

## Estimation en traitement d'antenne

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## Sommaire

Introduction

Le traitement d'antenne et analyse spectrale

Les techniques d'estimation :

Méthodes de Fourier (limitations du non paramétrique)

Méthodes de sous espace (potentiel et faiblesse)

Maximum de vraisemblance (complexité)

Sélection de l'ordre

Erreur de modèle et calibration

Radioastronomie

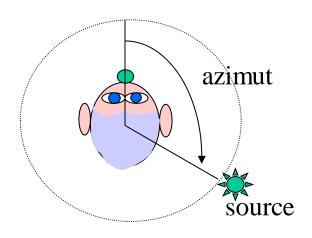
Géolocalisation

Localisation avec algorithmes parcimonieux

Application au cyclone Georges (complexité, robustesse, performances)



## - Array signal processing



### **Sensibility in Azimut:**

• phase and intensity

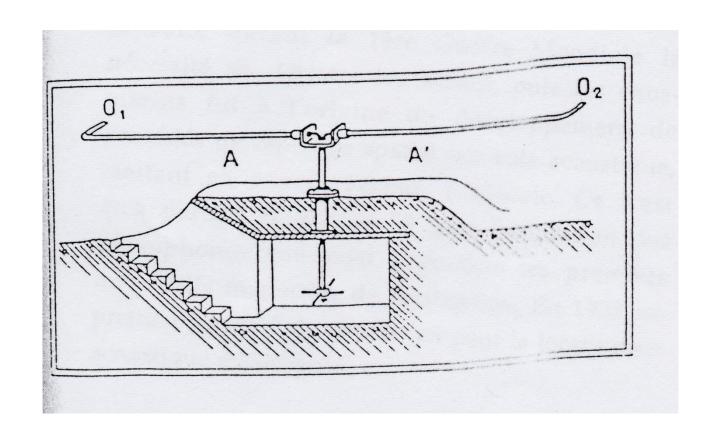


#### Systèmes et Applications des Technologies de l'Information et de l'Energie UMR - 8029



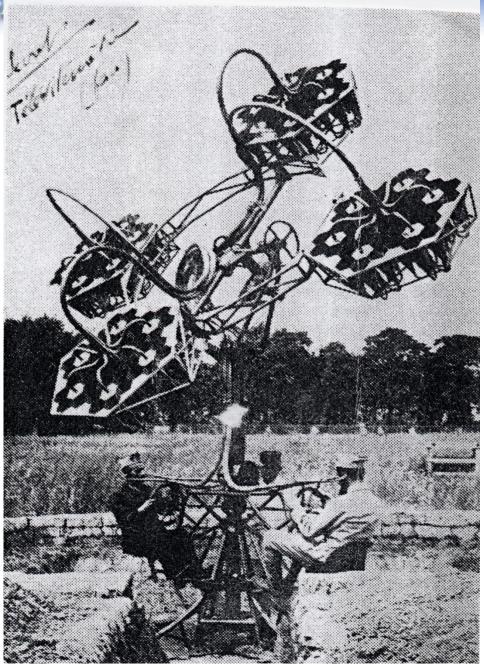






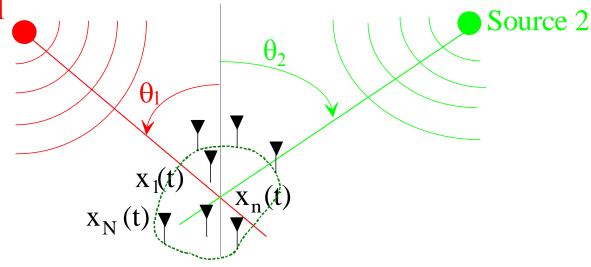


Systèmes et Applications des Technologies de l'



#### LE TRAITEMENT D'ANTENNE

### Source 1



#### **Estimation:**

- Localisation
- Détection
- Séparation
- Identification

$$\mathbf{x}(t) = \sum_{m=1}^{M} \mathbf{a}(\theta_m) \cdot s_m(t) + \mathbf{b}(t)$$

$$\mathbf{x}(t) = A(\theta) \cdot \mathbf{s}(t) + \mathbf{b}(t) = \mathbf{y}(t) + \mathbf{b}(t)$$



# Two Types of Beamformers

 Method 1: Single sensor with directional response due to reflector, aperture size, baffles, pipes, etc.



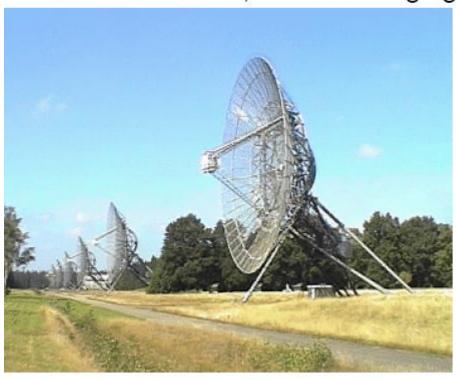
Green Bank Telescope, National Radio Astronomy Observatory, West Virginia.

100 m clear aperture. Largest fully steerable antenna in the world.



# Two Types of Beamformers (cont.)

 Method 2: Sensor arrays. Used in SONAR, RADAR, communications, medical imaging, radio astronomy, etc.

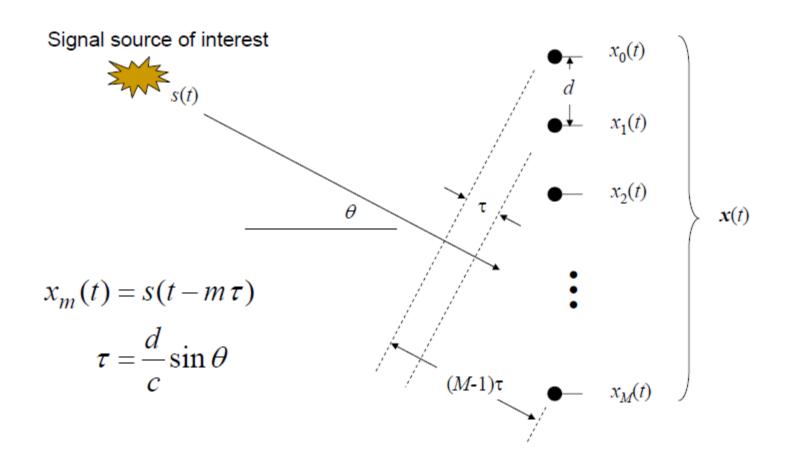




Line array of directional sensors
Westerbork Synthesis Array
Radio Telescope, (WSRT)
the Neterlands.

