

Assignment 3

04/02/2021

1. Determine the coefficients of FIR filter using rectangular window with following

$$H_d(e^{j\omega}) = \begin{cases} e^{-j3\omega} & -\pi/4 \leq \omega \leq \pi/4 \\ 0 & \pi/4 \leq \omega \leq \pi \end{cases}$$

Length of the filter is 7.

Ans1. Taking, I.F.T of $H_d(e^{j\omega})$,
$$h_d(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H_d(e^{j\omega}) \cdot e^{j\omega n} d\omega$$

$$h_d(n) = \frac{\sin \frac{\pi(n-3)}{4}}{\pi(n-3)}$$

- Considering a rectangular window,

$$W_r(n) = \begin{cases} 1 & n=0 \text{ to } 6 \\ 0 & \text{otherwise.} \end{cases}$$

Response of filter is $h(n) = h_d(n) \cdot w(n)$.

To determine the coefficients, find the value of $h_d(n)$ for n in the range 0 to 6. (As $w_r(n)$ is 1).

The values are 0.075, 0.159, 0.225, 0.25, 0.225, 0.159, 0.075.

2. Let $h_d[n]$, $-\infty < n < \infty$, denotes the impulse response samples of a zero-phase filter with frequency response $H_d(\exp\{j\omega\})$. It is known that the frequency response $H_t(\exp\{j\omega\})$ of the zero-phase FIR filter $h_t[n]$, $-M \leq n \leq M$, obtained by multiplying $h_d[n]$ with a rectangular window $w_R[n]$, $-M \leq n \leq M$, has the least integral-squared error Φ_R defined in the following equation.

$$\Phi_R = \sum_{n=-\infty}^{\infty} |h_t[n] - h_d[n]|^2$$

Let Φ_{Hann} denote the integral-squared error if a length $2M+1$ Hann window is used to develop FIR filter. Determine an expression for the excess error

$$\Phi_{\text{excess}} = \Phi_R - \Phi_{\text{Hann}}$$

Ans2.

$$\Phi_R = \frac{1}{2\pi} \int_{-\pi}^{\pi} \left| H_t(e^{j\omega}) - H_d(e^{j\omega}) \right|^2 d\omega, \text{ where } H_t(e^{j\omega}) = \sum_{n=-M}^M h_t[n] e^{-j\omega n}.$$

$$\begin{aligned} \text{Using Parseval's relation, we can write } \Phi_R &= \sum_{n=-\infty}^{\infty} |h_t[n] - h_d[n]|^2 \\ &= \sum_{n=-M}^M |h_t[n] - h_d[n]|^2 + \sum_{n=-\infty}^{-M-1} h_d^2[n] + \sum_{n=M+1}^{\infty} h_d^2[n]. \end{aligned}$$

$$\begin{aligned} \text{Now, } \Phi_{Haan} &= \sum_{n=-\infty}^{\infty} |h_d[n] \cdot w_{Haan}[n] - h_d[n]|^2 \\ &= \sum_{n=-M}^M \left| h_d[n] \left(\frac{1}{2} + \frac{1}{2} \cos\left(\frac{2\pi n}{2M+1}\right) \right) - h_d[n] \right|^2 + \sum_{n=-\infty}^{-M-1} h_d^2[n] + \sum_{n=M+1}^{\infty} h_d^2[n] \end{aligned}$$

$$\begin{aligned} \text{Hence, } \Phi_{Excess} &= \Phi_R - \Phi_{Haan} \\ &= \sum_{n=-M}^M |h_d[n] \cdot w_R[n] - h_d[n]|^2 - \sum_{n=-M}^M \left| h_d[n] \cdot \left(\frac{1}{2} + \frac{1}{2} \cos\left(\frac{2\pi n}{2M+1}\right) \right) - h_d[n] \right|^2 \\ &= - \sum_{n=-M}^M \left| \frac{h_d[n]}{2} \cos\left(\frac{2\pi n}{2M+1}\right) - \frac{h_d[n]}{2} \right|^2 = \frac{1}{2} (1 + 2M) \left| \cos\left(\frac{2\pi M}{2M+1}\right) - 1 \right|^2. \end{aligned}$$

