

ICLAB 2023 Fall

LAB 12 : Cell-Based APR Design Flow

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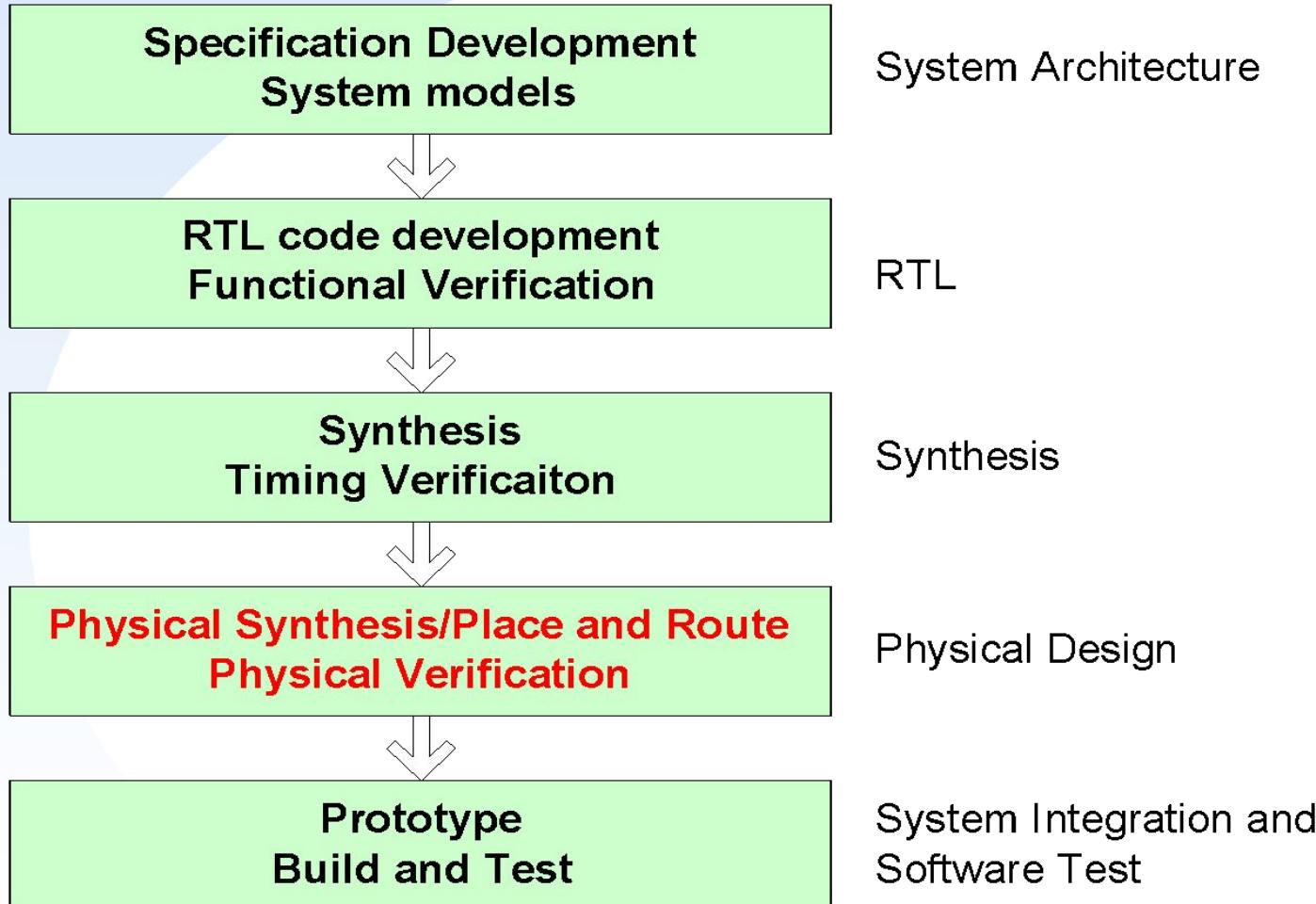
Outline

- ✓ **Section 1- What's APR**
- ✓ **Section 2- SOP of APR**



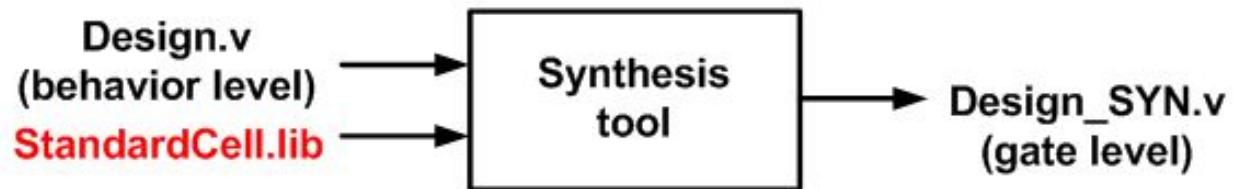
Introduction

✓ Cell-based Design Flow

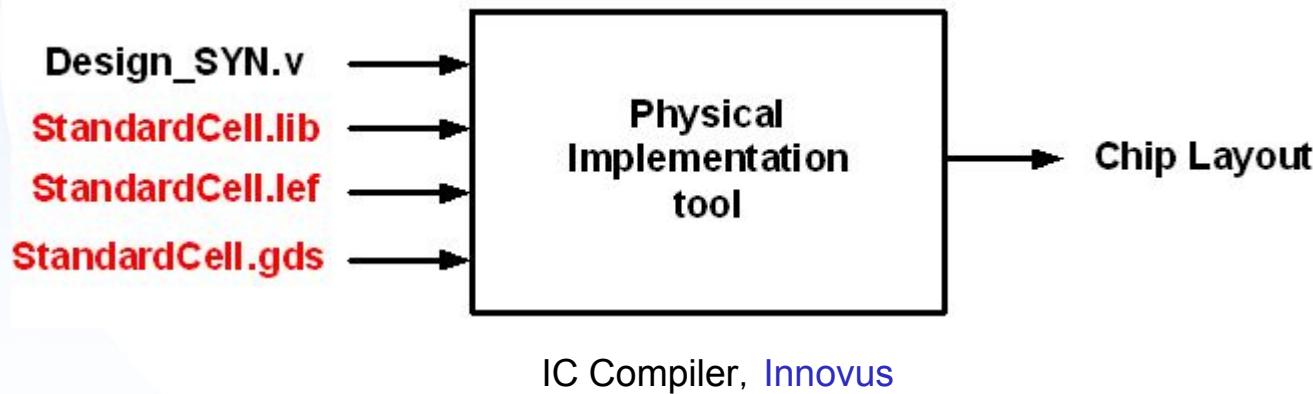


Review Cell-based Design Flow

- ✓ RTL coding
- ✓ Synthesis



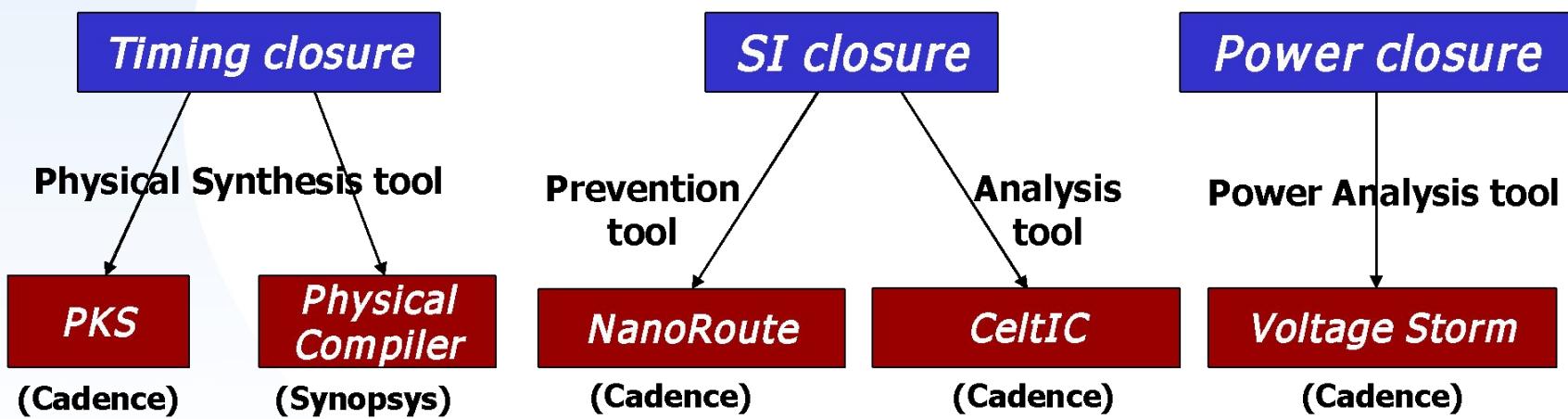
- ✓ Physical Implementation



Wiring Problem

✓ New problems due to large wire resistance and capacitance

- Timing closure
- Signal Integrity(SI) closure(crosstalk, ...)
- Power closure (IR drop, ...)



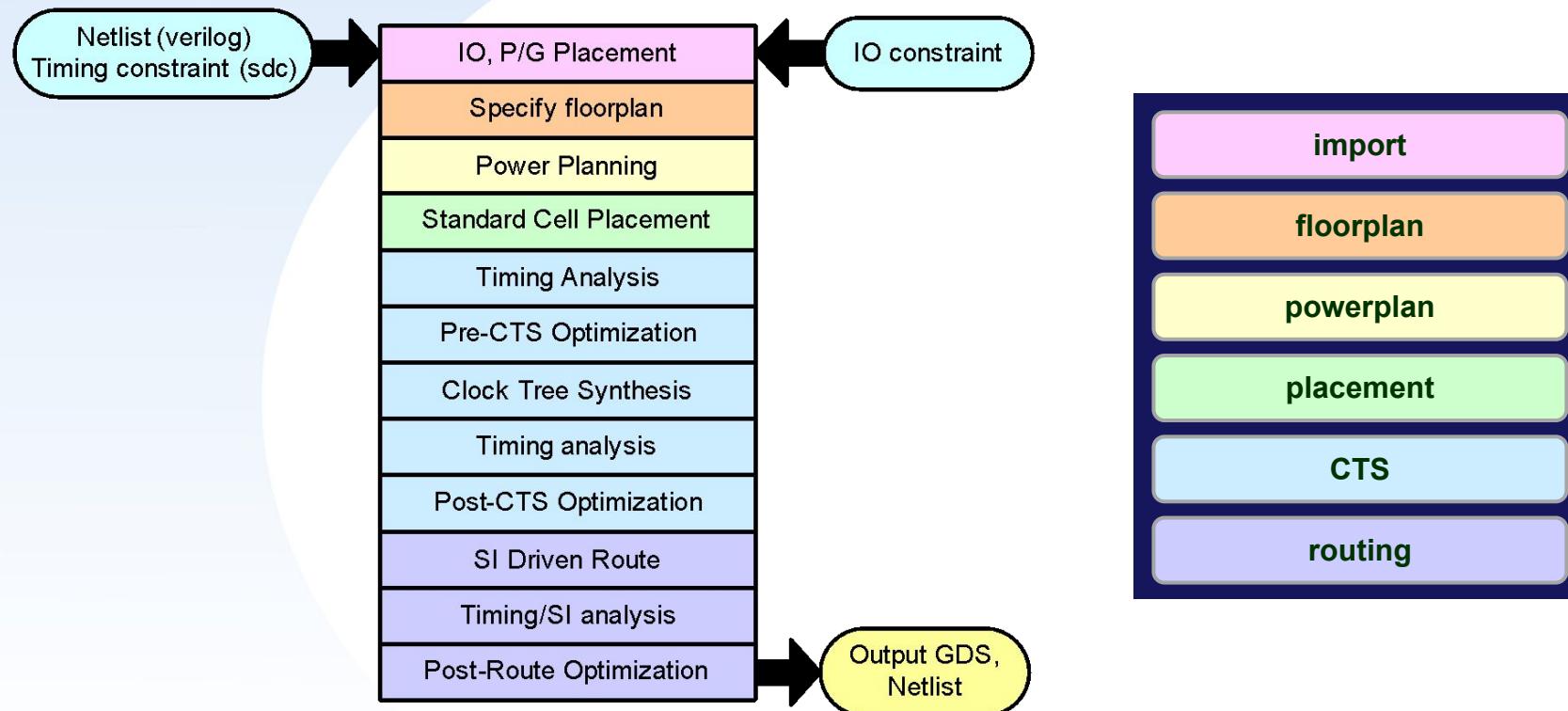
Outline

- ✓ Section 1- What's APR
- ✓ Section 2- SOP of APR



APR Flow

Innovus Design Flow



1. Data Preparation

✓ Library files

- Timing libraries (LIB)
 - slow.lib
 - fast.lib
 - umc18io3v5v_slow.lib
 - umc18io3v5v_fast.lib
 - HardMacro_slow.lib
 - HardMacro_fast.lib
- Physical libraries (LEF)
 - umc18_6lm.lef
 - umc18io3v5v_6lm.lef
 - umc18_6lm_antenna.lef
 - HardMacro.vclef
- RC extraction
 - umc18_1p6m.captbl
 - RCGen.tch
- CeltIC libraries
 - slow.cdb
 - fast.cdb

- GDSII layout

- umc18.gds2
- umc18io3v5v_61m.gds2
- HardMacro.gds2

✓ Timing constraint files (User Data)

- Generate timing constraint files from Design Complier
 - dc_shell-t> write_sdc CHIP.sdc

✓ Design netlist files (User Data)

- Synthesized design netlist
 - core_syn.v
- Chip design netlist
(combine core_syn.v&chip_shell.v)
 - CHIP_SYN.v
- IO pad location file
 - CHIP.io



1. Data Preparation – Library Files

✓ LIB Library : timing information

- Calculate cell delay according to different combinations of input slew rate and output loading capacitance
- Using SignalStorm Library Characterizer

✓ LEF Library : process and cell information

- Metal layer information, routing pitch, default direction
- Cell size, pin position, pin metal layer

✓ RC Extraction

- RC extraction for further power analysis, simulation etc

✓ CelIIC Library

- cdb model (noise model for each cell which will be used in SI optimization phase)

✓ GDSII

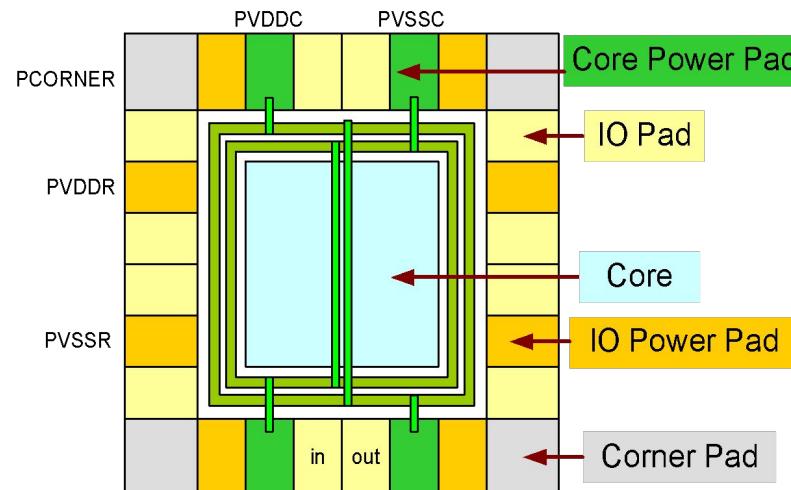
- planar geometric shapes and other information about the layout in hierarchical form



1. Data Preparation - Design Netlist

✓ Prepare chip design netlist: CHIP_SYN.v

- Add pad netlist to synthesized design netlist
- Combine *designName_SYN.v* and *CHIP_SHELL.v*
 - `cat CHIP_SHELL.v CORE_SYN.v > CHIP_SYN.v`
- **CHIP_SHELL.v** contains I/O, I/O power, and core power
 - Note: Refer to pad cell library datasheet
 - <evince ~iclabta01/umc018/Doc/umc18io3v5v.pdf> &



1. Data Preparation - Design Netlist

✓ Core/IO power pad selection (UMC 0.18)

– Pad cell categories

Core power pads: PVDDC, PVSSC

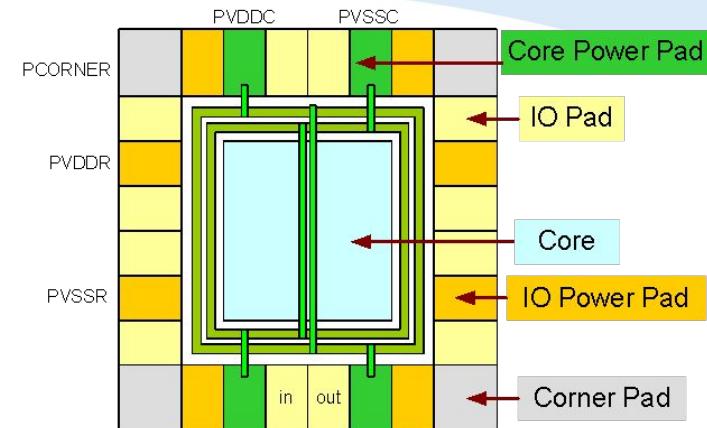
IO power pads: PVDDR, PVSSR

Special Cells
I/O filler pads: PFILL, ...