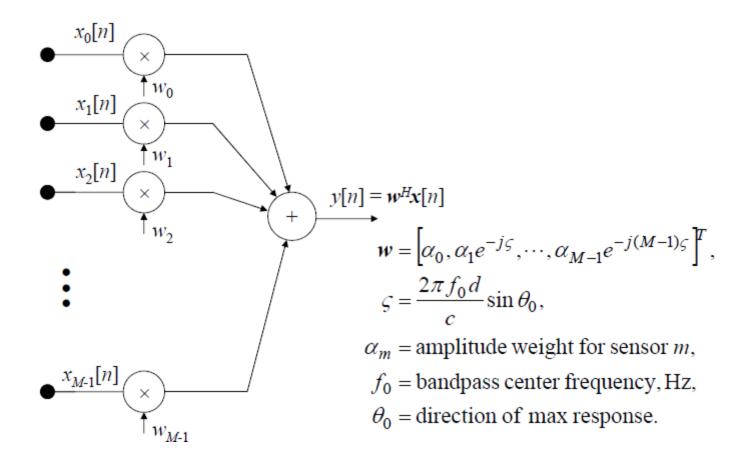


# The Uniform Line Array





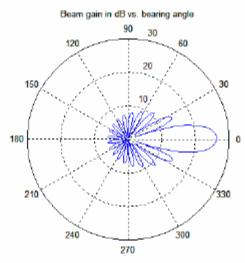
# Narrowband Phased Array



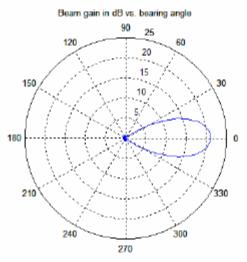


# Beamforming is **Spatial** Filtering

 Sensors in any wave propagation medium (acoustic, electromagnetic) can form a response pattern with higher sensitivity in desired directions.



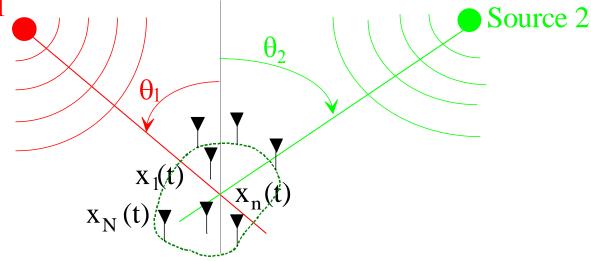
Pencil beam response, no windowing



Pencil beam response, Hamming window

#### LE TRAITEMENT D'ANTENNE





#### **Estimation:**

- Localisation
- Détection
- Séparation
- Identification

$$\mathbf{x}(t) = \sum_{m=1}^{M} \mathbf{a}(\theta_m) s_m(t) + \mathbf{b}(t)$$
$$= A(\boldsymbol{\theta}) \mathbf{s}(t) + \mathbf{b}(t)$$



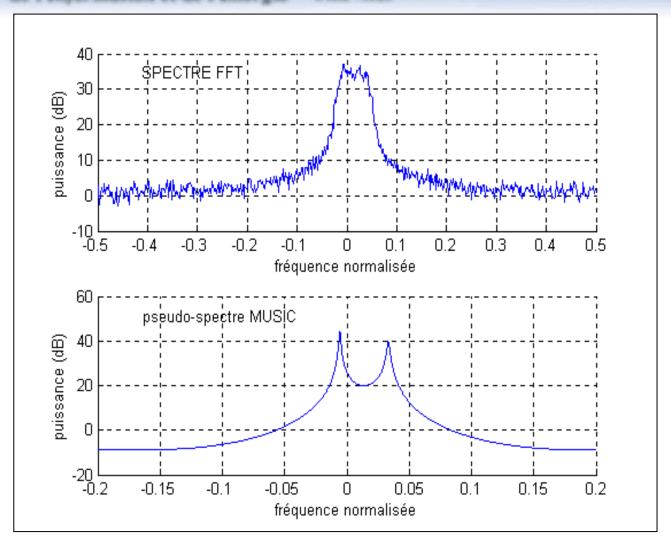
### Les méthodes de sous espace : la méthode MUSIC

$$\mathbf{x}\left(t\right)=\mathbf{y}(t)+\mathbf{b}(t)=\sum_{m=1}^{M}\mathbf{a}\left(\theta_{m}\right)s_{m}\left(t\right)+\mathbf{b}\left(t\right), \quad \ \mathbf{x}\left(t\right)\in\mathbb{C}^{N} \quad \text{et} \ \ \mathbf{y}(t)\in\mathbb{C}^{M} \qquad N>M$$

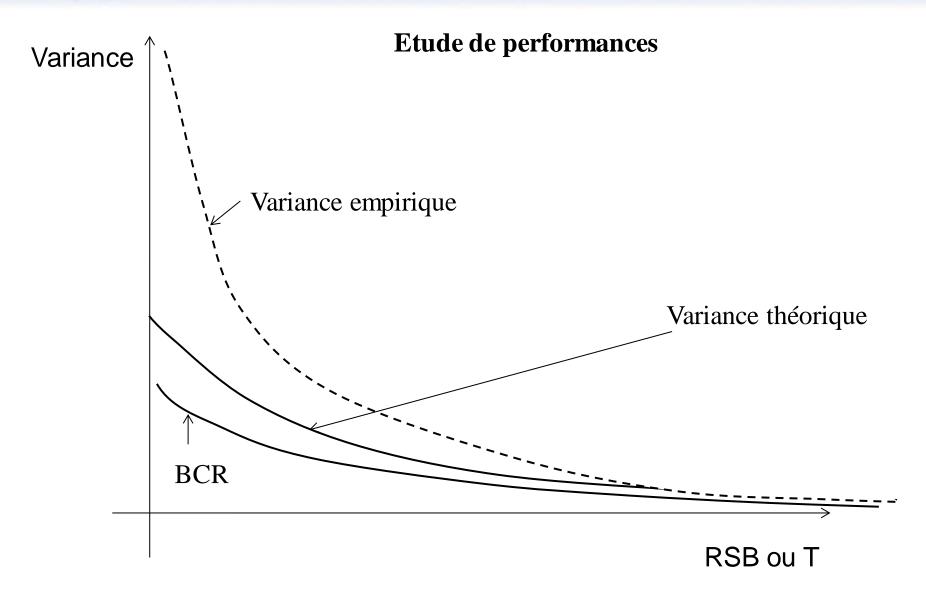
$$\hat{R}_{\mathbf{x}} = \sum_{t=1}^{T} \mathbf{x}(t) \mathbf{x}(t)^{H} = \hat{\mathbf{E}}_{S} \hat{\boldsymbol{\Lambda}}_{S} \mathbf{E}_{S}^{H} + \hat{\mathbf{E}}_{B} \hat{\boldsymbol{\Lambda}}_{B} \hat{\mathbf{E}}_{B}^{H}$$

$$R_{\mathbf{x}} = E\left[\mathbf{x}\left(t\right)\mathbf{x}\left(t\right)^{H}\right] = E\left[\stackrel{\wedge}{R}_{\mathbf{x}}\right] = \mathbf{E}_{S}\mathbf{\Lambda}_{S}\mathbf{E}_{S}^{H} + \mathbf{E}_{B}\mathbf{\Lambda}_{B}\mathbf{E}_{B}^{H}$$

