

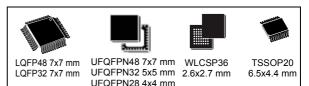
# STM32F042x4 STM32F042x6

ARM®-based 32-bit MCU, up to 32 KB Flash, crystal-less USB FS 2.0, CAN, 9 timers, ADC and comm. interfaces, 2.0 - 3.6 V

Datasheet - production data

#### **Features**

- Core: ARM<sup>®</sup> 32-bit Cortex<sup>®</sup>-M0 CPU, frequency up to 48 MHz
- Memories
  - 16 to 32 Kbytes of Flash memory
  - 6 Kbytes of SRAM with HW parity
- CRC calculation unit
- Reset and power management
  - Digital and I/Os supply: V<sub>DD</sub> = 2 V to 3.6 V
  - Analog supply: V<sub>DDA</sub> = from V<sub>DD</sub> to 3.6 V
  - Selected I/Os:  $V_{DDIO2}$  = 1.65 V to 3.6 V
  - Power-on/Power down reset (POR/PDR)
  - Programmable voltage detector (PVD)
  - Low power modes: Sleep, Stop, Standby
  - V<sub>BAT</sub> supply for RTC and backup registers
- Clock management
  - 4 to 32 MHz crystal oscillator
  - 32 kHz oscillator for RTC with calibration
  - Internal 8 MHz RC with x6 PLL option
  - Internal 40 kHz RC oscillator
  - Internal 48 MHz oscillator with automatic trimming based on ext. synchronization
- Up to 38 fast I/Os
  - All mappable on external interrupt vectors
  - Up to 24 I/Os with 5 V tolerant capability and 8 with independent supply V<sub>DDIO2</sub>
- 5-channel DMA controller
- One 12-bit, 1.0 µs ADC (up to 10 channels)
  - Conversion range: 0 to 3.6 V
  - Separate analog supply: 2.4 V to 3.6 V
- Up to 14 capacitive sensing channels for touchkey, linear and rotary touch sensors
- Calendar RTC with alarm and periodic wakeup from Stop/Standby



#### Nine timers

- One 16-bit advanced-control timer for six channel PWM output
- One 32-bit and four 16-bit timers, with up to four IC/OC, OCN, usable for IR control decoding
- Independent and system watchdog timers
- SysTick timer
- Communication interfaces
  - One I<sup>2</sup>C interface supporting Fast Mode Plus (1 Mbit/s) with 20 mA current sink, SMBus/PMBus and wakeup
  - Two USARTs supporting master synchronous SPI and modem control, one with ISO7816 interface, LIN, IrDA, auto baud rate detection and wakeup feature
  - Two SPIs (18 Mbit/s) with 4 to 16 programmable bit frames, one with I<sup>2</sup>S interface multiplexed
  - CAN interface
  - USB 2.0 full-speed interface, able to run from internal 48 MHz oscillator and with BCD and LPM support
- HDMI CEC, wakeup on header reception
- Serial wire debug (SWD)
- 96-bit unique ID
- All packages ECOPACK<sup>®</sup>2

#### Table 1. Device summary

Reference	Part number
STM32F042x4	STM32F042F4, STM32F042G4, STM32F042K4, STM32F042T4, STM32F042C4
STM32F042x6	STM32F042F6, STM32F042G6, STM32F042K6, STM32F042T6, STM32F042C6

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## 1 Introduction

This datasheet provides the ordering information and mechanical device characteristics of the STM32F042x4/x6 microcontrollers.

This document should be read in conjunction with the STM32F0xxxx reference manual (RM0091). The reference manual is available from the STMicroelectronics website <a href="https://www.st.com">www.st.com</a>.

For information on the ARM<sup>®</sup> Cortex<sup>®</sup>-M0 core, please refer to the Cortex<sup>®</sup>-M0 Technical Reference Manual, available from the www.arm.com website.





# 2 Description

The STM32F042x4/x6 microcontrollers incorporate the high-performance ARM® Cortex®-M0 32-bit RISC core operating at up to 48 MHz frequency, high-speed embedded memories (up to 32 Kbytes of Flash memory and 6 Kbytes of SRAM), and an extensive range of enhanced peripherals and I/Os. All devices offer standard communication interfaces (one I<sup>2</sup>C, two SPIs/one I<sup>2</sup>S, one HDMI CEC and two USARTs), one USB Full-speed device (crystal-less), one CAN, one 12-bit ADC, four 16-bit timers, one 32-bit timer and an advanced-control PWM timer.

The STM32F042x4/x6 microcontrollers operate in the -40 to +85 °C and -40 to +105 °C temperature ranges, from a 2.0 to 3.6 V power supply. A comprehensive set of power-saving modes allows the design of low-power applications.

The STM32F042x4/x6 microcontrollers include devices in seven different packages ranging from 20 pins to 48 pins with a die form also available upon request. Depending on the device chosen, different sets of peripherals are included.

These features make the STM32F042x4/x6 microcontrollers suitable for a wide range of applications such as application control and user interfaces, hand-held equipment, A/V receivers and digital TV, PC peripherals, gaming and GPS platforms, industrial applications, PLCs, inverters, printers, scanners, alarm systems, video intercoms and HVACs.

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Table 2. STM32F042x4/x6 device features and peripheral counts

Peripheral		STM32F042Fx STM32F042G		STM32F042K		STM32F042T		STM32F042C			
Flash memory (Kbyte)		16	32	16	32	16	32	16	32	16	32
SRAM	(Kbyte)					(	3				
Timers	Advanced control		1 (16-bit)								
Timers	General purpose		4 (16-bit) 1 (32-bit)								
	SPI [I <sup>2</sup> S] <sup>(1)</sup>				1	[1]				2	[1]
	I <sup>2</sup> C					,	1				
Comm.	USART		2								
interfaces	CAN	1									
	USB	1									
	CEC	1									
	t ADC f channels)	1 (9 ext. + 3 int.) (10 ext. + 3 int.)									
GP	lOs	1	6	2	4	2 2		3	0	3	8
Capacitive sensing channels		-	7	1	1	1	3 4	1	4	1	4
Max. CPU frequency		48 MHz									
Operating voltage		2.0 to 3.6 V									
Operating temperature		Ambient operating temperature: -40°C to 85°C / -40°C to 105°C Junction temperature: -40°C to 105°C / -40°C to 125°C									
Packages		TSS	OP20	UQF	PN28		P32 PN32	WLC	SP36		P48 PN48

<sup>1.</sup> The SPI interfaces can be used either in SPI mode or in  $I^2S$  audio mode.

POWER SWCLK Serial Wire SWDIO as AF VOLT.REG 3.3 V to 1.8 V  $V_{DD} = 2 \text{ to } 3.6 \text{ V}$ Debug op Flash GPL Up to 32 KB Flash On memory interface @ V<sub>DD</sub> CORTEX-M0 CPU 32-bit f<sub>MAX</sub> = 48 MHz OKIN ← SUPPLY SUPERVISION V<sub>DDIO2</sub> OKIN ◀ NRST SRAM 6 KB SRAM controller Bus matrix Reset ◀  $V_{DDA}$   $V_{SSA}$ POR/PDR Int ◀ NVIC V<sub>DD</sub> HSI14 PVD RC 14 MHz HSI RC 8 MHz PLLCLK PLL LSI GP DMA 5 channels XTAL OSC 4-32 MHz RC 40 kHz OSC\_IN HSI48 OSC\_OUT RC 48MHz Ind. Window WDG PA[15:0] < GPIO port A RESET & CLOCK CONTROL 录。 PB[15:0] GPIO port B  $V_{BAT}$  = 1.65 to 3.6 V @ V<sub>BA</sub> OSC32\_IN OSC32\_OUT decoder GPIO port C PC[15:13] < XTAL32 kHz System and peripheral clocks 1 TAMPER-RTC (ALARM OUT) Backup AHB RTC reg RTC interface CRS PF[11,1:0] < GPIO port F SYNC CRC 4 channels 4 channels 3 compl. channels BRK, ETR input as AF PWM TIMER 1 5 groups of Touch Analog switches 4 channels Sensing Controller AHR TIMER 2 32-bit 4 ch., ETR as AF SYNC APB 4 ch., ETR as AF TIMER 3 EXT. IT WKUP 38 AF TIMER 14 1 channel as AF USB D+, D-USB PHY @ V<sub>DDIO2</sub> 1 channel SRAM 768B TIMER 16 1 compl, BRK as AF SRAM 256B 1 channel Window WDG TIMER 17 1 compl, BRK as AF TX, RX as AF < **BxCAN** IR\_OUT as AF DBGMCU MOSI/SD RX, TX,CTS, RTS, USART1 MISO/MCK SCK/CK CK as AF SPI1/I2S1 RX, TX,CTS, RTS, CK as AF NSS/WS as AF SYSCFG IF USART2 MOSI MISO SCK NSS as AF SCL, SDA, SMBA (20 mA FM+) as AF I2C1 Temp. HDMI-CEC CEC as AF 10x AD input 12-bit ADC V<sub>SSA</sub> @ V<sub>DDA</sub> Power domain of analog blocks :  $V_{BAT}$  $V_{DD}$  $V_{DDA}$ V<sub>DDIO2</sub> MSv33178V4

Figure 1. Block diagram



#### 3 Functional overview

Figure 1 shows the general block diagram of the STM32F042x4/x6 devices.

# 3.1 ARM®-Cortex®-M0 core

The ARM<sup>®</sup> Cortex<sup>®</sup>-M0 is a generation of ARM 32-bit RISC processors for embedded systems. It has been developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced system response to interrupts.

The ARM<sup>®</sup> Cortex<sup>®</sup>-M0 processors feature exceptional code-efficiency, delivering the high performance expected from an ARM core, with memory sizes usually associated with 8- and 16-bit devices.

The STM32F042x4/x6 devices embed ARM core and are compatible with all ARM tools and software.

#### 3.2 Memories

The device has the following features:

- 6 Kbytes of embedded SRAM accessed (read/write) at CPU clock speed with 0 wait states and featuring embedded parity checking with exception generation for fail-critical applications.
- The non-volatile memory is divided into two arrays:
  - 16 to 32 Kbytes of embedded Flash memory for programs and data
  - Option bytes

The option bytes are used to write-protect the memory (with 4 KB granularity) and/or readout-protect the whole memory with the following options:

- Level 0: no readout protection
- Level 1: memory readout protection, the Flash memory cannot be read from or written to if either debug features are connected or boot in RAM is selected
- Level 2: chip readout protection, debug features (Cortex<sup>®</sup>-M0 serial wire) and boot in RAM selection disabled

#### 3.3 Boot modes

At startup, the boot pin and boot selector option bits are used to select one of the three boot options:

- boot from User Flash memory
- boot from System Memory
- boot from embedded SRAM

The boot pin is shared with the standard GPIO and can be disabled through the boot selector option bits. The boot loader is located in System Memory. It is used to reprogram the Flash memory by using USART on pins PA14/PA15, or PA9/PA10 or I<sup>2</sup>C on pins PB6/PB7 or through the USB DFU interface.



## 3.4 Cyclic redundancy check calculation unit (CRC)

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code from a 32-bit data word and a CRC-32 (Ethernet) polynomial.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

## 3.5 Power management

#### 3.5.1 Power supply schemes

- V<sub>DD</sub> = V<sub>DDIO1</sub> = 2.0 to 3.6 V: external power supply for I/Os (V<sub>DDIO1</sub>) and the internal regulator. It is provided externally through VDD pins.
- V<sub>DDA</sub> = from V<sub>DD</sub> to 3.6 V: external analog power supply for ADC, Reset blocks, RCs and PLL (minimum voltage to be applied to V<sub>DDA</sub> is 2.4 V when the ADC is used). It is provided externally through VDDA pin. The V<sub>DDA</sub> voltage level must be always greater or equal to the V<sub>DD</sub> voltage level and must be established first.
- V<sub>DDIO2</sub> = 1.65 to 3.6 V: external power supply for marked I/Os. V<sub>DDIO2</sub> is provided externally through the VDDIO2 pin. The V<sub>DDIO2</sub> voltage level is completely independent from V<sub>DD</sub> or V<sub>DDA</sub>, but it must not be provided without a valid supply on V<sub>DD</sub>. The V<sub>DDIO2</sub> supply is monitored and compared with the internal reference voltage (V<sub>REFINT</sub>). When the V<sub>DDIO2</sub> is below this threshold, all the I/Os supplied from this rail are disabled by hardware. The output of this comparator is connected to EXTI line 31 and it can be used to generate an interrupt. Refer to the pinout diagrams or tables for concerned I/Os list.
- V<sub>BAT</sub> = 1.65 to 3.6 V: power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when V<sub>DD</sub> is not present.

For more details on how to connect power pins, refer to Figure 13: Power supply scheme.

#### 3.5.2 Power supply supervisors

The device has integrated power-on reset (POR) and power-down reset (PDR) circuits. They are always active, and ensure proper operation above a threshold of 2 V. The device remains in reset mode when the monitored supply voltage is below a specified threshold,  $V_{POR/PDR}$ , without the need for an external reset circuit.

- The POR monitors only the V<sub>DD</sub> supply voltage. During the startup phase it is required that V<sub>DDA</sub> should arrive first and be greater than or equal to V<sub>DD</sub>.
- The PDR monitors both the V<sub>DD</sub> and V<sub>DDA</sub> supply voltages, however the V<sub>DDA</sub> power supply supervisor can be disabled (by programming a dedicated Option bit) to reduce the power consumption if the application design ensures that V<sub>DDA</sub> is higher than or equal to V<sub>DD</sub>.

The device features an embedded programmable voltage detector (PVD) that monitors the  $V_{DD}$  power supply and compares it to the  $V_{PVD}$  threshold. An interrupt can be generated when  $V_{DD}$  drops below the  $V_{PVD}$  threshold and/or when  $V_{DD}$  is higher than the  $V_{PVD}$ 



threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

#### 3.5.3 Voltage regulator

The regulator has two operating modes and it is always enabled after reset.

- Main (MR) is used in normal operating mode (Run).
- Low power (LPR) can be used in Stop mode where the power demand is reduced.

In Standby mode, it is put in power down mode. In this mode, the regulator output is in high impedance and the kernel circuitry is powered down, inducing zero consumption (but the contents of the registers and SRAM are lost).

#### 3.5.4 Low-power modes

The STM32F042x4/x6 microcontrollers support three low-power modes to achieve the best compromise between low power consumption, short startup time and available wakeup sources:

#### Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

#### • Stop mode

Stop mode achieves very low power consumption while retaining the content of SRAM and registers. All clocks in the 1.8 V domain are stopped, the PLL, the HSI RC and the HSE crystal oscillators are disabled. The voltage regulator can also be put either in normal or in low power mode.

The device can be woken up from Stop mode by any of the EXTI lines. The EXTI line source can be one of the 16 external lines, the PVD output, RTC, I2C1 USART1, USB or the CEC.

The CEC, USART1 and I2C1 peripherals can be configured to enable the HSI RC oscillator so as to get clock for processing incoming data. If this is used when the voltage regulator is put in low power mode, the regulator is first switched to normal mode before the clock is provided to the given peripheral.

#### Standby mode

The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire 1.8 V domain is powered off. The PLL, the HSI RC and the HSE crystal oscillators are also switched off. After entering Standby mode, SRAM and register contents are lost except for registers in the RTC domain and Standby circuitry.

The device exits Standby mode when an external reset (NRST pin), an IWDG reset, a rising edge on the WKUP pins, or an RTC event occurs.

Note: The RTC, the IWDG, and the corresponding clock sources are not stopped by entering Stop or Standby mode.



## 3.6 Clocks and startup

System clock selection is performed on startup, however the internal RC 8 MHz oscillator is selected as default CPU clock on reset. An external 4-32 MHz clock can be selected, in which case it is monitored for failure. If failure is detected, the system automatically switches back to the internal RC oscillator. A software interrupt is generated if enabled. Similarly, full interrupt management of the PLL clock entry is available when necessary (for example on failure of an indirectly used external crystal, resonator or oscillator).

Several prescalers allow the application to configure the frequency of the AHB and the APB domains. The maximum frequency of the AHB and the APB domains is 48 MHz.

Additionally, also the internal RC 48 MHz oscillator can be selected for system clock or PLL input source. This oscillator can be automatically fine-trimmed by the means of the CRS peripheral using the external synchronization.

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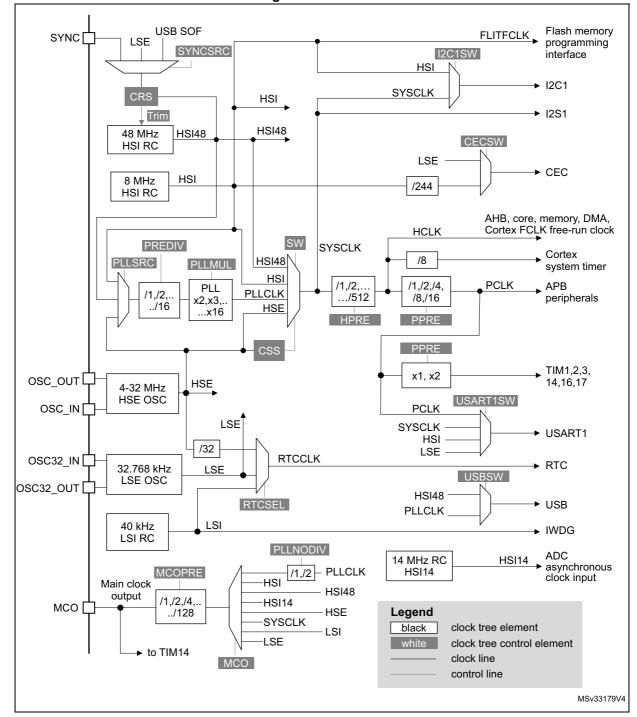


Figure 2. Clock tree

# 3.7 General-purpose inputs/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions.



The I/O configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

### 3.8 Direct memory access controller (DMA)

The 5-channel general-purpose DMAs manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers.

The DMA supports circular buffer management, removing the need for user code intervention when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with support for software trigger on each channel. Configuration is made by software and transfer sizes between source and destination are independent.

DMA can be used with the main peripherals: SPIx, I2Sx, I2Cx, USARTx, all TIMx timers (except TIM14) and ADC.

## 3.9 Interrupts and events

#### 3.9.1 Nested vectored interrupt controller (NVIC)

The STM32F0xx family embeds a nested vectored interrupt controller able to handle up to 32 maskable interrupt channels (not including the 16 interrupt lines of Cortex -M0) and 4 priority levels.

- Closely coupled NVIC gives low latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Closely coupled NVIC core interface
- Allows early processing of interrupts
- Processing of late arriving higher priority interrupts
- Support for tail-chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimal interrupt latency.

#### 3.9.2 Extended interrupt/event controller (EXTI)

The extended interrupt/event controller consists of 24 edge detector lines used to generate interrupt/event requests and wake-up the system. Each line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the internal clock period. Up to 38 GPIOs can be connected to the 16 external interrupt lines.

# 3.10 Analog-to-digital converter (ADC)

The 12-bit analog-to-digital converter has up to 10 external and 3 internal (temperature



sensor, voltage reference, VBAT voltage measurement) channels and performs conversions in single-shot or scan modes. In scan mode, automatic conversion is performed on a selected group of analog inputs.

The ADC can be served by the DMA controller.

An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

### 3.10.1 Temperature sensor

The temperature sensor (TS) generates a voltage  $V_{\text{SENSE}}$  that varies linearly with temperature.

The temperature sensor is internally connected to the ADC\_IN16 input channel which is used to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode.

Calibration value name	Description	Memory address	
TS_CAL1	TS ADC raw data acquired at a temperature of 30 °C (± 5 °C), V <sub>DDA</sub> = 3.3 V (± 10 mV)	0x1FFF F7B8 - 0x1FFF F7B9	
TS_CAL2	TS ADC raw data acquired at a temperature of 110 °C (± 5 °C), V <sub>DDA</sub> = 3.3 V (± 10 mV)	0x1FFF F7C2 - 0x1FFF F7C3	

Table 3. Temperature sensor calibration values

# 3.10.2 Internal voltage reference (V<sub>REFINT</sub>)

The internal voltage reference ( $V_{REFINT}$ ) provides a stable (bandgap) voltage output for the ADC.  $V_{REFINT}$  is internally connected to the ADC\_IN17 input channel. The precise voltage of  $V_{REFINT}$  is individually measured for each part by ST during production test and stored in the system memory area. It is accessible in read-only mode.

Table 4. Internal voltage reference calibration values

Calibration value name	Description	Memory address	
	Raw data acquired at a temperature of 30 °C (± 5 °C), V <sub>DDA</sub> = 3.3 V (± 10 mV)	0x1FFF F7BA - 0x1FFF F7BB	

## 3.10.3 V<sub>BAT</sub> battery voltage monitoring

This embedded hardware feature allows the application to measure the  $V_{BAT}$  battery voltage using the internal ADC channel ADC\_IN18. As the  $V_{BAT}$  voltage may be higher than  $V_{DDA}$ , and thus outside the ADC input range, the  $V_{BAT}$  pin is internally connected to a bridge divider by 2. As a consequence, the converted digital value is half the  $V_{BAT}$  voltage.

## 3.11 Touch sensing controller (TSC)

The STM32F042x4/x6 devices provide a simple solution for adding capacitive sensing functionality to any application. These devices offer up to 14 capacitive sensing channels distributed over 5 analog I/O groups.

Capacitive sensing technology is able to detect the presence of a finger near a sensor which is protected from direct touch by a dielectric (glass, plastic...). The capacitive variation introduced by the finger (or any conductive object) is measured using a proven implementation based on a surface charge transfer acquisition principle. It consists in charging the sensor capacitance and then transferring a part of the accumulated charges into a sampling capacitor until the voltage across this capacitor has reached a specific threshold. To limit the CPU bandwidth usage, this acquisition is directly managed by the hardware touch sensing controller and only requires few external components to operate. For operation, one capacitive sensing GPIO in each group is connected to an external capacitor and cannot be used as effective touch sensing channel.

The touch sensing controller is fully supported by the STMTouch touch sensing firmware library, which is free to use and allows touch sensing functionality to be implemented reliably in the end application.

Table 5. Capacitive sensing GPIOs available on STM32F042x4/x6 devices

Group	croup Capacitive sensing signal name				
	TSC_G1_IO1	PA0			
1	TSC_G1_IO2	PA1			
'	TSC_G1_IO3	PA2			
	TSC_G1_IO4	PA3			
	TSC_G2_IO1	PA4			
2	TSC_G2_IO2	PA5			
2	TSC_G2_IO3	PA6			
	TSC_G2_IO4	PA7			
	TSC_G3_IO2	PB0			
3	TSC_G3_IO3	PB1			
	TSC_G3_IO4	PB2			

Group	Capacitive sensing signal name	Pin name
	TSC_G4_IO1	PA9
4	TSC_G4_IO2	PA10
4	TSC_G4_IO3	PA11
	TSC_G4_IO4	PA12
	TSC_G5_IO1	PB3
5	TSC_G5_IO2	PB4
3	TSC_G5_IO3	PB6
	TSC_G5_IO4	PB7



Table 6. No. of capacitive sensing channels available on STM32F042x devices

	Number of capacitive sensing channels									
Analog I/O group	STM32F042Cx LQPF48 UQFPN48	STM32F042Tx WLCSP36	STM32F042Kx LQFP32 UQFPN32	STM32F042Gx UQFPN28	STM32F042Fx TSSOP20					
G1	3	3	3	3	3					
G2	3	3	3	3	3					
G3	2 2		1 2	1	0					
G4	3	3	3	1	1					
G5	3	3	3	3	0					
Number of capacitive sensing channels	14	14	13 14	11	7					

# 3.12 Timers and watchdogs

The STM32F042x4/x6 devices include up to five general-purpose timers and an advanced control timer.

