

Experimental implementation of a cognitive radio using OpenBTS , GNU Radio and spectrum sensing techniques

M. Tech. Dissertation

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by

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Abstract

Our goal is to set up a software defined cognitive radio using OpenBTS, GNU Radio and USRP kits. We decide on a frequency channel, to run our cognitive OpenBTS system in, beforehand. First we sense the presence of ongoing calls made by the primary users in the predefined frequency channel. The sensing is done by calculating the energy in that channel using a technique of energy detection called periodogram analysis. If the energy is above some predefined threshold then there are ongoing calls in that channel and hence we wait for the calls to end. As soon as the calls involving the primary users end the energy in that channel goes low. GNU Radio detects this change and it provides the ARFCN, corresponding to this channel, to the secondary BTS system and the secondary BTS starts using this ARFCN allowing secondary users to make calls and send SMSs.

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Chapter 1

Introduction

1.1 Motivation

Due to the rapid increase of mobile phones and other wireless communication devices, there is a need for efficient utilization of the available radio spectrum. The Spectrum Policy Task Force, a group under the Federal Communications Commission (FCC) in the United States, published a report in 2002 saying [1]:

“In many bands, spectrum access is a more significant problem than physical scarcity of spectrum, in large part due to legacy command-and-control regulation that limits the ability of potential spectrum users to obtain such access.”

If we scan the spectrum in metropolitan cities which are heavily used regions, we find that some frequency bands are unoccupied most of the time[6]. These are referred to as spectrum holes. A spectrum hole is a band of frequencies assigned to a primary user, but, at a particular time and specific geographic location, the band is not being utilized by that user[3].

This problem of inefficient utilization of spectrum can be solved by allowing secondary users which are non licensed, to access these spectrum holes. Cognitive radio which includes software defined radio, is a means to accomplish this by utilizing these spectrum holes intelligently and efficiently[4][2][5]. It uses one of the spectrum sensing techniques to identify the spectrum holes in the radio spectrum.

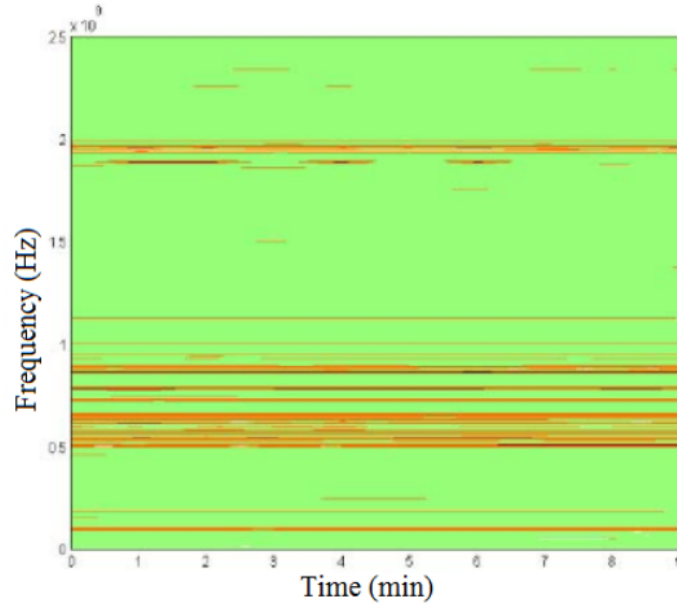


Figure 1.1: Frequency usage of Spectrum Band

1.2 Cognitive Radio

A cognitive radio is an intelligent radio whose primary objective is efficient utilization of the radio spectrum. It can be programmed and configured dynamically. It works on the principle of understanding-by-building to learn from the surrounding environment and adapt to changes in the RF stimuli by making corresponding changes in operating parameters. The transceiver is designed to find an unoccupied channel in the vicinity and utilize it for transmission. It enables coexistence of primary licensed users and secondary unlicensed users. Whenever a primary user wants to occupy the channel which is currently in use by secondary users, it finds some other unoccupied channel in the vicinity and secondary users migrate seamlessly to this new channel thus vacating the previously used channel for primary users.

1.3 Contribution of thesis

An experimental setup is developed which demonstrates the presence of secondary users along with primary users in the existing GSM network and utilizing the already existing resources there by increasing the total mobiles in the network.

1. A two frequency band cognitive system is developed where secondary users migrate to frequency f_2 if frequency f_1 is occupied and viceversa.
2. A two frequency system is extended to a four frequency system where we demonstrate that primary users are occupying two bands out of these four and secondary users occupy one out of the other two free bands.
3. We have used energy detection spectrum sensing technique and CUSUM peak detection technique to detect the presence of primary users. Band occupied by secondary users is continuously monitored to check if primary users are trying to occupy that band and as soon as the request from primary users is detected a new free band is found out in the vicinity and utilized by secondary users for transmission there by vacating the band for primary users .

