

Quality Enhancement of Voice Communication & Audio experience

Application Assignment

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Outline

Although we have sufficient resources and already existing technologies in the field of audio and voice enhancement, there are a lot of opportunities for implementing advanced algorithms using Digital Signal Processing for audio and voice enhancements.

Sound Effect Normalization

- SEN Basics and implementation

Acoustic Echo Cancellation

- Background

Mono to Stereo Conversion

- Definitions

- Conversion approach & Results

Outline

Sound Effect Normalization

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Sound Effect Normalization(SEN)

Sound Effect Normalization(SEN) can be used to improve speech intelligibility in movies through de-emphasis of background sounds and music. Most of the times it is applicable for television shows and sports where we have a lot of background noise due to people shouting and music playing.

Digital Audio Effects:- Audio effects can be classified by the way do their processing. We studied to of them:

- ▶ **Basic Filtering** : Lowpass, Highpass filter etc, Equaliser
- ▶ **Time Varying Filters** : Wah-wah

Basic Digital Audio Filtering Effects: Equalisers

- ▶ **Filters** by definition remove/attenuate audio from the spectrum above or below some cut-off frequency.
But for many audio applications this a little too restrictive.

Basic Digital Audio Filtering Effects:

Equalisers

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Equalisers, by contrast, enhance/diminish certain frequency bands whilst leaving others unchanged.

Basic Digital Audio Filtering Effects:

Equalisers

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But for many audio applications this is a little too restrictive.
- ▶ Therefore **Equalisers** are used.
Equalisers, by contrast, enhance/diminish certain frequency bands whilst leaving others unchanged.
They are built using a series of **Shelving** and **Peak** filters.
Usually second-order filters are employed.

Shelving Filter

Boost or cut the low or high frequency bands with a cut-off frequency, F_c and gain G .

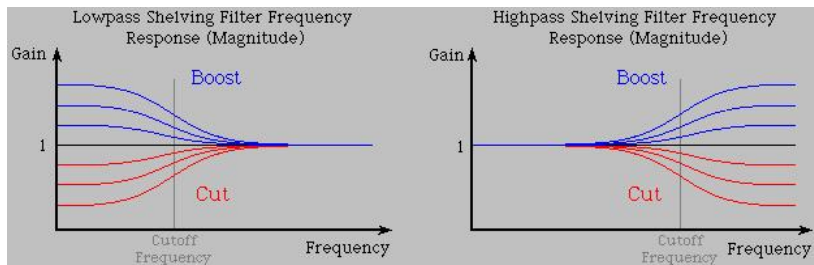


Figure: Shelving Filter Frequency Response

Shelving Filter (Cont.)

A first-order shelving filter may be described by the transfer function:

$$H(z) = 1 + H_0/2 * (1 + A(z)) \quad \text{.....(ForLowPass)}$$

$$H(z) = 1 + H_0/2 * (1 - A(z)) \quad \text{.....(ForHighPass)}$$

Shelving Filter (Cont.)

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where $A(z)$ is a first-order all pass filter passes all frequencies but modifies phase:

For Boost:

$$A(z) = \frac{z^{-1} + a_B}{1 + a_B z^{-1}}$$

Similarly, for Cut we use a_C

Shelving Filter (Cont.)

The gain, G , in dB can be adjusted accordingly:

$$H_0 = V_0 - 1$$

where $V_0 = 10^{(G/20)}$

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and the cut-off frequency for boost or cut are given by:

$$a_B = \frac{\tan(2\pi fc/fs) - 1}{\tan(2\pi fc/fs) + 1}$$
$$a_C = \frac{\tan(2\pi fc/fs) - V_0}{\tan(2\pi fc/fs) + V_0}$$

Peak Filter

Boost or cut mid-frequency bands with a cut-off frequency, F_c , a bandwidth, f_b and gain G .

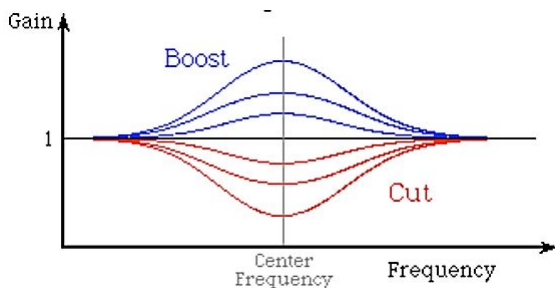


Figure: Peak Filter Frequency Response(Magnitude)

Peak Filter (Contd.)

A second-order peak filter may be described by the transfer function:

$$H(z) = 1 + H_0/2 * (1 + A(z))$$

Peak Filter (Contd.)