Corner pads: PCORNER

I/O pads:

P2A	P4A	P8A	P16A	P24A
P2C	P4C	P8C	P16C	P24C



```
PVDDC VDDC0 ();  
PVSSC GND0 ();  
P8C I_CLK [(.Y(C_clk), .P(clk), .A(1'b0), .ODEN(1'b0),  
.OCEN(1'b0), .PU(1'b1), .PD(1'b0), .CEN(1'b0), .CSEN(1'b1)];
```

– Core power pad

- One set core power pad (PVDDC along with PVSSC) can provide 40~50mA current

– IO power pad

- One set IO power pad (PVDDR along with PVSSR) can provide the power for
 - ◆ 3~4 output pads/PG pad set
 - ◆ 6~8 input pads pads/PG pad set



1. Data Preparation – I/O Files

✓ Prepare IO pad location file: CHIP.io

- Edit IO pad location file
 - CHIP.io
- Syntax
 - Pad: pad_instance_name direction [pad_type]
- Categories

Field	Usage			
pad_instance_name	Corresponding instance name of pad in the chip design netlist			
direction	Specify pad location in the floorplan	S (southern pads)		
		E (eastern pads)		
		W (western pads)		
		N (northern pads)		
pad_type	Specify pad type (only for pad filler and corner pad)	PFILL (pad fillers)		
		PCORNER (corner pads)		



1. Data Preparation – I/O Files

✓ Example of IO pad location file

CHIP_SYN.v

```
PVDDR pi11 ();  
PVSSR pi12 ();
```

CHIP.io

Version: 2

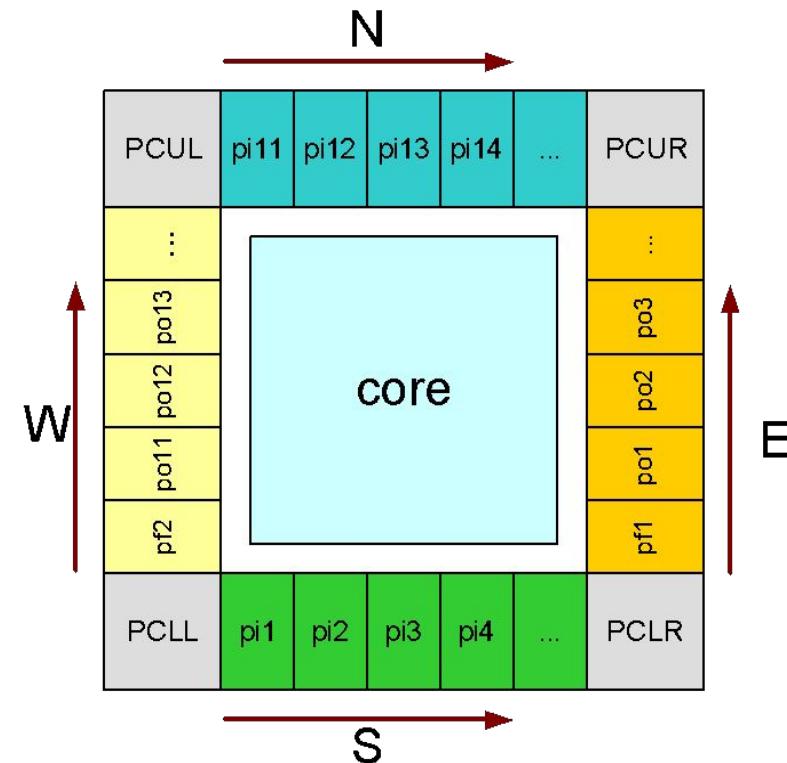
Pad: pi1 S
Pad: pi2 S
Pad: pi3 S
Pad: pi4 S
.....

Pad: pf1 E PFILL
Pad: po1 E
Pad: po2 E
Pad: po3 E
.....

Pad: pi11 N
Pad: pi12 N
Pad: pi13 N
Pad: pi14 N
.....

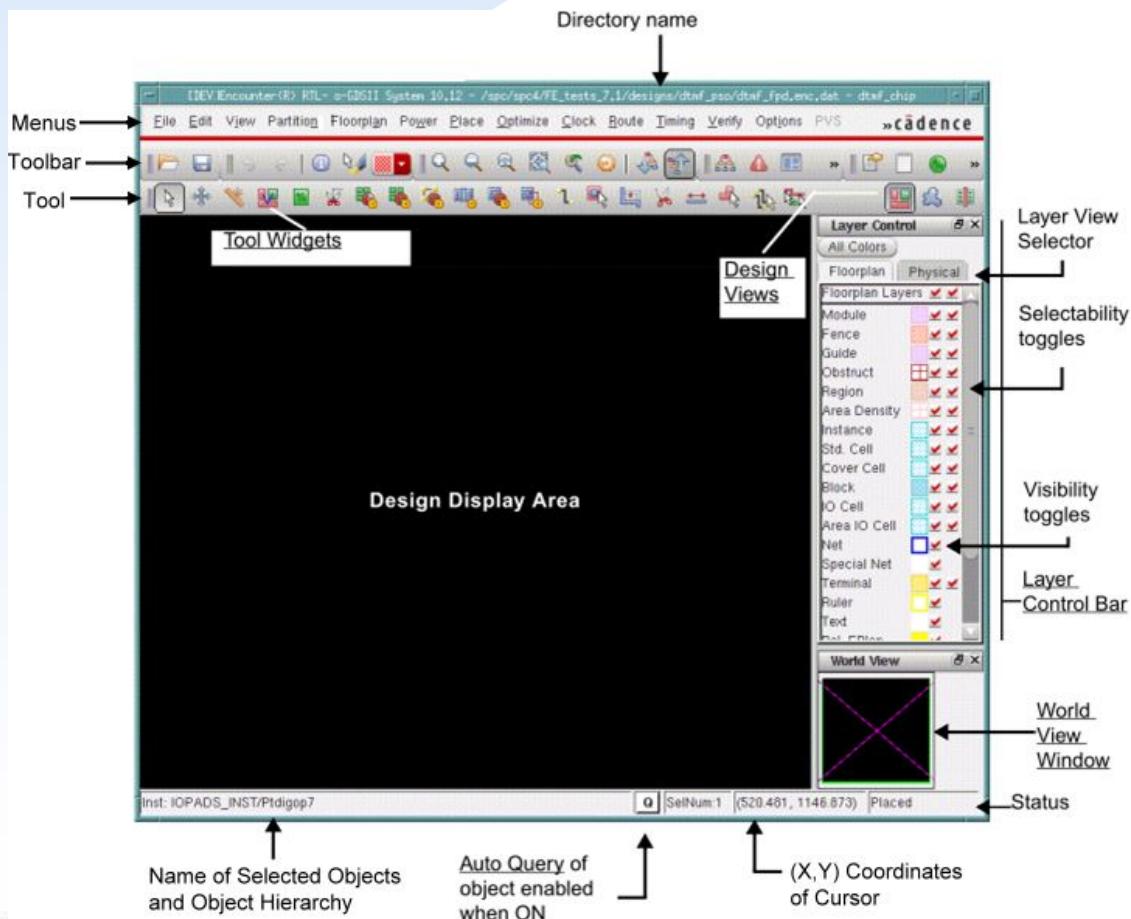
Pad: pf2 W PFILL
Pad: po11 W
Pad: po12 W
Pad: po13 W
.....

Pad: PCLL SW PCORNER
Pad: PCLR SE PCORNER
Pad: PCUL NW PCORNER
Pad: PCUR NE PCORNER



2. Importing Design – Invoke Innovus

- ✓ Command to start an Innovus session
 - **UNIX% innovus**



Note:

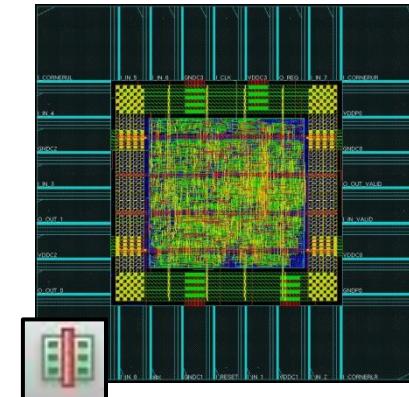
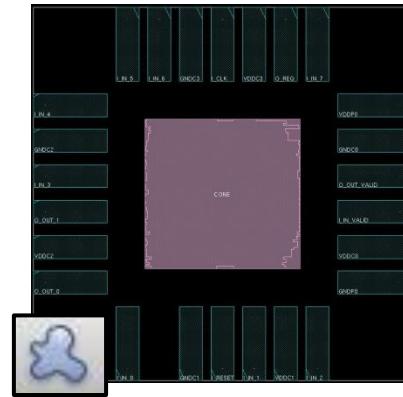
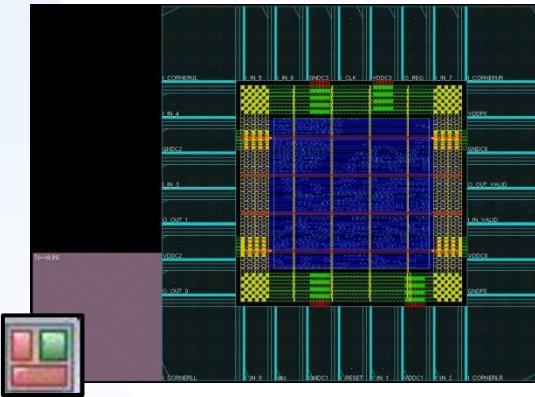
1. Do not use background execution!!!
(The terminal become the interface of command input while running soc innovus)
2. Log file name:
innovus.log
innovus.log1
innovus.log2
...
3. Command file name:
innovus.cmd
innovus.cmd1
innovus.cmd2
...



2. Importing Design – Innovus Views

✓ Design Views

-  Floorplan View: displays the hierarchical module and block guides, connection flight lines and floorplan objects
-  Amoeba View: display the outline of modules after placement
-  Placement View: display the detailed placements of cells.



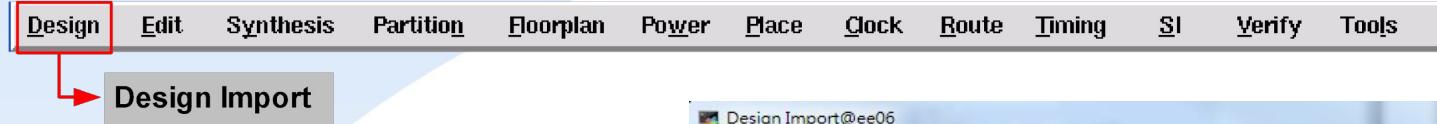
✓ Set design view

- innovus #> [setDrawMode fplan](#)

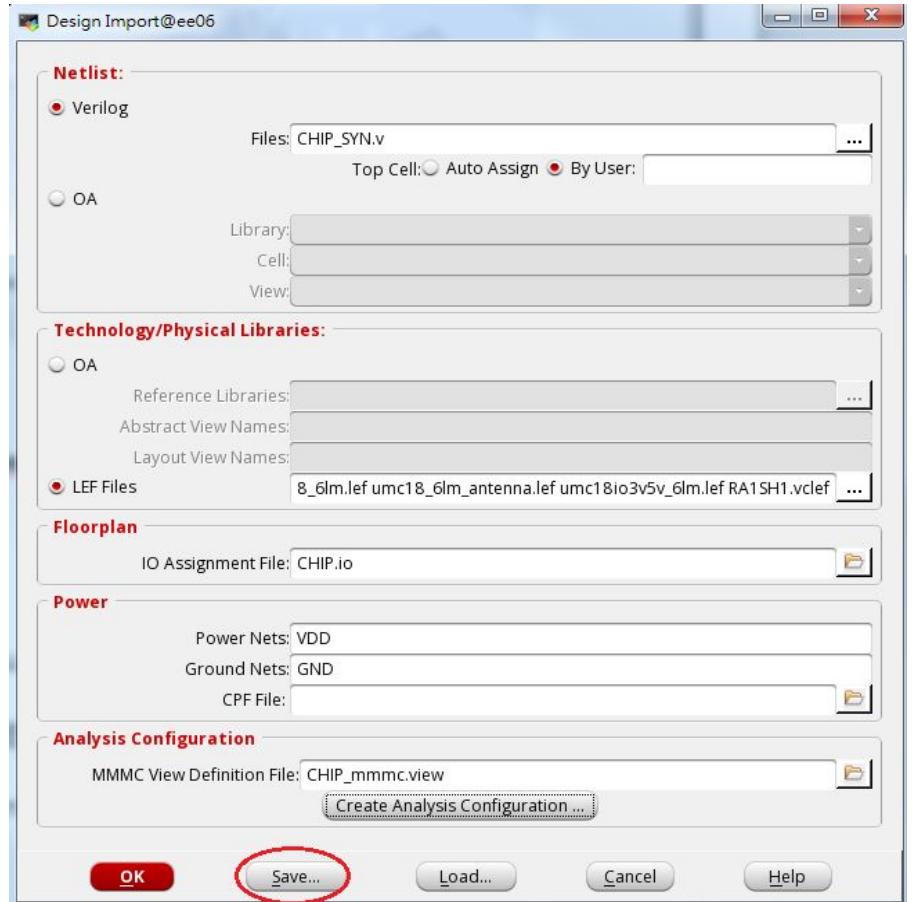


2. Importing Design

✓ Import design files into Innovus environment



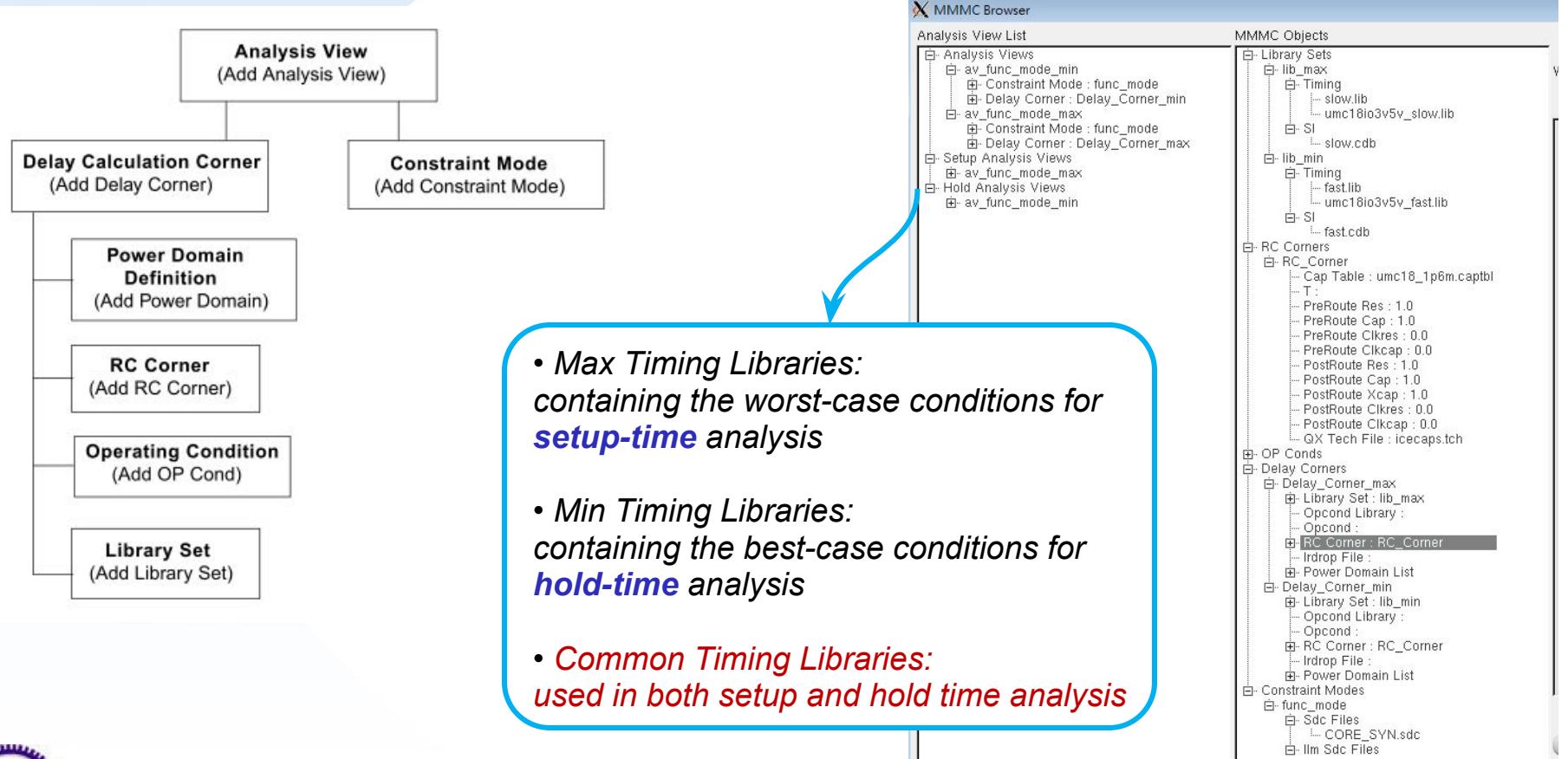
- Netlist:
By Verilog, and specifies the top cell by user
- Technology/Physical Libraries:
You must specify the technology LEF file first, then specify the standard cell LEF and block LEF. (Use LEF flow here)
- Floorplan
Specifies the I/O files to import.



