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DAYANANDA SAGAR COLLEGE OF ENGINEERING

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Internal Assessment Test - I

Course: MIMO Communication

Course Code: TE814 Semester: VIII 'A' &'B' Date; 24/02/2020 Maximum marks: 50 Duration: 90 Min

		Note: Answer 5 full questions.							
1	a)	A which process the redundancy received signals i)Combiner ii)Equalizer iii) PCM iv) All of mentioned							
	b)	Diversity is used to combat							
	.,	i)Fading ii) Co channel interference iii) Error burst iv) all of mentioned							
	c)	Diversity techniques may exploit the propagation, resulting in a diversity gain							
	,	i)Single path ii) Multipath iii) Narrow path iv) Fading path							
	d)	The path loss exponent of the free space environment i.e. when only line of sight exists, no scatters is							
		i)2 ii) 3 iii) 6 iv) None of the above							
	e)	In combiner, the first fully received and valid data packet will be immediately further processed, whereas the later arriving redundant packets will be immediately discarded after reception. i)Max-Ratio ii) Equal gain iii) Scanning/Switching iv) Selection							
	f)	Space diversity means using different physical parts for the signal, at a frequency i)Multiple ii) Single iii) Co channel iv) Orthogonal							
	ဋ)	order means how many degrees of freedom you can have in your design							
		i)Rank ii) Selection iii)Diversity iv) Uplink/Downlink							
	h)	The multipath spread is basically the difference between the shortest and the longest paths that							
		the transmitted signal goes through i)Time ii) Frequency iii) Both time and frequency iv)None of the above							
	:\								
	i)	If the channel is said to be frequency selective fading channel, then it exhibits interference ii)Intersymbol interference iii) No Intersymbol interference iii) Both a and b iv) none of the above							
	i)	Channel capacity, which is defined as the rate at which we can transmit information with an							
	.17	arbitrary probability of error							
		i)Maximum, High ii) Maximum, Low iii) Minimum, Low iv) None of the above							
		Define Channel Fading. Also discuss the wireless propagation mechanism							
	,	Describe about Soft Decision Decoding and Hard Decision Decoding.							
		Focus on Capacity of Noisy Channel.							
		Analyze the information Rate of Noisy Channel.							
	Disc	cuss the classification in Channel Fading (OR)							
		borate Diversity Techniques in Wireless Communication							
	Disc	cuss Beamforming technique with Smart Antennae							
		(OR)							
	Rric	ofly explain demodulation of signal from set of received signals in MIMO							

Solution Scheme Dale: 24/02/2020 Prepared by : Dr. Sayed Abdullogan Ist CIE Subject: MIMD Comm. Code: TE814 Sem 8th A 4B Q1. a) O Combiner — (i) 6 all of mentioned — (iv) @ Multipoth — (ii) 1 X10 1 a - 1 (e) Selection — (V) Boundle (ii)
Diversity (iii) (b) Time - (i) (i) Intersymbol Interference (i) (j) Maximumi, Low Q2. @ Channel Fading! The Time variation of received Signa The sime war due to changes in transmissions power due to changes in transmission power as fading.

medium or path is known as fading.

medium depends on atmospheric conditions and fading depends on atmospheric path obstackles.

Whiteless channels, signals are recein through multiple poths. There is significant time variations caused by the relative motion of transmitter and receives. Electromognetic waves travel through three different mechanismi (3)- Reflections - Refroction - Scottering (2.6) Soft Decision Coding Hard Decision Decoding HDD is relatively Simple 1) Difficult to Implement to Implement 2 Does not gives the 2) Gives the best performance best performance. OReceiver is simple. (3) Receiver is complex (4) Channel decoder directly (4) Max likelihood decoder picks the codeword closest works with the matched of filter output to perform to the received sequence in Hamming distance senge as The decoder output max likelihood decoding, (5) Here code constraint Bentaline Hard decisions are made based on matches filter are road considered or correlator outputs without. 3.a Capacity of Norsy Channel
W= message

page-2

1N= message n=noof channel X, X2 ... = Input channel Y, Y2 -- = Output channels C=Channel Capacity messege Sneeden X, X2 Schamel Y, Y2 1/2 decoder estimate I(";") -> Mutual Information $G = \lim_{n \to \infty} \frac{1}{n} \max_{p(x_1, x_n)} I(x_1, x_n - x_n) I(x_1, x_n)$ C = max I(x, y) (5)G = 69 (1+P) Information Rate of Noisy Channel Ci = lim I max I(X,...Xx; X... Yn)