1.4 ORGANIZATION

The rest of this document is organized as follows. Chapter 2 briefly describes the GSM architecture and its Um interface. Chapter 3 gives a literature survey on Universal Software Radio Peripheral (USRP N210) the hardware used in this project. Literature survey done on the GNU Radio software package and OpenBTS software is described in Chapter 4 and 5 respectively. Chapter 6 covers spectrum sensing techniques to detect the presence of primary users in the channel. Chapter 7 covers an implementation of cognitive radio using GNU Radio and OpenBTS. It describes the experimental setup for our project in the beginning followed by detailed description of what we have achieved in this project along with a flow chart of our work. The final chapter of this thesis is the conclusion of our project followed by future work.

Chapter 2

Overview of traditional GSM networks

2.1 What is GSM?

GSM, or Global System for Mobile Communications, is a European standard for the Mobile telecommunications and it is considered as one of the most popular standard worldwide. There are thirteen different frequency bands defined in GSM. However, the 850 MHz, 900 MHz, 1800 MHz, and 1900 MHz bands are the most commonly used. The frequency bands employed within each of the four ranges are similarly organized. They differ essentially only in the frequencies, such that various synergy effects can be taken advantage of; hence, here we give some details only for the usage in the 900 MHz band.

In the 900 MHz band, a total of 70 MHz bandwidth is allocated, two 25-MHz frequency bands for uplink and downlink and a 20 MHz unused guard band between them. The MS transmits in the 890 to 915 MHz range (uplink) and the BTS transmits in the 935 to 960 MHz (downlink) band. This corresponds to 124 duplex channels, where each channel within a BTS is referred to as an Absolute Radio Frequency Channel Number (ARFCN). This number describes a pair of frequencies, one uplink and one downlink, and is given a channel index between C0 and C123, with C0 designated as the beacon channel. An ARFCN could be used to calculate the exact frequency (in MHz) of the radio channel. In the GSM 900 band, this is computed by the following equations:

$$F_{uplink}(n) = 890 + 0.2 * n \quad 1 \leq n \leq 124$$

$$F_{downlink}(n) = F_{uplink}(n) + 45 \quad 1 \leq n \leq 124$$

Similar formulae are also defined for the other GSM frequency bands.

The principle component groups of a GSM network are as follows:

- The Mobile Station (MS)
- The Base Station System (BSS)
- The Network Switching System(NSS)



Figure 2.1: The conventional GSM architecture.

Source: http://www.hill2dot0.com/wiki/index.php?title=Image:G2407_{ }GSM-Architecture.jpg
[Accessed on Oct 22, 2013]

2.1.1 Mobile Station

The MS consists of two parts, the Mobile Equipment (ME) and Subscriber Identity module (SIM). The ME has an identity number called the International Mobile Equipment Identity

(IMEI) associated with it, which is unique for that particular device and permanently stored in it. The SIM card consists the International Mobile Subscriber Identity (IMSI) number which is used to identify the subscriber to the system. The IMEI and the IMSI are independent of each other and hence allow personal mobility.

2.1.2 Base Station System

The GSM Base Station System is the equipment located at a cell site. It comprises of a combination of digital and RF equipment. The BSS provides the link between the MS and the Mobile Services Switching Centre (MSC). The BSS consists mainly of:

The Base Transceiver Station (BTS) The BTS contains the RF components that provide the air interface for a particular cell. This is the part of the GSM network which communicates with the MS. The antenna is included as part of the BTS.

The Base Station Controller (BSC) The BSC provides the control for the BSS. The BSC communicates directly with the MSC. The BSC may control single or multiple BTSs. Crucial functions like radio channel link establishment, frequency hopping, and handovers from one cell to another.

2.1.3 Network Switching Subsystem

The Network Switching Subsystem includes the main switching functions of the GSM network. It also contains the databases required for subscriber data and mobility management. Its main function is to manage communications between the GSM network and other telecommunications networks. The main components of the Network Switching System are:

Mobile Services Switching Centre (MSC)

The MSC does call-switching and its overall purpose is the same as that of any telephone exchange. When the MSC provides the interface between the PSTN and the BSSs in the GSM network it will be known as a Gateway MSC. In this position it will provide the switching required for all MS originated or terminated traffic. Each MSC provides service to

MSs located within a defined geographic coverage area. One MSC is capable of supporting a regional capital with approximately one million inhabitants. The functions carried out by the MSC are: Call Processing, Operations and Maintenance Support, Internetwork Interworking and Billing

Home Location Register (HLR)

The HLR is a central database that contains the details of each mobile phone subscriber that is authorized to use the GSM core network. The IMSI of each SIM acts as a primary key to each HLR record. Each MSISDN is also a primary key to the HLR record. The HLR data is stored as long as a subscriber remains with the mobile phone operator. Data stored in the HLR against each IMSI are, GSM services that the subscriber has requested, GPRS settings to allow the subscriber to access packet services, current location of the subscriber, etc.

Visitor Location Register (VLR)

The VLR is a database of the subscribers who have roamed into the jurisdiction of the MSC which it serves. Each main base station in the network is served by exactly one VLR, hence a subscriber cannot be present in more than one VLR at a time. The data stored in the VLR has either been received from the HLR, or collected from the MS. Data stored includes: IMSI (the subscriber's identity number), authentication data, MSISDN, GSM services that the subscriber is allowed to access, the HLR address of the subscriber.