2. Importing Design – MMMC

✓ Multi-mode multi-corner (MMMC) timing analysis and optimization

- Press “Create Analysis Configuration” tab
- In MMMC environment you must specify the lower-level information first.



- *Max Timing Libraries:* containing the worst-case conditions for **setup-time** analysis
- *Min Timing Libraries:* containing the best-case conditions for **hold-time** analysis
- *Common Timing Libraries:* used in both setup and hold time analysis



2. Importing Design – MMMC

Multi-mode

run

test

Multi-corner

slow

fast

view1

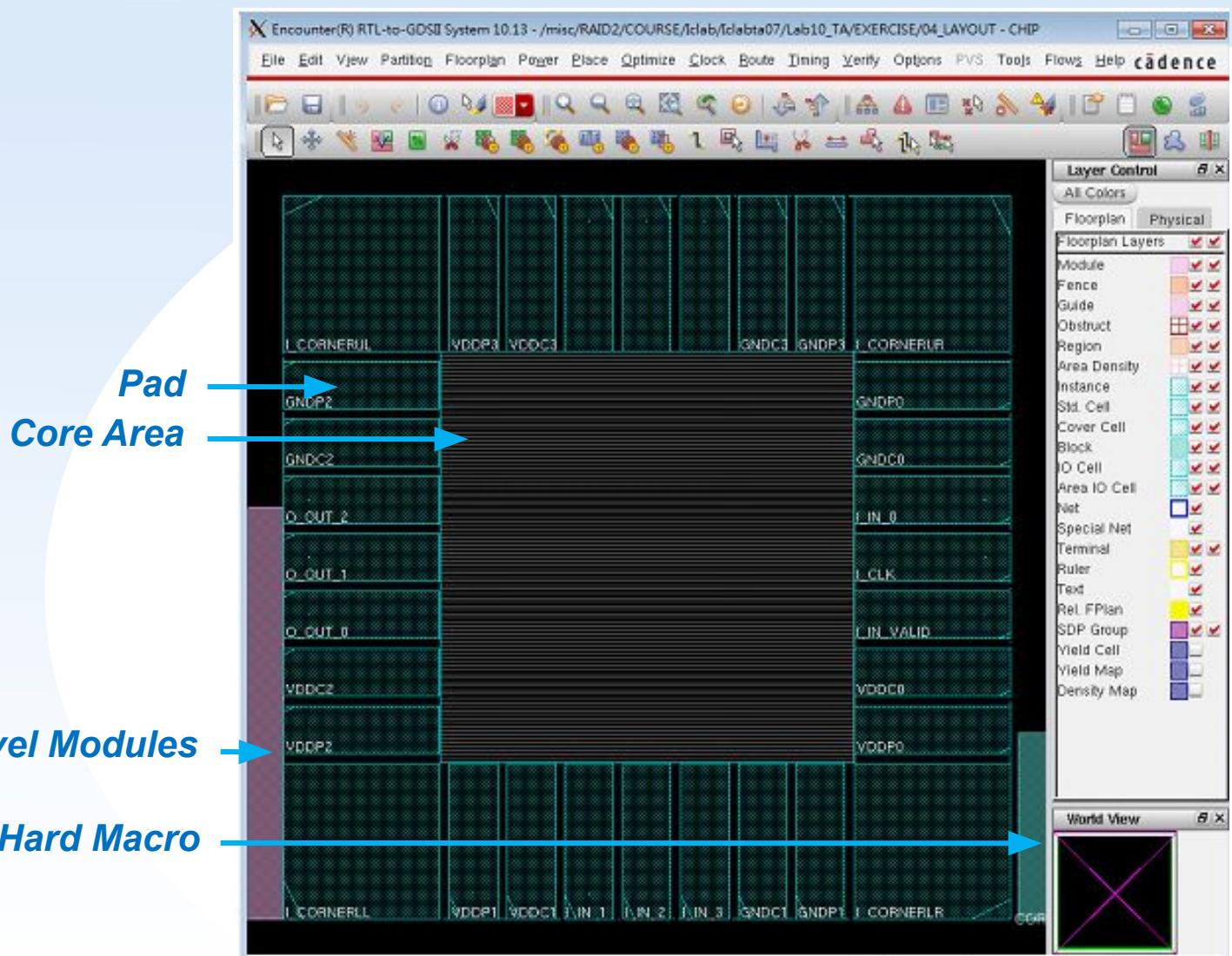
view2

view3

view4



2. Importing Design - Result



3. Floor Planning – Intro.

✓ Calculating core size, width and height

- When calculating core size of standard cells, the core utilization must be decided first. Usually the core utilization is higher than **70%**
- The core size can be calculated as follows

$$\text{Core Size of Standard Cell} = \frac{\text{standard cell area}}{\text{core utilization}}$$

- The recommended core shape is a square, i.e. Core Aspect Ratio = 1. Hence the width and height can be calculated as

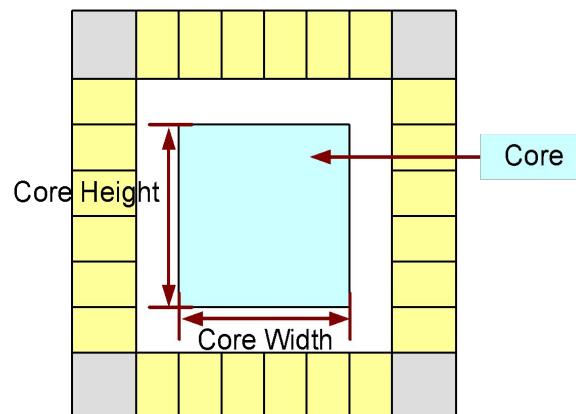
$$\text{Width} = \text{Height} = \sqrt{\text{Core Size of Standard Cell}}$$

- Note: because stripes and macros will be added, width and height are usually set larger than the value calculated above
- E.g.

- Standard cell area = 2,000,000
- Core utilization demanded = 85%
- No macros

$$\text{Core Size of Standard Cell} = \frac{2,000,000}{0.85} = 2,352,941$$

$$\text{Width} = \text{Height} = \sqrt{2,352,941} = 1534$$



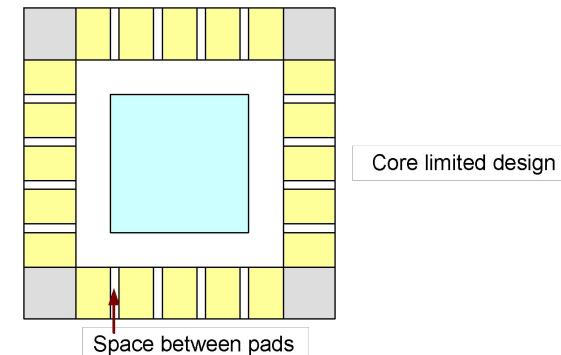
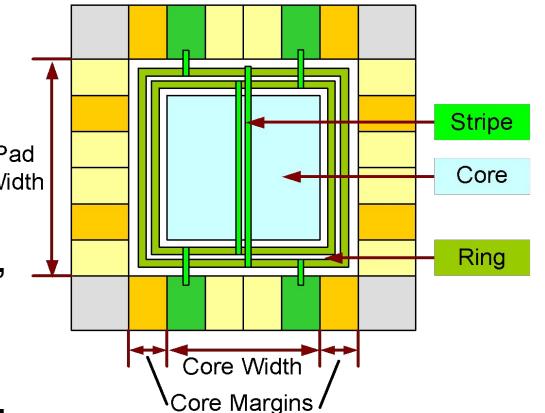
3. Floor Planning

✓ Core margins

- When setup the floorplan, remember to leave enough space for Power/Ground (P/G) ring
 - Note: the width needed for P/G ring will be discussed in power planning

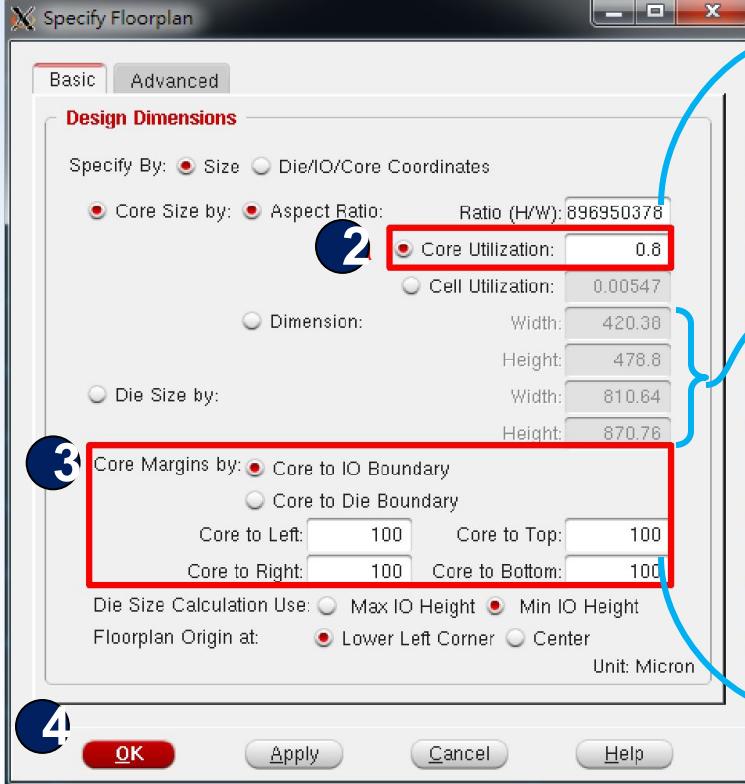
✓ Core limited design or Pad limited design

- Die size determination
 - When pad width > (core width + core margin), die size is decided by pads.
And it is called pad limited design
 - When pad width < (core width + core margin), die size is decided by core.
And it is called core limited design
- Adding pad filler
 - Provide power to pad
 - There should be no spacing between pads.
Pad filler is necessary for **core limited design**



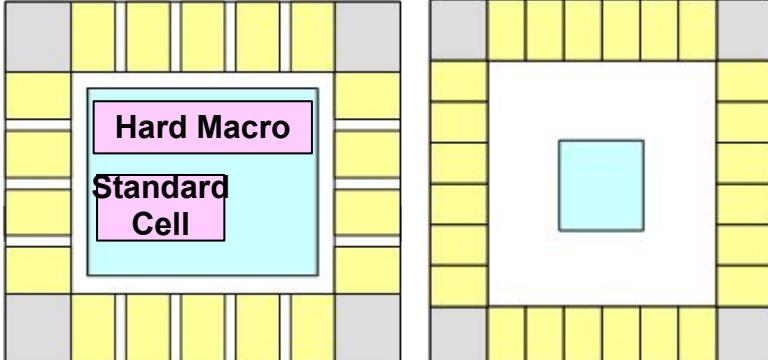
3. Floor Planning

1 **Specify Floorplan**



Some space are remained for routing.

Set a demanded value of width and height, when Hard Macro takes most of the place.



- Core Limited Design
- Pad Limited Design

Some space are remained for power rings.
(design dependent)



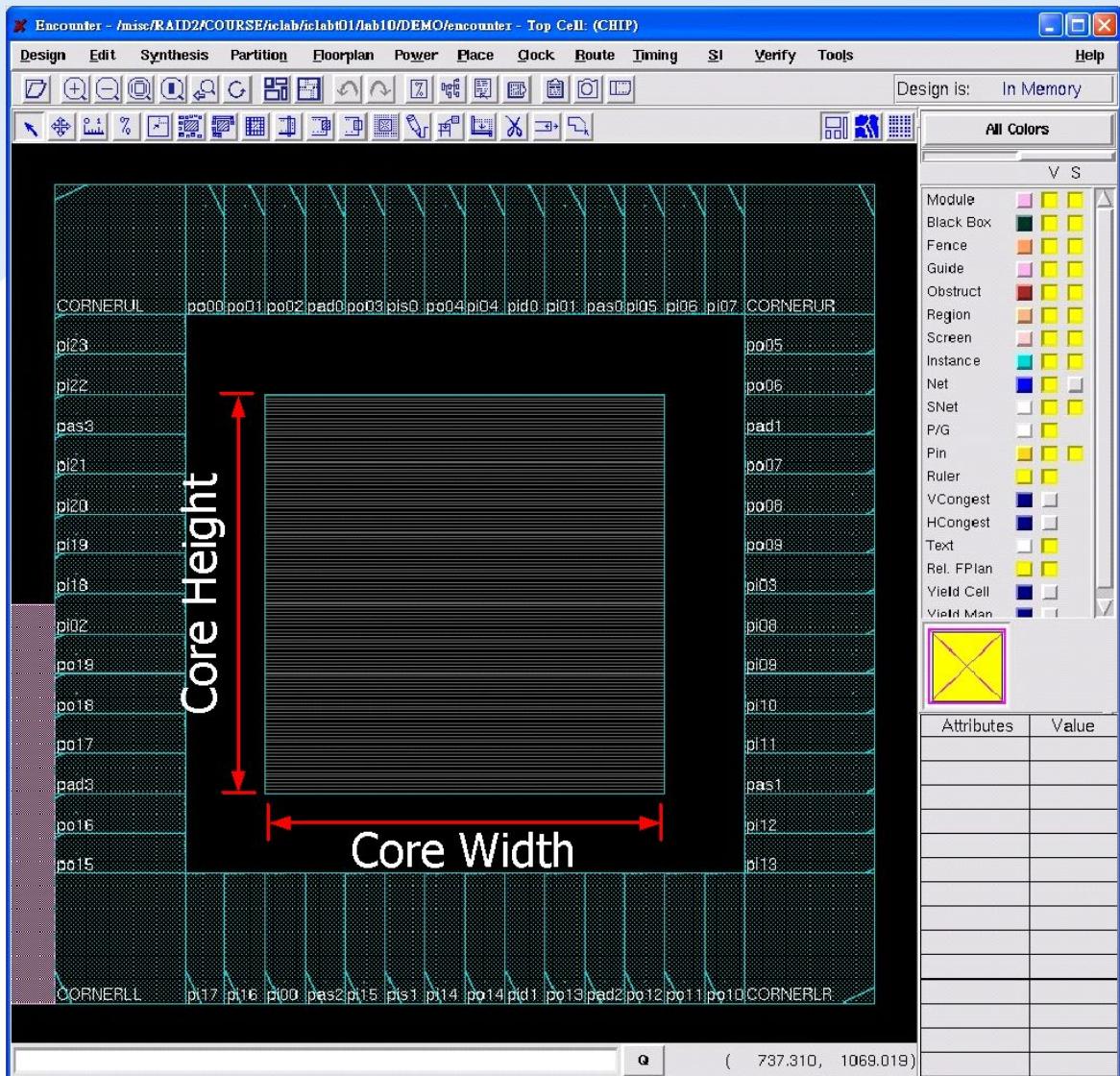
Text Command (Specify core width, height and core margin)

