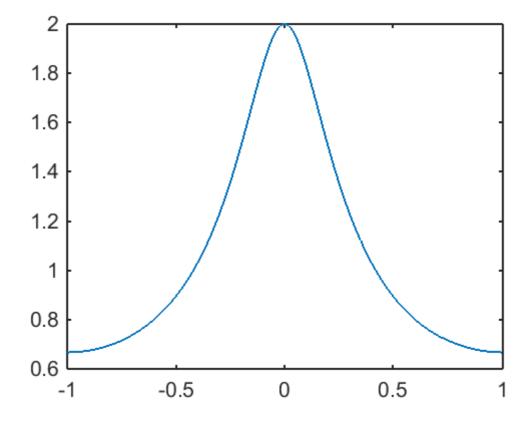
Out[6]:

```
In [9]: w = linspace(-1,1,2^8)*pi;
X = exp(-1j*(w'*n))*x';
plot(w/pi,abs(X))
```



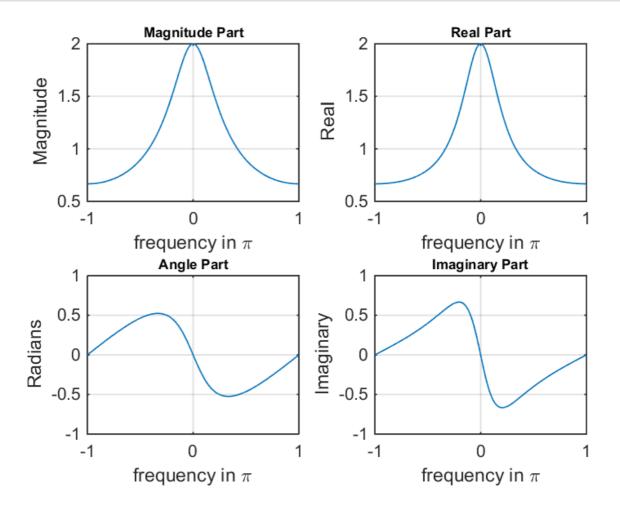
Out[9]:

In [11]: X = dtft(x,n,w);
plot(w/pi,abs(X))



Out[11]:

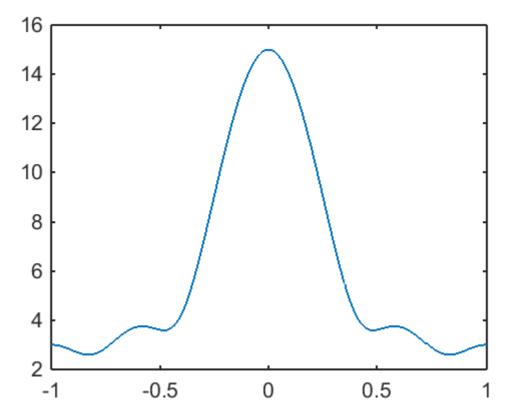
```
In [7]: %plot -s 800,600
        N = 500;
        w2 = [-1:1/N:1]*pi;
        w = w2(1:end-1);
        % closed form of DTFT
        X = \exp(1j*w)./(\exp(1j*w) - 0.5*ones(size(w)));
        magX = abs(X); angX = angle(X);
        realX = real(X);
                             imagX = imag(X);
        subplot(2,2,1); plot(w/pi,magX); grid
        xlabel('frequency in \pi'); title('Magnitude Part', 'fontsize',8); ylabel('Magnitude')
        subplot(2,2,3); plot(w/pi,angX); grid
        xlabel('frequency in \pi'); title('Angle Part', 'fontsize',8); ylabel('Radians')
        subplot(2,2,2); plot(w/pi,realX); grid
        xlabel('frequency in \pi'); title('Real Part', 'fontsize',8); ylabel('Real')
        subplot(2,2,4); plot(w/pi,imagX); grid
        xlabel('frequency in \pi'); title('Imaginary Part', 'fontsize',8); ylabel('Imaginary')
```



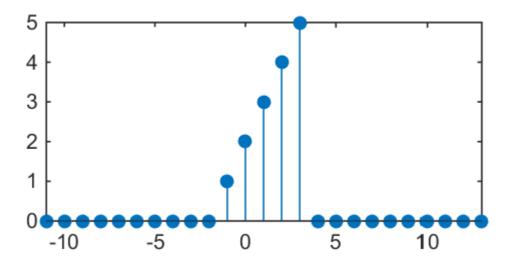
Out[7]:

