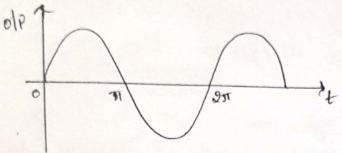


POWER AMPLISFIERS

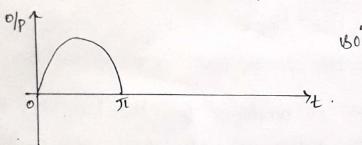
Amplifier classes (based on conduction angle).

class-A: o/p transiator brased at quircent current Ices and conducts for entire cycle of i/p rignal.

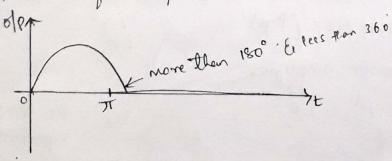


If the Apt and the 1/p cryan are schooled such that the ofp cryal is obtained for a full 1/p cycle.

class-B:- ofp transistor conducts for only half of each sine wave i/p cycle.



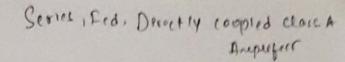
cless-AB: of p transistor is biased at a small a wiercent current and conducts for slightly more I than half a cycle.

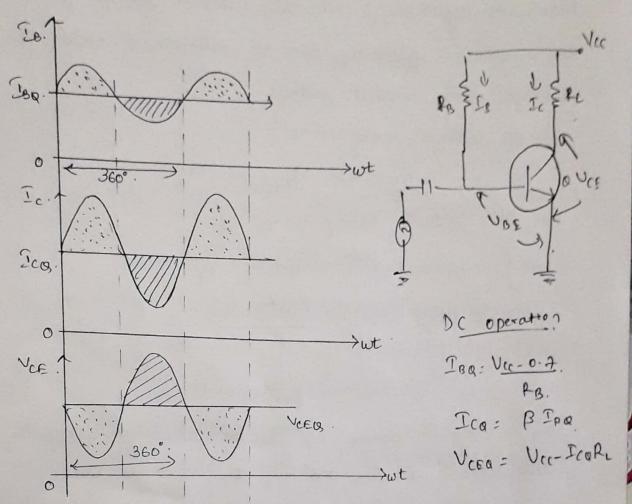


class-C: ofp transiplor conducts slightly lessthan too!

rolf a i/p cycle. ofp. | Lessthan 180°.

* The main feature of a large signal amplifier the oxt's power efficiency and impedance matching the ofp device. · Amplifier efficiency: Power efficiency of an amplifier is defined as the ratio of power ofp to the power i/p, improves going from class A to D. 1. efficiency (1) = Pout x 100. But Pin = Pout + Planser. 1. (n) = Pout ×100. Pout + Proposes Plones - is the power lost in the power transistor * max. 2 of class A amplifier for the largest ofp Voltage and current swing is only 05%. with a direct or series fed load and 50%. With a transformer Connection to the load. Class B' -> 78.5%. , class AB = 50% to 765 Class-16 → over 901. (High) => Class-A of power amplifier Telma) 100d tood tig: Graphical ÎB5. representation of Class-A. amplifie - IBA





Power Considerations

I/p power is given by,

The ofp power delivered to the load (R.) is given by

Efféciency (2): It prepresents the amount of ac Power delivered from the d.c. source.

Maximum efficiency: For the class. A Beriefs feel amplified the maximum efficiency can be determined using the maximum voltage and current swings.

For the voltage swing it is,

For the current swing it is,

max Ic(pp) = Vec Rc.

max $P_{o}(ac) = \frac{V_{ce}(V_{ce}/R_{c})}{8}$ $= \frac{V_{ce}^{2}}{8R_{c}}$

the maximum power 1/p can be calculated using the dc bias current set to one-half the maximum value.