- innovus #> [floorPlan -s coreW coreH coreToLeft coreToBottomcoreToRight coreToTop](#)



3. Floor Planning – Result

✓ Result

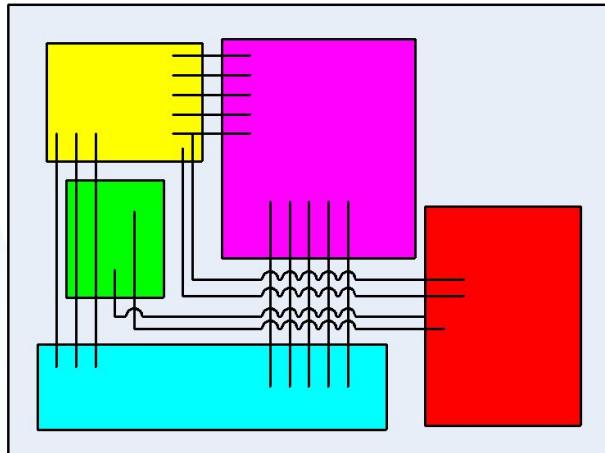


3. Floor Planning – Intro.

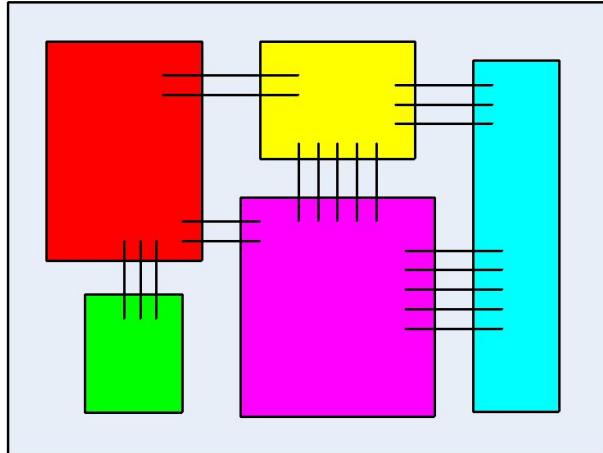


Floorplan Purposes

- Develop early physical layout to ensure design objective can be achieved
 - Minimum area for low cost
 - Minimum congestion for design routable
 - Estimate parasitic for delay calculation
 - Analysis power for reliability
- Gain early visibility into implementation issues
- Different floorplan, different performance



Bad floorplan



Good floorplan

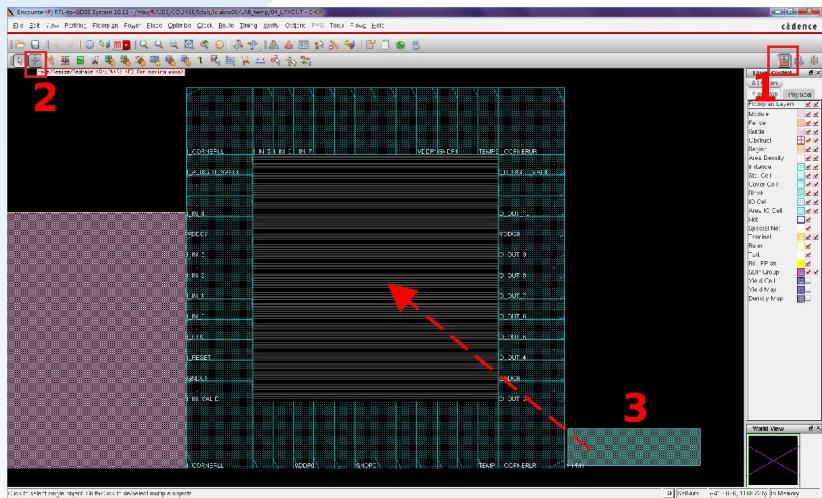


3. Floor Planning with Hard Macro



Macro Floorplanning

- Move macro blocks to proper position
- Place macro around the corner to improve routability



Block place issue:

- power issue
- noise issue
- route issue

In practice, if there are many hard macro, try to use automatic floorplan and the following tools

- Floorplan → Relative Floorplan → Define array constraint
- Floorplan → Placement toolbox

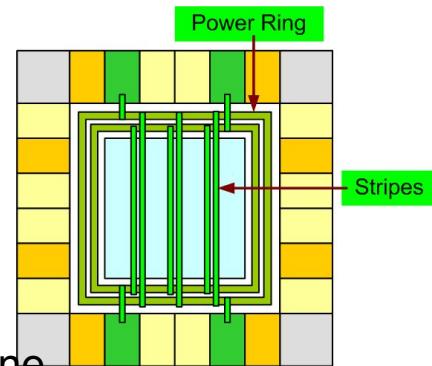


4. Power Planning – Intro.



Power issue

- Metal migration (also known as electro-migration)
 - Under **high currents**, electron collisions with metal grains cause the metal to move. The metal wire may be **open circuit** or **short circuit**.
 - Prevention: sizing power supply lines to ensure that the chip does not fail
 - Experience: make current density of power ring < **1mA/ μ m**
- IR drop
 - IR drop is the problem of voltage drop of the power and ground due to high current flowing through the power-ground resistive network
 - When there are excessive voltage drops in the power network or voltage rises in the ground network, the device will run at **slower speed**
 - IR drop can cause the chip to fail due to
 - ◆ Performance (circuit running slower than specification)
 - ◆ Functionality problem (setup or hold violations)
 - ◆ Unreliable operation (less noise margin)
 - ◆ Power consumption (leakage power)
 - ◆ Latch up
 - Prevention: adding stripes to avoid IR drop on cell's power line



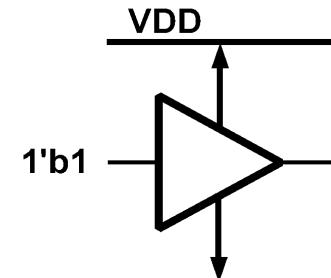
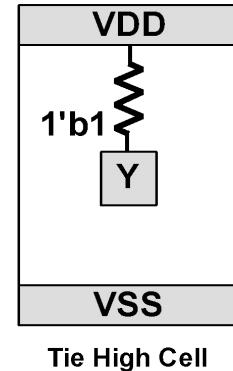
4.1. Power Planning - Connect/Define Global Nets

import
floorplan
powerplan
placement
CTS
routing

4.1. Connect/Define Global Nets

– Complete the form

- Add power nets and pins connection list
- Add ground nets and pins connection list
- Tie High: Connect all the 1'b1 to VDD
- Tie Low: Connect all the 1'b0 to GND



- After Applying, Only warnings of IO pads and padfiller are allowed
- Text Command

- innovus #> clearGlobalNets
- innovus #> globalNetConnect VDD -type pgpin -pin VDD -inst *
- innovus #> globalNetConnect VDD -type net -net VDD
- innovus #> globalNetConnect VDD -type tiehi -inst *
- innovus #> globalNetConnect GND -type pgpin -pin GND -inst *
- innovus #> globalNetConnect GND -type net -net GND
- innovus #> globalNetConnect GND -type tielo -inst *
- innovus #> globalNetConnect GND -type pgpin -pin VSS -inst *



4.1. Power Planning - Connect/Define Global Nets

import
floorplan
powerplan
placement
CTS
routing

Design Edit Synthesis Partition Floorplan **Power** Place Clock Route Timing SI Verify Tools

1

Connect Global Nets...

Create 6 global nets connection list

2 Connect ◆ Pins: VDD

To Global Net: VDD

3 Connect ◆ Net Basename: VDD

To Global Net: VDD

4 Connect ◆ Tie High

To Global Net: VDD

5 Connect ◆ Pins: GND

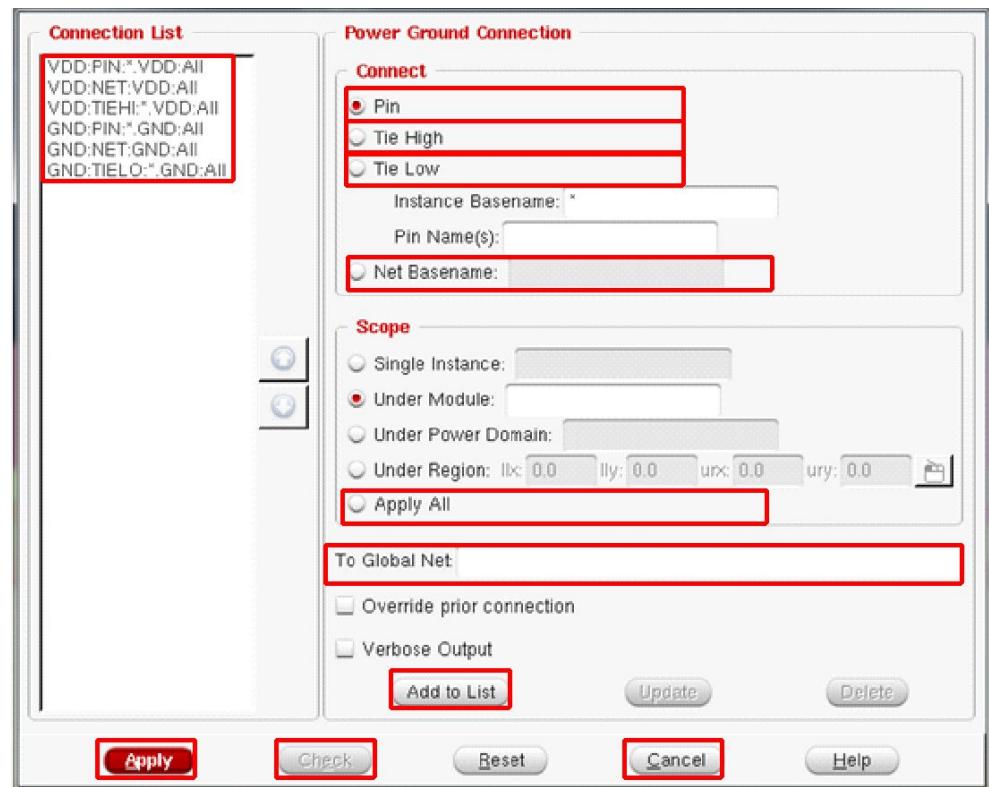
To Global Net: GND

6 Connect ◆ Net Basename: GND

To Global Net: GND

7 Connect ◆ Tie Low

To Global Net: GND



4. Power Planning – RING

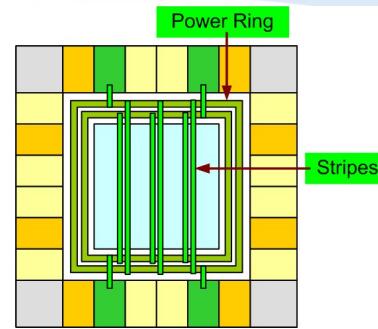
✓ Calculate power/ground ring width

– Experience

- Gate count = 70 k
- 4000 Flip-Flops
- 80% FF with dynamic gated clock
- Current needed = **0.2mA/MHz**
 - ◆ Note: the value should multiply with **1.8~2** for no gated design

– Example:

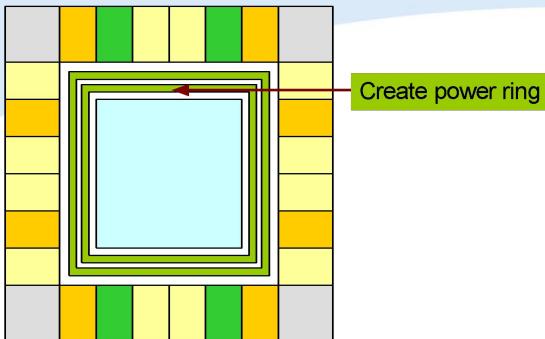
- Gate count = 200 k
- No gated clock
- Clock frequency = 20 MHz
- Current needed = $(200k/70k) * 0.2 \text{ mA/MHz} * 20\text{MHz} * 2 = 22.86 \text{ mA}$
- Current density < 1mA/ μm
- The Width of P/G Ring > 22.86 μm
- In order to avoid the slot rule of wide metal, the largest width is 20 μm (process dependent)
- Use two set P/G ring for this case



4.2. Power Planning - Core Power Ring

import
floorplan
powerplan
placement
CTS
routing

✓ 4.2. Core Power Ring



- Interleaving and Wire group

	Without Wire group	With Wire group
Without Interleaving	<p>Without wire groups</p> <p>VDD VDD GND GND</p>	<p>With wire groups</p> <p>VDD VDD GND GND</p>
Interleaving	<p>Without wire groups</p> <p>VDD GND VDD GND</p>	<p>With wire groups</p> <p>VDD GND VDD GND</p>



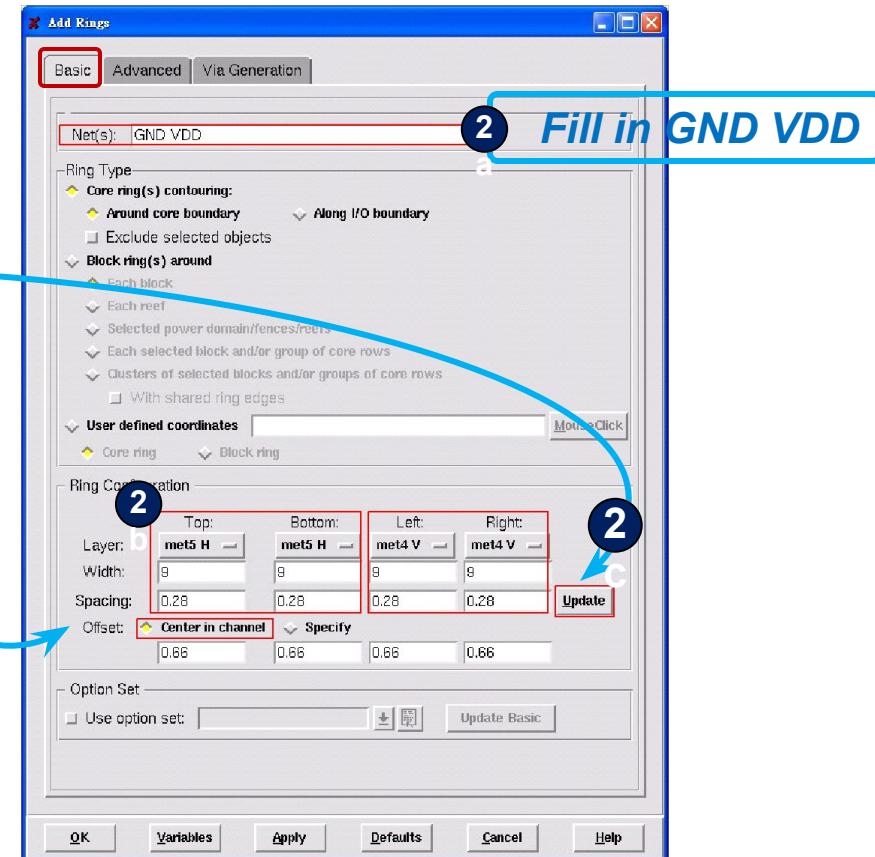
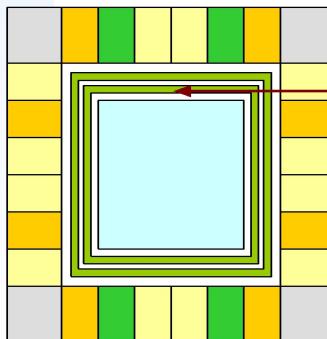
4.2. Power Planning - Core Power Ring

import
floorplan
powerplan
placement
CTS
routing



– In Basic tab

Field	Fill In
Top and Bottom Layer	Metal 3
Left and Right Layer	Metal 4
Width	All 9 (or the demanded value)
	Click Update
Offset	◆ Center in channel



4.2. Power Planning - Core Power Ring

- In Advanced tab

- You can create one more set of power ring

Field	Fill In
Wire Group	<input checked="" type="checkbox"/> Use wire group <input checked="" type="checkbox"/> Use wire group
Number of bits	The demanded value

