## Notation

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## 1 Font notation

$a, b, c, \ldots, A, B, C, \ldots$	Scalars
$a, b, c, \dots$	Vectors
$\overline{\mathbf{A},\mathbf{B},\mathbf{C},\dots}$	Matrices
$\mathcal{A}, \mathcal{B}, \mathcal{C}, \dots$	Tensors
$A, B, C, \ldots, A, B, C, \ldots, A, B, C, \ldots$	Sets

# 2 Signals and functions

## 2.1 Time indexing

x(t)	Continuous-time $t$
$x[n],x[k],x[m],x[i],\ldots$	Discrete-time $n, k, m, i, \ldots$ (parenthe-
$x_n, x_k, x_m, x_i, \dots$	sis should be adopted only if there
$x(n), x(k), x(m), x(i), \dots$	are no continuous-time signals in the
	context to avoid ambiguity)
$x[((n-m))_N][32], x((n-m))_N[26]$	Circular shift in $m$ samples within a
	N-samples window

## 2.2 Common signals

$\delta(t)$	Delta function
$\delta[n], \delta_{i,j}$	Kronecker function $(n = i - j)$
h(t), h[n]	Impulse response (continuous and
	discrete time)
$\tilde{x}[n], \tilde{x}(t)$	Periodic discrete- or continuous-time
	signal
$\hat{x}[n], \hat{x}(t)$	Estimate of $x[n]$ or $x(t)$
$\dot{x}[m]$	Interpolation of $x[n]$

#### 2.3 Common functions

$\mathcal{O}(\cdot), O(\cdot)$	Big-O notation
$\Gamma(\cdot)$	Gamma function
$Q(\cdot)$	Quantization function
$\operatorname{sgn}(\cdot)$	Signum function
$\tanh(\cdot)$	Hyperbolic tangent function
$I_{\alpha}(\cdot)$	Modified Bessel function of the first
	kind and order $\alpha$

n	Binomial coefficient
k	binomiai coenicient

## 2.4 Operations and symbols

$f:A\to B$	A function $f$ whose domain is $A$ and
	codomain is $B$
$\mathbf{f}:A \to \mathbb{R}^n$	A vector-valued function $\mathbf{f}$ , i.e., $n \geq 2$
$f^n, x^n(t), x^n[k]$	<i>n</i> th power of the function $f$ , $x[n]$ or
, , , , , , , , , , , , , , , , , , , ,	x(t)
$f^{(n)}, x^{(n)}(t)$	nth derivative of the function $f$ or
	x(t)
$f', f^{(1)}, x'(t)$	1th derivative of the function $f$ or
	x(t)
$f'', f^{(2)}, x''(t)$	2th derivative of the function $f$ or
	x(t)
$\underset{x \in \mathcal{A}}{\operatorname{argmax}} f(x)$	Value of $x$ that minimizes $x$
arg min f(x)	Value of $x$ that minimizes $x$
$f(\mathbf{x}) = \inf_{\mathbf{y} \in \mathcal{A}} g(\mathbf{x}, \mathbf{y})$	Infimum, i.e., $f(\mathbf{x}) =$
$\mathbf{y} \in \mathcal{A}$	$\min \{ g(\mathbf{x}, \mathbf{y}) \mid \mathbf{y} \in \mathcal{A} \land (\mathbf{x}, \mathbf{y}) \in \text{dom}(g) \},$
	which is the greatest lower bound of
	this set $[10]$
$f(\mathbf{x}) = \sup g(\mathbf{x}, \mathbf{y})$	Supremum, i.e., $f(\mathbf{x}) =$
$\mathbf{y} \in \mathcal{A}$	$\max \{ g(\mathbf{x}, \mathbf{y}) \mid \mathbf{y} \in \mathcal{A} \land (\mathbf{x}, \mathbf{y}) \in \text{dom}(g) \}.$
	which is the least upper bound of
	this set $[10]$
$f \circ g$	Composition of the functions $f$ and
	g
*	Convolution (discrete or continuous)
⊛ [17], N [32]	Circular convolution

## 2.5 Digital signal processing

$T_s[26], T[32]$	Sampling period
$f_s, F_s[26]$	Sampling frequency (in Hz or sam-
	ples per secod [26, chapter 3]), i.e.,
	$1/T_s$

Continuous linear frequency (in Hz). Apparently, there is no notation for the discrete linear frequency, we use $\omega$ only. However, in [26], the uppercase letters $F$ and $\Omega$ are used to denote the continuous-time frequency, while the lowercase $f$ and $\omega$ denote the discrete-time frequency (Oppenheim [32] does not do it!) $\Omega$ [26] Continuous angular frequency (in rad/s), that is, $2\pi f$ . $\Omega_s$ Sampling frequency (in rad/s), i.e., $2\pi f_s$ $\omega$ Discrete angular frequency, i.e., $\Omega T_s$ [26, eq (3.27)]. As $\omega$ is also used to denote continuous angular frequency outside the DSP context, it is always convenient to state that it denotes the discrete frequency when it does $W_N$ Twiddle factor, $e^{-j\frac{2\pi}{N}}$ [26] $N$ Number of samples in the DFT/FFT $R_N[n]$ Rectangular window used to cut off the discrete sequences [26] $\Omega_N$ [32], $B$ One-sided effective bandwidth of the continuous-time signal spectrum $\omega_s$ [26] Stop frequency
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continuous-time signal spectrum
$\omega_s$ [26] Stop frequency
$\omega_p$ [26] Pass frequency
$\Delta\omega$ [26] $\omega_s - \omega_p$
$\omega_c$ [26] Cutoff frequency
s(t) Impulse train
$\operatorname{gdr}\left[H(e^{j\omega})\right]$ [32] Group delay of $H(e^{j\omega})$
$\angle H(e^{j\omega})$ [32] Phase response of $H(e^{j\omega})$
$ H(e^{j\omega}) $ [32] Magnitude (or gain) of $H(e^{j\omega})$
$x_c(t)$ [32], $x(t)$ Continuous-time signal
$x_s(t)$ Sampled version of $x(t)$ , i.e., $x(t)s(t)$
$x_r(t)$ Reconstruction of $x(t)$ from interpo-
lation
$\tilde{x}[n]$ Periodic extension of the the aperi-
odic signal $x[n]$

#### 2.6 Transformations

$\mathcal{F}\left\{\cdot\right\}$ [32, section 2.9]	Fourier transform (FT)

OTFT), Discrete Fourier Trans-
rm (DFT), Discrete Fourier Series
OFS), respectively
aplace transform
transform
ilbert transform of $x(t)$ or $x[n]$
aplace transform of $x(t)$
ourier transform (FT) (in linear fre-
uency, Hz) of x(t)
ourier transform (FT) (in angular
equency, rad/sec) of $x(t)$
iscrete-time Fourier transform
OTFT) of $x[n]$
iscrete Fourier transform (DFT) or
st Fourier transform (FFT) of $x[n]$ ,
even the Fourier series (FS) of the
eriodic signal $x(t)$
iscrete Fourier series (DFS) of $\tilde{x}[n]$
transform of $x[n]$

# 3 Probability, statistics, and stochastic processes

## 3.1 Operators and symbols

$\mathbf{E}\left[\cdot\right],\mathbf{E}\left[\cdot\right]\left[31\right],E\left[\cdot\right],\mathbb{E}\left[\cdot\right]\left[16\right]$	Statistical expectation operator
$E_{u}[\cdot], E_{u}[\cdot][31], E_{u}[\cdot], \mathbb{E}_{u}[\cdot]$	Statistical expectation operator with
	respect to $u$
$\overline{\langle \cdot \rangle}$	Ensemble average
var [·] [31], VAR[·] [9, 25, 30, 34]	Variance operator
$\operatorname{var}_{u}\left[\cdot\right]\left[\cdot\right], \operatorname{VAR}_{u}\left[\cdot\right]$	Variance operator with respect to $u$
$cov[\cdot], COV[\cdot]$	Covariance operator [9]
$\operatorname{cov}_{u}[\cdot], \operatorname{COV}_{u}[\cdot]$	Covariance operator with respect to
	и
$\mu_x$	Mean of the random variable $x$
$\mu_x, m_x$	Mean vector of the random variable
	x [11]
$\mu_n$	nth-order moment of a random vari-
	able
$\frac{\sigma_{x}^{2}, \kappa_{2}}{\mathcal{K}_{x}, \mu_{4}}$	Variance of the random variable $x$
$\mathcal{K}_x, \mu_4$	Kurtosis (4th-order moment) of the
	random variable $x$