2.2 Um Interface

The Um interface is the air interface of the GSM mobile telephone standard. It is the interface between the MS and the BTS. It is called Um because it is the mobile analog to the U interface of ISDN. Um is defined in the GSM 04.xx and 05.xx series of specifications.

The layers of GSM are initially defined in GSM 04.01 Section 7 and roughly follow the OSI model. Um is defined in the lower three layers of the model.

2.2.1 Physical Layer (L1)

The Um physical layer is defined in the GSM 05.xx series of specifications, with the introduction and overview in GSM 05.01. For most channels, Um L1 transmits and receives 184-bit control frames or 260-bit vocoder frames over the radio interface in 148-bit bursts with one burst per timeslot. There are three sublayers:

Radiomodem

This is the actual radio transceiver. GSM uses 8PSK modulation with 1 bit per symbol which produces a 13/48 MHz (270.833 kHz or 270.833 K symbols/second) symbol rate and a channel spacing of 200 kHz. Since adjacent channels overlap, the standard does not allow adjacent channels to be used in the same cell. The standard defines several bands ranging from 400 MHz to 1990 MHz. GSM is frequency duplexed, meaning that the network and MS transmit on different frequencies, allowing the BTS to transmit and receive at the same time. Transmission from the network to the MS is called the “downlink” and that from the MS to the network is called the “uplink”. The GSM uplink and downlink bands are separated by 45 or 50 MHz. Uplink/downlink channel pairs are identified by an index called the ARFCN. Within the BTS, these ARFCNs are given arbitrary carrier indexes C0..Cn-1, with C0 designated as a Beacon Channel and always operated at constant power. The radio channel is time-multiplexed into 8 timeslots, each with a duration of 156.25 symbol periods. These 8 timeslots form a frame of 1,250 symbol periods. The capacity associated with a single timeslot on a single ARFCN is called a physical channel (PCH) and referred to as “CnTm” where n is a carrier index and m is a timeslot index (0-7). Each timeslot is occupied by a radio burst with a guard interval, two payload fields, tail bits, and a midamble.

Multiplexing and Timing

GSM uses TDMA to subdivide each radio channel into as many as 16 traffic channels or as many as 64 control channels. The multiplexing patterns are defined in GSM 05.02. Each physical channel is time-multiplexed into multiple logical channels according to the rules of GSM 05.02. Traffic channel multiplexing follows a 26-frame (0.12 second) cycle called a “multiframe”. Control channels follow a 51-frame multiframe cycle. The C0T0 physical

channel carries the synchronization channel(SCH), which encodes the timing state of the BTS to facilitate synchronization to the TDMA pattern.

FEC Coding

The coding sublayer provides forward error correction. As a general rule, each GSM channel uses a block parity code (usually a Fire code), a rate-1/2, 4th-order convolutional code and a 4-burst or 8-burst interleaver.

2.2.2 Data Link Layer(L2)

The Um data link layer, LAPDm, is defined in GSM 04.05 and 04.06. LAPDm is the mobile analog to ISDN's LAPD.

2.2.3 Network layer(L3)

The Um network layer is defined in GSM 04.07 and 04.08 and has three sublayers. A subscriber terminal must establish a connection in each sublayer before accessing the next higher sublayer.

Radio Resource (RR) This sublayer manages the assignment and release of logical channels on the radio.

Mobility Management (MM) This sublayer authenticates users and tracks their movements from cell to cell. It is normally terminated in the VLR or HLR.

Call Control (CC) This sublayer connects telephone calls and is taken directly from ITU-T Q.931. GSM 04.08 Annex E provides a table of corresponding paragraphs in GSM 04.08 and ITU-T Q.931 along with a summary of differences between the two. The CC sublayer is terminated in the MSC.

The access order is RR, MM, CC. The release order is the reverse of that. Note that none of these sublayers terminate in the BTS itself. The standard GSM BTS operates only in layers 1 and 2.

Chapter 3

OpenBTS

3.1 What is OpenBTS?

OpenBTS is a Unix application that uses a software radio to present a GSM Um interface to handsets and uses a SIP softswitch or PBX to connect calls. (You might even say that OpenBTS is a simplified form of IMS that works with 2G feature-phone handsets). The combination of the global -standard GSM air interface with low cost VoIP backhaul forms the basis of a new type of cellular network that can be deployed and operated at substantially lower cost than existing technologies in many applications, especially rural cellular deployments and private cellular networks in remote areas.

3.2 The OpenBTS Application Suite

A complete OpenBTS P2.8 installation comprises several distinct applications:

OpenBTS The actual OpenBTS application, containing most of the GSM stack above the radio modem.

Transceiver The software radio modem and hardware control interface.

Asterisk The VoIP PBX in the standard public release configuration.

Smqueue The store-and-forward server for text messaging.

Subscriber Registry A database of subscriber information that replaces both the Asterisk SIP registry and the GSM Home Location Register (HLR).