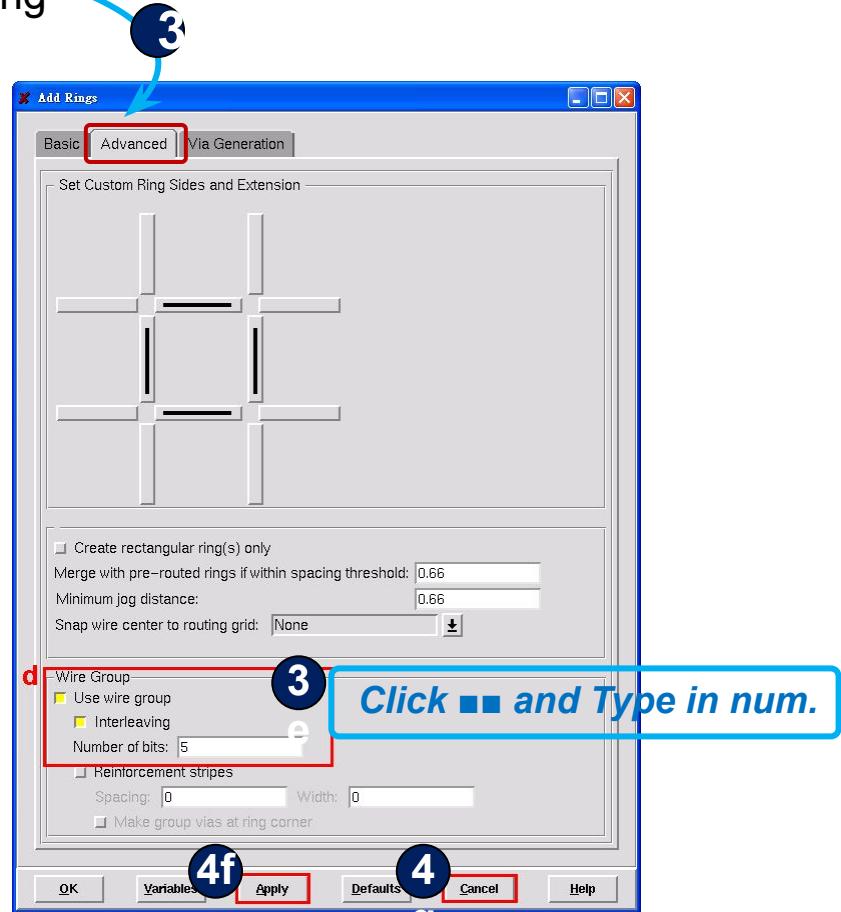
Note:

you can choose the option “Interleaving” to observe the difference in power ring created

- Click Apply

- Tips

- Click Apply then you can undo this step
- Click OK then undo is not allowed



4.3. Power Planning

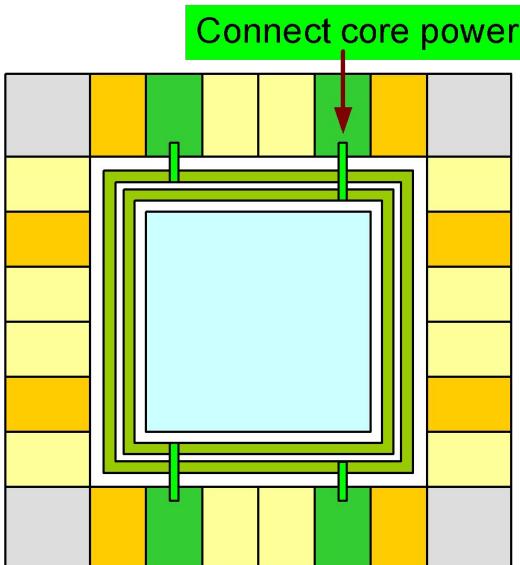
- Core Power Pin

import
floorplan
powerplan
placement
CTS
routing



4.3. Core Power Pin

- Connect from core power pads to power rings



- Note:
If global nets aren't correctly connected, core power pins will have trouble in generating.



4.3. Power Planning - Core Power Pin



✓ Connect core power pin

- Complete the form and click [Apply](#)

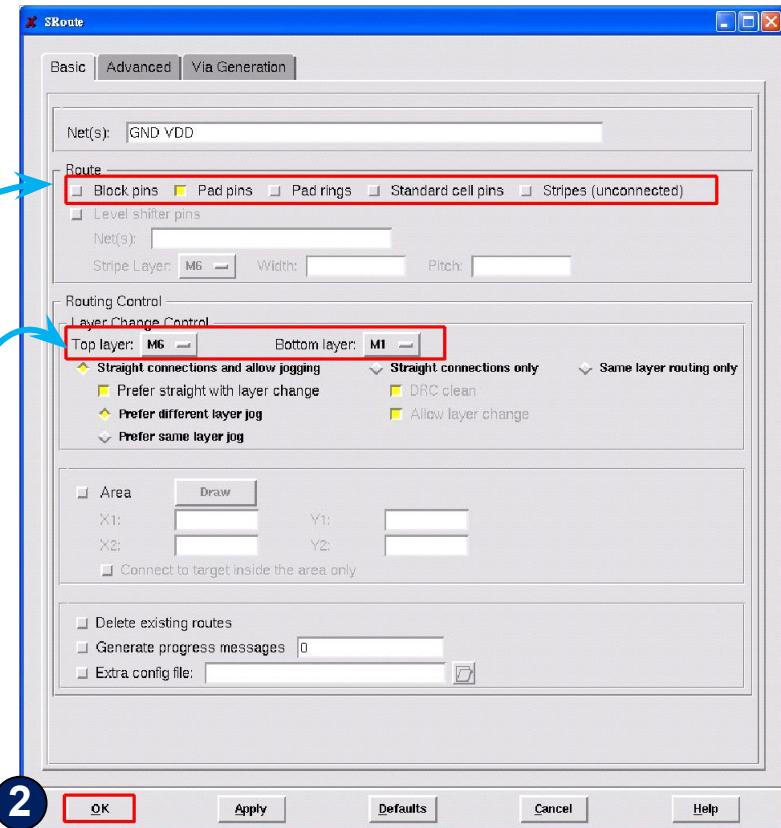
Field	Fill In
Block pins	<input type="checkbox"/> (off)
Pad pins	<input checked="" type="checkbox"/> (on) 2a
Pad rings	<input type="checkbox"/> (off)
Standard cell pins	<input type="checkbox"/> (off)
Stripes (unconnected)	<input type="checkbox"/> (off)

Choose M2 for Bottom layer (Suggestion)

Note:

Make sure that core power pads are connected to the power ring.

Adjust the top and bottom layer if needed.



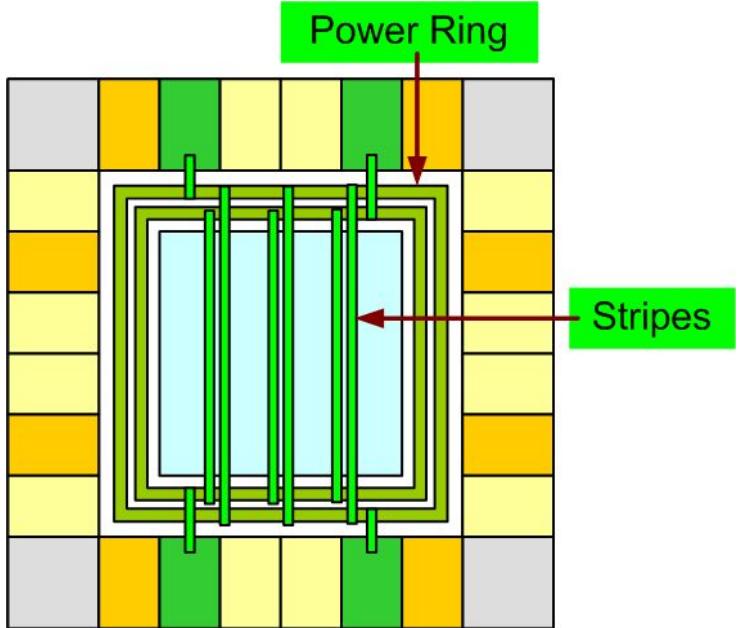
2



4. Power Planning – Stripes

✓ Calculate stripe set

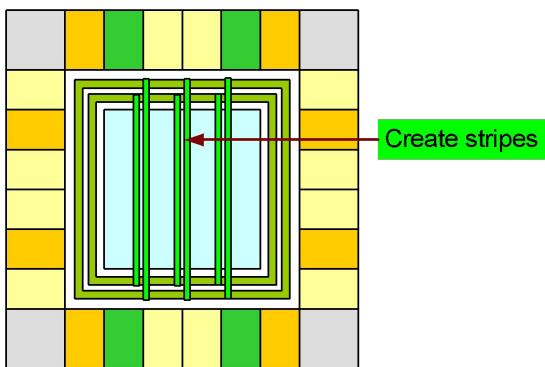
- Experience
 - Add one stripe set per 100 μm
- Example
 - Core width = height = 1600
 - Stripe set added = 15



4.4. Power Planning - Stripes



- Complete the form in Basic tab and click Apply

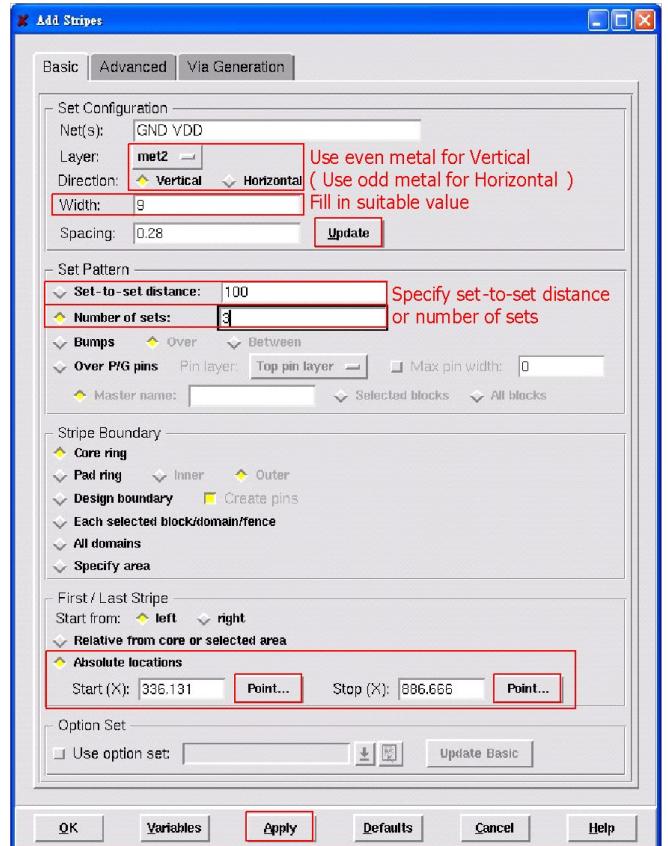


Note: after click point

Start (X): click the location of first stripe in design display area

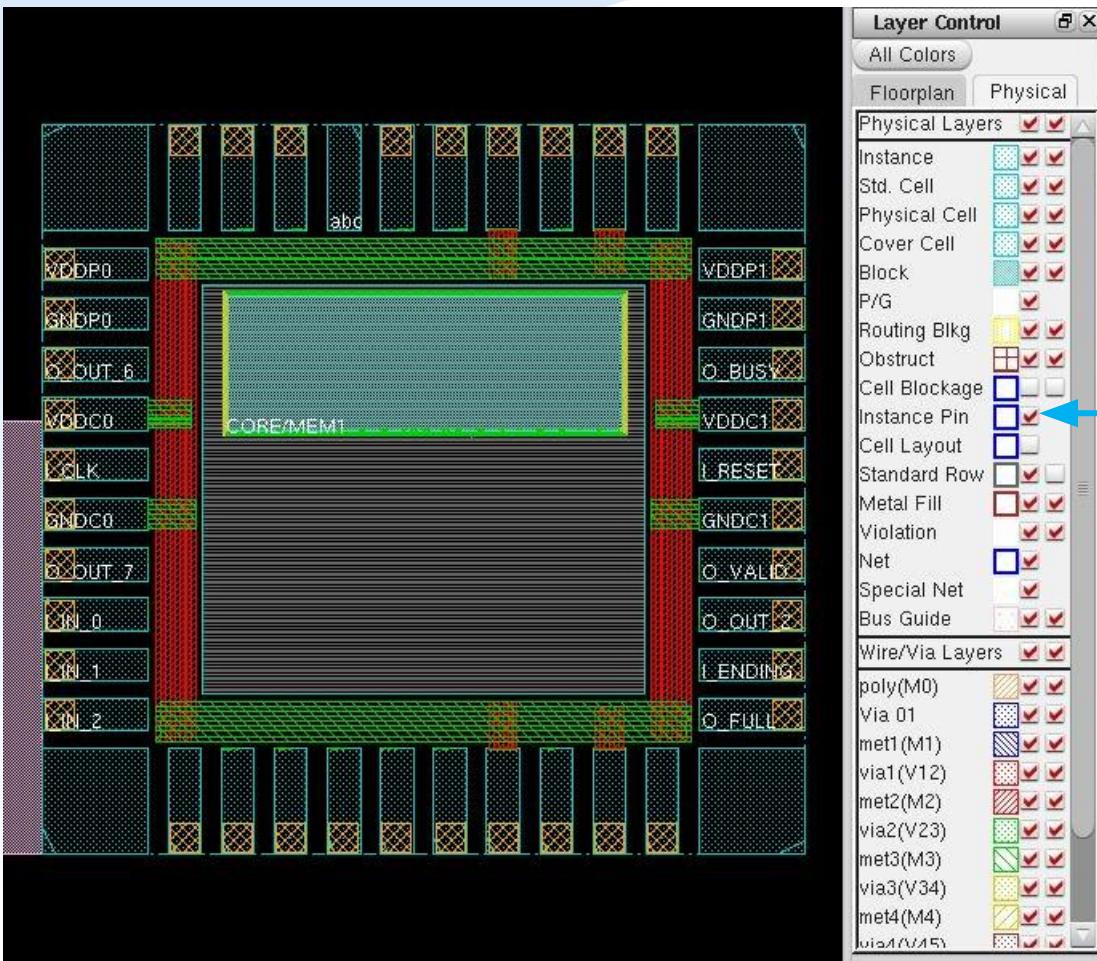
Stop (X): click the location of last stripe in design display area

Note: Specify Start (Y) and Stop (Y) for horizontal stripes



4. Power Planning

– Block Ring for Hard Macro



*To visualize the
block ring of
SRAM*

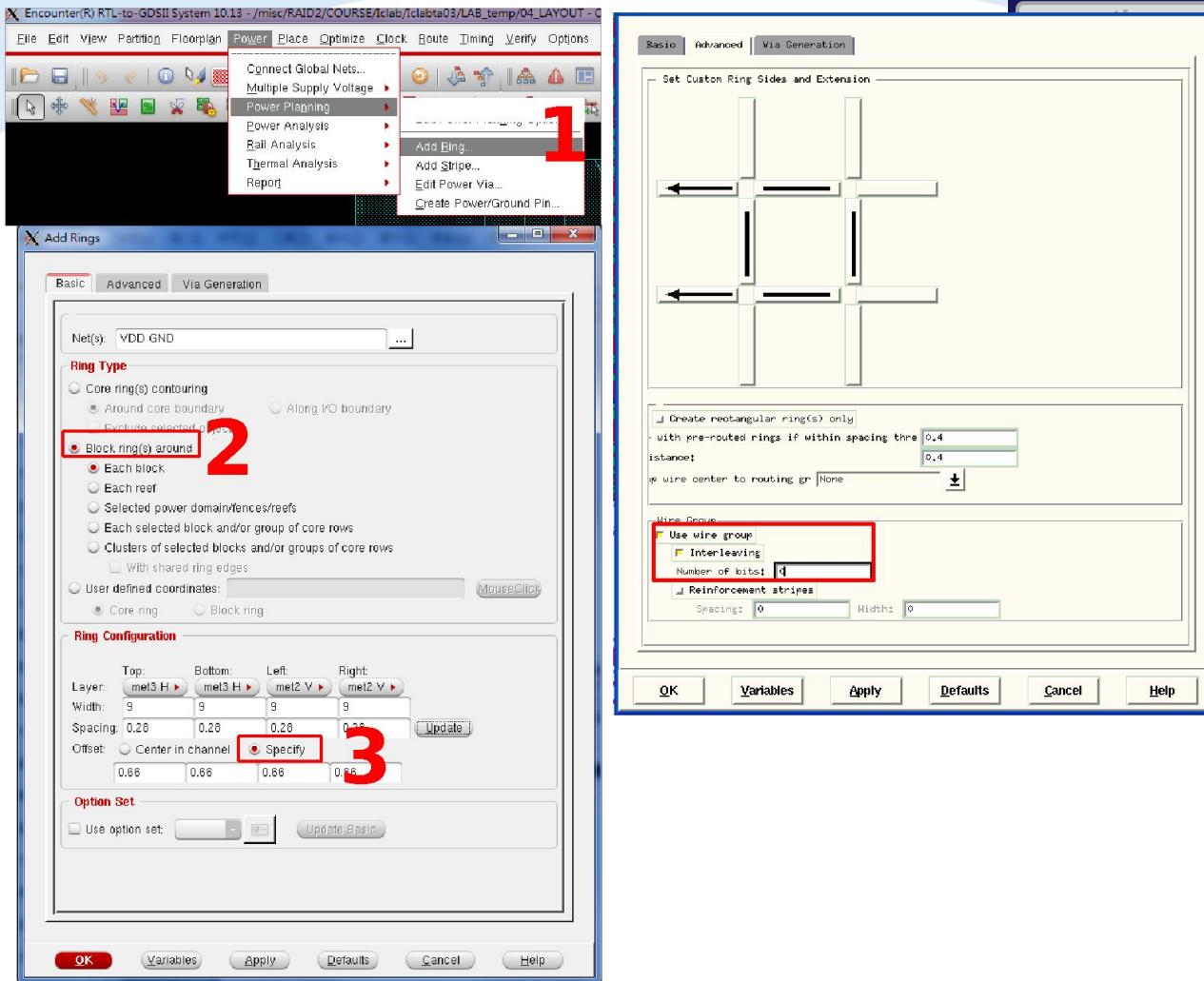
4. Power Planning

– Block Ring for Hard Macro



Add block ring

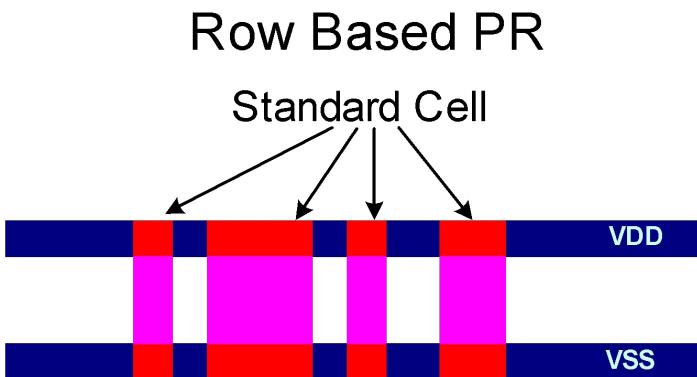
- Specify nets
- Block ring around
- Specify metals
- Width, offset
- Extension
- Wiregroups