1.2. DTFT of a numerical computation (using a definition)

```
In [13]: x = [0,1,2,3,4,5];
n = -2:3;
X = dtft(x,n,w);
plot(w/pi,abs(X))
```

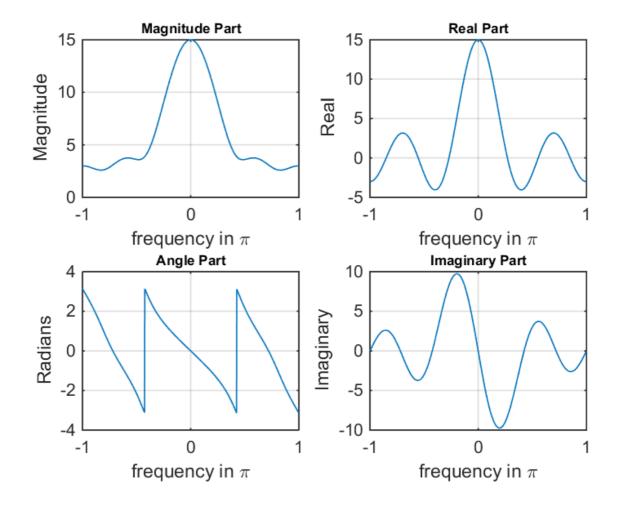


Out[13]:



Out[8]:

```
In [9]: %plot -s 800,600
                                      %% DTFT code
                                      \frac{1}{2} \frac{1}
                                      N = 500;
                                      k = -N:N-1; w = (pi/N)*k;
                                      X = x * (exp(-1j*pi/500)).^{(n'*k)};
                                                                                                                                                                                                                       % DTFT using matrix-vector product
                                      % plots
                                      magX = abs(X); angX = angle(X);
                                      realX = real(X);
                                                                                                                              imagX = imag(X);
                                      subplot(2,2,1); plot(w/pi,magX); grid
                                      xlabel('frequency in \pi'); title('Magnitude Part', 'fontsize',8); ylabel('Magnitude')
                                      subplot(2,2,3); plot(w/pi,angX); grid
                                      xlabel('frequency in \pi'); title('Angle Part', 'fontsize',8); ylabel('Radians')
                                      subplot(2,2,2); plot(w/pi,realX); grid
                                      xlabel('frequency in \pi'); title('Real Part', 'fontsize',8); ylabel('Real')
                                      subplot(2,2,4); plot(w/pi,imagX); grid
                                      xlabel('frequency in \pi'); title('Imaginary Part', 'fontsize',8); ylabel('Imaginary')
```

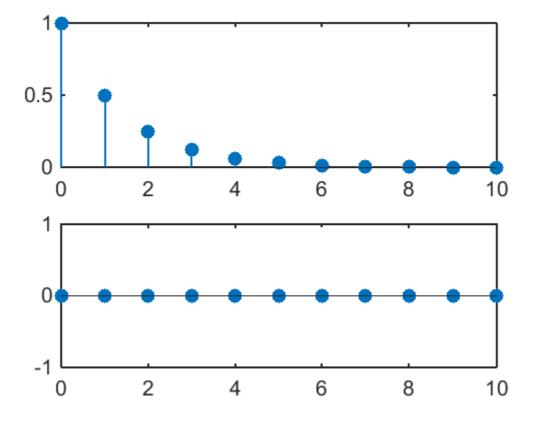


Out[9]:

Example 2

$$x[n] = \left(0.5e^{(j\pi/3)}
ight)^n$$

```
In [20]: %plot -s 560,420
n = 0:10;
x = (0.5).^n;
subplot(2,1,1), stem(n,real(x),'filled')
subplot(2,1,2), stem(n,imag(x),'filled')
```



Out[20]:

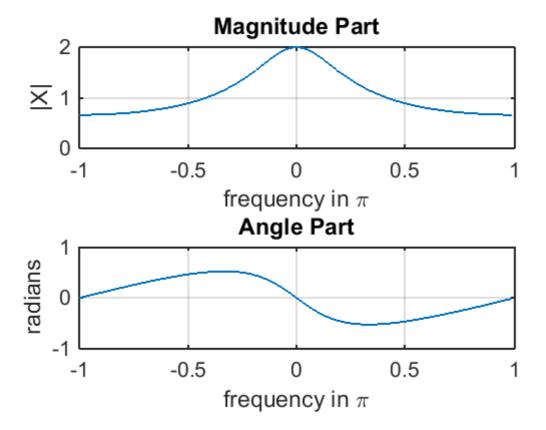
```
In [21]: N = 100;
k = -N:N-1; w = (pi/N)*k;

X = x*(exp(-1j*pi/N)).^(n'*k); % DTFT

magX = abs(X); angX = angle(X);

subplot(2,1,1); plot(w/pi,magX);grid
xlabel('frequency in \pi'); ylabel('|X|')
title('Magnitude Part')

subplot(2,1,2); plot(w/pi,angX);grid
xlabel('frequency in \pi'); ylabel('radians')
title('Angle Part')
```



Out[21]:

2. Numerical DTFT Computation

```
function X = dtft(x,n,w)
% X = dtft(x, n, w)
%
% X = DTFT values computed at w frequencies
% x = finite duration sequence over n
% n = sample position vector
% w = frequency location vector
X = exp(-1j*(w'*n))*x';
```

end

DTFT of the unit pulse

$$p[n] = egin{cases} 1 & -M \leq n \leq M \ 0 & ext{otherwise} \end{cases} \ P(\omega) = \sum_{n=-\infty}^{\infty} p[n] \, e^{-j\omega n} = \sum_{n=-M}^{M} e^{-j\omega n} = rac{e^{j\omega M} - e^{-j\omega(M+1)}}{1 - e^{-j\omega}} = rac{e^{-j\omega/2} \left(e^{j\omegarac{2M+1}{2}} - e^{-j\omega/2}
ight)}{e^{-j\omega/2} \left(e^{j\omega/2} - e^{-j\omega/2}
ight)} = rac{2j \, \sin\left(\omegarac{2M+1}{2}
ight)}{2j \, \sin\left(rac{\omega}{2}
ight)} \end{cases}$$

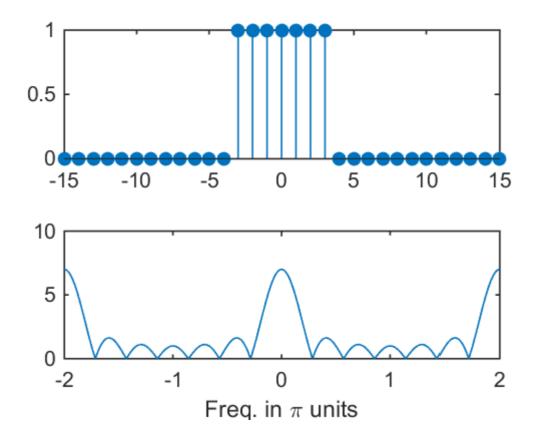
```
In [12]: %plot -s 560,420

x = [1 1 1 1 1 1 1];
n = -3:3;
w = linspace(-2,2,2^10)*pi;

X = dtft(x,n,w);

xd = zeros(1,31);
nd = -15:15;
[y,ny] = sigadd(xd,nd,x,n);

subplot(2,1,1), stem(ny,y,'filled')
subplot(2,1,2), plot(w/pi,abs(X)), xlabel('Freq. in \pi units')
```



Out[12]:

DTFT of triangle

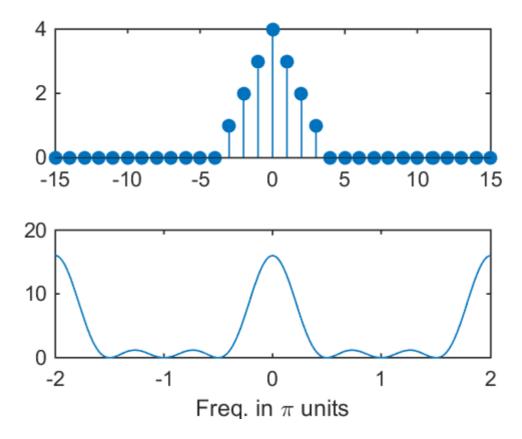
```
In [13]: %plot -s 560,420

x = [1 2 3 4 3 2 1];
n = -3:3;
w = linspace(-2,2,2^10)*pi;

X = dtft(x,n,w);

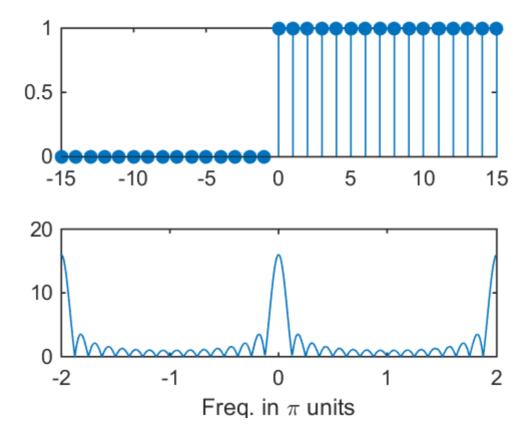
xd = zeros(1,31);
nd = -15:15;
[y,ny] = sigadd(xd,nd,x,n);

subplot(2,1,1), stem(ny,y,'filled')
subplot(2,1,2), plot(w/pi,abs(X)), xlabel('Freq. in \pi units')
```



