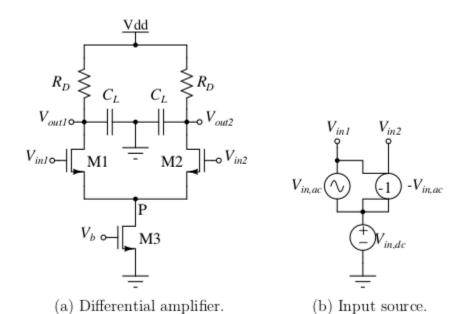
EE 660 VLSI Design Laboratory

Assignment III

Submitted by: L Sri Sai Swathi (2414202)

1. Design an NMOS common-source differential amplifier with resistive load as shown in Fig. 1a. You can make use of UMC65 models for the transistors. Given (W/L) 1,2,3 = $6\mu/0.06\mu$, R D = 1 k Ω , C L =100 fF, and V b = 0.55 V.



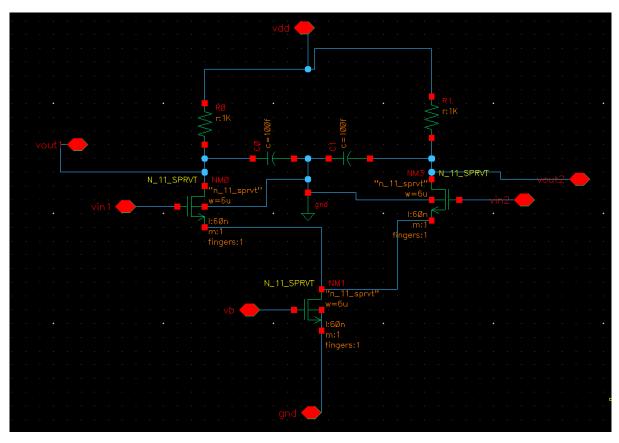


Figure: Schematic showing NMOS common-source differential amplifier with resistive load.

a) The configuration shown in Fig. 1b is used to generate the input for the amplifier. You can use the small signal amplitude of V in,ac to be $10\,mV$, and frequency to be $1\,kHz$. Decide the value of V in,dc , say $(V\ b1\)$, such that the differential gain is at least 2. Calculate the value of gm1 from the gain assuming the second order effects are absent.

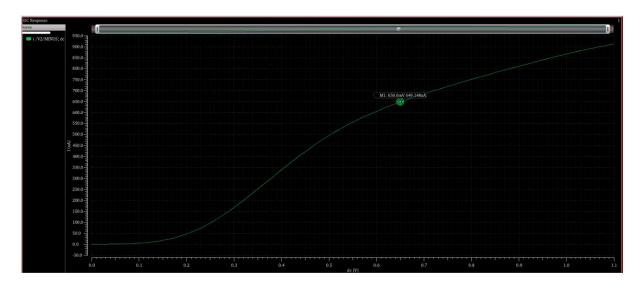


Figure:Plot showing vin dc vs I with vin dc as 500mv

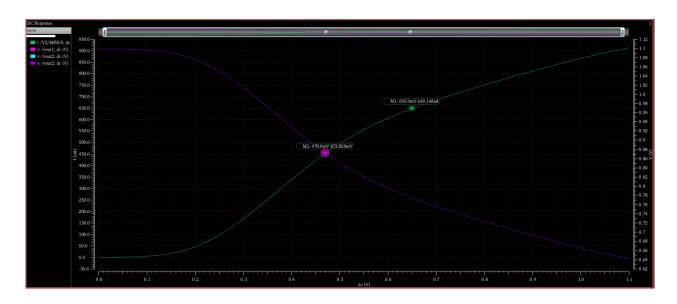


Figure:Plot showing vin dc vs I, vout1, vout2 with vin dc as 500mv

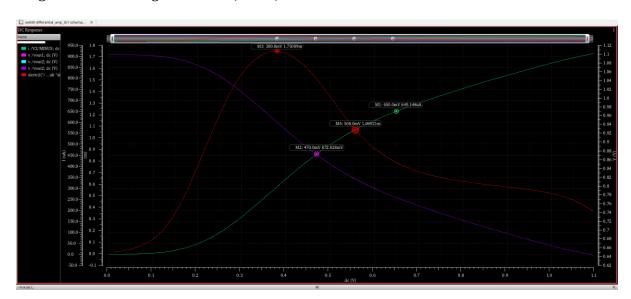


Figure:Plot showing vin dc vs I, vout1, gm with vin dc as 500mv

Observation : From the plot gm is identified to be 1.75m seimens.