*Table 7* compares the features of the different timers.

Table 7. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary outputs
Advanced control	TIM1	16-bit	Up, down, up/down	integer from 1 to 65536	Yes	4	3
	TIM2	32-bit	Up, down, up/down	integer from 1 to 65536	Yes	4	-
General	TIM3	16-bit	Up, down, up/down	integer from 1 to 65536	Yes	4	-
purpose	TIM14	16-bit	Up	integer from 1 to 65536	No	1	-
	TIM16 TIM17	16-bit	Up	integer from 1 to 65536	Yes	1	1

## 3.12.1 Advanced-control timer (TIM1)

The advanced-control timer (TIM1) can be seen as a three-phase PWM multiplexed on six channels. It has complementary PWM outputs with programmable inserted dead times. It

can also be seen as a complete general-purpose timer. The four independent channels can be used for:

- input capture
- output compare
- PWM generation (edge or center-aligned modes)
- one-pulse mode output

If configured as a standard 16-bit timer, it has the same features as the TIMx timer. If configured as the 16-bit PWM generator, it has full modulation capability (0-100%).

The counter can be frozen in debug mode.

Many features are shared with those of the standard timers which have the same architecture. The advanced control timer can therefore work together with the other timers via the Timer Link feature for synchronization or event chaining.

#### 3.12.2 General-purpose timers (TIM2, 3, 14, 16, 17)

There are five synchronizable general-purpose timers embedded in the STM32F042x4/x6 devices (see *Table 7* for differences). Each general-purpose timer can be used to generate PWM outputs, or as simple time base.

#### TIM2, TIM3

STM32F042x4/x6 devices feature two synchronizable 4-channel general-purpose timers. TIM2 is based on a 32-bit auto-reload up/downcounter and a 16-bit prescaler. TIM3 is based on a 16-bit auto-reload up/downcounter and a 16-bit prescaler. They feature 4 independent channels each for input capture/output compare, PWM or one-pulse mode output. This gives up to 12 input captures/output compares/PWMs on the largest packages.

The TIM2 and TIM3 general-purpose timers can work together or with the TIM1 advanced-control timer via the Timer Link feature for synchronization or event chaining.

TIM2 and TIM3 both have independent DMA request generation.

These timers are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 3 hall-effect sensors.

Their counters can be frozen in debug mode.

#### **TIM14**

This timer is based on a 16-bit auto-reload upcounter and a 16-bit prescaler.

TIM14 features one single channel for input capture/output compare, PWM or one-pulse mode output.

Its counter can be frozen in debug mode.

#### TIM16 and TIM17

Both timers are based on a 16-bit auto-reload upcounter and a 16-bit prescaler.

They each have a single channel for input capture/output compare, PWM or one-pulse mode output.



TIM16 and TIM17 have a complementary output with dead-time generation and independent DMA request generation.

Their counters can be frozen in debug mode.

#### 3.12.3 Independent watchdog (IWDG)

The independent watchdog is based on an 8-bit prescaler and 12-bit downcounter with user-defined refresh window. It is clocked from an independent 40 kHz internal RC and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. It is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

#### 3.12.4 System window watchdog (WWDG)

The system window watchdog is based on a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the APB clock (PCLK). It has an early warning interrupt capability and the counter can be frozen in debug mode.

#### 3.12.5 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard down counter. It features:

- a 24-bit down counter
- autoreload capability
- maskable system interrupt generation when the counter reaches 0
- programmable clock source (HCLK or HCLK/8)

# 3.13 Real-time clock (RTC) and backup registers

The RTC and the five backup registers are supplied through a switch that takes power either on  $V_{DD}$  supply when present or through the  $V_{BAT}$  pin. The backup registers are five 32-bit registers used to store 20 bytes of user application data when  $V_{DD}$  power is not present. They are not reset by a system or power reset, or at wake up from Standby mode.

The RTC is an independent BCD timer/counter. Its main features are the following:

- calendar with subseconds, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format
- automatic correction for 28, 29 (leap year), 30, and 31 day of the month
- programmable alarm with wake up from Stop and Standby mode capability
- on-the-fly correction from 1 to 32767 RTC clock pulses. This can be used to synchronize the RTC with a master clock
- digital calibration circuit with 1 ppm resolution, to compensate for quartz crystal inaccuracy
- two anti-tamper detection pins with programmable filter. The MCU can be woken up from Stop and Standby modes on tamper event detection
- timestamp feature which can be used to save the calendar content. This function can be triggered by an event on the timestamp pin, or by a tamper event. The MCU can be woken up from Stop and Standby modes on timestamp event detection
- reference clock detection: a more precise second source clock (50 or 60 Hz) can be used to enhance the calendar precision

The RTC clock sources can be:

- a 32.768 kHz external crystal
- a resonator or oscillator
- the internal low-power RC oscillator (typical frequency of 40 kHz)
- the high-speed external clock divided by 32

# 3.14 Inter-integrated circuit interface (I<sup>2</sup>C)

The I<sup>2</sup>C interface (I2C1) can operate in multimaster or slave modes. It can support Standard mode (up to 100 kbit/s), Fast mode (up to 400 kbit/s) and Fast Mode Plus (up to 1 Mbit/s) with 20 mA output drive.

It supports 7-bit and 10-bit addressing modes, multiple 7-bit slave addresses (two addresses, one with configurable mask). It also includes programmable analog and digital noise filters.

Aspect	Analog filter	Digital filter					
Pulse width of suppressed spikes	≥ 50 ns	Programmable length from 1 to 15 I2Cx peripheral clocks					
Benefits	Available in Stop mode	<ul><li>–Extra filtering capability vs.</li><li>standard requirements</li><li>–Stable length</li></ul>					
Drawbacks	Variations depending on temperature, voltage, process	Wakeup from Stop on address match is not available when digital filter is enabled.					

Table 8. Comparison of I<sup>2</sup>C analog and digital filters

In addition, I2C1 provides hardware support for SMBUS 2.0 and PMBUS 1.1: ARP capability, Host notify protocol, hardware CRC (PEC) generation/verification, timeouts verifications and ALERT protocol management. I2C1 also has a clock domain independent



from the CPU clock, allowing the I2C1 to wake up the MCU from Stop mode on address match.

The I2C peripheral can be served by the DMA controller.

Table 9. STM32F042x4/x6 I<sup>2</sup>C implementation

I <sup>2</sup> C features <sup>(1)</sup>	I2C1
7-bit addressing mode	X
10-bit addressing mode	Х
Standard mode (up to 100 kbit/s)	Х
Fast mode (up to 400 kbit/s)	Х
Fast Mode Plus with 20 mA output drive I/Os (up to 1 Mbit/s)	Х
Independent clock	Х
SMBus	Х
Wakeup from STOP	Х

<sup>1.</sup> X = supported.

# 3.15 Universal synchronous/asynchronous receiver/transmitter (USART)

The device embeds two universal synchronous/asynchronous receivers/transmitters (USART1, USART2) which communicate at speeds of up to 6 Mbit/s.

They provide hardware management of the CTS, RTS and RS485 DE signals, multiprocessor communication mode, master synchronous communication and single-wire half-duplex communication mode. USART1 supports also SmartCard communication (ISO 7816), IrDA SIR ENDEC, LIN Master/Slave capability and auto baud rate feature, and has a clock domain independent of the CPU clock, allowing to wake up the MCU from Stop mode.

The USART interfaces can be served by the DMA controller.

Table 10. STM32F042x4/x6 USART implementation

USART modes/features <sup>(1)</sup>	USART1	USART2
Hardware flow control for modem	Х	X
Continuous communication using DMA	Х	Х
Multiprocessor communication	Х	X
Synchronous mode	Х	X
Smartcard mode	Х	-
Single-wire half-duplex communication	Х	X
IrDA SIR ENDEC block	Х	-
LIN mode	Х	-
Dual clock domain and wakeup from Stop mode	Х	-
Receiver timeout interrupt	Х	-



USART modes/features <sup>(1)</sup>	USART1	USART2
Modbus communication	Х	-
Auto baud rate detection	X	-
Driver Enable	Х	Х

Table 10. STM32F042x4/x6 USART implementation (continued)

# 3.16 Serial peripheral interface (SPI) / Inter-integrated sound interface (I<sup>2</sup>S)

Up to two SPIs are able to communicate up to 18 Mbit/s in slave and master modes in full-duplex and half-duplex communication modes. The 3-bit prescaler gives 8 master mode frequencies and the frame size is configurable from 4 bits to 16 bits.

One standard I<sup>2</sup>S interface (multiplexed with SPI1) supporting four different audio standards can operate as master or slave at half-duplex communication mode. It can be configured to transfer 16 and 24 or 32 bits with 16-bit or 32-bit data resolution and synchronized by a specific signal. Audio sampling frequency from 8 kHz up to 192 kHz can be set by an 8-bit programmable linear prescaler. When operating in master mode, it can output a clock for an external audio component at 256 times the sampling frequency.

Table 11. STM32F042x4/x6 SPI/I<sup>2</sup>S implementation

SPI features <sup>(1)</sup>	SPI1	SPI2
Hardware CRC calculation	Х	Х
Rx/Tx FIFO	Х	Х
NSS pulse mode	Х	Х
I <sup>2</sup> S mode	X	-
TI mode	Х	Х

<sup>1.</sup> X = supported.

# 3.17 High-definition multimedia interface (HDMI) - consumer electronics control (CEC)

The device embeds a HDMI-CEC controller that provides hardware support for the Consumer Electronics Control (CEC) protocol (Supplement 1 to the HDMI standard).

This protocol provides high-level control functions between all audiovisual products in an environment. It is specified to operate at low speeds with minimum processing and memory overhead. It has a clock domain independent from the CPU clock, allowing the HDMI\_CEC controller to wakeup the MCU from Stop mode on data reception.

# 3.18 Controller area network (CAN)

The CAN is compliant with specifications 2.0A and B (active) with a bit rate up to 1 Mbit/s. It can receive and transmit standard frames with 11-bit identifiers as well as extended frames



<sup>1.</sup> X = supported.

with 29-bit identifiers. It has three transmit mailboxes, two receive FIFOs with 3 stages and 14 scalable filter banks.

### 3.19 Universal serial bus (USB)

The STM32F042x4/x6 embeds a full-speed USB device peripheral compliant with the USB specification version 2.0. The internal USB PHY supports USB FS signaling, embedded DP pull-up and also battery charging detection according to Battery Charging Specification Revision 1.2. The USB interface implements a full-speed (12 Mbit/s) function interface with added support for USB 2.0 Link Power Management. It has software-configurable endpoint setting with packet memory up-to 1 KB (the last 256 byte are used for CAN peripheral if enabled) and suspend/resume support. It requires a precise 48 MHz clock which can be generated from the internal main PLL (the clock source must use an HSE crystal oscillator) or by the internal 48 MHz oscillator in automatic trimming mode. The synchronization for this oscillator can be taken from the USB data stream itself (SOF signalization) which allows crystal-less operation.

## 3.20 Clock recovery system (CRS)

The STM32F042x4/x6 embeds a special block which allows automatic trimming of the internal 48 MHz oscillator to guarantee its optimal accuracy over the whole device operational range. This automatic trimming is based on the external synchronization signal, which could be either derived from USB SOF signalization, from LSE oscillator, from an external signal on CRS\_SYNC pin or generated by user software. For faster lock-in during startup it is also possible to combine automatic trimming with manual trimming action.

# 3.21 Serial wire debug port (SW-DP)

An ARM SW-DP interface is provided to allow a serial wire debugging tool to be connected to the MCU.



# 4 Pinouts and pin descriptions

Figure 3. LQFP48 package pinout

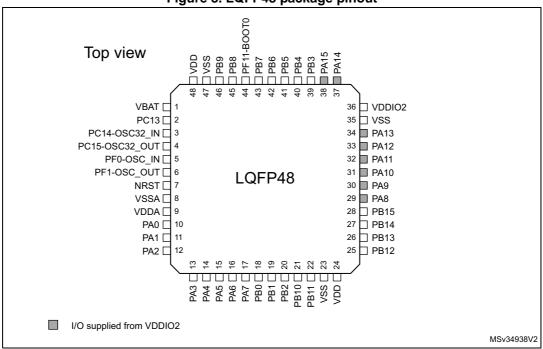
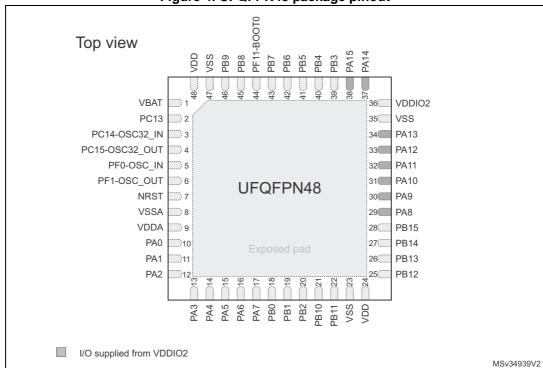


Figure 4. UFQFPN48 package pinout



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Top view 1 2 3 4 5 6 Α PA12 PA15 PB4 PB7 VDD PC13 PC14-OSC32 IN PF0-OSC IN PB8-В PA13 PA14 PB3 воото С PA10 PA11 PA4 PB6 OSC\_ OUT OUT D PA9 PB2 PA5 NRST VSS PA1 Ε VDDIO2 PA8 PA6 PA2 VDDA PB5 F VSS PB1 PB0 PA7 PA3 PA0 WLCSP36 I/O supplied from VDDIO2 MSv34201V3

Figure 5. WLCSP36 package pinout

1. The above figure shows the package in top view, changing from bottom view in the previous document versions.

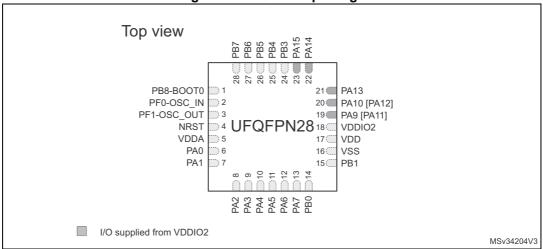
UVSS
| PB8-BOOT0
| PB7
| PB6
| PB6
| PB4
| PB3 Top view 32 31 33 29 28 27 26 25 VDD [ 24 PA14 PF0-OSC\_IN 23 PA13 PF1-OSC\_OUT 
NRST 22 PA12 21 PA11 LQFP32 VDDA □ 20 🗖 PA10 PA0 | PA1 | PA2 | 6 19 🔲 PA9 18 PA8 17 VDDIO2 8 6 4 6 4 6 9 PA3 C I/O supplied from VDDIO2 MSv34202V2

Figure 6. LQFP32 package pinout

PB8
PF11-BOOT0
PB7
PB6
PB5
PB4
PB3
PA15 Top view VDD 24 📖 PA14 0 PF0-OSC\_IN 23 🚛 PA13 PF1-OSC OUT 22 PA12 UFQFPN32 21 PA11 NRST VDDA PA0 19 PA9 PA1 18 📖 PA8 PA2 VDDIO2 17( VSS PA3 PA4 PA5 PA6 PA7 PB0 PB1 I/O supplied from VDDIO2 MSv34203V3

Figure 7. UFQFPN32 package pinout





1. Pin pair PA11/12 can be remapped in place of pin pair PA9/10 using the SYSCFG\_CFGR1 register.

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Top view TSSOP20 PA14 PB8-BOOT0 □ PF0-OSC\_IN □ 19 □ PA13 PF1-OSC\_OUT □ 18 PA10 [PA12] \_ NRST □ 4 17 → PA9 [PA11] ─ VDD VDDA ⊏ 16 15 □ vss PA0 □ PA1 □ □ PB1 13 PA7 PA2 □ 8 9 12 \_\_ PA6 PA3 ⊏ 10 11 PA4 □ □ PA5 MSv34205V2

Figure 9. TSSOP20 package

1. Pin pair PA11/12 can be remapped in place of pin pair PA9/10 using the SYSCFG\_CFGR1 register.

Table 12. Legend/abbreviations used in the pinout table

Na	me	Abbreviation	Abbreviation Definition					
Pin r	name	Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name						
		S	Supply pin					
Pin	type	I/O	Input / output pin					
		FT	5 V-tolerant I/O					
		FTf	FTf 5 V-tolerant I/O, FM+ capable					
		TTa 3.3 V-tolerant I/O directly connected to ADC						
I/O str	ucture	TC Standard 3.3 V I/O						
		RST	Bidirectional reset pin with embedded weak pull-up resistor					
No	tes	Unless otherwise reset.	specified by a note, all I/Os are set as floating inputs during and after					
Pin	Alternate functions	Functions selected through GPIOx_AFR registers						
functions	Additional functions	Functions directly	selected/enabled through peripheral registers					

Table 13. STM32F042x pin definitions

		Pin nı	umbe	rs				-		Pin functions		
LQFP48/UFQFPN48	WLCSP36	LQFP32	UFQFPN32	UFQFPN28	TSSPOP20	Pin name (function upon reset)	Pin type	I/O structure	Notes	Alternate function	Additional functions	
1	-	-	-	-	-	VBAT	S	-	-	Backup power s	upply	
2	A6	-	-	-	-	PC13	I/O	TC	(1) (2)	-	WKUP2, RTC_TAMP1, RTC_TS, RTC_OUT	
3	В6	-	1	-	-	PC14- OSC32_IN (PC14)	I/O	тс	(1) (2)	-	OSC32_IN	
4	C6	-	-	-	-	PC15- OSC32_OUT (PC15)	I/O	TC	(1) (2)	-	OSC32_OUT	
5	B5	2	2	2	2	PF0-OSC_IN (PF0)	I/O	FTf	-	CRS_SYNC I2C1_SDA	OSC_IN	
6	C5	3	3	3	3	PF1-OSC_OUT (PF1)	I/O	FTf	-	I2C1_SCL	OSC_OUT	
7	D5	4	4	4	4	NRST	I/O	RST	-	Device reset input / interr (active low		
8	D6	32	0	16	15	VSSA	S		(3)	Analog grou	nd	
9	E5	5	5	5	5	VDDA	S		-	Analog power s	upply	
10	F6	6	6	6	6	PA0	I/O	ТТа	-	USART2_CTS, TIM2_CH1_ETR, TSC_G1_IO1	RTC_ TAMP2, WKUP1, ADC_IN0,	
11	D4	7	7	7	7	PA1	I/O	ТТа	-	USART2_RTS, TIM2_CH2, TSC_G1_IO2, EVENTOUT	ADC_IN1	
12	E4	8	8	8	8	PA2	I/O	TTa	-	USART2_TX, TIM2_CH3, TSC_G1_IO3	ADC_IN2, WKUP4	
13	F5	9	9	9	9	PA3	I/O	ТТа	-	USART2_RX, TIM2_CH4, TSC_G1_IO4	ADC_IN3	

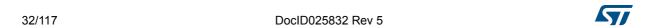


Table 13. STM32F042x pin definitions (continued)

	ı	Pin nı	umbe							Pin function	s
LQFP48/UFQFPN48	WLCSP36	LQFP32	UFQFPN32	UFQFPN28	TSSPOP20	Pin name (function upon reset)	Pin type	I/O structure	Notes	Alternate function	Additional functions
14	C3	10	10	10	10	PA4	I/O	TTa	-	SPI1_NSS, I2S1_WS, TIM14_CH1, TSC_G2_IO1, USART2_CK USB_NOE	ADC_IN4
15	D3	11	11	11	11	PA5	I/O	ТТа	-	SPI1_SCK, I2S1_CK, CEC, TIM2_CH1_ETR, TSC_G2_IO2	ADC_IN5
16	E3	12	12	12	12	PA6	I/O	TTa	-	SPI1_MISO, I2S1_MCK, TIM3_CH1, TIM1_BKIN, TIM16_CH1, TSC_G2_IO3, EVENTOUT	ADC_IN6
17	F4	13	13	13	13	PA7	I/O	ТТа	-	SPI1_MOSI, I2S1_SD, TIM3_CH2, TIM14_CH1, TIM1_CH1N, TIM17_CH1, TSC_G2_IO4, EVENTOUT	ADC_IN7
18	F3	14	14	14	-	PB0	I/O	ТТа	-	TIM3_CH3, TIM1_CH2N, TSC_G3_IO2, EVENTOUT	ADC_IN8
19	F2	15	15	15	14	PB1	I/O	ТТа	-	TIM3_CH4, TIM14_CH1, TIM1_CH3N, TSC_G3_IO3	ADC_IN9
20	D2	-	16	-	-	PB2	I/O	FT	-	TSC_G3_IO4	-
21	-	-	-	-	-	PB10	I/O	FTf	-	SPI2_SCK, CEC, TSC_SYNC, TIM2_CH3, I2C1_SCL	-
22	ı	-	-	-	-	PB11	I/O	FTf	-	TIM2_CH4, EVENTOUT, I2C1_SDA	-
23	F1	16	0	16	15	VSS	S	-	-	Ground	
24	-	-	-	17	16	VDD	S	-	-	Digital power supply	
25	-	-	-	-	-	PB12	I/O	FT	-	TIM1_BKIN, SPI2_NSS, EVENTOUT	-



Table 13. STM32F042x pin definitions (continued)

	I	Pin nı	umbe			e 13. 31W321 042X pin deminions				Pin functions			
LQFP48/UFQFPN48	WLCSP36	LQFP32	UFQFPN32	UFQFPN28	TSSPOP20	Pin name (function upon reset)	Pin type	I/O structure	Notes	Alternate function	Additional functions		
26	1	-	-	-	-	PB13	I/O	FTf	-	SPI2_SCK, TIM1_CH1N, I2C1_SCL	-		
27	1	1	ı	-	-	PB14	I/O	FTf	-	SPI2_MISO, TIM1_CH2N, I2C1_SDA	-		
28	ı	-	-	-	-	PB15	I/O	FT	-	SPI2_MOSI, TIM1_CH3N	WKUP7, RTC_REFIN		
29	E2	18	18	-	-	PA8	I/O	FT	(4)	USART1_CK, TIM1_CH1, EVENTOUT, MCO, CRS_SYNC	-		
30	D1	19	19	19	17	PA9	I/O	FTf	(4)	USART1_TX, TIM1_CH2, TSC_G4_IO1, I2C1_SCL	-		
31	C1	20	20	20	18	PA10	I/O	FTf	(4)	USART1_RX, TIM1_CH3, TIM17_BKIN, TSC_G4_IO2, I2C1_SDA	-		
32	C2	21	21	19 <sup>(5)</sup>	17 <sup>(5)</sup>	PA11	I/O	FTf	(4)	CAN_RX, USART1_CTS, TIM1_CH4, TSC_G4_IO3, EVENTOUT, I2C1_SCL	USB_DM		
33	A1	22	22	20 <sup>(5)</sup>	18 <sup>(5)</sup>	PA12	I/O	FTf	(4)	CAN_TX,USART1_RTS, TIM1_ETR, TSC_G4_IO4, EVENTOUT, I2C1_SDA	USB_DP		
34	B1	23	23	21	19	PA13	I/O	FT	(4) (6)	IR_OUT, SWDIO USB_NOE	-		
35	-	-	-	-	-	VSS	S	-	-	Ground			
36	E1	17	17	18	16	VDDIO2	S	-	-	Digital power supply			
37	B2	24	24	22	20	PA14	I/O	FT	(4) (6)	USART2_TX, SWCLK	-		