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$$H(z) = 1 + H_0/2 * (1 + A(z))$$

where $A(z)$ is a second-order all pass filter

For Boost:

$$A(z) = \frac{-a_B + (d - da_B)z^{-1} + z^{-2}}{1 + (d - da_B)z^{-1} + a_B z^{-2}}$$

Similarly for Cut, we use a_C

Peak Filter (Contd.)

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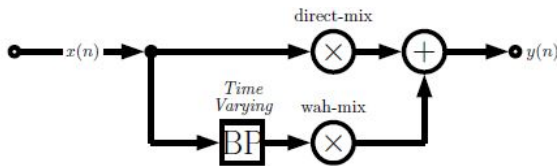
The center/cut-off frequency, d , is given by:

$$d = -\cos(2\pi fc/fs)$$

Time-varying Filters

Wah-wah : A bandpass filter with a time varying centre (resonant) frequency and a small bandwidth. Filtered signal mixed with direct signal.

The signal flow for a wah-wah is as follows:



where BP is a time varying frequency bandpass filter.

It alters the tone and frequencies of the audio signal to create a distinctive sound, mimicking the human voice saying the onomatopoeic name "wah-wah". Hence the name.

Time Varying Filter Implementation:

State Variable Filter

In time varying filters, independent control over the cut-off frequency and damping factor of a filter is needed.

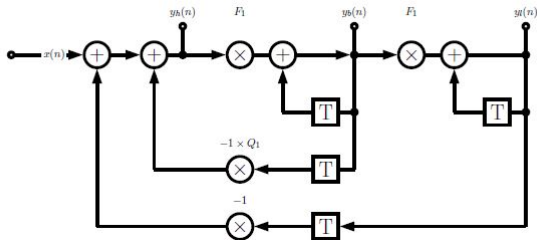
A **State Variable Filter** is implemented for the same.

One further advantage is that lowpass, bandpass and highpass filter output are obtained simultaneously.

A state variable filter is a type of active filter. It consists of one or more integrators, connected in some feedback configuration.

A state variable filter realizes the state-space model directly.

The State Variable Filter



where:

$x(n)$ = input signal

$yl(n)$ = lowpass signal

$yb(n)$ = bandpass signal

$yh(n)$ = highpass signal

and tuning coefficients F_1 and Q_1 related to the cut-off frequency, f_c , and damping, d :

$$F_1 = 2\sin(2\pi f_c / f_s), \text{ and } Q_1 = 2d$$

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Background & Motivation

- ▶ Echo is a phenomenon where a delayed and distorted version of an original sound or electrical signal is reflected back to the source. Since the advent of telephony echoes have been a problem in communication networks
- ▶ **Electrical** echo due to impedance mismatch
- ▶ **Acoustic** echo is due to the coupling between the loudspeaker and microphone
- ▶ **Did you know?**
 - ▶ The range of human hearing is generally considered to be 20 Hz to 20 kHz, but it is far more sensitive to sounds between 1 kHz and 4 kHz. For example, listeners can detect sounds as low as 0 dB SPL at 3 kHz, but require 40 dB SPL at 100 hertz (an amplitude increase of 100)

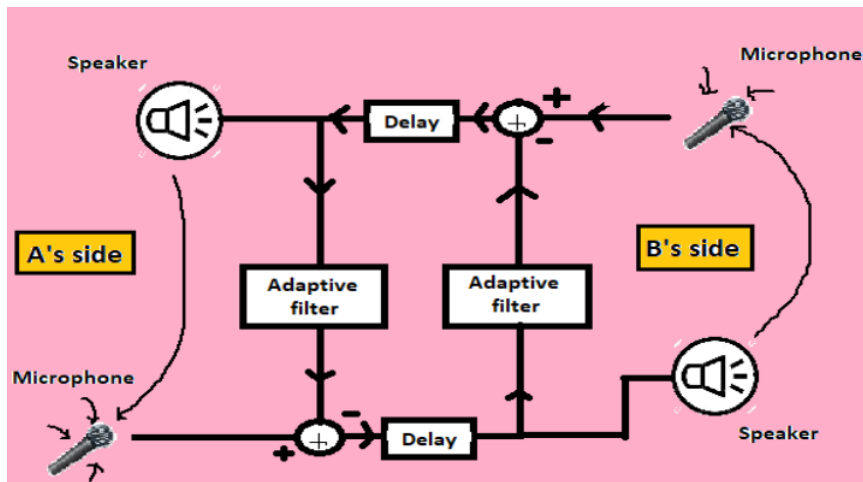


Figure: Basic block diagram of AEC Implementation

AEC Generation block

- ▶ A measured microphone signal $d(n)$ contains two signals:
The near-end speech signal $r(n)$ The far-end echoed speech signal $d(n)$
- ▶ The goal is to remove the far-end echoed speech signal from the microphone signal.