Other Servers Other optional GSM services, beyond speech and text messaging, are supported through external servers.

The OpenBTS and Transceiver applications must run inside each GSM/SIP access point. The Asterisk and the subscriber registry applications communicated through the filesystem and therefore must run on the same computer, but that computer can be remote from the access point. smqueue and the other servers can run anywhere and may have multiple instances.

3.2.1 OpenBTS

The OpenBTS application contains:

- L1 Time division multiplexing(TDM) functions (GSM 05.02)
- L1 Forward error correction(FEC) functions (GSM 05.03)
- L1 closed loop power and timing controls (GSM 05.08 and 05.10)
- L2 Link access protocol on Dm-channel (LAPDm) (GSM 04.06)
- L3 radio resource management functions (GSM 04.08)
- L3 GSM-SIP gateway for mobility management
- L3 GSM-SIP gateway for call control
- L4 GSM-SIP gateway for text messaging

The general design approach of OpenBTS is avoid implementing any function above L3, so at L3 or L4 every subprotocol of GSM is either terminated locally or translated through a gateway to some other protocol for handling by an external application. Similarly, OpenBTS itself does not contain any speech transcoding functions above the L1 FEC parts.

3.2.2 Transceiver

The transceiver application performs the radio modem functions of GSM 05.05 and manages the Gigabit Ethernet interface (USB2 interface, in case of USRP1 or older models) to the radio hardware.

3.2.3 Asterisk

OpenBTS uses a SIP switch or PBX to perform the call control functions that would normally be performed by the mobile switching center in a conventional GSM network, although in most network configurations this switching function is distributed over multiple switches. These switches also provide transcoding services. In OpenBTS P2.8 the standard SIP switch is Asterisk 1.8.

3.2.4 Subscriber Registry

OpenBTS uses a modified SIP registry as a substitute for the home location register found in a conventional GSM network. OpenBTS also relies on Asterisk for any transcoding functions.

3.2.5 Smqueue

Smqueue is a store-and-forward server that is used for text messaging in the OpenBTS system. Smqueue is required to send a text message from one MS to another, or to provide reliable delivery of text messages to an MS from any source.

3.2.6 Network Organization

In the simplest network, with a single access point, all of the applications in the suite run inside the access point on the same embedded computer.

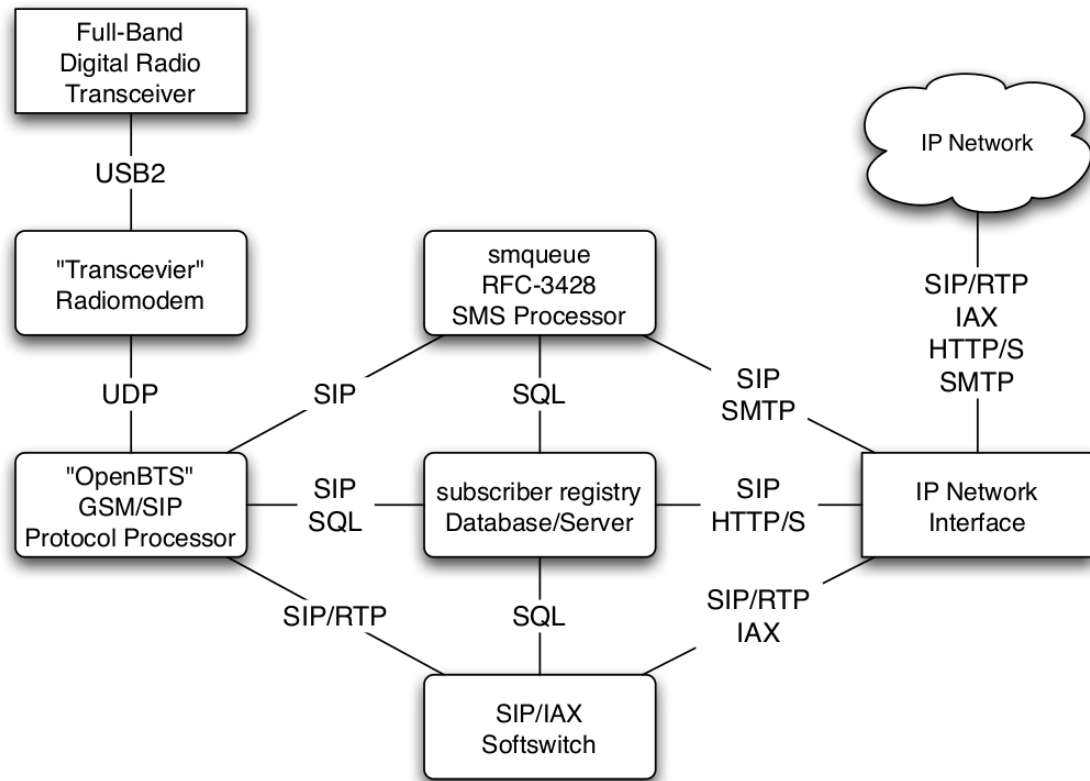


Figure 3.1: **Network Organization of OpenBTS:** The smqueue block handles SMSs. The subscriber registry database contains the details of all registered users and it maps the registered IMSIs to corresponding dialling numbers. The softswitch connects speech calls (e.g. Asterisk, FreeSwitch). The transceiver performs radio modem functions and manages the Gigabit Ethernet interface (USB2 interface, in case of USRP1 or older models) to the USRP device. The OpenBTS itself is the GSM implementation from the TDMA part of L1 up through L3 and the L3/L4 boundary. It has a SIP interface to communicate with the other blocks like smqueue, subscriber registry, etc.