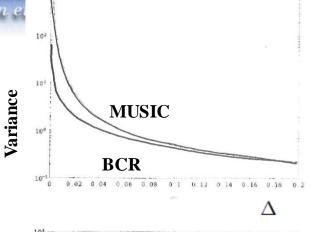
#### Systèmes et Applications des Technologies de l'Information et de l'Energie UMR - 8029



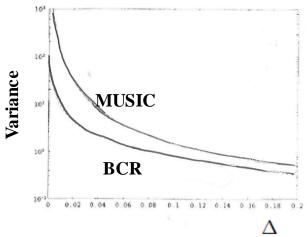




**BCR: Borne Cramer Rao** 



$$\rho = 0.5$$

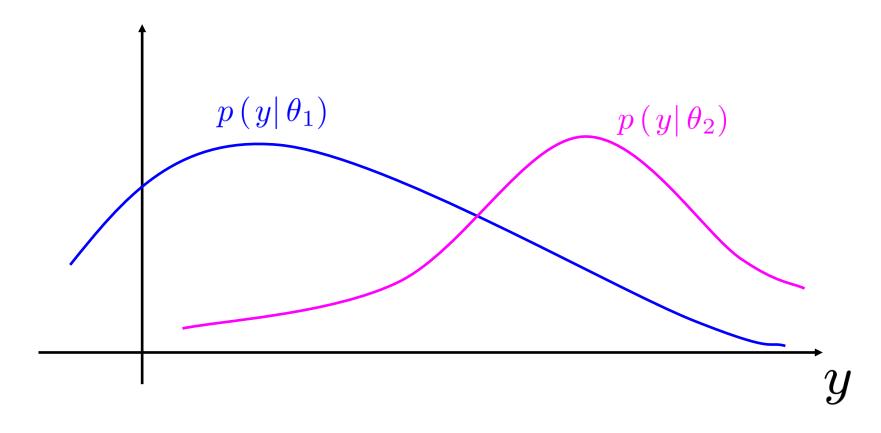


$$\rho = 0.9$$

$$\rho = 0.99$$

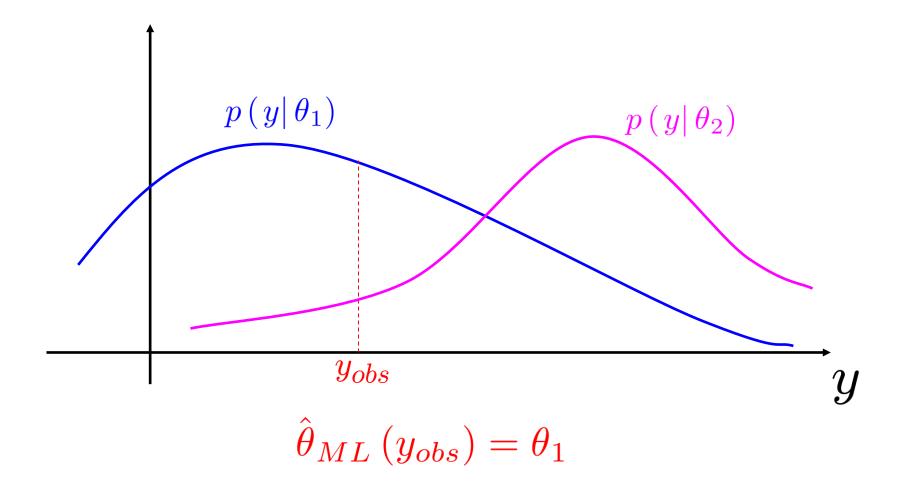


## Maximum de vraisemblance : principe



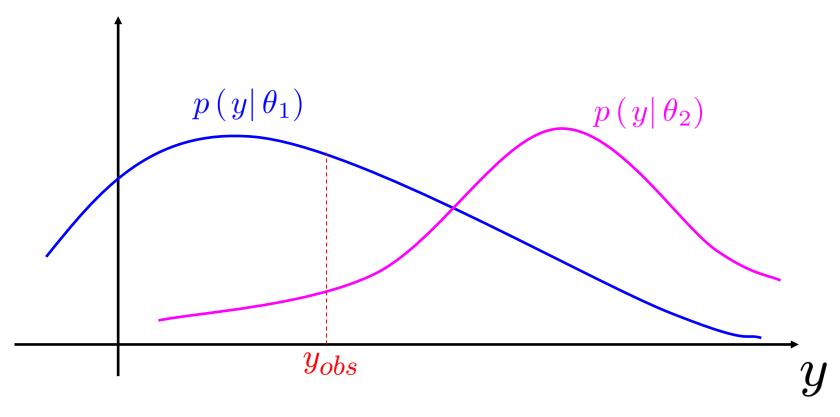


## Maximum de vraisemblance: principe





### Maximum de vraisemblance: principe



Généralisation

$$\hat{\boldsymbol{\theta}}_{ML}\left(\mathbf{y}_{obs}\right) = \arg\max_{\boldsymbol{\theta}} p\left(\left.\mathbf{y}_{obs}\right|\boldsymbol{\theta}\right)$$



#### Maximum de vraisemblance en traitement d'antenne

$$\mathbf{y}(t) = \mathbf{A}(\boldsymbol{\theta}) \mathbf{s}(t) + \mathbf{b}(t)$$
  $t = 1...N$ 

Hypothèses

$$\mathbf{s}\left(t
ight) \sim \mathcal{N}_{\mathbb{C}}\left(\mathbf{0}, \mathbf{\Sigma}_{s}
ight)$$
 aléatoires

Paramètres à estimer

$$oldsymbol{ heta}, \sigma^2, oldsymbol{\Sigma}_s$$

Maximum de Vraisemblance stochastique

$$s(t)$$
  $t = 1...N$  déterministes

Paramètres à estimer

$$\boldsymbol{\theta}, \sigma^2, \mathbf{s}_1, ..., \mathbf{s}_N$$

Maximum de Vraisemblance déterministe



### Maximum de Vraisemblance stochastique

$$\mathbf{y}\left(t
ight) = \mathbf{A}\left(oldsymbol{ heta}
ight)\mathbf{s}\left(t
ight) + \mathbf{b}\left(t
ight) \quad \mathbf{s}\left(t
ight) \sim \mathcal{N}_{\mathbb{C}}\left(\mathbf{0}, \mathbf{\Sigma}_{s}
ight)$$
 $\mathbf{y}\left(t
ight) \sim \mathcal{N}_{\mathbb{C}}\left(\mathbf{0}, \mathbf{\Sigma}_{y}
ight)$ 

$$\mathbf{\Sigma}_{y} = \mathbf{A}\left(\boldsymbol{\theta}\right) \mathbf{\Sigma}_{s} \mathbf{A}^{H}\left(\boldsymbol{\theta}\right) + \sigma^{2} \mathbf{I}$$

$$\begin{cases} \hat{\boldsymbol{\theta}}_{MVS} = \arg\min_{\boldsymbol{\theta}} \left| \mathbf{A} \left( \boldsymbol{\theta} \right) \hat{\mathbf{R}}_{s} \mathbf{A}^{H} \left( \boldsymbol{\theta} \right) + \hat{\sigma}^{2} \mathbf{I} \right| \\ \hat{\sigma}^{2} = \frac{1}{M - K} Trace \left\{ \mathbf{\Pi}_{\mathbf{A}}^{\perp} \left( \boldsymbol{\theta} \right) \hat{\boldsymbol{\Sigma}}_{y} \right\} \\ \hat{\mathbf{R}}_{s} = \mathbf{A}^{\dagger} \left( \hat{\boldsymbol{\Sigma}}_{y} - \hat{\sigma}^{2} \mathbf{I} \right) \mathbf{A}^{\dagger H} \end{cases}$$

