

Class: Robust Multivariate Control

Professor: Dr. Sean Humbert

TAs: Santosh Chaganti

Student: Steve Gillet

Date: April 24, 2025

Assignment: Homework 8

1.

In this problem you will analyze the robustness properties of a F-16 equipped with a lateral regulator. We consider the lateral dynamics (β, ϕ, p, r) augmented with aileron and rudder actuator dynamics (δ_a, δ_r) and a washout filter (x_w) in the yaw channel. The state vector of the augmented dynamics is $(\beta, \phi, p, r, \delta_a, \delta_r, x_w)$. The outputs of the model include (r_w, p, β, ϕ) where r_w is the washed out yaw rate. Assume a static controller $u = Ke$ where

$$K = \begin{bmatrix} -0.56 & -0.44 & -0.11 & -0.35 \\ -1.19 & -0.21 & -0.44 & 0.26 \end{bmatrix}.$$

An m-file with the augmented system A, B, C, D matrices and controller gains K is available for download on the course website. For the analysis, assume that the plant is subject to an unstructured inverse multiplicative output uncertainty $\tilde{G} = (I - W_o\Delta)^{-1}G$ and the desired performance metric is attenuation of measurement noise n at the output y , as described.

(a)

Draw a block diagram of the overall system including the weighting functions described below and label w, z, v, u, y_Δ and u_Δ .

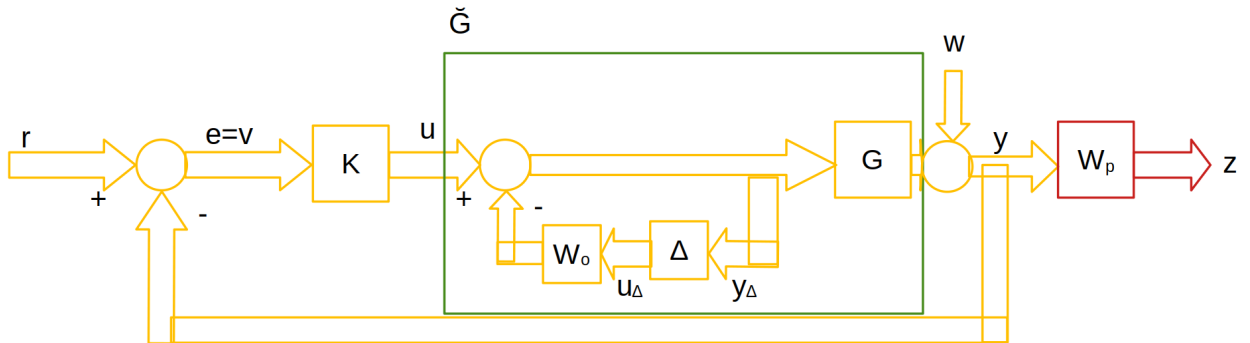


Figure 1: Block diagram of the overall system with labeled components.

(b)

Compute the generalized plant P and close the lower LFT (with $\Delta \neq 0$) to compute N . What are the corresponding tests for nominal performance and robust stability?

Writing y_Δ , v , z in terms of u_Δ , w , u :

$$\begin{aligned}y_\Delta &= u - W_o u_\Delta \\z &= W_p(w + G(u - W_o u_\Delta)) \\v &= -y \\&= -w - G(u - W_o u_\Delta)\end{aligned}$$

$$\begin{bmatrix} y_\Delta \\ z \\ v \end{bmatrix} = \begin{bmatrix} -W_o & 0 & I \\ -W_p G W_o & W_p & W_p G \\ G W_o & -I & -G \end{bmatrix} \begin{bmatrix} u_\Delta \\ w \\ u \end{bmatrix}$$

Closing lower LFT:

$$\begin{aligned}N = F_l(P, K) &= P_{11} + P_{12}K(I - P_{22}K)^{-1}P_{21} \\&= \begin{bmatrix} -W_o & 0 \\ -W_p G W_o & W_p \end{bmatrix} + \begin{bmatrix} I \\ W_p G \end{bmatrix} K(I + GK)^{-1} [G W_o \quad -I] \\&= \begin{bmatrix} -W_o + K(I + GK)^{-1}G W_o & -K(I + GK)^{-1} \\ -W_p G W_o + W_p G K(I + GK)^{-1}G W_o & W_p - W_p G K(I + GK)^{-1} \end{bmatrix} \\&= \begin{bmatrix} S_I W_o & -K S_O \\ -W_p S_O G W_o & W_p T_O \end{bmatrix}\end{aligned}$$

The test for robust stability is comes from the N_{11} block so $S_I W_o$. The test for nominal performance comes from the N_{22} block so $W_p T_O$. Need those H_∞ norms to be less than 1.

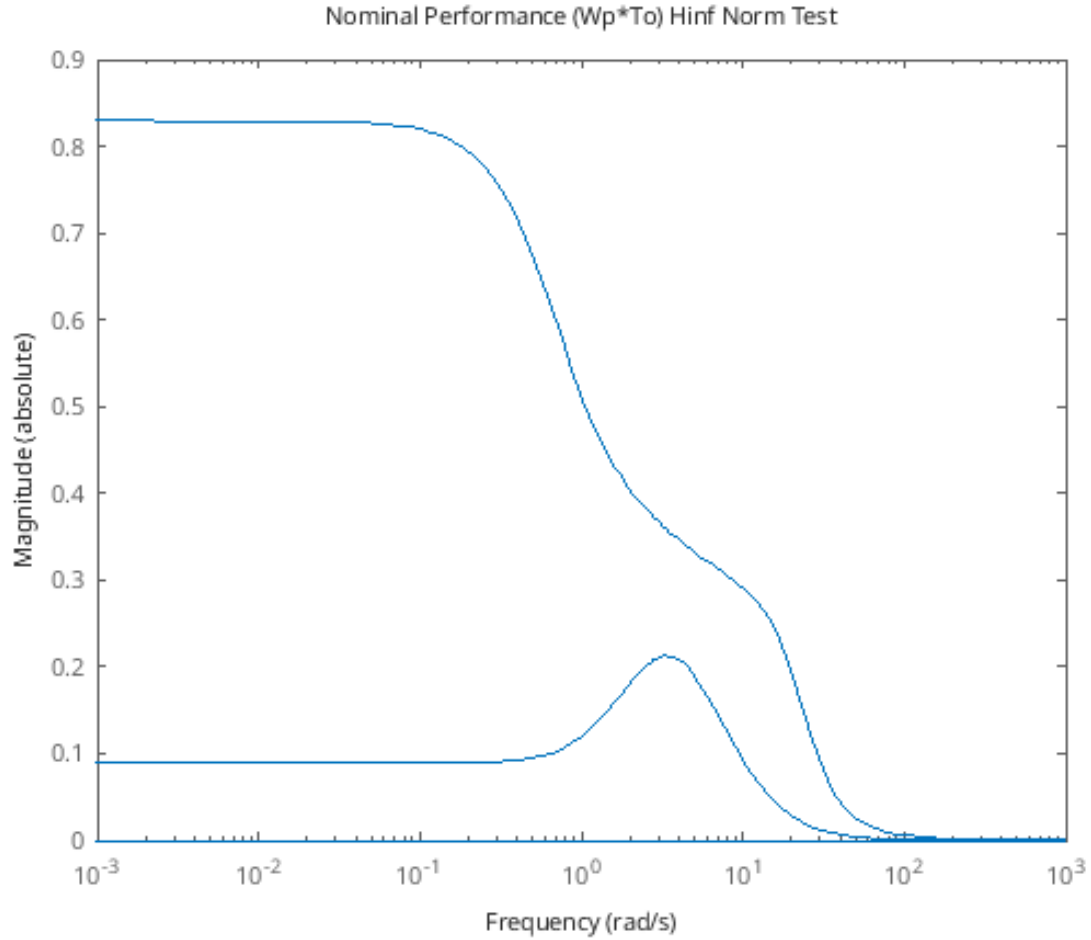
(c)