Table 13. STM32F042x pin definitions (continued)

	I	Pin nı	umbe	rs			-			Pin functions	
LQFP48/UFQFPN48	WLCSP36	LQFP32	UFQFPN32	UFQFPN28	TSSPOP20	Pin name (function upon reset)	Pin type	I/O structure	Notes	Alternate function	Additional functions
38	A2	25	25	23	-	PA15	I/O	FT	(4)	SPI1_NSS, I2S1_WS, USART2_RX, TIM2_CH1_ETR, EVENTOUT, USB_NOE	-
39	В3	26	26	24	-	PB3	I/O	FT	1	SPI1_SCK, I2S1_CK, TIM2_CH2, TSC_G5_IO1, EVENTOUT	-
40	A3	27	27	25	-	PB4	I/O	FT	-	SPI1_MISO, I2S1_MCK, TIM17_BKIN, TIM3_CH1, TSC_G5_IO2, EVENTOUT	-
41	E6	28	28	26	-	PB5	I/O	FT	-	SPI1_MOSI, I2S1_SD, I2C1_SMBA, TIM16_BKIN, TIM3_CH2	WKUP6
42	C4	29	29	27	-	PB6	I/O	FTf	-	I2C1_SCL, USART1_TX, TIM16_CH1N, TSC_G5_I03	-
43	A4	30	30	28	-	PB7	I/O	FTf	1	I2C1_SDA, USART1_RX, TIM17_CH1N, TSC_G5_IO4	-
44	-	-	31	-	-	PF11-BOOT0	I/O	FT	-	-	Boot memory selection
-	B4	31	-	1	1	PB8-BOOT0	I/O	FTf	1	I2C1_SCL, CEC, TIM16_CH1, TSC_SYNC, CAN_RX	Boot memory selection
45	-	-	32	-	-	PB8	I/O	FTf	-	I2C1_SCL, CEC, TIM16_CH1, TSC_SYNC, CAN_RX	-



Pin numbers										Pin functions		
LQFP48/UFQFPN48	WLCSP36	LQFP32	UFQFPN32	UFQFPN28	TSSPOP20	Pin name (function upon reset)	Pin type	I/O structure	Notes	Alternate function	Additional functions	
46	1	1	-	-	-	PB9	I/O	FTf	1	SPI2_NSS, I2C1_SDA, IR_OUT, TIM17_CH1, EVENTOUT, CAN_TX	·	
47	-	32	0	-	-	VSS	S	-	-	Ground		
48	A5	1	1	-	-	VDD	S	-	-	Digital power supply		

Table 13. STM32F042x pin definitions (continued)

- 4. PA8, PA9, PA10, PA11, PA12, PA13, PA14 and PA15 I/Os are supplied by VDDIO2.
- 5. Pin pair PA11/12 can be remapped in place of pin pair PA9/10 using SYSCFG\_CFGR1 register.
- 6. After reset, these pins are configured as SWDIO and SWCLK alternate functions, and the internal pull-up on the SWDIO pin and the internal pull-down on the SWCLK pin are activated.



PC13, PC14 and PC15 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIOs PC13 to PC15 in output mode is limited:

- The speed should not exceed 2 MHz with a maximum load of 30 pF.

<sup>-</sup> These GPIOs must not be used as current sources (e.g. to drive an LED).

<sup>2.</sup> After the first RTC domain power-up, PC13, PC14 and PC15 operate as GPIOs. Their function then depends on the content of the RTC registers which are not reset by the system reset. For details on how to manage these GPIOs, refer to the RTC domain and RTC register descriptions in the reference manual.

<sup>3.</sup> Distinct VSSA pin is only available on 48-pin packages. On all other packages, the pin number corresponds to the VSS pin to which VSSA pad of the silicon die is connected.



Table 14. Alternate functions selected through GPIOA\_AFR registers for port A

Pin name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PA0	-	USART2_CTS	TIM2_CH1_ETR	TSC_G1_IO1	-	-	-	-
PA1	EVENTOUT	USART2_RTS	TIM2_CH2	TSC_G1_IO2	-	-	-	-
PA2	-	USART2_TX	TIM2_CH3	TSC_G1_IO3	-	-	-	-
PA3	-	USART2_RX	TIM2_CH4	TSC_G1_IO4	-	-	-	-
PA4	SPI1_NSS, I2S1_WS	USART2_CK	USB_NOE	TSC_G2_IO1	TIM14_CH1	-	-	-
PA5	SPI1_SCK, I2S1_CK	CEC	TIM2_CH1_ETR	TSC_G2_IO2	-	-	-	-
PA6	SPI1_MISO, I2S1_MCK	TIM3_CH1	TIM1_BKIN	TSC_G2_IO3	-	TIM16_CH1	EVENTOUT	-
PA7	SPI1_MOSI, I2S1_SD	TIM3_CH2	TIM1_CH1N	TSC_G2_IO4	TIM14_CH1	TIM17_CH1	EVENTOUT	-
PA8	MCO	USART1_CK	TIM1_CH1	EVENTOUT	CRS_SYNC	-	-	-
PA9	-	USART1_TX	TIM1_CH2	TSC_G4_IO1	I2C1_SCL	MCO	-	-
PA10	TIM17_BKIN	USART1_RX	TIM1_CH3	TSC_G4_IO2	I2C1_SDA	-	-	-
PA11	EVENTOUT	USART1_CTS	TIM1_CH4	TSC_G4_IO3	CAN_RX	I2C1_SCL	-	-
PA12	EVENTOUT	USART1_RTS	TIM1_ETR	TSC_G4_IO4	CAN_TX	I2C1_SDA	-	-
PA13	SWDIO	IR_OUT	USB_NOE	-	-	-	-	-
PA14	SWCLK	USART2_TX	-	-	-	-	-	-
PA15	SPI1_NSS, I2S1_WS	USART2_RX	TIM2_CH1_ETR	EVENTOUT	-	USB_NOE	-	-

Table 15. Alternate functions selected through GPIOB\_AFR registers for port B

Pin name	AF0	AF1	AF2	AF3	AF4	AF5
PB0	EVENTOUT	TIM3_CH3	TIM1_CH2N	TSC_G3_IO2	-	-
PB1	TIM14_CH1	TIM3_CH4	TIM1_CH3N	TSC_G3_IO3	-	-
PB2	-	-	-	TSC_G3_IO4	-	-
PB3	SPI1_SCK, I2S1_CK	EVENTOUT	TIM2_CH2	TSC_G5_IO1	-	-
PB4	SPI1_MISO, I2S1_MCK	TIM3_CH1	EVENTOUT	TSC_G5_IO2	-	TIM17_BKIN
PB5	SPI1_MOSI, I2S1_SD	TIM3_CH2	TIM16_BKIN	I2C1_SMBA	-	-
PB6	USART1_TX	I2C1_SCL	TIM16_CH1N	TSC_G5_IO3	-	-
PB7	USART1_RX	I2C1_SDA	TIM17_CH1N	TSC_G5_IO4	-	-
PB8	CEC	I2C1_SCL	TIM16_CH1	TSC_SYNC	CAN_RX	-
PB9	IR_OUT	I2C1_SDA	TIM17_CH1	EVENTOUT	CAN_TX	SPI2_NSS
PB10	CEC	I2C1_SCL	TIM2_CH3	TSC_SYNC	-	SPI2_SCK
PB11	EVENTOUT	I2C1_SDA	TIM2_CH4	-	-	-
PB12	SPI2_NSS	EVENTOUT	TIM1_BKIN	-	-	-
PB13	SPI2_SCK	-	TIM1_CH1N	-	-	I2C1_SCL
PB14	SPI2_MISO	-	TIM1_CH2N	-	-	I2C1_SDA
PB15	SPI2_MOSI	-	TIM1_CH3N	-	-	-

## Table 16. Alternate functions selected through GPIOF\_AFR registers for port F

Pin name	AF0	AF1
PF0	CRS_SYNC	I2C1_SDA
PF1	-	I2C1_SCL



#### 5 **Memory mapping**

To the difference of STM32F042x6 memory map in Figure 10, the two bottom code memory spaces of STM32F042x4 end at 0x0000 3FFF and 0x0800 3FFF, respectively.

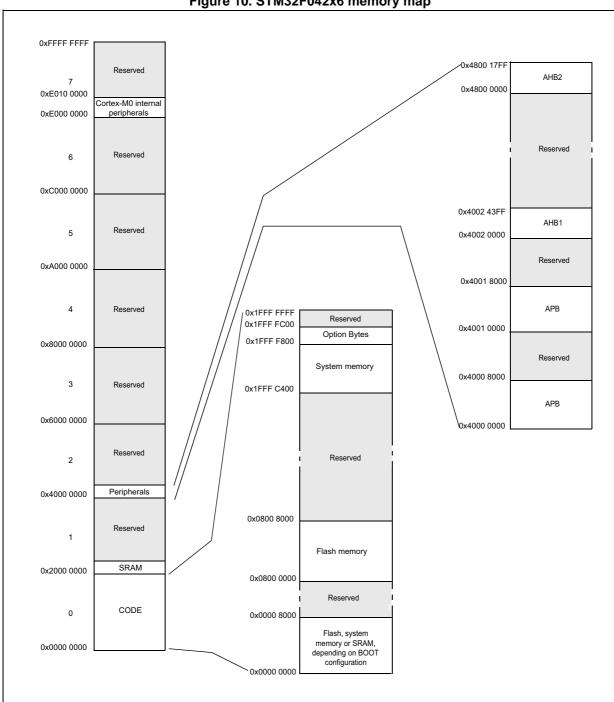


Figure 10. STM32F042x6 memory map

MSv34924V2

Table 17. STM32F042x4/x6 peripheral register boundary addresses

Bus	Boundary address	Size	Peripheral
	0x4800 1800 - 0x5FFF FFFF	~384 MB	Reserved
	0x4800 1400 - 0x4800 17FF	1 KB	GPIOF
	0x4800 0C00 - 0x4800 13FF	2 KB	Reserved
AHB2	0x4800 0800 - 0x4800 0BFF	1 KB	GPIOC
	0x4800 0400 - 0x4800 07FF	1 KB	GPIOB
	0x4800 0000 - 0x4800 03FF	1 KB	GPIOA
	0x4002 4400 - 0x47FF FFFF	~128 MB	Reserved
	0x4002 4000 - 0x4002 43FF	1 KB	TSC
	0x4002 3400 - 0x4002 3FFF	3 KB	Reserved
	0x4002 3000 - 0x4002 33FF	1 KB	CRC
	0x4002 2400 - 0x4002 2FFF	3 KB	Reserved
AHB1	0x4002 2000 - 0x4002 23FF	1 KB	Flash memory interface
	0x4002 1400 - 0x4002 1FFF	3 KB	Reserved
	0x4002 1000 - 0x4002 13FF	1 KB	RCC
	0x4002 0400 - 0x4002 0FFF	3 KB	Reserved
	0x4002 0000 - 0x4002 03FF	1 KB	DMA
	0x4001 8000 - 0x4001 FFFF	32 KB	Reserved
	0x4001 5C00 - 0x4001 7FFF	9 KB	Reserved
	0x4001 5800 - 0x4001 5BFF	1 KB	DBGMCU
	0x4001 4C00 - 0x4001 57FF	3 KB	Reserved
	0x4001 4800 - 0x4001 4BFF	1 KB	TIM17
	0x4001 4400 - 0x4001 47FF	1 KB	TIM16
	0x4001 3C00 - 0x4001 43FF	2 KB	Reserved
	0x4001 3800 - 0x4001 3BFF	1 KB	USART1
APB	0x4001 3400 - 0x4001 37FF	1 KB	Reserved
	0x4001 3000 - 0x4001 33FF	1 KB	SPI1/I2S1
	0x4001 2C00 - 0x4001 2FFF	1 KB	TIM1
	0x4001 2800 - 0x4001 2BFF	1 KB	Reserved
Ī	0x4001 2400 - 0x4001 27FF	1 KB	ADC
	0x4001 0800 - 0x4001 23FF	7 KB	Reserved
Ī	0x4001 0400 - 0x4001 07FF	1 KB	EXTI
	0x4001 0000 - 0x4001 03FF	1 KB	SYSCFG
	0x4000 8000 - 0x4000 FFFF	32 KB	Reserved

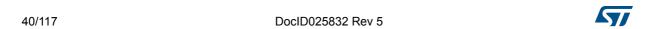


Table 17. STM32F042x4/x6 peripheral register boundary addresses (continued)

Bus	Boundary address	Size	Peripheral
	0x4000 7C00 - 0x4000 7FFF	1 KB	Reserved
	0x4000 7800 - 0x4000 7BFF	1 KB	CEC
	0x4000 7400 - 0x4000 77FF	1 KB	Reserved
	0x4000 7000 - 0x4000 73FF	1 KB	PWR
	0x4000 6C00 - 0x4000 6FFF	1 KB	CRS
	0x4000 6800 - 0x4000 6BFF0	1 KB	Reserved
	0x4000 6400 - 0x4000 67FF	1 KB	BxCAN
	0x4000 6000 - 0x4000 63FF	1 KB	USB/CAN RAM
	0x4000 5C00 - 0x4000 5FFF	1 KB	USB
	0x4000 5800 - 0x4000 5BFF	1 KB	Reserved
	0x4000 5400 - 0x4000 57FF	1 KB	I2C1
	0x4000 4800 - 0x4000 53FF	3 KB	Reserved
A DD	0x4000 4400 - 0x4000 47FF	1 KB	USART2
APB	0x4000 3C00 - 0x4000 43FF	2 KB	Reserved
	0x4000 3800 - 0x4000 3BFF	1 KB	SPI2
	0x4000 3400 - 0x4000 37FF	1 KB	Reserved
	0x4000 3000 - 0x4000 33FF	1 KB	IWDG
	0x4000 2C00 - 0x4000 2FFF	1 KB	WWDG
	0x4000 2800 - 0x4000 2BFF	1 KB	RTC
	0x4000 2400 - 0x4000 27FF	1 KB	Reserved
	0x4000 2000 - 0x4000 23FF	1 KB	TIM14
	0x4000 0800 - 0x4000 1FFF	6 KB	Reserved
	0x4000 0400 - 0x4000 07FF	1 KB	TIM3
	0x4000 0000 - 0x4000 03FF	1 KB	TIM2

## 6 Electrical characteristics

#### 6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V<sub>SS</sub>.

#### 6.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at  $T_A = 25$  °C and  $T_A = T_A$ max (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean  $\pm 3\sigma$ ).

## 6.1.2 Typical values

Unless otherwise specified, typical data are based on  $T_A = 25$  °C,  $V_{DD} = V_{DDA} = 3.3$  V. They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean  $\pm 2\sigma$ ).

# 6.1.3 Typical curves

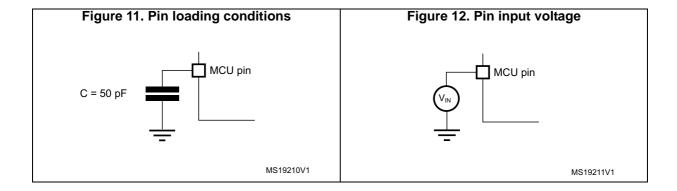
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

### 6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in *Figure 11*.

#### 6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in Figure 12.



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# 6.1.6 Power supply scheme

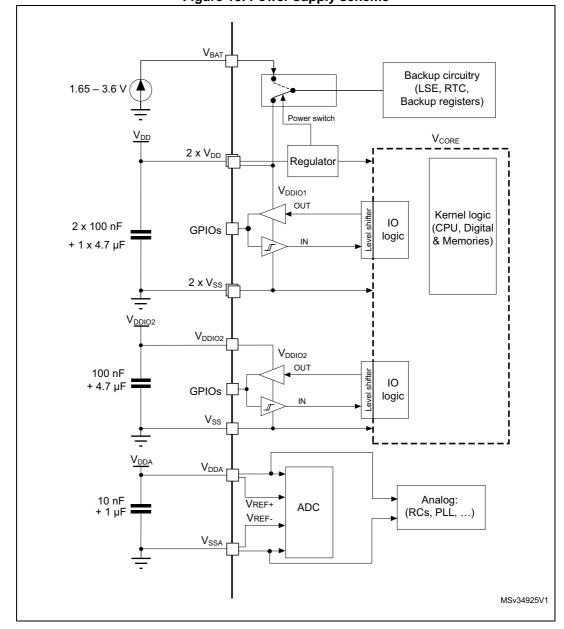


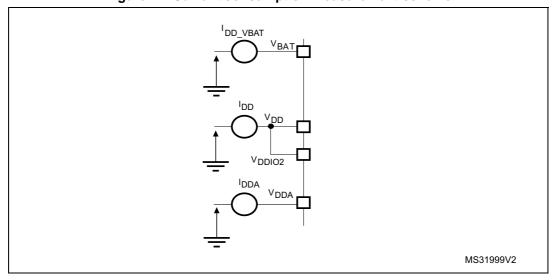
Figure 13. Power supply scheme

Caution:

Each power supply pair ( $V_{DD}/V_{SS}$ ,  $V_{DDA}/V_{SSA}$  etc.) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure the good functionality of the device.

# 6.1.7 Current consumption measurement

Figure 14. Current consumption measurement scheme



# 6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in *Table 18: Voltage characteristics*, *Table 19: Current characteristics* and *Table 20: Thermal characteristics* may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Table 18. Voltage characteristics<sup>(1)</sup>

Symbol	Ratings	Min	Max	Unit
$V_{DD}$ – $V_{SS}$	External main supply voltage	- 0.3	4.0	V
V <sub>DDIO2</sub> -V <sub>SS</sub>	External I/O supply voltage	- 0.3	4.0	V
V <sub>DDA</sub> -V <sub>SS</sub>	External analog supply voltage	- 0.3	4.0	V
V <sub>DD</sub> –V <sub>DDA</sub>	Allowed voltage difference for $V_{DD} > V_{DDA}$	-	0.4	٧
V <sub>BAT</sub> -V <sub>SS</sub>	External backup supply voltage	- 0.3	4.0	V
	Input voltage on FT and FTf pins	V <sub>SS</sub> - 0.3	V <sub>DDIOx</sub> + 4.0 <sup>(3)</sup>	٧
$V_{IN}^{(2)}$	Input voltage on TTa pins	V <sub>SS</sub> - 0.3	4.0	V
	Input voltage on any other pin	V <sub>SS</sub> - 0.3	4.0	V
$ \Delta V_{DDx} $	Variations between different V <sub>DD</sub> power pins	-	50	mV
V <sub>SSx</sub> - V <sub>SS</sub>	Variations between all the different ground pins	-	50	mV
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (human body model) see Section 6.3.12: Electrical sensitivity characteristics			-

All main power (V<sub>DD</sub>, V<sub>DDA</sub>) and ground (V<sub>SS</sub>, V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range.

<sup>2.</sup> V<sub>IN</sub> maximum must always be respected. Refer to *Table 19: Current characteristics* for the maximum allowed injected current values.

Valid only if the internal pull-up/pull-down resistors are disabled. If internal pull-up or pull-down resistor is enabled, the maximum limit is 4 V.

**Table 19. Current characteristics** 

Symbol	Ratings		Unit
ΣI <sub>VDD</sub>	ΣI <sub>VDD</sub> Total current into sum of all VDD power lines (source) <sup>(1)</sup>		
ΣI <sub>VSS</sub>	Total current out of sum of all VSS ground lines (sink) <sup>(1)</sup>	-120	
I <sub>VDD(PIN)</sub>	Maximum current into each VDD power pin (source) <sup>(1)</sup>	100	
I <sub>VSS(PIN)</sub>	Maximum current out of each VSS ground pin (sink) <sup>(1)</sup>	-100	
	Output current sunk by any I/O and control pin	25	
I <sub>IO(PIN)</sub>	Output current source by any I/O and control pin	-25	
	Total output current sunk by sum of all I/Os and control pins <sup>(2)</sup>	80	
$\Sigma I_{IO(PIN)}$	Total output current sourced by sum of all I/Os and control pins <sup>(2)</sup>	-80	mA
	Total output current sourced by sum of all I/Os supplied by VDDIO2	-40	
	Injected current on FT and FTf pins	-5/+0 <sup>(4)</sup>	
I <sub>INJ(PIN)</sub> <sup>(3)</sup>	Injected current on TC and RST pin	± 5	1
	Injected current on TTa pins <sup>(5)</sup>	± 5	
ΣΙ <sub>ΙΝJ(PIN)</sub>	Total injected current (sum of all I/O and control pins) <sup>(6)</sup>	± 25	

- 1. All main power (VDD, VDDA) and ground (VSS, VSSA) pins must always be connected to the external power supply, in the permitted range.
- 2. This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count QFP packages.
- 3. A positive injection is induced by  $V_{IN} > V_{DDIOx}$  while a negative injection is induced by  $V_{IN} < V_{SS}$ .  $I_{INJ(PIN)}$  must never be exceeded. Refer to *Table 18: Voltage characteristics* for the maximum allowed input voltage values.
- 4. Positive injection is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.
- On these I/Os, a positive injection is induced by V<sub>IN</sub> > V<sub>DDA</sub>. Negative injection disturbs the analog performance of the device. See note <sup>(2)</sup> below *Table 56: ADC accuracy*.
- When several inputs are submitted to a current injection, the maximum ΣI<sub>INJ(PIN)</sub> is the absolute sum of the positive and negative injected currents (instantaneous values).

**Table 20. Thermal characteristics** 

Symbol	Ratings	Value	Unit
T <sub>STG</sub>	Storage temperature range	-65 to +150	°C
T <sub>J</sub>	Maximum junction temperature	150	°C



# 6.3 Operating conditions

# 6.3.1 General operating conditions

Table 21. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit	
f <sub>HCLK</sub>	Internal AHB clock frequency	-	0	48	MHz	
f <sub>PCLK</sub>	Internal APB clock frequency	-	0	48	IVITZ	
V <sub>DD</sub>	Standard operating voltage	-	2.0	3.6	V	
V <sub>DDIO2</sub>	I/O supply voltage	Must not be supplied if V <sub>DD</sub> is not present	1.65	3.6	V	
V	Analog operating voltage (ADC not used)	Must have a potential equal	$V_{DD}$	3.6	\/	
$V_{DDA}$	Analog operating voltage (ADC used)	to or higher than V <sub>DD</sub>	2.4	3.6	V	
$V_{BAT}$	Backup operating voltage		1.65	3.6	V	
		TC and RST I/O	-0.3	V <sub>DDIOx</sub> +0.3	V	
$V_{IN}$	I/O input voltage	TTa I/O	-0.3	V <sub>DDA</sub> +0.3 <sup>(1)</sup>		
		FT and FTf I/O	-0.3	5.5 <sup>(1)</sup>		
		LQFP48	-	364		
	Power dissipation at T <sub>A</sub> = 85 °C	UFQFPN48	-	606		
		WLCSP36	-	313		
$P_{D}$	for suffix 6 or T <sub>A</sub> = 105 °C for	LQFP32	-	351	mW	
	suffix 7 <sup>(2)</sup>	UFQFPN32	-	526	1	
		UFQFPN28	-	170		
		TSSOP20	-	263		
	Ambient temperature for the	Maximum power dissipation	-40	85	00	
т.	suffix 6 version	Low power dissipation <sup>(3)</sup>	-40	105	°C	
TA	Ambient temperature for the	Maximum power dissipation	-40	105	°C	
	suffix 7 version	Low power dissipation <sup>(3)</sup>	-40	125	°C	
т.	lunction towns are true are a	Suffix 6 version	-40	105	°C	
TJ	Junction temperature range	Suffix 7 version	-40	125	°C	

<sup>1.</sup> For operation with a voltage higher than  $V_{DDIOx}$  + 0.3 V, the internal pull-up resistor must be disabled.

