Reconfigurable hardware programming in a protocol processor unit

- SUNIL KALLUR RAMEGOWDA



Master's Degree Project School of Information and Communication Technology Royal Institute of Technology(KTH) Stockholm,Sweden 2015



Ericsson AB
Department of ASIC/FPGA
Stockholm,Sweden.

This thesis is carried out in cooperation with Ericsson AB.

I would like to thank my manager Mr. Pierre Rohdin G, for selecting and providing me an opportunity to pursue Master thesis at Ericsson. I wish to extend sincere thanks to my senior supervisor Mr. Tume Wihamre for all his guidance and support to make me understand and solve the research question. Its my pleasure to thank my immediate supervisors Katarina Nilsson and Orri Tomasson who helped me all along my thesis by guiding patiently, providing constructive feedback, arranging weekly meetings and reviewing my work. I would also take this opportunity to thank other thesis students Mehdi Taabouri, Jing Zhang X and Marcus Andersson who were next to me all the time, for all the discussions at my work place.

Acknowledgement

First of all i would like to thank Dr. Johnny Öberg for being my supervisor and examiner and for his support, guidance, advice throughout the research project. I wish to thank EIT ICT Labs Master school for providing me an opportunity to study Master's in Embedded platforms. I am pleased to thank all the Professors of AES dept, TU Berlin and ICT school, KTH. I extend my gratitude to the coordinators Nina Reinecke at TU Berlin and Helena Törnkvist at KTH for all the administrative help.

Last but not the least, i would like to thank my parents and sisters for their unconditional support, both financially and emotionally throughout my degree. In particular i am grateful for my 2 year old nephew who made them happy all the time in my absence. I would also extend my thanks to all those who sponsored my studies including Vijaya Bank, India.

Reconfigurable hardware programming in a protocol processor unit

Sunil Kallur Ramegowda October 25, 2015

Abstract

Reconfigurable hardware architectures have been a research topic for many years. Programming such architectures requires manual low level coding or the design of custom compilers to generate the required configuration data for the architecture.

A protocol processor, in general processes the packets according to the protocol. There are number of protocols like Ethernet and CPRI to define how the data has to be sent and received between the source and destination points. Data packets can be processed using generic processors programmed in software, but hardware processing is always faster and energy efficient.

A compiler/mapper is investigated in this thesis work. The language application is developed using a parser generator tool called Antlr. The grammar is written in Extended Backus Naur Form (EBNF) and the corresponding language is used to describe the architecture and the protocols. The tool will generate a hardware model and its interconnections in SystemC TLM2.0 based on the protocol description. The hardware model is connected to Universal Verification Methodology (UVM) driver and functionality is verified by simulation. The Ethernet protocol is described using the developed language and the complete framework is verified.

Contents

A	bstra	ct		i
$\mathbf{C}_{\mathbf{c}}$	onter	$_{ m nts}$		iv
\mathbf{A}	bbre	viation	s	\mathbf{v}
Li	st of	Figure	es	viii
Li	st of	Tables	S	ix
1	Intr	oducti	ion	1
	1.1	Hardw	vare and Programming	. 3
	1.2		figurable Systems	
		1.2.1	Granularity	
		1.2.2	Reconfiguration Models	
		1.2.3	Reconfiguration rate	
	1.3	Purpo	se of this thesis	
	1.4	-	em Description	
	1.5	Goals	· · · · · · · · · · · · · · · · · · ·	. 9
	1.6	Limits	s on scope	. 9
	1.7		ure of the thesis	
2	Bac	kgrour	ad	11
	2.1	_	Grain Reconfigurable systems	. 11
	2.2		e Grain Reconfigurable systems	
	2.3		evel synthesis	
	2.4	Protoc		
		2.4.1	Ethernet	
		2.4.2	Xio-s	. 23
		2.4.3	CPRI	
	2.5	Gramı	mar and Language	. 25
		2.5.1	Parser	
		2.5.2	Backus-Naur Form	. 26
3	Mo	deling		28
	3.1	_	uction	. 28
	3.2		n-Level Modeling	
		3.2.1	Language comparison	
		3.2.2	Transaction Level Modeling	
		3.2.3	SystemC	

CONTENTS

3.3	Verification Methodology	
3.4		
		35
a =		36
3.5	Comparison of 3 Protocols	38
Frey	v j a	40
4.1	Freyja Architecture	40
	4.1.1 Switch port	40
	4.1.2 Operator units	40
	4.1.3 Switch wrapper	41
	4.1.4 Overall architecture	42
4.2	Parser implementation	42
4.3	Operator Instantiation	45
4.4	Context switching	46
4.5	Memory	47
4.6	Transaction handling	48
4.7	Error handling	48
Ana	lysis	49
	v .	49
5.2		51
	· · ·	54
		54
		55
	•	55
6.4	Timing annotation	56
Con	clusion and Future work	57
7.1	Reconfigurable architecture	57
7.2	Protocol sharing	58
7.3	Language framework	59
7.4	Limitations	59
7.5	Future work	60
eferei	nces	61
pene	dix	66
•		
	3.4 3.5 Frey 4.1 4.2 4.3 4.4 4.5 4.6 4.7 Ana 5.1 5.2 Test 6.1 6.2 6.3 6.4 Con 7.1 7.2 7.3 7.4 7.5 eferer	3.4 High level description of the protocol 3.4.1 Related work 3.4.2 Antlr 3.5 Comparison of 3 Protocols Freyja 4.1 Freyja Architecture 4.1.1 Switch port 4.1.2 Operator units 4.1.3 Switch wrapper 4.1.4 Overall architecture 4.2 Parser implementation 4.3 Operator Instantiation 4.4 Context switching 4.5 Memory 4.6 Transaction handling 4.7 Error handling Analysis 5.1 Ethernet 5.2 Xio-s Test system 6.1 Complete Test System 6.2 Payload and blocking transport 6.3 Input and output 6.4 Timing annotation Conclusion and Future work 7.1 Reconfigurable architecture 7.2 Protocol sharing 7.3 Language framework 7.4 Limitations

CONTENTS	CONTENTS
B Grammar file : freyja_arch.g4	69
C Language Recognition terms	71

Abbreviations Abbreviations

Abbreviations

ASIC Application Specific Integrated Circuit.

ASIP Application Specific Instruction-set Processors.

BNF Backus Naur Form.

CLB Configurable Logic Block.

CPRI Common Public Radio Interface.

CRC Cyclic Redundancy Check.

DRRA Dynamic Re-programmable Resource Array.

EBNF Extended Backus Naur Form.

EDA Electronic Design Automation.

FCS Frame Check Sequence.

FPGA Field Programmable Gate Arrays.

GPP General Purpose Processor.

HDLs Hardware Description Languages.

HLS High Level Synthesis.

IEEE Institute of Electrical and Electronics Engineers.

MAC Media Access Control.

OSI Open Systems Interconnect model.

PE Processing Element.

PLA Programmable Logic Array.

RBS Radio Base Station.

Abbreviations Abbreviations

RTL Register Transfer Level.

SFD Start of Frame Delimiter.

 ${f TLM}$ Transaction Level Modeling.

 ${f UVM}$ Universal Verification Methodology.

VHDL Very High Speed Integrated Circuit Hardware Description Language.

List of Figures

1	Domains and level of description in Gajski Y-chart 4
2	Flexibility Vs Performance of Hardware Classes
3	Reconfiguration Models
4	Hardware Software Design Gaps versus Time [1] 8
5	Field Programmable Gate Array [2]
6	Configurable Logic block [2]
7	Basic Coarse grain reconfigurable architecture [3]
8	Coarse Grain Reconfigurable blocks [3]
9	Ethernet Transmit
10	Ethernet Receive
11	Ethernet Encoder block
12	Ethernet Decoder block
13	Xio-s Transmitter
14	Xio-s Receiver
15	Xio-s Frame format Type 1
16	Xio-s Frame format Type 2
17	CPRI Transmitter
18	CPRI Receiver
19	Parser
20	Hardware description languages and abstraction levels [4] 30
21	Layered SystemC Architecture
22	Initiator and Target sockets
23	Many to many binding
24	UVM Sequencer
25	Antlr: Input, Grammar and Parsetree
26	Transmitter of all 3 protocols
27	Receiver of all 3 protocols
29	Switch port
28	Switch based Freyja network topology
30	Freyja operator
31	Freyja Switch Wrapper
32	Freyja Architecture for 4 operators
33	Antlr generated Files
34	CRC operator in Freyja Architecture
35	Freyja Operator Instantiation
36	Freyja Memory content definition
37	Freyja protocol Interconnection
38	Freyja CRC operator Control Block code segment 50
39	Frevja Reconfiguration constants

LIST OF FIGURES

	•	٠			52
ion					52
on					53
t shown)) .				54
					55
					55
					56
	ion on t shown)	ion on	ion	ion	ion

List of Tables

1	OSI Model	1
2	Coarse Grain Reconfiguration Mechanism [3]	16
3	Ethernet raw frame	21
4	Preamble and SFD	21
5	Control Word	22

1 | Introduction

A set of rules define the communication strategy between digital systems. There are many rules which makes the communication possible between systems. Over the decades the rules have evolved into standards. Open Systems Interconnect model (OSI) is an international effort to facilitate communications among different manufacturers and technologies. OSI reference model partition communication systems into 7 abstraction layers. It addresses the interconnection requirement of an open systems environment.

	Layers	Functions	Examples
7	Application	High level APIs	Mail, IE, Firefox
6	Presentation	Character code translation, Data conversion, Data compression etc	ASCII, JPEG
5	Session	Session establishment between processes running on different systems.	HTTP, SMTP
4	Transport	Acknowledgement, Segmentation, Multiplexing etc	TCP, UDP
3	Network	Addressing, Routing, Traffic control etc	IPv4, IPv6
2	Data Link	Error free data transfer from one node to another	PPP, IEEE 802.2
1	Physical	Transmission and reception in physical medium.	DSL

Table 1: OSI Model

The rules are called as *protocols* in communication systems. They help in exchange of information between the digital systems. Most of the higher layers (4 to 7) are processed through generic computer systems and the lower layers with dedicated embedded hardware. General Purpose Processor (GPP) found in embedded systems cover a broad range from 8 bit low cost microcontrollers, to 32 bit RISC

1 INTRODUCTION

microprocessors, to today's high performance processors with Digital Signal Processing (DSP) enhancements. These can be considered as hybrid processors which incorporate signal processing features into embedded hardware that unify control and signal processing in a single core. In the communication domain, a network processor, is a programmable microprocessor optimized for processing network data packets. Common functions include header parsing, pattern matching, bit manipulation, packet modification, shift and data movement. Software programmability of network processors allows it to be more flexible across a range of applications. Even though all network processors are programmable, the user might not be given access for programming, restricting the programmability with vendors. Other GPPs can be programmed for protocol processing as they are less expensive. Solutions based on GPP or Application Specific Integrated Circuit (ASIC) exist [5][6] for protocol processing. GPP will have more flexibility but are less energy efficient when compared to ASIC which is less flexible but the most energy efficient. Application Specific Instruction-set Processors (ASIP) or domain specific processors are more suitable for the protocol processing task and depending on their architectural characteristics they allow varying degrees of trade-off between flexibility and energy-efficiency [7].

Another class of hardware architecture called as Reconfigurable architectures, are devices that contain programmable function blocks and programmable interconnects between function blocks. The most mature class of reconfigurable architectures is Field Programmable Gate Arrays (FPGA) which are considered fine-grained reconfigurable architectures. Recent advance research in these field have led to the Coarse grain reconfigurable architectures and reconfigurable computing platforms. Hardware resource usage and performance varies depending on the reconfigurable architecture and its level of abstraction [8]. The design of coarse grain reconfigurable hardware architecture requires the compiler to produce the configuration or the hardware compatible code [9]. These files can be produced on run time when the application is running or in a static way before execution. The complexity of the system depends on the selected design.

The reconfigurable hardware is modeled in SystemC language using Transaction Level Modeling (TLM). The files required for the reconfiguration is obtained by parsing the description of protocols which is described in high level language. The Antlr tool is used for building the base parser file for the defined grammar and then the required functions are implemented to output the complete system and configuration data.

1.1 Hardware and Programming

Until late 1960s, ICs were designed, optimized, and laid out by hand. Gate level simulation appeared in the early 1970s, and cycle based simulation became available by 1979. Place and Route, schematic circuit capture, formal verification and static timing analysis techniques were introduced during the 1980s. Meanwhile Gajski and Kuhn introduced the Y chart for describing the hardware as 3 different domains namely Behavioural Domain, the Structural Domain and the Physical Domain [10]. It represents different levels of hardware abstraction which are indicated by concentric circles as illustrated in Fig. 1. The three axes represents the 3 different domains and the synthesis can be viewed as a process of transformation from one axis to another and/or from higher level to lower level. Behavioural domain describes the functional behavior of the system. Structural domain maps the hardware into subsystems and how they are interconnected. Geometrical domain represents the geometric properties of the system and its subsystems. Each intersection points of domains and concentric circles represents different abstraction levels with respect to their domains. Simulation tools to represent the hardware are widely adapted during 1980s based on hardware description languages such as Verilog(1986) and Very High Speed Integrated Circuit Hardware Description Language (VHDL)(1987).

In the software domain, similar to the hardware domain, progress in design methodologies for programming the hardware have helped with today's sophisticated hardware independent compiler environment. Machine code was once the only means of programming the hardware. In 1950s, assembly level programming was introduced which abstracted the programming one step higher. With the compilers becoming more intelligent, High level languages evolved and associated compilation techniques were developed to improve software productivity. High level languages like Python, Perl, Java, PHP etc are platform independent and provide the user with flexibility and portability by hiding details of hardware architecture. Languages like C, C++ abstracts from assembly level and together with C compilers it helps in efficient object code generation. Programmers create interfaces to the C library so the routines can be used from higher level languages. With the new hardware architecture becoming more complex in nature, the software applications are developed to provide good solutions which are in parallel becoming more complex.