Source: <https://wush.net/trac/rangepublic/attachment/wiki/WikiStart/PReleaseManual.pdf> [Accessed on Oct 22, 2013]

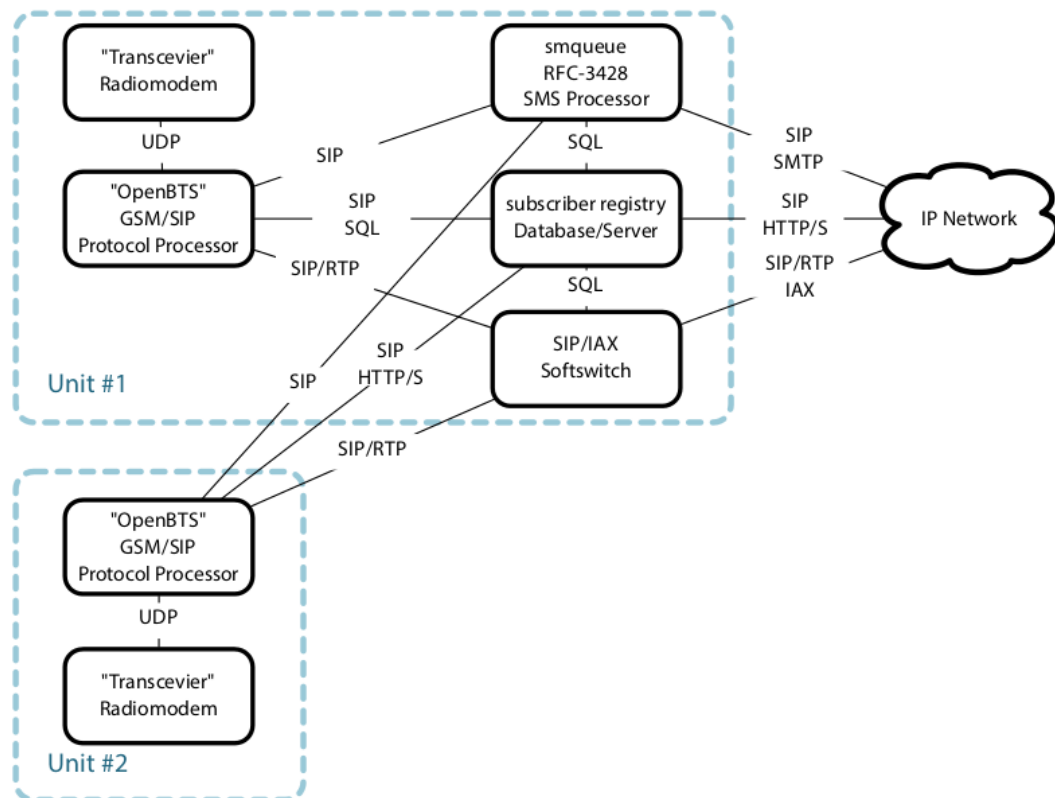


Figure 3.2: Two access points with unit #1 providing servers for both.

Source: <https://wush.net/trac/rangepublic/attachment/wiki/WikiStart/OpenBTS-4.0-Manual.pdf>
 [Accessed on May 27, 2014]

Chapter 4

Spectrum Sensing

Due to limited availability of spectrum resource, there is a serious impact on the emerging mobile applications. Hence there is a need to efficiently utilize the available radio spectrum. The problem right now is not the physical scarcity of the radio spectrum rather the inefficient use of the spectrum. Solution to this is cognitive radio. The major problems in cognitive radio is detection of primary users in the licensed spectrum and enable secondary users to quit the frequency band as quickly as possible if the corresponding primary radio emerges to avoid interference to primary users. This technique is called spectrum sensing. Spectrum sensing is the first step to implement cognitive radio system.

There are various methods for local spectrum sensing proposed by researchers.