Out[13]:

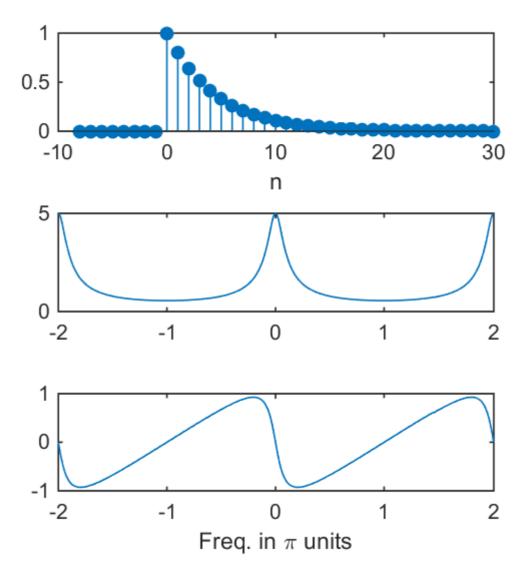
DTFT of pulse



Out[14]:

DTFT of a one-sided exponential

$$h[n] = lpha^n u[n] \qquad \longleftrightarrow \qquad H(\omega) = rac{1}{1 - lpha e^{-j\omega}}$$

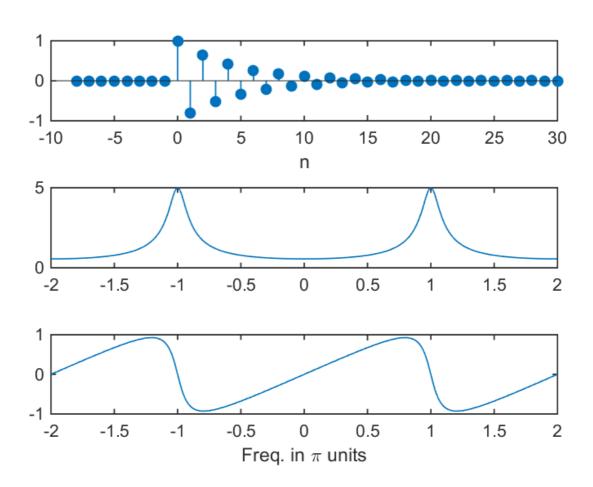


3. Property:

DTFT and Modulation

```
e^{-j\omega_0 n}x[n] \longleftrightarrow X(\omega-\omega_0) \ e^{-j\omega_0 n}x[n] = (-1)^nx[n] \qquad 	ext{when} \quad \omega_0 = rac{2\pi}{N}rac{N}{2} = \pi
```

```
In [16]:
         %plot -s 800,600
         N = 30;
         nd = -8:N;
         xd = zeros(size(nd));
         x = zeros(1,N);
         for i = 1:N
             x(i) = (-0.8)^{(i-1)};
         end
         n = 0:N-1;
         %w = linspace(-1,1,2^10)*pi;
         w = linspace(-2,2,2^10)*pi;
         X = dtft(x,n,w);
         subplot(3,1,1), stem(nd,sigadd(xd,nd,x,n),'filled'), xlabel('n')
         subplot(3,1,2), plot(w/pi,abs(X)),
         subplot(3,1,3), plot(w/pi,phase(X)), xlabel('Freq. in \pi units')
```