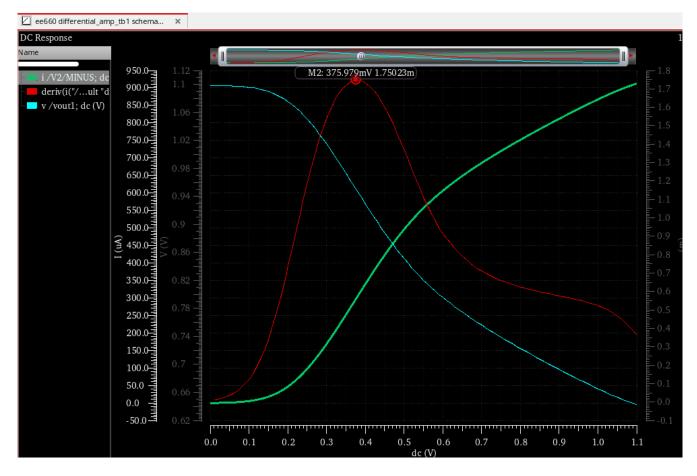


Figure:Plot showing I, vout1, gm with vin dc as 650mv We have also tried increasing vindc from 500mV to 750mV and observed gm to be 1.75mS.

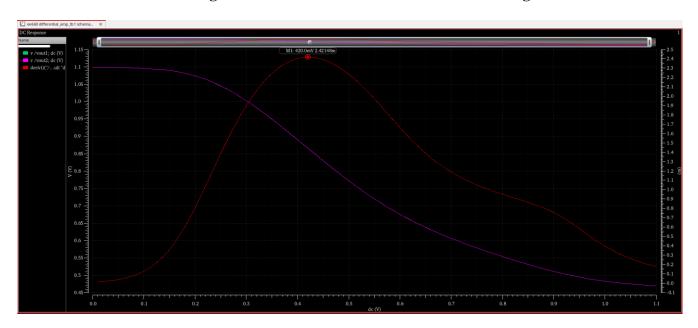


Figure:Plot showing d vout /d vindc.

Observation: From the above plots, choosing 650mv as dc bias point. We have also observed that gm as 1.75mS. And dvout/dvin as 2.4

Below are the simulation results of DC operating point analysis at dc voltage 0.5v

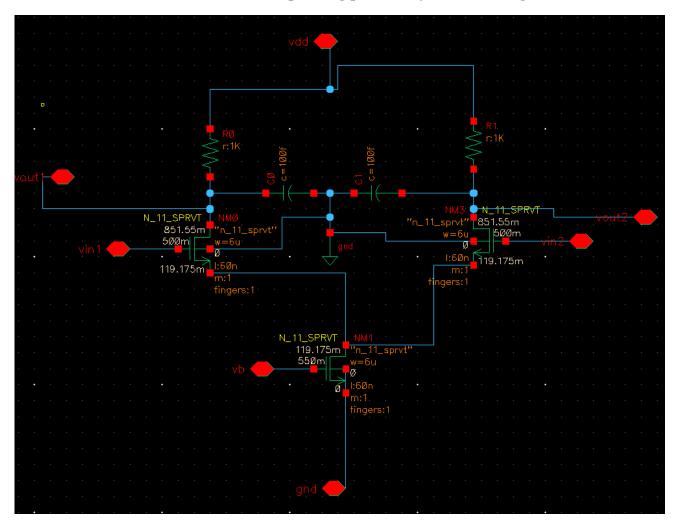


Figure: Schematic showing NMOS common-source differential amplifier with resistive load with all the operating point values.