### 6.3.2 Operating conditions at power-up / power-down

The parameters given in *Table 22* are derived from tests performed under the ambient temperature condition summarized in *Table 21*.



<sup>2.</sup> If  $T_A$  is lower, higher  $P_D$  values are allowed as long as  $T_J$  does not exceed  $T_{Jmax}$ . See Section 7.8: Thermal characteristics.

In low power dissipation state, T<sub>A</sub> can be extended to this range as long as T<sub>J</sub> does not exceed T<sub>Jmax</sub> (see Section 7.8: Thermal characteristicsSection 7.8: Thermal characteristics).

Table 22. Operating conditions at power-up / power-down

Symbol	Parameter	Conditions	Min	Max	Unit
t <sub>VDD</sub>	V <sub>DD</sub> rise time rate	_	0	8	
	V <sub>DD</sub> fall time rate	_	20	8	υοΛ/
t <sub>VDDA</sub>	V <sub>DDA</sub> rise time rate		0	8	μs/V
	V <sub>DDA</sub> fall time rate	_	20	8	

# 6.3.3 Embedded reset and power control block characteristics

The parameters given in *Table 23* are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*.

Table 23. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>POR/PDR</sub> <sup>(1)</sup>	reset threshold	Falling edge <sup>(2)</sup>	1.80	1.88	1.96 <sup>(3)</sup>	V
YPOR/PDR		Rising edge	1.84 <sup>(3)</sup>	1.92	2.00	V
V <sub>PDRhyst</sub>	PDR hysteresis	-	-	40	-	mV
t <sub>RSTTEMPO</sub> <sup>(4)</sup>	Reset temporization	-	1.50	2.50	4.50	ms

<sup>1.</sup> The PDR detector monitors  $V_{DD}$  and also  $V_{DDA}$  (if kept enabled in the option bytes). The POR detector monitors only  $V_{DD}$ .

Table 24. Programmable voltage detector characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V	PVD threshold 0	Rising edge	2.1	2.18	2.26	V
V <sub>PVD0</sub>	F VD tillesiloid 0	Falling edge	2	2.08	2.16	V
V	PVD threshold 1	Rising edge	2.19	2.28	2.37	V
V <sub>PVD1</sub>	PVD (nreshold 1	Falling edge	2.09	2.18	2.27	V
V	PVD threshold 2	Rising edge	2.28	2.38	2.48	V
V <sub>PVD2</sub>		Falling edge	2.18	2.28	2.38	V
V	PVD threshold 3	Rising edge	2.38	2.48	2.58	V
V <sub>PVD3</sub>		Falling edge	2.28	2.38	2.48	V
V	PVD threshold 4	Rising edge	2.47	2.58	2.69	V
V <sub>PVD4</sub>	PVD threshold 4	Falling edge	2.37	2.48	2.59	V
V	DVD throshold 5	Rising edge	2.57	2.68	2.79	V
V <sub>PVD5</sub>	PVD threshold 5	Falling edge	2.47	2.58	2.69	V



<sup>2.</sup> The product behavior is guaranteed by design down to the minimum  $V_{POR/PDR}$  value.

<sup>3.</sup> Data based on characterization results, not tested in production.

<sup>4.</sup> Guaranteed by design, not tested in production.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V	PVD threshold 6	Rising edge	2.66	2.78	2.9	V
V <sub>PVD6</sub>	F VD tillesiloid o	Falling edge	2.56	2.68	2.8 V 3 V 2.9 V	V
V	PVD threshold 7	Rising edge	2.76	2.88	3	V
V <sub>PVD7</sub>	F VD tillesiloid 7	Falling edge	2.66	2.78	2.9	V
V <sub>PVDhyst</sub> <sup>(1)</sup>	PVD hysteresis	-	-	100	-	mV
I <sub>DD(PVD)</sub>	PVD current consumption	-	-	0.15	0.26 <sup>(1)</sup>	μΑ

Table 24. Programmable voltage detector characteristics (continued)

### 6.3.4 Embedded reference voltage

The parameters given in *Table 25* are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*.

	Table 201 Elliboa	aoa iiitoiiiai roioroiiot	voltag	•		
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{REFINT}$	Internal reference voltage	-40 °C < T <sub>A</sub> < +105 °C	1.2	1.23	1.25	V
t <sub>START</sub>	ADC_IN17 buffer startup time	-	-	-	10 <sup>(1)</sup>	μs
t <sub>S_vrefint</sub>	ADC sampling time when reading the internal reference voltage	-	4 <sup>(1)</sup>	-	-	μs
$\Delta V_{REFINT}$	Internal reference voltage spread over the temperature range	V <sub>DDA</sub> = 3 V	-	-	10 <sup>(1)</sup>	mV
T <sub>Coeff</sub>	Temperature coefficient	-	- 100 <sup>(1)</sup>	-	100 <sup>(1)</sup>	ppm/°C

Table 25. Embedded internal reference voltage

# 6.3.5 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in *Figure 14: Current consumption measurement scheme*.

All Run-mode current consumption measurements given in this section are performed with a reduced code that gives a consumption equivalent to CoreMark code.

<sup>1.</sup> Guaranteed by design, not tested in production.

<sup>1.</sup> Guaranteed by design, not tested in production.

### Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in analog input mode
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time is adjusted to the f<sub>HCLK</sub> frequency:
  - 0 wait state and Prefetch OFF from 0 to 24 MHz
  - 1 wait state and Prefetch ON above 24 MHz
- When the peripherals are enabled f<sub>PCLK</sub> = f<sub>HCLK</sub>

The parameters given in *Table 26* to *Table 28* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*.

Table 26. Typical and maximum current consumption from  $V_{DD}$  supply at  $V_{DD}$  = 3.6 V

	er.			All	peripher	als enab	led <sup>(1)</sup>	All peripherals disabled				
Symbol	Parameter	Conditions	f <sub>HCLK</sub>		M	lax @ T <sub>A</sub>	(2)		М	ax @ T <sub>A</sub>	(2)	Unit
S	Par			Тур	25 °C	85 °C	105 °C	Тур	25 °C	85 °C	105 °C	
		HSI48	48 MHz	20.3	23.2	23.4	24.6	12.7	14.4	14.4	14.7	
	ory		48 MHz	20.2	22.9	23.0	23.9	12.6	14.1	14.3	14.4	
	Run mode, HSE bypass, PLL on P	32 MHz	14.0	16.0	16.1	16.7	8.7	9.5	9.7	10.3		
	Run mode, Flash mem		24 MHz	11.0	13.5	13.7	13.8	6.9	7.6	7.8	8.2	
		HSE bypass,	8 MHz	3.9	5.2	5.3	5.6	2.6	3.1	3.2	3.3	
I <sub>DD</sub>	current in	PLL off	1 MHz	0.9	1.3	1.5	1.8	0.7	1.0	1.1	1.3	mA
	curr		48 MHz	20.5	23.1	23.3	23.6	12.8	14.6	14.6	15.0	
	Supply current in le executing from	HSI clock, PLL on	32 MHz	14.3	15.6	15.9	17.0	8.6	9.5	9.7	10.0	
	Sul Sul		24 MHz	11.2	13.6	13.8	14.8	6.9	7.4	7.5	7.7	
	8	HSI clock, PLL off	8 MHz	4.1	5.2	5.3	5.6	2.6	3.1	3.1	3.3	

Table 26. Typical and maximum current consumption from  $V_{DD}$  supply at  $V_{DD}$  = 3.6 V (continued)

	7.			All	peripher	als enab	oled <sup>(1)</sup>	All	periphe	rals disa	abled		
Symbol	Parameter	Conditions	f <sub>HCLK</sub>		M	lax @ T <sub>A</sub>	(2)		М	ax @ T <sub>A</sub>	(2)	Unit	
Sy	Para			Тур	25 °C	85 °C	105 °C	Тур	25 °C	85 °C	105 °C		
		HSI48	48 MHz	19.3	21.9	22.1	23.7	11.9	13.4	13.6	13.7		
		HSE bypass, PLL on	48 MHz	19.2	21.8 <sup>(3)</sup>	22.0	22.1 <sup>(3)</sup>	11.7	13.3 <sup>(3)</sup>	13.5	13.7 <sup>(3)</sup>		
	ode,		32 MHz	13.4	15.8	15.9	16.0	7.9	8.8	8.9	9.7		
	= E		24 MHz	10.3	12.6	13.0	13.4	6.2	8.0	8.2	8.3		
	n Ru g fro	HSE bypass,	8 MHz	3.6	4.1	4.3	4.4	2.0	2.1	2.1	2.5		
	ent i uting	PLL off	1 MHz	8.0	0.9	0.9	1.1	0.4	0.5	0.6	0.8		
		HSI clock,	48 MHz	19.5	22.0	22.1	22.5	11.8	13.6	13.8	13.9		
		ode	HSI clock, PLL on	32 MHz	13.5	16.3	16.4	16.6	8.0	8.8	9.1	9.9	
			24 MHz	10.5	12.8	13.0	13.8	6.5	8.0	8.1	8.4		
		HSI clock, PLL off	8 MHz	3.7	4.7	5.0	5.3	2.1	2.3	2.4	3.0	mA	
I <sub>DD</sub>		HSI48	48 MHz	12.4	15.1	16.3	16.7	3.0	3.2	3.3	3.4	IIIA	
			48 MHz	12.3	15.0 <sup>(3)</sup>	16.0	16.2 <sup>(3)</sup>	2.9	3.2 <sup>(3)</sup>	3.3	3.4 <sup>(3)</sup>		
	pode	HSE bypass, PLL on	32 MHz	8.5	10.6	11.2	11.7	1.9	2.1	2.2	2.5		
	u də:		24 MHz	6.5	8.1	8.5	8.7	1.6	1.8	1.8	1.9		
	Sle	HSE bypass,	8 MHz	2.3	3.0	3.1	3.2	0.7	0.8	0.8	0.9		
	nt ir	PLL off	1 MHz	0.4	0.4	0.4	0.6	0.1	0.3	0.3	0.4		
			48 MHz	12.4	15.3	15.7	15.9	3.0	3.0	3.2	3.4		
		HSI clock, PLL on	32 MHz	8.6	10.7	11.3	11.6	2.1	2.2	2.2	2.5		
		PLL on _	24 MHz	6.6	8.4	8.7	8.9	1.6	1.6	1.7	1.9		
		HSI clock, PLL off	8 MHz	2.4	3.2	3.4	3.6	0.6	0.8	0.9	1.0		

<sup>1.</sup> USB is kept disabled as this IP functions only with a 48 MHz clock.

<sup>2.</sup> Data based on characterization results, not tested in production unless otherwise specified.

<sup>3.</sup> Data based on characterization results and tested in production (using one common test limit for sum of  $I_{DD}$  and  $I_{DDA}$ ).

Table 27. Typical and maximum current consumption from the  $V_{\text{DDA}}$  supply

					V <sub>DDA</sub>	= 2.4 V	1		V <sub>DDA</sub>	= 3.6 \	/	
Symbol	Para- meter	Conditions (1)	f <sub>HCLK</sub>	Tun	Max @ T <sub>A</sub> <sup>(2)</sup>		Tvn	М	A <sup>(2)</sup>	Unit		
				iyp	25 °C	85 °C	105 °C	Тур	25 °C	85 °C	105 °C	
		HSI48	48 MHz	309	325	332	342	317	334	338	344	
	Supply current in	Run or Sleep mode, code HSE bypass, PLL off	48 MHz	148	167 <sup>(3)</sup>	176	179 <sup>(3)</sup>	161	181 <sup>(3)</sup>	193	197 <sup>(3)</sup>	
current in			32 MHz	102	119	124	126	111	128	135	137	
	Run or		24 MHz	80	95	99	100	88	102	106	108	
	•		8 MHz	2.7	3.7	4.2	4.5	3.5	4.7	5.2	5.5	
I <sub>DDA</sub>	code executing		1 MHz	2.7	3.7	4.2	4.2	3.6	4.7	5.2	5.5	μΑ
	from		48 MHz	220	242	251	254	242	264	275	279	
	Flash memory	HSI clock, PLL on	32 MHz	173	193	200	202	191	211	219	221	
	or RAM		24 MHz	151	169	175	177	167	184	191	193	
	OI TO WI	HSI clock, PLL off	8 MHz	72	82	85	85	82	92	95	95	

Current consumption from the V<sub>DDA</sub> supply is independent of whether the digital peripherals are enabled or disabled, being in Run or Sleep mode or executing from Flash memory or RAM. Furthermore, when the PLL is off, I<sub>DDA</sub> is independent from the frequency.

<sup>2.</sup> Data based on characterization results, not tested in production unless otherwise specified.

<sup>3.</sup> Data based on characterization results and tested in production (using one common test limit for sum of I<sub>DD</sub> and I<sub>DDA</sub>).

Table 28. Typical and maximum consumption in Stop and Standby modes

_			Typical allu li				V <sub>DD</sub> = V				Max <sup>(1)</sup>	)			
Symbol	Parameter		Conditions	2.0 V	2.4 V	2.7 V	3.0 V	3.3 V	3.6 V	T <sub>A</sub> = 25°C	T <sub>A</sub> = 85°C	T <sub>A</sub> = 105°C	Unit		
	Supply current in	mod	julator in run de, all illators OFF	14.3	14.5	14.6	14.7	14.8	14.9	21.0	47.0	64.0			
I <sub>DD</sub>	Stop mode	pow	julator in low- ver mode, all illators OFF	2.9	3.1	3.2	3.3	3.4	3.5	6.5	32.0	44.0			
	Supply current in	LSI ON	ON and IWDG	0.8	0.9	1.1	1.2	1.3	1.5	-	-	-			
	Standby mode	LSI OFF	OFF and IWDG	0.6	0.7	0.8	0.9	1.0	1.1	2.0	2.5	3.0			
	Supply current in Stop mode	Supply Z		z	Regulator in run mode, all oscillators OFF	2.0	2.1	2.2	2.4	2.5	2.7	3.5	3.5	4.5	
		nonitoring apow d	Regulator in low-power mode, all oscillators OFF	2.0	2.1	2.2	2.4	2.5	2.7	3.5	3.5	4.5	μА		
	Supply current in	V <sub>DE</sub>	LSI ON and IWDG ON	2.4	2.6	2.8	3.0	3.1	3.4	-	-	-			
	Standby mode		LSI OFF and IWDG OFF	1.9	2.0	2.1	2.3	2.4	2.5	3.4	3.5	4.5			
I <sub>DDA</sub>	Supply	4	Regulator in run mode, all oscillators OFF	1.3	1.3	1.3	1.4	1.4	1.5	-	-	-			
	current in Stop mode	current in O Regulator in	low-power mode, all oscillators	1.3	1.3	1.3	1.4	1.4	1.5	-	-	-			
	Supply current in	V <sub>DD</sub>	LSI ON and IWDG ON	1.7	1.8	1.8	2.0	2.1	2.2	-	-	-			
	Standby mode		LSI OFF and IWDG OFF	1.1	1.2	1.2	1.3	1.3	1.4	-	-	-			

<sup>1.</sup> Data based on characterization results, not tested in production unless otherwise specified.

		r Conditions	Typ @ V <sub>BAT</sub>									
Symbol	Parameter		1.65 V	1.8 V	2.4 V	2.7 V	3.3 V	3.6 V	T <sub>A</sub> = 25 °C	T <sub>A</sub> = 85 °C	T <sub>A</sub> = 105 °C	Unit
I <sub>DD_VBAT</sub> dom	RTC domain	LSE & RTC ON; "Xtal mode": lower driving capability; LSEDRV[1:0] = '00'	0.5	0.5	0.6	0.7	0.9	1.1	1.2	1.5	2.0	
	supply current	LSE & RTC ON; "Xtal mode" higher driving capability; LSEDRV[1:0] = '11'	0.8	0.9	1.1	1.2	1.4	1.5	1.6	2.0	2.6	μА

Table 29. Typical and maximum current consumption from the V<sub>BAT</sub> supply

### **Typical current consumption**

The MCU is placed under the following conditions:

- V<sub>DD</sub> = V<sub>DDA</sub> = 3.3 V
- All I/O pins are in analog input configuration
- The Flash memory access time is adjusted to f<sub>HCLK</sub> frequency:
  - 0 wait state and Prefetch OFF from 0 to 24 MHz
  - 1 wait state and Prefetch ON above 24 MHz
- When the peripherals are enabled, f<sub>PCLK</sub> = f<sub>HCLK</sub>
- PLL is used for frequencies greater than 8 MHz
- AHB prescaler of 2, 4, 8 and 16 is used for the frequencies 4 MHz, 2 MHz, 1 MHz and 500 kHz respectively



<sup>1.</sup> Data based on characterization results, not tested in production.

Table 30. Typical current consumption, code executing from Flash memory, running from HSE 8 MHz crystal

Symbol	Parameter	£	Typical con Run i	sumption in mode		sumption in mode	Unit	
Symbol	raiailletei	f <sub>HCLK</sub>	Peripherals enabled	Peripherals disabled	Peripherals enabled	Peripherals disabled	Ollit	
		48 MHz	20.7	12.8	12.3	3.4		
		36 MHz	15.9	9.9	9.5	2.7		
		32 MHz	14.3	9.0	8.5	2.5		
	Current	24 MHz	11.0	7.1	6.6	2.1		
I==	consumption	16 MHz	7.7	5.0	4.7	1.6	mA	
'DD	from V <sub>DD</sub> supply		8 MHz	4.3	3.0	2.7	1.2	ША
		4 MHz	2.6	2.0	1.7	0.9		
		2 MHz	1.8	1.5	1.2	0.8		
		1 MHz	1.4	1.2	1.0	0.8		
		500 kHz	1.2	1.1	0.8	0.7		
		48 MHz		16	3.3			
		36 MHz		12	4.3			
		32 MHz		11	1.9			
	Current	24 MHz		87	7.1			
,	consumption	16 MHz		62	2.5		μA	
I <sub>DDA</sub>	from V <sub>DDA</sub> supply	8 MHz		2	.5		μΑ	
	зирріу	4 MHz		2	.5			
		2 MHz		2	.5			
		1 MHz		2	.5			
		500 kHz		2	.5			

### I/O system current consumption

The current consumption of the I/O system has two components: static and dynamic.

#### I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in *Table 50: I/O static characteristics*.

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt



trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

Caution:

Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

#### I/O dynamic current consumption

In addition to the internal peripheral current consumption measured previously (see *Table 32: Peripheral current consumption*), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the I/O supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DDIOx} \times f_{SW} \times C$$

where

 $I_{\mbox{\scriptsize SW}}$  is the current sunk by a switching I/O to charge/discharge the capacitive load

V<sub>DDIOx</sub> is the I/O supply voltage

f<sub>SW</sub> is the I/O switching frequency

C is the total capacitance seen by the I/O pin:  $C = C_{INT} + C_{EXT} + C_{S}$ 

C<sub>S</sub> is the PCB board capacitance including the pad pin.

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.



Table 31. Switching output I/O current consumption

Symbol	Parameter	Conditions <sup>(1)</sup>	I/O toggling frequency (f <sub>SW</sub> )	Тур	Unit
			4 MHz	0.07	
		V <sub>DDIOx</sub> = 3.3 V	8 MHz	0.15	
		$C = C_{INT}$	16 MHz	0.31	
			24 MHz	0.53	
			48 MHz	0.92	
			4 MHz	0.18	
		V <sub>DDIOx</sub> = 3.3 V	8 MHz	0.37	
		C <sub>EXT</sub> = 0 pF	16 MHz	0.76	
		$C = C_{INT} + C_{EXT} + C_{S}$	24 MHz	1.39	
			48 MHz	2.188	
			4 MHz	0.32	
		V <sub>DDIOx</sub> = 3.3 V	8 MHz	0.64	
		C <sub>EXT</sub> = 10 pF	16 MHz	1.25	
		$C = C_{INT} + C_{EXT} + C_{S}$	24 MHz	2.23	
low	I/O current		48 MHz	4.442	mA
'500	I <sub>SW</sub> I/O current consumption		4 MHz	0.49	1177
		$V_{DDIOx} = 3.3 V$ $C_{EXT} = 22 pF$	8 MHz	0.94	
		$C = C_{INT} + C_{EXT} + C_{S}$	16 MHz	2.38	
		INT EXT 0	24 MHz	3.99	
			4 MHz	0.64	
		$V_{DDIOx} = 3.3 \text{ V}$ $C_{EXT} = 33 \text{ pF}$	8 MHz	1.25	
		$C = C_{INT} + C_{EXT} + C_{S}$	16 MHz	3.24	
		INT EXT 0	24 MHz	5.02	
		V <sub>DDIOx</sub> = 3.3 V	4 MHz	0.81	
		$C_{EXT} = 47 \text{ pF}$	8 MHz	1.7	
		$C = C_{INT} + C_{EXT} + C_{S}$ $C = C_{int}$	16 MHz	3.67	
		V <sub>DDIOx</sub> = 2.4 V	4 MHz	0.66	
		$C_{EXT} = 47 \text{ pF}$	8 MHz	1.43	
		$C = C_{INT} + C_{EXT} + C_{S}$	16 MHz	2.45	
		C = C <sub>int</sub>	24 MHz	4.97	

<sup>1.</sup> C<sub>S</sub> = 7 pF (estimated value).



### On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in *Table 32*. The MCU is placed under the following conditions:

- All I/O pins are in analog mode
- All peripherals are disabled unless otherwise mentioned
- The given value is calculated by measuring the current consumption
  - with all peripherals clocked off
  - with only one peripheral clocked on
- Ambient operating temperature and supply voltage conditions summarized in Table 18:
   Voltage characteristics

Table 32. Peripheral current consumption

		morar varront voncampuon	
	Peripheral	Typical consumption at 25 °C	Unit
	BusMatrix <sup>(1)</sup>	2.2	
	CRC	1.9	
	DMA	5.1	
	Flash memory interface	15.0	
	GPIOA	8.2	
AHB	GPIOB	7.7	µA/MHz
	GPIOC	2.1	
	GPIOF	1.8	
	SRAM	1.1	
	TSC	4.9	
	All AHB peripherals	49.8	

Table 32. Peripheral current consumption (continued)

	Peripheral	Typical consumption at 25 °C	Unit
	APB-Bridge <sup>(2)</sup>	2.9	
	ADC <sup>(3)</sup>	3.9	
	CAN	12.9	
	CEC	1.5	
	CRS	1.0	
	DBG (MCU Debug Support)	0.2	
	I2C1	3.6	
	PWR	1.4	
	SPI1	8.5	
	SPI2	6.1	
APB	SYSCFG	1.8	µA/MHz
711 15	TIM1	15.1	μινινιιίΣ
	TIM2	16.8	
	TIM3	11.7	
	TIM14	5.5	
	TIM16	7.0	
	TIM17	6.9	
	USART1	17.8	
	USART2	5.6	
	USB	4.9	
	WWDG	1.4	
	All APB peripherals	136.7	

<sup>1.</sup> The BusMatrix is automatically active when at least one master is ON (CPU, DMA).