$\kappa_n$	nth-order cumulant of a random vari-
	able
$\rho_{x,y}$	Pearson correlation coefficient be-
	tween $x$ and $y$
$a \sim P$	Random variable $a$ with distribution
	P
$\overline{\mathcal{R}}$	Rayleigh's quotient

## 3.2 Stochastic processes

PS: Estimated terms come with a hat above of it

$r_{\scriptscriptstyle X}( au)$ [31], $R_{\scriptscriptstyle X}( au)$	Autocorrelation function of the signal
	x(t) or $x[n]$
$S_x(f), S_x(j\omega)$	Power spectral density (PSD) of $x(t)$
	in linear $(f)$ or angular $(\omega)$ frequency
$S_{x,y}(f), S_{x,y}(j\omega)$	Cross PSD of $x(t)$ and $y(t)$ in linear
	or angular $(\omega)$ frequency
$R_{x}$	(Auto)correlation matrix of $\mathbf{x}(n)$
$r_{x,d}(\tau), R_{x,d}(\tau)$	Cross-correlation between $x[n]$ and
	d[n] or $x(t)$ and $d(t)$ [31]
$\overline{R_{xy}}$	Cross-correlation matrix of $\mathbf{x}(n)$ and
	$\mathbf{y}(n)$
$\mathbf{r}_{xd} [24],  \mathbf{p}_{xd} [16]$	Cross-correlation vector between
	$\mathbf{x}(n)$ and $d(n)$
$c_{x}(\tau), C_{x}(\tau)$	Autocovariance function of the signal
	x(t) or $x[n]$ [31]
$\mathbf{C}_{\mathbf{x}}, \mathbf{K}_{\mathbf{x}}, \mathbf{\Sigma}_{\mathbf{x}}, \operatorname{cov}\left[\mathbf{x}\right]$	(Auto)covariance matrix of <b>x</b> [9, 25,
	30, 34, 41
$\tilde{\mathbf{C}}_{\mathbf{x}}[34]$	Pseudocovariance matrix of $\mathbf{x}$
$c_{xy}(\tau), C_{xy}(\tau)$	Cross-covariance function of the sig-
	$\operatorname{nal} x(t) \text{ or } x[n] [31]$
$C_{xy}, K_{xy}, \Sigma_{xy}$	Cross-covariance matrix of <b>x</b> and <b>y</b>

#### 3.3 Functions

$Q(\cdot)$	Q-function, i.e., $P[\mathcal{N}(0,1) > x]$ [34]
$\operatorname{erf}(\cdot)$	Error function [34]
$\operatorname{erfc}(\cdot)$	Complementary error function i.e.,
	$\operatorname{erfc}(x) = 2Q(\sqrt{2}x) - \operatorname{erf}(x)$ [34]
P[A]	Probability of the event or set $A$ [30]

$p(\cdot), f(\cdot)$	Probability density function (PDF)
	or probability mass function (PMF)
	[30]
$p(x \mid A)$	Conditional PDF or PMF [30]
$F(\cdot)$	Cumulative distribution function
	(CDF)
$\Phi_{X}(\omega), M_{X}(j\omega), E\left[e^{j\omega X}\right]$	First characteristic function (CF) of
	x [34, 40]
$M_X(t), \Phi_X(-jt), E\left[e^{tx}\right]$	Moment-generating function (MGF)
	of $x [34, 40]$
$\Psi_{x}(\omega), \ln \Phi_{x}(\omega), \ln E\left[e^{j\omega x}\right]$	Second characteristic function
$K_x(t), \ln E\left[e^{tx}\right], \ln M_x(t)$	Cumulant-generating function
	(CGF) of $x$ [25]

## 3.4 Distributions

$\mathcal{N}(\mu,\sigma^2)$	Gaussian distribution of a random
	variable with mean $\mu$ and variance $\sigma^2$
$\mathcal{CN}(\mu, \sigma^2)$	Complex Gaussian distribution of a
	random variable with mean $\mu$ and
	variance $\sigma^2$
$\mathcal{N}(\mu, \Sigma)$	Gaussian distribution of a vector ran-
	dom variable with mean $\mu$ and co-
	variance matrix $\Sigma$
$\mathcal{CN}(oldsymbol{\mu},oldsymbol{\Sigma})$	Complex Gaussian distribution of a
	vector random variable with mean $\mu$
	and covariance matrix $\Sigma$
$\frac{\mathcal{U}(a,b)}{\chi^2(n),\chi^2_n}$	Uniform distribution from a to b
$\chi^2(n), \chi_n^2$	Chi-square distribution with $n$ degree
	of freedom (assuming that the Gaus-
	sians are $\mathcal{N}(0,1)$ )
$\exp(\lambda)$	Exponential distribution with rate
	parameter $\lambda$
$\Gamma(\alpha, \beta)$	Gamma distribution with shape pa-
	rameter $\alpha$ and rate parameter $\beta$
$\Gamma(\alpha, \theta)$	Gamma distribution with shape pa-
	rameter $\alpha$ and scale parameter $\theta$ =
	1/eta
$\overline{\mathrm{Nakagami}(m,\Omega)}$	Nakagami-m distribution with shape
	parameter or fading figure $m$ and
	spread, scale, or shape parameter $\Omega$
Rayleigh( $\sigma$ )	Rayleigh distribution with scale pa-
	rameter $\sigma$

$\operatorname{Rayleigh}(\Omega)$	Rayleigh distribution with the second
	moment $\Omega = E\left[x^2\right] = 2\sigma^2$
$Rice(s,\sigma)$	Rice distribution with noncentrality
	parameter s and $\sigma$ . $s^2$ represent the
	specular component power
$\operatorname{Rice}(\Omega, K), \operatorname{Rice}(A, K)$	Rice distribution with Rice factor
	$K = s^2/2\sigma^2$ and scale parameter $\Omega =$
	$A = s^{2} + 2\sigma^{2} = 2\sigma^{2}(K+1)$ ( $\Omega$ is pref-
	ered over A)

# 4 Machine learning, optimization theory, and statistical signal processing

#### 4.1 Matrix Calculus (in denominator layout)

PS: Estimated terms come with a hat above of it

$\mathbf{g}, \nabla f, \frac{\partial f}{\partial \mathbf{w}}$	Gradient descent vector, used in the steepest descent method, also called gradient descent method or deterministic gradient method.
<b>g</b> if the gradient vector is $\nabla f$ (or $\hat{\mathbf{g}}$ if the gradient vector is $\mathbf{g}$ [24])	Stochastic gradient descent (SGD) vector, i.e., instantaneous approximation of gradient descent vector
$\mathbf{g}_{\mathbf{x}}, \nabla_{\mathbf{w}} f, \frac{\partial f}{\partial \mathbf{w}}$	Gradient descent vector with respect $\mathbf{w}$ [9]
$\mathbf{J}, \frac{\partial \mathbf{y}^{\top}}{\partial \mathbf{x}},   abla \mathbf{y}^{\top}  [24]$	Jacobian matrix.
$ \frac{\mathbf{J}, \frac{\partial \mathbf{y}^{T}}{\partial \mathbf{x}}, \nabla \mathbf{y}^{T} [24]}{\mathbf{H}, \frac{\partial^{2} f}{\partial \mathbf{w}^{2}}, \nabla^{2} f [24], \nabla \nabla f [9]} $	Hessian matrix. The notation $\nabla^2$ is sometimes used in Matrix Calculus to denote the second-order vector derivative. However, it must be avoided since, in Vector Calculus, $\nabla^2$ also denotes the Laplacian operator which in turn may be scalar or vector Laplacian operator depending on whether $f$ is scalar- or vector-valued, respectively. Some discussion about can be found in [5–7]