Simulation result of first mosfet M1:

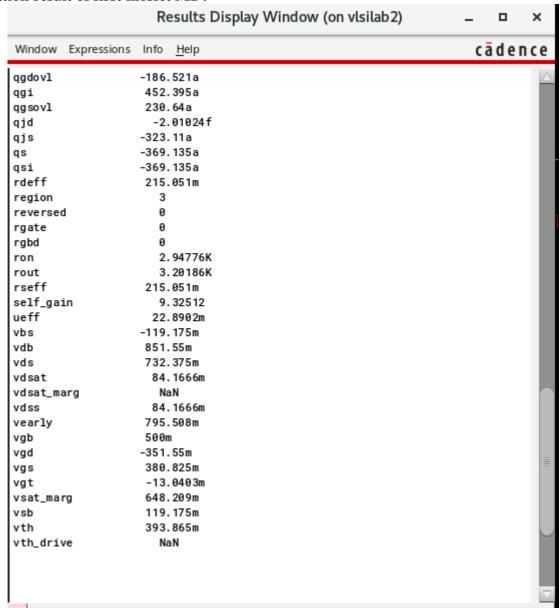


Figure: Simulation result of DC operating point analysis.

Observation: M1 is in region 3 at vindc =500mv

Simulation result of mosfet M2:

	Results Display Window (on vlsilab2)	×
Window Expressions		cādenc
ıgdovl	-186.521a	
qgi	452.395a	
qgsovl	230.64a	
ıjd	-2.01024f	
ıjs	-323.11a	
įs	-369.135a	
Įsi	-369.135a	
rdeff	215.051m	
region	3	
reversed	0	
rgate	0	
rgbd	0	
ron	2.94776K	
rout	3.20186K	
rseff	215.051m	
self_gain	9.32512	
ueff	22.8902m	
/bs	-119.175m	
/db	851.55m	
/ds	732.375m	
/dsat	84.1666m	
/dsat_marg	NaN	
/dss	84.1666m	
/early	795.508m	
/gb	500m	
/gd	-351.55m	
/gs	380.825m	
/gt	-13.0403m	
/sat_marg	648.209m	
/sb	119.175m	
/th	393.865m	
vth_drive	NaN	

Figure: Simulation result of DC operating point analysis.

Observation: M2 is in region 3 at vindc =500mv

Simulation result of mosfet M3:

	Results Display Window (on v	vlsilab2) _ 🗆 ×
Window Expressi	ons Info <u>H</u> elp	cādence
qdi	-276. 119a	
qg	1.22036f	
qgdovl	260.889a	
qgi	1.22036f	
qgsovl	332.69a	
qjd	-322.687a	
qjs	-297.94z	
qs	-631.032a	
qsi	-631.032a	
rdeff	215.051m	
region	1	
reversed	0	
rgate	0	
rgbd	0	
ron	239.838	
rout	441.636	
rseff	215.051m	
self_gain	1.42659	
ueff	21.4591m	
vbs	0	
vdb	119.175m	
vds	119.175m	
vdsat	141.858m	
vdsat_marg	NaN	
vdss	141.858m	
vearly	219.448m	
vgb	550m	
vgd	430.825m	
vgs	550m	
vgt	124.093m	
vsat_marg	-22.6832m	
vsb	-0	
vth	425.907m	
vth_drive	NaN	
14		

Figure: Simulation result of DC operating point analysis.

Observation: M2 is in region 1 at vindc =500mv. Since all 3 are not in saturation increasing vin, dc to 650mv.

Below are the simulation results of DC operating point analysis at dc voltage 650mv:

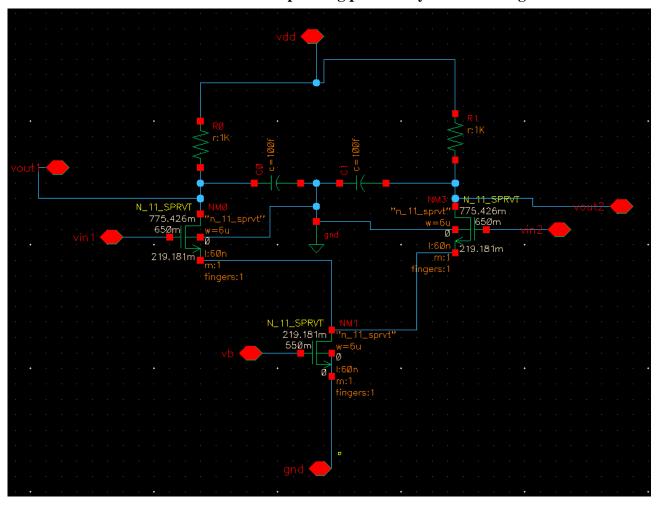


Figure: Schematic showing NMOS common-source differential amplifier with resistive load with all the operating point values.

