

CLC404 Wideband, High-Slew Rate, Monolithic Op Amp

General Description

The CLC404 is a high-speed, monolithic op amp that combines low power consumption (110mW typical, 120mW maximum) with superior large signal performance. Operating off of $\pm\,5V$ supplies, the CLC404 demonstrates a large-signal bandwidth (5V $_{\rm pp}$ output) of 165MHz. The bandwidth performance, along with other speed characteristics such as rise and fall time (2.1ns for a 5V step), is nearly identical to the small signal performance since slew rate is not a limiting factor in the CLC404 design.

With its 175MHz bandwidth and 10ns settling (to 0.2%), the CLC404 is ideal for driving ultra-fast flash A/D converters. The 0.5° deviation from linear phase, coupled with -53dBc 2nd harmonic distortion and -60dBc 3rd harmonic distortion (both at 20MHz), is well suited for many digital and analog communication applications. These same characteristics, along with 70mA output current, differential gain of 0.07%, and differential phase at 0.03°, make the CLC404 an appropriate high-performance solution for video distribution and line driving applications.

Constructed using an advanced, complementary bipolar process and proven current feedback topologies, the CLC404 provides performance far beyond that of other monolithic op amps. The CLC404 is available in several versions to meet a variety of requirements. A three-letter suffix determines the version:

CLC404AJP -40°C to +85°C 8-pin plastic DIP CLC404AJE -40°C to +85°C 8-pin plastic SOIC

CLC404ALC -40°C to +85°C dice

dice qualified to Method 5008,

MIL-STD-883, Level B

CLC404AJM5 -40°C to +85°C 5-pin SOT

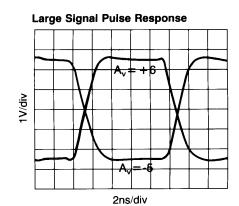
DESC SMD number: 5962-90994

Features

- 165MHz large signal bandwidth (5V_{pp})
- 2600V/µs slew rate■ Low power: 110mW
- Low distortion: -53dBc at 20MHz
- 10ns settling to 0.2%
- 0.07% diff. gain, 0.03% diff. phase

Applications

- Fast A/D conversion
- Line drivers
- Video distribution
- High-speed communications
- Radar, IF processors



Pinout Pinout SOT23-5 DIP & SOIC 5 | Vcc **Not Connected** Not Connected 7 v_{inv} +V_{CC} 2 V_{EE} 6 Vout V_{non-inv} 5 Not Connected $V_{non-inv}$

CLC404 Electrical Characteristics ($A_v = +6$, $V_{CC} = \pm 5V$, $R_g & R_L = 100\Omega$, $R_f = 500\Omega$; unless specified)

PARAMETER	CONDITIONS	TYP	MAX	& MIN RA	rings	UNITS	SYMBOL
Ambient Temperature	CLC404AJ	+25°C	-40°C	+25°C	+85°C		
FREQUENCY DOMAIN RESPO -3dB bandwidth -3dB large signal gain flatness peaking peaking rolloff linear phase deviation	VSE Vout<2V _{pp} Vout<5V _{pp} Vout<2V _{pp} <40MHz >40MHz <75MHz DC to 75MHz	175 165 0 0 0.2 0.5	>150 >140 <0.4 <0.7 <1.0 <1.0	>140 >140 <0.3 <0.5 <1.1 <1.0	>120 >110 <0.4 <0.7 <1.3 <1.2	MHz MHz dB dB dB	SSBW LSBW GFPL GFPH GFR LPD
TIME DOMAIN RESPONSE rise and fall time settling time to ±0.2% overshoot slew rate (measured at A _v +2)¹	2V step 5V step 2V step 2V step	2.0 2.1 10 5 2600	<2.4 <2.6 <15 <15 >2000	<2.4 <2.6 <15 <12 >2000	<2.9 <3.2 <15 <15 >2000	ns ns ns % V/µs	TRS TRL TS OS SR
DISTORTION AND NOISE RES 2nd harmonic distortion 3rd harmonic distortion equivalent input noise noise floor integrated noise differential gain ² differential phase ²	PONSE 2V _{pp} , 20MHz 2V _{pp} , 20MHz >1MHz 1MHz to 200MHz	-53 -60 -159 40 0.07 0.03	<-40 <-50 <-157 <45 —	<-45 <-50 <-157 <45 —	<-45 <-50 <-156 <50 —	dBc dBc dBm(1Hz) μV %	HD2 HD3 SNF INV DG DP
* input offset voltage average temperature coefficie input bias current average temperature coefficie input bias current average temperature coefficie vinput bias current average temperature coefficie power supply rejection ratio common mode rejection ratio supply current	non-inverting ent inverting	2 30 15 150 15 150 52 50	< ±9.0 < ±50 < ±44 < ±275 < ±40 < ±275 > 45 > 44 < 12	<±5.0	<±10.0 <±50 <±22 <±200 <±22 <±200 >45 >44 <12	mV μV/°C μA nA/°C μA nA/°C dB dB mA	VIO DVIO IBN DIBN IBI DIBI PSRR CMRR ICC
miscellaneous performation non-inverting input output impedance output voltage range common mode input range for output current	resistance capacitance at DC no load	1000 1 0.1 ±3.3 ±2.2 ±60	>250 <2 <0.3 >±2.8 >±1.4 >±35	>500 <2 <0.2 >±3.0 >±1.8 >±50	>1000 <2 <0.2 >±3.0 >±2.0 >±50	kΩ pF Ω V V mA	RIN CIN RO VO CMIR IO

Min/max ratings are based on product characterization and simulation. Individual parameters are tested as noted. Outgoing quality levels are determined from tested parameters.