## 4.2 Statistics: estimation and detection theory

X	output
W	Parameters
$p(\mathbf{x} \mid \mathbf{w}), l(\mathbf{x} \mid \mathbf{w})[30]$	Likelihood function
$\ln p(\mathbf{x} \mid \mathbf{w})$	Log-likelihood function
$\Lambda(\mathbf{x})[30], \frac{p(\mathbf{x} H_1)}{p(\mathbf{x} H_0)}$ [27, 30], $L(\mathbf{x})$ [12,	Likelihood ratio function (also called
[27]	likelihood ratio test (LRT) [27])
$\Lambda_l(\mathbf{x}), \mathcal{L}(\mathbf{x})$ [12], $l(\mathbf{x})$ [27]	Log-likelihood ratio (LLR [27]) func-
	tion
$\hat{ ho}_{x,y}$	Estimated Pearson correlation coeffi-
	cient between $x$ and $y$
$\overline{\mathcal{R}_k}$	kth Decision region
$x(t) \stackrel{m.s.e}{=} y(t)$	x(t) equals $y(t)$ is the mean square er-
	ror sense, that is $E[ x(t) - y(t) ^2] = 0$
$x(t) = 1. i. m. \sum_{i=1}^{N} x_i \phi_i(t) [42]$	$\lim_{N\to\infty} \mathbf{E} \left  \left  x(t) - \sum_{i=1}^{N} x_i \phi_i(t) \right ^2 \right  = 0$
$N{ ightarrow}\infty$	(l.i.m stands for "limit in the mean").
	It is analogous to the $\stackrel{m.s.e}{=}$ notation,
	but denoting that they equal in the
	MSE sense only when $N \to \infty$

# 4.3 Signals, (hyper)parameters, system performance, and criteria

N	Number of instances (or samples),
	i.e., $n \in \{1, 2,, N\}$
$N_{ m trn}$	Number of instances in the training
	set, i.e., $n \in \{1, 2,, N_{\text{trn}}\}$
$N_{ m tst}$	Number of instances in the test set,
	i.e., $n \in \{1, 2, \dots, N_{\text{tst}}\}$
$N_{ m val}$	Number of instances in the validation
	set, i.e., $n \in \{1, 2,, N_{\text{val}}\}$
$N_e$	Number of epochs
$N_a$	Number os attributes
K [24]	Number of classes (which is the num-
	ber of outputs in multiclass prob-
	lems). Use $k$ to iterate over it
L	Number of layers, i.e., the depth of
	the network. Use $l$ to iterate over it

$M_l, m_l [24], J [24]$	Number of neurons at the $l$ th layer. You might prefer $J$ in the case of the single-layer perceptron (use $j$ to iterate over it). If you want to iterate through it, a sensible variation of Haykin notation is $M_l$ , where $m_l$
	can be used as an iterator. $m_0$ is the length of the input vector without the bias.
$\mathbf{x}(n), \mathbf{x}_n$	Input signal (in $\mathbb{R}^{N_a+1}$ )
$x_0(n)$	Dummy input of the bais, which is usually $\pm 1$ . $+1$ is preferred [9, 24].
$\varphi(\cdot)[24], h(\cdot)[9]$	Activation function
$\frac{\varphi(\cdot)[24], \ h(\cdot)[9]}{\varphi'(v_{m_l}^{(l)}(n))[24], \ \frac{\partial y_{m_l}^{(l)}(n)}{\partial v_{m_l}^{(l)}(n)}} \ [24]$	Partial derivative of the activation function with respect to $v_{m_l}^{(l)}(n)$ $(m_l$ neuron at $l$ th layer)
$y_{m_l}^{(l)}(n), \varphi\left(v_{m_l}^{(l)}(n)\right)[24], \mathbf{t}_{m_l}^{(l)}(n)[9]$	Output signal (target) of the $m_l$ th neuron at the $l$ th layer
$\mathbf{y}^{(l)}(n)$	Output signal of the <i>l</i> th layer
	Output of the neural network
	Desired label (in case of supervised learning). For multiclass classification, one-hot encoding is usually used. For binary (scalar) classification, however antipodal encoding, i.e., {-1,1} is more recommended [24].
$e_{m_l}(n)$	Error signal of the neuron $m_l$ at the
	Ith layer
$\frac{\mathbf{e}(n), \mathbf{d}(n) - \mathbf{y}(n)}{\mathbf{w}_{m_l}^{(l)}(n), \boldsymbol{\theta}_{m_l}^{(l)}(n)}$	Error signal
$\mathbf{w}_{m_{l}}^{(l)}(n), \mathbf{\theta}_{m_{l}}^{(l)}(n) = \begin{bmatrix} w_{m_{l},0}^{(l)}(n) & w_{m_{l},1}^{(l)}(n) & \dots & w_{m_{l},m_{l-1}}^{(l)}(n) \end{bmatrix}$	Parameters, coefficients, or synaptic weights vector in the <i>l</i> th layer. In
	the case of Single Layer Perceptrons or adaptive filters, the superscript is omitted
$w_{m_l,0}^{(l)}(n), b_{m_l}^{(l)}(n)$	Bias (the first term of the weight vector) of the <i>l</i> th layer
$\mathbf{W}(n), \begin{bmatrix} \mathbf{w}(1) & \mathbf{w}(2) & \cdots & \mathbf{w}(N) \end{bmatrix}^{T}$	Matrix of the synaptic weights
$\widetilde{\mathbf{W}}(n)$	Matrix of the synaptic weights, but without the bias

$v_{m_l}^{(l)}(n), \mathbf{w}_{m_l}^{(l)\top}(n)\mathbf{y}_{m_{l-1}}^{(l-1)}(n)$	Induced local field or activation potential. At the first layer $\mathbf{y}_{m_0}^{(0)}(n) = \mathbf{x}(n)$ [9]
$\mathbf{v}^{(l)}(n), \mathbf{W}^{(l)}(n)\mathbf{y}_{m_{l-1}}^{(l-1)}(n)$	Vector of the local fields at the <i>l</i> th
	layer
$\mathbf{w}^{\star}, \mathbf{w}_{o}, \mathbf{\theta}^{\star}, \mathbf{\theta}_{o}$	Optimum value of the parameters,
	coefficients, or synaptic weights vec-
	tor ( $\mathbf{w}^*$ is also used [9] but it is not
	recommended as it may be confused
	with the conjugation operator)
$\delta_{m_l}^{(l)}(n),rac{\partial \mathscr{E}(n)}{\partial  u_{m_l}^{(l)}(n)}$	Local gradient of the $m_l$ th neuron of
$\partial v_{m_l}(n)$	the $l$ th layer.
$\boldsymbol{\delta}^{(l)}(n)$	Vector of the local gradients of all
	neurons at the $l$ th layer
$\mathbf{X}, \begin{bmatrix} \mathbf{x}(1) & \mathbf{x}(2) & \cdots & \mathbf{x}(N) \end{bmatrix}$	Data matrix [24]
$\eta(n)$	Learning rate hyperparameter [24]
$\mathscr{R}$	Bayes risk or average risk [24]
$c_{ij}, C_{ij}$	Misclassification cost in deciding in
-33	favor of class $\mathscr{C}_i$ (represented in the
	subspace $\mathcal{H}_i$ ) when the $\mathcal{C}_i$ is the true
	class (used in Bayes classifiers/detec-
	tors) [12, 24]
$\mathscr{C}_k[24],  \mathcal{C}_k[9]$	kth class
$ \begin{array}{c c} \mathscr{C}_k[24],  \mathscr{C}_k[9] \\ \mathscr{T}[24],  \mathbb{X}[22] \end{array} $	Training set, i.e., the set $\{\mathbf{x}(n), d(n)\}$
	that is used in the training phase.
$\mathcal{H}_k$	Subspace of the training vector be-
	longing to the class $\mathcal{C}_k$
$\mathcal{H}$	Complete space of the input vector,
	i.e., $\mathcal{H}_1 \cup \mathcal{H}_2 \cup \cdots \mathcal{H}_K$
$\mathscr{X}$ [24]	Set of all vectors in the training,
	batch, validation, or test dataset that
	were misclassified
$\mathscr{E}(\mathbf{w}), \mathscr{E}(\mathbf{w}(n)), \mathscr{E}(n)$	Cost function or objective function
	(the way it is written depends on the
	purpose of the text)
$J(\mathbf{w}), J(\mathbf{w}(n)), J(n)$	Alternative to the cost function
$\Delta \mathcal{E}(\mathbf{w}(n)), \Delta \mathcal{E}(n), \mathcal{E}(\mathbf{w}(n+1))$ -	Cost function or objective function
$\mathscr{E}(\mathbf{w}(n))$	(the way it is written depends on the
	purpose of the text)
$\mathscr{E}_{\mathrm{av}}(\cdot)[24]$	Error energy averaged over the train-
	ing sample or the empirical risk