Borne de Cramér-Rao Stochastique

$$\mathbf{BCRS}\left(\boldsymbol{\theta}_{0}\right) = \frac{\sigma^{2}}{2N} \left[ \operatorname{Re}\left\{\mathbf{H}\left(\boldsymbol{\theta}_{\mathbf{0}}\right) \odot \boldsymbol{\Sigma}_{s} \mathbf{A}^{H}\left(\boldsymbol{\theta}_{0}\right) \boldsymbol{\Sigma}_{y}^{-1} \mathbf{A}\left(\boldsymbol{\theta}_{\mathbf{0}}\right) \boldsymbol{\Sigma}_{s} \right\} \right]^{-1}$$



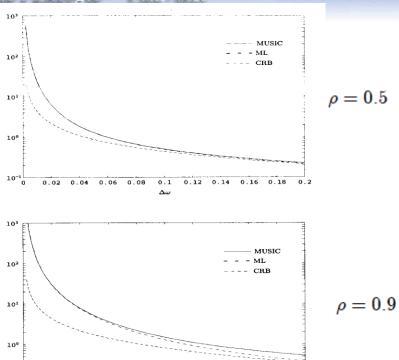
### Maximum de Vraisemblance déterministe

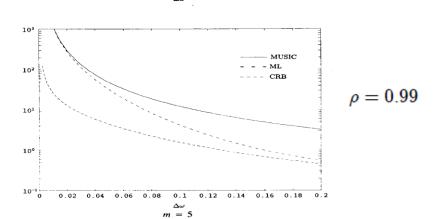
$$\begin{split} \mathbf{y}\left(t\right) &= \mathbf{A}\left(\boldsymbol{\theta}\right)\mathbf{s}\left(t\right) + \mathbf{b}\left(t\right) \quad \mathbf{s}\left(t\right) \quad t = 1...N \\ \mathbf{y}\left(t\right) &\sim \mathcal{N}_{\mathbb{C}}\left(\mathbf{A}\left(\boldsymbol{\theta}\right)\mathbf{s}\left(t\right), \boldsymbol{\Sigma}_{y}\right) \\ \boldsymbol{\Sigma}_{y} &= \sigma^{2}\mathbf{I} \\ \begin{cases} \hat{\boldsymbol{\theta}}_{MVD} = \arg\min_{\boldsymbol{\theta}} Trace\left\{\boldsymbol{\Pi}_{\mathbf{A}}^{\perp}\left(\boldsymbol{\theta}\right)\hat{\boldsymbol{\Sigma}}_{y}\right\} \\ \hat{\boldsymbol{\Sigma}}_{y} &= \frac{1}{N}\sum_{t=1}^{N}\mathbf{y}\left(t\right)\mathbf{y}^{H}\left(t\right) \end{cases} \end{split}$$

Borne de Cramér-Rao Déterministe

$$\mathbf{BCRD}\left(\boldsymbol{\theta}_{0}\right) = \frac{\sigma^{2}}{2N} \left[ \operatorname{Re}\left\{ \mathbf{H}\left(\boldsymbol{\theta}_{\mathbf{0}}\right) \odot \hat{\boldsymbol{\Sigma}}_{s} \right\} \right]^{-1}$$

#### Systèmes et Applications des Technologies de l'Information et de l'Energie UMR - 8029



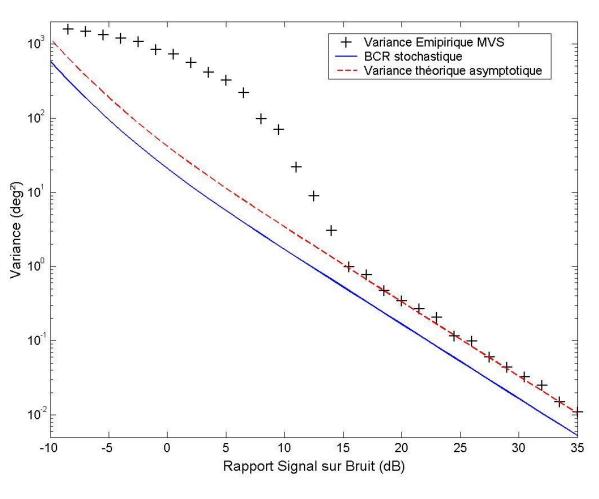


0.02 0.04 0.06 0.08 0.1 0.12 0.14 0.16 0.18 0.2



## Performances asymptotiques

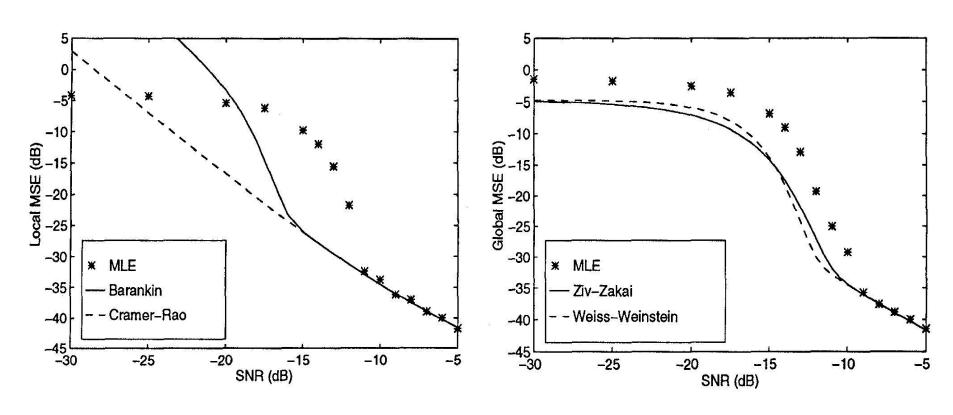
### Cas particulier : scénario une source



- ALU  $(\lambda/2)$ : M = 4 capteurs
- -N = 2 observations
- 1 source 0°
- 1000 Monte Carlos