1.2 Reconfigurable Systems

In 1960 Gerald Estrin, proposed the idea of a fixed plus variable structure computer [11]. It consisted of a fixed processor and an array of reconfigurable hardware which was controlled by the fixed processor. Even though the idea was demonstrated with a proof, the industry did not consider to further innovate in these field and till 1980's

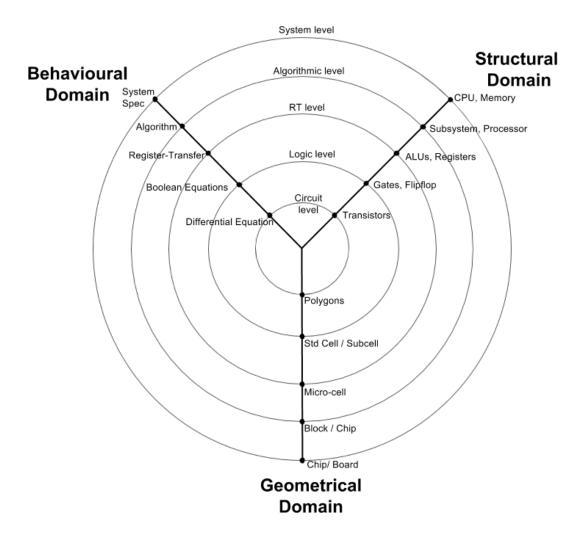


Figure 1: Domains and level of description in Gajski Y-chart

there were no significant developments. In 1985, the reconfigurable Programmable Logic Array (PLA) was patented [12]. Innovation in PLA's further continued with the commercially available FPGA in today's market.

In the field of computer architecture, designers make decisions based on flexibility and performance requirement [13]. ASIC are the least flexible in terms of adapting for any change in the application and GPP are the most flexible as they are independent of the application and the core can be programmed to make the required algorithm work at the cost of higher power and lower efficiency. ASIC and GPP lies in extreme corners of the graph between Flexibility Vs Performance as in Fig. 2. Reconfigurable architectures are intended to fill the gap and provide more flexibility in terms of

hardware and potentially higher performance than software [13].

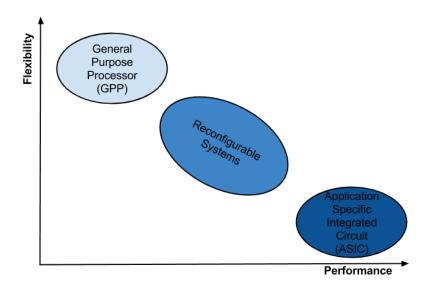


Figure 2: Flexibility Vs Performance of Hardware Classes

1.2.1 Granularity

Reconfigurable devices like FPGA have the Configurable Logic Block (CLB) which can be configured to map the required functionality. The complexity of the function is not a concern but the number of inputs and output of the function has to be considered based on the FPGA architecture. This level of granularity in implementing the functions is called Fine grained Reconfigurable Architecture as it provides the reconfigurable granularity till lowest possible level. These reconfigurable devices are not energy efficient and the execution speed is less than the ASIC counterpart. Another type of reconfigurable devices are the coarse grained reconfigurable architectures. These devices have the granularity at function levels. They will configure the function blocks to achieve efficient algorithm implementation. The function blocks can vary from constant block to complex functions which are commonly used by the application.

1.2.2 Reconfiguration Models

The reconfigurable architectures need configuration of hardware. This can be at compile time or at runtime of an application as in Fig. 3.

In the compile time reconfiguration model, the reconfigurable hardware system is configured at compile time and will be static during the application run time.

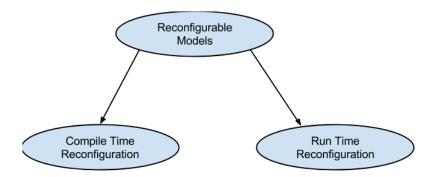


Figure 3: Reconfiguration Models

In this model the programmable logic can be configured to perform some specific task like hardware accelerators to achieve high performance. An FPGA configured to perform floating point multiplication together with a GPP will accelerate the performance of the application if the GPP doesn't have a Floating Point Unit. In the run-time reconfiguration model, the reconfiguration hardware is configured at run time and will be dynamically programmed to perform different tasks. The decision for making such dynamic reconfiguration has to be embedded coupled with the application and hence it increase the overhead. The Dynamic Re-programmable Resource Array (DRRA) fabric developed at KTH Electronic System Dept is an example of this model [14].

1.2.3 Reconfiguration rate

The Fine graine systems will have more reconfiguration data (in FPGAs it is in term of bit streams) which leads to more time and the Coarse Grained Reconfigurable systems will have comparatively less blocks as they have higher granularity and will contain less reconfiguration data. Hence the Coarse grain architecture will take less time to re-configure. This depends on the dynamic reconfiguration architecture whether the complete fabric is reconfigured or partially reconfigured during runtime.

1.3 Purpose of this thesis

Ericsson AB [15] is a market leader in the radio base station equipments. There are different protocols being used for communication in the Radio Base Station (RBS) units. Ethernet, explained by Institute of Electrical and Electronics Engineers (IEEE) in 802.3 standard, defines the protocol for 10Gbit transfer which is mainly used for communication between the silicon chips. Other protocols include Common Public Radio Interface (CPRI), Serial Rapid IO (SRIO), Xio-s (Erics-

son Specific protocol) for reliable communication between chips at high data rate. Most of these Media Access Control (MAC) layer protocols share common hardware functions. Ericsson design and manufacture custom ASIC chips for Baseband signal processing and Radio signal processing. Data is received and transmitted from these chips by one of the agreed protocols. A reconfigurable hardware can be designed to route the data using different protocols.

The reconfigurable architecture requires a new hardware and software co-design. The reconfiguration details are extracted based on the hardware design and the compiler/mapper should be able to produce such reconfiguration. This is accomplished by using Grammar based technique i.e by defining a language based on EBNF grammar and then describing the protocols using this language. The overall architecture and working principle will be explained in further chapters.

The thesis deals with understanding the reconfigurable architecture and identifying the configuration details to make the system work for different protocols. The description in high level language is used to extract these configuration details and to verify the complete system using a test bench.

The purpose of this thesis is to investigate an approach of a compiler or mapper to describe the protocols in high level language and then map it to hardware blocks and their interconnection. This involves showing the proof of concept by SystemC TLM simulation models. The Individual hardware blocks are modeled in SystemC and can vary from a simple block to complex functions of the protocols. This thesis work serves as a proof for the project in Ericsson AB to further investigate the feasibility of developing such architectures.

1.4 Problem Description

As explained in the previous section reconfigurable architecture with lowest granularity are available for different applications. They still face the challenges of lower speed, high energy consumption and the compatibility of tool chains between different vendors. The lowest reconfigurable granularity can be designed at bit, block or function level. When targeting hybrid architectures to improve either performance, cost or speed, the application must be partitioned in such a way that certain repetitive or computation intensive functions are mapped on a reconfigurable hardware. Such mapping is not simple as it requires deep understanding of both hardware and software design [16]. The know how of the process to build the complete system which solves the practical problems is of great importance. Fig. 4 from the International Technology Roadmap for semiconductors shows how hardware and software design productivity has lagged Moore's law. The need of configuration data for the

reconfigurable hardware requires the design of new software tools.

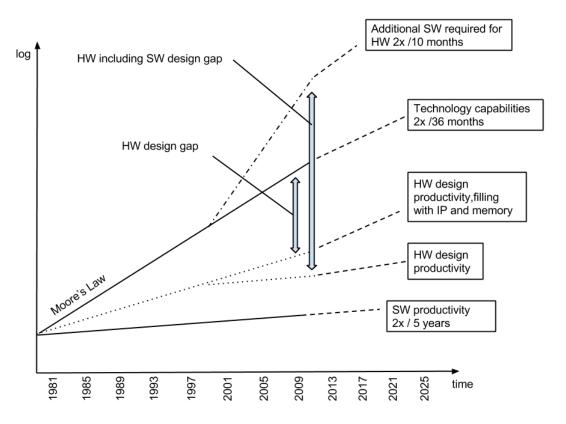


Figure 4: Hardware Software Design Gaps versus Time [1].

The RBS receives and transmit data using different protocols. Identifying errors, providing security are some of the main features for reliable communications. Algorithms used by different protocols for providing such features differ by minor polynomial change. Identifying such functions is required for efficient reconfigurable hardware design. The designer need to have the overall understanding of hardware and software for such a system. How the reconfigurable hardware can be programmed to accomplish the protocol processing in an efficient way by designing the framework using high level description needs to be explored. High level language development provides flexibility for the designer to describe in their own syntax. The development time need to be evaluated for such an approach. This allows the user with minimal know how about hardware to reconfigure and show result in short time.

1.5 Goals

The thesis goal is to achieve the below milestones:

- Understand the reconfigurable hardware architecture designed at Ericsson AB
- Understand Ethernet, Xio-s and CPRI protocols
- Define a language to describe the protocols in high level description
- Identifying how to represent the reconfiguration information
- Mapping the description of language to hardware and interconnections
- Integrating Ethernet protocol for the complete system
- Verifying the system by simulation

1.6 Limits on scope

The thesis focus more on showing the proof of concept considering one to two protocols i.e Ethernet and Xio-s. The language will be designed such that it is easy with minor modification to extend for other protocols like CPRI. Developing and integrating the TLM models for all the protocol's will not be feasible in this time line. The architectural changes required are suggested but not changed as the focus is more on describing the protocol in high level description.

1.7 Structure of the thesis

The thesis is organized to provide required details for understanding the overall work. The first chapter gives a brief introduction to the reader about the topic of investigation, limitations and goals. The rest of the thesis is structured as follows.

Chapter 2: This chapter will describe the background about the topic. It is summarized in three sections starting with Reconfigurable hardware architectures and their terminologies. The three different protocols considered in protocol processors are discussed in detail. The last section explains the meaning of Grammar, Language and parser in computer science world.

Chapter 3: This chapter describes the research methodology inculcated in carrying out this thesis work. It starts with the research process and then the introduction of the reconfigurable hardware architecture under study i.e Freyja architecture details. Then the process of building a language application is explained

using Antlr tool. The comparison of three protocols under study and their common functions are also discussed.

Chapter 4: The chapter starts with the details of building the framework. It also discuss how the architectural requirements are mapped using the high level description. Mapping of Freyja operators, their memory contents, interconnections and error handling are discussed.

Chapter 5: This chapter explains the details of complete test system. How the overall system can be represented using the input file and how the data is processed and handled within the system.

Chapter 6: This chapter explains the integration of Ethernet protocol functions into the system of files auto generated from the language application. The challenges and the mapping of input description are discussed in detail. The Xio-s protocol is also considered but part of the functions are integrated to show the context switching and highlighting the control block changes.

Chapter 7: This chapter concludes the thesis by explaining the outcome of connecting 3 dots namely reconfigurable hardware, protocols and Grammar based high level description framework. The limitations encountered and the future work is discussed further.

2 | Background

To understand the reconfigurable hardware and its terminologies, this chapter explains in detail about different architecture classes and its meaning. Protocols and their common functions are also explained which helps in designing the reconfigurable protocol processor. The further sections explains the meaning of grammar and language and its terms.

2.1 Fine Grain Reconfigurable systems

Fine grain reconfigurable logic arrays have evolved into commercially available FP-GAs. They are scalable chip architecture based on a 2D array of simple computational cells with individually configurable processing functions and an electronically configurable interconnect structure allowing complex application circuits to be built from the available cells [17].

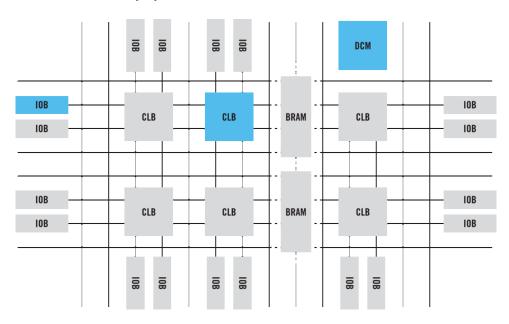


Figure 5: Field Programmable Gate Array [2]

Fig. 5 shows the FPGA architecture with a matrix of Configurable Logic Blocks (CLBs) connected through programmable interconnects. The CLB is the basic logic unit in a FPGA wherein the number of such units in each device varies depending on cost. Each CLB consists of a configurable switch matrix with 4 or 6 inputs, some

selection circuitry and flip-flops. The highly flexible switch matrix can be configured to handle combinatorial logic, shift registers or RAM [2]. IOBs refers to the Basic select IO structure. The interconnect routes the signals between CLBs and to and from IOs. Most of the modern FPGA boards include the Embedded block RAM and Digital Clock Management (DCM).

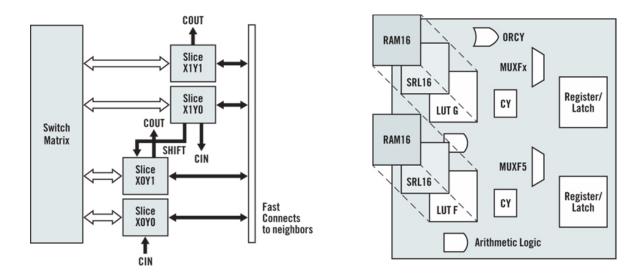


Figure 6: Configurable Logic block [2]

The introduction of FPGA led to the research of optimizing and reusing the reconfigurable logic blocks. The Electronic Design Automation (EDA) tools compile and synthesize Hardware Description Languages (HDLs) to create a physical design in terms of the FPGA's resources. They are attractive compared to the ordinary CPUs because of less power consumption per computation and price per performance [18]. Despite these merits, FPGAs have still achieved limited applicability in industries. The main reason is that most software programmers lacks the knowledge of hardware description languages which are used to describe the complex algorithms in FPGAs. FPGA programming requires a more rigorous development process, involving training the programmers beyond application level, resulting in an increase in development cost and time to market.

To overcome the problems stated above, research on mapping high level description to Register Transfer Level (RTL) level started. The growing capabilities of silicon technology and the increasing complexity of applications in recent decades have also forced design methodologies and tools to move to higher abstraction levels. The methodology is called High Level Synthesis (HLS). It enables the automatic synthesis of high level, untimed or partially timed specification (SystemC) to a low-

level cycle accurate RTL specifications for efficient implementations in ASICs or FPGAs. During the 1990s, the first generation of commercial HLS tools were available [19]. HLS tools helped the designer to use High Level Languages (HLLs) for FPGA programming similar to processor programming. The latest generation of HLS tools, in most cases, uses either ANSI C, C++, or languages such as SystemC that are based on C or C++ that add hardware-specific constructs such as timing, hardware hierarchy, interface ports, signals, explicit specification of parallelism, and others [19]. Some of the commercial HLS tools are Mentor's Catapult C, Forte's Cynthesizer, Cadence C-to-Silicon, NEC's CyberWorkbench and a new French company Synflow tool has its own language named Cx.