$\overline{\rho}$	Distance of the margin of separation
	between two classes (Support Vector
	Machine, SVM)
$g(\cdot)$	Discriminant function, i.e., $g(\mathbf{w}^*) = 0$

# 5 Linear Algebra

## 5.1 Common matrices and vectors

$\mathbf{W}, \mathbf{D}$	Diagonal matrix
P	Projection matrix; Permutation ma-
	trix
J	Jordan matrix
$\mathbf{L}$	Lower matrix
U	Upper matrix
$\overline{\mathbf{C}}$	Cofactor matrix
$C_{\mathbf{A}}$ , cof $(\mathbf{A})$	Cofactor matrix of A
S	Symmetric matrix
Q	Orthogonal matrix
$\overline{\mathrm{I}_N}$	$N \times N$ -dimensional identity matrix
$0_{M  imes N}$	$M \times N$ -dimensional null matrix
$\overline{oldsymbol{0}_N}$	N-dimensional null vector
$1_{M  imes N}$	$M \times N$ -dimensional ones matrix
$\overline{1_N}$	N-dimensional ones vector
0	Null matrix, vector, or tensor (di-
	mensionality understood by context)
1	Ones matrix, vector, or tensor (di-
	mensionality understood by context)

## 5.2 Indexing

$x_{i_1,i_2,,i_N}, [\mathcal{X}]_{i_1,i_2,,i_N}$	Element in the position
	$(i_1, i_2, \ldots, i_N)$ of the tensor $\mathcal{X}$
$\mathcal{X}^{(n)}$	nth tensor of a nontemporal sequence
$\mathbf{x}_n, \mathbf{x}_{:n}$	nth column of the matrix $X$
$\mathbf{x}_{n}$ :	nth row of the matrix $X$
$\mathbf{X}_{i_1,,i_{n-1},:,i_{n+1},,i_N}$	Mode- $n$ fiber of the tensor $\mathcal{X}$
$\mathbf{x}_{:,i_2,i_3}$	Column fiber (mode-1 fiber) of the
	thrid-order tensor $\mathcal{X}$
$\mathbf{x}_{i_1,:,i_3}$	Row fiber (mode-2 fiber) of the thrid-
	order tensor $\mathcal{X}$

$\mathbf{x}_{i_1,i_2,:}$	Tube fiber (mode-3 fiber) of the
1, 2,	thrid-order tensor $\mathcal{X}$
$\overline{\mathbf{X}_{i_1,:,:}}$	Horizontal slice of the thrid-order
	tensor $\mathcal{X}$
$\mathbf{X}_{:,i_2,:}$	Lateral slices slice of the thrid-order
	tensor $\mathcal{X}$
$\overline{\mathbf{X}_{i_3},\mathbf{X}_{:,:,i_3}}$	Frontal slices slice of the thrid-order
	tensor $\mathcal{X}$

## 5.3 General operations

$\langle \mathbf{a}, \mathbf{b}  angle  , \mathbf{a}^ op \mathbf{b}, \mathbf{a} \cdot \mathbf{b}$	Inner or dot product
$\mathbf{a} \circ \mathbf{b}, \mathbf{a} \mathbf{b}^{ op}$	Outer product
8	Kronecker product
·	Hadamard (or Schur) (elementwise)
	product
.⊙n	nth-order Hadamard power
$\odot \frac{1}{n}$	nth-order Hadamard root
$\bigcirc \frac{1}{n}$	nth-order Hadamard root Hadamard (or Schur) (elementwise)
	Hadamard (or Schur) (elementwise)
Ø	Hadamard (or Schur) (elementwise) division

## 5.4 Operations with matrices and tensors

$\mathbf{A}^{-1}$	Inverse matrix
$\mathbf{A}^{+}, \mathbf{A}^{\dagger}$	Moore-Penrose left pseudoinverse
$\mathbf{A}^{T}, \mathbf{A}^{T}, \mathbf{A}^{t}, \mathbf{A}^{t} $ [37]	Transpose
$\mathbf{A}^{-\top}$	Transpose of the inverse, i.e.,
	$\left(\mathbf{A}^{-1}\right)^{\top} = \left(\mathbf{A}^{\top}\right)^{-1} \left[21, 33\right]$
<b>A</b> *	Complex conjugate
$\mathbf{A}^{H}$	Hermitian
$\ \mathbf{A}\ _{\mathrm{F}}$	Frobenius norm
$\ \mathbf{A}\ $	Matrix norm
$ \mathbf{A} , \det(\mathbf{A})$	Determinant
$\operatorname{diag}\left(\mathbf{A}\right)$	The elements in the diagonal of <b>A</b>
	Vectorization: stacks the columns of
	the matrix <b>A</b> into a long column vec-
	tor

$\mathbf{E}_d\left[\mathbf{A}\right]$	Extracts the diagonal elements of a
	square matrix and returns them in a
	column vector
$\mathbf{E}_{l}\left[\mathbf{A} ight]$	Extracts the elements strictly below
	the main diagonal of a square matrix
	in a column-wise manner and returns
	them into a column vector
$\mathbf{E}_{u}\left[\mathbf{A}\right]$	Extracts the elements strictly above
	the main diagonal of a square matrix
	in a column-wise manner and returns
	them into a column vector
$\mathbf{E}_b\left[\mathbf{A}\right]$	Block vectorization operator: stacks
	square block matrices of the input
	into a long block column matrix
unvec (A)	Reshapes a column vector into a ma-
	trix
$-\operatorname{tr}\{\mathbf{A}\}$	trace
$X_{(n)}$	$n$ -mode matricization of the tensor $\mathcal X$

## 5.5 Operations with vectors

$\ \mathbf{a}\ $	$l_1$ norm, 1-norm, or Manhattan norm
$\ \mathbf{a}\ , \ \mathbf{a}\ _2$	$l_2$ norm, 2-norm, or Euclidean norm
$\ \mathbf{a}\ _p$	$l_p$ norm, $p$ -norm, or Minkowski norm
$\ \mathbf{a}\ _{\infty}$	$l_{\infty}$ norm, $\infty$ -norm, or Chebyshev
	norm
diag (a)	Diagonalization: a square, diagonal matrix with entries given by the vec-
	tor a

## 5.6 Decompositions

Λ	Eigenvalue matrix [39]
Q	Eigenvectors matrix; Orthogonal ma-
	trix of the QR decomposition[39]
R	Upper triangular matrix of the QR
	decomposition[39]
U	Left singular vectors[39]
$\overline{\mathbf{U}_r}$	Left singular nondegenerated vectors
Σ	Singular value matrix
$\Sigma_r$	Singular value matrix with nonzero
	singular values in the main diagonal

$\Sigma^+$	Singular value matrix of the pseu-
-	
	doinverse [39]
$\Sigma_r^+$	Singular value matrix of the pseu-
	doinverse with nonzero singular val-
	ues in the main diagonal
V	Right singular vectors [39]
$\overline{\mathrm{V}_r}$	Right singular nondegenerated vec-
	tors
$eig(\mathbf{A})$	Set of the eigenvalues of A [13, 30,
	33]
$[A, B, C, \ldots]$	CANDECOMP/PARAFAC (CP) de-
	composition of the tensor $\mathcal{X}$ from the
	outer product of column vectors of <b>A</b> ,
	$\mathbf{B},\mathbf{C},\ldots$
$\boxed{\llbracket \boldsymbol{\lambda}; \mathbf{A}, \mathbf{B}, \mathbf{C}, \ldots \rrbracket}$	Normalized CANDE-
	COMP/PARAFAC (CP) decom-
	position of the tensor $\mathcal{X}$ from the
	outer product of column vectors of
	$A, B, C, \dots$