4.5. Standard Cell Power Connection

import
floorplan
powerplan
placement
CTS
routing

- ✓ Connect standard cell power line



- Text Command
- ```
innovus #> sroute -noBlockPins -noPadRings -noPadPins -noStripes -jogControl
{ preferWithChanges differentLayer }
```



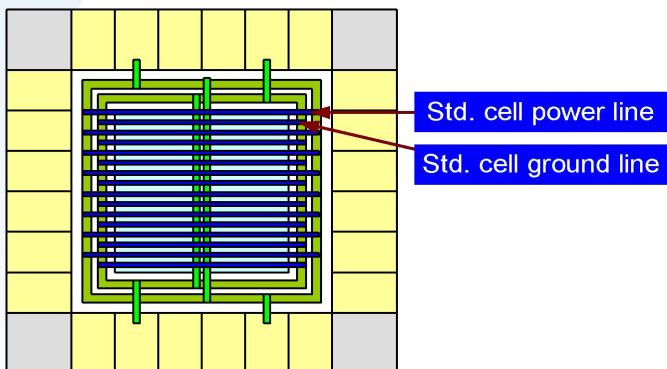
# 4.5. Standard Cell Power Connection



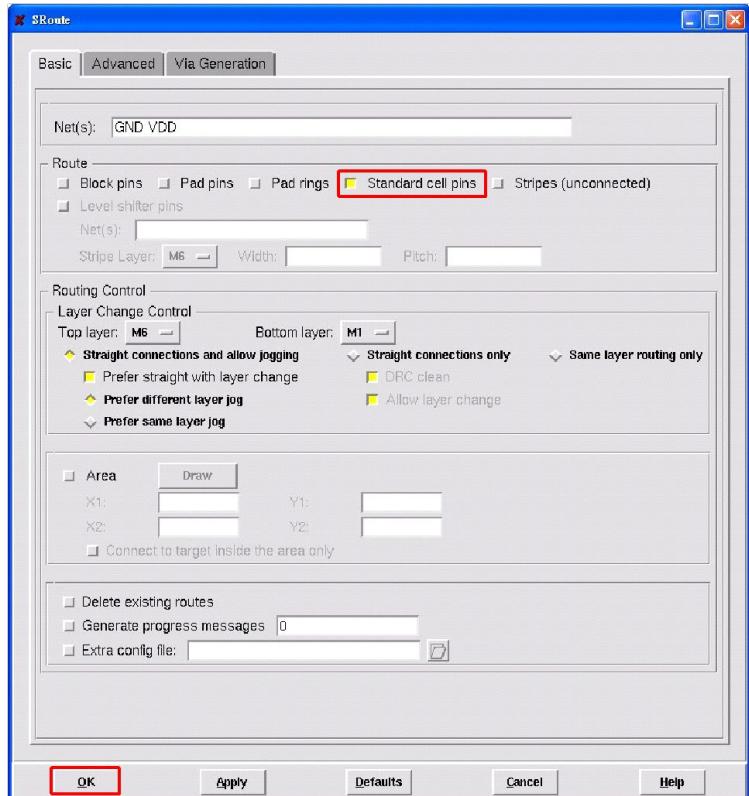
## Connect standard cell power line

Design Edit Synthesis Partition Floorplan Power Place Clock Route Timing SI Verify Tools

| Field                 | Fill In                                  |
|-----------------------|------------------------------------------|
| Block pins            | <input type="checkbox"/> (off)           |
| Pad pins              | <input type="checkbox"/> (off)           |
| Pad rings             | <input type="checkbox"/> (off)           |
| Standard cell pins    | <input checked="" type="checkbox"/> (on) |
| Stripes (unconnected) | <input type="checkbox"/> (off)           |



Special Route ...



### Text Command

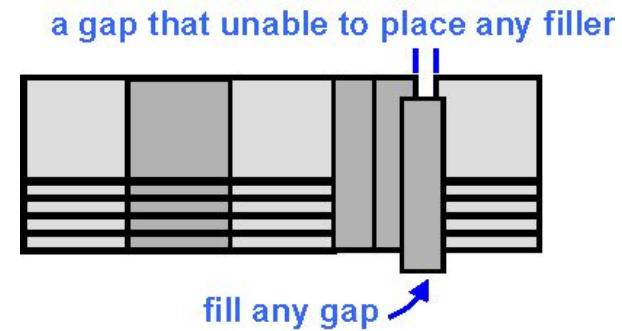
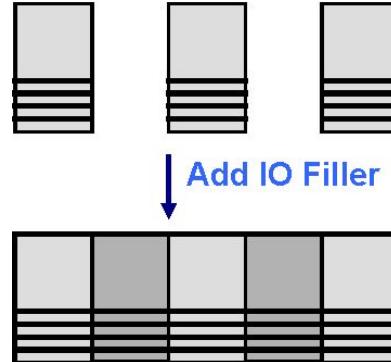
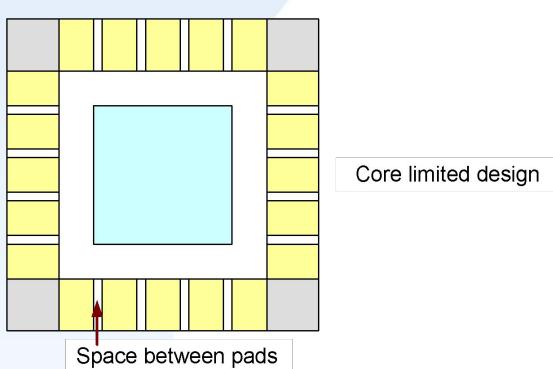
```
innovus #> sroute -noBlockPins -noPadRings -noPad Pins -noStripes -jogControl
{ preferWithChanges differentLayer }
```



# 4.6 Add PAD Filler

## ✓ Adding pad filler

- There should be no spacing between pads.  
Therefore adding pad filler is necessary for core limited design
- PAD filler must be added before detail route, or there may have some DRC/LVS violations after PAD filler insertion



### - Text Command

From wider  
fillers to  
narrower ones

- innovus #> addIoFiller -cell PFILL -prefix IOFILLER
- innovus #> addIoFiller -cell PFILL\_9 -prefix IOFILLER
- innovus #> addIoFiller -cell PFILL\_1 -prefix IOFILLER
- innovus #> addIoFiller -cell PFILL\_01 -prefix IOFILLER -fillAnyGap

*Only the smallest filler  
can use  
-fillAnyGap option*



# 5. Standard Cell Placement

## ✓ Specify placement blockage



- Choose the metal layers of stripe.  
Then there will be no cell placed under stripe
- Text Command  
`innovus #> setPrerouteAsObs {2 3}`



- Add placement blockage for macros
- Text Command  
`innovus #> createObstruct /lx /ly urx ury`

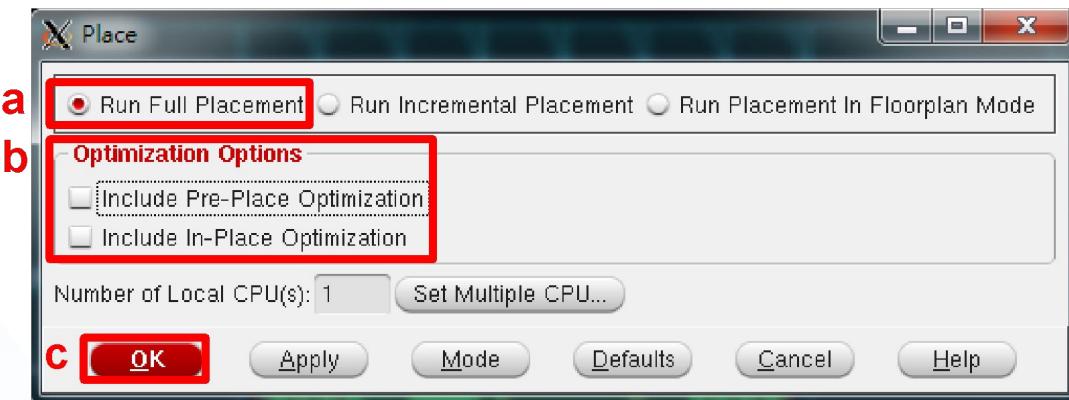


# 5. Standard Cell Placement

## ✓ Place standard cells

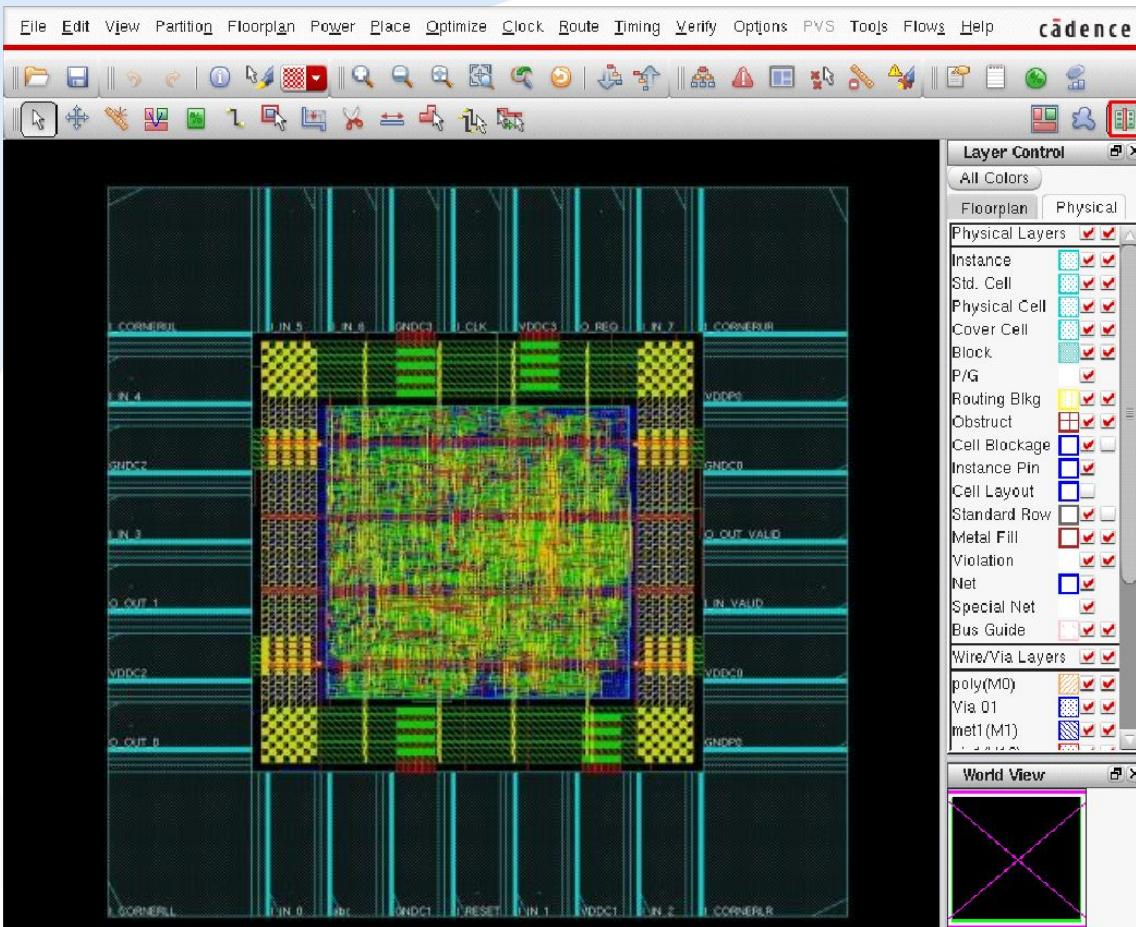


- Run Full Placement
- Optimization Options
  - Include Pre-Place Optimization
  - Include In-Place Optimization
- Click OK button



# 5. Standard Cell Placement – Result

import  
floorplan  
powerplan  
**placement**  
CTS  
routing



# 6. Pre-CTS Timing Analysis

## ✓ To check timing

- innovus #> timeDesign –preCTS

## ✓ Operations

- Trial route
- RC extraction
- Timing analysis
- Generates reports are saved in ./timingReports/ , including
  - \_preCTS.cap
  - \_preCTS.fanout
  - \_preCTS.tran
  - \_preCTS\_all.tarpt

## ✓ Reports

- Congestion distribution
- Timing Summary
- If the timing is met, pre-CTS optimization can be skipped

| Congestion distribution: |      |       |      |        |
|--------------------------|------|-------|------|--------|
| Remain                   | cntH | cntV  |      |        |
| -2:                      | 5    | 0.02% | 0    | 0.00%  |
| -1:                      | 10   | 0.03% | 0    | 0.00%  |
| 1:                       | 10   | 0.03% | 21   | 0.06%  |
| 2:                       | 10   | 0.03% | 58   | 0.17%  |
| 3:                       | 0    | 0.00% | 88   | 0.25%  |
| 4:                       | 33   | 0.10% | 379  | 1.09%  |
| 5:                       | 0    | 0.00% | 4227 | 12.18% |

| Setup mode       | all    | reg2reg | in2reg | reg2out | in2out | clkgate |
|------------------|--------|---------|--------|---------|--------|---------|
| WNS (ns):        | -0.744 | N/A     | -0.744 | N/A     | N/A    | N/A     |
| TNS (ns):        | -7.291 | N/A     | -7.291 | N/A     | N/A    | N/A     |
| Violating Paths: | 16     | N/A     | 16     | N/A     | N/A    | N/A     |
| All Paths:       | 40     | N/A     | 40     | N/A     | N/A    | N/A     |