<sup>2.</sup> The APB Bridge is automatically active when at least one peripheral is ON on the Bus.

<sup>3.</sup> The power consumption of the analog part (I<sub>DDA</sub>) of peripherals such as ADC is not included. Refer to the tables of characteristics in the subsequent sections.

### 6.3.6 Wakeup time from low-power mode

The wakeup times given in *Table 33* are the latency between the event and the execution of the first user instruction. The device goes in low-power mode after the WFE (Wait For Event) instruction, in the case of a WFI (Wait For Interruption) instruction, 16 CPU cycles must be added to the following timings due to the interrupt latency in the Cortex M0 architecture.

The SYSCLK clock source setting is kept unchanged after wakeup from Sleep mode. During wakeup from Stop or Standby mode, SYSCLK takes the default setting: HSI 8 MHz.

The wakeup source from Sleep and Stop mode is an EXTI line configured in event mode. The wakeup source from Standby mode is the WKUP1 pin (PA0).

All timings are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*..

Symbol	Parameter	Conditions	Typ @Vdd = VddA						Unit
Symbol	rarameter	Conditions	= 2.0 V	= 2.4 V	= 2.7 V	= 3 V	= 3.3 V	Max	Onic
t <sub>WUSTOP</sub> Wakeup from Stomode	Wakeup from Stop	Regulator in run mode	3.2	3.1	2.9	2.9	2.8	5	
	mode	Regulator in low power mode	7.0	5.8	5.2	4.9	4.6	9	ue
t <sub>WUSTANDBY</sub>	Wakeup from Standby mode	-	60.4	55.6	53.5	52	51	-	μs
t <sub>WUSLEEP</sub>	Wakeup from Sleep mode	-		4 SY	SCLK cy	cles		-	

Table 33. Low-power mode wakeup timings

#### 6.3.7 External clock source characteristics

#### High-speed external user clock generated from an external source

In bypass mode the HSE oscillator is switched off and the input pin is a standard GPIO.

The external clock signal has to respect the I/O characteristics in *Section 6.3.14*. However, the recommended clock input waveform is shown in *Figure 15: High-speed external clock source AC timing diagram*.

	rable of ringin opeca external acor clock characteristics				
Symbol	Parameter <sup>(1)</sup>	Min	Тур	Max	Unit
f <sub>HSE_ext</sub>	User external clock source frequency	-	8	32	MHz
V <sub>HSEH</sub>	OSC_IN input pin high level voltage	0.7 V <sub>DDIOx</sub>	-	$V_{DDIOx}$	V
V <sub>HSEL</sub>	OSC_IN input pin low level voltage	V <sub>SS</sub>	-	0.3 V <sub>DDIOx</sub>	V
$\begin{array}{c} t_{w(\text{HSEH})} \\ t_{w(\text{HSEL})} \end{array}$	OSC_IN high or low time	15	1	-	ns
t <sub>r(HSE)</sub> t <sub>f(HSE)</sub> OSC_IN rise or fall time		-	-	20	113

Table 34. High-speed external user clock characteristics



1. Guaranteed by design, not tested in production.

Figure 15. High-speed external clock source AC timing diagram

#### Low-speed external user clock generated from an external source

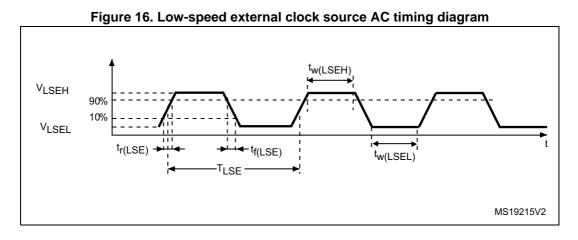
In bypass mode the LSE oscillator is switched off and the input pin is a standard GPIO.

The external clock signal has to respect the I/O characteristics in *Section 6.3.14*. However, the recommended clock input waveform is shown in *Figure 16*.

Symbol	Parameter <sup>(1)</sup>	Min	Тур	Max	Unit
f <sub>LSE_ext</sub>	User external clock source frequency	-	32.768	1000	kHz
$V_{LSEH}$	OSC32_IN input pin high level voltage	0.7 V <sub>DDIOx</sub>	-	$V_{DDIOx}$	V
$V_{LSEL}$	OSC32_IN input pin low level voltage	$V_{SS}$	-	0.3 V <sub>DDIOx</sub>	V
$\begin{matrix} t_{w(\text{LSEH})} \\ t_{w(\text{LSEL})} \end{matrix}$	OSC32_IN high or low time	450	-	-	ns
$t_{r(LSE)}$ $t_{f(LSE)}$	OSC32_IN rise or fall time	-	-	50	113

Table 35. Low-speed external user clock characteristics

<sup>1.</sup> Guaranteed by design, not tested in production.



### High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 32 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 36*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter	Conditions <sup>(1)</sup>	Min <sup>(2)</sup>	Тур	Max <sup>(2)</sup>	Unit
f <sub>OSC_IN</sub>	Oscillator frequency	-	4	8	32	MHz
R <sub>F</sub>	Feedback resistor	-	-	200	-	kΩ
		During startup <sup>(3)</sup>	-	-	8.5	
		$V_{DD}$ = 3.3 V, Rm = 30 $\Omega$ , CL = 10 pF@8 MHz	-	0.4	-	
		$V_{DD}$ = 3.3 V, Rm = 45 $\Omega$ , CL = 10 pF@8 MHz	-	0.5	-	
I <sub>DD</sub>	HSE current consumption	$V_{DD}$ = 3.3 V, Rm = 30 $\Omega$ , CL = 5 pF@32 MHz	-	0.8	-	mA
		$V_{DD}$ = 3.3 V, Rm = 30 $\Omega$ , CL = 10 pF@32 MHz	-	1	-	
		$V_{DD}$ = 3.3 V, Rm = 30 $\Omega$ , CL = 20 pF@32 MHz	-	1.5	-	
9 <sub>m</sub>	Oscillator transconductance	Startup	10	-	-	mA/V
t <sub>SU(HSE)</sub> <sup>(4)</sup>	Startup time	V <sub>DD</sub> is stabilized	-	2	-	ms

Table 36. HSE oscillator characteristics

- 1. Resonator characteristics given by the crystal/ceramic resonator manufacturer.
- 2. Guaranteed by design, not tested in production.
- 3. This consumption level occurs during the first 2/3 of the  $t_{\text{SU(HSE)}}$  startup time
- 4. t<sub>SU(HSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

For  $C_{L1}$  and  $C_{L2}$ , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 20 pF range (Typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 17*).  $C_{L1}$  and  $C_{L2}$  are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of  $C_{L1}$  and  $C_{L2}$ . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing  $C_{L1}$  and  $C_{L2}$ .

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.



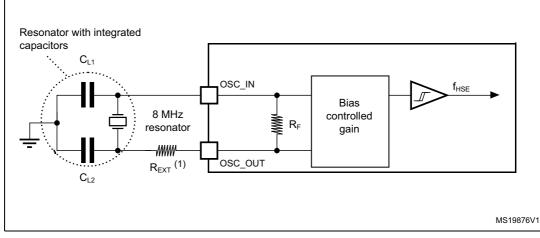


Figure 17. Typical application with an 8 MHz crystal

1.  $R_{\text{EXT}}$  value depends on the crystal characteristics.

### Low-speed external clock generated from a crystal resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in *Table 37*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Symbol	Parameter Conditions <sup>(1)</sup>		Min <sup>(2)</sup>	Тур	Max <sup>(2)</sup>	Unit
		low drive capability	-	0.5	0.9	
	LSE current consumption	medium-low drive capability	-	-	1	
I <sub>DD</sub>	LSE current consumption	medium-high drive capability	-	-	1.3	μA
		high drive capability	-	-	1.6	
		low drive capability	5	-	-	
~	Oscillator	medium-low drive capability	8	-	-	
9 <sub>m</sub>	transconductance	medium-high drive capability	15	-	-	μA/V
		high drive capability	25	-	-	
t <sub>SU(LSE)</sub> <sup>(3)</sup>	Startup time	V <sub>DDIOx</sub> is stabilized	-	2	-	S

Table 37. LSE oscillator characteristics ( $f_{LSE} = 32.768 \text{ kHz}$ )

t<sub>SU(LSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal and it can vary significantly with the crystal manufacturer



Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".

<sup>2.</sup> Guaranteed by design, not tested in production.

Note:

For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

Resonator with integrated capacitors

OSC32\_IN

Drive programmable amplifier

OSC32\_OUT

OSC32\_OUT

MS30253V2

Figure 18. Typical application with a 32.768 kHz crystal

Note:

An external resistor is not required between OSC32\_IN and OSC32\_OUT and it is forbidden to add one.

#### 6.3.8 Internal clock source characteristics

The parameters given in *Table 38* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*. The provided curves are characterization results, not tested in production.



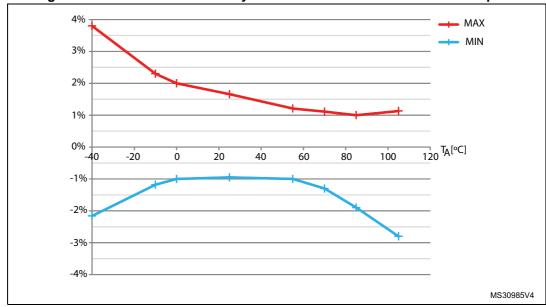
# High-speed internal (HSI) RC oscillator

Table 38. HSI oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSI</sub>	Frequency	-	-	8	-	MHz
TRIM	HSI user trimming step	-	-	-	1 <sup>(2)</sup>	%
DuCy <sub>(HSI)</sub>	Duty cycle	-	45 <sup>(2)</sup>	-	55 <sup>(2)</sup>	%
		T <sub>A</sub> = -40 to 105°C	-2.8 <sup>(3)</sup>	-	3.8 <sup>(3)</sup>	
	Accuracy of the HSI oscillator	T <sub>A</sub> = -10 to 85°C	-1.9 <sup>(3)</sup>	-	2.3 <sup>(3)</sup>	
ACC		T <sub>A</sub> = 0 to 85°C	-1.9 <sup>(3)</sup>	-	2 <sup>(3)</sup>	%
ACC <sub>HSI</sub>		T <sub>A</sub> = 0 to 70°C	-1.3 <sup>(3)</sup>	-	2 <sup>(3)</sup>	70
		T <sub>A</sub> = 0 to 55°C	-1 <sup>(3)</sup>	-	2 <sup>(3)</sup>	
		$T_A = 25^{\circ}C^{(4)}$	-1	-	1	
t <sub>su(HSI)</sub>	HSI oscillator startup time	-	1 <sup>(2)</sup>	-	2 <sup>(2)</sup>	μs
I <sub>DDA(HSI)</sub>	HSI oscillator power consumption	-	-	80	100 <sup>(2)</sup>	μΑ

- 1.  $V_{DDA} = 3.3 \text{ V}$ ,  $T_A = -40 \text{ to } 105^{\circ}\text{C}$  unless otherwise specified.
- 2. Guaranteed by design, not tested in production.
- 3. Data based on characterization results, not tested in production.
- 4. Factory calibrated, parts not soldered.

Figure 19. HSI oscillator accuracy characterization results for soldered parts



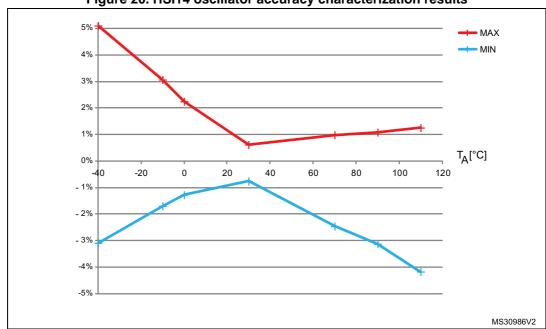
# High-speed internal 14 MHz (HSI14) RC oscillator (dedicated to ADC)

Table 39. HSI14 oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSI14</sub>	Frequency	-	-	14	-	MHz
TRIM	HSI14 user-trimming step	-	-	-	1 <sup>(2)</sup>	%
DuCy <sub>(HSI14)</sub>	Duty cycle	-	45 <sup>(2)</sup>	-	55 <sup>(2)</sup>	%
	Accuracy of the HSI14 oscillator (factory calibrated)	$T_A = -40 \text{ to } 105 ^{\circ}\text{C}$	-4.2 <sup>(3)</sup>	-	5.1 <sup>(3)</sup>	%
ACC		T <sub>A</sub> = -10 to 85 °C	$-3.2^{(3)}$	-	3.1 <sup>(3)</sup>	%
ACC <sub>HSI14</sub>		T <sub>A</sub> = 0 to 70 °C	-2.5 <sup>(3)</sup>	-	2.3 <sup>(3)</sup>	%
		T <sub>A</sub> = 25 °C	-1	-	1	%
t <sub>su(HSI14)</sub>	HSI14 oscillator startup time	-	1 <sup>(2)</sup>	-	2 <sup>(2)</sup>	μs
I <sub>DDA(HSI14)</sub>	HSI14 oscillator power consumption	-	-	100	150 <sup>(2)</sup>	μΑ

- 1.  $V_{DDA}$  = 3.3 V,  $T_{A}$  = -40 to 105 °C unless otherwise specified.
- 2. Guaranteed by design, not tested in production.
- 3. Data based on characterization results, not tested in production.

Figure 20. HSI14 oscillator accuracy characterization results



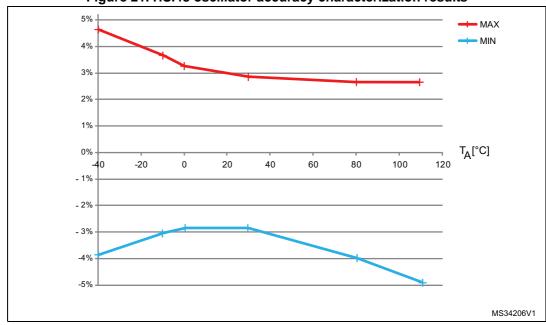
# High-speed internal 48 MHz (HSI48) RC oscillator

Table 40. HSI48 oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSI48</sub>	Frequency	-	-	48	-	MHz
TRIM	HSI48 user-trimming step	-	0.09 <sup>(2)</sup>	0.14	0.2 <sup>(2)</sup>	%
DuCy <sub>(HSI48)</sub>	Duty cycle	-	45 <sup>(2)</sup>	-	55 <sup>(2)</sup>	%
	Accuracy of the HSI48	$T_A = -40 \text{ to } 105 ^{\circ}\text{C}$	-4.9 <sup>(3)</sup>	-	4.7 <sup>(3)</sup>	%
ACC		$T_A = -10 \text{ to } 85 ^{\circ}\text{C}$	-4.1 <sup>(3)</sup>	-	3.7 <sup>(3)</sup>	%
ACC <sub>HSI48</sub>	oscillator (factory calibrated)	T <sub>A</sub> = 0 to 70 °C	-3.8 <sup>(3)</sup>	-	3.4 <sup>(3)</sup>	%
		T <sub>A</sub> = 25 °C	-2.8	-	2.9	%
t <sub>su(HSI48)</sub>	HSI48 oscillator startup time	-	-	-	6 <sup>(2)</sup>	μs
I <sub>DDA(HSI48)</sub>	HSI48 oscillator power consumption	-	-	312	350 <sup>(2)</sup>	μА

- 1.  $V_{DDA}$  = 3.3 V,  $T_{A}$  = -40 to 105 °C unless otherwise specified.
- 2. Guaranteed by design, not tested in production.
- 3. Data based on characterization results, not tested in production.

Figure 21. HSI48 oscillator accuracy characterization results



### Low-speed internal (LSI) RC oscillator

Table 41. LSI oscillator characteristics<sup>(1)</sup>

Symbol	Parameter	Min	Тур	Max	Unit
f <sub>LSI</sub>	Frequency	30	40	50	kHz
t <sub>su(LSI)</sub> <sup>(2)</sup>	LSI oscillator startup time	-	-	85	μs
I <sub>DDA(LSI)</sub> <sup>(2)</sup>	LSI oscillator power consumption	-	0.75	1.2	μΑ

<sup>1.</sup>  $V_{DDA}$  = 3.3 V,  $T_A$  = -40 to 105 °C unless otherwise specified.

#### 6.3.9 PLL characteristics

The parameters given in *Table 42* are derived from tests performed under ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*.

Table 42. PLL characteristics

Symbol	Parameter		Unit		
Symbol	Farameter	Min	Тур	Max	Offic
f	PLL input clock <sup>(1)</sup>	1 <sup>(2)</sup>	8.0	24 <sup>(2)</sup>	MHz
f <sub>PLL_IN</sub>	PLL input clock duty cycle	40 <sup>(2)</sup>	-	60 <sup>(2)</sup>	%
f <sub>PLL_OUT</sub>	PLL multiplier output clock	16 <sup>(2)</sup>	-	48	MHz
t <sub>LOCK</sub>	PLL lock time	-	-	200 <sup>(2)</sup>	μs
Jitter <sub>PLL</sub>	Cycle-to-cycle jitter	-	-	300 <sup>(2)</sup>	ps

Take care to use the appropriate multiplier factors to obtain PLL input clock values compatible with the range defined by f<sub>PLL OUT</sub>.

# 6.3.10 Memory characteristics

## Flash memory

The characteristics are given at  $T_A$  = -40 to 105 °C unless otherwise specified.

Table 43. Flash memory characteristics

Symbol	Parameter	Conditions	Min	Тур	Max <sup>(1)</sup>	Unit
t <sub>prog</sub>	16-bit programming time	T <sub>A</sub> = - 40 to +105 °C	40	53.5	60	μs
t <sub>ERASE</sub>	Page (1 KB) erase time	T <sub>A</sub> = - 40 to +105 °C	20	-	40	ms
t <sub>ME</sub>	Mass erase time	T <sub>A</sub> = - 40 to +105 °C	20	-	40	ms
	Supply ourrant	Write mode	-	-	10	mA
I <sub>DD</sub> Suppi	Supply current	Erase mode	-	-	12	mA

<sup>1.</sup> Guaranteed by design, not tested in production.



<sup>2.</sup> Guaranteed by design, not tested in production.

<sup>2.</sup> Guaranteed by design, not tested in production.

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Unit
N <sub>END</sub>	Endurance	$T_A = -40 \text{ to } +105 ^{\circ}\text{C}$	10	kcycle
		1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 85 °C	30	
t <sub>RET</sub>	Data retention	1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 105 °C	10	Year
		10 kcycle <sup>(2)</sup> at T <sub>A</sub> = 55 °C	20	

Table 44. Flash memory endurance and data retention

#### 6.3.11 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

#### Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports). the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- **Electrostatic discharge (ESD)** (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to V<sub>DD</sub> and V<sub>SS</sub> through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in *Table 45*. They are based on the EMS levels and classes defined in application note AN1709.

Symbol	Parameter	Conditions	Level/ Class
V <sub>FESD</sub>	Voltage limits to be applied on any I/O pin to induce a functional disturbance	$V_{DD}$ = 3.3 V, LQFP48, $T_{A}$ = +25 °C, $f_{HCLK}$ = 48 MHz, conforming to IEC 61000-4-2	3B
V <sub>EFTB</sub>	Fast transient voltage burst limits to be applied through 100 pF on V <sub>DD</sub> and V <sub>SS</sub> pins to induce a functional disturbance	$V_{DD}$ = 3.3 V, LQFP48, $T_A$ = +25°C, $f_{HCLK}$ = 48 MHz, conforming to IEC 61000-4-4	4B

Table 45. EMS characteristics

# Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.



<sup>1.</sup> Data based on characterization results, not tested in production.

<sup>2.</sup> Cycling performed over the whole temperature range.

#### Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (for example control registers)

#### **Prequalification trials**

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

#### **Electromagnetic Interference (EMI)**

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Max vs. [fHSE/fHCLK] **Monitored** Conditions Unit Symbol Parameter frequency band 8/48 MHz 0.1 to 30 MHz -9  $V_{DD} = 3.6 \text{ V}, T_A = 25 ^{\circ}\text{C},$ 30 to 130 MHz 9 dBuV LQFP48 package Peak level  $S_{EMI}$ compliant with 130 MHz to 1 GHz 17 IEC 61967-2 **EMI Level** 3

Table 46. EMI characteristics

### 6.3.12 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

#### Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the JESD22-A114/C101 standard.



Symbol	Ratings	Conditions	Packages	Class	Maximum value <sup>(1)</sup>	Unit	
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (human body model)	T <sub>A</sub> = +25 °C, conforming to JESD22-A114	All	2	2000	V	
V <sub>ESD(CDM)</sub>	Electrostatic discharge voltage (charge device model)	T <sub>A</sub> = +25 °C, conforming to ANSI/ESD STM5.3.1	All	C4	500	V	

Table 47. ESD absolute maximum ratings

#### Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin.
- A current injection is applied to each input, output and configurable I/O pin.

These tests are compliant with EIA/JESD 78A IC latch-up standard.

Table 48. Electrical sensitivities

Symbol Parameter		Conditions	Class	
LU	Static latch-up class	T <sub>A</sub> = +105 °C conforming to JESD78A	II level A	

# 6.3.13 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below  $V_{SS}$  or above  $V_{DDIOX}$  (for standard, 3.3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

#### Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of the -5  $\mu$ A/+0  $\mu$ A range) or other functional failure (for example reset occurrence or oscillator frequency deviation).

The characterization results are given in Table 49.

Negative induced leakage current is caused by negative injection and positive induced leakage current is caused by positive injection.



<sup>1.</sup> Data based on characterization results, not tested in production.

Symbol	Description	Functional susceptibility		Unit	
	Description	Negative injection	Positive injection	Onit	
I <sub>INJ</sub>	Injected current on PA12 pin	-0	+5		
	Injected current on PA9, PB3, PB13, PF11 pins with induced leakage current on adjacent pins less than 50 µA	-5	NA	mA	
	Injected current on PB0, PB1 and all other FT and FTf pins	-5	NA		
	Injected current on all other TC, TTa and RST pins	-5	+5		

Table 49. I/O current injection susceptibility

### 6.3.14 I/O port characteristics

### General input/output characteristics

Unless otherwise specified, the parameters given in *Table 50* are derived from tests performed under the conditions summarized in *Table 21: General operating conditions*. All I/Os are designed as CMOS- and TTL-compliant.