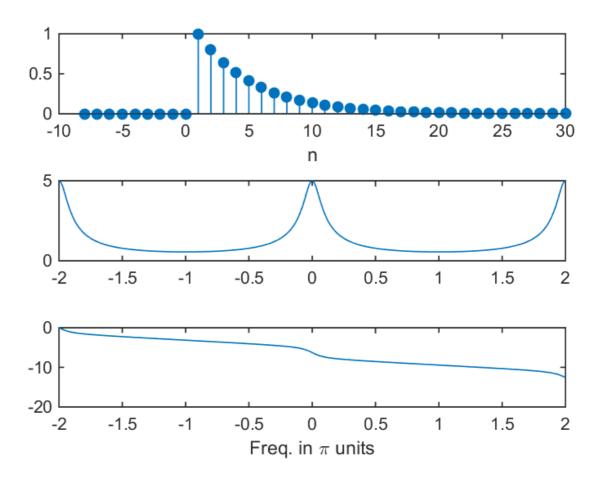


DTFT and Time Shift

$$x[n-m] \qquad \longleftrightarrow \qquad e^{-j\omega m}X(\omega)$$

- · same amplitude
- phase changed (linearly $-\angle \omega m$)

```
In [17]:
         %plot -s 800,600
         N = 30;
         nd = -8:N;
         xd = zeros(size(nd));
         x = zeros(1,N);
         for i = 1:N
             x(i) = 0.8^{(i-1)};
         end
         m = 1;
                          % m = 1 => one sample delay or shift
         n = 0+m:N-1+m;
         [y,ny] = sigadd(xd,nd,x,n);
         % w = linspace(-1,1,2^10)*pi;
         w = linspace(-2,2,2^10)*pi;
         X = dtft(x,n,w);
         subplot(3,1,1), stem(ny,y,'filled'), xlabel('n')
         subplot(3,1,2), plot(w/pi,abs(X)),
         subplot(3,1,3), plot(w/pi,phase(X)), xlabel('Freq. in \pi units')
```

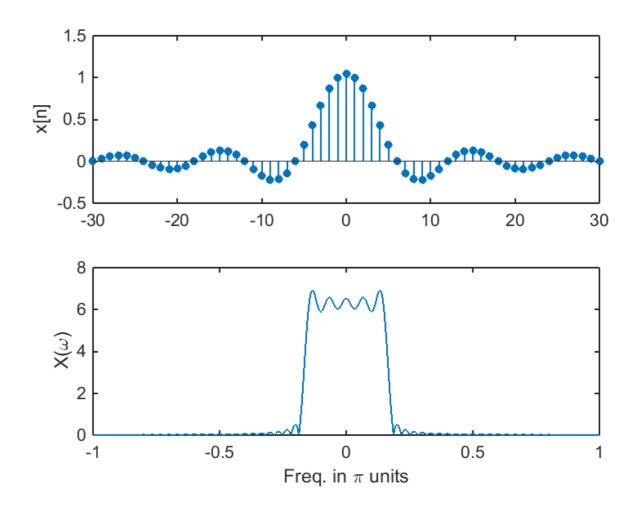


4. Filters

Ideal lowpass filter and discrete-time sinc function

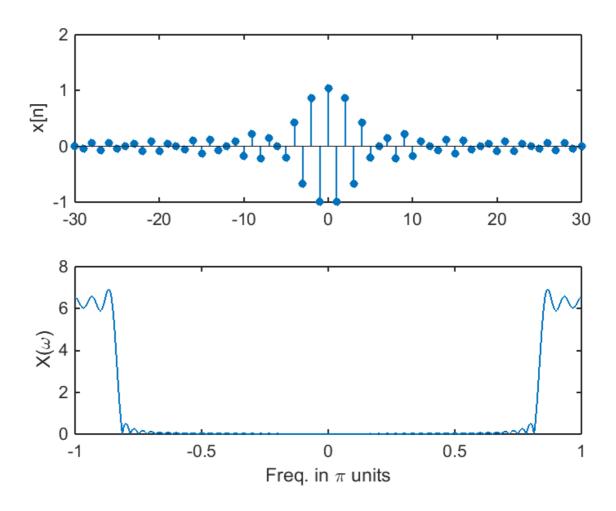
Impulse Response of the Ideal Lowpass Filter

onse of the ideal Lowpass Filter
$$h[n] = 2\omega_c rac{sin(\omega_c n)}{\omega_c n} \qquad \longleftrightarrow \qquad H(\omega) = egin{cases} 1 & -\omega_c \leq \omega \leq \omega_c \ 0 & ext{otherwise} \end{cases}$$
 $h[n] = \int_{-\pi}^{\pi} H(\omega) \, e^{j\omega n} rac{d\omega}{2\pi} = \int_{-\omega_c}^{\omega_c} e^{j\omega n} rac{d\omega}{2\pi} = rac{e^{j\omega_c n} - e^{-j\omega_c n}}{jn} = 2\omega_c rac{\sin(\omega_c n)}{\omega_c n} = \sin(x) = \frac{\sin(\pi x)}{\pi x}$