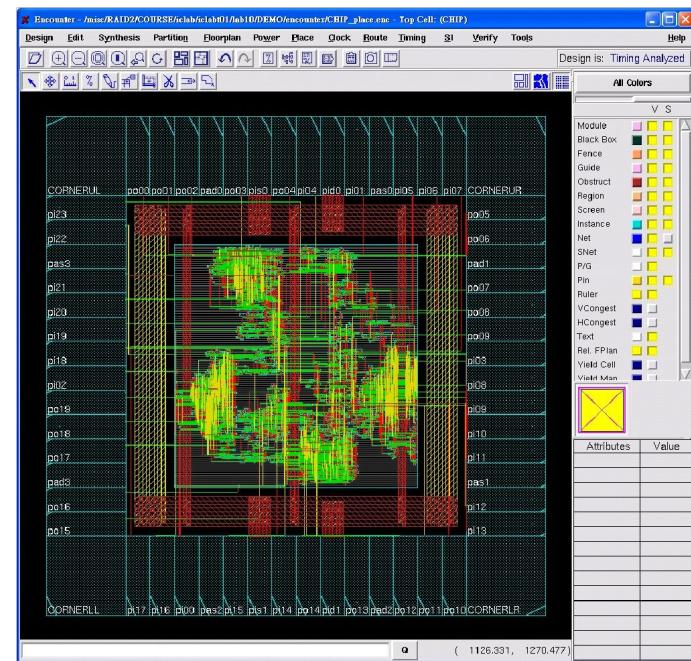
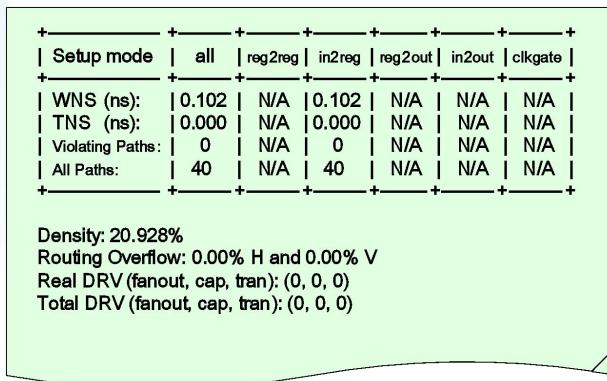
Negative slack should be fixed  
Density: 19.85%  
Routing Overflow: 0.00% H and 0.00% V  
Real DRV (fanout, cap, tran): (0, 1, 0)  
Total DRV (fanout, cap, tran): (0, 1, 0)

Total DRV = real DRV + clock network DRV



# 6. Pre-CTS Optimization

- ✓ To optimize timing placed design for the first time with ideal clocks
  - innovus #> optDesign –preCTS
- ✓ To further optimize a design after above command execution
  - innovus #> optDesign –preCTS –incr
- ✓ Operation
  - Repair Setup slack, Setup times
  - Repair Design rule violations (DRVs) and remaining DRVs



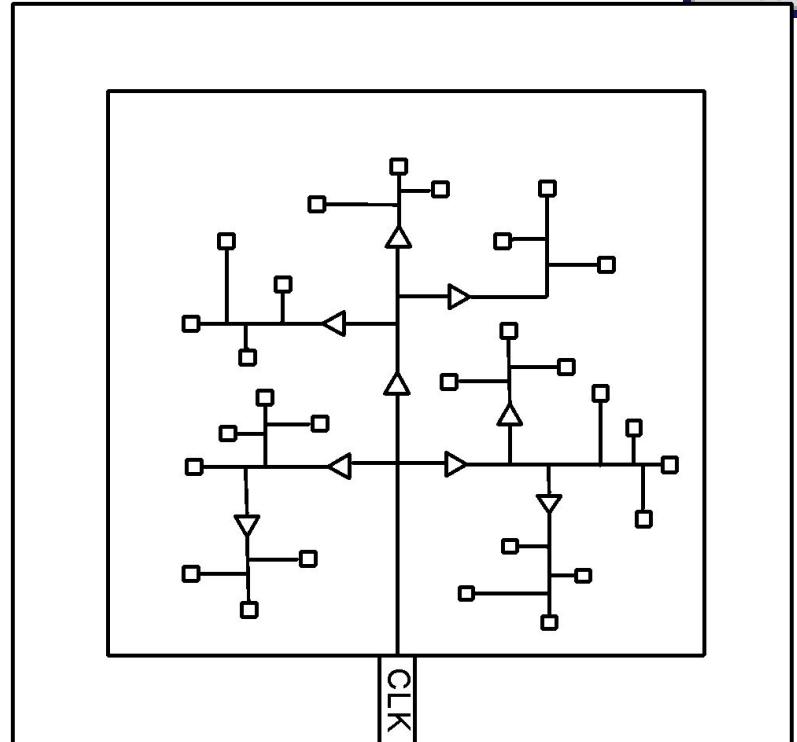
# 7. Clock Tree Synthesis - Clock Problem

import  
floorplan  
powerplan  
placement  
**CTS**  
routing



## Clock problem

- ⌘ Heavy clock net loading
- ⌘ Long clock insertion delay
- ⌘ Clock skew
- ⌘ Skew across clocks
- ⌘ Clock to signal coupling effect
- ⌘ Clock is power hungry
- ⌘ Electromigration on clock net



**Clock is one of the most important treasure in a chip, do not take it as other use.**



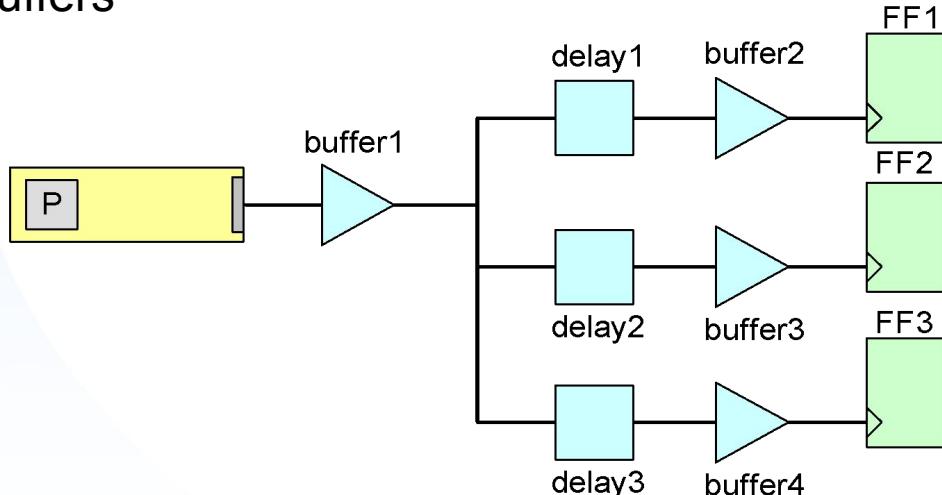
# 7. Clock Tree Synthesis

✓ **The goal of clock tree synthesis includes**

- Creating clock tree spec file
- Building a buffer distribution network
- Routing clock nets using CTS-NanoRoute

✓ **In automatic CTS mode, Innovus will do the following things**

- Build the clock buffer tree according to the clock tree specification file
- Balance the clock phase delay with appropriately sized, inserted clock buffers



# 8. Post-CTS Timing Analysis

## - Text Command

### ✓ To check setup time

- innovus #> `timeDesign –postCTS`

### ✓ To check hold time

- innovus #> `timeDesign –postCTS –hold`

### ✓ Reports

- Generated timing reports are saved in `./timingReports/`, including  
`_postCTS.cap` , `_postCTS.fanout` , `_postCTS.tran` , `_postCTS_all.tarpt`"

### ✓ To correct setup violations and design rule violation

- innovus #> `optDesign –postCTS`

### ✓ To correct hold violations

- innovus #> `optDesign –postCTS –hold`

### ✓ Operation

- Repair Setup slack, Setup times
- Repair Design rule violations (DRVs)

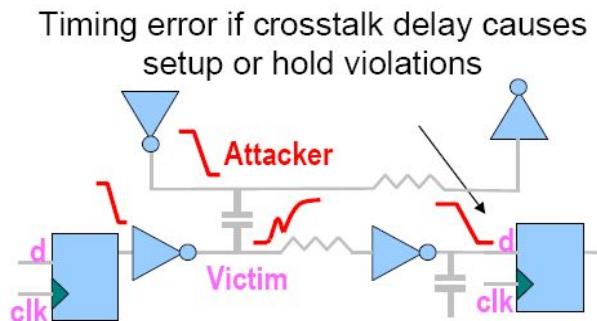
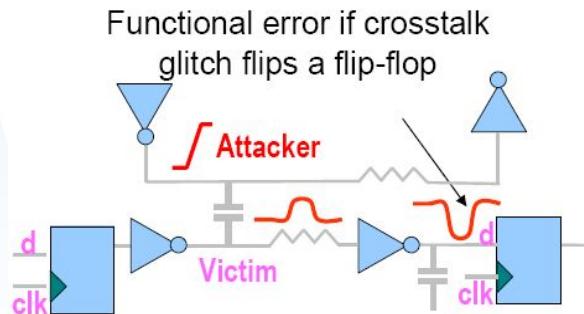




# 10. NanoRoute – Intro.

## ✓ Signal Integrity (SI) Issue

- Crosstalk
- Charge sharing
- Supply noise
- Leakage
- Propagated noise
- Overshoot
- Under shoot



## ✓ SI closure

- Occur when a design is free from SI induced functional glitch and timing failure





# Appendix: Crosstalk

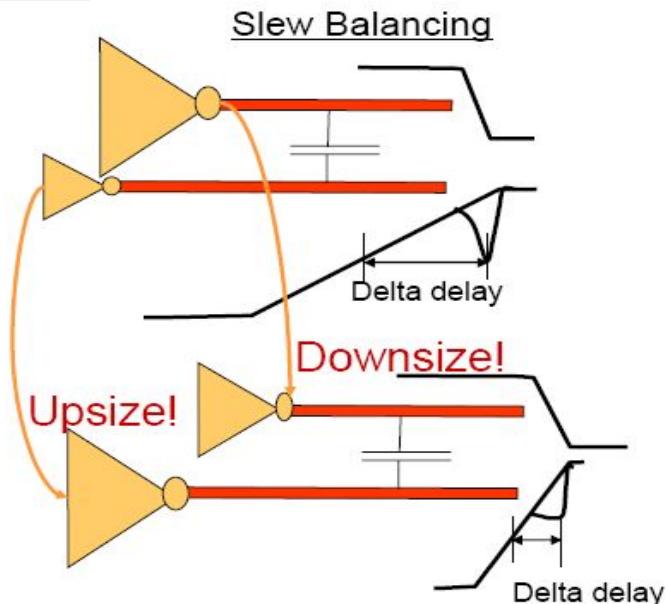
## ✓ Crosstalk prevention – Placement solution

- Insert buffer in lines
- Upsize driver
- Congestion optimization

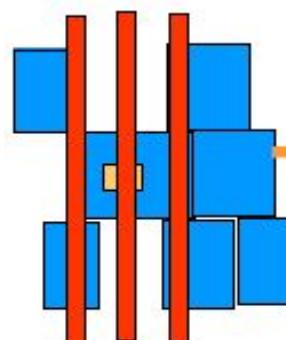
## ✓ Crosstalk prevention – Routing solution

- Limit length of parallel nets
- Wider routing grid
- Shield special nets

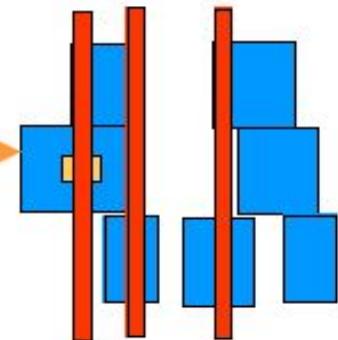
• Reduce coupling capacitance



Original placement



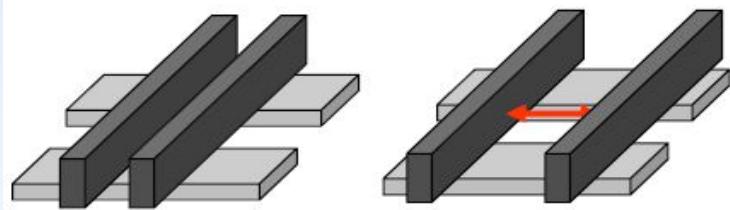
Optimized placement



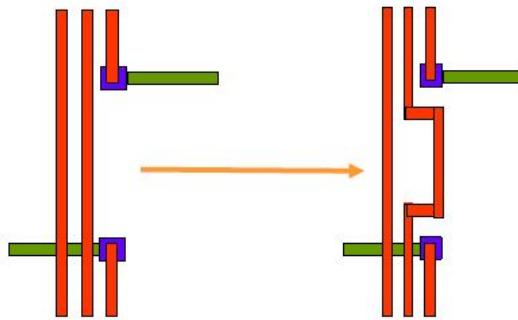
# 10. NanoRoute – Intro.

## ✓ SI prevention (cont.)

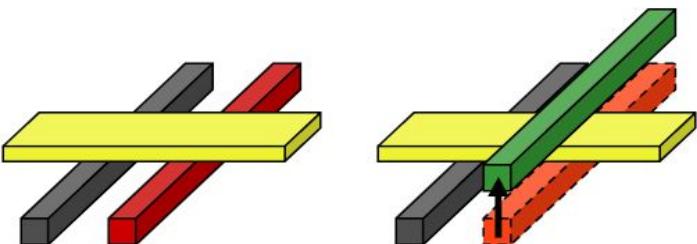
- Routing-based SI prevention
  - Wiring Spacing



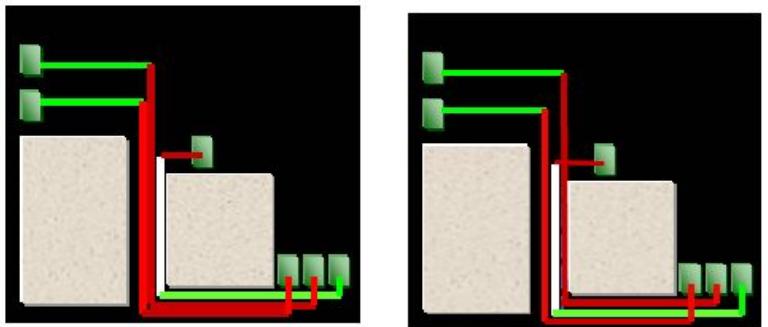
- Parallel Wires Reducing



- Layer Switching



- Net Re-ordering



# 10. NanoRoute – Process Antenna Effect

import  
floorplan  
powerplan  
placement  
CTS  
routing

## ✓ Antenna Effect

- In a chip manufacturing process, metal is initially deposited so it covers the entire chip
- Then, the unneeded portions of the metal are removed by etching, typically in plasma (charged particles).
- The exposed metal collect charge from plasma and form voltage potential.
- If the voltage potential across the gate oxide becomes large enough, the current can damage the gate oxide.