Assume a performance weighting function to be $W_p(s) = w_p(s)I$ where $w_p(s) = (s/M + \omega_B^*)/(s + \omega_B^* A)$, with $A = 4$; $M = 10$; $\omega_B^* = 3$. Does the system satisfy this nominal performance criterion you derived in part (b)?

I created N in Matlab using the weighting functions and state space models including the state space model for the controller included in the hint below and the state space model for the plant provided.

```
1 plantSS = ss(A,B,C,D);
2 controllerSS = ss(cA,cB,cC,cD);
3
4 Ti = controllerSS*plantSS*inv(eye(size(controllerSS*plantSS)) + controllerSS*plantSS);
5 To = plantSS*controllerSS*inv(eye(size(plantSS*controllerSS)) + plantSS*controllerSS);
6 So = inv(eye(size(plantSS*controllerSS)) + plantSS*controllerSS);
7 Si = inv(eye(size(controllerSS*plantSS)) + controllerSS*plantSS);
8
9 s = tf('s');
10 Wp = (s/10 + 3)/(s+3*4)*eye(4);
11 Wo = (0.02*s+0.05)/(0.02*s/0.4+1)*eye(2);
12
13 N = [Si*Wo -K*So; -Wp*So*plantSS*Wo Wp*To];
```

I then plotted the sigma plots and computed the H_∞ norm for the N_{22} block.

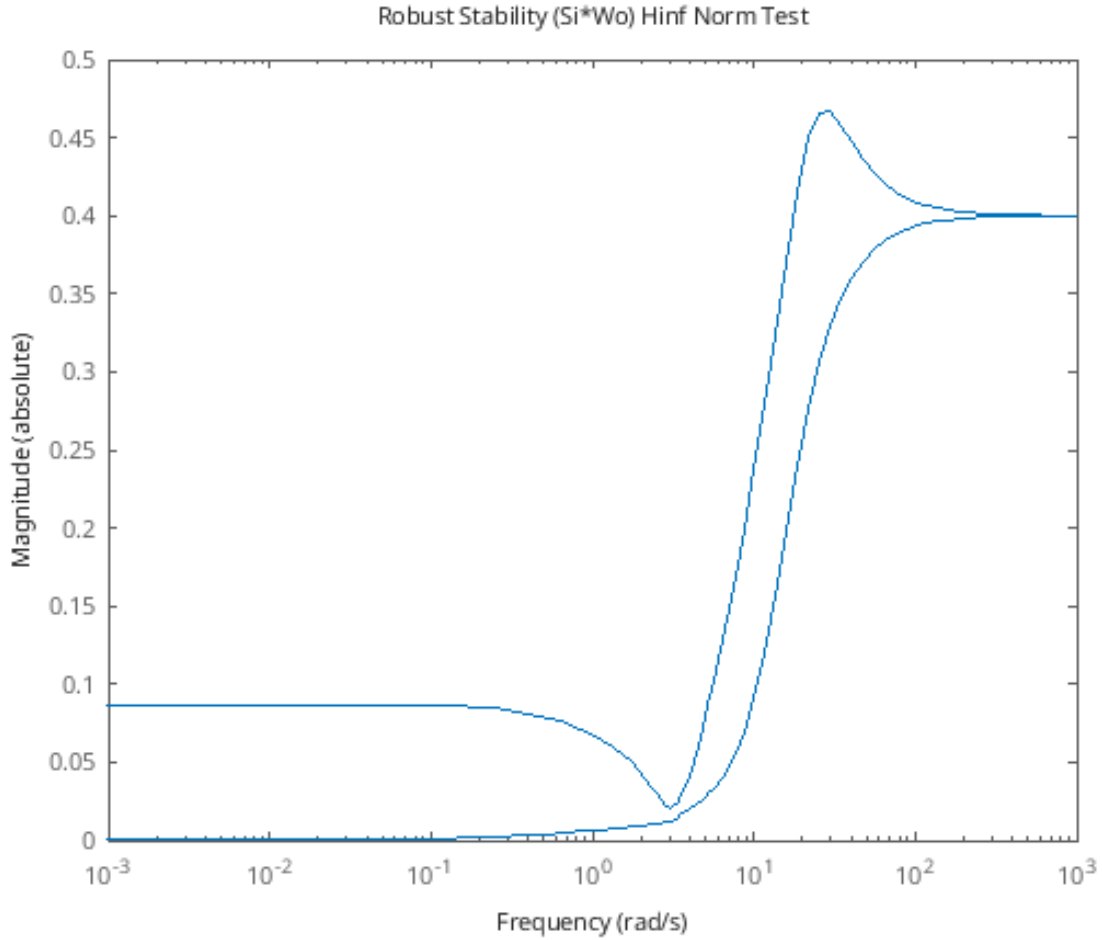


The H_∞ norm is 0.8292 which is less than 1 so the system satisfies the nominal performance criterion.