## 5.7 Spaces and sets

#### 5.7.1 Common spaces and sets

$\mathbb R$	Set of real numbers
a,b	Closed interval of a real set from $a$ to
	b
(a,b)	Opened interval of a real set from $a$
	to b
[a,b),(a,b]	Half-opened intervals of a real set
	from $a$ to $b$
$\mathbb{C}$	Set of complex numbers
$\mathbb{I}, j\mathbb{R}$	Set of imaginary numbers
$\mathbb{Q}$	Set of rational number
$\mathbb{R}\setminus\mathbb{Q}$	Set of irrational number
$\mathbb{Z}$	Set of integer number
N	Set of natural numbers
$\overline{\{1,2,\ldots,n\}}$	Discrete set containing the integer el-
	ements $1, 2, \ldots, n$
$\mathbb{B} = \{0, 1\}$	Boolean set
Ø	Empty set
$\mathbb{K} \in \{\mathbb{R}, \mathbb{C}\}$	Real or complex space (field)

$\mathbb{K}^{I_1 \times I_2 \times \cdots \times I_N}$	$I_1 \times I_2 \times \cdots \times I_N$ -dimensional real (or complex) space
$\mathbb{K}_{+}$	Nonnegative real (or complex) space
	[10]
K++	Positive real (or complex) space, i.e.,
	$\mathbb{K}_{++} = \mathbb{K}_+ \setminus \{0\} \ [10]$
$\overline{U}$	Universe
$2^A$	Power set of A

#### 5.7.2 Convex sets (or spaces)

$\mathbb{S}^n$ [15], $\mathcal{S}^n$ [10]	Conic set of the symmetric matrices in $\mathbb{R}^{n \times n}$
$\mathbb{S}^{n\perp}$ [15]	Conic set of the skew-symmetric (also called antisymmetric) matrices in $\mathbb{R}^{n \times n}$
$\mathbb{S}^n_+$ , $\mathcal{S}^n_+$	Conic set of the symmetric positive semidefinite matrices in $\mathbb{R}^{n \times n}$ [10]
$\mathbb{S}^n_{++}, \mathcal{S}^n_{++}$	Conic set of the symmetric positive definite matrices in $\mathbb{R}^{n\times n}$ , i.e., $\mathbb{S}^n_{++} = \mathbb{S}^n_+ \setminus \{0\}$ [10]
$\mathbb{H}^n$	Set of all hermitian matrices in $\mathbb{C}^{n\times n}$
conv C	Convex hull
aff C	Affune hull
$\overline{\mathcal{R}}$	Ray
$\mathcal{H}$	Hyperplane
$\mathcal{H}_+, \mathcal{H}$	Positive/negative halfspace
$B(\mathbf{x}_c, r)$	Euclidean ball with radium $r$ and
	centered at $\mathbf{x}_c$
$\mathcal{E}$	Ellipsoid
C K	Norm cone
	Proper cone
<i>K</i> *	Dual cone
$\mathcal{P}$	Polyhedra
S	Simplex
$C_{\alpha}$	$\alpha$ -sublevel set
-epi $f$	Epigraph of the function $f$
hypo $f$	Hypograph of the function $f$

#### 5.7.3 Spaces from matrices or vectors

$\mathrm{span}\left\{\mathbf{a}_{1},\mathbf{a}_{2},\ldots,\mathbf{a}_{n}\right\}$	Vector space spanned by the argu-
	ment vectors [21]
$C(\mathbf{A})$ , columnspace( $\mathbf{A}$ ), range( $\mathbf{A}$ ),	Columnspace, range or image, i.e.,
$\operatorname{span} \{\mathbf{A}\}, \operatorname{image}(\mathbf{A})$	the space span $\{\mathbf{a}_1, \mathbf{a}_2, \dots, \mathbf{a}_n\}$ , where
	$\mathbf{a}_i$ is the ith column vector of the ma-
	trix <b>A</b> [31, 39]
$C(\mathbf{A}^{H})$	Row space (also called left
	columnspace) [31, 39]
$N(\mathbf{A})$ , nullspace( $\mathbf{A}$ ), null( $\mathbf{A}$ ), kernel( $\mathbf{A}$	Nullspace (or kernel space) [31, 39,
	40]
$N(\mathbf{A}^{H})$	Left nullspace
rank A	Rank, that is, $\dim(\operatorname{span}\{A\}) =$
	$\dim (C(\mathbf{A}))$ [31]
nullity (A)	Nullity of $\mathbf{A}$ , i.e., dim $(N(\mathbf{A}))$

## 5.8 Set operations

A + B	Set addition (Minkowski sum), i.e.,
	$\{\mathbf{v} \in \mathbb{R}^n \mid \mathbf{v} = \mathbf{x} + \mathbf{y}, \ \forall \ \mathbf{x} \in \mathcal{X} \land \mathbf{y} \in \mathcal{Y}\}$
	[28]
A - B	Minkowski difference, i.e.,
	$\{\mathbf{v} \in \mathbb{R}^n \mid \mathbf{v} = \mathbf{x} - \mathbf{y}, \ \forall \ \mathbf{x} \in \mathcal{X} \land \mathbf{y} \in \mathcal{Y}\}$
$A\ominus B$	Pontryagin difference, i.e.,
	$\{\mathbf{v} \in \mathbb{R}^n \mid \mathbf{v} + \mathbf{y} \in \mathcal{X}, \ \forall \ \mathbf{y} \in \mathcal{Y}\} $ [28]
$A \setminus B, A - B$	Set difference or set subtraction, i.e.,
	$A \setminus B = \{x   x \in A \land x \notin B\}$ the set con-
	taining the elements of $A$ that are not
	in $B[\overline{36}]$
$A \cup B$	Set of union
$A \cap B$	Set of intersection
$A \times B$	Cartesian product
$A^n$	$A \times A \times \cdots \times A$
	n  times
$A^{\perp}$	Orthogonal complement of $A$ , e.g.,
	$N(\mathbf{A}) = C(\mathbf{A}^{T})^{\perp} [10]$
$a \perp b$	a is orthogonal to b
a ∠ b	<b>a</b> is not orthogonal to <b>b</b>
$A \oplus B$	Direct sum, i.e., each $\mathbf{v} \in$
	$\{\sum \mathbf{a}_i \mid \mathbf{a}_i \in S_i, i = 1, \dots, k\}$ has a
	unique representation of $\sum \mathbf{a}_i$ with
	$\mathbf{a}_i \in S_i$ . That is, they expand to a
	space. Note that $\{S_i\}$ might not be
	orthogonal each other [21]

$A\overset{\perp}{\oplus} B$	Direct sum of two spaces that are orthogonal and span a $n$ -dimensional
	space, e.g., $C(\mathbf{A}^{T}) \stackrel{\perp}{\oplus} C(\mathbf{A}^{T})^{\perp} =$
	$\mathbb{R}^n$ (this decomposition of $\mathbb{R}^n$ is
	called the orthogonal decomposition
	induced by $\mathbf{A}$ ) [10]
$\overline{A}, A^c$	Complement set (given $U$ )
#A,  A	Cardinality of A
$a \in A$	a is element of A
$a \notin A$	a is not element of A