Absolute Maximum Ratings

Miscellaneous Ratings

V _{cc}	±7V
output is short circuit protected	to
ground, but maximum reliability	
maintained if lout does not excee	ed 60mA
maintained in lout does not excee	5u 00111A
common mode input voltage	$\pm V_{cc}$
differential input voltage	10Ŭ
junction temperature	+ 150°C
operating temperature range	
, A.I:	- 40°C to + 85°C
, , , , , , , , , , , , , , , , , , , ,	05:04- 450:0
storage temperature range	– 65°C to + 150°C
lead solder duration (+ 300°C)	10 sec
ESD rating (human body model)	500V

Reliability Information	
Transistor count	28

+ 2 to + 21, - 1 to - 20 recommended gain range:

NOTES:

2

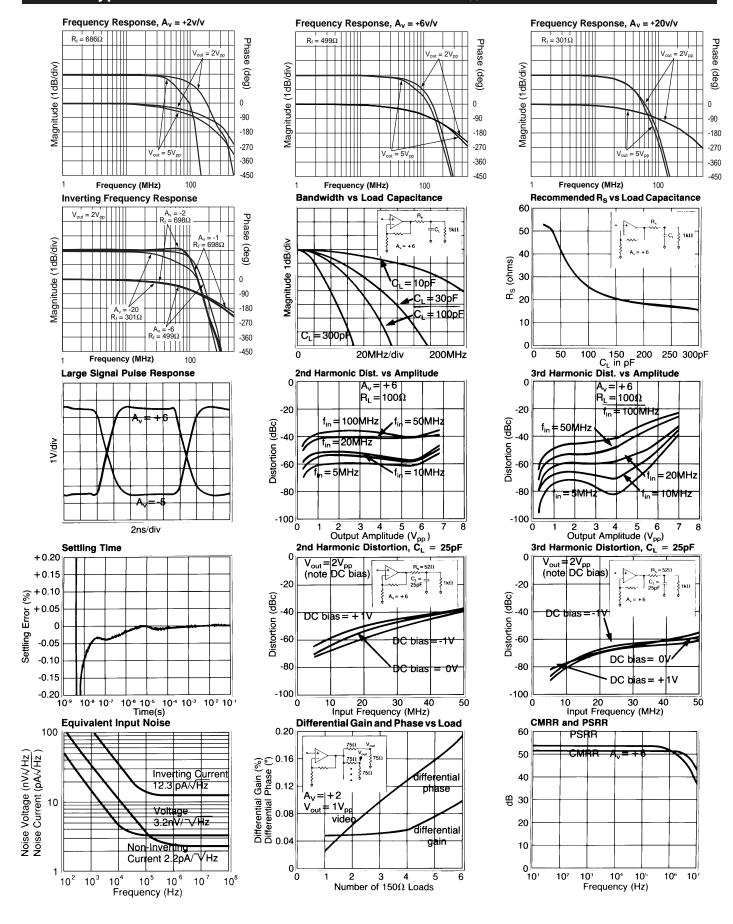
100% tested at + 25°C, sample at + 85°C.

note 1: note 2:

See the text on the back page of the datasheet. Differential gain and phase measured at A $_{\rm V}$ + 2, R $_{\rm f}$ 500 Ω , R $_{\rm L}$ 150 Ω 1V $_{\rm pp}$ equivalent video signal, 0-100 IRE, 40 IRE $_{\rm pp}$, 0IRE = 0 volts, at 75 Ω load and 3.58MHz. See text.

Package Thermal Resistance					
Package	θ_{JC}	θ_{JA}			
AJP	65°C/W	120°C/W			
AJE	60°C/W	140°C/W			

CLC404 Typical Performance Characteristics (T_A = 25°, A_V = +6, V_{CC} = ±5V, R_L = 100Ω, R_f = 500Ω; unless specified)



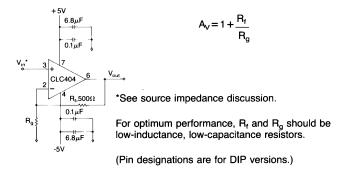


Figure 1: recommended non-inverting gain circuit

Slew Rate

Slew rate limiting is a nonlinear response which occurs in amplifiers when the output voltage swing approaches hard, abrupt limits in the speed at which it can change. In most applications, this results in an easily identifiable "slew rate" as well as a dramatic increase in distortion for large signal levels. The CLC404 has been designed to provide enough slew rate to avoid slew rate limiting in almost all circuit configurations. The large signal bandwidth of 165MHz, therefore, is nearly the same as the 175MHz small signal bandwidth. The result is a low-distortion, linear system for both small signals and large signals.

