Chapter 8 II Lyapunov-Based Learning Control

1 General Nonlinear System

In the subsequent analysis, we will attack the control problem of the following nonlinear system

$$\dot{e} = f(t) - u_1 \tag{1}$$

where $f(t) \in \mathcal{L}_{\infty}$. Assume that the system unknown nonlinearity f(t) is periodic and bounded as follows

$$f(t) = f(t - T), \quad |f(t)| \le f_o \tag{2}$$

where $T \in \Re^1$ denotes the known period and f_o represents a positive bounding constant.

Question: As the unknown function is periodic, in other words, repeatable, is it possible to somehow learn such a signal so that to compensate it by a feedforward term in the control input? That is the design motivation of the learning control.

Remark 1 Compared to the adaptive control of learning the system parameters, the subsequently designed learning control intends to learn the uncertain nonlinear function completely. As a result, the LP (linear parameterization) condition for the adaptive control is not needed, instead, some assumption regarding the property of the nonlinear function f(t) is required to make the learning process possible.

2 Learning Control

For the uncertain nonlinear system of (1), design the following learning controller

$$u_1 = \hat{f}(t) + k_1 e \tag{3}$$

where the feedforward term $\hat{f}(t) \in \Re^1$ represents a learning based estimate for f(t) that is generated on-line via the following expression

$$\hat{f}(t) = \hat{f}(t - T) + k_2 e \tag{4}$$

Remark 2 Obviously, the present of the first term $\hat{f}(t)$ in the learning controller (3) aims to cancel the uncertain nonlinearity f(t) in the system dynamics (1); hence, the performance of this controller largely depends on whether f(t) can be learned successfully or not.

Substituting the controller $u_1(t)$ into the system dynamics (1) yields

$$\dot{e} = f(t) - \hat{f}(t) - k_1 e
= \tilde{f}(t) - k_1 e$$
(5)

where $\tilde{f}\left(t\right)\in\Re^{1}$ denotes the following defined learning error

$$\tilde{f}(t) = f(t) - \hat{f}(t) \tag{6}$$

After substituting (4) into (6) for $\hat{f}(t)$ and performing some mathematical manipulation

$$\tilde{f}(t) = f(t) - \hat{f}(t)
= f(t) - \hat{f}(t-T) - k_2 e
= f(t-T) - \hat{f}(t-T) - k_2 e
= \tilde{f}(t-T) - k_2 e$$
(7)

we obtain

$$\tilde{f}(t-T) = \tilde{f}(t) + k_2 e \tag{8}$$

Define the Lyapunov function as follows

$$V = \frac{1}{2}e^2 + \frac{1}{2k_2} \int_{t-T}^t \left[\tilde{f}(\sigma) \right]^2 d\sigma \geqslant 0 \tag{9}$$

Taking the time derivative of (9) yields

$$\dot{V} = e\dot{e} + \frac{1}{2k_2} \left\{ \left[\tilde{f}(t) \right]^2 - \left[\tilde{f}(t-T) \right]^2 \right\}
= e\dot{e} + \frac{1}{2k_2} \left\{ \left[\tilde{f}(t) + \tilde{f}(t-T) \right] \left[\tilde{f}(t) - \tilde{f}(t-T) \right] \right\}
= e\dot{e} + \frac{1}{2k_2} \left[2\tilde{f}(t) + k_2 e \right] \left[-k_2 e \right]
= e \left[\tilde{f}(t) - k_1 e \right] - \frac{e}{2} \left[2\tilde{f}(t) + k_2 e \right]
= e \left[\tilde{f}(t) - k_1 e \right] - e\tilde{f}(t) - \frac{k_2}{2} e^2
= -\left(k_1 + \frac{k_2}{2} \right) e^2 \le 0$$
(10)

where (5) and (8) have been utilized.

2.0.1 Signal Chasing

Question: Can we show all the signals bounded so far?

