1

Control Systems

G V V Sharma*

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Abstract—The objective of this manual is to introduce control system design at an elementary level.

Download python codes using

svn co https://github.com/gadepall/school/trunk/ control/ketan/codes

1 Frequency Response Analysis

- 1.1 Polar Plot
- 1.2 Direct and Inverse Polar Plot
- 1.3 Bode Plot
- 1.3.1. Sketch the bode magnitude and phase plots for the closed loop (negative feedback) system given by:

$$G(s) = \frac{100(s+2)(s+4)}{s^2 - 3s + 10}$$
 (1.3.0.1)

$$H(s) = \frac{1}{s} \tag{1.3.0.2}$$

*The author is with the Department of Electrical Engineering, Indian Institute of Technology, Hyderabad 502285 India e-mail: gadepall@iith.ac.in. All content in this manual is released under GNU GPL. Free and open source.

Solution: The closed loop transfer function of the system is given by:

$$G_m(s) = \frac{Y(s)}{X(s)} = \frac{G(s)}{1 + G(s) \cdot H(s)}$$

$$(1.3.0.3)$$

$$\implies G_m(s) = \frac{100s(s+2)(s+4)}{s^3 + 97s^2 + 610s + 800}$$

$$(1.3.0.4)$$

Evaluate at $s = j\omega$:

$$G_m(j\omega) = \frac{100j\omega(j\omega + 2)(j\omega + 4)}{(j\omega)^3 + 97(j\omega)^2 + 610(j\omega) + 800}$$

$$= \frac{-600\omega^2 + j(800\omega - 100\omega^3)}{800 - 97\omega^2 + j(610\omega - \omega^3)}$$
(1.3.0.6)

From (1.3.0.6):

$$|G_m(j\omega)| = \frac{\sqrt{(600\omega^2)^2 + (800\omega - 100\omega^3)^2}}{\sqrt{(800 - 97\omega^2)^2 + (610\omega - \omega^3)^2}}$$

$$(1.3.0.7)$$

$$\frac{G_m(j\omega) = \tan^{-1}\left(\frac{\omega^2 - 8}{6\omega}\right) - \tan^{-1}\left(\frac{610\omega - \omega^3}{800 - 97\omega^2}\right)$$

$$(1.3.0.8)$$

The following code plots the bode magnitude and phase plots in Fig. 1.3.1:

1.3.2. Compute the gain margin of the system. **Solution:**

$$G(j\omega)H(j\omega) = \left(\frac{100(j\omega + 2)(j\omega + 4)}{(j\omega)^2 - 3j\omega + 10}\right)\left(\frac{1}{j\omega}\right)$$

$$(1.3.0.9)$$

$$= \frac{100(-\omega^2 + 8 + j6\omega)}{3\omega^2 + j(10\omega - \omega^3)}$$

$$(1.3.0.10)$$

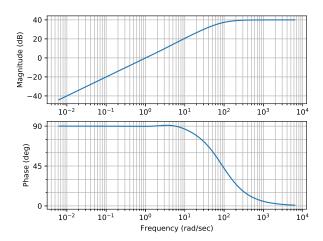


Fig. 1.3.1: Bode plot for $G_m(j\omega)$

Using (1.3.0.10)

$$|G(j\omega)H(j\omega)| = \frac{100\sqrt{(8-\omega^2)^2 + (6\omega)^2}}{\sqrt{(3\omega^2)^2 + (10\omega - \omega^3)^2}}$$
(1.3.0.11)

$$\underline{\langle G(j\omega)H(j\omega) = \tan^{-1}\left(\frac{6\omega}{8-\omega^2}\right)} - \tan^{-1}\left(\frac{10-\omega^2}{3\omega}\right) \quad (1.3.0.12)$$

At the phase crossover frequency ω_{pc} :

$$\left| \underline{/G} (j\omega) H(j\omega) \right| = 180 \qquad (1.3.0.13)$$

$$\implies \tan^{-1}\left(\frac{6\omega_{pc}}{8-\omega_{pc}^2}\right) - \tan^{-1}\left(\frac{10-\omega_{pc}^2}{3\omega_{pc}}\right) = 180$$
(1.3.0.14)

Solving the above equation:

$$\frac{6\omega_{pc}}{8 - \omega_{pc}^2} = \frac{10 - \omega_{pc}^2}{3\omega_{pc}}$$
 (1.3.0.15)

$$\implies \omega_{pc} = 5.8 rad/sec$$
 (1.3.0.16)

$$|G(j\omega)H(j\omega)|_{\omega=\omega_p c} = 26.9dB \quad (1.3.0.17)$$

Gain Margin (GM:

$$GM = 0 - \left| G(j\omega) H(j\omega) \right|_{\omega = \omega_p c} dB$$

$$= -26.9dB \qquad (1.3.0.18)$$

1.3.3. Compute the phase margin of the system. **Solution:**

At the gain crossover frequency ω_{gc} :

$$\left| G(j\omega) H(j\omega) \right|_{\omega = \omega_{nc}} = 1 \qquad (1.3.0.20)$$

From (1.3.0.11), At ω_{gc} :

$$10^{4} \left(\left(8 - \omega^{2} \right)^{2} + (6\omega)^{2} \right) = \left(3\omega^{2} \right)^{2} + \left(10\omega - \omega^{3} \right)^{2}$$
(1.3.0.21)

$$\implies \omega_{gc} = 100.15 rad/sec$$
 (1.3.0.22)

Substitute ω_{gc} in (1.3.0.12):

$$\underline{/G}(J\omega)H(J\omega)_{\omega=\omega_{gc}} = 265^{\circ}$$
 (1.3.0.23)

Phase Margin *PM*:

$$PM = 180^{\circ} - \underline{/G} (j\omega) H (j\omega)_{\omega = \omega_{gc}}$$

$$(1.3.0.24)$$

$$= 180^{\circ} - 265^{\circ} = -85^{\circ}$$

$$(1.3.0.25)$$

1.3.4. Verify the values of Gain Margin and Phase Margin using a python plot.

Solution:

The following code is used to verify the gain and phase margins:

1.3.5. Comment on the stability of the system **Solution:**

As both the Gain Margin (GM) and Phase Margin (PM) are found to be negative, the system is unstable.

- 2 Stability in Frequency Domain
- 2.1 Nyquist Criterion
 - 3 Design in Frequency Domain
 - 4 PID Controller Design
- 4.1 Introduction

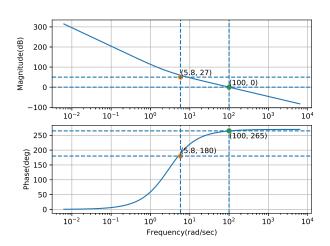


Fig. 1.3.2: Bode plot for $G(j\omega)H(j\omega)$