### **BORNES MINIMALES**



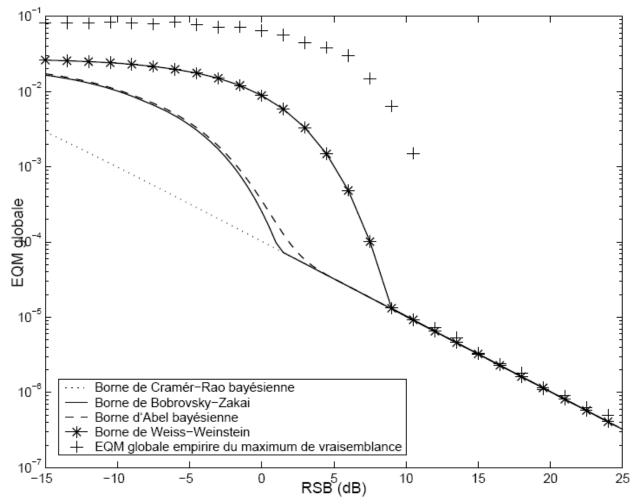
Paramètres Déterministes

Paramètres aléatoires



## Bornes minimales en traitement du signal

Application à la synchronisation (analyse spectrale)



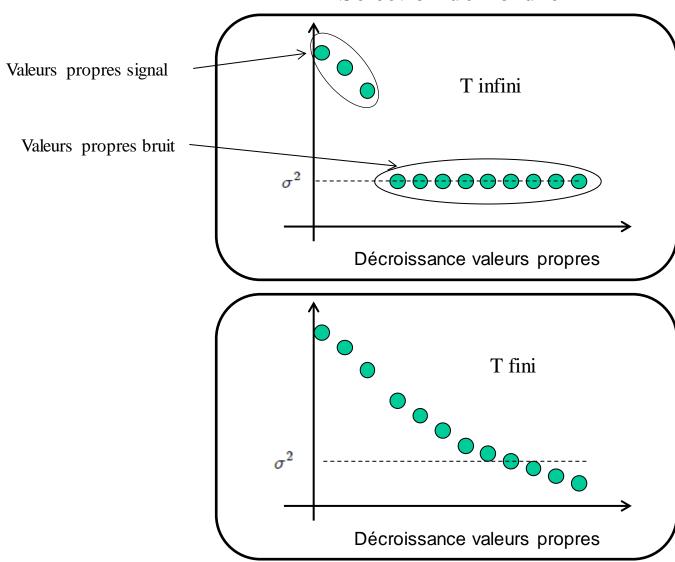
20 observations

Maximum de Vraisemblance

EQM Globale



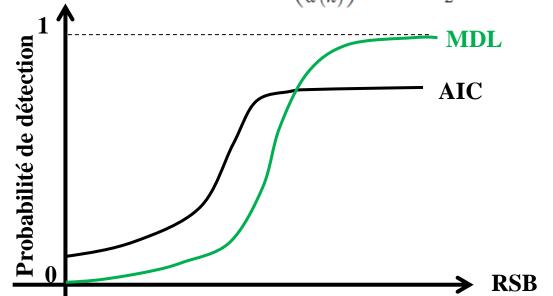
#### Sélection de l'ordre



## Sélection fondée sur la théorie de l'information

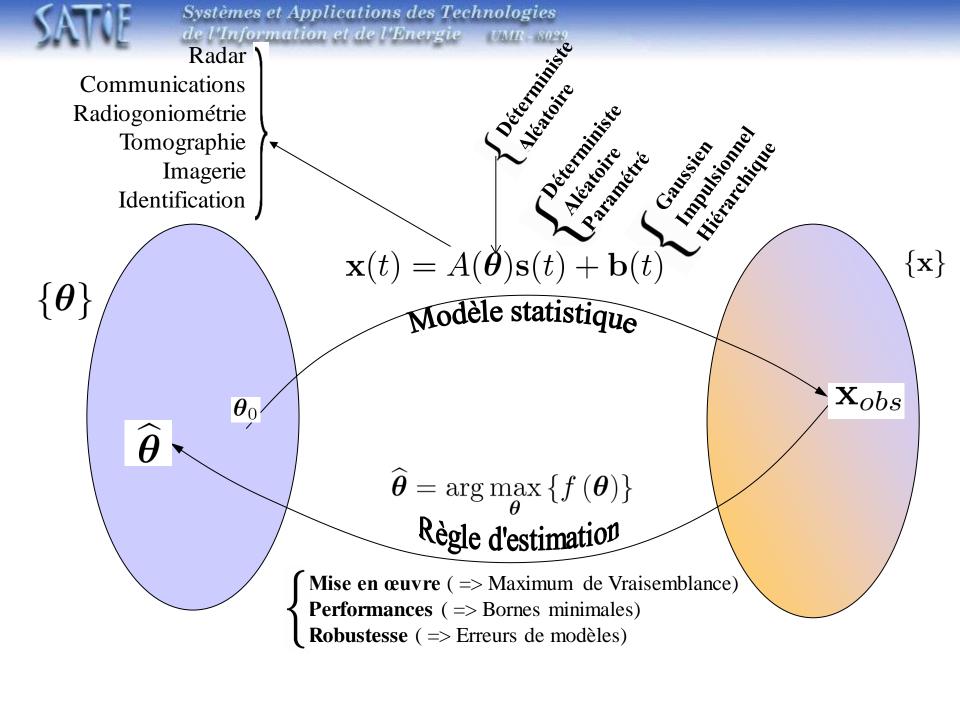
**Akaike(74)** 
$$AIC(k) = -LV(k) + NPL = -\log\left(\frac{g(k)}{a(k)}\right)^{T(N-k)} + k(2N-k)$$

**Rissanen et Schwartz(78)** 
$$MDL(k) = -\log\left(\frac{g(k)}{a(k)}\right)^{T(N-k)} + \frac{1}{2}k(2N-k)\log(T)$$

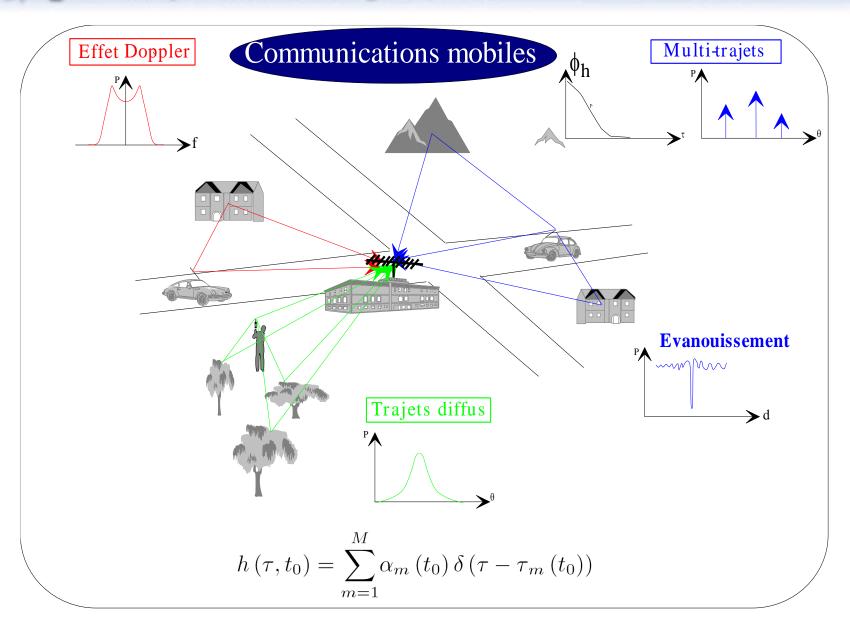


Réglage de la probabilité de fausse alarme —— Théorie de la décision

Rapport de vraisemblance 
$$RV\left(k\right) = \left(\frac{a\left(k\right)}{g\left(k\right)}\right)^{T\left(N-k\right)} \lessgtr s\left(k\right)$$