Based on (9) and (10), we know that $e(t) \in \mathcal{L}_{\infty} \Longrightarrow x(t) \in \mathcal{L}_{\infty} \stackrel{?????}{\Longrightarrow} u_1(t) \in \mathcal{L}_{\infty}$ or $\tilde{f}(t)$ or $\hat{f}(t) \in \mathcal{L}_{\infty}$?

2.0.2 Error Tracking

From (9) and (10), and if we can further show that $\dot{e} \in \mathcal{L}_{\infty}$, then it is straightforward to conclude that

$$\lim_{t \to \infty} e(t) = 0. \tag{11}$$

Now, it still comes to the question of the boundedness of the signal \dot{e} . Unfortunately, with the controller proposed in (3), there is no way to prove that $\dot{e} \in \mathcal{L}_{\infty}$.

Question: Can we modify (3) to make it bounded? If yes, how?

2.1 Saturated Learning Controller

Apparently, a choice to ensure the boundedness of u_1 is to saturate the controller itself or the learning based signal $\hat{f}(t)$. Based on this thought, the learning based estimate $\hat{f}(t)$ is modified as follows

$$\hat{f}(t) = sat_{f_o}\left(\hat{f}(t-T)\right) + k_2e \tag{12}$$

where the saturation function $sat_{f_o}(\cdot)$ is defined in the following manner

$$sat_{f_o}(\varepsilon) = \begin{cases} \varepsilon & \text{for } |\varepsilon| \le f_o \\ sgn(\varepsilon) \delta_o & \text{for } |\varepsilon| > f_o \end{cases}$$
 (13)

Remark 3 The structure of (12) and the definition of (13) provide a bounded estimation signal $\hat{f}(t)$ in the sense that

$$\left|\hat{f}\left(t\right)\right| \le f_o + k_2 \tag{14}$$

As the controller has been changed, we need to repeat the stability analysis for the new controller (12). To do that, we first perform the following calculation

$$f(t-T) - sat_{f_o}\left(\hat{f}(t-T)\right) = f(t) - \left[\hat{f}(t) - k_2 e\right]$$

$$= f(t) - \hat{f}(t) + k_2 e$$
(15)

To show the stability of the closed-loop system, we choose the following Lyapunov function

$$V_{2} = \frac{1}{2}e^{2} + \frac{1}{2k_{2}} \int_{t-T}^{t} \left[f(\sigma) - sat_{f_{o}} \hat{f}(\sigma) \right]^{2} d\sigma \geqslant 0$$

$$(16)$$

and calculate its time derivative as follows

$$\dot{V}_{2} = e\dot{e} + \frac{1}{2k_{2}} \left\{ \left[f(t) - sat_{fo}\hat{f}(t) \right]^{2} - \left[f(t-T) - sat_{fo}\hat{f}(t-T) \right]^{2} \right\} \\
= e \left[\tilde{f}(t) - k_{1}e \right] + \frac{1}{2k_{2}} \left\{ \left[f(t) - sat_{fo}\hat{f}(t) \right]^{2} - \left[f(t) - \hat{f}(t) + k_{2}e \right]^{2} \right\} \\
= e \tilde{f}(t) - k_{1}e^{2} + \frac{1}{2k_{2}} \left\{ \left[f(t) - sat_{fo}\hat{f}(t) \right]^{2} - \left[f(t) - \hat{f}(t) \right]^{2} \right\} \\
- \frac{1}{2k_{2}} \left[k_{2}^{2}e^{2} + 2k_{2}e \left(f(t) - \hat{f}(t) \right) \right] \\
\leq - \left(k_{1} + \frac{k_{2}}{2} \right) e^{2} \leq 0$$
(17)

where the following fact has been utilized (see Appendix for the proof)

$$\left\{ \left[f\left(t\right) - sat_{fo}\hat{f}\left(t\right) \right]^{2} - \left[f\left(t\right) - \hat{f}\left(t\right) \right]^{2} \right\} \le 0 \tag{18}$$

2.1.1 Signal Chasing

Based on (16) and (17), we know that $e(t) \in \mathcal{L}_{\infty} \Longrightarrow x(t) \in \mathcal{L}_{\infty} \Longrightarrow u_1(t) \in \mathcal{L}_{\infty}$ since $\hat{f}(t)$, $\tilde{f}(t) \in \mathcal{L}_{\infty} \Longrightarrow$ all the signals during closed-loop operation remain bounded.