FPGA requires the reconfiguration of programmable fabric either before the program execution or partial reconfiguration on the fly when the program is running. The streaming and multimedia application requires the reconfiguration of fabric on the fly to optimally utilize the fabric. As FPGAs have granularity at the lowest possible bit level, the reconfiguration data is huge and the time for the hardware to reconfigure increases with the complexity. Most of the data handled in real world is either in Byte or word width and reconfiguring the interconnect at bit level can be abstracted to higher level. This reduces the overhead of reconfiguring each bits. Similarly the commonly used combinatorial logics can be defined as standards blocks for the custom design with the standard set of reconfiguration details. These features will reduce the amount of data and time required for reconfiguration and such a hardware class is called Coarse Grain Reconfigurable Logic.

2.2 Coarse Grain Reconfigurable systems

To overcome the limitations of Fine grained reconfigurable systems, new architectures were explored. Optimally designed processing elements which perform word level data processing are configured with few configuration bits at word level. Due to the word level reconfiguration, a small number of configuration bits is required resulting into a massive reduction of configuration data, memory needs, and reconfiguration time [20]. Even the interconnections, since they are grouped in buses, they are configured by a single control signal instead of separate control signal for each wire. Also, because few programmable switches are used for configuration purposes and the Processing Element (PE)'s are optimally-designed hardwired units; high performance, small area, and low power consumption are achieved [20].

A generic architecture of Coarse grain reconfigurable system is illustrated in Fig. 7. It consists of a set of Reconfigurable units (RU), a programmable interconnect, a configuration memory and a controller. The coarse grain reconfigurable part will be designed to take the computationally intensive parts of application

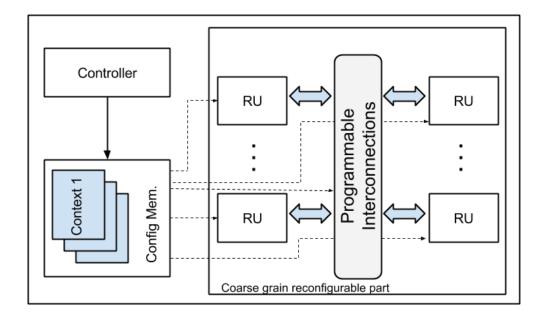


Figure 7: Basic Coarse grain reconfigurable architecture [3]

and in most cases coupled to the main processor which takes care of other tasks. Coarse grain architectures are always optimized for the target domain. The number of reconfigurable units, their design, the interconnection network, memory and controller are tailored to the domain's needs. Memory will hold the control (configuration information) bits that are used to program the reconfigurable units and interconnection network. The configuration memory may store multiple configuration contexts but one context active at a time [20]. The controller is responsible to control the loading of configuration context from the main memory to configuration memory, to monitor the execution process of the reconfigurable hardware and to activate reconfiguration contexts.

The programmable interconnection network ensures the communication of data between the computing reconfigurable units. The wires are grouped into buses and are configured by single configuration bits. The interconnection network can be realized by a crossbar, mesh or a mesh variation structure.

The reconfigurable units are the processing units, which are domain specific hardwired units, perform useful operation to accomplish the application requirements. The operations might refer to logic or arithmetic operation. They reconfigure autonomously based on the control information and are therefore different from the CLB's of fine grain architecture. Each unit is configured at word level, configuration bits reconfigures the entire unit and not each slice at bit level. Theoretically, the granularity might vary from bit to any word length. In most of the practical architectures which are designed, the granularity is either 8bits or above, as the processing of data happens with the word length of 8bits or more.

The reconfigurable units in the architecture are optimized hardware units for the application domain. It can be designed to perform any word level arithmetic or logical operations. As coarse grain reconfigurable systems target at a specific domain, the RU's are designed with the operation required by the domain in mind. The reconfigurable units are hardwired units optimized to perform for specific application domain to improve performance, area and power consumption.

GPP with programmed embedded software is the classical approach which tightly couples data and control flow for many applications. They can be coupled with the general purpose embedded processors in different ways. The design depends on the performance and requirements of the application. Three different scenarios are illustrated in Fig. 8. These additional hardware units coupled to the processor will accelerate the performance or increase the throughput of the overall system. This illustration also helps to understand the different ways of reconfiguring the fabric.

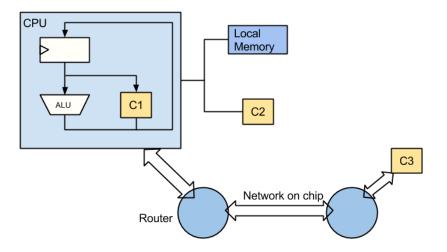


Figure 8: Coarse Grain Reconfigurable blocks [3]

Register mapped reconfigurable blocks gives the tightest integration with embedded software. They are created by modifying the micro-architecture of the embedded core. For example as in Fig. 8, C1 is an additional data path next to ALU. For programming such an architecture, it should be visible in the instruction set of the embedded core. Since it requires tight coupling, the exploiting the parallelism with the introduction of new hardware unit becomes cumbersome. The tools need to

provide such features for the user to map the application for the new hardware architecture.

Memory mapped reconfiguration is the next variant illustrated as C2 in Fig. 8. This is achieved by providing a memory interface to the reconfigurable block. It results in looser coupling between software and the reconfigurable block. A set of shared memory locations can convey control and data flow orientated information. As explained in [3], reconfigurable blocks need to share memory address space with other peripherals. The control and data flow information might need to be routed through the CPU which increases the bottleneck.

In Network mapped reconfigurable blocks (C3), the network packets have the control and data flow information. In this case integration of embedded software and reconfigurable blocks can be done using communication primitives. As discussed in [3], Network mapped systems deviates more from the classic sequential model and hence the programming model is more complicated. Table.2 shows the comparison of different coarse grain reconfigurable systems.

Mapping	Architecture Strategy	Reconfiguration Mechanism	Data flow / Control flow Coupling	Energy Efficiency Improvement
Register- Mapped	Custom Datapath	Custom Instructions	Tightly Synchronized	Low
Memory- Mapped	Co processor	Memory mapped instructions	Loosely Synchronized	Medium
Network- Mapped	Peer processor	Configuration Packets	Uncoupled	High

Table 2: Coarse Grain Reconfiguration Mechanism [3]

2.3 High level synthesis

High level synthesis is the process of translating from algorithmic to Register transfer level description as illustrated in Fig. 1. Transforming application level programs which are described in high level languages directly into register transfer level will not require the programmer to know the detailed hardware architecture. High level synthesis tools explore the design space based on the information from high level

abstraction with more degrees of freedom. There were many academic and industry projects showing the concept of High Level Synthesis (HLS) [21] [22]. They operate on internal models known as control/data flow graphs (CDFG) and produces a RTL model of the hardware implementation [23]. The computationally intensive applications like DSP algorithms were the target in the initial research of HLS tools.

A brief introduction to the academic and industry standard high level synthesis tools are discussed here.

GAUT [24] is an academic High level synthesis tool dedicated to Digital signal processing applications. It is an open source tool which takes a pure C function along with the constraints to extract the potential parallelism and data dependencies. After the allocation, scheduling and binding tasks it will generate the RTL code.

Cynthesizer [19] is a pin and protocol accurate SystemC model used as synthesis input of HLS tool. It produces the optimized RTL for specified target technology identified by the user in the form of a .lib file.

A new approach called as *interactive synthesis methodology* is explained in [25], which takes C language as input and allow the user to control and change the synthesis decisions about scheduling, allocation and binding by using Graphical user interface (GUI) at any time.

DRESC [26] presents a retargetable compiler for a family of coarse grain reconfigurable architectures. It focuses on loop level parallelization for different segments of application code and uses modulo scheduling algorithm for mapping into hardware blocks.

A generic compilation framework for architectures based on dataflow execution paradigm is explained in [27]. It describes a method to transform applications described in HLL to Data flow graphs (DFGs) and technique to optimize the same.

All the above tools help the designers in exploring the hardware architectures with more degrees of freedom. In other words, the hardware architecture changes based on the designer input. In most of the coarse grain architecture designs, the synthesis tools are developed for the custom optimized hardware architectures by assuming that designer know the RTL implementation of the hardware. Few such cases are discussed below:

VESYLA (VEctorizing SYmbolic Language Assembler) [28] is a semiautomatic

framework for implementing DSP functions. This tool is relatively custom design for the coarse grain reconfigurable architecture designed at KTH called as DRRA (Dynamically Reconfigurable Resource Array). It takes an untimed C specification of a DSP function with pragmas and generates configurate for DRRA architecture.

RaPiD (Reconfigurable Pipeline Datapath) [29] architecture is a coarse grain architecture that allows pipelined computational structures to be constructed from an array of arithmetic units, registers and memories. Programming is performed using RaPiD-C, a C-like language with extensions to explicitly specify parallelism, data movement and partitioning [4]. The compilation process produces a structural specification with components specific to the underlaying architecture.

PipeRench [30] is a coarse grain reconfigurable system consisting of stages organized in a pipeline structure. It uses a *pipeline reconfiguration* technique to provide fast partial and dynamic reconfiguration and it also provides runtime scheduling of configuration and data streams. Programming such a complex model is performed using source language called *dataflow intermediate language* (DIL) which is a single assignment language with C operators. After parsing, the compiler inlines all modules, unrolls all loops, and generates a straight-line, single-assignment code [4].

Pleiades [31] is a coarse grain reconfigurable template with heterogeneous processing elements, optimized for a given domain of algorithms and which allows runtime reconfiguration. Detailed architecture is discussed in [4], which indicates the mapping concept divided into 2 parts. The first task is to create the template instance and the other is to map an algorithm onto a processor instance. A reference of power and performance with respect to a general purpose processor is used to derive the resources for the architecture.

Montium [32] is a coarse grain reconfigurable architecture which resembles a VLIW architecture. It is optimized for the typical operations like correlation, finite impulse response (FIR) filters, matrix and vector multiplication, Max-Log-MAP decoding, 8x8 point Discrete Cosine Transform (DCT) and Fast Fourier Transform (FFT). Programming this architecture consists of transforming C to directly architecture dependent code provided kernel written in C code is available. Otherwise the HLL C is translated into CDFG and then clustering and finally into Montium C code.

The design of custom hardware architectures require a framework for producing the executable code and/or reconfiguration data. Most of the EDA tools are not matured to explore the coarse grain architectures. The projects discussed above

develop their own framework for mapping application into their architecture.

2.4 Protocols

The communication between chips in Radio Base station equipments has many protocols to fulfill the requirements of the specification. The protocols differ by standards. The most commonly used protocols in the wireless base stations are discussed briefly in below section.

Gb Ethernet (GbE): Is used for low rate signaling paths for which latency is not critical. It supports communication over a wide variety of physical media and transmission ranges, and is therefore well-suited for inter-module communication.

10Gb Ethernet (10GbE): It is increasingly being used for critical low-latency, fast-path communication. The data rate is consistent with the throughput and latency requirement of typical macrocell base stations. Like all Ethernet variants 10GbE can support communication over a wide variety of physical media and transmission ranges and is well suited for inter module communication.

Serial Rapid I/O (SRIO): It is used for low latency, fast path communication, particularly within the Base-Band Unit (BBU). SRIO is flexible serial interface standard which is well suited to these requirements. It offers the possibility of very high throughput (5 Gb/s and 6.25 Gb/s) per lane (v2) with low latency. The range of components which support SRIO is more limited than Ethernet but many components associated with wireless baseband processing have adopted it as their principal high-speed data interface. SRIO is less widely used on general-purpose processors (GPPs); hence, its utilization is often limited to the BBU application.

Peripheral Component Interconnect Express (PCIe): It is sometimes used for both slow- and fast-path communication within the base station. It is able to support high data rates (5 Gb/s per lane and 8 Gb/s) and low latency. It's use is less widespread in wireless base stations than either Ethernet or SRIO, and it is normally used to provide connectivity to a GPP.

Processor local buses: They have been used to provide intra-module communication between a host GPP and its peripherals. Such interfaces are used to support both slow- and fast-path communication. This type of architecture is now rarely used in new designs and has largely been superseded by serial interfaces and/or Ethernet-based interconnect systems.

Proprietary interfaces like Xio-s: They are used to provide slow- and fast-path communication within the base station when the system designer has control of both ends of the link.

The next section will explain in detail about the 3 protocols which are of study for the reconfigurable protocol processor. It will explain about different operations performed at each level of protocol processing.

2.4.1 Ethernet

Ethernet is a widely used protocol for data communication. It is typically used in Local Area Network (LAN) applications. IEEE organization has standardized the protocol and revises it according to the technological advancement. The recent standard available is from 2012 [33] and it defines the protocol for different applications.

Ethernet Transmit

The Ethernet transmit sequence is shown in Fig. 9.

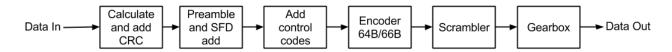


Figure 9: Ethernet Transmit

The *DataIn* is from the higher layers of the protocols which contains the data to be transmitted using MAC layer protocol. Physical layer protocols are out of this thesis scope and not explained.

Ethernet Receive

The Ethernet receive sequence is shown in Fig. 10. DataIn is from the physical transmission layer and Dataout is to the higher protocol layers.



Figure 10: Ethernet Receive

Ethernet Raw frame

The Ethernet raw frame format is represented as in Table. 3. Each of these frames enter the transmitter *Datain* as in Fig. 9 and in the receiver *Dataout* as shown in Fig. 10.

Mac Destination Address	Mac Source Address	802.1Q VLAN tag	Ethertype/length	Payload
6 octets	6 octets	4 octets	2 octets	42-1500 octet

Table 3: Ethernet raw frame

A brief functional description of each blocks in the transmitter and receiver section is explained below.

CRC

Cyclic Redundancy Check (CRC) is used to detect errors incurred during the physical transmission. The CRC value is computed by dividing the data to be transmitted with the pre-defined CRC polynomial stored in the memory[34]. The remainder of the division is known as the Frame Check Sequence (FCS).

In the transmitter side, FCS is computed for the incoming data and appended as last 4 bytes (32bits). In the Ethernet receive, the FCS is again computed for the incoming data and is compared with the FCS field for any errors. This block will not change the incoming data apart from appending the FCS field.

Preamble and SFD

Preamble is added at the start of the frame to indicate the new Ethernet frame. This block will not change the incoming data apart from adding the Preamble(7bytes) and Start of Frame Delimiter (SFD) at the beginning of the frame(1byte).

Preamble	10101010 10101010 10101010 10101010 10101010 10101010 10101010
SFD	10101011

Table 4: Preamble and SFD

In the receiver, the Preamble and SFD are identified and deleted.