(d)

Assume an uncertainty weighting function to be $W_o(s) = w_o(s)I$ where $w_o(s) = (\tau s + r_0)/(\tau s/r_\infty + 1)$, with $\tau = 0.02$, $r_0 = 0.05$, $r_\infty = 0.4$, and full (unstructured) uncertainty. Is the system robustly stable using the criterion you derived in part (b)? Hint: to compute the state space model for the controller, assume $D = K$ (static gain) and the A, B, C matrices are all zeros with appropriate dimensions.

I used the same technique to compute and plot the robust stability for the system using $W_p T_O$ from the N_{22} block for robust stability. The H_∞ norm is 0.4679 which is less than 1 so the system satisfies the robust stability criterion.



2.

Consider a model of the longitudinal dynamics of a missile

$$\begin{aligned} \frac{d}{dt} \begin{bmatrix} \alpha \\ q \end{bmatrix} &= \begin{bmatrix} Z_\alpha & Z_q \\ M_\alpha & 0 \end{bmatrix} \begin{bmatrix} \alpha \\ q \end{bmatrix} + \begin{bmatrix} Z_{\delta_u} \\ M_{\delta_u} \end{bmatrix} \delta_u \\ n_z &= \begin{bmatrix} Z_\alpha & 0 \end{bmatrix} \begin{bmatrix} \alpha \\ q \end{bmatrix} + Z_{\delta_u} \delta_u \end{aligned}$$

where the states are the angle of attack α , the pitch rate q , n_z is the output (normal acceleration), δ_u is the tail deflection input. Assume the following variations in stability derivatives: $M_\alpha = M_\alpha^0 + \delta_\alpha$ and $Z_q = Z_q^0 + \delta_q$. Generate the analytical state space model as discussed in lecture for the extended plant N by "pulling out the deltas" for this case of structured uncertainty. Assume $w = \delta_u$, $z = n_z$ and $\Delta = \text{diag}\{\delta_\alpha, \delta_q\}$.

I started with the state space block diagram from the input to the output showing how the different states, dynamics, and uncertainties are related.

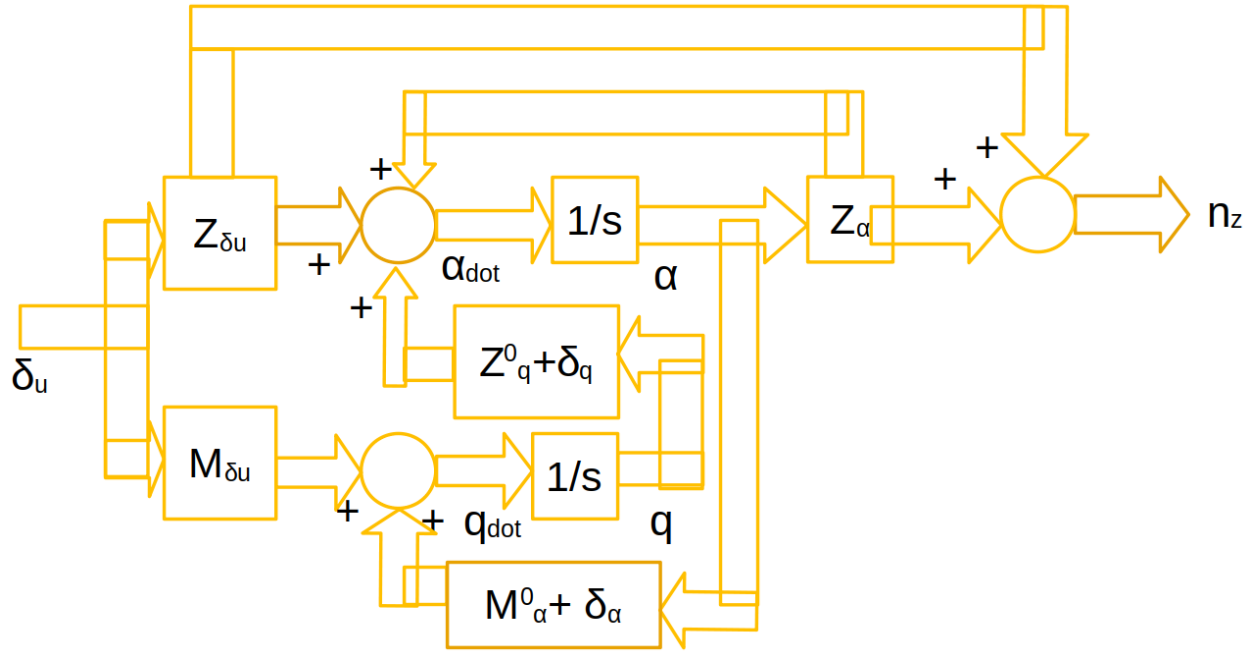


Figure 2: State Space Block Diagram of Longitudinal Missile Dynamics

Then I "pulled out the deltas" so that those uncertainties can be treated as inputs and outputs.