## 5.9 Inequalities

$\mathcal{X} \le 0$	Nonnegative tensor
$\mathbf{a} \leq_K \mathbf{b}$	Generalized inequality meaning that
	$\mathbf{b} - \mathbf{a}$ belongs to the conic subset $K$ in
	the space $\mathbb{R}^n[10]$
$a <_K b$	Strict generalized inequality meaning
	that $\mathbf{b} - \mathbf{a}$ belongs to the interior of
	the conic subset $K$ in the space $\mathbb{R}^n[10]$
$a \le b$	Generalized inequality meaning that
	$\mathbf{b} - \mathbf{a}$ belongs to the nonnegative or-
	thant conic subset, $\mathbb{R}^n_+$ , in the space
	$\mathbb{R}^n.[10]$
a < b	Strict generalized inequality meaning
	that $\mathbf{b} - \mathbf{a}$ belongs to the positive or-
	thant conic subset, $\mathbb{R}^n_{++}$ , in the space
	$\mathbb{R}^n[10]$
$A \leq_K B$	Generalized inequality meaning that
	${\bf B}-{\bf A}$ belongs to the conic subset $K$
	in the space $\mathbb{S}^n[10]$
$A \prec_K B$	Strict generalized inequality meaning
	that $\mathbf{B} - \mathbf{A}$ belongs to the interior of
	the conic subset $K$ in the space $\mathbb{S}^n[10]$
$A \leq B$	Generalized inequality meaning that
	$\mathbf{B} - \mathbf{A}$ belongs to the positive semidef-
	inite conic subset, $\mathbb{S}_{+}^{n}$ , in the space
	$\mathbb{S}^n[10]$
A < B	Strict generalized inequality meaning
	that $\mathbf{B} - \mathbf{A}$ belongs to the positive or-
	thant conic subset, $\mathbb{S}_{++}^n$ , in the space
	$\mathbb{S}^n[10]$

# 6 Communication systems

## 6.1 Common symbols

В	One-sided bandwidth of the base-
	band signal, in Hz
W	One-sided bandwidth of the base-
	band signal, in rad/s
$N_0$	Noise density, in ???
$x_i$	Real or in-phase part of x
$x_q$	Imaginary or quadrature part of $x$
$f_c, f_{RF}$	Carrier frequency (in Hertz)
$f_L$	Carrier frequency in L-band (in
	Hertz)
$\frac{f_{IF}}{f_s}$	Intermediate frequency (in Hertz)
$f_s$	Sampling frequency or sampling rate (in Hertz)
$T_s$	Sampling time interval/duration/pe-
	riod
R	Bit rate
T	Bit interval/duration/period
$T_c$	Chip interval/duration/period
$T_{sy}, T_{sym}$	Symbol/signaling[34] interval/dura-
29. 29	tion/period
$S_{RF}$	Transmitted signal in RF
$S_{FI}$	Transmitted signal in FI
$S, S_l$	Lowpass (or baseband) equivalent
	signal or envelope complex of trans-
	mitted signal
$r_{RF}$	Received signal in RF
$r_{FI}$	Received signal in FI
$r, r_l$	Lowpass (or baseband) equivalent
	signal or envelope complex of re-
	ceived signal
$\phi$	Signal phase
$\phi_0$	Initial phase
$\overline{\eta_{RF}, w_{RF}}$	Noise in RF
$\eta_{FI}, w_{FI}$	Noise in FI
$\overline{\eta}, w$	Noise in baseband
τ	Timing delay
$\Delta  au$	Timing error (delay - estimated)
$\overline{\varphi}$	Phase offset
$\Delta \varphi$	Phase error (offset - estimated)
$f_d$	Linear Doppler frequency

$\Delta f_d$	Frequency error (Doppler frequency -
	estimated)
ν	Angular Doppler frequency
$\Delta \nu$	Frequency error (Doppler frequency -
	estimated)
$\gamma, A$	Transmitted signal amplitude
$\gamma_0, A_0$	Combined effect of the path loss and
	antenna gain

## ${\bf 6.2} \quad {\bf Fading\ multipath\ channels}$

$t \stackrel{\mathcal{F}}{\leftrightarrow} \lambda \ [34]$	Support temporal of the signal. $\lambda$ is obtained after taking the Fourier transform on $t$ .
$\tau \stackrel{\mathcal{F}}{\leftrightarrow} f \ [34]$	Second support temporal of the signal $(c(t))$ varies with with the input at the time $\tau$ ). $f$ is obtained after taking the Fourier transform on $\tau$ .
$c(t,\tau) [34]$	Complex envelope of the channel response at the time t due to an impulse
C(f, t) [94]	applied at the $t - \tau$
$\frac{C(f,t) [34]}{\alpha(t,\tau) [34]}$	Transfer function of $c(t,\tau)$ in $\tau$
$\alpha(t,\tau)$ [54]	Attenuation of $c(t,\tau)$ , i.e., $c(t,\tau) = \alpha(t,\tau)e^{e\pi f_c\tau}$
$R_c(\tau_1, \tau_2, \Delta t)$ [34]	Autocorrelation function of
, , , , , , , , , , , , , , , , , , ,	$c(t,\tau)$ , i.e., $R_c(\tau_1,\tau_2,\Delta t) =$
	$\mathrm{E}\left[c^*(t,\tau_1),c^*(t+\Delta t,\tau_2)\right]$
$R_c(\tau, \Delta t)$ [34]	Autocorrelation function of $c(t, \tau)$ as-
	suming uncorrelated scattering
$R_c(\tau), R_c(\tau, \Delta t)\Big _{\Delta t=0}$ [34]	Multipath intensity profile or delay
-24-0	power spectrum
$R_C(\Delta f, \Delta t), R_C(f_1, f_2; \Delta t),$	Spaced-frequency, spaced-time corre-
$\mathrm{E}\left[C(f_1,t),C(f_2,t+\Delta t)\right],$	lation function $(\Delta f = f_2 - f_1)$
$\mathcal{F}_{\tau}\left\{R_{c}(\tau,\Delta t)\right\} [20]$	
$R_C(\Delta f),  R_C(\Delta f, \Delta t) _{\Delta t=0}  [34],$	Spaced-frequency correlation func-
$\mathcal{F}\left\{R_c(\tau)\right\} [20]$	tion
$(\Delta f)_c$	Coherence bandwidth of $c(t)$ , that
	is, the frequency interval in which
	$R_C(\Delta f)$ is nonzero [34]
$T_m$	Multipath spread of the channel, that
	is, the time interval in which $R_c(\tau)$ is
	nonzero $(T_m \approx 1/(\Delta f)_c)$ [34]

$R_C(\Delta t), R_C(\Delta f, \Delta t)\Big _{\Delta f=0}$	Spaced-time correlation function [34]
$S_C(\lambda)$ [34], $\mathcal{F}\{R_C(\Delta t)\}$ [20]	Doppler power spectrum
$(\Delta t)_c$	Coherence time of $c(t)$ , that is, the
	time interval in which $R_C(\Delta t)$ is
	nonzero [34]
$B_m$	Multipath spread of the channel, that
	is, the frequency interval in which
	$S_c(\lambda)$ is nonzero $(B_d \approx 1/(\Delta t)_c)$ [34]
$S_C(\tau,\lambda)$ [34], $\mathcal{F}_{\Delta f,\Delta t} \left\{ R_C(\Delta f,\Delta t) \right\}$	Scattering function
[20]	

## 7 Discrete mathematics

## 7.1 Quantifiers, inferences

$\forall$	For all (universal quantifier) [23]
3	There exists (existential quantifier)
	[23]
∄	There does not exist [23]
∃!	There exists an unique [23]
$\exists_n$	There exists exactly n [36]
€	Belongs to [23]
∉	Does not belong to [23]
:	Because [23]
<u> ,:</u>	Such that, sometimes that parenthe-
	ses is used [23]
$\overline{}$ ,,(·)	Used to separate the quantifier with
	restricted domain from its scope, e.g.,
	$\forall x < 0 (x^2 > 0) \text{ or } \forall x < 0, x^2 > 0$
	[23]
··	Therefore [23]

## 7.2 Propositional Logic

$\neg a$	Logical negation of $a$ [36]
$a \wedge b$	Conjunction (logical AND) operator
	between $a$ and $b[36]$
$a \lor b$	Disjunction (logical OR) operator be-
	tween $a$ and $b[36]$
$a \oplus b$	Exclusive OR (logical XOR) operator
	between $a$ and $b[36]$