$$e(n) = d(n) - \hat{y}(n)$$

$e(n)$ in turns adjust the coefficients

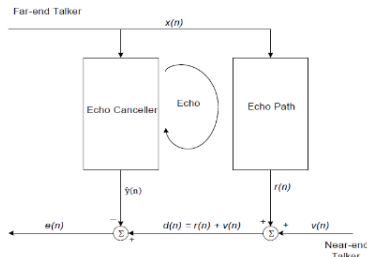


Figure: AEC Generation

Adaptive filter

- ▶ The closed loop digital adaptive filter uses feedback in the form of an error signal to refine its transfer function
- ▶ Least mean squares (LMS) algorithms are a class of adaptive filter used to mimic a desired filter by finding the filter coefficients that relate to producing the least mean square of the error signal
- ▶ [Here] the adaptive filter tends to minimize

$$[e(n)]^2$$

and accordingly adjust

$$h^*[n]$$

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What is Mono & Stereo sound?

- ▶ Stereo uses two (or more) independent audio channels to reproduce sound creating an illusion of **multi-directional** audible perspective
- ▶ Mono-aural sound is a single channel sound intended to reproduce sound perceived as coming from **single** position
- ▶ The practical difference is that mono has only one spatial dimension; something can be either close to (loud) or far away (quiet) from the listener. Stereo has two dimensions: close/far and left/right, giving a sense of space while listening to a stereo track

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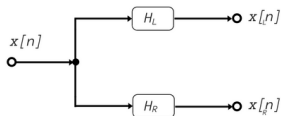


Figure: Basic blockscheme of the pseudo-stereo system

► Down-mix compatibility

$$\begin{aligned}x_M[n] &= x_L[n] + x_R[n] \\&= x[n] * h_L[n] + x[n] * h_R[n] \\&= x[n - D]\end{aligned}$$

$$\begin{aligned}\Rightarrow X_M(e^{j\Omega}) &= X(e^{j\Omega})H_L(e^{j\Omega}) + X(e^{j\Omega})H_R(e^{j\Omega}) \\&= X(e^{j\Omega})e^{-j\Omega D}\end{aligned}$$

$$\Rightarrow H_L(e^{j\Omega}) + H_R(e^{j\Omega}) = e^{-j\Omega D}$$

Continued...

- ▶ For a neutral sounding stereo output without coloration artifacts the sum of the magnitude frequency responses must be constant

$$\left| H_L(e^{j\Omega}) \right| + \left| H_R(e^{j\Omega}) \right| = 1$$

- ▶ For obtaining real-valued impulse responses both pseudo-stereo filters must feature conjugate symmetry

$$H_{L/R}(e^{j\Omega}) = H_{L/R}^*(e^{-j\Omega})$$

- ▶ Left channel decorrelation filter

$$H_L[k] = (1/2 + (1/\pi) \arctan(\omega^2 R[k])) e^{-j(2\pi kD/N)}$$

where $R[k]$ is DFT of standard Gaussian Noise.

Time domain implementation

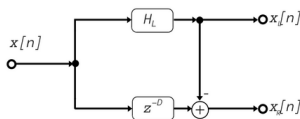


Figure: Blockscheme of the time-domain realization

- To get the left channel FIR filter impulse response, we have

$$h_L[n] = \mathcal{F}^{-1}(H_L[k])$$

- For the right channel,

$$x_R[n] = x[n - D] - x_L[n]$$

where

$$D = (N + 1)/2$$

is the group delay of the FIR filter of length N .

Frequency domain implementation

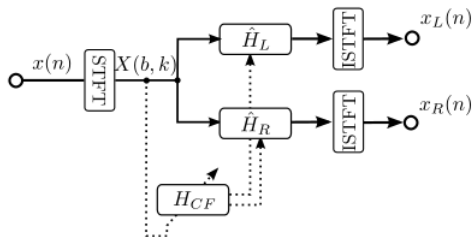


Figure: Blockscheme of the frequency-domain realization

- ▶ The time-frequency representations of the output channels are computed with a dot-wise product and a following overlapping inverse STFT synthesis of the left and right output channel.

$$X_{L/R}(b, k) = X(b, k) \cdot H_{L/R}[k]$$

References I



Dave Marshall

Digital Audio Effects

Dave.Marshall_CM0268_PDF_10_CM0268_Audio_FX



Markus Christoph

Acoustic echo cancellation

United States Patent US12710092 (2009)



Marco Fink, Sebastian Kraft, Udo Zolzer,

Downmix-compatible conversion from mono to stereo in
time- and Frequency-domain

*Proc. of the 18 th Int. Conference on Digital Audio Effects
(DAFx-15), Trondheim, Norway, Nov 30 - Dec 3, 2015*

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AEC in MATLAB
MATLAB Examples



Wikipedia
Adaptive Filtering



Some interesting forums for discussion on DSP
Music Stack Exchange
Science Encyclopedia
DSP Stack Exchange