Simulation result of first mosfet M1:

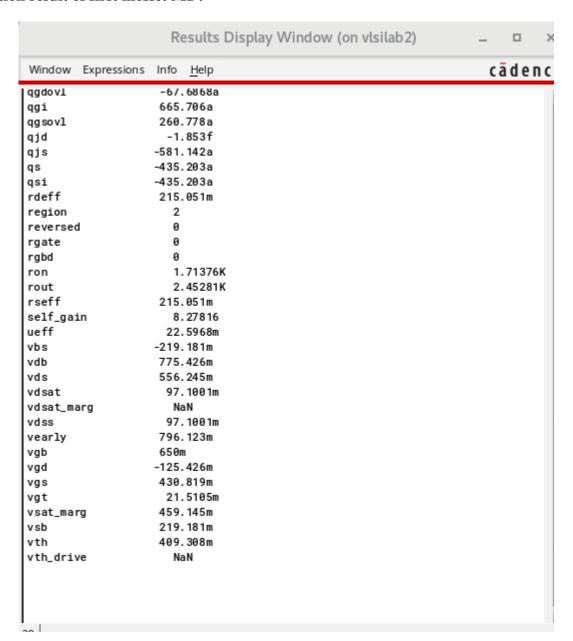


Figure: Simulation result of DC operating point analysis.

Observation: M1 is in region 2 at vindc =650mv

Simulation result of mosfet M2:

		Re	esults I	Display W	Vindow	(on vls	ilab2)	_		×
Window	Expressions	Info	<u>H</u> elp					cā	d e i	n c e
qdi		67.	3513a							
qg			706a							
qgdovl			6868a							
qgi			706a							
qgsovl			778a							
qjd			853f							
qjs			142a							
qs			203a							
qsi			203a							
rdeff			.051m							
region .		2								
reversed		9								
rgate		9								
rgbd		9	712761							
ron			71376K							
rout rseff			.45281K .051m							
self_gai			27816							
ueff			. 5968m							- 1
vbs			. 181m							
vdb			.426m							- 1
vds			245m							
vdsat			1001m							
vdsat_ma	ra		aN							
vdss	. 9		1001m							- 1
vearly			123m							
vgb		650n	1							
vgd			426m							- 1
vgs			819m							- 1
vgt		21.	5105m							
vsat_mar	g	459.	145m							
vsb	_	219.	181m							
vth		409.	308m							
vth_driv	e	Na	aΝ							

Figure: Simulation result of DC operating point analysis.

Observation: M2 is in region 2 at vindc =650mv

Simulation result of mosfet M3:

	Results Display Window (on vlsilab2)	_ 0 3
Window Expressions	Info <u>H</u> elp	cādeno
qdi	-205.163a	
qg	1.16997 f	
qgdovl	200.626a	
qgi	1.16997 f	
qgsovl	332.67a	
qjd	-580.613a	
qjs	-389.224z	
qs	-643.979a	
qsi	-643.979a	
rdeff	215.051m	
region	2	
reversed	0	
rgate	0	
rgbd	0	
ron	337.644	
rout	911.308	
rseff	215.051m	
self_gain	4.09135	
ueff	21.5372m	
vbs	0	
vdb	219.181m	
vds	219.181m	
vdsat	144.923m	
vdsat_marg	NaN	
vdss	144.923m	
vearly	591.575m	
vgb	550m	
vgd	330.819m	
vgs	550m	
vgt	130.209m	
vsat_marg	74.2584m	
vsb	-0	
vth	419.791m	
vth_drive	NaN	
20		

Figure: Simulation result of DC operating point analysis.

Observation: M3 is in region 2 at vindc =650mv. It is observed that all the 3 mosfets are in saturation region .i.e region2