Control codes

The Add control codes block will add the control codes for the incoming data such that the Encoder block can use the 8 octets to encode the data based on this control codes. Idles are added to the data if the length of the data is not equal to 8 octets. Each bit in the control word represents whether the octet is data, terminate or an idle octet.

Type	Idel	Idel	Terminate	Data 4	Data 3	Data 2	Data 1	Data 0
Control	1	1	1	0	0	0	0	0
\mathbf{word}	1	L	1	U	U	U	0	U

Table 5: Control Word

In the receiver, the complete process is reversed. It will search for the control word and delete before forwarding to the next block.

Encoder 64/66B and Decoder 66B/64B

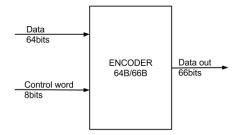


Figure 11: Ethernet Encoder block

The Encoder block is represented as shown in Fig. 11. The 8 octet data is encoded using the control word into 66 bit output. The first 2 bits of the output are called sync header which is used for the synchronization from the receiver. The sync header "10" correspond to data and "01" corresponds to control codes

In the decoder the sync header is used to synchronize the 66bit data. The process of decoder is the reverse interpretation of the encoder module.

Scrambler and Descrambler

This block is used to randomize the signal so that long sequence of 1's and 0's are eliminated. This is performed using the Scrambler polynomial.

The De-scrambler will take the scrambled input and will output the unscrambled data.

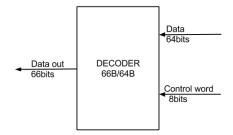


Figure 12: Ethernet Decoder block

Gearbox

This block is used to switch output rates. The incoming data is transmitted at different rates based on the clock frequency.

2.4.2 Xio-s

Xio-s is Ericsson proprietary protocol used for communication between chips.

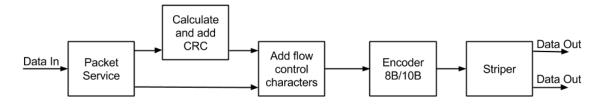


Figure 13: Xio-s Transmitter

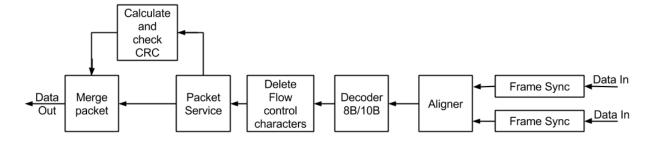


Figure 14: Xio-s Receiver

There are 7 different types of packet services for Xio-s protocol. This assigns the packet to proper channel based on the service type. As an example, if the service type is format 1 then it uses only CRC16 channel or if the service type is format II then it uses both CRC16 and CRC32 channels.

	Service type [4:0]	Data	CRC [15:0]	
--	-----------------------	------	---------------	--

Figure 15: Xio-s Frame format Type 1

	Service type [4:0]	Data 1	CRC 1 [15:0]	Data 2	CRC 2 [31:0]	
--	-----------------------	--------	-----------------	--------	-----------------	--

Figure 16: Xio-s Frame format Type 2

CRC

There are 16bit and 32bit CRC calculations required in Xio-s protocol. So the polynomials for CRC16 and CRC32 are stored in the memory and the function is used according to the service type.

In the receiver, the CRC is again computed and compared with the received bytes.

Flow control characters

The flow control character will add the control word similar to Ethernet protocol in each channel. These are used for indication of start and end of frames.

In the receiver the flow and control characters are identified and deleted.

Encoder/Decoder

The encoder module will encode one octet at a time to 10bits. So for 8 octects it outputs 80bits The encoder 8B/10B is invented by IBM and famous for short run length and DC balance.

The decoder module does the reverse of encoder and thus the output of decoder will be the same as the input of encoder.

Striper

It is used to split the 80bits incoming data into 40bits of 2 physical channels to increase the data transfer rate.

Frame Sync

This block is used to synchronize the receiving data. It is performed using the special character in the transmitted data called as k28.5 character.

Aligner

This block is used in the receiver if striper is used in the transmitter side. It aligns the two incoming 40bits channels into one 80bits channel.

2.4.3 CPRI

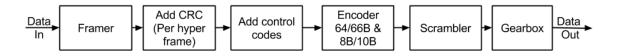


Figure 17: CPRI Transmitter

CPRI is an industry co-operation aimed at defining a publicly available specification for the key internal interface of radio base stations between Radio Equipment Control (REC) and the Radio Equipment (RE) [35]. It is the co-operating work of Ericsson AB, Huawei Technologies Co.Ltd, NEC corporation, Nortel Networks SA and Siemens AG. All the blocks in this protocol are similar to the blocks in Ethernet protocol. The frame structure is similar to the Xio-s protocol.

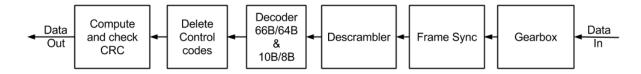


Figure 18: CPRI Receiver

2.5 Grammar and Language

A Grammar is used to describe the syntax of a language, that is, all possible legal sentences or combination of words that make up the language. More formally, Grammar G describes all allowed legal sequences of strings. This is called the language(G) of the grammar. The language in turn is made of sequence of elements which can be letters, numbers or special symbols. For example, in the word "KTH" the capital letters K, T, H are to be recognized and then the word needs to be formed.

This is performed by the Lexer which recognizes the letter and forms the token. To be able to recognize words, Lexer need some special constructs. These special constructs make-up the language that can be recognized by regular expressions. For example, regular expression [0-9] recognizes a single letter in the range from 0 to 9.

2.5.1 Parser

The Lexer scans the input character streams and forms the valid tokens. The Parser takes tokens as inputs and then based on the parsing rules in the grammar, decides the parsing strategy. The parser output can be used either to create an interpreter or a compiler.



Figure 19: Parser

Here, a language application is built to output the reconfigurable hardware architecture. Hence the parser output is used to translate it to the required output.

2.5.2 Backus-Naur Form

In computer science world, Backus Naur Form (BNF) is the notational technique for context free grammars. It is a set of derivation rules to define the language.

For example,

$$\langle \text{ int} \rangle ::= \langle \text{ DIGIT } \rangle \mid \langle \text{ int } \rangle \langle \text{ DIGIT } \rangle$$

$$\langle \text{ DIGIT } \rangle ::= [0-9]$$

In the above Grammar, \langle int \rangle on the left hand side is called as non terminal and the \langle DIGIT \rangle is called as terminal. So the sequences of digits like 9999... can be parsed by representing grammar as above BNF code. An extension to BNF grammar with more operators to write the syntax is called as EBNF. The grammar above can be rewritten in EBNF as below

$$\langle \text{ int } \rangle ::= \langle \text{ DIGIT } \rangle *$$

Here "+" means one or more occurrences of digits. Similarly "*" operator means 0 or more occurrences.

3 Modeling

This chapter explains the research process. It further explains in detail about the system level modeling and its languages. It is followed by the section which explains the verification of the developed system. Why and how a new language is developed for programming the architecture at Ericsson is explained in the next section. The last section gives the configurable functions in protocol processing to achieve a reconfigurable hardware architecture.

3.1 Introduction

The hardware architecture and its interconnections need to be understood to design the protocol processor. The reconfigurable architecture developed at Ericsson AB is called Freyja architecture. The hardware architecture details has to be abstracted out to define it in high level description. Similarly the protocols details has to be abstracted using the same description to reconfigure the hardware for different protocols. Mapping between them requires a custom set of configuration data to define and reconfigure the architecture according to the requirement. To accomplish this, the research involves the following major steps:

- Study the Reconfigurable hardware architecture
- Identify the common protocol functions
- Develop a grammar to describe the details in high level language
- Test the developed system

To study the reconfigurable hardware architecture developed at Ericsson, the language used to develop the architecture should be known. Why such a language is used to develop the architecture need to be understood. So the next section will provide details about the different system level modeling languages and their abstraction details.

3.2 System-Level Modeling

Hardware system requires modeling concepts that allow designers to explore different aspects of the system. These concepts include communication, time, structure, hierarchy etc. As explained in Fig. 1, there are different domains of abstraction and

views in which development phases varies. For example in software, designers carry performance exploration to identify bottlenecks with respect to time, excluding the lower level hardware details. Y chart maps the different modeling styles of hardware as 3 different domains. In order to represent the hardware in these domains at different abstraction levels, an executable model is required. The designer should be able to model different aspects of hardware, it should be scalable, possible to simulate, test and verify the functions. The Register transfer level description of the hardware is synthesized into gate level netlist by logic synthesis tools. This level of hardware abstraction describes signal transition between synchronously clocked registers. Traditional modeling languages have emphasized the hardware design flow, whereas the recent languages have incorporated concepts for efficient modeling at system level. With traditional languages like VHDL/Verilog which are more used for rtl level, it takes too much time to develop for architectural exploration and software development. They also have poor performance i.e takes longer time to simulate complex systems and are not available early in the development phase. In this section, Transaction Level Modeling (TLM) is introduced which is an alternative to traditional approaches for efficient abstraction and exploration of architectural details. Transaction level modeling is supported in languages such as SystemC and SpecC by abstract channels that connect communicating modules.

3.2.1 Language comparison

Several languages have emerged to address various aspects of system design. Fig. 20 illustrates the comparison of different languages. Although Ada and Java have proven their value, C/C++ is predominately used today for embedded system software. The Hardware description languages, VHDL and Verilog, are used for simulating and synthesizing digital circuits. With the increase in design complexity, with multi million gates being fabricated, the increase in pressure to get design out faster with first time design success is also higher. Many new languages are used in industries which helps in designing the higher abstraction models of hardware to be used for function verification and software validation. Systemverilog helped the designer with more constructs compared to Verilog, which address many hardware-oriented system design issues.

SystemC is an ANSI standard C++ class library for system and hardware design for use by designers and architects who need to address complex systems that are hybrid between hardware and software [36]. SystemC provide an event-driven simulation kernel in C++, together with signals, events, and synchronization primitives, deliberately mimicking the hardware description languages VHDL and Verilog. Matlab and several other tools are widely used for capturing system requirements

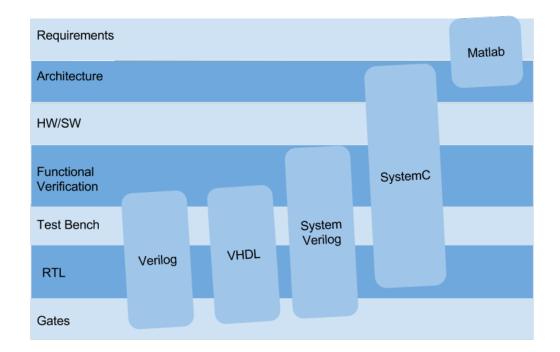


Figure 20: Hardware description languages and abstraction levels [4]

and developing signal processing algorithms.

3.2.2 Transaction Level Modeling

Transaction level models use abstract channels to model communication between concurrent processes in the system using function calls [37]. In RT level each bit is instantiated as a port and hence in TLM level it can be abstracted to a byte, word or according to the architecture requirement. Also the signal assignments of individual bits in RT level are abstracted with function calls which either sends or receive the transaction. A transaction can be defined as exchange of information between processes in the simulation. It may contain complex data structure with control and data information, burst of data or a simple word of data packet. Time within TLM components are modeled as untimed, approximately or cycle accurate [38].

Transaction level modeling complements RT-level modeling, in order to perform following activities [4]:

 Hardware micro-architecture exploration and starting point for more detailed hardware modeling.

- System level architectural exploration, such as selecting communication and processing components and HW/SW partitioning.
- Virtual platform for software development.
- Reference model for hardware functional verification.

Transaction level modeling is supported by languages like SystemC and SpecC. The communicating modules which are either initiators or targets are connected through channels. An initiator will start sending the transaction while the target will respond to the initiator request. Arbitration and routing algorithms are necessary for multiple initiators and targets to communicate in the system. The arbitration algorithm selects the initiator to be given access to the channel, and the routing algorithm assures that the correct target is addressed by transaction information. All these concepts are used to model the interconnects found in SoCs.

3.2.3 SystemC

SystemC is a system design language that has evolved in response to a pervasive need for a language that improves overall productivity for designers of electronic systems [4]. The open SystemC initiative (OSCI) formed in 1999, maintained a simulation library for SystemC. In December 5.2011, Accellera Systems Initiative was formed by the merger of Accellera and the Open SystemC Initiative. It is a C++ library that contains routines and macros to simulate concurrent processes using HDL like semantic. SystemC is now the IEEE Std. 1666 for system-level modeling, design and verification. It offers real productivity gains for the designer by providing hardware and software components design in parallel, but at a high level of abstraction. This higher level of abstraction gives the design team a fundamental understanding early in the design process of the intricacies and interactions of the entire system and enables better system trade offs, better and earlier verification, and over all productivity gains through reuse of early system models as executable specifications [4].

Fig. 21 illustrates the layered architecture of SystemC. The shaded blocks are the SystemC core language standard, build upon standard C++. The layers above the standards are additional libraries available for system-level design and verification. A SystemC module encapsulates processes, which describe behavior, and communicates through ports and channels with other SystemC modules. Processes are used to describe concurrency and wait-statements are used to halt process execution for a specific time or until an event occurs. The OSCI SystemC TLM library contains ports, interfaces, channels, and also data structures used to represent request and response in an initiator-to-target communication scenario. The C++ notion of

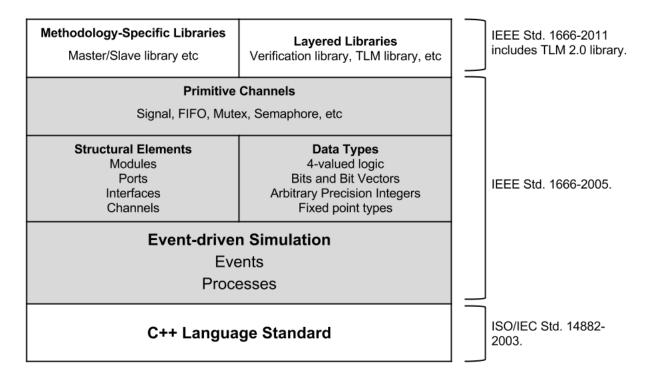


Figure 21: Layered SystemC Architecture

header (.h file) is used for entity and the notion of implementation (.cpp file) for the architecture. Hardware by nature is concurrent and modeling it through simulation on a host processor is an illusion. It is accomplished by simulating each concurrent unit (defined as SC_METHOD, SC_THREAD or SC_CTHREAD). Each unit is allowed to execute until simulation of the other unit is required. The simulation of concurrency is the same for SystemC, Verilog, VHDL or any other Hardware Description Languages. The simulator kernel will handle these tasks.