Symbol **Parameter Conditions** Min Max Unit Тур 0.3 V<sub>DDIOx</sub>+0.07<sup>(1)</sup> TC and TTa I/O Low level input FT and FTf I/O 0.475 V<sub>DDIOx</sub>-0.2<sup>(1)</sup> ٧  $V_{IL}$ voltage All I/Os 0.3 V<sub>DDIOx</sub> TC and TTa I/O 0.445 V<sub>DDIOx</sub>+0.398<sup>(1)</sup> High level input  $V_{\text{IH}}$ FT and FTf I/O  $0.5 V_{\rm DDIOx} + 0.2^{(1)}$ ٧ voltage All I/Os 0.7 V<sub>DDIOx</sub> 200(1) TC and TTa I/O Schmitt trigger  $V_{hys}$ mV hysteresis FT and FTf I/O  $100^{(1)}$ TC, FT and FTf I/O TTa in digital mode ± 0.1  $V_{SS} \le V_{IN} \le V_{DDIOx}$ TTa in digital mode 1 Input leakage  $V_{\text{DDIO}x} \le V_{\text{IN}} \le V_{\text{DDA}}$ μΑ  $I_{lkq}$ current<sup>(2)</sup> TTa in analog mode ± 0.2  $V_{SS} \le V_{IN} \le V_{DDA}$ FT and FTf I/O 10  $V_{DDIOx} \le V_{IN} \le 5 V$ Weak pull-up equivalent resistor  $V_{IN} = V_{SS}$ 25 40 55 kΩ  $R_{PU}$ 

Table 50. I/O static characteristics



Symbol	Parameter	Conditions	Min	Тур	Мах	Unit
R <sub>PD</sub>	Weak pull-down equivalent resistor <sup>(3)</sup>	V <sub>IN</sub> = - V <sub>DDIOx</sub>	25	40	55	kΩ
C <sub>IO</sub>	I/O pin capacitance	-	-	5	-	pF

- 1. Data based on design simulation only. Not tested in production.
- The leakage could be higher than the maximum value, if negative current is injected on adjacent pins. Refer to Table 49: I/O current injection susceptibility.
- 3. Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimal (~10% order).

All I/Os are CMOS- and TTL-compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements is shown in *Figure 22* for standard I/Os, and in *Figure 23* for 5 V-tolerant I/Os. The following curves are design simulation results, not tested in production.



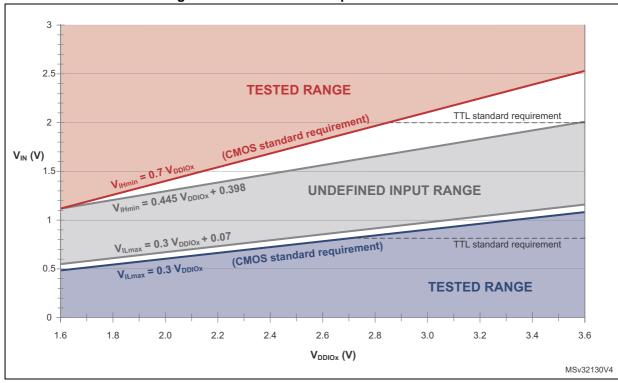
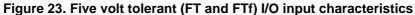
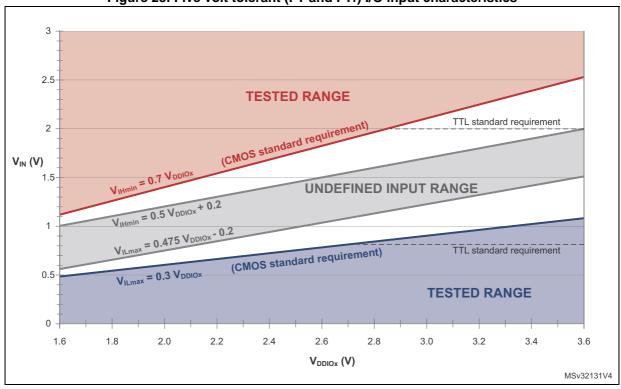


Figure 22. TC and TTa I/O input characteristics





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### **Output driving current**

The GPIOs (general purpose input/outputs) can sink or source up to +/-8 mA, and sink or source up to +/- 20 mA (with a relaxed  $V_{OL}/V_{OH}$ ).

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in Section 6.2:

- The sum of the currents sourced by all the I/Os on V<sub>DDIOX</sub>, plus the maximum consumption of the MCU sourced on V<sub>DD</sub>, cannot exceed the absolute maximum rating ΣI<sub>VDD</sub> (see *Table 18: Voltage characteristics*).
- The sum of the currents sunk by all the I/Os on V<sub>SS</sub>, plus the maximum consumption of the MCU sunk on V<sub>SS</sub>, cannot exceed the absolute maximum rating ΣI<sub>VSS</sub> (see Table 18: Voltage characteristics).

### **Output voltage levels**

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*. All I/Os are CMOS- and TTL-compliant (FT, TTa or TC unless otherwise specified).

Table 51. Output voltage characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit	
V <sub>OL</sub>	Output low level voltage for an I/O pin	CMOS port <sup>(2)</sup>	-	0.4	.,	
V <sub>OH</sub>	Output high level voltage for an I/O pin	$ I_{IO}  = 8 \text{ mA}$ $V_{DDIOx} \ge 2.7 \text{ V}$	V <sub>DDIOx</sub> -0.4	-	V	
V <sub>OL</sub>	Output low level voltage for an I/O pin	TTL port <sup>(2)</sup>	-	0.4	.,	
V <sub>OH</sub>	Output high level voltage for an I/O pin	I <sub>IO</sub>   = 8 mA V <sub>DDIOx</sub> ≥ 2.7 V	2.4	-	V	
V <sub>OL</sub> <sup>(3)</sup>	Output low level voltage for an I/O pin	I <sub>IO</sub>   = 20 mA	-	1.3	V	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	V <sub>DDIOx</sub> ≥ 2.7 V	V <sub>DDIOx</sub> -1.3	-	'	
V <sub>OL</sub> <sup>(3)</sup>	Output low level voltage for an I/O pin	I <sub>IO</sub>   = 6 mA	-	0.4	V	
V <sub>OH</sub> <sup>(3)</sup>	Output high level voltage for an I/O pin	V <sub>DDIOx</sub> ≥ 2 V	V <sub>DDIOx</sub> -0.4	-	V	
V <sub>OL</sub> <sup>(4)</sup>	Output low level voltage for an I/O pin	II I = 4 m A	-	0.4	V	
V <sub>OH</sub> <sup>(4)</sup>	Output high level voltage for an I/O pin	I <sub>IO</sub>   = 4 mA	V <sub>DDIOx</sub> -0.4	-	V	
V <sub>OLFm+</sub> <sup>(3)</sup>	Output low level voltage for an FTf I/O pin in Fm+ mode	I <sub>IO</sub>   = 20 mA V <sub>DDIOx</sub> ≥ 2.7 V	-	0.4	٧	
	Thir mode	I <sub>IO</sub>   = 10 mA	-	0.4	V	

The I<sub>IO</sub> current sourced or sunk by the device must always respect the absolute maximum rating specified in *Table 18:* Voltage characteristics, and the sum of the currents sourced or sunk by all the I/Os (I/O ports and control pins) must always respect the absolute maximum ratings ΣI<sub>IO</sub>.

- 2. TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.
- 3. Data based on characterization results. Not tested in production.
- 4. Data based on characterization results. Not tested in production.

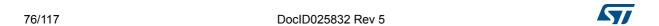


### Input/output AC characteristics

The definition and values of input/output AC characteristics are given in *Figure 24* and *Table 52*, respectively. Unless otherwise specified, the parameters given are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*.

Table 52. I/O AC characteristics<sup>(1)(2)</sup>

OSPEEDRy [1:0] value <sup>(1)</sup>	Symbol	Parameter	Conditions	Min	Max	Unit
	f <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>		-	2	MHz
	t <sub>f(IO)out</sub>	Output fall time	C <sub>L</sub> = 50 pF, V <sub>DDIOx</sub> ≥ 2 V	-	125	20
x0	t <sub>r(IO)out</sub>	Output rise time		-	125	ns
XU	f <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>		-	1	MHz
	t <sub>f(IO)out</sub>	Output fall time	C <sub>L</sub> = 50 pF, V <sub>DDIOx</sub> < 2 V	-	125	ns
	t <sub>r(IO)out</sub>	Output rise time	1	-	125	115
	f <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>		-	10	MHz
	t <sub>f(IO)out</sub>	Output fall time	C <sub>L</sub> = 50 pF, V <sub>DDIOx</sub> ≥ 2 V	-	25	7.5
01	t <sub>r(IO)out</sub>	Output rise time			25	ns
01	f <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>		-	4	MHz
	t <sub>f(IO)out</sub>	Output fall time	C <sub>L</sub> = 50 pF, V <sub>DDIOx</sub> < 2 V		62.5	ns
	t <sub>r(IO)out</sub>	Output rise time		-	62.5	115
			$C_L = 30 \text{ pF}, V_{DDIOx} \ge 2.7 \text{ V}$	- 50		
	f <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>	$C_L = 50 \text{ pF}, V_{DDIOx} \ge 2.7 \text{ V}$	-	30	MHz
		ut Maximum frequency.	$C_L = 50 \text{ pF}, 2 \text{ V} \le \text{V}_{\text{DDIOx}} < 2.7 \text{ V}$	-	20	IVIITZ
			C <sub>L</sub> = 50 pF, V <sub>DDIOx</sub> < 2 V	-	10	
			$C_L = 30 \text{ pF}, V_{DDIOx} \ge 2.7 \text{ V}$	-	5	
11		Output fall time	$C_L = 50 \text{ pF}, V_{DDIOx} \ge 2.7 \text{ V}$	-	8	1
11	t <sub>f(IO)</sub> out	Output fail time	$C_L = 50 \text{ pF}, 2 \text{ V} \le \text{V}_{\text{DDIOx}} < 2.7 \text{ V}$	-	12	
			C <sub>L</sub> = 50 pF, V <sub>DDIOx</sub> < 2 V	-	25	ne
			$C_L = 30 \text{ pF}, V_{DDIOx} \ge 2.7 \text{ V}$	-	5	ns
	t	Output rise time	C <sub>L</sub> = 50 pF, V <sub>DDIOx</sub> ≥ 2.7 V	-	8	
	t <sub>r(IO)out</sub>	Output rise time	$C_L = 50 \text{ pF}, 2 \text{ V} \le \text{V}_{DDIOx} < 2.7 \text{ V}$	-	12	
			C <sub>L</sub> = 50 pF, V <sub>DDIOx</sub> < 2 V	-	25	



OSPEEDRy [1:0] value <sup>(1)</sup>	Symbol	Parameter	Conditions	Min	Max	Unit
	f <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>		-	2	MHz
	$t_{f(IO)out}$	Output fall time	$C_L = 50 \text{ pF}, V_{DDIOx} \ge 2 \text{ V}$	-	12	no
Fm+ configuration	t <sub>r(IO)out</sub>	Output rise time	]		34	ns
(4)	f <sub>max(IO)out</sub>	Maximum frequency <sup>(3)</sup>			0.5	MHz
	t <sub>f(IO)out</sub>	Output fall time	C <sub>L</sub> = 50 pF, V <sub>DDIOx</sub> < 2 V	-	16	ne
	t <sub>r(IO)out</sub>	Output rise time			44	ns
-	t <sub>EXTIpw</sub>	Pulse width of external signals detected by the EXTI controller	-	10	1	ns

Table 52. I/O AC characteristics<sup>(1)(2)</sup> (continued)

- The I/O speed is configured using the OSPEEDRx[1:0] bits. Refer to the STM32F0xxxx RM0091 reference manual for a description of GPIO Port configuration register.
- 2. Guaranteed by design, not tested in production.
- 3. The maximum frequency is defined in Figure 24.
- When Fm+ configuration is set, the I/O speed control is bypassed. Refer to the STM32F0xxxx reference manual RM0091 for a detailed description of Fm+ I/O configuration.

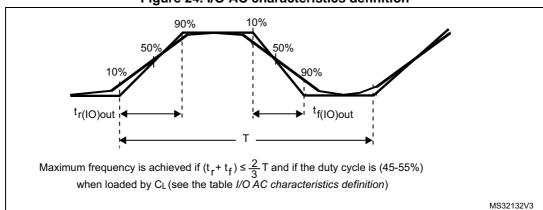


Figure 24. I/O AC characteristics definition

### 6.3.15 NRST pin characteristics

The NRST pin input driver uses the CMOS technology. It is connected to a permanent pull-up resistor,  $R_{\text{PU}}$ .

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in *Table 21: General operating conditions*.

SymbolParameterConditionsMinTypMaxUnit $V_{\text{IL(NRST)}}$ NRST input low level voltage--- $0.3 \text{ V}_{\text{DD}} + 0.07^{(1)}$  $V_{\text{IH(NRST)}}$ NRST input high level voltage- $0.445 \text{ V}_{\text{DD}} + 0.398^{(1)}$ --

Table 53. NRST pin characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>hys(NRST)</sub>	NRST Schmitt trigger voltage hysteresis	-	-	200	-	mV
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(2)</sup>	V <sub>IN</sub> = V <sub>SS</sub>	25	40	55	kΩ
V <sub>F(NRST)</sub>	NRST input filtered pulse	-	-	-	100 <sup>(1)</sup>	ns
V	NRST input not filtered pulse	2.7 < V <sub>DD</sub> < 3.6	300 <sup>(3)</sup>	-	-	ns
V <sub>NF(NRST)</sub>		2.0 < V <sub>DD</sub> < 3.6	500 <sup>(3)</sup>	-	-	115

Table 53. NRST pin characteristics (continued)

- 1. Data based on design simulation only. Not tested in production.
- 2. The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimal (~10% order).
- 3. Data based on design simulation only. Not tested in production.

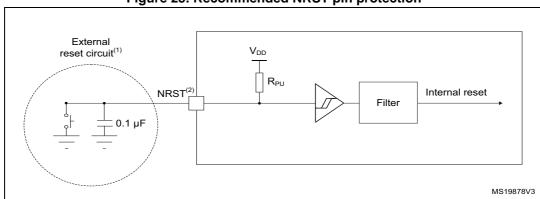


Figure 25. Recommended NRST pin protection

- 1. The external capacitor protects the device against parasitic resets.
- 2. The user must ensure that the level on the NRST pin can go below the  $V_{\rm IL(NRST)}$  max level specified in *Table 53: NRST pin characteristics*. Otherwise the reset will not be taken into account by the device.

### 6.3.16 12-bit ADC characteristics

Unless otherwise specified, the parameters given in *Table 54* are derived from tests performed under the conditions summarized in *Table 21: General operating conditions*.

Note: It is recommended to perform a calibration after each power-up.

**Symbol Conditions** Unit **Parameter** Min Тур Max Analog supply voltage for 2.4 3.6 ٧  $V_{DDA}$ ADC ON Current consumption of  $V_{DDA} = 3.3 V$ 0.9 mA I<sub>DDA</sub> (ADC) the ADC<sup>(1)</sup> ADC clock frequency 0.6 14 MHz  $f_{ADC}$  $f_S^{(2)}$ Sampling rate 12-bit resolution 0.043 1 MHz

Table 54. ADC characteristics



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Table 54. ADC characteristics (continued)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>TRIG</sub> <sup>(2)</sup>	External trigger frequency	f <sub>ADC</sub> = 14 MHz, 12-bit resolution	-	-	823	kHz
1140		12-bit resolution	-	-	17	1/f <sub>ADC</sub>
V <sub>AIN</sub>	Conversion voltage range	-	0	-	$V_{DDA}$	V
R <sub>AIN</sub> <sup>(2)</sup>	External input impedance	See Equation 1 and Table 55 for details	-	-	50	kΩ
R <sub>ADC</sub> <sup>(2)</sup>	Sampling switch resistance	-	-	-	1	kΩ
C <sub>ADC</sub> <sup>(2)</sup>	Internal sample and hold capacitor	-	-	-	8	pF
t <sub>CAL</sub> <sup>(2)(3)</sup>	Calibration time	f <sub>ADC</sub> = 14 MHz		5.9		μs
<sup>L</sup> CAL` /\*'	Calibration time	-		83		1/f <sub>ADC</sub>
	ADC_DR register ready latency	ADC clock = HSI14	1.5 ADC cycles + 2 f <sub>PCLK</sub> cycles	-	1.5 ADC cycles + 3 f <sub>PCLK</sub> cycles	-
W <sub>LATENCY</sub> (2)(4)		ADC clock = PCLK/2	-	4.5	-	f <sub>PCLK</sub> cycle
		ADC clock = PCLK/4	-	8.5	-	f <sub>PCLK</sub> cycle
		$f_{ADC} = f_{PCLK}/2 = 14 \text{ MHz}$	0.196		μs	
		f <sub>ADC</sub> = f <sub>PCLK</sub> /2	5.5			1/f <sub>PCLK</sub>
t <sub>latr</sub> <sup>(2)</sup>	Trigger conversion latency	$f_{ADC} = f_{PCLK}/4 = 12 \text{ MHz}$	0.219		μs	
		$f_{ADC} = f_{PCLK}/4$	10.5		1/f <sub>PCLK</sub>	
		f <sub>ADC</sub> = f <sub>HSI14</sub> = 14 MHz	0.179	-	0.250	μs
Jitter <sub>ADC</sub>	ADC jitter on trigger conversion	f <sub>ADC</sub> = f <sub>HSI14</sub>	-	1	-	1/f <sub>HSI14</sub>
t <sub>S</sub> <sup>(2)</sup>	Sampling time	f <sub>ADC</sub> = 14 MHz	0.107	-	17.1	μs
is	Camping time	-	1.5	-	239.5	1/f <sub>ADC</sub>
t <sub>STAB</sub> <sup>(2)</sup>	Stabilization time	-	14		1/f <sub>ADC</sub>	
t <sub>CONV</sub> <sup>(2)</sup>	Total conversion time	f <sub>ADC</sub> = 14 MHz, 12-bit resolution	1	-	18	μs
'CONV'	(including sampling time)	12-bit resolution	14 to 252 (t <sub>S</sub> for sampling +12.5 for successive approximation)		1/f <sub>ADC</sub>	

<sup>1.</sup> During conversion of the sampled value (12.5 x ADC clock period), an additional consumption of 100  $\mu$ A on I<sub>DD</sub> and 60  $\mu$ A on I<sub>DD</sub> should be taken into account.

<sup>2.</sup> Guaranteed by design, not tested in production.

<sup>3.</sup> Specified value includes only ADC timing. It does not include the latency of the register access.

<sup>4.</sup> This parameter specify latency for transfer of the conversion result to the ADC\_DR register. EOC flag is set at this time.

### Equation 1: R<sub>AIN</sub> max formula

$$R_{AIN} < \frac{T_S}{f_{ADC} \times C_{ADC} \times \ln(2^{N+2})} - R_{ADC}$$

The formula above (Equation 1) is used to determine the maximum external impedance allowed for an error below 1/4 of LSB. Here N = 12 (from 12-bit resolution).

Table 55.  $R_{AIN}$  max for  $f_{ADC} = 14$  MHz

T <sub>s</sub> (cycles)	t <sub>S</sub> (μs)	R <sub>AIN</sub> max (kΩ) <sup>(1)</sup>
1.5	0.11	0.4
7.5	0.54	5.9
13.5	0.96	11.4
28.5	2.04	25.2
41.5	2.96	37.2
55.5	3.96	50
71.5	5.11	NA
239.5	17.1	NA

<sup>1.</sup> Guaranteed by design, not tested in production.

Table 56. ADC accuracy<sup>(1)(2)(3)</sup>

Symbol	Parameter	Test conditions	Тур	Max <sup>(4)</sup>	Unit
ET	Total unadjusted error		±1.3	±2	
EO	Offset error	f <sub>PCLK</sub> = 48 MHz,	±1	±1.5	
EG	Gain error	$f_{ADC}$ = 14 MHz, $R_{AIN}$ < 10 kΩ $V_{DDA}$ = 3 V to 3.6 V	±0.5	±1.5	LSB
ED	Differential linearity error	T <sub>A</sub> = 25 °C	±0.7	±1	
EL	Integral linearity error		±0.8	±1.5	
ET	Total unadjusted error		±3.3	±4	
EO	Offset error	f <sub>PCLK</sub> = 48 MHz,	±1.9	±2.8	
EG	Gain error	$f_{ADC}$ = 14 MHz, $R_{AIN}$ < 10 kΩ $V_{DDA}$ = 2.7 V to 3.6 V	±2.8	±3	LSB
ED	Differential linearity error	T <sub>A</sub> = - 40 to 105 °C	±0.7	±1.3	
EL	Integral linearity error		±1.2	±1.7	
ET	Total unadjusted error		±3.3	±4	
EO	Offset error	f <sub>PCLK</sub> = 48 MHz,	±1.9	±2.8	
EG	Gain error	$f_{ADC}$ = 14 MHz, $R_{AIN}$ < 10 kΩ $V_{DDA}$ = 2.4 V to 3.6 V	±2.8	±3	LSB
ED	Differential linearity error	T <sub>A</sub> = 25 °C	±0.7	±1.3	
EL	Integral linearity error		±1.2	±1.7	

<sup>1.</sup> ADC DC accuracy values are measured after internal calibration.

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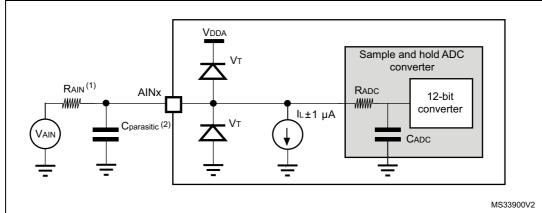
- ADC Accuracy vs. Negative Injection Current: Injecting negative current on any of the standard (non-robust) analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to standard analog pins which may potentially inject negative current
  - Any positive injection current within the limits specified for I<sub>INJ(PIN)</sub> and ΣI<sub>INJ(PIN)</sub> in Section 6.3.14 does not affect the ADC accuracy.
- Better performance may be achieved in restricted V<sub>DDA</sub>, frequency and temperature ranges.
- Data based on characterization results, not tested in production.

5 6 7

Eg (1) Example of an actual transfer curve 4095 (2) The ideal transfer curve (3) End point correlation line 4094 4093 ET = Total Unajusted Error: maximum deviation between the actual and ideal transfer curves. Eo = Offset Error: maximum deviation between the first actual transition and the first ideal one. 6 Eg = Gain Error: deviation between the last 5 ideal transition and the last actual one. Fi 4 ED = Differential Linearity Error: maximum deviation between actual steps and the ideal ones. 3 EL = Integral Linearity Error: maximum deviation 2 between any actual transition and the end point 1 LSB IDEAL correlation line. VDDA 0 4093 4094 4095 4096

Figure 26. ADC accuracy characteristics





- Refer to Table 54: ADC characteristics for the values of RAIN, RADC and CADC
- $C_{parasitic}$  represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7 pF). A high  $C_{parasitic}$  value will downgrade conversion accuracy. To remedy this,  $f_{ADC}$  should be reduced.

#### General PCB design guidelines

Power supply decoupling should be performed as shown in Figure 13: Power supply scheme. The 10 nF capacitor should be ceramic (good quality) and it should be placed as close as possible to the chip.