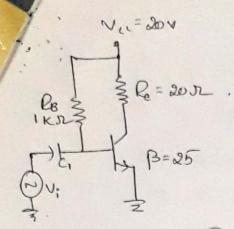
maximum Pi(dc) = Vcc (maximum Ic) = Vcc (Vcc/Ra)

 $\frac{P_{i}(dc)^{2}}{2Rc} = \frac{N_{cc}^{2}}{2Rc}$ $\frac{P_{o}(ac)_{max}}{P_{i}(dc)_{max}} \times 100^{\circ} / .$

= \frac{\fir}{\fint}}}}}}{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\fir}{\fint}}}}}}{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\fir}{\fir}}}}}}{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac{\frac

= 25.1.

1) calculate the input power, output power and efficiency of the amplifier.



$$\frac{1}{R_B} = \frac{V_{cc} - 0.7V}{R_B} = \frac{90V - 0.7V}{1 \text{ KR}} = 19.3 \text{ mA}.$$

$$\frac{1}{C_B} = \frac{1}{B} \frac{1}{B_B} = \frac{25(19.3 \text{ mA})}{1 \text{ KR}} = 482.5 \text{ mA} \pm 10.48 \text{ A}.$$

$$\frac{1}{C_B} = \frac{1}{V_{cc}} \frac{1}{C_C} = \frac{1}{2} \frac{1}{V_{cc}} = \frac{1000 \text{ mA}}{1000 \text{ mA}} = \frac{10.4 \text{ V}}{1000 \text{ mA}}$$

$$\frac{1}{C_C} = \frac{1}{V_{cc}} \frac{1}{C_C} = \frac{1000 \text{ mA}}{R_c} = \frac{10.4 \text{ V}}{R_c}$$

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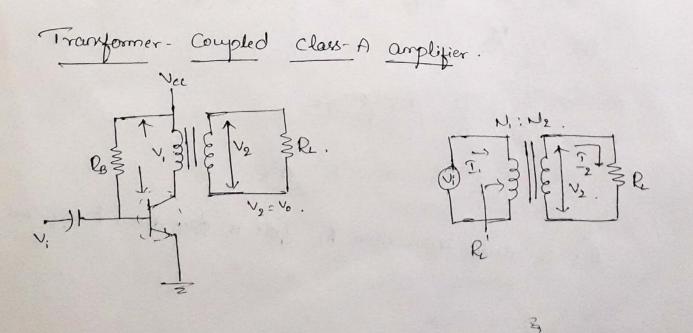
$$\frac{1}{C_C} = \frac{1}{V_{cc}} \frac{1}{C_C} = \frac{1000 \text{ mA}}{R_c} = \frac{10.4 \text{ V}}{R_c}$$

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$$\frac{1}{C_C} = \frac{10.4 \text{ V}}{R_c} = \frac{10.4 \text{ V}}{R_c}$$

$$\frac{1}{C_C} = \frac{10.4$$



Ni- no of turns on primary.

Nz- no of turns on secondary.

Vi- vtg applied to primary.

Vz- vtg on secondary.

Voltage transformation :-

$$\frac{V_2}{V_i} = \frac{N_2}{N_i}$$

current transformation:

$$\frac{\mathcal{I}_2}{\mathcal{I}_1} = \frac{N_1}{N_2}$$

Impedance matching:

Re is connected across the transformer secondary. This impedance is changed by the transformer when viewed at the primary side (Ri). Ri is reflected impedance at the primary side:

$$\frac{R_{L}}{R_{L}} = \frac{R_{2}}{R_{L}} = \frac{V_{2}/T_{2}}{V_{1}/T_{1}} = \frac{V_{2}}{V_{1}} \cdot \frac{T_{1}}{T_{2}} = \frac{N_{2}}{N_{2}} \cdot \frac{N_{2}}{N_{2}} = \left(\frac{N_{2}}{N_{1}}\right)^{2}.$$
Let $\frac{N_{2}}{N_{1}} = \sqrt{\alpha}$.