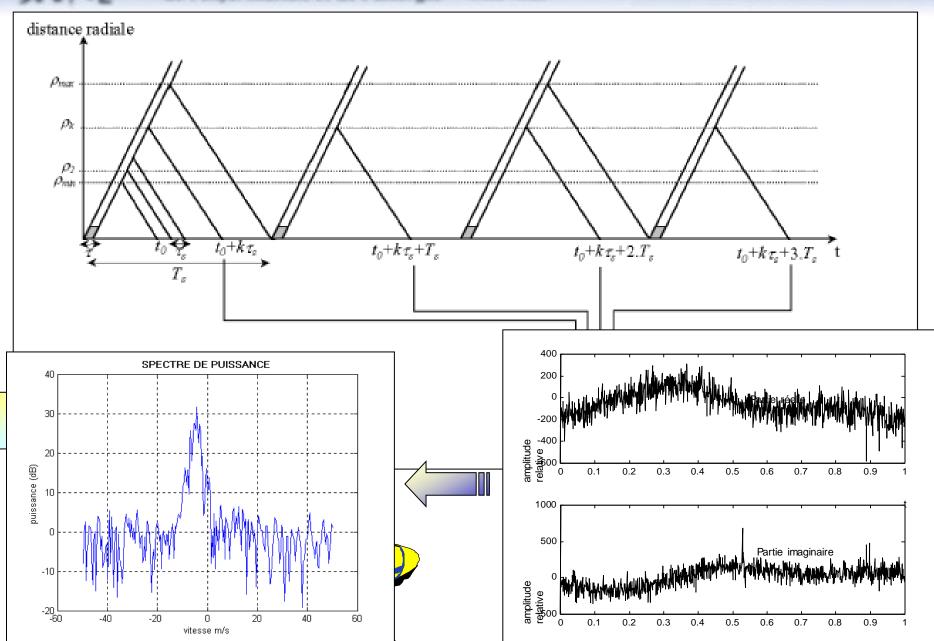




## Estimation d'échos Doppler par radars strato-troposphériques Le cyclone Georges

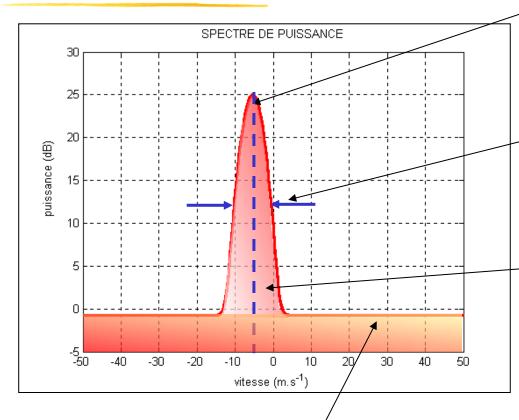








#### Paramètres d'intérêt



#### Puissance de bruit

Paramètre de nuisance

#### **Vent moyen (Doppler)**

Renseigne sur la valeur et le sens du vent radial

#### Ecart-type de l'écho

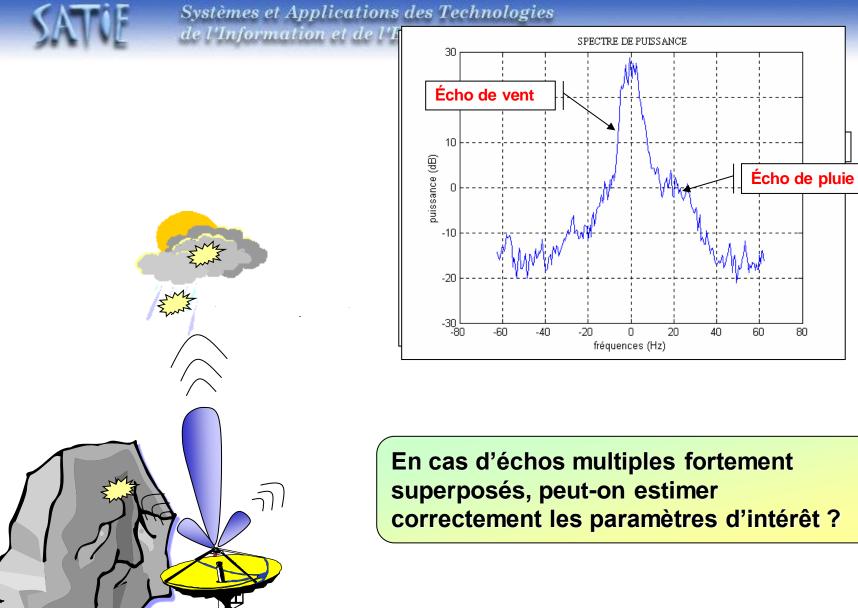
Mesure la dispersion des vitesses par rapport à la vitesse moyenne

#### Puissance de l'écho

Donne une estimation de la réflectivité qui est reliée à l'intensité de la turbulence

$$P_0$$
 ;  $v_0 = -\frac{\lambda}{2} f_0$  ;  $\sigma_0$ 

3 paramètres d'intérêt à estimer



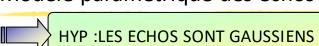
En cas d'échos multiples fortement correctement les paramètres d'intérêt?