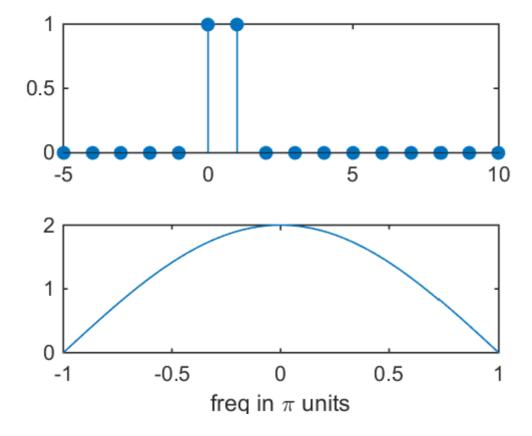
Out[1]:

```
In [2]: %plot -s 800,600
        wc = pi/6;
        N = 30;
        n = -N:N;
        d = zeros(1,length(n));
        h = zeros(1,length(n));
        d(N+1) = d(N+1) + 1;
        for i = 1:length(n)
            h(i) = (-1)^{(i-N-1)*2*wc*sinc(1/pi*wc*(i-N-1))};
        end
        %h = sigadd(d,n,-h,n);
        w = linspace(-1,1,2^12)*pi;
        %w = linspace(0,2,2^10)*pi;
        X = dtft(h,n,w);
        subplot(2,1,1), stem(n,h,'filled','markersize',4), ylabel('x[n]')
        subplot(2,1,2), plot(w/pi,abs(X)), ylabel('X(\omega)')
        xlabel('Freq. in \pi units'), ylabel('X(\omega)')
```



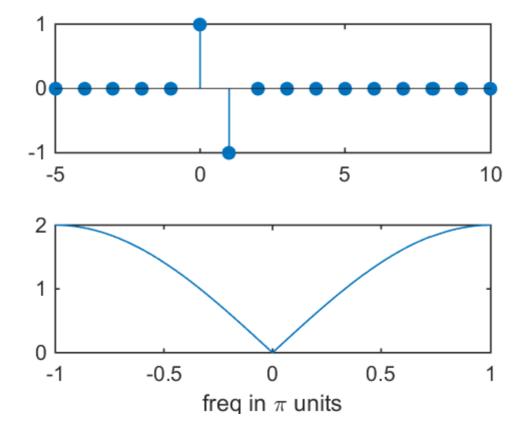
Out[2]:

Linear Filters: Low-Pass



Out[20]:

Linear Filters: High-Pass



Out[21]:

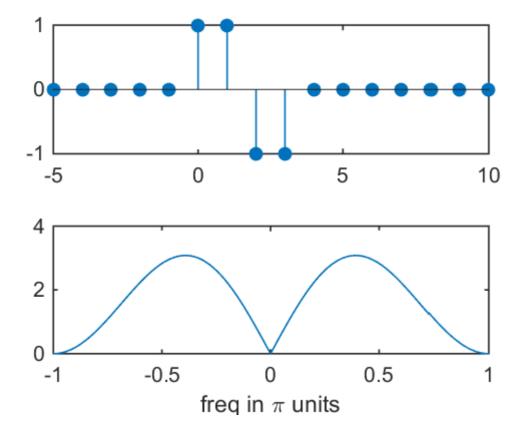
Linear Filters: Band-Pass

```
In [22]: %plot -s 560,420
    nd = -5:10;
    hd = zeros(1,length(nd));

N = 4;
    h = [1 1 -1 -1];
    n = 0:N-1;

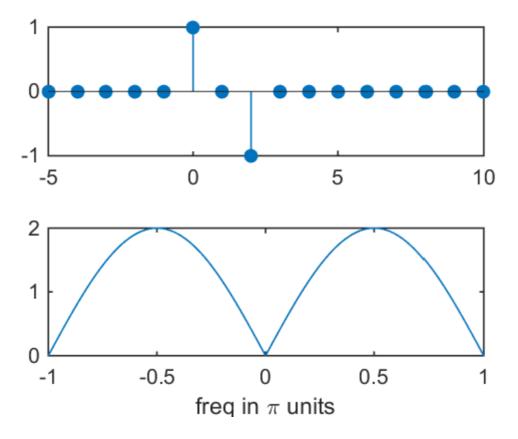
w = linspace(-1,1,2^10)*pi;
X = dtft(h,n,w);

[y,ny] = sigadd(hd,nd,h,n);
subplot(2,1,1), stem(ny,y,'filled')
subplot(2,1,2), plot(w/pi,abs(X)), xlabel('freq in \pi units')
```