$$\Rightarrow \frac{R_i}{R_i} = \frac{R_i}{R_i} = \left(\frac{N_i}{N_i}\right)^2 = a^2$$

$$\Rightarrow$$
 $R_1 = \alpha^2 R_2$ or $R_1' = \alpha^2 R_1$

Pie es always hegher than Fe, for a step down transformed

No - 20 of turns on primary

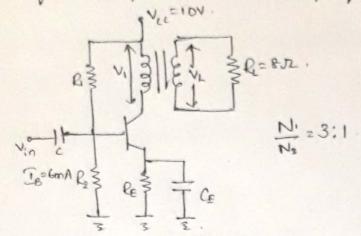
No - 20 of turns on recordary.

Vo - vig applied to primary

No - vig on recordary.

$$P_{ac} = \frac{V_1^2 rms}{R_L} = \frac{V_2^2 m}{2R_L}$$

1) For the cxt shown below, calculate the dc input power Power descipated by the transistor & efficiency of the for the input current moing of 4 mA.



$$P_{0}(dc) = V_{cc}I_{cg} = 10v \times 140xmA = 14W$$

$$P_{0} = P_{in}(dc) - P_{0}(ac)$$

$$P_{0}(oc) = \left(V_{CEmax} - V_{CEmin}\right)\left(I_{Cmax} - I_{Cmin}\right)$$

$$8$$

$$= \left(18 \cdot 3 - 1 \cdot 7V\right)\left(255 \text{ mA} - 25 \text{ mA}\right)$$

$$P_{0}(ac) = 0.477W$$

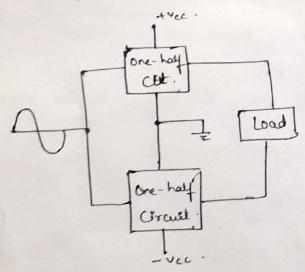
To get a full even ocross the load, a pair of transestors is used in a acc B operation. The two transaction conduct on allerate half eyeles of the operation and full cycle across the load to obtained. The five teamisters induction celterate bay your of the super extent come of the right right & a full cycle across les load es obtained. The two transferous are edentical in characteristics and called material According ons al when both the transitors are of same type to estee n-p-ndep-n-p

then the class could as puch pull class B auplifor exp

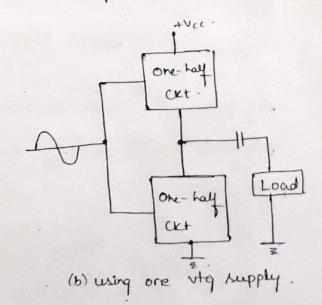
e) Complementary symmetry class B.

Jans-B Amplifier

Clau-B amplifier provider of p for the only one-half of the signal cycle. To Obtain output for the full cycle of signal, it is necessary to me two transistors and have each conduct on opposite half-cycles, the combined operation providing a full cycle of output signal. Since one part of the cet pushes the signal high during one half-cycle and the other part pulls the signal low during the other half-cycle, the cet is referred to as a push-pull cet. The below fig. show, a diagram for push-pull operation. The power transistors used in the push-pull circuit are capable of delivering the desired power to the load, and the class-B operation of these transistors provides greater efficiency than say possible using a single transistor in class-A, operation.



(a) wing two vtg supplies.



Input (DC) power:

Pilac) = Vec Tac:

where Ise if the average (or) de current drawn from power supplies. In class-B operation, the current drawn for a higher power supply has the form of a fault-wave rectified signal, whereas that drawn from two power supplies has the form of a half-wave rectified signal from each supply. In either case, the value of the average current drawn can be expressed as

$$T_{dc} = \frac{2}{\pi t} T_{(p)}.$$

$$P_{i(dc)} = v_{cc} \left(\frac{2}{\pi} T_{(p)} \right)$$

Top - Peak value of the ofp current with.

output (Ac) power:

$$P_{o}(ac) = \frac{V_{L}^{2}(rms)}{R_{L}}$$

$$P_{o}(ac) = \frac{V_{L}^{2}(p-p)}{8R_{L}} = \frac{V_{L}^{2}(p)}{2R_{L}}$$

The larger the rms (or) peak ofp vtg, the larger is the power delivered to the load.