Slew rate and large signal performance in the CLC404 can best be understood by first comparing the small and large signal performance plots at a gain of +6. In the CLC404, there is almost no difference between large and small signal performance at this gain. Large signal performance in the CLC404 at a gain of +6 is not slew rate limited. (In an amplifier which is slew limiting, the large signal response rolloff has an abrupt break indicating the onset of slew rate limitation.)

The CLC404 reaches slew rate limits only for low non-inverting gains. In other words, slew rate limiting is constrained by common mode voltage swings at the input. (This is different from traditional slew rate constraints.) The large-signal frequency response plot at a gain of +2 shows a break in the response, which shows that slew rate limit has been reached. Note also that the frequency response plots at gain of +21 show that the large signal and small signal responses are nearly identical.

Differential Gain and Phase

Differential gain and phase are measurements useful primarily in composite video channels. Differential gain and phase are measured by monitoring the gain and phase of a high frequency carrier (3.58MHz for NTSC composite video) as the output of the amplifier is swept over a range of DC voltages. Any changes in gain and phase at the carrier frequency are the desired measurement, differential gain and phase.

Specifications for the CLC404 include differential gain and phase. The test signals used are based on a $1V_{pp}$ video level. Test conditions used are the following:

DC sweep range: 0 to 100 IRE units (black to white)
Carrier: 3.58MHz at 40 IRE units peak to peak

The amplifier is specified for a gain of ± 2 , and 150Ω load (for a backmatched 75Ω system). IRE amplitudes are referred to 75Ω , at the load of a video system. This is a

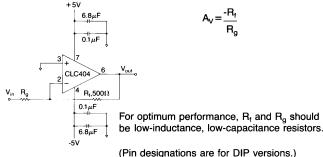


Figure 2: recommended inverting gain circuit

different condition from the rest of the specifications $(A_v = +6, R_l = 100\Omega)$.

Source Impedance

For best results, source impedance in the non-inverting circuit configuration (see Figure 1) should be kept below $3k\Omega$ Above $3k\Omega$ it is possible for oscillation to occur, depending on other circuit parasitics. Depending on the signal source, a resistor with a value of less than $3k\Omega$ may be used to terminate the non-inverting input to ground.

Feedback Resistor

In current-feedback op amps, the value of the feedback resistor plays a major role in determining amplifier dynamics. It is important to select the correct value resistor. The CLC404 provides optimum performance with a 500Ω feedback resistor. Furthermore, the specifications shown on the previous pages are valid only when a 500Ω feedback resistor is used. Selection of an incorrect value can lead to severe rolloff in frequency-response (if the resistor value is too large) or peaking or oscillation (if the value is too low).

Printed Circuit Layout

As with any high frequency device, a good PCB layout will enhance performance. Ground plane construction and good power supply bypassing close to the package are critical to achieving full performance. In the non-inverting configuration, the amplifier is sensitive to stray capacitance to ground at the inverting input. Hence, the inverting node connections should be small with minimal coupling to the ground plane. Shunt capacitance across the feedback resistor should not be used to compensate for this effect.

Parasitic or load capacitance directly on the output will introduce additional phase shift in the loop degrading the loop phase margin and leading to frequency response peaking. A small series resistor before the capacitance effectively decouples this effect. The graphs on the preceding page illustrate the required resistor value and resulting performance vs. capacitance.

Precision buffed resistors (PRP8351 series from Precision Resistive Products) with low parasitic reactances were used to develop the data sheet specifications. Precision carbon composition resistors will also yield excellent results. Standard spirally-trimmed RN55D metal film resistors will work with a slight decrease in bandwidth due to their reactive nature at high frequencies.

Evaluation PC boards (part numbers CLC730013 for through-hole and CLC 730027 for SOIC) for the CLC404 are available.

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National Semiconductor Corporation

1111 West Bardin Road Arlington, TX 76017 Tel: 1(800) 272-9959 Fax: 1(800) 737-7018

National Semiconductor Europe

Fax: (+49) 0-180-530 85 86 E-mail: europe.support.nsc.com Deutsch Tel: (+49) 0-180-530 85 85 English Tel: (+49) 0-180-532 78 32 Francais Tel: (+49) 0-180-532 93 58 Italiano Tel: (+49) 0-180-534 16 80

National Semiconductor Hong Kong Ltd.

2501 Miramar Tower 1-23 Kimberley Road Tsimshatsui, Kowloon Hong Kong Tel: (852) 2737-1600

Fax: (852) 2736-9960

National Semiconductor Japan Ltd.

Tel: 81-043-299-2309 Fax: 81-043-299-2408

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