Out[22]:

Linear Filters: Band-Stop



Out[23]:

5. High-density spectrum and high-resolution spectrum

$$x[n] = \cos(0.48\pi n) + \cos(0.52\pi n)$$

```
In [24]: N = 100;
n = 0:N-1;
x = cos(0.48*pi*n) + cos(0.52*pi*n);
```

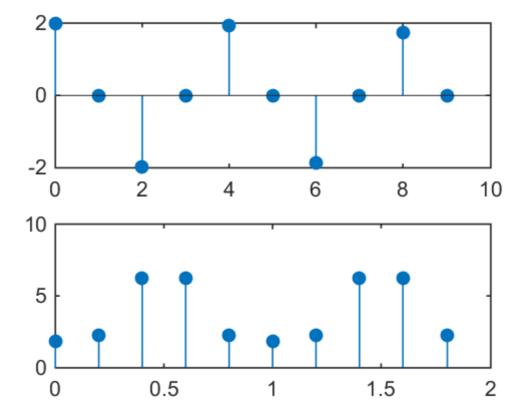
Out[24]:

use only 10-point DFT of x[n]

```
In [25]: %plot -s 560,420

n1 = 0:9;
y1 = x(1:10);
Y1 = dft(y1,10);
k1 = n1;
w1 = 2*pi/10*k1;

subplot(2,1,1), stem(n1,y1,'filled')
subplot(2,1,2), stem(w1/pi,abs(Y1),'filled')
```



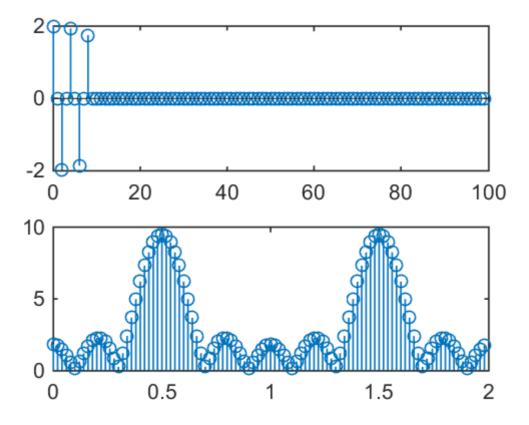
Out[25]:

pad 90 zeros to obtain a dense spectrum

```
In [26]: %plot -s 560,420

n2 = 0:99;
y2 = [x(1:10), zeros(1,90)];
Y2 = dft(y2,N);
k2 = n2;
w2 = 2*pi/100*k2;

subplot(2,1,1), stem(n2,y2)
subplot(2,1,2), stem(w2/pi,abs(Y2))
```



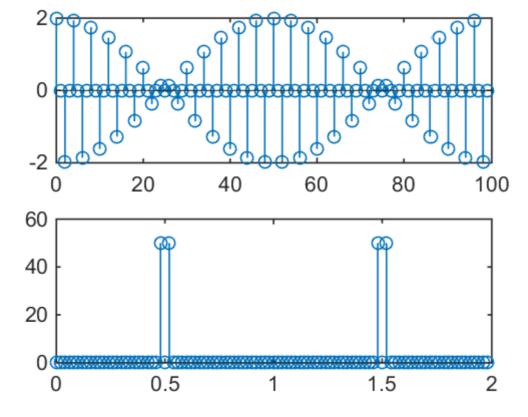
Out[26]:

use 100 samples of x[n]

```
In [27]: %plot -s 560,420

n3 = n;
y3 = x;
y3 = dft(y3,100);
k3 = n3;
w3 = 2*pi/100*k3;

subplot(2,1,1), stem(n3,y3)
subplot(2,1,2), stem(w3/pi,abs(Y3))
```



Out[27]:

conclusion:

padding more zeros to the 100-point sequence will result in a smoother rendition of the spectrum but will not reveal any new information.