Efficiency:

$$P_{i(dc)} = \frac{P_{o}(ac)}{P_{i(dc)}} \times 100\%.$$

$$= \frac{V_{i}(p)}{V_{cc}(\frac{3}{5} - \frac{7}{10})} \times 100\%.$$

$$= \frac{V_{i}(p)}{V_{cc}(\frac{3}{5} - \frac{7}{10})} \times 100\%.$$

The larger the peak voltage, the higher is the ext efficiency, up to a maximum value when VL(p)= Vee, this maximum efficiency then being

maximum efficiency = \frac{17}{4} x 100% = 78.5%.

· Power discipated by output transistors:

The power dissipated (as heat) by the output power transitions is the difference b/s the input power debrered by the supplies and the ofp power delivered to the load.

where Pea 1x the power dissipated by the two power transite ... The power dissipated by each transistor is,

There a class B amplifier providing a 20-v peak signal to a 16-z load (speaker) and a power supply of Vc=30v determine the input power, output power & cut efficiency.

$$I_L(p) = \frac{V_L(p)}{R_L} = \frac{20V}{16\pi} = \frac{1.25 \, \text{A}}{1}$$

$$P_0(ac) = \frac{V_L^2(p)}{2RL} = \frac{(20)^2}{2(16)} = 12.5W$$

Maximum power considerations

for class B, the maximum output power 9x delivered to the load when $V_{\nu}(p) = V_{\nu}(p)$.

max. Po(ac) =
$$\frac{V_{cc}^2}{2R_L}$$

The corresponding peak ac current Icp is

max
$$1.2 = \frac{P_0(ac)}{P_1(dc)} \times 100\%$$
 = $\frac{\frac{1}{\sqrt{L}} \sqrt{2} L}{\sqrt{100}} = \frac{\sqrt{L}}{4} \times 100\%$

D. For a class-B amplifier using a supply of $V_{cc}=300 \, \xi_{c}$ driving a load of 1602, determine the maximum 1/p power output power and transistor dissipation.

maxi output power,
$$P_{c}(ac) = \frac{V_{cc}^2}{2R_L} = \frac{30V^2}{2(16)} = 28.125 \text{ m}$$
.

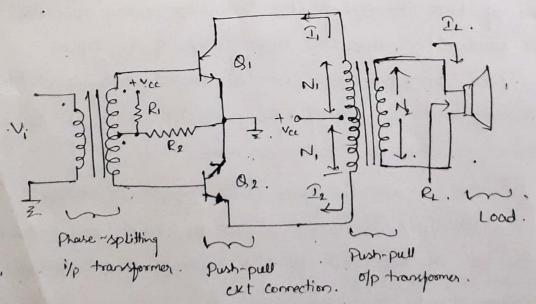
Efficiency in also given by

1.7 = 78.54. VL(P)

Class-B Amplifier Circuits

- O Transpormer-Coupled push-pull Ckt.
- 1 Complementary Symmetry Ckt.
- 3 Quari-complementary push-pull amplifier:

Transformer - coupled Push-pull circuits:



The ckt wer a centre-tapped appul transformer to produce opposite-polarity signals to the two transistor inputs and an output transformer to drive the load in a push-pull mode of operation.

During the first half-cycle of operation, transiptor gi driven in to conduction, whereas transister Bz is driven of The aurent I, through the transformer results in the first-ra cycle of signal to the load.

During the second half-cycle of the input signal, Q2 conduct where as Q1 stays off, the current Is through the transform resulting in the second half-cycle to the load. The overall signal developed across the load then varies over the full cycle of signal operation.