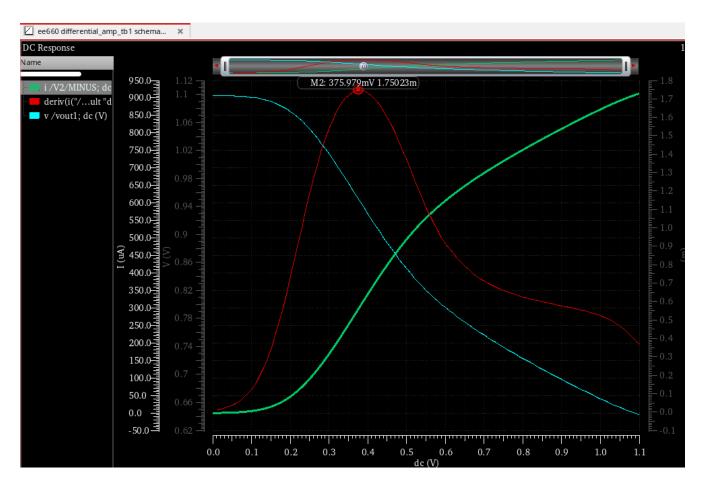


Figure:Plot showing I, vout1, gm with vin dc as 650mv

(b) Perform the operating point analysis with V in,dc = V b1 and check if all the transistors are in

saturation region of operation. If not, find out another value for V b1 . Calculate V out1,max and V out1,min , and hence the maximum differential output voltage swing. From this calculate the maximum input swing that can be applied to the amplifier

Performed transient analysis at Vin,dc=650mV, small signal voltage is 10mV, frequency =1KHz

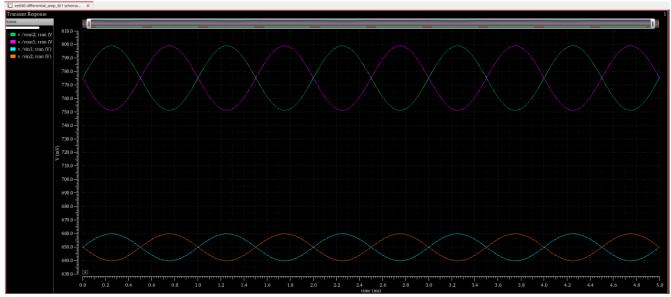


Figure: Plot showing Vin1, Vin2, Vout1, Vout2

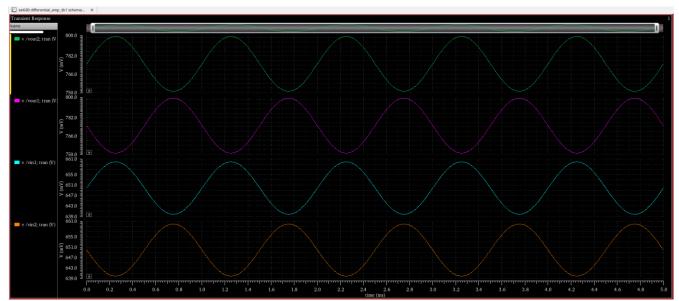


Figure:Plot showing Vin1,Vin2,Vout1,Vout2

Observation: From the above plot it is evident that Vin1, Vout1 are out of phase and Vin2, Vout2 are out of phase.

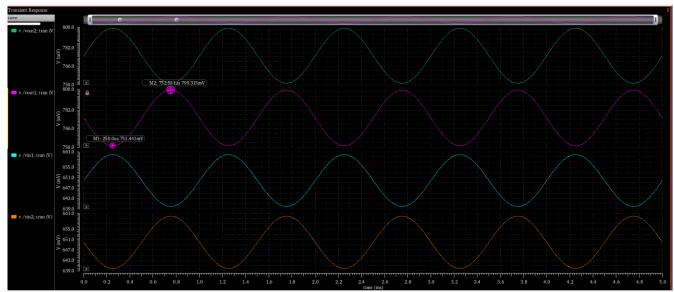


Figure: Plot showing Vin1, Vin2, Vout1, Vout2

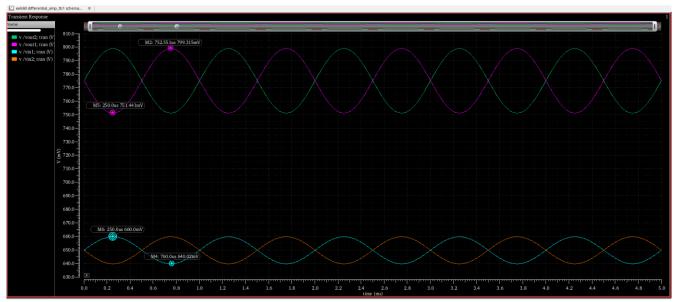


Figure: Plot showing Vin1, Vout1 with min and max points highlighted

Observation:

Vout1max=799.31mV

Vout1min= 751.44mv

Vout,swing=47.87mV

Vdiff swing =2x(Vout1max-vout1min)=2 x47.87=95.74mV

Vin,pp=20mv

Av = Vout1,pp/Vin,pp = 2.39

Vin,swing =vout,swing /2.39=20mV

(c) Perform the AC analysis of the circuit and find out the 3 dB frequency. What is the phase difference between the input and output at this cut-off frequency? Is it possible for you to increase the bandwidth without reducing the gain by changing R D and/or W/L ratios?