In TLM2.0, an *initiator* is a module that initiates the transaction and a *target* is a module that responds to transaction initiated by other module. The same module can act as an initiator and as a target for example in the case of modeling a bus or a router. TLM2.0 uses sockets to send/receive transaction.

An interconnect is a component which doesn't modify the transaction but merely forward the transaction. Fig. 22 illustrates the different ways of function calls for data exchange through transaction. The TLM2.0 library utils provide the simple initiator and simple target sockets for user convenience. These are used in modeling the reconfigurable architecture discussed in further section together with sockets which allow transactions to be sent to multiple destinations.

The transaction being routed can be represented by a simple data structure

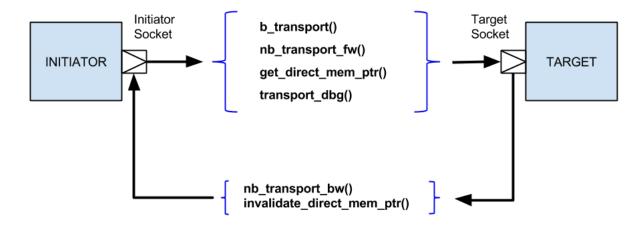


Figure 22: Initiator and Target sockets

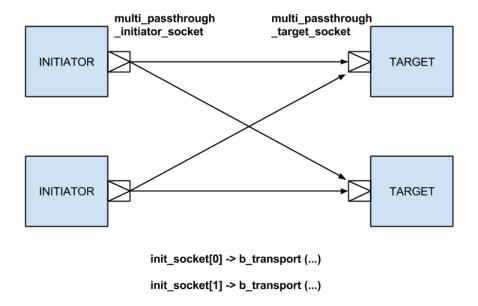


Figure 23: Many to many binding

provided by TLM2.0 utils called as Generic payload. It has standard set of bus attributes like command, address, data, byte enables, streaming width, and response status. The command can be either write or read, data attribute points to a data buffer within the initiator and the data length attribute will give the length of the data array. The streaming width attribute specifies the width of a streaming burst where the address repeats itself. This Generic payload is modeled in each of the reconfigurable operator blocks for initiating the transaction and the attributes are

handled accordingly.

3.3 Verification Methodology

While in the early days, digital designs were verified by looking at waveforms and performing manual checks. This is a very tedious and time consuming task for the present day complex systems. So automating the verification process using better and efficient framework is always preferred. The SystemVerilog language came to aid many verification engineers. Its features like classes, covergroups and constraints have helped verifying complex digital systems. Later Verification Methodologies started to appear. UVM is one of the methodologies that was created for the need to automate verification. UVM is a verification methodology based on the best features of Open Verification Methodology (OVM) and Verification Methodology Manual (VMM) [39].

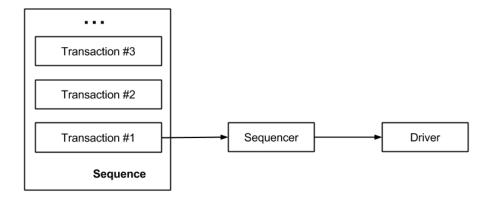


Figure 24: UVM Sequencer

The first step in verifying the designed system is by defining what kind of data to be sent to the design under test. Detailed architecture of UVM is not discussed as it is out of the thesis scope. Only features relevant to the work are summarized. A transaction can be built form uvm_transaction or uvm_sequence_item classes similar to the transaction in SystemC. The "random" keyword is used to generate transaction according to the constraints. The sequencer is responsible for sending the transaction to the driver component which in turn connects to the Design Under Test.

3.4 High level description of the protocol

In general, how the reconfigurable architectures are programmed and the related work in this field is discussed briefly in Section 2.2. wherein it explained different tools and framework developed for computationally intensive DSP architectures. As coarse grain architectures are targeted for specific domain to optimize the architecture design, the synthesis tools designed for protocol processing are discussed here.

3.4.1 Related work

As discussed in previous section, SystemC enables the specification of hierarchically structured communication protocols using the concepts of interfaces and channels. SystemC ^{sv} [40] discuss the way to model the complete protocol environment by extending the SystemC language. It allows specifying the protocol which generates the controller for producer and consumer and send the data through abstract channels. The tool is called COSYNE (Controller Synthesis Environment).

Clairvoyant [41] describes the protocols using grammar based specification. It will generate an FSM described in VHDL which can be used by the logic synthesis tools. It was further commercialized and extended as Synopsys Protocol compiler. ProGrIL [42] is the Protocol Grammar Interface Language based on Context free LL(1) grammar. The configuration file will contain the information about the widths and depths of various parts of the grammar processors such as buses and memories.

ProGram [43] [44], input specification is similar to the production based specification and the output is VHDL code of register transfer level. It is distinct in specification description when compared to the other approaches wherein the production rules are independent of port widths. The port widths are specified as design space constraints and the synthesis tool will generate different designs with varying performance for different values of constraints.

All the above discussed methodologies tries to develop a generic tool to map the protocol described in high level description directly to register transfer level. This includes the synthesis tools exploring design space for the optimized hardware architecture with respect to area and performance. As discussed in Section. 2.2, for the computationally intensive DSP applications, there were tools which maps the high level description into specific architectures. Many case studies related to custom coarse grain architectures were also discussed which developed their own set of kernels, software, configuration data for programming such architectures. The architecture being developed at Ericsson is a custom architecture which requires a programming tool to generate the reconfiguration data and program it according to user definition. The architecture components are implemented in SystemC language and it requires a new tool to program it for different configurations. A new high level language is developed to describe the custom architecture.

High level language description requires the design of Lexer and parser as explained in Section 2.5. They can be designed using languages like C, C++, java, C# and more. Based on the application requirement, a parsing strategy needs to be decided. As the machine resources grew in today's world, researchers have developed more complex and powerful Non Deterministic parsing strategies. Today both "bottom-up" and "top-down" approaches exist. Debugging a "bottom-up" parsers are hard to understand and debug compared to the "top-down" parsers.

There are many tools like ANTLR4, APG, AXE, YACC etc which helps to build a language application. They provide the user with Lexer and Parser implementations for the matched grammar defined by the user. This helps in reducing time for building the language in a short time. Antlr is one such tool which helps in designing the recursive descent top-down parser and a clear error recovery mechanism.

3.4.2 Antlr

ANTLR4 accepts as inputs any context-free grammar that does not contain indirect or hidden left-recursion [45]. Antlr4 generates a recursive descent top down parser. It uses ALL(*) production prediction function. ALL(*) prediction mechanism launches sub parsers at decision point and they operate in pseudo parallel to explore all possibilities of input combinations. Antlr 4 currently generates parsers in Java or c# and the previous version supports even C and C++.

Antlr4 grammar use yacc-like syntax with EBNF operators like Kleene star(*) and token literals in single quotes. Both Lexical and syntactic rules are specified in the same grammar file. The Lexical rules are specified in capital letters which distinguishes them from others.

Fig. 25 illustrates ANTLRs yacc-like metalanguage. Antlr4 automatically rewrites the rule to be non left recursive and unambiguous. The grammar analysis is performed at parse time and caching results in lookahead DFA for efficiency [45].

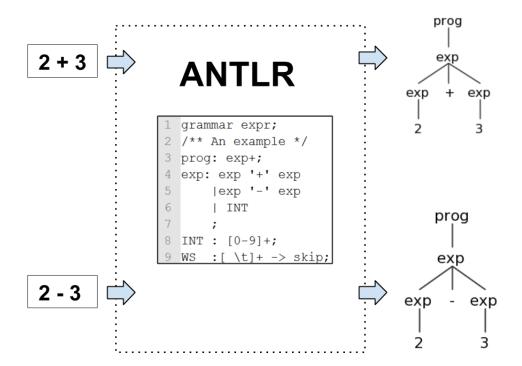


Figure 25: Antlr: Input, Grammar and Parsetree

Parse tree listners and Visitors

Antlr provides 2 tree walking mechanisms in its runtime library. In parse tree Listners, ANTLR generates a parsetreeListner subclass specific to each grammar with entry and exit methods for each rule. This is suitable for applications wherein the complete tree need to be invoked from root till the last leaf node in-order. This is the default method in ANTLR. The advantage of using parsetreeListner mechanism is that it generates automatic APIs for the walker sequence and it is easier to build the language application.

In many circumstances, the designer need to control the walk sequence. It is not possible to design a language application with visitor method which generates APIs from left to right leaf node. The parse tree Visitor mechanism is used when the tree walking needs to be controlled. Option -visitor makes ANTLR4 to generate a visitor interface from a grammar with a visit method per rule. Then in the application-specific code, a visitor implementation can be called.

3.5 Comparison of 3 Protocols

The parser generator tools need to be used to describe the protocols and produce the reconfigurable files for the hardware architecture. The protocols of the MAC layer considered in this work are explained in section 2.4.

The common functions in each protocol are highlighted with the same color for the boundary line in Figure 26 & Figure 27. Each of the operators are implemented in SystemC language and they receive and process the TLM2.0 transaction to perform the corresponding tasks.

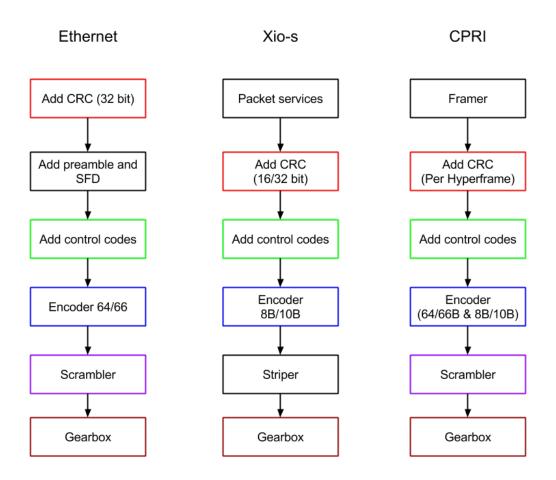


Figure 26: Transmitter of all 3 protocols

The operators can be multiplexed and a generic operator unit can be designed for the blocks with same color boundary line. For example, the FCS computation is required in all 3 protocols. They require 16/32 bit polynomial to compute the check-sum value. The memory unit is stored with polynomials and control data required for informing the CRC operator to perform a CRC16 or CRC32 operation. Since the CRC check operation performed in the receiver stage also require the computation of FCS, a single operator unit can be designed and configured to perform different operation. The CRC operator will be designed such that it is used to transmit the received transaction by appending the FCS value according to the protocol and if the transaction is from receiver stage then the same hardware is used to compute the checksum value and then compared with the received checksum value. The operator switches its context based on the control information received form memory block. A common hardware architecture can be designed with minimal reconfiguration to perform the protocol processing of different protocols stated above.

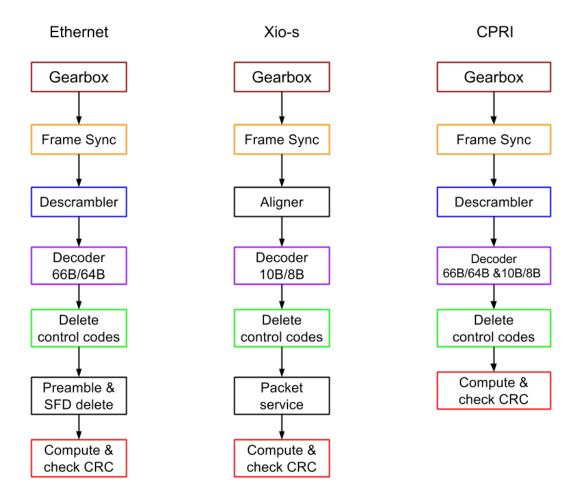


Figure 27: Receiver of all 3 protocols

4 | Freyja

The reconfigurable architecture called Freyja is explained in detail in the first section and then a language description is developed to define the Freyja architecture which maps different protocol operators and their interconnections. The high level description accounts for different operator instantiation, transaction routing, constants in memory and error handling. Each of these is explained in detail in this chapter.

4.1 Freyja Architecture

As in Figure 28, Freyja architecture is a reconfigurable protocol processor. It consists of different protocol operators which are connected through the central switch. This switch based network topology can be configured to process the data based on protocols. The Ring bus (RB) interface the Freyja with the higher layers of protocol and it issues the data frames of different protocols as tokens. The Common Memory Interface (CMI) is used to fetch the data from memory. The physical interface is represented by the Serializer/Deserializer block i.e SERDES. The details of each of the Freyja architecture blocks are discussed below.

4.1.1 Switch port

Freyja Switch port is the smallest unit in the architecture which receives the transactions from different functions and then forwards it to the internally connected next switch port based on the destination of the transaction. It consists of a simple initiator socket, simple target socket, Multipass through initiator socket and Multi pass through target socket. It is represented as in Figure 29

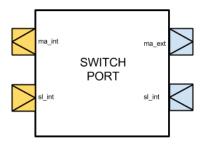


Figure 29: Switch port

4.1.2 Operator units

Freyja Operator units are designed to contain a control, process and memory blocks. Each operator function is implemented in the process block and the transaction routing and context switching is performed in the control block. The memory block stores the constants required for the process

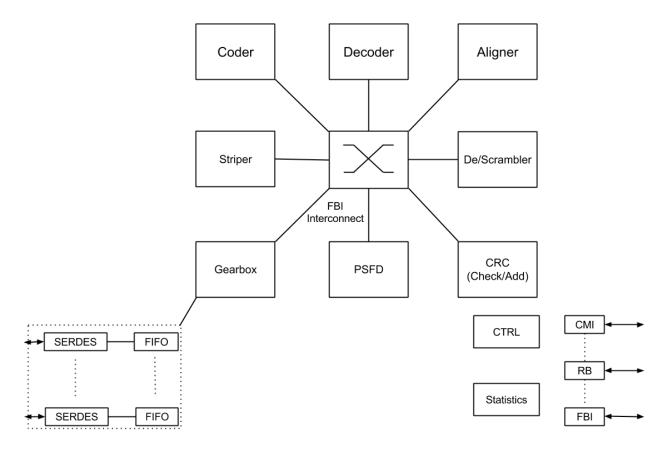


Figure 28: Switch based Freyja network topology

block. It can also be used to store the result and then the control block can access the results. As in Figure 30, the 3 components of each operator is encapsulated with one simple initiator socket and one target socket which initiates the transaction and receives the transaction from the switch respectively. The transaction received is routed to the control block where it sends the transaction to memory block about the received transaction. Based on the required operator, the memory block will send the transaction to the process block. Meanwhile the received transaction is sent to the process block from the control block to perform the required operation.