Ci = lim p(x1,x2...xn) I(X,...Xx; X... Yn) Maximum mutal Informations

Par memoryless channels i.e of the ontput-mpt

relationship can be willen as

P(y1, y2--yn) x1,x12-.xn)= TP(yi/xi) G= max I (X; Y) => mutual Information $I(X;Y) = E\left[\log \frac{P(X,Y)}{P(X),P(Y)}\right]$ E> expectation is over joint density of | Page-3

Q H.g. Classification of Channel Fading Path Loss Fading: - It is Large Scale fad.
We can write received Power Po as d' Separation between trensmitter and $P_r = \frac{c}{d^n} P_t$ Pt -> transmitted power. n -> Path loss exponent in range of 2 to 6. Shadowing :- 91 is Large scale fading Here received Signal Strength is averaged over a long period of time of order of few seconds or minutes. PraBm = PraBm + XrdB dBm, Xo -> Zero mean sandom varrette. rans > U=U Brook Frequency Randing Selective Fading B.W=W. Bc = coherence B-W If Signal Bandwidth Wis Significantly Smaller than the wherence B. H Be of thomas all frequency components see the same all frequency components see the same effective channel; e C(fit) for the frequency range of interest will be independent of the page-4

yi(t)=C(O;t) & x(l) darty = C(0; t) x((t)) M & Bc > multipath spread of channel is significantly smaller than Signal time durthin, 1-t. -> No Inter symbol Interference between consecutive symbols. -> Different frequency components of undergo different channel fachs. Flat Frequency fading

Flat Frequency fading

(CO,nT) = constant

M>Bc; C(O,t) & duration is smaller

tm>Ts => Symbol time is is smaller

than Signal time.

All frequency components of Signal

andergo same channel fades. Symbol duration Ts is significantly

Symbol duration Ts is significantly

smaller than coherence time of channels

channels than coherence time of channels

smaller than coherence time of channels

channels th Slow. Fading / Fast Fading Fast Fading

Symbol durature Ts is greater than

Symbol durature of charmed

wherence time of charmed

coherence time of charmed

coherence time of charmed

Annel differs over some symbol

Raulo, oh Ind-Kayleigh Fading pege-5

Kaylaigh Fading channel crefficient has zero mean.

channel gain has absolute value Rayley

C = Complex Gaussian Random variable

Rician Fadrig Rician Fadring
Channel gain has non zero mean
C = multiplicative term = complex Gaussian Landon Variable Q5 Diversity techniques ian Diversity techniques collectively refer to methods of improving this performance by effectively transmitting I performance by effectively transmitting I the Same information multiple times where each replico sees a different ideally independent channel. Trequency Diversity Space Diversity Time Diversity. Transmit Diversity Receive Diversity Chamel Coding Explanations. Q6. Beam Forming techniques with Smart Antenna Consider a Scenario of small number of local scallerers and comm by page - 6.

In Such scenario multiple antenna elements can be employed at transmitter in order to provide directionality for electromognettic waves, hence improving the effective SNR of channel. A fundamental approach sho is the transmission of Same Signal from
each of antenna elements with a certain
gain and phase shift.
This can be Implemented Analog of
Digital. Digital Beam forming is
A more blowble Different techniques can be used to Different techniques. of more florable find Suitable coefficients and of antenna beams from beedback to make coefficients read be comprised to make effective antenna patterns have main beam in this required directures. (4) Also Interference suppression can be accomplished using beam forming. Demodulation from Signal Received from Set of MIMO Maximum Ratio Combining Assun channel coefficients are perfectly available at Maxi Likelihood decision rule en symbol transmilled is 2 = aigmax p(y, y2 - y/h, h2. h,xi) Page-7.

p(·1·) -> conditional joint Pd.f Since chattenel goins, transmitted signal, received Signal are independent. $\hat{\chi} = \underset{j=1,2...M}{\text{argminRe}} \left\{ \left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right\} - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] + \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] - \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] + \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] + \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] + \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] + \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] + \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] + \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] + \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{l} \right) \chi_{j}^{*} \right] + \frac{1}{2} \left[\left(\sum_{l=1}^{\infty} h_{l}^{*} \gamma_{$ SNR of L branch diversity with MRC each branch. Hence error probabilities and onlage will be reduced. Works with branch with best channel · Selection Combining; condition for any given transmission condition for any given transmission picked (2) Equal Gain Combining; -Signals of all tranches are cophased by multiplyingwith e-j/h! using basebend molahing -Signals Summed logether to form equivalent channel output Switch and Stay Combining: -We use particular brench for demodition until SNR of it falls below threshold; Then The of it of the other branch which has largest SNR is choosen and mainteined till it also falls below threshold. This procedure is Carried. page-8