Below are simulation results of the AC response analysis:

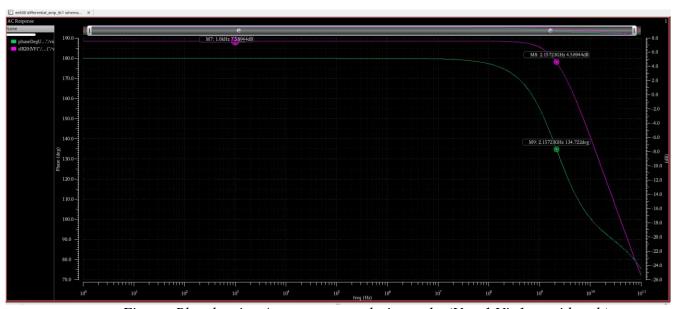


Figure: Plot showing Ac response analysis results (Vout1, Vin1 considered)

Observations: From the above Ac Analysis plots we can say that:

- 1.Low frequency gain is 7.589dB
- 2. 3dB cuttoff frequency is **2.157GHz**
- 3. Phase difference between the input and output at the 3 dB frequency: 134.72 Degrees.

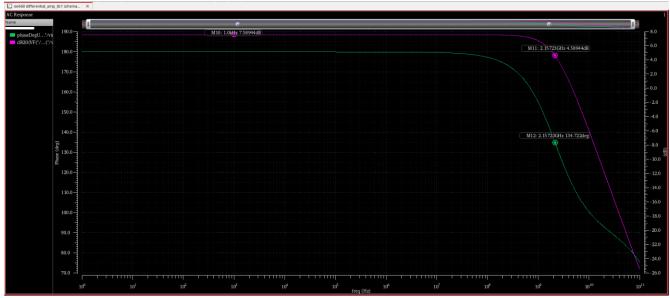


Figure: Plot showing Ac response analysis results (Vout2, Vin2 considered)

AC response Vout2,vin2

Observations: From the above Ac Analysis plots we can say that :

- 1.Low frequency gain is **7.589dB**
- 2. 3dB cuttoff frequency is **2.157GHz**
- 3. Phase difference between the input and output at the 3 dB frequency: 134.72 Degrees.

Is it possible for you to increase the bandwidth without reducing the gain by changing R D and/or W/L ratios?

Trial 1:When RD is increased from 1k to 1.5k



Figure :Plot showing Frequency and Phase response obtained by performing AC analysis.

Observations: From the above Ac Analysis plots we can say that :

- 1.Low frequency gain is **9.424dB**
- 2. 3dB cuttoff frequency is 1.689GHz

Bandwidth got reduced.

Trial 2: Changed Aspect ratios i.e M1,M2 has width 9um and M3 has width 8um



Figure :Plot showing Frequency and Phase response obtained by performing AC analysis.

Observations: From the above Ac Analysis plots we can say that :

- 1.Low frequency gain is **9.33dB**
- 2. 3dB cuttoff frequency is **2.43GHz**

Trial 3:when all three transistors has width as 9um



Figure :Plot showing Frequency and Phase response obtained by performing AC analysis.

Observations: From the above Ac Analysis plot we can say that :

- 1.Low frequency gain is **9.48dB**
- 2. 3dB cuttoff frequency is **2.53GHz**

Trial4: when m1, m2 has w = 6um and m3 has w = 9um



Figure :Plot showing Frequency and Phase response obtained by performing AC analysis.

- 1.Low frequency gain is **8.40dB**
- 2. 3dB cuttoff frequency is 2.36GHz



Trial5: when m1, m2, m3 has w = 9um and Rd = 0.5Kohms

Figure :Plot showing Frequency and Phase response obtained by performing AC analysis.