The following section describes three important methods:

1. Energy detection
2. Matched filter detection
3. Cyclostationarity detection

4.1 Energy Detection

Measuring the energy of a particular band is one of the simplest techniques to detect the presence of primary users in that band. It is one of the most widely used technique to

detect spectrum holes as it requires no a priori knowledge of the primary radio. Apart from this major advantage the technique is very cost efficient and less complex compared to other techniques. We calculate the energy of the received radio spectrum and this energy is compared with a predefined energy detection threshold to conclude whether primary user is present or absent in the frequency of interest. This technique is an optimal one when we have absolutely no knowledge of the user occupying the channel in advance. The following block diagram describes energy detection technique:

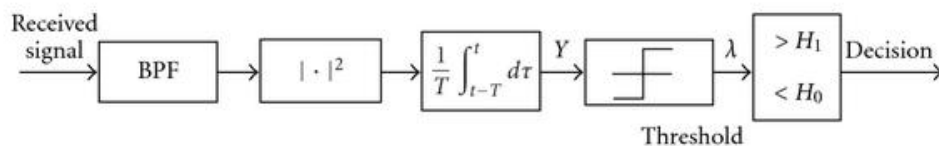


Figure 4.1: Block diagram of Energy Detection implementation. *Source: <http://www.hindawi.com/journals/ijvt/2011/630467.fig.003.jpg> [Accessed on Oct 22, 2013]*

As described in the energy detection block diagram, it is basically a hypothesis testing problem with two possible hypothesis H_0 and H_1 . Hypothesis H_1 concludes the presence of primary users in the band of interest and hypothesis H_0 concludes their absence. And energy detection technique is basically about distinguishing between these two hypotheses[?].

$$x(t) = n(t); \quad H_0$$

$$x(t) = hs(t) + n(t); \quad H_1$$

where $x(t)$ is signal received by secondary user and $s(t)$ is primary radio signal, $n(t)$ is additive white Gaussian noise (AWGN) and h is the amplitude gain of the channel. $s(t)$ and $n(t)$ are assumed to be independent of each other. Signal detection is performed using an energy detector and compute decision statistics Y which corresponds to energy collected in observation time Y and bandwidth W and comparing this statistics to a predetermined threshold. Energy detection is implemented using average periodogram analysis which is covered in later part of this report.

4.2 Matched filter detection

Matched filter is a linear filter used to match a particular transit waveform with the reference signal. The output is maximum when the match happens. When there is a priori knowledge of primary radio, matched filter technique is applied. Matched filter operation is equivalent to correlation operation in which the incoming signal is convolved with a filter whose impulse response is mirror and shifted version of reference signal. This output is then compared with the threshold for primary user detection. It is mathematically defined as:

$$Y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$$

x is the unknown signal and h is the impulse response of matched filter which is matched to reference signal for maximizing SNR.

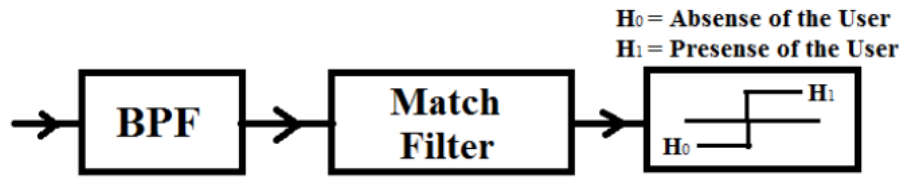


Figure 4.2: Block diagram of Matched Filter implementation.

There is a constraint on this technique. We need to have a prior information about the primary radio to perform matched filtering. However matched filter requires demodulation of primary signal which means it has information of primary radio both at the PHY layer and the MAC layer like operating frequency, modulation type, packet format bandwidth etc. But the cumbersome part is it has to achieve coherency with primary user by means of timing and carrier synchronization. This coherent detection is still achievable since primary signals have pilots, preambles etc to serve the purpose. The advantage of matched filter detection is when the information of the primary user signal is known, it is optimal detection in stationary Gaussian noise. But the performance of matched filter detection depends on the accuracy of the information of primary radio. This technique also requires cognitive radio to have dedicated receiver for every type of primary user which in turn results in complex hardware and large power consumption[?].

4.3 Cyclostationarity detection

This technique utilizes periodicity property of the received signal to detect presence of primary users. Periodicity property is generally exhibited by the communication signals due to sinusoidal carriers, pulse train, hopping sequences etc. Due to this underlying periodicity most of the communication signals can be modeled as cyclostationary processes. This technique detects a random primary signal with a particular modulation type in a background of noise and other modulated signals.

Cyclostationary feature detection is robust to noise and is a better performer than energy based detection in low SNR regions. This technique also requires a priori knowledge of the primary signal. Also this technique is computationally highly complex and the sensing time is also quite long. Due to these reasons it is less commonly used than energy based detection.

Block diagram of cyclostationary detection is given in figure 4.3 [?].

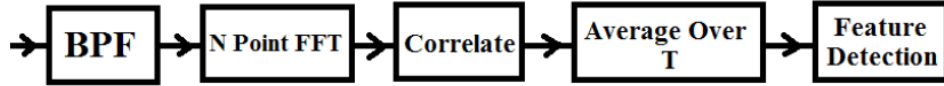


Figure 4.3: Block diagram of Cyclostationary Detection implementation.