3. Repeat above problem if a Hamming window is used instead.

Ans3.

$$\Phi_R = \sum_{n=-\infty}^{\infty} |h_t[n] - h_d[n]|^2 \text{ and } \Phi_{Hamm} = \sum_{n=-\infty}^{\infty} |h_d[n] \cdot w_{Hamm}[n] - h_d[n]|^2.$$

$$\begin{aligned} \text{Therefore, } \Phi_{Excess} &= \Phi_R - \Phi_{Hamm} = - \sum_{n=-M}^M \left| h_d[n] \left(0.46 \cos\left(\frac{2\pi n}{2M+1}\right) - 0.46 \right) \right|^2 \\ &= - \sum_{n=-M}^M \left| 0.46 h_d[n] \left(\cos\left(\frac{2\pi n}{2M+1}\right) - 1 \right) \right|^2 = 0.46(2M+1) \left| \cos\left(\frac{2\pi M}{2M+1}\right) - 1 \right|^2. \end{aligned}$$

4. Determine the coefficients $\{h(n)\}$ of a linear-phase FIR filter of length $M=15$ which has a symmetric unit sample response and a frequency response that satisfies the condition

$$H_r\left(\frac{2\pi k}{15}\right) = \begin{cases} 1, & k = 0, 1, 2, 3 \\ 0 & k = 4, 5, 6, 7 \end{cases}$$

Ans4.

$$M = 15. H_r\left(\frac{2\pi k}{15}\right) = \begin{cases} 1, & k = 0, 1, 2, 3 \\ 0, & k = 4, 5, 6, 7 \end{cases}$$

$$H_r(w) = h\left(\frac{M-1}{2}\right) + 2 \sum_{n=0}^{\frac{M-3}{2}} h(n) \cos w\left(\frac{M-1}{2} - n\right)$$

$$h(n) = h(M-1-n)$$

$$h(n) = h(14-n)$$

$$H_r(w) = h(7) + 2 \sum_{n=0}^6 h(n) \cos w(7-n)$$

Solving the above eqn yields,

$$h(n) = \{0.3189, 0.0341, -0.1079, -0.0365, 0.0667, 0.0412, -0.0498, 0.4667, 0.4667, -0.0498, 0.0412, 0.0667, -0.0365, -0.1079, 0.0341, 0.3189\}$$

5. Design a 5-tap FIR band reject filter with a lower cutoff frequency of 2,000 Hz, an upper cutoff frequency of 2,400 Hz, and a sampling rate of 8,000 Hz using the Hamming window method. Determine the transfer function.

Ans5.

$2N+1 = 5$ which leads that $N=2$, $h(n)$ is given by

$$h(n) = \begin{cases} \frac{\pi - \omega_H + \omega_L}{\pi}, & n = 0 \\ -\frac{\sin(\omega_H n)}{\pi n} + \frac{\sin(\omega_L n)}{\pi n} & n \neq 0 \end{cases}$$

Then the five samples of $h(n)$ shifted by 2 is equal to

$$h(2) = 0.9,$$

$$h(1) = h(3) = 0.01558$$

$$h(0) = h(4) = 0.09355$$

For Hamming window with $M=5$

$$w(n) = 0.54 - 0.46 \cos\left(\frac{n\pi}{2}\right)$$

and for $n = 0, 1, 2, 3, 4$

$$w_{ham}(2) = 1,$$

$$w_{ham}(1) = w_{ham}(3) = 0.54,$$

$$w_{ham}(0) = w_{ham}(4) = 0.08,$$

Then the windowed impulse response coefficients are

$$h_w(2) = 0.9,$$

$$h_w(1) = h_w(3) = 0.00841$$

$$h_w(0) = h_w(4) = 0.00748$$

Convert to z-domain

$$H(z) = 0.00748 + 0.00841z^{-1} + 0.9z^{-2} + 0.00841z^{-3} + 0.00748z^{-4}$$

6. What are the filter coefficients for a 3-tap FIR low pass filter with a cutoff frequency of 800 Hz and a sampling rate of 8,000 Hz using Hamming window? Also determine the transfer function and difference equation of the designed FIR system.