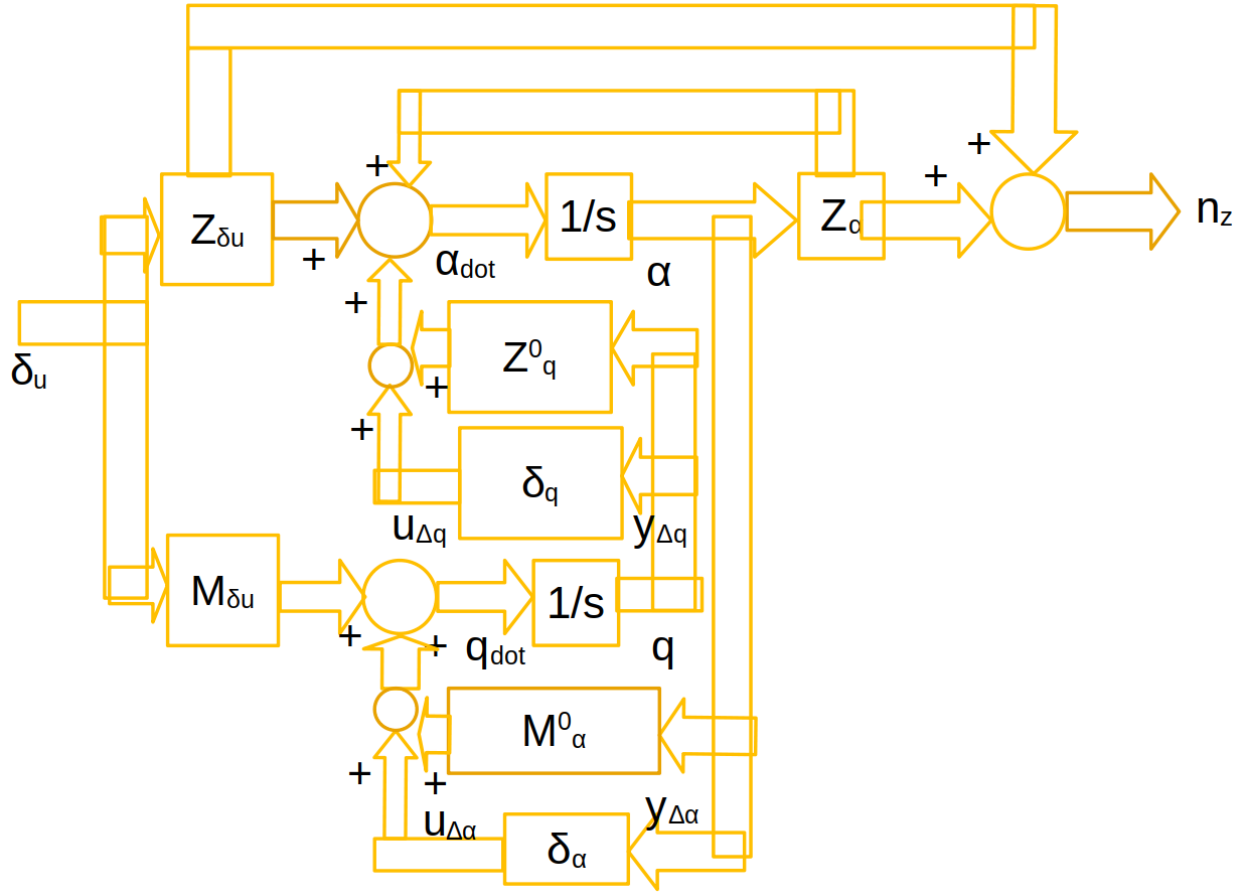


Figure 3: Block Diagram with Deltas Pulled Out

Then I wrote out the state space equations based on what I had done in the block diagrams, starting with including the u_{Δ} s and y_{Δ} s as separate inputs and outputs and then stacking them together to get the full analytical state space equation in (d) and (e).

$$\begin{aligned}
\text{(a)} \quad \begin{bmatrix} \dot{\alpha} \\ \dot{q} \end{bmatrix} &= \begin{bmatrix} Z_\alpha & Z_q^0 \\ M_\alpha^0 & 0 \end{bmatrix} \begin{bmatrix} \alpha \\ q \end{bmatrix} + \begin{bmatrix} u_{\Delta_q} \\ u_{\Delta_\alpha} \end{bmatrix} + \begin{bmatrix} Z_{\delta_u} \\ M_{\delta_u} \end{bmatrix} \delta_u \\
\text{(b)} \quad y_{\Delta_q} &= q, \quad y_{\Delta_\alpha} = \alpha \\
\text{(c)} \quad n_z &= \begin{bmatrix} Z_\alpha & 0 \end{bmatrix} \begin{bmatrix} \alpha \\ q \end{bmatrix} + Z_{\delta_u} \delta_u \\
\text{(d)} \quad \begin{bmatrix} \dot{\alpha} \\ \dot{q} \end{bmatrix} &= \begin{bmatrix} Z_\alpha & Z_q^0 \\ M_\alpha^0 & 0 \end{bmatrix} \begin{bmatrix} \alpha \\ q \end{bmatrix} + \begin{bmatrix} 1 & 0 & Z_{\delta_u} \\ 0 & 1 & M_{\delta_u} \end{bmatrix} \begin{bmatrix} u_{\Delta_q} \\ u_{\Delta_\alpha} \\ \delta_u \end{bmatrix} \\
\text{(e)} \quad \begin{bmatrix} Y_{\Delta_\alpha} \\ Y_{\Delta_q} \\ n_z \end{bmatrix} &= \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ Z_\alpha & 0 \end{bmatrix} \begin{bmatrix} \alpha \\ q \end{bmatrix} + \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & Z_\delta \end{bmatrix} \begin{bmatrix} u_{\Delta_q} \\ u_{\Delta_\alpha} \\ \delta_u \end{bmatrix} \\
\text{(f)} \quad N &= (sI - A)^{-1} B \\
&= \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ Z_\alpha & 0 \end{bmatrix} (sI - \begin{bmatrix} Z_\alpha & Z_q^0 \\ M_\alpha^0 & 0 \end{bmatrix})^{-1} \begin{bmatrix} 1 & 0 & Z_{\delta_u} \\ 0 & 1 & M_{\delta_u} \end{bmatrix} \\
&= \begin{bmatrix} 1 & 0 \\ 0 & 1 \\ Z_\alpha & 0 \end{bmatrix} \begin{bmatrix} s - Z_\alpha & s - Z_q^0 \\ s - M_\alpha^0 & s \end{bmatrix}^{-1} \begin{bmatrix} 1 & 0 & Z_{\delta_u} \\ 0 & 1 & M_{\delta_u} \end{bmatrix}
\end{aligned}$$

Then you can see where I wrote out what that Transfer Function would look like from the input to the output with the uncertainties.

3.

Consider the following lateral flight dynamics model for the A-4D aircraft:

$$\frac{d}{dt} \begin{bmatrix} \beta \\ r \end{bmatrix} = \begin{bmatrix} Y_\beta/u_0 & -(1 - Y_r/u_0) \\ N_\beta & N_r \end{bmatrix} \begin{bmatrix} \beta \\ r \end{bmatrix} + \begin{bmatrix} 0 & Y_{\delta_r}/u_0 \\ N_{\delta_a} & N_{\delta_r} \end{bmatrix} \begin{bmatrix} \delta_a \\ \delta_r \end{bmatrix}.$$

Assume the sea level reference flight condition (Condition 1) from the attached data sheet, and Mach 1 at sea level is 1126 ft/s (use this to compute u_0). Build the state space A and B matrices using the regular stability and control derivatives ($Y_\beta, N_\beta, N_r, Y_{\delta_r}, N_{\delta_a}, N_{\delta_r}$), not the 'primed' derivatives, and assume $Y_r = 0$. Also assume the C matrix is the identity so that the outputs are the two states β and r .

(a)

Using the nominal plant, synthesize a H_∞ controller using `hinfyn` with a disturbance rejection performance weighting function $W_p(s) = w_p(s)I$ where $w_p(s) = (s/M + \omega_B^*) / (s + \omega_B^* A)$, with $A = 0.005$; $M = 2$; $\omega_B^* = 1$. Does the closed loop system satisfy nominal performance? Verify with a plot of $\partial[S(j\omega)]$ and $1/|w_p(j\omega)|$.