✓ Antenna Ratio = 
$$\frac{\text{Area of process antennas on a node}}{\text{Area of gates to the node}}$$

## ✓ Problem Repair

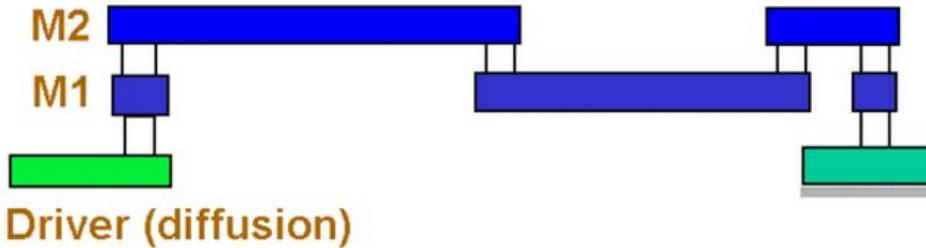
- Add jumper (change metal layer)
- Add antenna cell (diode)
- Add dummy transistor



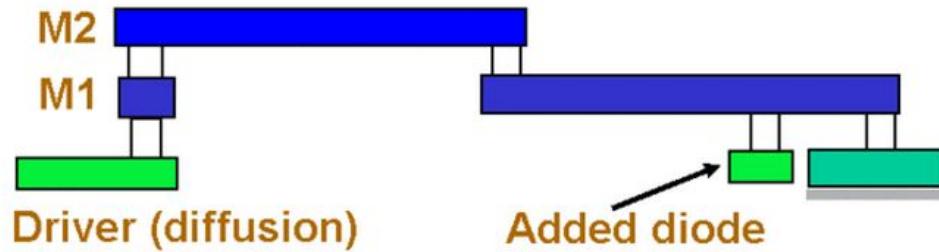
# 10. NanoRoute – Process Antenna Effect

import  
floorplan  
powerplan  
placement  
CTS  
routing

- ✓ Add jumper



- ✓ Add diode



- ✓ If after fixing antenna violation, there still exist violations
  - Larger chip area
  - Place module precisely



# 10. NanoRoute – GUI

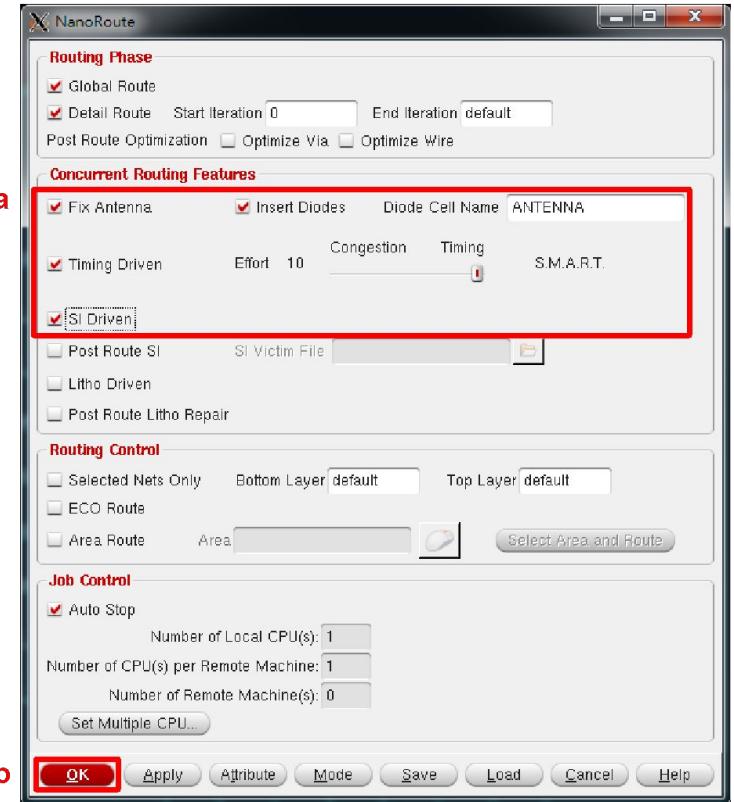


## NanoRoute

- Routing the design without creating DRC or LVS violations
- Routing the design without degrading timing or creating signal integrity violations
- Complete the form and click [OK](#)

| Field                       | Fill In                                              |
|-----------------------------|------------------------------------------------------|
| Concurrent Routing Features | ◆ Fix Antenna                                        |
|                             | ◆ Insert Diodes                                      |
|                             | Diode Cell Name <input type="text" value="ANTENNA"/> |
|                             | ◆ Timing Driven                                      |
|                             | ◆ SI Driven                                          |

Add I/O Pad Filler before Routing!!



# 11. Post-Route Timing Analysis

- ✓ **To check setup time**
  - innovus #> `timeDesign –postRoute`
- ✓ **To check hold time**
  - innovus #> `timeDesign –postRoute –hold`
- ✓ **Operation**
  - Repair Setup slack, Setup times
  - Repair Design rule violations (DRVs)  
and remaining DRVs
- ✓ **Reports**
  - Generated timing reports are saved in `./timingReports/` , including
    - `postRoute.cap`
    - `_postRoute.fanout`
    - `_postRoute.tran`
    - `_postRoute_all.tarpt`

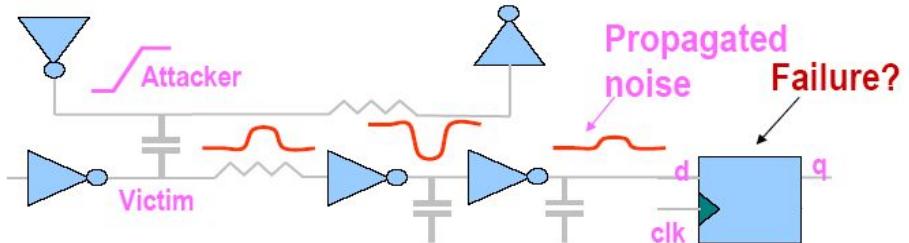


# 12. Post-Route Timing Analysis (include SI)

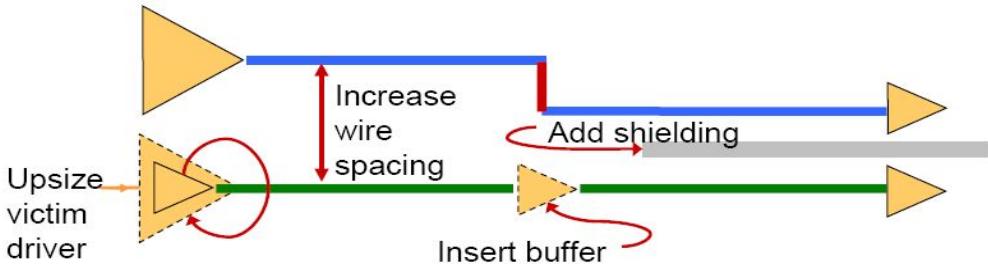
import  
floorplan  
powerplan  
placement  
CTS  
routing

## ✓ Text Command

- To do glitch noise analysis
  - innovus #> timeDesign –postRoute –si



## ✓ SI repair techniques



## ✓ Text command

- To correct setup violations and design rule violations
  - innovus #> optDesign –postRoute –si
- To correct hold violations
  - innovus #> optDesign –postRoute –si –hold

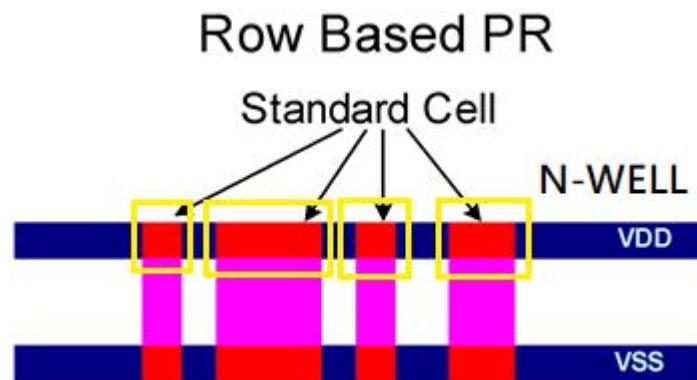


# 13. Add Filler Cells

## ✓ Purpose of adding filler cells

- Fill all the gaps between standard cell instances
- Provide decoupling capacitances to complete connections in the standard cell rows

## ✓ Metal filler inserted after routing, but before parasitic extraction and GDSII out



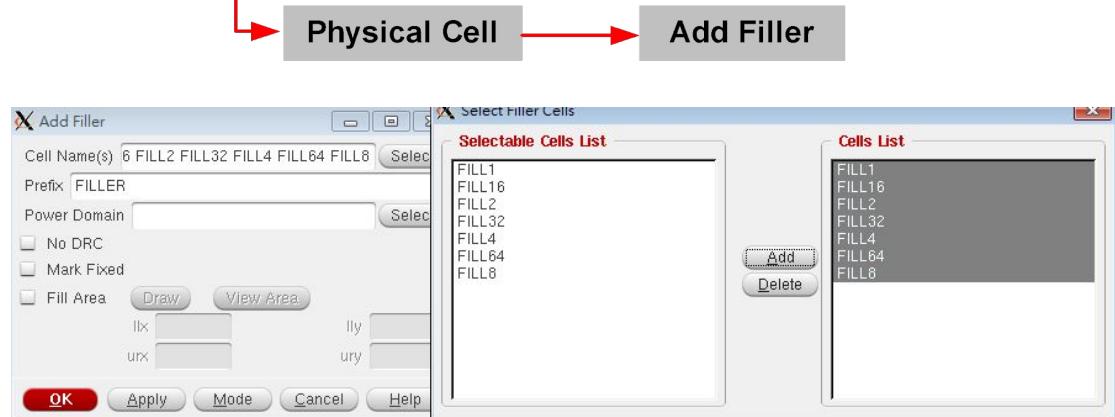
# 13. Add Filler Cells

✓ Add filler cell into the free space of core design



Why add dummy

- meet minimize metal density rule
- prevent over etching
- prevent sagging in local area
- improve yield
- reduce on chip variation



✓ Text command

- innovus #> addFiller -cell FILL64 -prefix filler64
- innovus #> addFiller -cell FILL32 -prefix filler32
- innovus #> addFiller -cell FILL16 -prefix filler16
- innovus #> addFiller -cell FILL8 -prefix filler8
- innovus #> addFiller -cell FILL4 -prefix filler4
- innovus #> addFiller -cell FILL2 -prefix filler2
- innovus #> addFiller -cell FILL1 -prefix filler1



# 14. Before Output: LVS & DRC check

## ✓ LVS(layout verse schemectic)



## ✓ DRC(design rule check)



## ✓ Reminder

- Recommended to perform the two checks after power planning
- Check your design before streaming out



# 14. LVS v.s. DRC v.s. DRV

## ✓ LVS: layout verse schematic

- The signals are connect correctly

## ✓ DRC: design rule check

- Metal density
- Size of poly, metals, vias

## ✓ DRV: design rule violation

- Fan-out
- Load
- Driving capability
- Wire length



# 14. Output Files

✓ The following files will be generated after APR design flow

- CHIP.v
  - Netlist file for post-layout gate-level simulation
- CHIP.sdf
  - Design timing file for post-layout gate-level simulation
- CHIP.def
  - Design exchange format relevant to physical layout
- CHIP.gds
  - GDSII stream file for Calibre-DRC (Design Rule Check)



# 14. Output Data

## ✓ Export Netlist for LVS and simulation

- File -> Save -> Netlist ...
- innovus #> **saveNetlist CHIP\_LAYOUT.v**
- innovus #> **saveNetlist –includePhysicalInst –excludeLeafCell CHIP\_LVS.v**

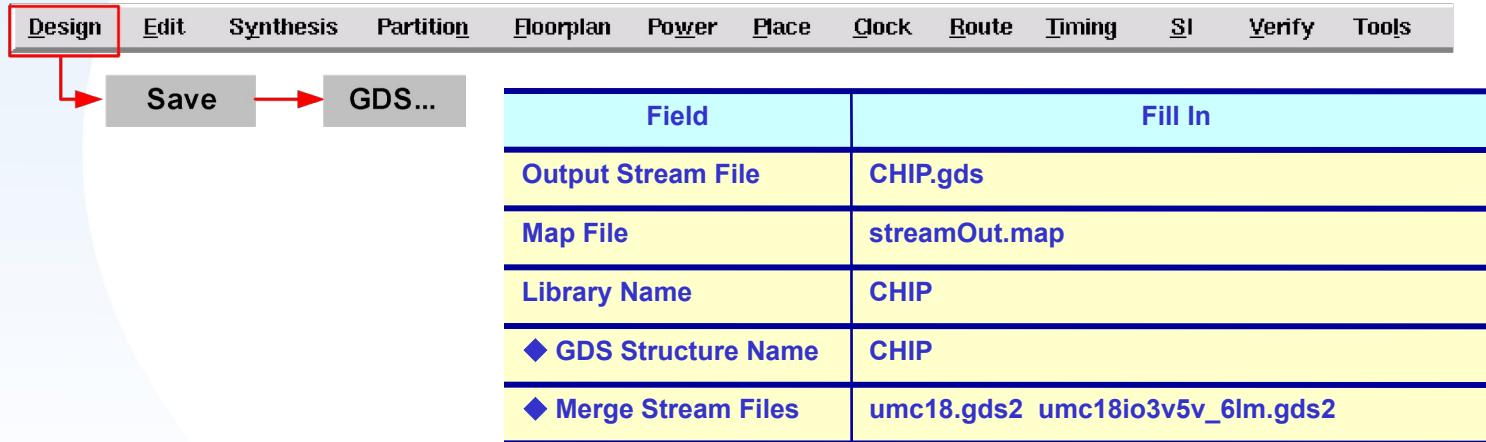
## ✓ Save sdf for post layout simulation

- Innovus #> **setAnalysisMode -checkType setup -analysisType bcwc -cppr none -clockGatingCheck 1 -timeBorrowing 1 -domain clock -useOutputPinCap 1**
- innovus #> **write\_sdf -edges check\_edge CHIP.sdf**



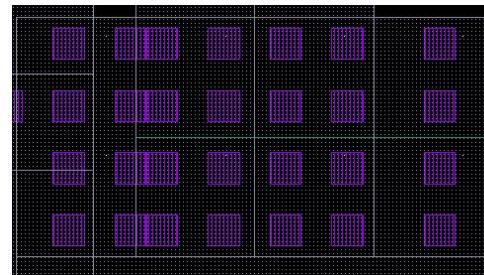
# 14. Output Data

- ✓ Export GDSII stream file for DRC, LVS, LPE (Layout Parameter Extraction), and tape out
  - CHIP.gds
- ✓ Merging GDSII files (UMC 0.18)
  - Merge the GDSII files of standard cells (and macros if any) into complete layout and output to a single GDSII file for hierarchical designs



# Appendix: More on APR

- ✓ DRC & LVS in Virtuoso with streamout GDS
  - ✓ DRC Result
    - PO.R.3
    - M2.R.1
    - VIA3.S.1



## OVERALL COMPARISON RESULTS

# #####  
# #  
# # CORRECT #  
# #  
#####

Warning: Ambiguity points were found and resolved arbitrarily.

| CELL SUMMARY     |        |        |
|------------------|--------|--------|
| Result           | Layout | Source |
| -----<br>CORRECT | CHIP   | CHIP   |

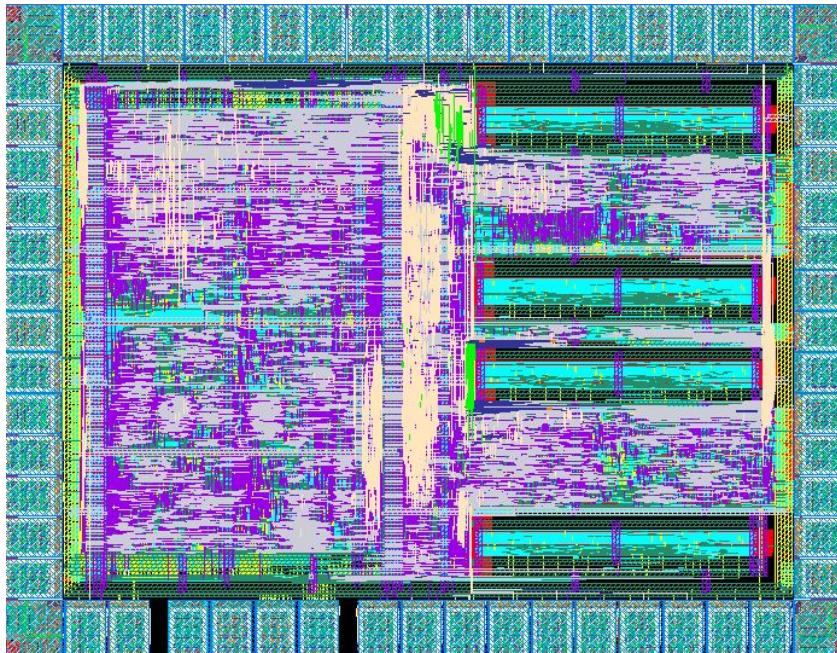


# Appendix: More on APR

## ✓ Macro Layout

- Pad -> Pin
- No Shell

## ✓ Different Power Domain



# Appendix

- ✓ **Modify the Memory.lef file**
  - ◆ Change “core” to “core\_5040”
  - ◆ Change “ME4” to “metal4” and so on
  - ◆ Change “VI1” to “via”
  - ◆ Change “VI2” to “via2” and so on

```
RECT 0.000 0.140 213.900 294.000 ;
LAYER VI1 ;
RECT 0.000 0.140 213.900 294.000 ;
LAYER VI2 ;
RECT 0.000 0.140 213.900 294.000 ;
LAYER VI3 ;
RECT 0.000 0.140 213.900 294.000 ;
```

```
NAMESCASESENSITIVE ON ;
MACRO TEST
CLASS BLOCK ;
FOREIGN TEST 0.000 0.000 ;
ORIGIN 0.000 0.000 ;
SIZE 213.900 BY 294.000 ;
SYMMETRY x y r90 ;
SITE core ;
PIN VCC
 DIRECTION INOUT ;
 USE POWER ;
 SHAPE ABUTMENT ;
PORT
 LAYER ME4 ;
 RECT 212.780 282.580 213.900 285.820 ;
 LAYER ME3 ;
 RECT 212.780 282.580 213.900 285.820 ;
 LAYER ME2 ;
 RECT 212.780 282.580 213.900 285.820 ;
 LAYER ME1 ;
 RECT 212.780 282.580 213.900 285.820 ;
END
```