Complementary - Symmetry Circuits

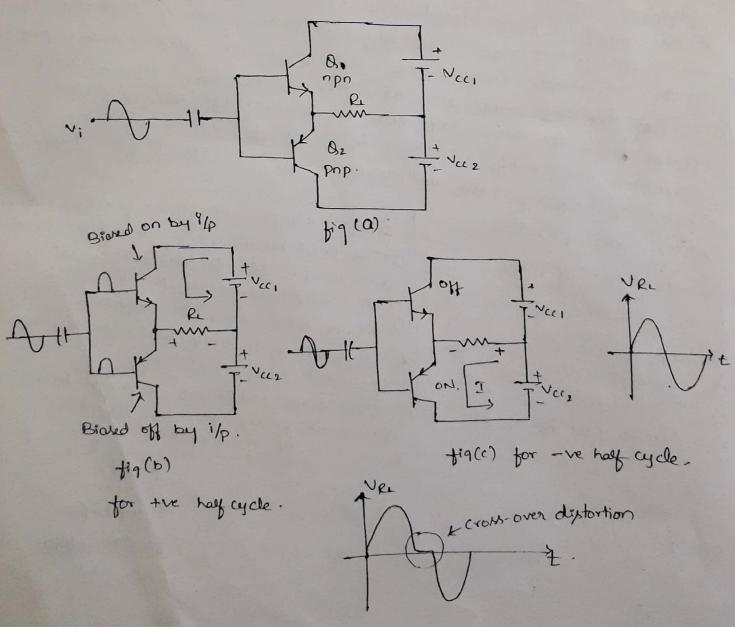
Using complementary transplory i.e. npn & pnp, it is possible to obtain a full cycle of across a load using hay-cycles of operation from each transplor on whom in fig (a). Where as a single input signal is applied to the base of both transplors, the transistor being opposite type, will conduct on opposite half-cycles of the input.

The npn transiptor will be biased into conduction by the half cycle of signal, the pnp transiptor is biased into conduction when the input goes the.

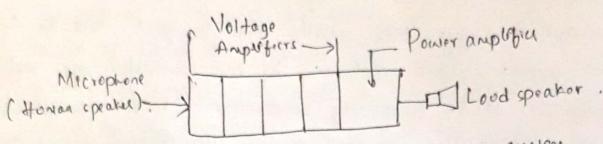
During a complete cycle of the input, a complete cycle of of prignal is developed across the load. one disadvantage of the ckt is the need for two seperate via supplies, and another disadvantage is present of cross-over distortion in the output rignal.

Cross-over distortion refers to the fact that during the signal cross-over from positive to negative, there is some Minearity in the output signal. This results from the fact that the ext doesnot provide exact switching of one transistor off and the other on at the Zero-vtg conduction. Both transistors may be positially off so that the off vtg doesnot follow the input around the Zero-vtg condition.

Brasing the transistors in class AB improves this operation by biasing both transistors to be 'on' for more than half a cycle.



Concept of large segnal Amplefreation



Poblec address System or amplifying system.

The system consists of many stages connected in cascade their basecally of is a moithstage amplifier. The enport is sound signal of a human speakor and the output is given to the loudspeaker which es an amprehed enpot segnal. The input and the entermedia stages are small segnal amplifiers : The sufficient voltage goes es obtained by all the intermediate stages. Hence these stages are ralled voltage amplifices

But the last stage gross an octpor to the load like a load 3 peaker. Hence the last stage must be capabale of delevery an appreciable amount of a.c. power to the load. So at next be capable of hardley large voltage or corrent swings or en other.

words large signals

Applications of PA:

& Public address systems

2) Radro recesous

Tope playele.

>) (attiode ray tobies etc.

Desadorning es of serres fed class A amp 1) the load resistance is derectly consoled and carries the Q collector corner. thes causes weretage of power) Power decepation es work +> Effectively Ps very poor, due to large decepation

+) The output impedance us high hence. Out cannot be out for los inspedance loade such as loads pealeds

3 us 4 49261 8 . 20 latelles