$a \rightarrow b$	Implication (or conditional) state-
	ment[36]
$a \leftrightarrow b$	Bi-implication (or biconditional)
	statement, i.e., $(a \rightarrow b) \land (b \rightarrow a)$
	[36]
$a \equiv b, a \iff b, a \Leftrightarrow b$	Logical equivalence, i.e., $a \leftrightarrow b$ is a
	tautology[36]

## 7.3 Operations

a	Absolute value of $a$
log	Base-10 logarithm or decimal loga-
	$\operatorname{rithm}$
ln	Natual logarithm
$\operatorname{Re}\left\{ x\right\}$	Real part of x
$\operatorname{Im}\left\{ x\right\}$	Imaginary part of x
۷٠	Phase (complex argument)
$x \mod y$	Remainder, i.e., $x - y \lfloor x/y \rfloor$ , for $y \neq 0$
x div y	Quotient [36]
$x \equiv y \pmod{m}$	Congruent, i.e., $m \setminus (x - y)$ [36]
$\operatorname{frac}(x)$	Fractional part, i.e., $x \mod 1$ [23]
$a \setminus b$ [23, Section 4.1], $a \mid b$ [36]	b is a positive integer multiple of $a \in$
	$\mathbb{Z}$ , i.e., $\exists ! \ n \in \mathbb{Z}_{++} \mid b = na$
a ∤ b [23, Section 4.1], a ∤ b [36]	b is not a positive integer multiple of
	$a \in \mathbb{Z}$ , i.e., $\not\equiv n \in \mathbb{Z}_{++} \mid b = na$
[·]	Ceiling operation [23]
[.]	Floor operation [23]

## 8 Vector Calculus

$\nabla f[38]$ , grad $f[35]$	Vector differential operator (Nabla symbol), i.e., $\nabla f$ is the gradient of the scalar-valued function $f$ , i.e., $f$ : $\mathbb{R}^n \to \mathbb{R}$
t,(u,v)	Parametric variables commonly used,
	t for one variable, $(u, v)$ for two vari-
	ables[38]
$l(x, y, z)$ [35], $\mathbf{r}(x, y, z)$ [38], $x\hat{\mathbf{i}} + y\hat{\mathbf{j}} + z\hat{\mathbf{k}}$	Vector position, i.e., $(x, y, z)$ .
l(t)	Vector position parametrized by $t$ ,
	i.e., $(x(t), y(t), z(t))$ [35, 38]

1/(4) 31 /34	First desirative of 1(4) is the
$\mathbf{l}'(t), \mathrm{d}\mathbf{l}/\mathrm{d}t$	First derivative of $\mathbf{l}(t)$ , i.e., the tangent vector of the curve
	tangent vector of the curve $(x(t), y(t), z(t))$ [38]
$\mathbf{u}(t)[29] \ \mathbf{T}(t)[38], \ \mathrm{dl}(t)[35]$	Tangent unit vector of $\mathbf{l}(t)$ , i.e.,
$\mathbf{u}(t)[29] 1(t)[30],  \mathbf{u}(t)[30]$	
$-\frac{1}{\left( \frac{y'(t)}{y'(t)} - \frac{y'(t)}{y'(t)} \right)}$	$\mathbf{u}(t) = \mathbf{l}'(t)/ \mathbf{l}'(t) $
$\mathbf{n}(t), \left(\frac{y'(t)}{ l'(t) }, -\frac{x'(t)}{ l'(t) }\right)$	Normal vector of $\mathbf{l}(t)$ , i.e.,
-	$\mathbf{n}(t) \perp \mathbf{T}(t)[38]$
C	Contour that traveled by $\mathbf{l}(t)$ , for $a \le t \le b$ [38]
L, L(C)	Total length of the contour $C$
	(which can be defined the vector
	$\hat{l}$ , parametrized by $t$ ), i.e., $L_C =$
	$\int_a^b  \mathbf{l}'(t)   \mathrm{d}t [38]$
s(t)	Length of the arc, which can be de-
~(*)	fined by the vector $\mathbf{l}$ and $t$ , that is,
ds	$s(t) = \int_{a}^{t}  \mathbf{l}'(u)   \mathrm{d}u \ (s(b) = L)[38]$ Differential operator of the length of
us	the contour $C$ , i.e., $ds =  \mathbf{l}'(t)  dt$ [38]
$\int_C f(1)  \mathrm{d}s,  \int_a^b f(1(t))  1'(t)   \mathrm{d}t$	Line integral of the function $f: \mathbb{R}^n \to$
$\int_C \int (\mathbf{I})  \mathrm{d}\mathbf{s}, \ \int_a \int (\mathbf{I}(t))  \mathbf{I}(t)   \mathrm{d}t$	$\mathbb{R}$ along the contour $C$ . In the context
	of integrals in the complex plane, it
	is also called "contour integral"
$\theta$ [35]	Angle between the contour $C$ and the
. [60]	vector field ${f F}$
$\int_C \mathbf{F} \cdot d\mathbf{l}, \ \int_a^b \mathbf{F}(\mathbf{l}(t)) \cdot \mathbf{l}'(t)  dt \ [8, 38],$	Line integral of vector field <b>F</b> along
$ \int_{C} \mathbf{F} \cdot \mathbf{u}  ds, \int_{C} \mathbf{F} \cos \theta  ds  [35] $ $ \int_{C} \mathbf{F} \cdot d\mathbf{u}  [35] $	the contour $C$
$\int_{C} \mathbf{F} \cdot d\mathbf{u} \ [35]$	In the field of electromagnetics, it is
	common to apply the line integral be-
	tween the vector field ${f F}$ and the unit
	vector $\mathbf{u}(t)$ . Therefore, this line inte-
	gral may appear as well
$\int_{2}^{\mathbf{b}} \mathbf{F}, \int_{2}^{\mathbf{b}} \mathbf{F} \cdot d\mathbf{l}$	Alternative notation to the line inte-
va va	gral, where the parametric variable $t$
	goes from $a$ to $b$ , making $r$ goes from
	$\mathbf{l}(a) = \mathbf{a} \text{ to } \mathbf{l}(b) = \mathbf{b} [8]$
$\oint_C, \oint_C$	Line integral along the closed contour
	C. The arrow indicates the contour
	integral orientation, which is counter-
	clockwise, by default. In the context
	of integrals in the complex plane, it is
	also called "closed contour integral".

$ \not\vdash_S $	Surface integral over the closed sur-
	face $S$
l(u, v)	Vector position
	(x(u, v), y(u, v), z(u, v)) parametrized
	by $(u, v)$
$\mathbf{l}_u$	$(\partial x/\partial u, \partial y/\partial u, \partial z/\partial u)$
$l_{\nu}$	$(\partial x/\partial v, \partial y/\partial v, \partial z/\partial v)$
$\mathrm{d}A$	Differential operator of a 2D area
	(denoted by $D$ or $R$ ) in the $\mathbb{R}^2$ do-
	main. This differential operator can
	be solved in different ways (rectangu-
	lar, polar, cylindric, etc) [38]
D,R	Integration domain in which dA is
	integrated, i.e., $\iint_D f  dA$ . R is pre-
	ferred when the integration domain
	is a rectangle, while $\stackrel{\circ}{D}$ is used when
	it has nonrectangular shape [38]
S	Smooth surface $S \subset \mathbb{R}^3$ , i.e., a 2D
	area in a 3D space
$\mathrm{d}S$ , $ \mathbf{l}_u \times \mathbf{l}_v $ $\mathrm{d}A$	Differential operator of a 2D area in
/   W V	a 3D domain (an surface). Note that
	$dS =  \mathbf{l}_u \times \mathbf{l}_v  dA$ should be accompa-
	nied with the change of the integra-
	tion interval(from $S$ to $D$ )
$A(S), \iint_S dS, \iint_D  \mathbf{l}_u \times \mathbf{l}_v  dA$	Area of the surface $S$ parametrized by
$J_{J} = J_{J} = J_{J$	(u, v), in which dA is the area defined
	in the $D$ domain (which is form by
	the <i>u</i> -by- <i>v</i> graph)
$\mathrm{d}V$	Differential operator of a shape vol-
	ume (denoted by $E$ ) in $\mathbb{R}^3$ domain,
	i.e., $\iiint_F dV = V$
E	Integration domain in which $dV$ is in-
	tegrated, i.e., $\iiint_E f  dV$ [38]
$V, \iint_D f  \mathrm{d}A, \iiint_E f  \mathrm{d}V$	Volume of the function $f$ over the re-
W.E	gions $D$ (in the case of double inte-
	grals) or $E$ (in the case of triple inte-
	grals)
$\iint_{S} f  \mathrm{d}S, \iint_{D} f  \mathbf{l}_{u} \times \mathbf{l}_{v}   \mathrm{d}A$	Surface integral over $S$
$\iint_{S} f  dS, \iint_{D} f   \mathbf{l}_{u} \times \mathbf{l}_{v}    dA$ $\mathbf{n}(u, v), \frac{\mathbf{l}_{u}(u, v) \times \mathbf{l}_{v}(u, v)}{ \mathbf{l}_{u}(u, v) \times \mathbf{l}_{v}(u, v) }$	Normal vector of of the smooth surface $S$
$\iint_{\mathbf{S}} \mathbf{F} \cdot \mathbf{n}  \mathrm{d}S$ , $\iint_{\mathbf{S}} \mathbf{F} \cdot \mathrm{d}\mathbf{S}$ ,	Flux integral of vector field ${f F}$ through
$\iint_D \mathbf{F} \cdot (\mathbf{l}_u \times \mathbf{l}_v)  \mathrm{d}A$	the smooth surface $S$ ( $\mathbf{n} dS \triangleq d\mathbf{S}$ )