4.1.3 Switch wrapper

The switch wrapper instantiates the switch ports and their interconnections. Based on the number of operators, the required number of switch ports is instantiated.

As in Figure 31, each switch port can send the transaction to any of the other switch port through the internal multi pass through socket. The other switch ports can receive the transaction using the multi pass through receive socket. A trans-

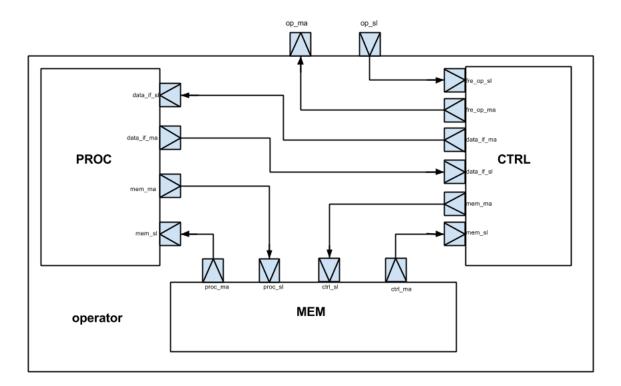


Figure 30: Freyja operator

action source and destination cannot be the same operator as there is no such interconnection.

4.1.4 Overall architecture

Assuming there are 4 operators in the Freyja architecture, an illustration of the system is shown in Figure 32.

The overall Frejya architecture consists of switch wrapper instantiating switch ports and the operator blocks for each of the operator functions. A transaction originating from operator 1 as shown in Figure 32 with orange box can be routed to any of the other operators as destination based on the address of the payload header data. This will be assigned in the control block of the operator where the transaction is originated.

4.2 Parser implementation

Antlr tool is used as a parser generator to develop the language. The grammar is defined using EBNF. Antlr parser use a new parsing technology called Adaptive LL(*). This parsing strategy combines the simplicity, efficiency and predictability of

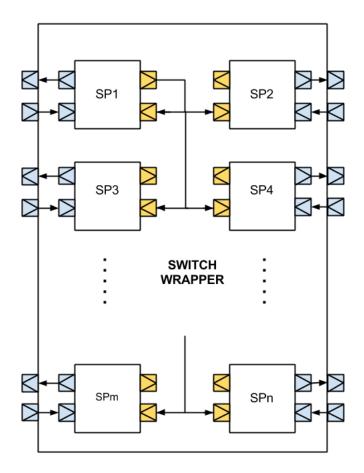


Figure 31: Freyja Switch Wrapper

conventional top-down LL(k) parser with the power of GLR like mechanism to make parsing decision. It performs the grammar analysis dynamically at runtime rather statically. The grammar analysis is moved to parse time which helps to handle any non-left recursive context free grammar.

A brief description of the generated file are stated below:

Freyja_archInitparser.java: contains the parser class definition specific to grammar Freyja_arch that recognizes our Freyja protocol processor language syntax.

Freyja_ArchInitLexer.java: This file contains the lexer class definition by analyzing the lexical rules in the grammar.

Freyja_archInit.tokens: Antlr generates a token type number to each token in the grammar and store these values in this file.

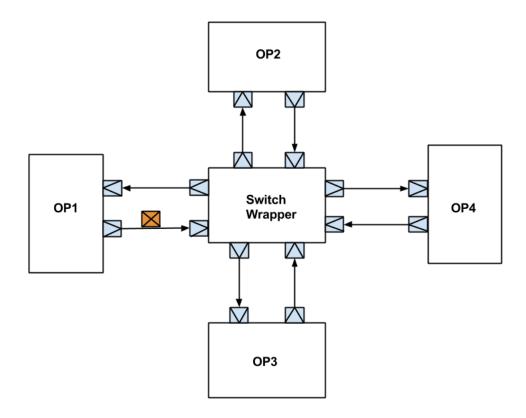


Figure 32: Freyja Architecture for 4 operators

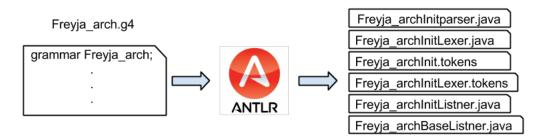


Figure 33: Antlr generated Files

Freyja_ArchInitListner.java, Freyja_archBaseListner.java: Antlr parser builds a tree walker that can trigger the callback events to a listener objects. Freyja_archInitListner is the interface that describes the callbacks and Freyja_ArchBaseListner is a set of empty default implementations.

The backend functions are implemented to produce the Freyja architecture based

on the language described.

4.3 Operator Instantiation

As explained in the Figure 34, each of the operator in Freyja architecture consists of a control block to take of the transaction routing and error handling and a memory block for storing the constants and results and the process block to perform the operator function.

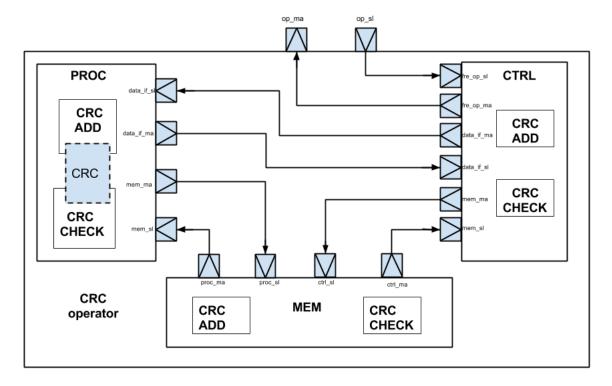


Figure 34: CRC operator in Freyja Architecture

In a protocol processor consisting of operators for transmitting and receiving the data frames, the functions performed during the transmitting stage might need to be performed in the receiver stage. For instance, FCS is appended during the transmitting stage and in the receiver stage the computation of CRC is performed again. In addition, it performs the comparison of the received FCS value and the newly computed value to identify if the received data is erroneous or not.

The control block of each operator has to identify the destination operator within the process block to forward the payload to exact operator. The language described has to consider this feature when instantiating the operator and defining the control blocks for each operator. An example of CRC block with CRC ADD and CRC check operators are discussed further.

As in Figure 34, the CRC function to compute the CRC is used by both CRC ADD and CRC CHECK operators. The control block has to check if the destination operator is CRC add or CRC check and forwards the data to be processed to the correct operator and also should indicate the memory block to forward the required constants for the process blocks. This is further explained in the next section.

The operator is defined as below:

```
Operator CHECKSUM{

sc_name : crc;

op_type : crc_add,crc_check;

op_mem : 0x01,0x04,0xc1,0x1d,0xb7,0x20,0x00,0x04,0x03,0x04,0x00,

0x00,0x01,0x59,0x53,0x10,0x03,0x02,0x04;

errorid : 02,03;

}
```

Figure 35: Freyja Operator Instantiation

The operator name CHECKSUM identifies the unique operator in the Freyja architecture. The sc_name field is used to create the systemC file with fre_op_ctrl_crc, fre_op_proc_crc, fre_op_mem_crc files which are the control, process and memory blocks respectively. Each operator definition also takes care of including a switch port and modifying the switch wrapper functions to take care of the interconnections.

4.4 Context switching

The context switching in the process block is performed based on the destination operator for the data packet. The interconnect information is by the control block to determine the protocol and destination operator for the data packet. The control block will then forwards the transaction to the process block where the context switching between the operators are performed. The operator type field will indicate the process block to switch between the operators.

The control block also indicates the memory block to send the required constants for the process block. For the CRC operator, as in Figure 34, the control block will determine the protocol and the operator type and then informs the memory blocks 4 FREYJA 4.5 Memory

with the transaction payload. The control block will send the received transaction to the process block. The memory block will decodes the operator and then sends the constants required for the operator through a transaction. The process block will do the context switching of operator between CRC add and CRC check to obtain the result according to the destination operator.

Once the results are obtained, the process block will send it to the control block. The control block will forward the transaction to the next block.

4.5 Memory

The memory block is designed to store the constants required for the operators and also to store the results. But currently the results are not stored in memory, as only one protocol is implemented.

The constants to be stored in the memory are indicated in the op_mem field. Each operator is allowed store all the constants in the memory and based on the operator in use the constants can be forwarded to the process block through a transaction. As in operator checksum definition, the memory contains 19 bytes which are used by the CRC16 and CRC 32 operators.

Memory transaction for each protocol are described in the language as below

```
memory {
crc_add :00 to 08;
crc_check :00 to 07,09;
scram :00 to 07:
descram :00 to 07;
}
```

Figure 36: Freyja Memory content definition

As in code above the first 8 bytes of contents are sent as a transaction from memory for the CRC ADD operation in CHECKSUM operator. The different forms of notation can be observed in the CRC check wherein the memory bytes can be specified with individual address. Each protocol with the memory code as above can send transaction for different operators. In the above code, the crc add,crc check,scram and descram send the transaction to their respective blocks.

4.6 Transaction handling

The transaction routing for each protocol is described in the input file using the below syntax

CONNECT: <source operator> to <Destination operator>

Based on these all the control blocks will be configured with the destination address for the payload. The switch interconnect will send the transaction to the next operator based on this address. This mapping defines the transaction routing for each protocols and is used to build the destination address automatically using the input description.

4.7 Error handling

Any erroneous packets can be sent to the error handler from each of the operator. The control block will check for the process block result, if there are any errors then it informs the control block through the FBI header and the control block will send the payload to error handler for further processing and will not be sent to the destination operator.

The error id for each operator can be implicitly assumed or even possible to mention as an explicit number.

5 | Analysis

This chapter explains the analysis of the framework through the integration of Ethernet protocol into the Freyja architecture. Ethernet operators are integrated into the system of files generated from the Antlr tool. To show the context switching between different protocols, Xio-s CRC16 operator is also integrated. The first section analyses the process and results of Ethernet integration and section 6.2 explains the Xio-s protocol integration.

5.1 Ethernet

Ethernet Transmit and Receive operations are explained in Fig. 9 and Fig. 10. The *Datain* in the transmit is passed from the UVM test bench. The UVM sequence is generated with first byte indicating the protocol and the second byte containing the address of the operator. Once the transaction reaches the Switch wrapper, it is decoded and forwarded to the first operator as defined by the input file.

Appendix A shows the complete description of Ethernet protocol. The interconnection between different Ethernet operator is explained in this section.

A part of the Ethernet description is shown in code below:

```
Tx_PATH {

CONNECT: proc_tx to crc_add;

CONNECT: crc_add to pre_sfd_add;

CONNECT: pre_sfd_add to cc_a;

CONNECT: cc_a to enc;

CONNECT: enc to scram;

CONNECT: scram to gb_tx;

CONNECT: gb_tx to gb_rx;

}
```

Figure 37: Freyja protocol Interconnection

The operators required by the protocols are first instantiated by defining them in the description. The string "protocol" indicates the definition of new protocol in Freyja architecture. The new protocol definition consists of 3 segments: Fields, Tx_path, Rx_Path.

• The Fields section is defined to indicate the details of transaction contents in each protocol

5 ANALYSIS 5.1 Ethernet

- Tx_Path is used to define the transmitter sequence and
- Rx Path is used to define the Receiver sequence

The Tx_Path and Rx_Path are for the user to distinguish between the Transmitter and Receiver sequence for each protocol. In Freyja architecture the interconnections have no difference w.r.t Tx_Path and Rx_Path. The interconnect information is used to code the control block of each operator to forward the transaction to the correct destination.

A part of control block of CRC operator for the description of Ethernet protocol is shown below:

```
//part of fre_op_ctrl_crc.cpp block

if (type == ethernet) {
    if (op_type == crc_add) {
        data[0] = ethernet;
        sz = sz + 2;
        data[1] = pre_sfd_add;
        gp.set_address(crc_add_2_pre_sfd_add);
    }

else if (op_type == crc_check) {
        data[0] = ethernet;
        sz = sz + 2;
        data[1] = proc;
        gp.set_address(crc_check_2_proc);

}
```

Figure 38: Freyja CRC operator Control Block code segment

In the above code the CRC operator's control block initially check for the protocol and then for the type of the operator. As explained in Fig. 34,the CRC operator multiplex 2 operations. CRC add and CRC check operators are having the same destination operator control block. They share the common memory,hardware for CRC computation and control function. The extra logic required to perform the CRC add and CRC compare are different.

The destination address of the transaction is obtained from the input file. This information is compiled in the form of constants in SystemC environment. The same information can be stored in shared or in the operator memory based on architecture requirement.

5 ANALYSIS 5.2 Xio-s

The parser implemented will generate these constants which are required for the reconfiguration of hardware. The context here is the possibility to send the transaction to different operators from the same source.

```
//constants for the protocol : ethernet
  const unsigned char ethernet = 0;
  const int proc_2_crc_add=0;
  const int crc_add_2_pre_sfd_add=1;
5 const int pre_sfd_add_2_cc_a=2;
 const int cc a 2 enc=3;
       int enc 2 scram=4;
 const
       int scram_2_gb=5;
  const
       int gb_2_descram=5;
       int descram 2 \text{ dec}=7;
 const
       int dec_2_cc_d=8;
  const
  const
       int cc_d_2_pre_sfd_del=2;
       int pre_sfd_del_2_crc_check=1;
       int crc check 2 proc=0;
  const
```

Figure 39: Freyja Reconfiguration constants

The constants are formed using simple syntax as below

```
\langle source operator \rangle _2_ \langle Destination operator \rangle
```

The constant value assigned is calculated by the parser based on the Multipass through ports of switch port to which the source and destination operators are connected.

The CRC operator also contains 19 bytes of memory elements. These 19 bytes are used by both CRC32 and CRC16 Operators. Ethernet protocol uses only CRC32 and hence only crc_add and crc_check uses first 9 bytes of the memory elements. The parser supports the access of contiguous or individual element. After the control block informs the memory about the protocol and operator, the memory block will send the transaction to the process block which includes the constants required by the operator.