I generated the H_∞ controller using `sysic` and `hinfyn` as shown in the code below:

```

1 Ybeta = -110.94;
2 u0 = 1126*0.4;
3 Yr = 0;
4 Nbeta = 15.16;
5 Nr = -0.639;
6 Ydeltar = 19.65;
7 Ndeltaa = 0.334;
8 Ndeltar = -6.732;
9
10 A = [Ybeta/u0 -(1-Yr/u0); Nbeta Nr];
11 B = [0 Ydeltar/u0; Ndeltaa Ndeltar];
12 C = eye(size(B,1));

```

```

13 D = 0;
14
15 G = ss(A,B,C,D);
16
17 M = 2;
18 Aweight = 0.005;
19 omegaB = 1;
20
21 s = tf('s');
22 Wp = tf([s/M+omegaB], [s+omegaB*Aweight])*eye(size(C,1));
23
24 systemnames = 'G Wp';
25 inputvar = '[u(2); w(2)]';
26 outputvar = '[Wp; -G-w]';
27 input_to_G = '[u]';
28 input_to_Wp = '[G+w]';
29 sysoutname = 'p';
30 sysic;
31 Pnominal = minreal(ss(P));
32
33 m = size(B,2);
34 p = size(C,1);
35 [K, CL, gamma, info] = hinfsyn(Pnominal,p,m,'method','ric','Tolgam',1e-3,'DISPLAY','on');
36
37 So = inv(eye(2)+G*K);
38 figure(1); clf;
39 sigma(So, 'm', inv(Wp), 'r--', {1e-3, 1e2});
40 grid on; legend('\sigma(So)', '\sigma(1/Wp)'); title('\sigma(So) vs. |1/Wp(jw)|');

```

Here is the plot of 'sigma' of So (the disturbance rejection transfer function) and $\frac{1}{w_p}$ (the weighting function).

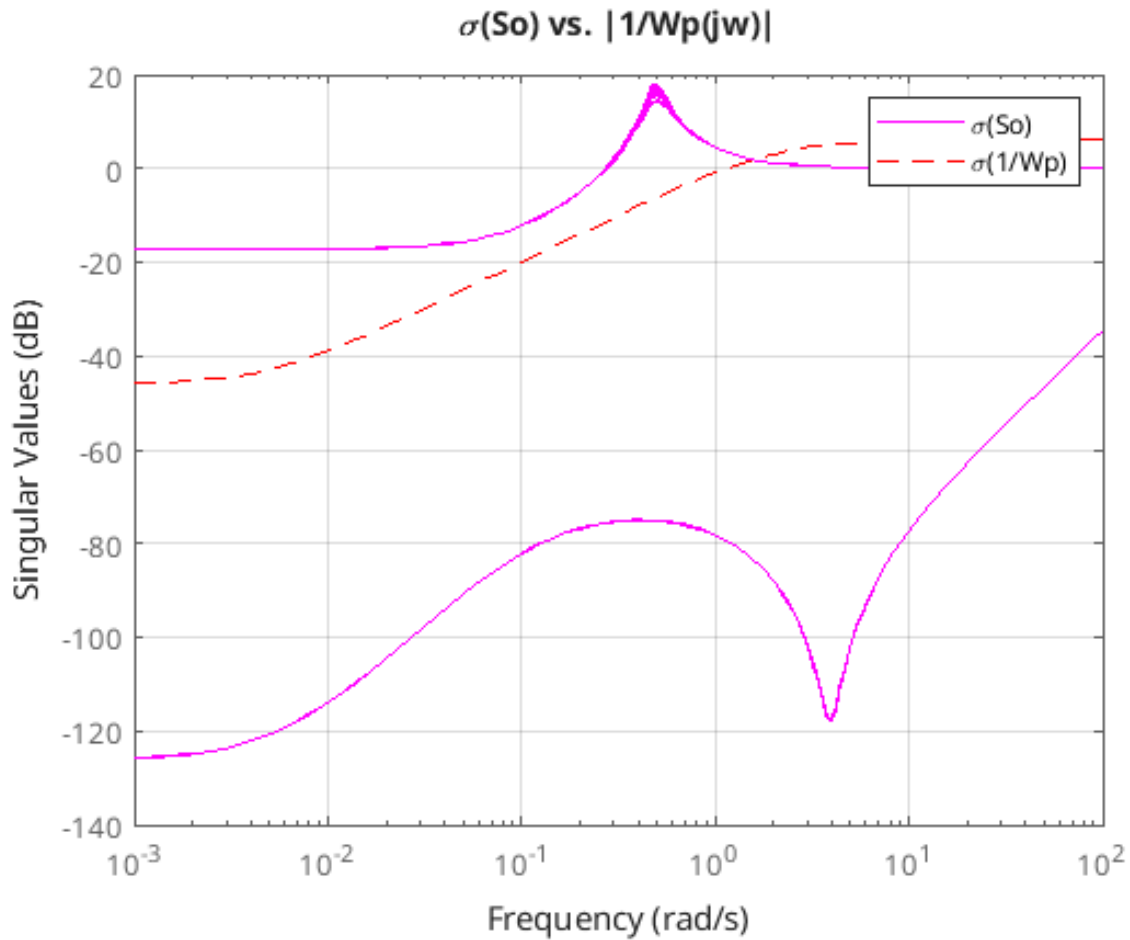


Figure 4: Nominal Performance Plot

The max singular value is not less than the weighting function for all frequencies so I don't think it does satisfy nominal performance, leading me to suspect some misstep in the process.

(b)

Now allow the parameter Y_β to have 20% real uncertainty and compute $\mu(N_{11})$. Use the MATLAB command `mussv` and generate the N and Δ blocks using the commands `lft` and `lftdata` as was done in lecture. Does the controller you designed provide robust stability for the perturbation in Y_β ?