Observation:

- 1.Low frequency gain is 6.08dB
- 2. 3dB cuttoff frequency is 3.94GHz

Conclusion: It is not possible for us to increase the bandwidth without reducing the gain by changing R D and/or W/L ratios

(d) Perform the transient analysis of the circuit with the input small signal voltage assumed to be 10 mV and 20 mV, and plot the differential input and output waveforms.

Performed transient analysis at Vin,dc=650mV, small signal voltage is 10mV, frequency =1KHz

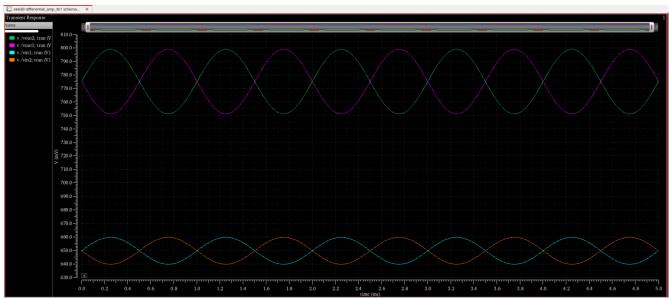


Figure: Plot showing Vin1, Vin2, Vout1, Vout2

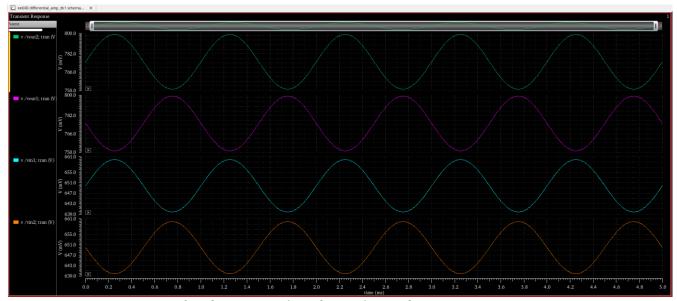


Figure: Plot showing Vin1, Vin2, Vout1, Vout2

Observation: From the above plot it is evident that Vin1 ,Vout1 are out of phase and Vin2 ,Vout2 are out of phase.

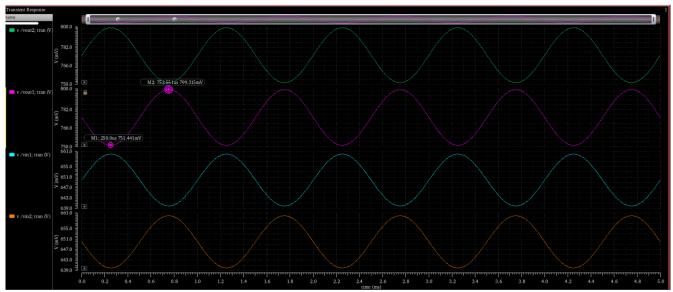


Figure: Plot showing Vin1, Vin2, Vout1, Vout2

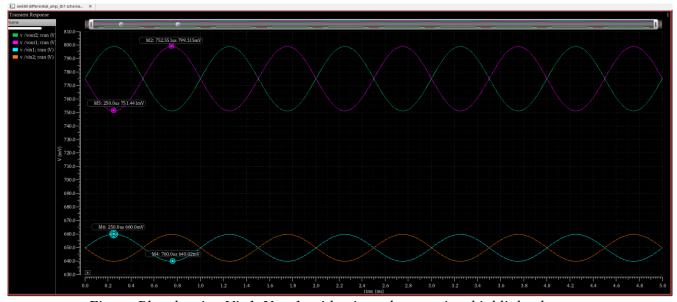


Figure:Plot showing Vin1, Vout1 with min and max points highlighted

Observation:

Vout1max=799.31mV

Vout1min= 751.44mv

Vdiff swing =2x(Vout1max-vout1min)=2 x47.87=95.74mV

Vin,pp=20mv

Av = Vout1,pp/Vin,pp = 2.39

Below are the results when input small signal input voltage is 20mV:

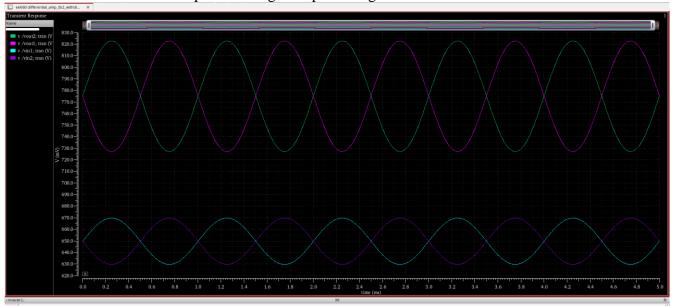


Figure: Plot showing Vin1, Vin2, Vout1, Vout2

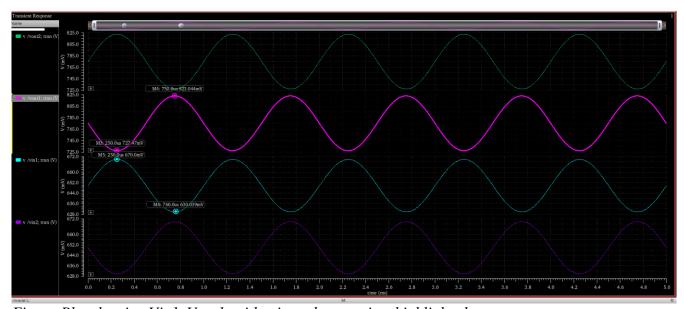


Figure:Plot showing Vin1, Vout1 with min and max points highlighted

Observation:

Vout1max=823mV

Vout1min= 727mv

Vdiff swing =2x(Vout1max-vout1min)=2 x96=192mV

Vin,pp=40mv

Av = Vout1,pp/Vin,pp = 2.4

e) Remove the VCVS from Fig. 1b and connect V in 2 to V in,dc . Perform the transient analysis of the circuit with the input small signal voltage assumed to be 10 mV. Plot the waveforms of V out and V out 2, and explain the operation of the circuit.

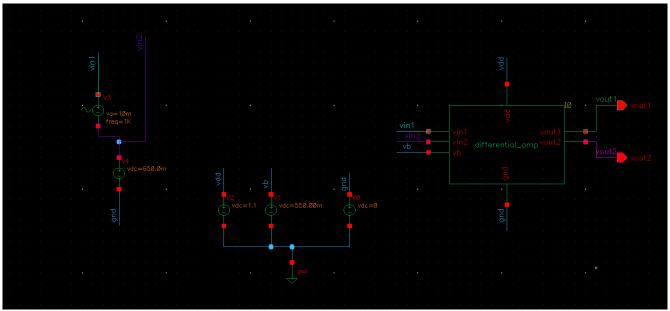


Figure: Schematic showing differential amplifier with vin2 connected to vin, dc and vin1 to ac input

Transient Analysis Simulation Results:

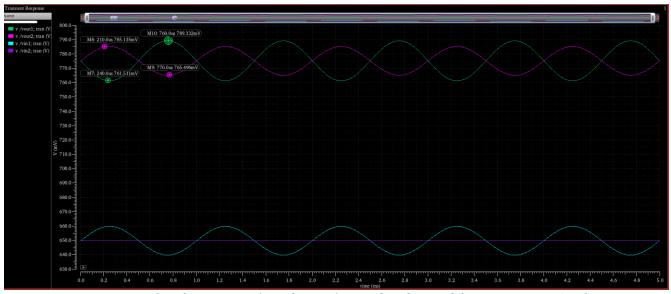


Figure: Plot showing Vin1, Vin2, Vout1, Vout2 obtained from Transient Analysis

Explanation of above differential Circuit operation:

Here Vin1 is sinusoidal voltage source with 10mv ac amplitude and 0.65v dc voltage, whereas Vin2 has only dc voltage of 0.65V .Since ,vin2 is fixed circuit will respond to the changes in the vin1. The tail current source (M3) provides a fixed total current that is split between M1 and M2.When Vin1 increases above Vin2, M1 draws more current, and M2 draws less. Since the total current remains constant, the current flowing through M1 and M2 adjusts in an opposite manner. This current variation directly affects the drain voltages Vout1 and Vout2.

When Vin1 increases (positive half-cycle of AC input)

M1 draws more current \rightarrow Voltage drop across RD increases \rightarrow Vout1 decreases.

M2 draws less current \rightarrow Voltage drop across RD decreases \rightarrow Vout2 increases.

When Vin1 decreases (negative half-cycle of AC input)

M1 draws less current \rightarrow Voltage drop across RD decreases \rightarrow Vout1 increases.

M2 draws more current \rightarrow Voltage drop across RD increases \rightarrow Vout2 decreases.