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### 6.3.17 Temperature sensor characteristics

Table 57. TS characteristics

Symbol	Parameter	Min	Тур	Max	Unit
T <sub>L</sub> <sup>(1)</sup>	V <sub>SENSE</sub> linearity with temperature	-	± 1	± 2	°C
Avg_Slope <sup>(1)</sup>	Average slope	4.0	4.3	4.6	mV/°C
V <sub>30</sub>	Voltage at 30 °C (± 5 °C) <sup>(2)</sup>	1.34	1.43	1.52	V
t <sub>START</sub> <sup>(1)</sup>	ADC_IN16 buffer startup time	-	-	10	μs
t <sub>S_temp</sub> <sup>(1)</sup>	ADC sampling time when reading the temperature	4	-	-	μs

<sup>1.</sup> Guaranteed by design, not tested in production.

### 6.3.18 V<sub>BAT</sub> monitoring characteristics

Table 58. V<sub>BAT</sub> monitoring characteristics

Symbol	Parameter	Min	Тур	Max	Unit
R	Resistor bridge for V <sub>BAT</sub>	-	2 x 50	-	kΩ
Q	Ratio on V <sub>BAT</sub> measurement	-	2	-	-
Er <sup>(1)</sup>	Error on Q	-1	-	+1	%
t <sub>S_vbat</sub> <sup>(1)</sup>	ADC sampling time when reading the V <sub>BAT</sub>	4	-	-	μs

<sup>1.</sup> Guaranteed by design, not tested in production.

### 6.3.19 Timer characteristics

The parameters given in the following tables are guaranteed by design.

Refer to Section 6.3.14: I/O port characteristics for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 59. TIMx characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
t <sub>res(TIM)</sub>	Timer resolution time	-	-	1	-	t <sub>TIMxCLK</sub>
res(TIM)	Time resolution time	f <sub>TIMxCLK</sub> = 48 MHz	-	20.8	-	ns
f	Timer external clock	-	-	f <sub>TIMxCLK</sub> /2	-	MHz
f <sub>EXT</sub>	frequency on CH1 to CH4	f <sub>TIMxCLK</sub> = 48 MHz	-	24	-	MHz
	16-bit timer maximum	-	ı	2 <sup>16</sup>	ı	t <sub>TIMxCLK</sub>
tury count	period	f <sub>TIMxCLK</sub> = 48 MHz	-	1365	ı	μs
t <sub>MAX_</sub> COUNT	32-bit counter	-	-	2 <sup>32</sup>	-	t <sub>TIMxCLK</sub>
	maximum period	f <sub>TIMxCLK</sub> = 48 MHz	-	89.48	-	s

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Measured at V<sub>DDA</sub> = 3.3 V ± 10 mV. The V<sub>30</sub> ADC conversion result is stored in the TS\_CAL1 byte. Refer to Table 3: Temperature sensor calibration values.

Prescaler divider	PR[2:0] bits	Min timeout RL[11:0]= 0x000	Max timeout RL[11:0]= 0xFFF	Unit		
/4	0	0.1	409.6			
/8	1	0.2	819.2			
/16	2	0.4	1638.4			
/32	3	0.8	3276.8	ms		
/64	4	1.6	6553.6			
/128	5	3.2	13107.2			
/256	6 or 7	6.4	26214.4			

Table 60. IWDG min/max timeout period at 40 kHz (LSI)<sup>(1)</sup>

These timings are given for a 40 kHz clock but the microcontroller internal RC frequency can vary from 30 to 60 kHz. Moreover, given an exact RC oscillator frequency, the exact timings still depend on the phasing of the APB interface clock versus the LSI clock so that there is always a full RC period of uncertainty.

	Table 01: WWD0 Hill/Hax timeout value at 40 Hill (1 0211)						
Prescaler	WDGTB	Min timeout value	Max timeout value	Unit			
1	0	0.0853	5.4613				
2	1	0.1706	10.9226	ms			
4	2	0.3413	21.8453	1115			
8	3	0.6826	43.6906				

Table 61. WWDG min/max timeout value at 48 MHz (PCLK)

### 6.3.20 Communication interfaces

### I<sup>2</sup>C interface characteristics

The I<sup>2</sup>C interface meets the timings requirements of the I<sup>2</sup>C-bus specification and user manual rev. 03 for:

- Standard-mode (Sm): with a bit rate up to 100 kbit/s
- Fast-mode (Fm): with a bit rate up to 400 kbit/s
- Fast-mode Plus (Fm+): with a bit rate up to 1 Mbit/s.

The I<sup>2</sup>C timings requirements are guaranteed by design when the I2Cx peripheral is properly configured (refer to Reference manual).

The SDA and SCL I/O requirements are met with the following restrictions: the SDA and SCL I/O pins are not "true" open-drain. When configured as open-drain, the PMOS connected between the I/O pin and  $V_{\rm DDIOx}$  is disabled, but is still present. Only FTf I/O pins support Fm+ low level output current maximum requirement. Refer to Section 6.3.14: I/O port characteristics for the I<sup>2</sup>C I/Os characteristics.

All I<sup>2</sup>C SDA and SCL I/Os embed an analog filter. Refer to the table below for the analog filter characteristics:

Table 62. I<sup>2</sup>C analog filter characteristics<sup>(1)</sup>

Symbol	Parameter	Min	Max	Unit
t <sub>AF</sub>	Maximum width of spikes that are suppressed by the analog filter	50 <sup>(2)</sup>	260 <sup>(3)</sup>	ns

- 1. Guaranteed by design, not tested in production.
- 2. Spikes with widths below t<sub>AF(min)</sub> are filtered.
- 3. Spikes with widths above  $t_{\text{AF}(\text{max})}$  are not filtered

### SPI/I<sup>2</sup>S characteristics

Unless otherwise specified, the parameters given in *Table 63* for SPI or in *Table 64* for I<sup>2</sup>S are derived from tests performed under the ambient temperature, f<sub>PCLKx</sub> frequency and supply voltage conditions summarized in *Table 21: General operating conditions*.

Refer to Section 6.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI and WS, CK, SD for I<sup>2</sup>S).

Table 63. SPI characteristics<sup>(1)</sup>

Symbol	Parameter	Parameter Conditions		Max	Unit
f <sub>SCK</sub>	SPI clock frequency	Master mode	-	18	MHz
1/t <sub>c(SCK)</sub>	SPI Clock frequency	Slave mode	-	18	IVI□Z
t <sub>r(SCK)</sub> t <sub>f(SCK)</sub>	SPI clock rise and fall time	Capacitive load: C = 15 pF	-	6	ns
t <sub>su(NSS)</sub>	NSS setup time	Slave mode	4Tpclk	-	
t <sub>h(NSS)</sub>	NSS hold time	Slave mode	2Tpclk + 10	-	
t <sub>w(SCKH)</sub> t <sub>w(SCKL)</sub>	SCK high and low time	Master mode, f <sub>PCLK</sub> = 36 MHz, presc = 4	Tpclk/2 -2	Tpclk/2 + 1	
t <sub>su(MI)</sub>	Data input actus time	Master mode	4	-	
t <sub>su(SI)</sub>	Data input setup time	Slave mode	5	-	
t <sub>h(MI)</sub>	Data input hold time	Master mode	4	-	
t <sub>h(SI)</sub>	Data input hold time	Slave mode	5	-	ns
t <sub>a(SO)</sub> <sup>(2)</sup>	Data output access time	Slave mode, f <sub>PCLK</sub> = 20 MHz	0	3Tpclk	
t <sub>dis(SO)</sub> <sup>(3)</sup>	Data output disable time	Slave mode	0	18	
t <sub>v(SO)</sub>	Data output valid time	Slave mode (after enable edge)	-	22.5	
t <sub>v(MO)</sub>	Data output valid time	Master mode (after enable edge)	-	6	
t <sub>h(SO)</sub>	Data output hold time	Slave mode (after enable edge)	11.5	-	
t <sub>h(MO)</sub>	Todia output noid time	Master mode (after enable edge)	2	-	
DuCy(SCK)	SPI slave input clock duty cycle	Slave mode	25	75	%

- 1. Data based on characterization results, not tested in production.
- 2. Min time is for the minimum time to drive the output and the max time is for the maximum time to validate the data.
- 3. Min time is for the minimum time to invalidate the output and the max time is for the maximum time to put the data in Hi-Z

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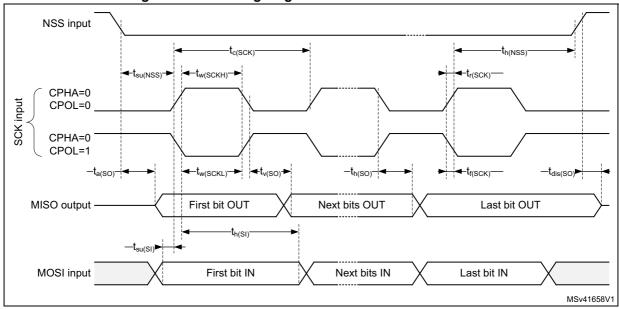
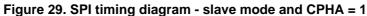
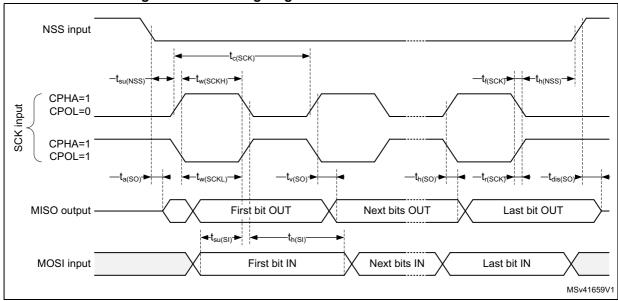


Figure 28. SPI timing diagram - slave mode and CPHA = 0





1. Measurement points are done at CMOS levels: 0.3  $V_{\rm DD}$  and 0.7  $V_{\rm DD}$ .

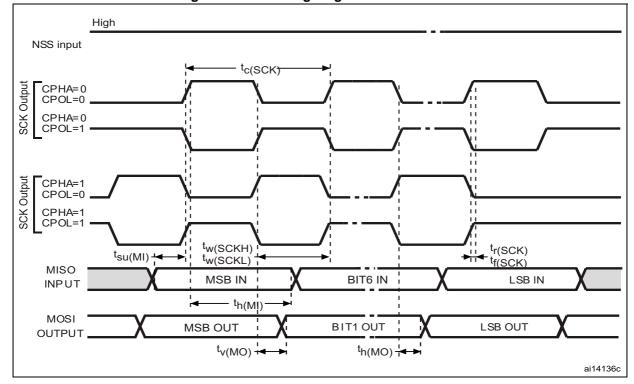


Figure 30. SPI timing diagram - master mode

1. Measurement points are done at CMOS levels: 0.3  $V_{\rm DD}$  and 0.7  $V_{\rm DD}$ .

Table 64. I<sup>2</sup>S characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit
f <sub>CK</sub>	I <sup>2</sup> S clock frequency	Master mode (data: 16 bits, Audio frequency = 48 kHz)	1.597	1.601	MHz
1/t <sub>c(CK)</sub>		Slave mode	0	6.5	
t <sub>r(CK)</sub>	I <sup>2</sup> S clock rise time	Capacitive load C = 15 pE	-	10	
t <sub>f(CK)</sub>	I <sup>2</sup> S clock fall time	- Capacitive load C <sub>L</sub> = 15 pF	-	12	
t <sub>w(CKH)</sub>	I <sup>2</sup> S clock high time	Master f <sub>PCLK</sub> = 16 MHz, audio	306	-	
t <sub>w(CKL)</sub>	I <sup>2</sup> S clock low time	frequency = 48 kHz	312	-	
t <sub>v(WS)</sub>	WS valid time	Master mode	2	-	ns
t <sub>h(WS)</sub>	WS hold time	Master mode	2	-	
t <sub>su(WS)</sub>	WS setup time	Slave mode	7	-	
t <sub>h(WS)</sub>	WS hold time	Slave mode	0	-	
DuCy(SCK)	I <sup>2</sup> S slave input clock duty cycle	Slave mode	25	75	%

	` ,				
Symbol	Parameter	Conditions	Min	Max	Unit
t <sub>su(SD_MR)</sub>	Data input setup time	Master receiver	6	-	
t <sub>su(SD_SR)</sub>	Data input setup time	Slave receiver	2	-	
t <sub>h(SD_MR)</sub> <sup>(2)</sup>	Data input hold time	Master receiver	4	-	
t <sub>h(SD_SR)</sub> (2)	Data input noid time	Slave receiver	0.5	-	no
$t_{v(SD\_MT)}^{(2)}$	Data output valid time	Master transmitter	-	4	ns
t <sub>v(SD_ST)</sub> <sup>(2)</sup>	Data output valid time	Slave transmitter	-	20	
t <sub>h(SD_MT)</sub>	Data output hold time	Master transmitter	0	-	
t <sub>h(SD_ST)</sub>	Data output Hold tillle	Slave transmitter	13	-	

Table 64. I<sup>2</sup>S characteristics<sup>(1)</sup> (continued)

- 1. Data based on design simulation and/or characterization results, not tested in production.
- 2. Depends on  $f_{PCLK}$ . For example, if  $f_{PCLK}$  = 8 MHz, then  $T_{PCLK}$  = 1/ $f_{PLCLK}$  = 125 ns.

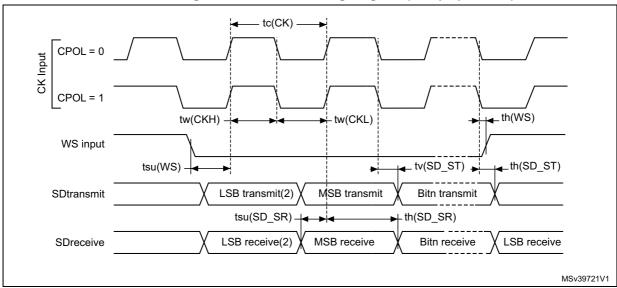


Figure 31. I<sup>2</sup>S slave timing diagram (Philips protocol)

- 1. Measurement points are done at CMOS levels: 0.3 ×  $V_{DDIOx}$  and 0.7 ×  $V_{DDIOx}$ .
- 2. LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

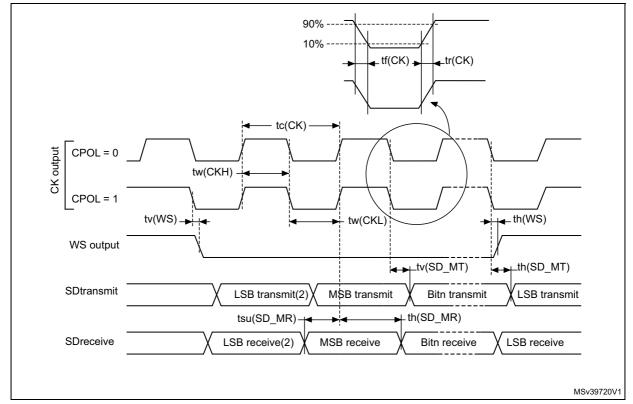


Figure 32. I<sup>2</sup>S master timing diagram (Philips protocol)

- 1. Data based on characterization results, not tested in production.
- LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

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#### **USB** characteristics

The STM32F042x4/x6 USB interface is fully compliant with the USB specification version 2.0 and is USB-IF certified (for Full-speed device operation).

Table 65. USB electrical characteristics

Symbol	Parameter	Conditions	Min.	Тур	Max.	Unit	
V <sub>DDIO2</sub>	USB transceiver operating voltage	-	3.0 <sup>(1)</sup>	-	3.6	V	
t <sub>STARTUP</sub> (2)	USB transceiver startup time	-	-	-	1.0	μs	
R <sub>PUI</sub>	Embedded USB_DP pull-up value during idle	-	1.1	1.26	1.5	5	
R <sub>PUR</sub>	Embedded USB_DP pull-up value during reception	-	2.0	2.26	2.6	kΩ 2.6	
Z <sub>DRV</sub> <sup>(2)</sup>	Output driver impedance <sup>(3)</sup>	Driving high and low	28	40	44	Ω	

The STM32F042x4/x6 USB functionality is ensured down to 2.7 V but not the full USB electrical characteristics which are degraded in the 2.7-to-3.0 V voltage range.

### CAN (controller area network) interface

Refer to Section 6.3.14: I/O port characteristics for more details on the input/output alternate function characteristics (CAN\_TX and CAN\_RX).

<sup>2.</sup> Guaranteed by design, not tested in production.

<sup>3.</sup> No external termination series resistors are required on USB\_DP (D+) and USB\_DM (D-); the matching impedance is already included in the embedded driver.

#### **Package information** 7

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: www.st.com. ECOPACK® is an ST trademark.

#### LQFP48 package information 7.1

LQFP48 is a 48-pin, 7 x 7 mm low-profile quad flat package.

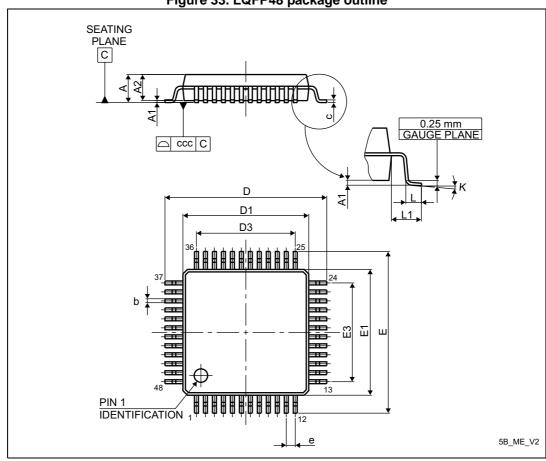


Figure 33. LQFP48 package outline

1. Drawing is not to scale.

Table 66. LQFP48 package mechanical data

Symbol	millimeters inches <sup>(1)</sup>			inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Max
А	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
С	0.090	-	0.200	0.0035	-	0.0079
D	8.800	9.000	9.200	0.3465	0.3543	0.3622
D1	6.800	7.000	7.200	0.2677	0.2756	0.2835
D3	-	5.500	-	-	0.2165	-
E	8.800	9.000	9.200	0.3465	0.3543	0.3622
E1	6.800	7.000	7.200	0.2677	0.2756	0.2835
E3	-	5.500	-	-	0.2165	-
е	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080			0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

9.70 ai14911d

Figure 34. Recommended footprint for LQFP48 package

1. Dimensions are expressed in millimeters.

### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

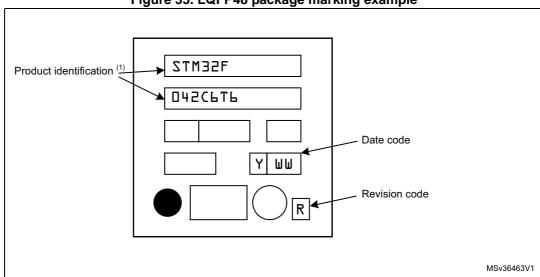


Figure 35. LQFP48 package marking example

 Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.



## 7.2 UFQFPN48 package information

UFQFPN48 is a 48-lead, 7x7 mm, 0.5 mm pitch, ultra-thin fine-pitch quad flat package.

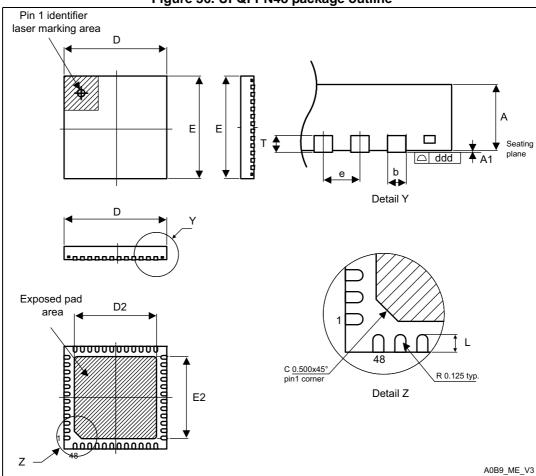


Figure 36. UFQFPN48 package outline

- 1. Drawing is not to scale.
- 2. All leads/pads should also be soldered to the PCB to improve the lead/pad solder joint life.
- There is an exposed die pad on the underside of the UFQFPN package. It is recommended to connect and solder this back-side pad to PCB ground.

Table 67. UFQFPN48 package mechanical data

Cumbal		millimeters			inches <sup>(1)</sup>	
Symbol	Min	Тур	Max	Min	Тур	Max
А	0.500	0.550	0.600	0.0197	0.0217	0.0236
A1	0.000	0.020	0.050	0.0000	0.0008	0.0020
D	6.900	7.000	7.100	0.2717	0.2756	0.2795
E	6.900	7.000	7.100	0.2717	0.2756	0.2795
D2	5.500	5.600	5.700	0.2165	0.2205	0.2244
E2	5.500	5.600	5.700	0.2165	0.2205	0.2244
L	0.300	0.400	0.500	0.0118	0.0157	0.0197
Т	-	0.152	-	-	0.0060	-
b	0.200	0.250	0.300	0.0079	0.0098	0.0118
е	-	0.500	-	-	0.0197	-
ddd	-	-	0.080	-	-	0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

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1. Dimensions are expressed in millimeters.

### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

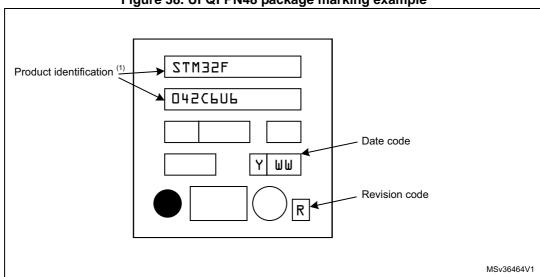


Figure 38. UFQFPN48 package marking example

Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet
qualified and therefore not yet ready to be used in production and any consequences deriving from such
usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering
samples in production. ST Quality has to be contacted prior to any decision to use these Engineering
Samples to run qualification activity.

## 7.3 WLCSP36 package information

WLCSP36 is a 36-ball, 2.605 x 2.703 mm, 0.4 mm pitch wafer-level chip-scale package.

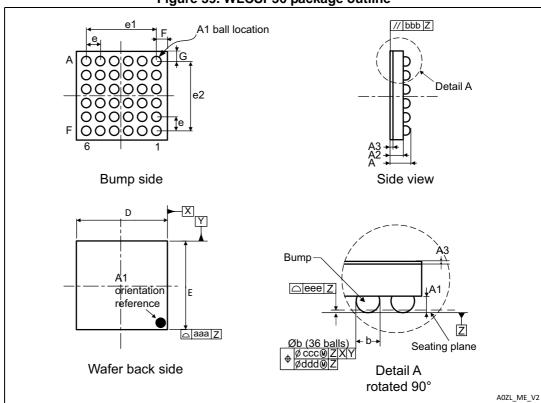


Figure 39. WLCSP36 package outline

1. Drawing is not to scale.

Table 68. WLCSP36 package mechanical data

Symbol		millimeters			inches <sup>(1)</sup>	
Symbol	Min	Тур	Max	Min	Тур	Max
А	0.525	0.555	0.585	0.0207	0.0219	0.0230
A1	-	0.175	-	-	0.0069	-
A2	-	0.380	-	-	0.0150	-
A3 <sup>(2)</sup>	-	0.025	-	-	0.0010	-
b <sup>(3)</sup>	0.220	0.250	0.280	0.0087	0.0098	0.0110
D	2.570	2.605	2.640	0.1012	0.1026	0.1039
E	2.668	2.703	2.738	0.1050	0.1064	0.1078
е	-	0.400	-	-	0.0157	-
e1	-	2.000	-	-	0.0787	-
e2	-	2.000	-	-	0.0787	-



	Table 68. WLCSP36 package mechanical data (continued)						
Cumb al		millimeters			inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Max	
F	-	0.3025	-	-	0.0119	-	
G	-	0.3515	-	-	0.0138	-	
aaa	-	-	0.100	-	-	0.0039	
bbb	-	-	0.100	-	-	0.0039	
ccc	-	-	0.100	-	-	0.0039	
ddd	-	-	0.050	-	-	0.0020	
666	_	_	0.050	_	_	0.0020	

Table 68 WI CSP36 package mechanical data (continued)

- 1. Values in inches are converted from mm and rounded to 4 decimal digits.
- 2. Back side coating.
- 3. Dimension is measured at the maximum bump diameter parallel to primary datum Z.