5.2 Xio-s

Xio-s protocol is breifly explained in section 2.4.2. The Xio-s protocol CRC block is integrated into the system. The control block will check for the type of protocol, prioritized according to the description in the input file. Similarly Operators are prioritized in the order they have been defined in the input file. Defining Xio-s after

5 ANALYSIS 5.2 Xio-s

Ethernet protocol description will change the control block code segment as shown below.

```
//part of fre_op_ctrl_crc.cpp block
if (type == ethernet) {
    if (op_type == crc_add) {
        ...
}
else if (op_type == crc_check) {
        ...
}
else if (type == xios) {
    if (op_type == crc_add) {
        ...
}
else if (op_type == crc_add) {
        ...
}
else if (op_type == crc_check) {
        ...
}
```

Figure 40: Freyja CRC Ctrl block with 2 protocols

As illustrated in Fig. 13 and Fig. 14, Xio-s protocol requires 2 interfaces to transmit the transaction from one operator to multiple operators. The transactions are sent from packet service block to two other operators both in transmission and during receiving stage. This feature is included in the input description. The Tx_path and Rx_Path in each protocol description will allow the user to define multiple destination operators for the same source operator. The code below depicts a scenario of Xio-s protocol assuming ps, crc_add_16 and cc_a_xio are the operators described in the input file.

```
Tx_PATH {
CONNECT: ps to crc_add_16;
CONNECT: ps to cc_a_xio;
}
```

Figure 41: Freyja One source to multiple destination interconnection

Similar logic applies to the receiver section which can be parsed to receive the transaction from different source operators. Assuming **fs1**,**fs2** and **aligner** are Xios protocol operators, the below code indicate the way to receive transaction from multiple sources.

5 ANALYSIS 5.2 Xio-s

```
Rx_PATH {
2 CONNECT: fs1 to aligner;
3 CONNECT: fs2 to aligner;
4 }
```

Figure 42: Freyja multiple source to one destination interconnection

But the Freyja Operator as explained in section 4.1.2 consists of only one interface. This need the change in architecture of operator units and a single interface will not be possible to model the Xio-s protocol. The operators can be customized by defining more strings in the operator instantiation stage. Once the architecture is modeled, the parsed input description can be used to output the architecture to match the interface requirement.

6 | Test system

This chapter describes the test system, testing strategy, input protocol file, discussion about output obtained.

6.1 Complete Test System

As in Figure 43, the first switch port is connected to the UVM test bench. The UVM environment will drive the input signals according to the Freyja architecture. The Freyja interconnect consists of 4 bytes header field and the data payload as in Figure 44. In SystemC implementation the first byte is considered to hold the unique protocol id and the second byte to have the unique operator id, the third byte and fourth byte are for Flow control and Context/error handling.

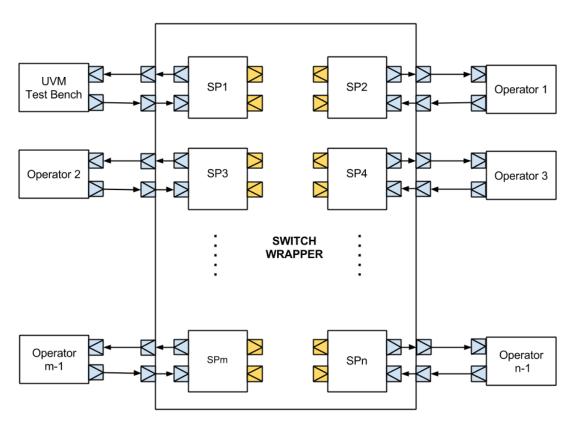


Figure 43: Complete Test System(*Multiport interconnections not shown)

The data bit width is suggested to be 80bits. This complete data is randomly

generated with constraints from UVM sequencer. The UVM driver will drive the input signals to the Freyja architecture. The data is sent as a transaction payload.

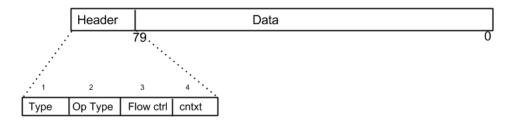


Figure 44: FBI Interconnect

6.2 Payload and blocking transport

Each generic payload transaction has a standard set of bus attributes: Command Address Data Byte enables Streaming width Response status The default values are set for the unused attributes. The address is initially set in the test bench and in the Freyja architecture; each operator control block will modify the address field based on the destination operator address. The data field is set to the result of the operator process blocks.

6.3 Input and output

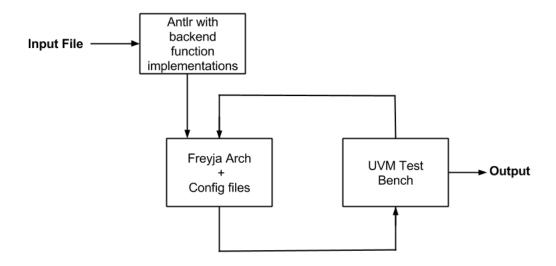


Figure 45: Input and Output of the system

The input file is the description of the protocol using the language defined by the EBNF grammar. The input file is parsed by the Antlr tool; the backend functions are triggered during parsing to output the Freyja architecture and the reconfigurable files. The UVM test bench will send the transaction to the Freyja architecture. The transaction is initiated from the UVM to the Freyja architecture and the final transaction is received back in the UVM. An example of instantiating 4 operator blocks as in Figure 32 is shown below:

```
Operator OPERATOR1
2 sc_name : op1;
3 op_type : op11,op12;
4 errorid : 01,02;
5 }
  Operator OPERATOR2{
8 sc_name : op2;
9 op_type : op21,op22,op23;
10 \text{ errorid} : 03,04,05;
11 }
   Operator OPERATOR3
   sc name : op3;
14
15 op_type : op31;
16 errorid : 06;
  }
  Operator OPERATOR4{
20 sc_name : op4;
op_type : op41,op42;
22 errorid : 07;
23 }
```

Figure 46: Freyja Overall architecture definition

6.4 Timing annotation

The transaction is sent through the socket using the b_transport method of the TLM2.0 blocking transport interface, which passes its transaction arguments by reference and has no return value. The b_transport also carries the timing annotation which is not configured as the main interest as of now is to model the functionality of the target and not modeling any timing detail.

7 | Conclusion and Future work

This chapter concludes the thesis by describing the learning outcomes of reconfigurable architectures used for protocol processing and the framework design by grammar based language application. It also describes the limitations encountered during the thesis work and further section describes the future work specific to the thesis and in general to the research topic.

7.1 Reconfigurable architecture

Ericsson AB investigated Switch and Mesh based network topology for efficient protocol processing. Based on performance, cost, scalability and other internal factors switch based network topology named as Freyja is considered for further implementation and research. The reconfigurable hardware implemented using systemC consisted of a central switch with all the operators connected to it as explained in section 4.1.

The design of a new architecture require manually setting up simulations, estimation of resources and synthesizing the hardware which consists of system, logic and physical synthesis. This is a time consuming process and decreases the productivity of the research. Research project called TACO(Tools for Application-specific Hardware/Software Codesign) at university of Turku, explains the functions, features and capabilities required by such a tool [46]. One such requirement is importing architecture details in the tool from systemC top level files. The same approach is applied in Freyja architecture by abstracting the systemC details into the high level description.

The main objective of this work is to show the complete process of mapping from high level description to the system level implementation. The features mapped consisted of different operator instantiation, interconnections, memory, error handling and routing to multiple destinations from each operators for different protocols. In general protocol processors are more complex including features like Flow control, FIFOs, arbitration mechanism, latency of each computation, dead lock avoidance etc. Mapping of such finer details in the high level description will be straight forward with the approach presented in this work. For example the FIFOs can be instantiated similar to the Operator instantiation by extending the grammar. The FIFO block can be further configured with the variable depth of the buffer and signaling the overloading because of back pressure. How these details are represented

in high level description depends on the architecture implementation. Another example of extension is to extend the grammar to hold the latency of each operator.

The same framework can be used at different levels of synthesis to output the required files for the architecture. This reduces the time to reconfigure and setup the environment. For all these to accomplish, the reconfigurable architecture should be known. The framework can be adapted based on the architecture. The approach in this work shows the concept of mapping few core features which are easy to extend for more refined details.

Apart from abstracting high level architecture details, the core functions of each operator can also be represented using high level description. More research in academia and industry have already proved the concept. These tools input from Algorithms to Register Transfer level (ALU, Reg & Mux). This transformation is called as High level synthesis [47] [19].

7.2 Protocol sharing

As illustrated in section 3.5 the protocols share common functions. The features that can be shared are discussed at system level based on algorithms. The gain of sharing and reducing the hardware cost comes with the need of reconfiguration and back pressure in the system. The buffer length required to handle this and further details can be abstracted to the higher level to produce the reconfiguration content for the architecture. Automating the calculation of these resource utilization through simulation and importing the result to configure the hardware will reduce the architecture design time.

The implemented work shows the concept of mapping one resource to different protocols. Sharing requires arbitration algorithms to decide who can access the operator at any instant of time. It can be based on priority. However mathematical tools like Matlab can be used to implement the equations and calculate the resource usage. More utilized resources like CRC computation can be replicated for avoiding the back pressure. TACO project [46] provides a similar research conclusion for the protocol processor which exploits the resource sharing at logic levels. The tool identifies the common hardware blocks and creates multiple cores of them to meet the constraints. Similar approach can be extended with Freyja architecture for frequently used operators.

Some bottlenecks would be the memory interface and token access from higher layers. The system needs to be configured with different feasible combinations for

higher throughput. In future, these details can be abstracted to high level description for configure the overall system.

7.3 Language framework

Programmers are building domain-specific languages, configuration file formats, network protocols and numerous data file formats as well as traditional programming language compilers and interpreters. The development of such systems can be made faster by using the language building tools.

Programmers tend to avoid using language tools, resorting to adhoc methods, partly because of the raw and low level interface to these tools [48]. Using grammar based approach to build parsers will offer a more natural, high fidelity, robust and maintainable means of encoding a language-related problem. Most grammar development is done today with a simple text editor. The ANTLR4 parser generator [49] attempts to make grammars more accessible to the programmer by generating recursive descent parsers that are very similar to what programmer would build by hand.

Antlr4 supports rapid grammar developement by using ANTLR's built in interpreter, thus, providing immediate feedback during development [50]. The parse tree associated with matching input (as in Fig. 19), helps in debugging for error in grammar definition. If the input sequence is not in the language recognized by specified start rule, Antlr4 inserts an error node into the parse tree to indicate where the problem occurred.

All the above features of Antlr4 tool helps to build the language application in a more structured way. Providing the framework in multiple languages makes the programmer comfortable to build with their skilled language. The most encountered problems are resolved with the community support and the book written from the author Terrence Parr [49].

7.4 Limitations

The thesis started with the focus of developing the complete framework and to show the concept of using grammar based language application in hardware/software co design. The reconfigurable architecture modeled in systemC abstracted more hardware details. Though language is developed to include all future requirements, the implementations in the backend has to be extended for the new features of the hardware. Even though 3 protocols and their functions are discussed, it is not targeted to integrate all the protocols. As stated in section 7.1 the high level description is targeted with respect to current Freyja architecture. The Antlr4 tool

supports java language to output the framework and support for other languages are yet to be released.

7.5 Future work

Freyja architecture is still under implementation and in future, the complete details of the architecture can be abstracted into high level description by extending the grammar. Features like multiple core instantiations and transaction from different sources to same destination (required for Xio-s protocol) are possible to represent but still requires the backend implementations to modify according to the interconnection defined in future. Once the reconfigurable file format required by RTL hardware implementations are defined, the same framework can output such files to speedup the design process. As explained in section 1.2 the granularity of Freyja architecture is at algorithmic level. Common hardware blocks at logic levels and exploiting the reconfigurability for freyja architecture can be considered further and then the grammar can be extended similar to TACO protocol processor [46]. The header as explained in Fig. 44, can be described in the high level description which provides control in manipulating the length of each field.

The integration of Xio-s and CPRI protocols into the framework is straight forward. The process block i.e core functions of each operator needs to be modified to integrate these protocols into the system. Ethernet integration is set as an example. The grammar supports definition of multiple protocols in the language. The implementation of protocol functions doesn't require any modifications in the grammar.

Antlr4 is made to output the framework in Java language and the tool (in future!) will support other languages like c,c++. Based on the project requirement the language can be selected. Java has good commands for File IO operations and handling the objects. Major backend implementation deals with storing the parsed data in the form of Hashmap and Linkedlist. These facts can be considered when building similar concept for other reconfigurable architectures.

In general, the framework can be extended for all levels of synthesis while designing the hardware. The development environment will be heterogeneous at different levels and a common framework for producing the top level configurable files will reduce the errors. Once the architecture is specified, the tool can be made to generate SystemC model for simulation, a Matlab model for estimation of resources and VHDL model for synthesis of the architecture. The resource estimation can also be integrated within the tool by developing an interpreter using the parsed input.

References

- [1] M. Keating, "The future of design," in *The Simple Art of SoC Design*, pp. 171–180, Springer New York, 2011.
- [2] "Xilinx, all programmable." http://www.xilinx.com/fpga/, Aug. 2015. [Online; accessed 1-Sept-2015].
- [3] P. Schaumont, K. Sakiyama, A. Hodjat, and I. Verbauwhede, "Embedded software integration for coarse-grain reconfigurable systems," in *Parallel and Distributed Processing Symposium*, 2004. Proceedings. 18th International, pp. 137–, April 2004.
- [4] J. D. David C. Black, System C: From the Ground Up. Springer US, 2004.
- [5] D. Szczesny, A. Showk, S. Hessel, A. Bilgic, U. Hildebrand, and V. Frascolla, "Performance analysis of lte protocol processing on an arm based mobile platform," in *System-on-Chip*, 2009. SOC 2009. International Symposium on, pp. 056–063, Oct 2009.
- [6] A. Abnous and J. Rabaey, "Ultra-low-power domain-specific multimedia processors," in VLSI Signal Processing, IX, 1996., [Workshop on], pp. 461–470, Oct 1996.
- [7] K. Keutzer, S. Malik, and A. Newton, "From asic to asip: the next design discontinuity," in *Computer Design: VLSI in Computers and Processors*, 2002. Proceedings. 2002 IEEE International Conference on, pp. 84–90, 2002.
- [8] M. Badawi, A. Hemani, and Z. Lu, "Customizable coarse-grained energy-efficient reconfigurable packet processing architecture," in *Application-specific Systems, Architectures and Processors (ASAP), 2014 IEEE 25th International Conference on*, pp. 30–35, June 2014.
- [9] Y. Kaneta, S. Yoshizawa, S. Minato, H. Arimura, and Y. Miyanaga, "Dynamic reconfigurable bit-parallel architecture for large-scale regular expression matching," in *Field-Programmable Technology (FPT)*, 2010 International Conference on, pp. 21–28, Dec 2010.
- [10] D. Gajski and R. Kuhn, "Guest editors' introduction: New vlsi tools," *Computer*, vol. 16, pp. 11–14, Dec 1983.
- [11] G. Estrin, "Reconfigurable computer origins: the ucla fixed-plus-variable (f+v) structure computer," *Annals of the History of Computing, IEEE*, vol. 24, pp. 3–9, Oct 2002.