Ans6.

$$\omega_c = 2\pi f_c T_s = 2\pi \times 800 / 8000 = 0.2\pi$$

The coefficients of the lowpass filter

$$h(0) = 0.2, h(-1) = h(1) = 0.1871$$

For Hamming window with $M = 3$

$$w(n) = 0.54 + 0.46 \cos(n\pi)$$

and for $n = 0, 1, 2$

$$w_{ham}(0) = 1, w_{ham}(1) = w_{ham}(2) = 0.08$$

Then the windowed impulse response coefficients are

$$h_w(0) = 0.2, h_w(1) = h_w(2) = 0.01497$$

convert to z-domain

$$H(z) = 0.01497 + 0.2z^{-1} + 0.01497z^{-2}$$

Then the difference equation is

$$y(n) = 0.01497x(n) + 0.2x(n-1) + 0.01497x(n-2)$$

7. A Hilbert Transform is a filter with frequency response $H_d(w) = -j \text{sign}(w)$

A] Roughly plot the magnitude and phase plot of the filter.

B] Determine $h_d(n)$.

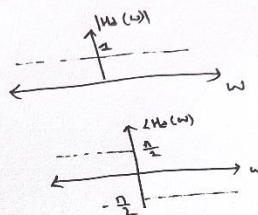
Ans7.

1. Given, $H_d(w) = -j \text{sign}(w)$

$$\therefore |H_d(w)| = 1 \quad \forall w$$

$$\angle H_d(w) = -\frac{\pi}{2} \quad \text{if } w > 0$$

$$= \frac{\pi}{2} \quad \text{if } w < 0$$



$$\begin{aligned} \rightarrow h_d(m) &= \frac{1}{2\pi} \int_{-\pi}^{\pi} -j \text{sign}(w) e^{j\omega m} d\omega \\ &= \frac{1}{2\pi} \int_{-\pi}^0 j e^{j\omega m} d\omega + \frac{1}{2\pi} \int_0^{\pi} -j e^{j\omega m} d\omega \\ &= \frac{1}{2\pi} \times \frac{j}{jm} [e^{j\omega m}]_{-\pi}^0 + \frac{1}{2\pi} \times \frac{-j}{jm} [e^{j\omega m}]_0^{\pi} \\ &= \frac{1}{2\pi m} [1 - e^{-jm\pi}] + \frac{1}{2\pi m} [1 - e^{jm\pi}] = \frac{1}{2\pi m} [1 - e^{-jm\pi} + 1 - e^{jm\pi}] = \frac{1}{2\pi m} \times 2 [1 - \cos m\pi] \\ &= \frac{1}{\pi m} \times 2 \sin^2\left(\frac{m\pi}{2}\right) \end{aligned}$$

$h_d(m) = \frac{1}{m\pi} \times 2 \sin^2\left(\frac{m\pi}{2}\right)$
 $= 0 \quad (m=0) \rightarrow \text{when } m \neq 0$

8. A Digital Filter is defined by the difference equation $y[n] = 0.99y[n-1] + x[n]$.

A] Determine the filter transfer function.

B] What is filter impulse function?

C] Is it LPF or HPF? (hint: check pole-zero plot)

Ans8.

$$3. \quad y(n) = 0.99 y(n-1) + x(n)$$

$$\rightarrow Y(z) = 0.99 z^{-1} Y(z) + X(z)$$

$$\text{or, } Y(z) (1 - 0.99 z^{-1}) = X(z)$$

$$\text{or, } H(z) = \frac{Y(z)}{X(z)} = \frac{1}{1 - 0.99 z^{-1}}$$

$$\therefore h(n) = (0.99)^n u(n)$$

[improving causality for practical implementation]

$$\rightarrow H(z) = \frac{1}{1 - 0.99 z^{-1}}$$

$$= \frac{z}{z - 0.99}$$

\hookrightarrow Pole at $z = 0.99$

$$\text{or, } z \approx 1$$

\downarrow
taking into Fourier domain by $z = e^{j\omega}$

\downarrow

Pole is close to $\omega = 0$

\downarrow

large values near low frequencies

\downarrow
LPF