80

#### Systèmes et Applications des Technologies

de l'Information et de l'Energie

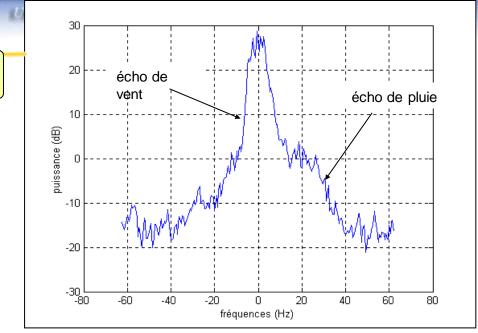
### Modèle paramétrique des échos radar



Spectre de puissance

$$P_{s}(f) = \sum_{i=1}^{N} S_{i}(f) + \sigma_{b}^{2},$$

$$S_i(f) = \frac{P_i}{\sqrt{2\pi}.\sigma_i} e^{-\frac{1}{2}\left(\frac{f-f_i}{\sigma_i}\right)^2}$$



$$\mathbf{x} = \mathbf{y} + \mathbf{b}$$
  $\begin{cases} \mathbf{y} \text{ vecteur aléatoire gaussien centré} \\ \mathbf{x} = \mathbf{y} + \mathbf{b} \end{cases}$ 

**b** bruit blanc complexe gaussien indépendant de **y**, de variance  $\sigma_b^2$ 

$$\mathbf{R}_{x}(\mathbf{\mu}) = E(\mathbf{x}\mathbf{x}^{H}) = \sum_{i=1}^{N} \mathbf{R}_{y_{i}}(P_{i}, \omega_{i}, \sigma_{i}^{2}) + \sigma_{b}^{2}\mathbf{I}$$

$$\mathbf{R}_{\mathbf{y}_{i}}(P_{i}, \omega_{i}, \sigma_{i}^{2}) = P_{i}.\mathbf{A}(\omega_{i})\mathbf{B}(\sigma_{i}^{2})\mathbf{A}^{H}(\omega_{i})$$

$$\mathbf{A}(\omega_i) = diag(1, e^{j\omega_i T_s}, ..., e^{j(m-1)\omega_i T_s})$$

$$\mathbf{B}_{k,l}(\sigma_i^2) = e^{-2\pi^2 \sigma_i^2 (k-l)^2 T_s^2}$$

Vecteur paramètre

$$\mathbf{\mu} = \left[ P_1 \ \omega_1 \ \sigma_1^2 \dots P_N \ \omega_N \ \sigma_N^2 \ \sigma_b^2 \right]^T$$



Systèmes et Applications des Technologies de l'Information et de l'Energie

#### Estimation conjointe des moments spectraux



#### **EXPLOITER LA STRUCTURE DE**



Facile à calculer



Porte toute l'information statistique



Elle dépend des paramètres d'intérêt

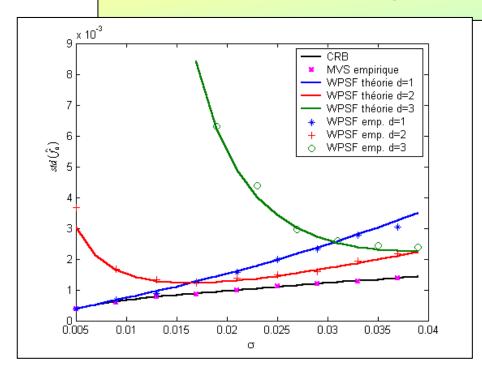
#### **DEUX APPROCHES RETENUES**

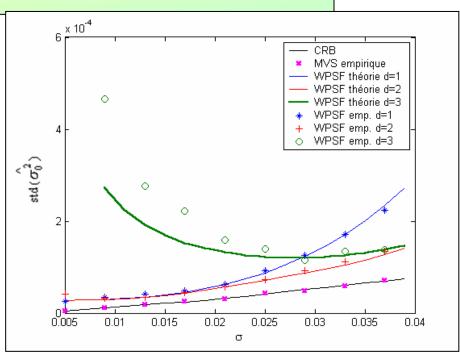
- ESTIMATEUR DU MAXIMUM DE VRAISEMBLANCE (MV)
- GENERALISATION MULTI-DIMENSIONNELLE DES METHODES DE SOUS-ESPACE