REFERENCES

[12] D. Page and L. Peterson, "Re-programmable pla," Apr. 2 1985. US Patent 4,508,977.

- [13] A. Waza, R. N. Mir, and H. N. ud din, "Reconfigurable architectures," *Journal of Advanced Computer Science & Technology*, vol. 1, no. 4, pp. 337–346, 2012.
- [14] M. Shami and A. Hemani, "Partially reconfigurable interconnection network for dynamically reprogrammable resource array," in ASIC, 2009. ASICON '09. IEEE 8th International Conference on, pp. 122–125, Oct 2009.
- [15] "Ericsson ab." http://www.ericsson.com/, 2015. [Online; accessed 1-March-2015].
- [16] E. Panainte, K. Bertels, and S. Vassiliadis, "Instruction scheduling for dynamic hardware configurations [m-jpeg encoder case study]," in *Design, Automation and Test in Europe, 2005. Proceedings*, pp. 100–105 Vol. 1, March 2005.
- [17] F. Mayer-Lindenberg, "High-level fpga programming through mapping process networks to fpga resources," in *Reconfigurable Computing and FPGAs*, 2009. ReConFig '09. International Conference on, pp. 302–307, Dec 2009.
- [18] A. DeHon, "The density advantage of configurable computing," *Computer*, vol. 33, pp. 41–49, Apr 2000.
- [19] P. Coussy, D. Gajski, M. Meredith, and A. Takach, "An introduction to high-level synthesis," *Design Test of Computers, IEEE*, vol. 26, pp. 8–17, July 2009.
- [20] G. Theodoridis, D. Soudris, and S. Vassiliadis, "A survey of coarse-grain reconfigurable architectures and cad tools," in Fine- and Coarse-Grain Reconfigurable Computing (S. Vassiliadis and D. Soudris, eds.), pp. 89–149, Springer Netherlands, 2007.
- [21] G. Genest, R. Chamberlain, and R. Bruce, "Programming an fpga-based super computer using a c-to-vhdl compiler: Dime-c," in *Adaptive Hardware and Systems*, 2007. AHS 2007. Second NASA/ESA Conference on, pp. 280–286, Aug 2007.
- [22] J. Mena, R. Deken, J. Coker, M. Johnstone, S. Ramirez, and P. Frey, "High level synthesis of a front end filter and dsp engine for analog to digital conversion x2013; a case study," in *VLSI Test Symposium (VTS)*, 2010 28th, pp. 252–252, April 2010.
- [23] R. Bergamaschi, "Bridging the domains of high-level and logic synthesis," Computer-Aided Design of Integrated Circuits and Systems, IEEE Transactions on, vol. 21, pp. 582–596, May 2002.

REFERENCES

[24] http://hls-labsticc.univ-ubs.fr/, 2015. [Online; accessed 20-Sept-2015].

- [25] D. Shin, A. Gerstlauer, R. Domer, and D. Gajski, "An interactive design environment for c-based high-level synthesis of rtl processors," Very Large Scale Integration (VLSI) Systems, IEEE Transactions on, vol. 16, pp. 466–475, April 2008.
- [26] B. Mei, S. Vernalde, D. Verkest, H. De Man, and R. Lauwereins, "Dresc: a retargetable compiler for coarse-grained reconfigurable architectures," in Field-Programmable Technology, 2002. (FPT). Proceedings. 2002 IEEE International Conference on, pp. 166–173, Dec 2002.
- [27] M. Alle, K. Varadarajan, A. Fell, S. K. Nandy, and R. Narayan, "Compiling techniques for coarse grained runtime reconfigurable architectures," in *Proceed*ings of the 5th International Workshop on Reconfigurable Computing: Architectures, Tools and Applications, ARC '09, (Berlin, Heidelberg), pp. 204–215, Springer-Verlag, 2009.
- [28] O. Malik, A. Hemani, and M. Shami, "High level synthesis framework for a coarse grain reconfigurable architecture," in NORCHIP, 2010, pp. 1–6, Nov 2010.
- [29] D. Cronquist, P. Franklin, S. Berg, and C. Ebeling, "Specifying and compiling applications for rapid," in FPGAs for Custom Computing Machines, 1998. Proceedings. IEEE Symposium on, pp. 116–125, Apr 1998.
- [30] S. Goldstein, H. Schmit, M. Budiu, S. Cadambi, M. Moe, and R. Taylor, "Piperench: a reconfigurable architecture and compiler," *Computer*, vol. 33, pp. 70–77, Apr 2000.
- [31] A. Abnous, K. Senoy, M. Wan, and J. Rabaey, "Evaluation of a low-power reconfigurable dsp architecture," in *proceedings 5 th Reconfigurable Architectures workshop (RAW98), March 30*, pp. 55–60, Springer, 1998.
- [32] P. Heysters and G. Smit, "Mapping of dsp algorithms on the montium architecture," in *Parallel and Distributed Processing Symposium*, 2003. Proceedings. International, pp. 6 pp.—, April 2003.
- [33] "Ieee standard for ethernet." https://standards.ieee.org/about/get/802/802.3.html, 2012. [Online; accessed 1-Feb-2015].
- [34] W. Peterson and D. Brown, "Cyclic codes for error detection," *Proceedings of the IRE*, vol. 49, pp. 228–235, Jan 1961.

REFERENCES REFERENCES

[35] "Cpri specification overview." http://www.cpri.info/spec.html, 2014. [On-line; accessed 1-Feb-2015].

- [36] "Ieee standard system c language reference manual," *IEEE Std 1666-2005*, pp. 0 1–423, 2006.
- [37] "Tlm 2.0 transaction level modeling library and whitepaper." http://accellera.org/downloads/standards/systemc, 2009. [Online; accessed 1-Feb-2015].
- [38] C. Shin, P. Grun, N. Romdhane, C. Lennard, G. Madl, S. Pasricha, N. Dutt, and M. Noll, "Enabling heterogeneous cycle-based and event-driven simulation in a design flow integrated using the SPIRIT consortium specifications," vol. 11, September 2007.
- [39] "Standard universal verification methodology class reference." http://accellera.org/downloads/standards/uvm, 2014. [Online; accessed 1-Feb-2015].
- [40] R. Siegmund and D. Muller, "A novel synthesis technique for communication controller hardware from declarative data communication protocol specifications," in *Design Automation Conference*, 2002. Proceedings. 39th, pp. 602–607, 2002.
- [41] A. Seawright and F. Brewer, "Clairvoyant: a synthesis system for production-based specification," Very Large Scale Integration (VLSI) Systems, IEEE Transactions on, vol. 2, pp. 172–185, June 1994.
- [42] R. Bloks, "A grammar based approach towards the automatic implementation of data communication protocols in hardware," in *Ph. D thesis, Eindhoven University of Technology, Sept 1993*, 1993.
- [43] J. Oberg, A. Kumar, and A. Hemani, "Grammar-based hardware synthesis from port-size independent specifications," *Very Large Scale Integration (VLSI) Systems, IEEE Transactions on*, vol. 8, pp. 184–194, April 2000.
- [44] J. Oberg, A. Kumar, and A. Hemani, "Grammar-based hardware synthesis of data communication protocols," in *System Synthesis*, 1996. Proceedings., 9th International Symposium on, pp. 14–19, Nov 1996.
- [45] S. H. Terence Parr and K. Fisher, "Adaptive ll(*) parsing: The power of dynamic analysis." http://www.antlr.org/papers/allstar-techreport.pdf, 2014. [Online; accessed 1-March-2015].

REFERENCES

[46] S. Virtanen, T. Lundstríom, and J. Lilius, "A design tool for the taco protocol processor development framework," in *n Proceedings of the 18 IEEE NorChip conference*, 6-7 November 2000, Turku, Finland.

- [47] A. M. Philippe Coussy, *High-Level Synthesis: From Algorithm to Digital Circuit*. Springer Netherlands, 2008.
- [48] J. Bovet and T. Parr, "Antlrworks: An antlr grammar development environment," *Softw. Pract. Exper.*, vol. 38, pp. 1305–1332, Oct. 2008.
- [49] T. Parr, The Definitive ANTLR Reference: Building Domain-Specific Languages. Pragmatic Bookshelf, 2007.
- [50] M. Daly, Y. Mandelbaum, D. Walker, M. Fernández, K. Fisher, R. Gruber, and X. Zheng, "Pads: An end-to-end system for processing ad hoc data," in Proceedings of the 2006 ACM SIGMOD International Conference on Management of Data, SIGMOD '06, (New York, NY, USA), pp. 727–729, ACM, 2006.

Appendices

A | Ethernet Protocol Description

```
FunctionUnits FREYJA:
  Operator PROCESSOR{
5 sc_name : proc;
6 op_type : proc_tx,proc_rx;
<sup>7</sup> errorid : 01,13;
8 }
10 Operator CHECKSUM{
sc_name : crc;
12 op_type : crc_add,crc_check;
{}_{13} \ \mathbf{op\_mem} \ : \ 0x01,0x04,0xc1,0x1d,0xb7,0x20,0x00,0x04,0x03,
0x04,0x00,0x00,0x01,0x59,0x53,0x10,0x03,0x02,0x04;
15 errorid : 02,03;
16 }
  Operator PRE_SFD{
19 sc_name : pre_sfd_add;
op_type : pre_sfd_add,pre_sfd_del;
21 errorid : 04,05;
  Operator CORRECTING_CODES_ADD{
25 sc_name : cc_a;
26 op_type : cc_a;
27 errorid : 06;
28 }
30 Operator ENCODER{
31 sc_name : enc_64_66;
32 op_type : enc;
зз errorid : 07;
34 }
  Operator SCRAMBLER{
37 sc_name : scram;
38 op_type : scram,descram;
39 op_mem : 0x01,0x00,0x00,0x00,0x80,0x00,0x00,0x04;
40 errorid : 08;
41 }
```

```
Operator GEARBOX_TX{
44 sc_name : gb_tx;
op_type : gb_tx;
46 errorid : 09;
47 }
  Operator GEARBOX_RX{
50 sc_name : gb_rx;
op_type : gb_rx;
52 errorid : 13;
53 }
55 Operator FRAMESYNC{
sc_name : fs;
57 op_type : fs;
58 errorid : 10;
59 }
  Operator DECODER{
62 sc_name : dec_66_64;
63 op_type : decdr;
64 errorid : 11;
  Operator CORRECTING_CODES_DEL{
68 sc_name : cc_d;
69 op_type : cc_d;
70 errorid : 12;
71 }
  protocol ETHERNET
75 fields : DA[06]
76 SA[06]
  VLAN:0x8100
  VLAN[02]
  Len[02]
80 PL[20]
  memory {
84 crc_add
            :00 to 08;
85 crc_check :00 to 07,09;
86 scram :00 to 07:
87 descram :00 to 07;
88 }
  Tx_PATH {
```

```
92 CONNECT: proc_tx to crc_add;
93 CONNECT: crc_add to pre_sfd_add;
94 CONNECT: pre_sfd_add to cc_a;
95 CONNECT: cc_a to enc;
96 CONNECT: enc to scram;
97 CONNECT: scram to gb_tx;
98 CONNECT: gb_tx to gb_rx;
99 }
100
  RX_PATH {
103 CONNECT: gb_rx to descram;
104 CONNECT: descram to decdr;
105 CONNECT: decdr to cc_d;
106 CONNECT: cc_d to pre_sfd_del;
107 CONNECT: pre_sfd_del to crc_check;
108 CONNECT: crc_check to proc_rx;
```

B | Grammar file : freyja_arch.g4

```
grammar freyja_arch;
   freyja: fus protocols;
   fus: 'FunctionUnits' ID': 'operators;
   operators:operator*;
5 operator:'operator ' ID'{' operator_fields '}';
   operator_fields: filename op_types opmem* memgrp* errorid;
   filename: 'sc_name: 'ID';';
  op_types:'op_type:' ID ids;
10 opmem: 'op_mem : ' HEX (',' HEX)* ';';
  memgrp:'mem_grp:' N ns;
   errorid: 'errorid: 'N ns;
  ns: ';' | ','N ns;
ids: ';' | ','ID ids;
   protocols: protocol*;
  protocol: 'protocol' P_NAME p_details;
  p_details: packet memory* tx_packet rx_packet;
packet: 'fields : ' (p_field)*;
p_field: hex_def | o_def;
  o_def: name position;
   position: '['N']';
  hex_def: name':' hex_val;
25 hex_val:HEX;
  name: ID;
  memory: 'memory' '{' memcontents '}';
   memcontents: (opername ':' mem bytes ';')*;
mem_bytes: range (',' range)*;
  range: (n1 'to' n2) \mid n1;
  opername: ID;
  n1: N;
  n2: N;
  tx_packet: 'Tx_PATH' '{' path* '}';
rx_packet: 'Rx_PATH' '{' path* '}';
  path: 'CONNECT' ':' id1 'to' id2 ';';
40 id1: ID;
  id2: ID;
   P_NAME: 'ETHERNET'| 'Xio' | 'CPRI' | 'SRIO' ;
45 HEX:'0x' HEXNUM+;
```

B GRAMMAR FILE : FREYJA_ARCH.G4

```
HEXNUM: ('0'..'9'|'a'...'f'|'A'...'F');

N: [0-9][0-9];

ID: [a-zA-Z_0-9]+;

WS: (''|[\t\r\n]+) -> skip;
```

C | Language Recognition terms

Language

A Language is a set of valid sentences which are composed of phrases, sub-phrases and so on.

Grammar

A Grammar formally defines the syntax rules of a language.

Syntax tree or Parse tree

This is a tree structure representation of a sentence. The leaves of the tree are symbols or tokens of the sentence

Token

A token is a symbol in a language like Identifier, keyword or an operator symbol.

Lexer

It performs lexical analysis by conversting input character streams into Tokens.

Parser

A Parser checks the sentence structure against the rules of a grammar.

Top down parser

It is a type of parsing strategy where one first looks at the highest level OF Parse tree (root) and works down the parse tree by reaching the leaf nodes.

Bottom up parser

In this parsing strategy the input text is processed from the lowest level to highest level (root).

Recursive Descent Parser

It is a kind of top down parser built from a set of mutually recursive procedures where each such procedure usually implements one of the productions of the grammar.

LookAhead parser

It defines the number of tokens accessible to the parser in making decisions at each point.

Left recursion

A grammar is left recursive if there a non terminal symbol which is derived in such a way that it exist as the left most symbol.

context free grammar

A formal grammar is context free when its production rules can be applied regardless of the context of a non terminal.