I generated the uncertain Y_β using the 'ureal' function and then 'lft' and 'lftdata' to generate N and Δ then grabbed N_{11} to test robust stability.

```
Ybeta = ureal('Ybeta', Ybeta, 'Percentage', 20);
```

```
...
```

```
CL = lft(P,K);
[N, Delta, Blkstruct] = lftdata(CL);
szDelta = size(Delta);
N11 = N(1:szDelta(2), 1:szDelta(1));
```

```

omega = logspace(-3,2);
N11frd = frd(N11, omega);
mu = mussv(N11frd, Blkstruct);

bodeOpt = bodeoptions;
bodeOpt.PhaseVisible = 'off';
bodeOpt.XLim = [1e-1 1e1];
bodeOpt.MagUnits = 'abs';
figure(2); clf;
bodeplot(mu(1,1), 'bo', mu(1,2), 'r-', bodeOpt);
grid on;
xlabel('Frequency (rad/s)');
ylabel('Mu(N11) Upper and Lower Bounds');
title('Robust Stability Plot');

```

Then I used mussv and bodeplot to plot the upper and lower bounds of the $\mu(N_{11})$ function.

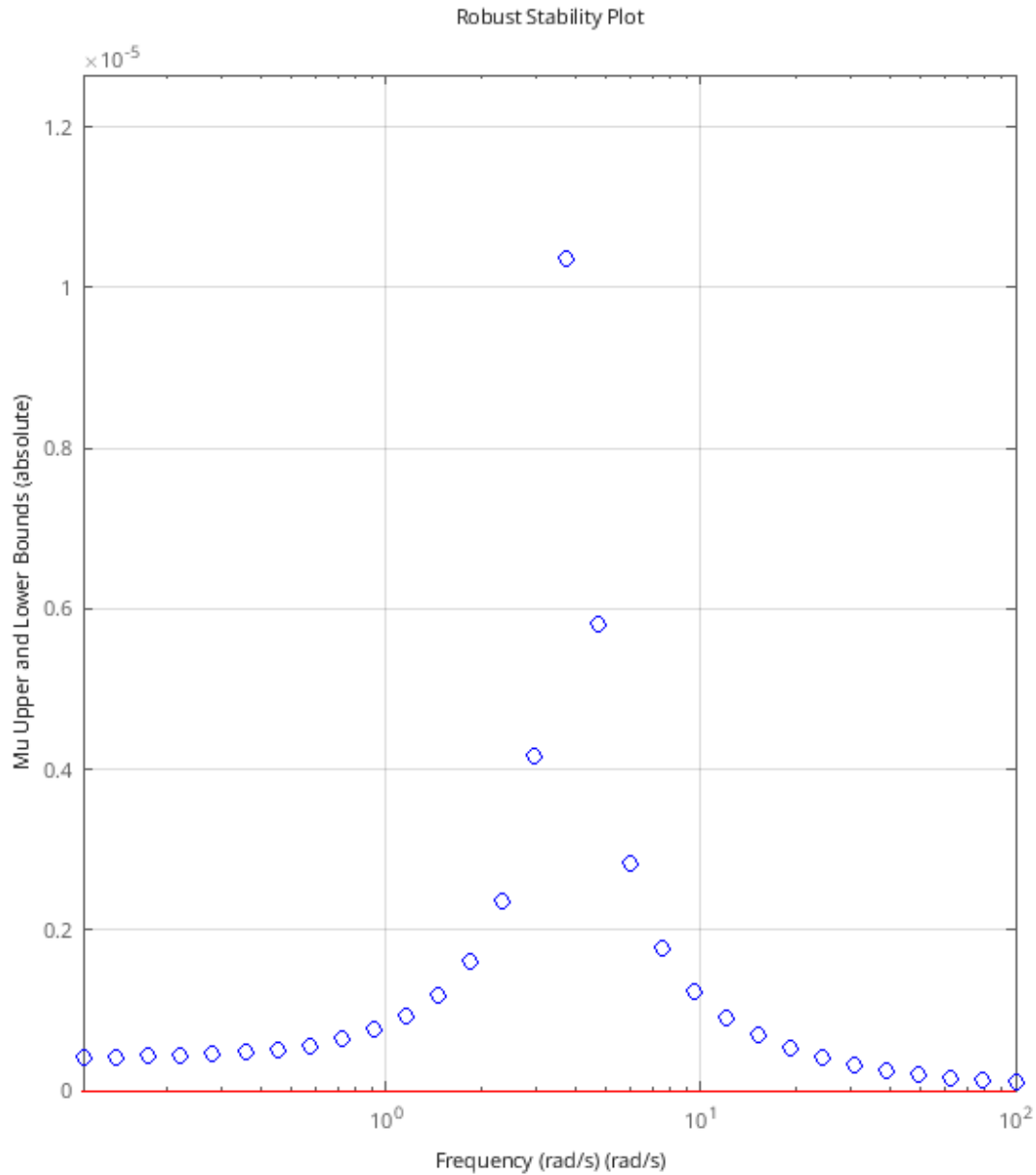


Figure 5: Robust Stability Plot

The controller does appear to satisfy robust stability by being less than 1 for all frequencies.

(c)

Now compute the robust performance test $\mu(N)$ using the appropriate unstructured Δ_p and structured Δ blocks as was discussed in lecture. Does the controller you designed provide robust performance for the perturbation in Y_β ?

For Robust Performance, I used a similar process except with all of N and a structured and unstructured uncertainty block structure.

```

1 Nfrd = frd(N, omega);
2 Blkstruct = [1 0; 2 2];
3 muN = mussv(Nfrd, Blkstruct);
4 figure(3); clf;
5 bodeplot(muN(1,1), 'bo', muN(1,2), 'r--', bodeOpt);
6 grid on;
7 xlabel('Frequency (rad/s)');
8 ylabel('Mu(N) Upper and Lower Bounds');
9 title('Robust Performance Plot');

```

Here is the plot:

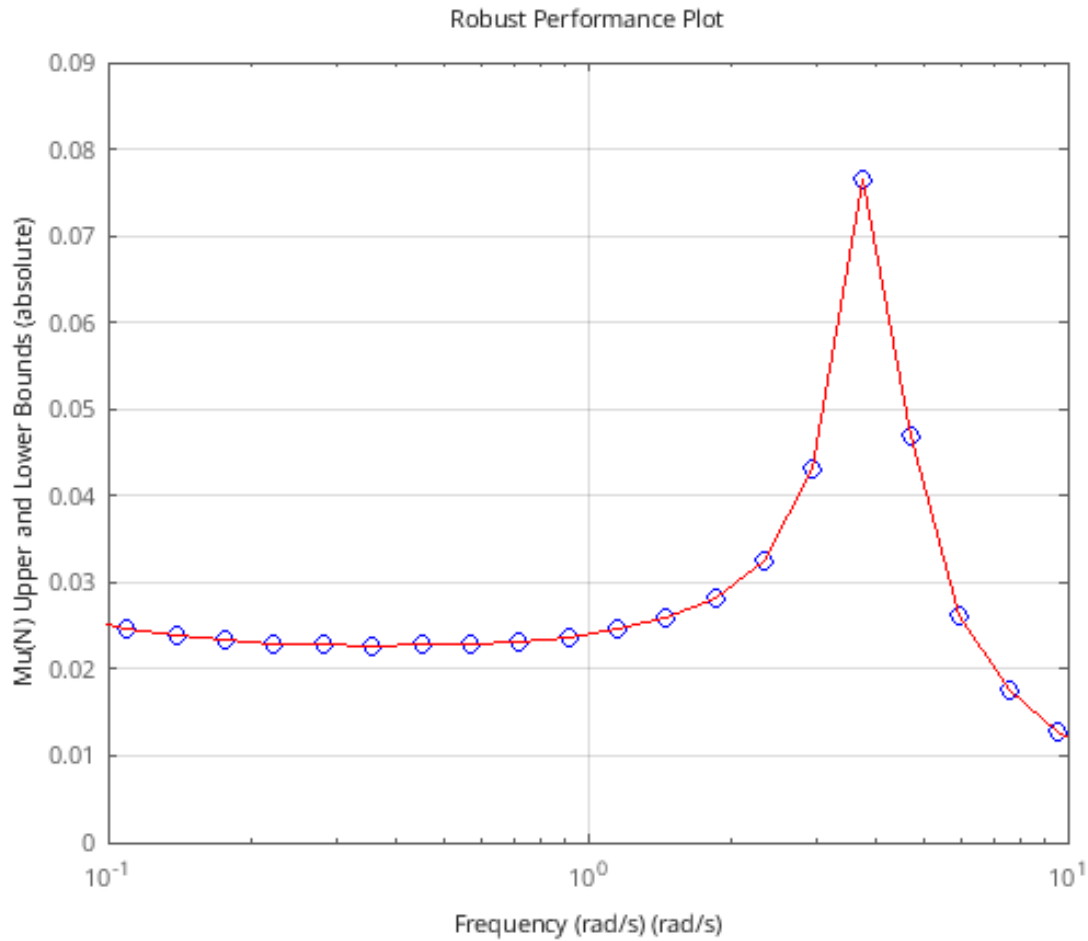


Figure 6: Robust Performance Plot

Robust performance appears to be satisfied since the bounds remain below 1 for all frequencies.

(d)

Now add an additional perturbation of 30% real uncertainty of the parameter N_β and check to see if your controller provides robust stability and robust performance for this uncertainty case.

Lastly I added uncertainty to N_β using the same process as for Y_β and then generated all of the plots again.