# SAVIE

### Simulations – cas de réalisations indépendantes

# Performances empiriques et théoriques des estimateurs WPSF et MVS dans le cas d'un écho gaussien



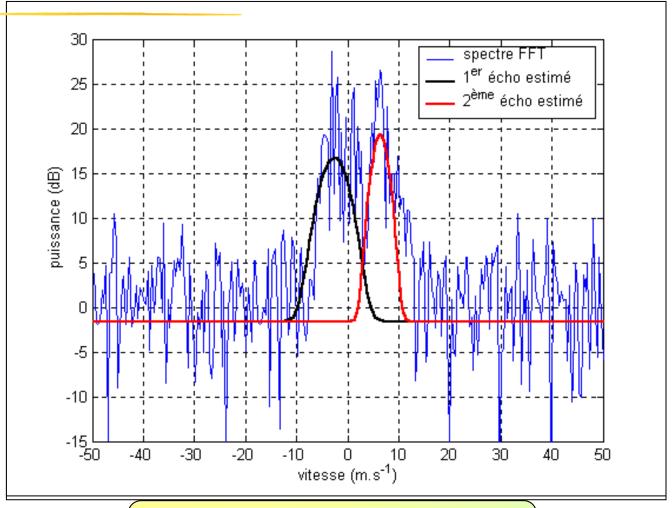




Choix de *d* crucial

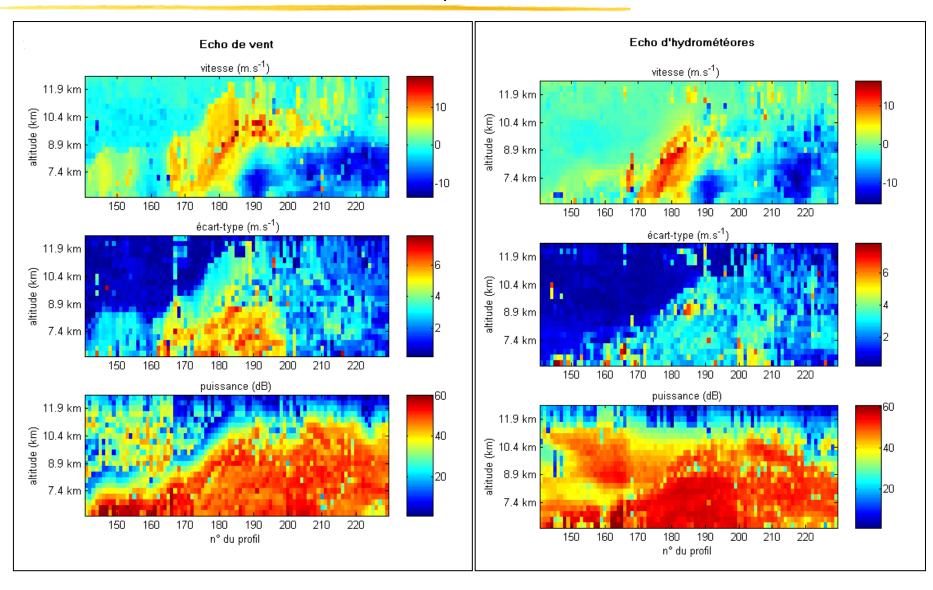
Non optimal

Validation sur données réelles

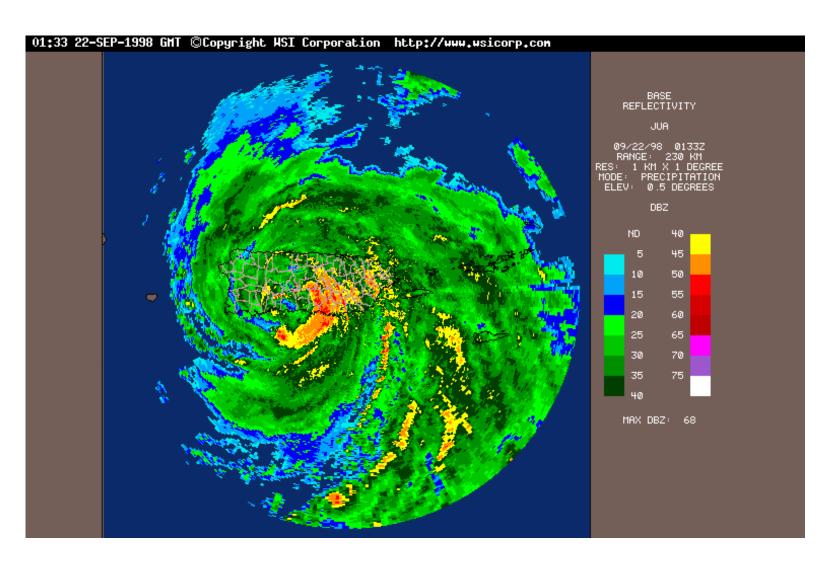


Estimation MVS d'un signal UHF avec *M*=16384, *N*<sub>coh</sub>=8 et *N*<sub>inc</sub>=8

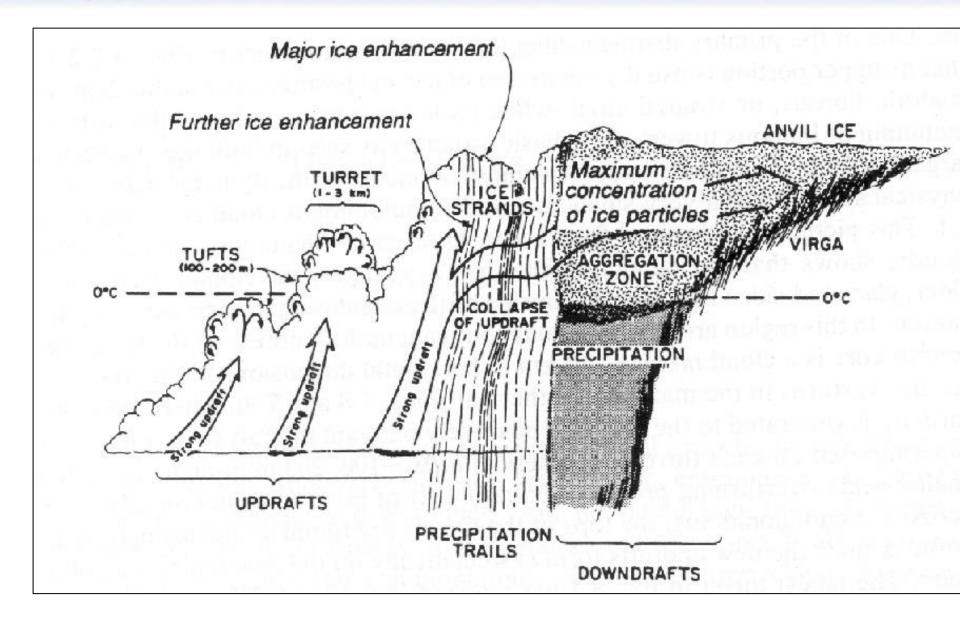
#### Restitution MVS des échos UHF de vent et de pluie



### Validation sur les données réelles du cyclone Georges









## Issues

- Performance
- Computational complexity
- Robustness



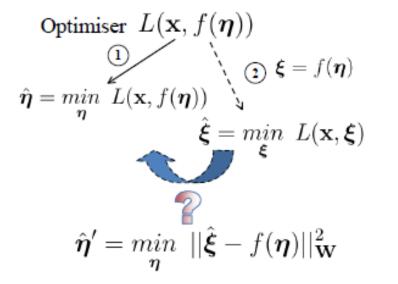
# Radar: Space Surveillance



- 5184 transmit elements
- 4660 receive elements



### The extended invariance principle (EXIP)



- f, fonction injective (dim(ξ)≥dim(η))
- Choix naturel: MCP.
- Choix du poids W?
- ► Comportement de  $\hat{\eta}'$ ?

### EXIP [Stoica et al., 1989]

Si 
$$f(\hat{\eta}) \to \hat{\xi}$$
, alors  $\hat{\eta}'$  converge en probabilité vers  $\hat{\eta}$  avec  $\mathbf{W} = E\left[\frac{\partial^2 L(\mathbf{x}, \boldsymbol{\xi})}{\partial \boldsymbol{\xi} \partial \boldsymbol{\xi}^T}\right]$ .



### Estimation avec ajustement d'une matrice de covariance structurée

$$\mathbf{x}(t) = A(\boldsymbol{\theta}) \mathbf{s}(t) + \mathbf{b}(t)$$
  $t = 1, ..., T$ 

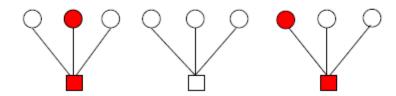
$$\hat{R} = \sum_{t=1}^{T} \mathbf{x}(t) \mathbf{x}(t)^{H}$$

$$\begin{cases}
\hat{r} = vec(\hat{R}) \Rightarrow \hat{\gamma} = J\hat{r} \\
\gamma(\theta) = Jr(\theta)
\end{cases} \implies \hat{\theta} = \arg\min_{\theta} (\hat{\gamma} - \gamma(\theta))^{H} C^{-1} (\hat{\gamma} - \gamma(\theta))$$



### Blood tests (Dorfman 1943)

N individuals. Probability of infection p << 1. Pool the tests:



N/k groups of k persons, two-stage testing

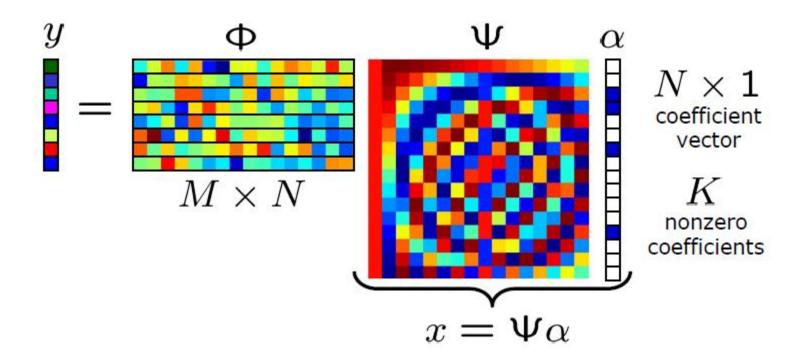
$$\overline{T} = \frac{N}{k} + \left[1 - (1-p)^k\right] \frac{N}{k} k$$

Minimal for  $k \sim \frac{1}{\sqrt{p}}$  , gives  $\overline{T} \sim 2\sqrt{p}N$ 



## CS Paradigm

• K-term quality from just  $K \log(N)$  measurements



· Robust and widely applicable