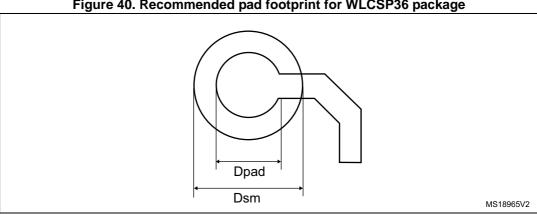


Figure 40. Recommended pad footprint for WLCSP36 package

Table 69. WLCSP36 recommended PCB design rules

Dimension	Recommended values
Pitch	0.4 mm
Dpad	260 µm max. (circular) 220 µm recommended
Dsm	300 μm min. (for 260 μm diameter pad)
PCB pad design	Non-solder mask defined via underbump allowed

### **Device marking**

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

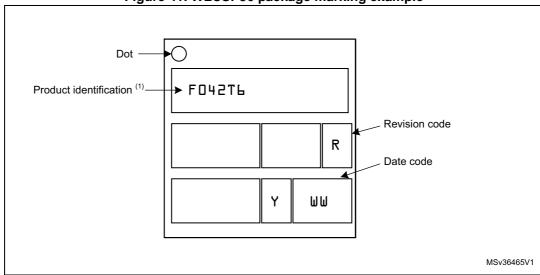


Figure 41. WLCSP36 package marking example

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.



## 7.4 LQFP32 package information

LQFP32 is a 32-pin, 7 x 7 mm low-profile quad flat package.

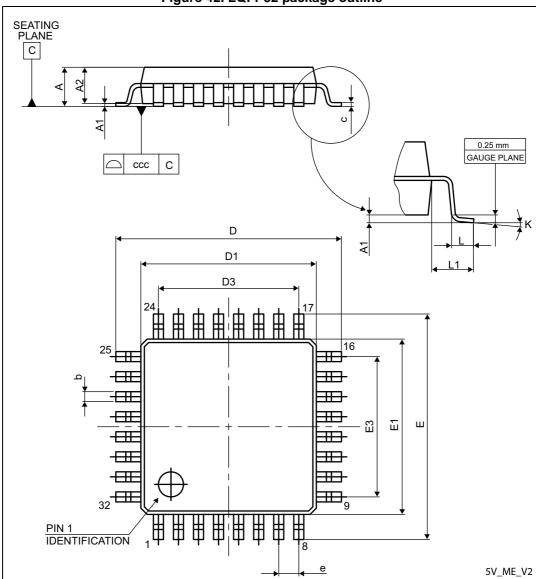


Figure 42. LQFP32 package outline

1. Drawing is not to scale.

Table 70. LQFP32 package mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
Symbol	Min	Тур	Max	Min	Тур	Max
А	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.300	0.370	0.450	0.0118	0.0146	0.0177
С	0.090	-	0.200	0.0035	-	0.0079
D	8.800	9.000	9.200	0.3465	0.3543	0.3622
D1	6.800	7.000	7.200	0.2677	0.2756	0.2835
D3	-	5.600	-	-	0.2205	-
E	8.800	9.000	9.200	0.3465	0.3543	0.3622
E1	6.800	7.000	7.200	0.2677	0.2756	0.2835
E3	-	5.600	-	-	0.2205	-
е	-	0.800	-	-	0.0315	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.100	-	-	0.0039

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 43. Recommended footprint for LQFP32 package 9.70 5V\_FP\_V2

1. Dimensions are expressed in millimeters.

### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

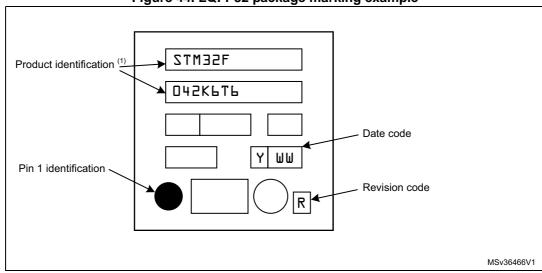


Figure 44. LQFP32 package marking example

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.

### 7.5 UFQFPN32 package information

UFQFPN32 is a 32-pin, 5x5 mm, 0.5 mm pitch ultra-thin fine-pitch quad flat package.

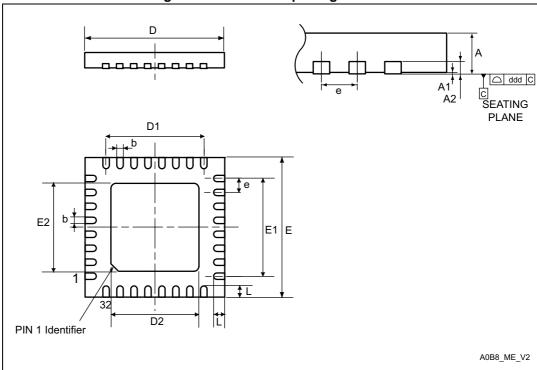


Figure 45. UFQFPN32 package outline

- 1. Drawing is not to scale.
- 2. All leads/pads should also be soldered to the PCB to improve the lead/pad solder joint life.
- There is an exposed die pad on the underside of the UFQFPN package. This pad is used for the device ground and must be connected. It is referred to as pin 0 in Table: Pin definitions.

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Table 71. UFQFPN32 package mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Тур	Max	Min	Тур	Max
Α	0.500	0.550	0.600	0.0197	0.0217	0.0236
A1	0.000	0.020	0.050	0.0000	0.0008	0.0020
A3	-	0.152	-	-	0.0060	-
b	0.180	0.230	0.280	0.0071	0.0091	0.0110
D	4.900	5.000	5.100	0.1929	0.1969	0.2008
D1	3.400	3.500	3.600	0.1339	0.1378	0.1417
D2	3.400	3.500	3.600	0.1339	0.1378	0.1417
E	4.900	5.000	5.100	0.1929	0.1969	0.2008
E1	3.400	3.500	3.600	0.1339	0.1378	0.1417
E2	3.400	3.500	3.600	0.1339	0.1378	0.1417
е	-	0.500	-	-	0.0197	-
L	0.300	0.400	0.500	0.0118	0.0157	0.0197
ddd	-	-	0.080	-	-	0.0031

<sup>1.</sup> Values in inches are converted from mm and rounded to 4 decimal digits.

3.80 5.30 3.80 0.75 3.80

Figure 46. Recommended footprint for UFQFPN32 package

1. Dimensions are expressed in millimeters.

A0B8\_FP\_V2

### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

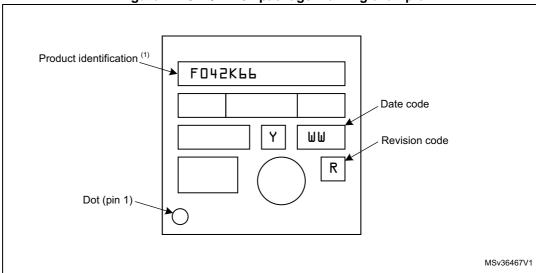


Figure 47. UFQFPN32 package marking example

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.



#### **UFQFPN28** package information 7.6

UFQFPN28 is a 28-lead, 4x4 mm, 0.5 mm pitch, ultra-thin fine-pitch quad flat package.

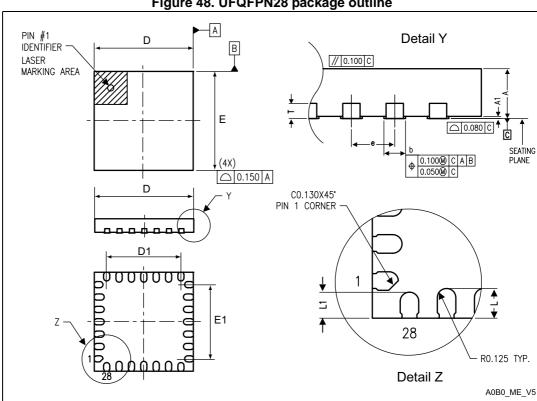


Figure 48. UFQFPN28 package outline

1. Drawing is not to scale.

Table 72. UFQFPN28 package mechanical data<sup>(1)</sup>

Symbol	millimeters			inches		
Symbol	Min	Тур	Max	Min	Тур	Max
А	0.500	0.550	0.600	0.0197	0.0217	0.0236
A1	-	0.000	0.050	-	0.0000	0.0020
D	3.900	4.000	4.100	0.1535	0.1575	0.1614
D1	2.900	3.000	3.100	0.1142	0.1181	0.1220
Е	3.900	4.000	4.100	0.1535	0.1575	0.1614
E1	2.900	3.000	3.100	0.1142	0.1181	0.1220
L	0.300	0.400	0.500	0.0118	0.0157	0.0197
L1	0.250	0.350	0.450	0.0098	0.0138	0.0177
Т	-	0.152	-	-	0.0060	-
b	0.200	0.250	0.300	0.0079	0.0098	0.0118
е	-	0.500	-	-	0.0197	-

1. Values in inches are converted from mm and rounded to 4 decimal digits.

3.30 4.30 0.21 0.20 3.20 0.20 3.30 0.50 0.50 0.50 0.50

Figure 49. Recommended footprint for UFQFPN28 package

1. Dimensions are expressed in millimeters.



### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

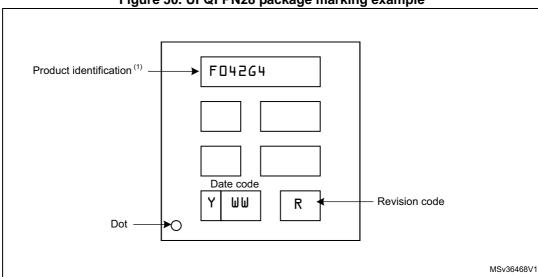


Figure 50. UFQFPN28 package marking example

<sup>1.</sup> Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.

## 7.7 TSSOP20 package information

TSSOP20 is a 20-lead thin shrink small-outline, 6.5 x 4.4 mm, 0.65 mm pitch, package.

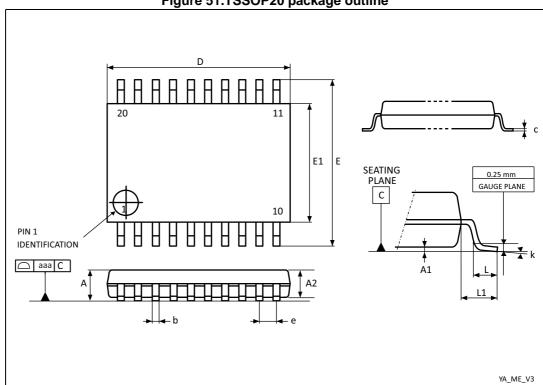


Figure 51.TSSOP20 package outline

1. Drawing is not to scale.

Table 73. TSSOP20 package mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
Symbol	Min.	Тур.	Max.	Min.	Тур.	Max.
А	-	-	1.200	-	-	0.0472
A1	0.050	-	0.150	0.0020	-	0.0059
A2	0.800	1.000	1.050	0.0315	0.0394	0.0413
b	0.190	-	0.300	0.0075	-	0.0118
С	0.090	-	0.200	0.0035	-	0.0079
D <sup>(2)</sup>	6.400	6.500	6.600	0.2520	0.2559	0.2598
E	6.200	6.400	6.600	0.2441	0.2520	0.2598
E1 <sup>(3)</sup>	4.300	4.400	4.500	0.1693	0.1732	0.1772
е	-	0.650	-	-	0.0256	
L	0.450	0.600	0.750	0.0177	0.0236	0.0295

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inches<sup>(1)</sup> millimeters **Symbol** Min. Тур. Max. Min. Тур. Max. L1 1.000 0.0394 0° 8° 0° 8° k 0.100 0.0039 aaa

Table 73. TSSOP20 package mechanical data (continued)

- 1. Values in inches are converted from mm and rounded to four decimal digits.
- 2. Dimension "D" does not include mold flash, protrusions or gate burrs. Mold flash, protrusions or gate burrs shall not exceed 0.15mm per side.
- 3. Dimension "E1" does not include interlead flash or protrusions. Interlead flash or protrusions shall not exceed 0.25mm per side.

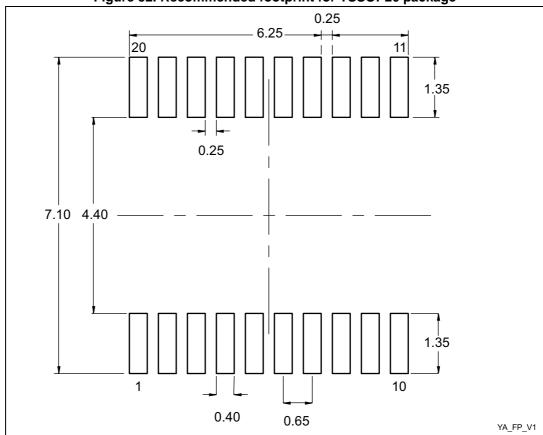


Figure 52. Recommended footprint for TSSOP20 package

1. Dimensions are expressed in millimeters.

### **Device marking**

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

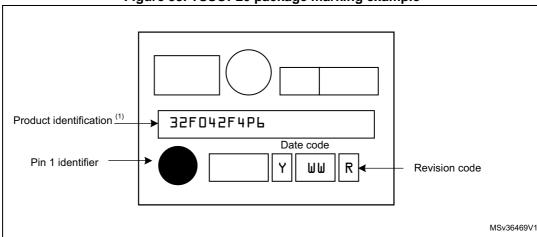


Figure 53. TSSOP20 package marking example

1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.



### 7.8 Thermal characteristics

The maximum chip junction temperature (T<sub>J</sub>max) must never exceed the values given in *Table 21: General operating conditions*.

The maximum chip-junction temperature,  $T_J$  max, in degrees Celsius, may be calculated using the following equation:

$$T_J \max = T_A \max + (P_D \max x \Theta_{JA})$$

#### Where:

- T<sub>A</sub> max is the maximum ambient temperature in °C,
- ullet  $\Theta_{JA}$  is the package junction-to-ambient thermal resistance, in °C/W,
- $P_D$  max is the sum of  $P_{INT}$  max and  $P_{I/O}$  max ( $P_D$  max =  $P_{INT}$  max +  $P_{I/O}$ max),
- P<sub>INT</sub> max is the product of I<sub>DD</sub> and V<sub>DD</sub>, expressed in Watts. This is the maximum chip internal power.

P<sub>I/O</sub> max represents the maximum power dissipation on output pins where:

$$P_{I/O}$$
 max =  $\Sigma (V_{OL} \times I_{OL}) + \Sigma ((V_{DDIOx} - V_{OH}) \times I_{OH})$ ,

taking into account the actual  $V_{OL}$  /  $I_{OL}$  and  $V_{OH}$  /  $I_{OH}$  of the I/Os at low and high level in the application.

Symbol	Parameter	Value	Unit
	Thermal resistance junction-ambient LQFP48 - 7 mm x 7 mm	55	
	Thermal resistance junction-ambient UFQFPN48 - 7 mm x 7 mm	33	
$\Theta_{JA}$	Thermal resistance junction-ambient WLCSP36 2.6 mm x 2.7 mm	64	
	Thermal resistance junction-ambient LQFP32 - 7 mm x 7 mm	57	°C/W
	Thermal resistance junction-ambient UFQFPN32 - 5 mm x 5 mm	38	
	Thermal resistance junction-ambient UFQFPN28 - 4 mm x 4 mm	118	
	Thermal resistance junction-ambient TSSOP20 - 6.5 mm x 6.4 mm	76	

Table 74. Package thermal characteristics

### 7.8.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org

### 7.8.2 Selecting the product temperature range

When ordering the microcontroller, the temperature range is specified in the ordering information scheme shown in *Section 8: Ordering information*.

Each temperature range suffix corresponds to a specific guaranteed ambient temperature at maximum dissipation and, to a specific maximum junction temperature.

As applications do not commonly use the STM32F042x4/x6 at maximum dissipation, it is useful to calculate the exact power consumption and junction temperature to determine which temperature range will be best suited to the application.

The following examples show how to calculate the temperature range needed for a given application.



## 8 Ordering information

For a list of available options (memory, package, and so on) or for further information on any aspect of this device, please contact your nearest ST sales office.

Table 75. Ordering information scheme

STM32 F Example: 042 С 6 XXX **Device family** STM32 = ARM-based 32-bit microcontroller **Product type** F = General-purpose **Sub-family** 042 = STM32F042xxPin count F = 20 pins G = 28 pins K = 32 pins T = 36 pinsC = 48 pins User code memory size 4 = 16 Kbyte 6 = 32 Kbyte **Package** P = TSSOP T = LQFPU = UFQFPN Y = WLCSP Temperature range  $6 = -40 \text{ to } 85 ^{\circ}\text{C}$  $7 = -40 \text{ to } 105 \,^{\circ}\text{C}$ **Options** xxx = code ID of programmed parts (includes packing type)



TR = tape and reel packing blank = tray packing

# 9 Revision history

**Table 76. Document revision history** 

Date	Revision	Changes	
25-Feb-2014	1	Initial release.	
03-Apr-2014	2	Added the sample engineering sections for all the packages in the chapter Package information:  Updated tables:  STM32F042x4/x6 USART implementation: added one table footnote.  STM32F042x pin definitions,  Current characteristics,  Typical and maximum current consumption from VDD supply at VDD = 3.6 V,  Typical and maximum current consumption from the VDDA supply,  Typical and maximum current consumption from the VBAT supply,  Typical current consumption, code executing from Flash, running from HSE 8 MHz crystal,  Flash memory characteristics,  I/O static characteristics,  I/O current injection susceptibility,  EMS characteristics,  Updated figures:  UFQFPN32 32-pin package pinout,  UQFPN28 28-pin package,  Power supply scheme,  TC and TTa I/O input characteristics,  Five volt tolerant (FT and FTf) I/O input characteristics.  LQFP48 marking example (package top view),  UFQFPN28 marking example (package top view),  UFQFPN32 marking example (package top view),  TSSOP20 marking example (package top view)	
26-Oct-2015	3	Cover page: number of I/Os and timers updated.  Updates in Section 2: Description:  - updated Figure 1: Block diagram  Updates in Section 3: Functional overview:  - updated Figure 2: Clock tree  - addition of the number of complementary outputs for the advanced control timer and for TIM16, TIM17 general purpose timers in Table 7: Timer feature comparison	



Table 76. Document revision history (continued)

Date	Revision	Changes
		- Table 9: STM32F042x4/x6 I <sup>2</sup> C implementation - adding 20 mA
		Updates in Section 4: Pinouts and pin descriptions
		<ul> <li>Table 12: Legend/abbreviations used in the pinout table - removing</li> <li>"I" pin type</li> </ul>
		Updates in Section 5: Memory mapping:
		<ul> <li>Figure 10: STM32F042x6 memory map, x4 difference described in text</li> </ul>
		Updates in Section 6: Electrical characteristics:
		<ul> <li>the condition "Regulator in run mode, all oscillators OFF" in Table 28:</li> <li>Typical and maximum consumption in Stop and Standby modes,</li> </ul>
		- footnote for V <sub>IN</sub> max value in <i>Table 18: Voltage characteristics,</i>
		- footnote for max V <sub>IN</sub> in <i>Table 21: General operating conditions</i> ,
		t <sub>START</sub> parameter definition in <i>Table 25: Embedded internal</i> reference voltage
	3	<ul> <li>addition of t<sub>START</sub> parameter in <i>Table 25: Embedded internal</i> reference voltage, removal of -40°C to 85°C condition and the associated footnote</li> </ul>
		<ul> <li>Table 26: Typical and maximum current consumption from VDD supply at VDD = 3.6 V: removing "code executing from Flash or RAM"</li> </ul>
26-Oct-2015		<ul> <li>removal of the min value for t<sub>START</sub> parameter in <i>Table 57: TS</i></li> <li>characteristics</li> </ul>
		<ul> <li>the typical value for R parameter in Table 58: VBAT monitoring characteristics</li> </ul>
		<ul> <li>removal of Res<sub>TM</sub> parameter line from <i>Table 59: TIMx characteristics</i> and putting all values in new Typ column, substitution of t<sub>COUNTER</sub> with t<sub>MAX_COUNT</sub>, values defined as powers of two</li> </ul>
		V <sub>ESD(CDM)</sub> class in <i>Table 47: ESD absolute maximum ratings</i>
		- reorganization of <i>Table 64: I<sup>2</sup>S characteristics a</i> nd filling max value
		of t <sub>v(SD_ST)</sub> – adding definition of levels in <i>Figure 32: l</i> <sup>2</sup> S master timing diagram ( <i>Philips protocol</i> )
		Updates in Section 7: Package information:
		- heading and display of columns in <i>Table 68: WLCSP36 package mechanical data.</i> ,
		- Figure 38: UFQFPN48 package marking example
		- Figure 41: WLCSP36 package marking example  - Figure 41: WLCSP36 package marking example
		- Figure 50: UFQFPN28 package marking example
		- Figure 41: WLCSP36 package marking example
		- Figure 51: TSSOP20 package outline - correcting GAGE to GAUGE
		- removing "die 445" from <i>Table 74: Package thermal characteristics</i>
		Updates in Section 8: Part numbering:
		<ul> <li>adding tray packing to options</li> </ul>



Table 76. Document revision history (continued)

Date	Revision	Changes
16-Dec-2015	4	Section 3: Functional overview:  - Figure 2: Clock tree modified  Section 4: Pinouts and pin descriptions:  - Package pinout figures updated (look and feel)  - Figure 5: WLCSP36 package pinout- now presented in top view  - Table 13: STM32F042x pin definitions - note 3 added; CIMP1_OUT and USART4_CTS removed  - Table 15: Alternate functions selected through GPIOB_AFR registers for port B - change of I2C2_SDA and I2C2_SCL to I2C1_SDA and I2C1_SCL  Section 5: Memory mapping:  - Table 17: STM32F042x4/x6 peripheral register boundary addresses - change of "SYSCFG + COMP" to "SYSCFG"  Section 6: Electrical characteristics:  - Table 50: I/O static characteristics - removed note  - Section 6.3.16: 12-bit ADC characteristics - changed introductory sentence  Section 7: Package information:  - Figure 49: Recommended footprint for UFQFPN28 package distance between corner pads added
10-Jan-2017	5	Section 6: Electrical characteristics:  - Table 37: LSE oscillator characteristics (fLSE = 32.768 kHz) - information on configuring different drive capabilities removed. See the corresponding reference manual.  - Table 25: Embedded internal reference voltage - V <sub>REFINT</sub> values  - Figure 28: SPI timing diagram - slave mode and CPHA = 0 and Figure 29: SPI timing diagram - slave mode and CPHA = 1 enhanced and corrected  Section 8: Ordering information:  - The name of the section changed from the previous "Part numbering"



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