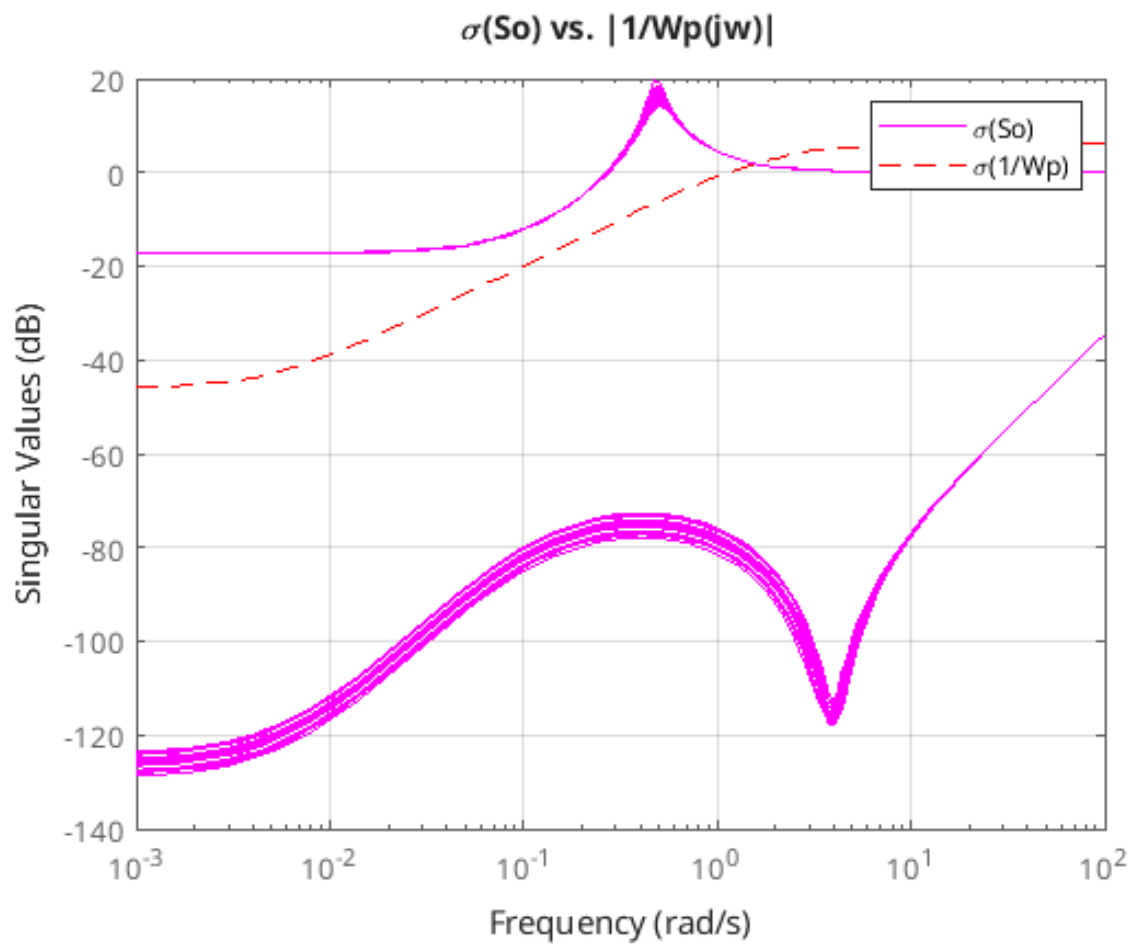


Figure 7: Nominal Performance Plot

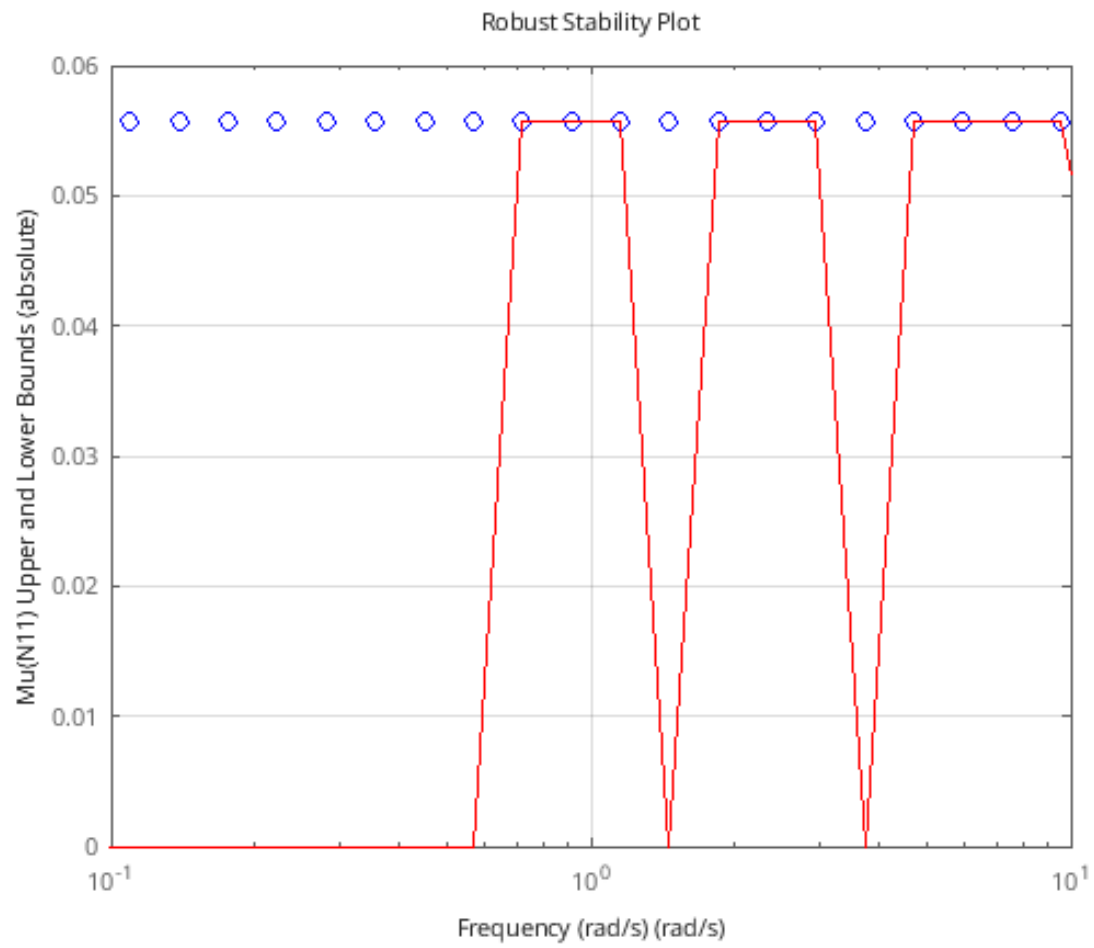


Figure 8: Robust Stability Plot

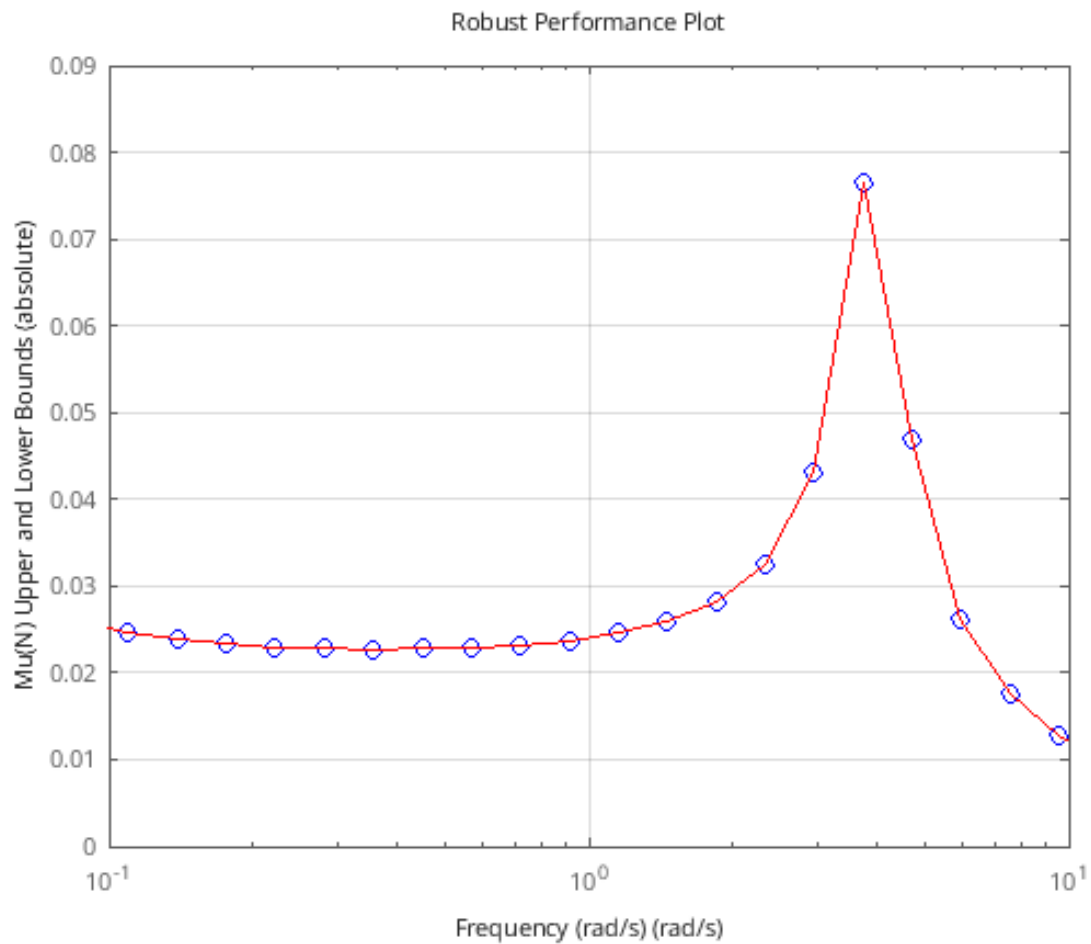


Figure 9: Robust Performance Plot

Again all of the tests appear to be satisfied except for nominal performance, which implies that none of them are actually satisfied.