2.1.2 Error Tracking

From (16) and (17), and the fact that $\dot{e} \in \mathcal{L}_{\infty}$, it is straightforward to conclude that

$$\lim_{t \to \infty} e(t) = 0. \tag{19}$$

2.1.3 Tracking Result

Asymptotically tracking

Question: Is the controller of (3) continuous?

Question: Can the estimation signal $\hat{f}(t)$ learn the unknown nonlinearity f(t)? Extended

Barbalat's Lemma to equation (5).

3 Comparision of Several Nonlinear Controllers

Nonlinear Control	Assumption Needed	Result	Something Related
Adaptive	LP: $f(x, t, \theta) = Y(x, t) \theta$	Asymptotic	Identification (PE Needed)
Sliding Mode	$ f(x,t,\theta) \le \rho(x ,t)$	Exponential	Chattering
High Frequency	$ f(x,t,\theta) \le \rho(x ,t)$	GUUB	Chattering
High Gain	$ f(x,t,\theta) \le \rho(x ,t)$	GUUB	Large Control Input
Learning	$f(t) = f(t - T), f(t) \le f_o$	Asymptotic	Nonlinearity Learned

A Inequality Proof

After expanding (18) and cancelling common terms, the inequality of (18) can be rewritten as follows

 $\left(sat_{f_0}\hat{f}(t)\right)^2 - 2f(t) \cdot sat_{f_0}\hat{f}(t) \le \left(\hat{f}(t)\right)^2 - 2f(t) \cdot \hat{f}(t) \tag{20}$

We can then substract $\left[\left(\hat{f}\left(t\right)\right)^{2}-2f\left(t\right)\cdot\hat{f}\left(t\right)\right]$ from both sides of (20) and factor the resulting expression to obtain the following equivalent inequality

$$\left(sat_{f_0}\hat{f}(t) - \hat{f}(t)\right)\left(sat_{\delta_0}\hat{f}(t) + \hat{f}(t) - 2f(t)\right) \le 0.$$
(21)

To show the inequality of (21), we then need to divide the proof into three possible cases as follows.

A.1 Case 1: $\left| \hat{f}(t) \right| \leq f_o$

Based on the definition of the function (13), it is easy to see that for this case

$$sat_{f_0}\hat{f}(t) = \hat{f}(t); \qquad (22)$$

hence, the inequality (21) holds for Case 1.

A.2 Case 2: $\hat{f}(t) > f_o$

From the definition of the function (13), it is clear that for this case

$$sat_{f_0}\hat{f}(t) = f_0 \tag{23}$$

which leads to the following fact

$$sat_{f_0}\hat{f}(t) - \hat{f}(t) \le 0, \quad sat_{f_0}\hat{f}(t) + \hat{f}(t) \ge 2f_0.$$
 (24)

The inequality of (21) then directly follows from (24) and the fact that

$$f\left(t\right) \le f_0. \tag{25}$$

A.3 Case 3: $\hat{f}(t) < -f_o$

From the definition of the function (13), it is clear that for this case

$$sat_{f_0}\hat{f}(t) = -f_0 \tag{26}$$

which leads to the following fact

$$sat_{f_0}\hat{f}(t) - \hat{f}(t) \ge 0, \quad sat_{f_0}\hat{f}(t) + \hat{f}(t) \le -2f_0.$$
 (27)

The inequality of (21) then directly follows from (27) and the fact that

$$f(t) \ge -f_0. \tag{28}$$

Therefore, (18) is true for all the three possible cases.