The detection is done by finding a unique cyclic frequency of the spectral correlation function of the received signal[?]. The spectral correlation function is the Fourier transform of the cyclic autocorrelation function the spectral correlation function is defined as:

$$S(f, \alpha) = \int_{-\infty}^{\infty} R_x^{\alpha}(\tau) e^{-2\pi f \tau} d\tau$$

Where the cyclic autocorrelation function is defined by:

$$R_x^{\alpha} = E\{x(t + \tau)x^*(t - \tau)e^{-2\pi\alpha t}\}$$

Here $x(t)$ is the signal received and α is the cyclic frequency. The spectral correlation function is also termed as cyclic spectrum. This is a two dimensional transform unlike power spectral density which is 1 dimensional. For successful detection under cyclo-stationary based

spectrum sensing, we need a priori knowledge of the cyclo-stationary features of the received signal only. However matched filtering is the optimal solution when we completely know about received signal in advance. This technique doesn't work well when the underlying noise is stationary. More over channel fading destroys the property of cyclo stationarity of the received signal and is also susceptible to sampling clock offset.

4.4 Comparision of various spectrum detection techniques

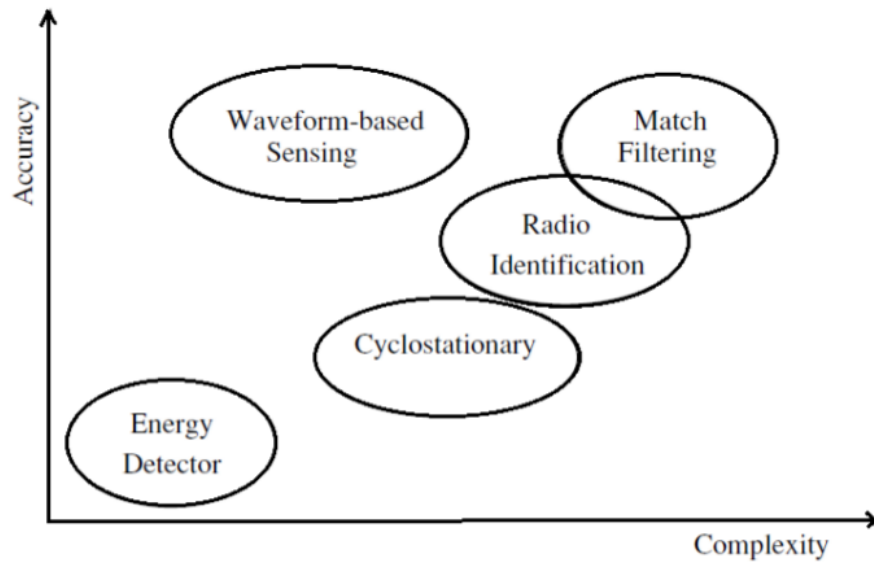


Figure 4.4: Comparison of sensing methods

Figure 4.4 compares various spectrum sensing techniques on the basis of their accuracy and complexity. We can see that energy detection technique is the least accurate and least complex of all where as matched filtering is most complex and highly accurate. Other techniques lie in between with some having more accuracy and some less complex. There is no ideal detector that suits all occasions. Thus decisions, compromises and tradeoffs must be made depending on primary radio type, transmission and propagation characteristics, characteristics of secondary user receiver, and etc[?].

4.5 Implementaton of energy detection technique

We have used energy detection technique for spectrum sensing in our project for the detection of primary users in the band of interest. Average periodigram analysis is a method to implement energy detection technique of spectrum sensing. This section describes average periodogram analysis. Implementation of wide band spectrum analyzer using this technique is also described in this section.

4.5.1 Average periodogram analysis

Average Periodogram analysis estimates the power spectrum of the received signal and it is based on the Discrete Fourier Transform (DFT) of finite length segments of signal. In this technique signal is sectioned into finite length segments and periodogram of each segment is calculated which are also referred to as modified periodograms. Then an average of all these modified periodogram is calculated[?].

Let $X[n]; n = 0, 1 \dots L - 1$ be the discrete time signal which is divided into M finite length segments of equal length, where N is the length of each segment i.e. $MN = L$; $X_r[n]; n = 0, 1 \dots N - 1$ is the r th segment and $W[n]; n = 0, 1 \dots N - 1$ is the window applied to each segment. The modified periodogram for the r th segment is,

$$I_r[k] = \frac{1}{NU} |V_r[k]|^2 \quad k = 0, 1 \dots N - 1$$

where $V_r[k]$ is a N point DFT and U is normalization factor i.e. , $V_r[k] = DFT\{W[n] * X_r[n]\}$ and $U = \frac{1}{N}(\sum_{n=0}^{N-1} (W[n])^2)$ The PSD of $X[n]$ sequence is then the time averaged periodogram estimate ,

$$I[k] = \frac{1}{M} \left| \sum_{r=0}^{M-1} X_r[k] \right|$$

4.6 Wide band spectrum analyzer

GNU radio packages provide a tool for wide band spectrum sensing called `usrp_spectrum_sense.py`. The program is provided in the appendix. It is used as a basic code for wide band spectrum

analyzer implementation. The output of this code is the magnitude squared of the FFT. This means for each FFT bin the output is $Y[i] = \text{re}[X[i]] * \text{re}[X[i]] + \text{im}[X[i]] * \text{im}[X[i]]$. We can calculate the power by taking square root of the output. We need N time samples of $x(t)$ sampled at a sampling frequency of F_s to use N point complex FFT $X(\omega)$ analysis. An appropriate window function is to be selected to reduce spectral leakage and applied to these time samples. The output of the complex FFT will represent the frequency spectrum content as follows: The first value of the FFT output ($\text{bin}0 == X[0]$) is the passband centre frequency. The first half of FFT spectrum ($X[1]$ to $X[N/2 - 1]$) contains the positive baseband frequencies, which corresponds to the passband spectrum from centre frequency to $+F_s/2$. The second half of the FFT ($X[N/2]$ to $X[N - 1]$) contains the negative baseband frequencies, i.e. from $-F_s/2$ to centre frequency.

For our project purpose, we collected 1024 samples using a tuner centered at uplink frequency of our interest, say 900 MHz. 1024 is chosen as the number of FFT points because the number of FFT points has to be a power of 2 for the fast execution of the FFT algorithm. Default sampling frequency is set as 10 MHz. The frequency resolution is therefore: $10 \text{ MHz} / 1024 = 9.7656 \text{ MHz}$. The decimation is defined as dsp rate divided by sample rate. The UHD driver requires the decimation value to be an even number. The dsp rate is the actual hardware-level sampling rate of the USRP kit. It is the rate at which the USRP device takes analog samples from the external world and converts them to digital form. The dsp rate of the USRP is 100 MHz. Hence we chose sampling frequency to be 10 MHz which gives a decimation value of: $100 \text{ MHz} / 10 \text{ MHz} = 10$.

Chapter 5

Implementation of cognitive radio using OpenBTS

In our project we are able to successfully demonstrate the coexistence of primary users and secondary users in the same frequency channel in the GSM band. In order to accomplish this, we have implemented a cognitive radio which detects the spectrum holes in the radio spectrum and enables secondary users to utilize these for communication. An experimental setup has been developed for this demonstration using OpenBTS and GNU radio software and USRP N210 as hardware.

Experimental setup diagram:

5.1 Description of setup

The figure above describes the experimental setup for a two-frequency system. The primary system has only one USRP as an RF front and it runs OpenBTS. The secondary system has two USRP kits connected to it and one of them runs OpenBTS and the other GNURadio. This secondary system has cognitive capabilities. To provide cognitive capabilities it was required that OpenBTS and GNU radio run together in the same system and talk to each other which was challenging. Secondary system continuously senses the frequency band of interest and does decision making depending upon the analysis of the data collected and changes its parameters accordingly so that primary and secondary users coexist. The

spectrum sensing is accomplished by using GNU radio. Also we made GNU radio and OpenBTS coordinate to behave in appropriate manner and take dynamic decisions as and when required to make over all system behave in a cognitive manner.

First a two-frequency cognitive system is developed. For this two GSM bands are used with centre frequency 945 MHz (F1) and 950 MHz (F2). Secondary users are made to occupy one of these two bands say F1. Then we make the primary users enter the same band. This results in an increase in energy levels in this band which is sensed by the secondary system as it continuously scans this band. Immediately secondary users are shifted to other frequency band (F2) there by vacating F1 for primary users. Hence a two-frequency cognitive system demonstrating coexistence of a pair of primary and secondary users is accomplished.

The whole technique is described using a flow graph below:

Flow graph here :

Now this two frequency system is expanded to a four frequency system where we have $F1 = 936$ MHz, $F2 = 943$ MHz, $F3 = 950$ MHz, $F4 = 957$ MHz. The experimental setup is also expanded with two primary systems and one secondary system. Each primary system has an USRP kit running OpenBTS and the secondary system has two USRP kits connected to it, one for OpenBTS and the other one for GNURadio, as we had previously in the two-frequency system.

Figure for 4 frequency system here:

Here we make a pair of primary users occupy one of the four frequency channel say F2. We make secondary users use a frequency channel say F1. Now the other pair of primary users try to enter frequency F1 for communication. This is sensed by the secondary system and it tries to migrate secondary users to F2 which also happens to be occupied already. Our secondary system detects that F2 is occupied and therefore moves on to find a spectrum hole in the four-frequency spectrum. It finds that frequency F3 is unoccupied and thus allows secondary users to enter F3 and utilize it for communication. The difference between a four-frequency system and a two-frequency system is that the secondary system in a four-frequency system has to first check the presence of primary users before switching into a particular frequency channel. This was not the case in the two-frequency system. In the two-frequency system we assumed that the other band is always unoccupied at the time of switching as only a pair of primary users existed and thereby only a single band is always

unoccupied.

The following flow graph describes the four frequency cognitive system:

Flow graph here

The spectrum sensing is done by energy detection technique and it was required that a proper threshold be set for decision making. A number of readings were taken to decide the noise level, energy level when only primary users were active and also energy levels when both primary users and secondary users were active in the same band for a short duration of time. The threshold value depends on the power transmitted by the users and their distance from the USRP kit which is RF front for GNURadio. This distance dependency can be removed by setting the threshold quite lower than required so that even if the users move far away the decision making is not affected.

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