$ \oint_{S} \mathbf{F} \cdot \mathbf{n}  dS, \oint_{S} \mathbf{F} \cdot d\mathbf{S}, $	Flux integral of vector field ${f F}$ through
$\iint_{D} \mathbf{F} \cdot (\mathbf{l}_{u} \times \mathbf{l}_{v})  \mathrm{d}A$	the smooth and closed surface $S$
JJD	$(\mathbf{n}  \mathrm{d} S \triangleq \mathrm{d} \mathbf{S})$
$\nabla \times \mathbf{F}$ , curl $\mathbf{F}$	Curl (rotacional) of the vector field ${f F}$
$\nabla \cdot \mathbf{F}, \operatorname{div} \mathbf{F}$	Divercence of the vector field ${f F}$
$\nabla^2 f, \nabla \cdot (\nabla f), \Delta f,$	Scalar Laplacian operator (per-
$\partial^2 f/\partial x^2 + \partial^2 f/\partial y^2 + \partial^2 f/\partial z^2$	formed on a scalar-valued function
	$f: \mathbb{R}^n \to \mathbb{R}$
$\overline{\nabla^2 \mathbf{F}, \nabla \times \nabla \times \mathbf{F} - \nabla(\nabla \cdot \mathbf{F}), \Delta \mathbf{F},}$	Vector Laplacian operator (per-
$(\partial^2 \mathbf{F}/\partial x^2, \partial^2 \mathbf{F}/\partial y^2, \partial^2 \mathbf{F}/\partial z^2)$	formed on a vector field, i.e., a
	vector-valued function, $\mathbf{F} : \mathbb{R}^n \to$
	$\mathbb{R}^n$ ). $\nabla^2$ denotes the scalar (vector)
	Laplacian if the function is scalar-
	valued (vector-valued). The notation
	$\Delta$ must be avoided as it is overused
	in many contexts

# 9 Electromagnetic waves

$\Phi$	Electric flux (scalar) (in V m)
Н	Magnetic field vector (in A/m)
В	Magnetic flux density vector (in
	$Wb/m^2 = T$
$\Phi[14]$	Magnetic flux
$q_{ m f},q_{ m free},Q_{ m free}[18]$	Free electric charge (in C)
$q_{ m b},q_{ m bound},Q_{ m bound}[18]$	Bound electric charge (in C)
$q, q_{\rm f} + q_{\rm b}$	Electric charge (in C)
$\rho_{\rm f}[1], \rho_{\rm free}$ [18]	Free electric charge density
$\rho_{\rm b}[1], \rho_{\rm bound}$ [18]	Electric charge density
$\rho, \rho_{\rm f} + \rho_{\rm b}$	Electric charge density (it can be
	in $C/m^3$ , $C/m^2$ or $C/m$ depending
	whether it is a volume, surface, or
	line shapes)
<b>f</b> [35], <b>F</b> [2]	Electrostatic force (Coulomb force),
	$(\text{in kg m/s}^2).$
ε	Electric permittivity(in F/m). If the
	medium is isotropic, it is a scalar. If
	it is anisotropic, it is a tensor. [35]
$\overline{\varepsilon_r}$	Relative electric permittivity or di-
	electric constant (in F/m) [35]
$\epsilon_0$	Electric permittivity in vacuum,
	$8.854 \times 10^{-12} \mathrm{F/m}$ [35]

Electric field vector (in V/m)
Electric conductivity (in S/m)
Electric current density vector (in
$\mathrm{A/m^2})$
Magnetization current density vector
$(in A/m^2)$
Electric flux density, electric dis-
placement, or electric induction vec-
tor (in $C/m^2$ )
Electric potential energy
Electric potential (in voltage, V).
However, keep in mind that there is
a subtle difference between both def-
initions [4]
Electric flux (in V m)
Electric flux ( <b>D</b> -field flux)
Electric polarization of the material
$(in C/m^2)$
Electric susceptibility (for linear and
isotropic materials)
Magnetic permeability
Magnetic permeability in vacuum

# ${\bf 10}\quad {\bf Generic\ mathematical\ symbols}$

	Q.E.D.
<u>_</u>	Equal by definition
:=, ←	Assignment [36]
<b>≠</b>	Not equal
$\infty$	Infinity
j	$\sqrt{-1}$

# 11 Abbreviations

wrt.	With respect to
st.	Subject to
iff.	If and only if
EVD	Eigenvalue decomposition, or eigen-
	decomposition [31]
SVD	Singular value decomposition
CP	CANDECOMP/PARAFAC

#### References

- URL: https://en.wikipedia.org/wiki/Electric\_displacement\_ field#Definition.
- [2] URL: https://en.wikipedia.org/wiki/Coulomb%27s\_law.
- [3] URL: https://en.wikipedia.org/wiki/Electric\_potential.
- [4] URL: https://physics.stackexchange.com/a/300937/368410.
- [5] Libavius (https://math.stackexchange.com/users/1020990/libavius). Which is the correct vector calculus notation for the Hessian? Mathematics Stack Exchange. URL:https://math.stackexchange.com/q/4560326 (version: 2023-02-15). eprint: https://math.stackexchange.com/q/4560326. URL: https://math.stackexchange.com/q/4560326.
- [6] maple (https://math.stackexchange.com/users/51601/maple). Does the symbol ∇² has the same meaning in Laplace Equation and Hessian Matrix? Mathematics Stack Exchange. URL:https://math.stackexchange.com/q/1353761 (version: 2022-07-29). eprint: https://math.stackexchange.com/q/1353761.
  URL: https://math.stackexchange.com/q/1353761.
- [7] Rubem Pacelli (https://math.stackexchange.com/users/817590/rubem-pacelli). Ambiguity over the notation ∇²: vector Laplacian operator (Vector Calculus) vs. second directional derivative (Matrix Calculus). Mathematics Stack Exchange. URL:https://math.stackexchange.com/q/4693212 (version: 2023-05-05). eprint: https://math.stackexchange.com/q/4693212. URL: https://math.stackexchange.com/q/4693212.
- [8] TM Apostol. Calculus, 2nd Edn., Vol. 2. 1967.
- [9] Christopher M Bishop and Nasser M Nasrabadi. *Pattern Recognition and Machine Learning*. Vol. 4. 4. Springer, 2006.
- [10] Stephen Boyd, Stephen P. Boyd, and Lieven Vandenberghe. Convex Optimization. Cambridge university press, 2004.
- [11] Robert Grover Brown and Patrick YC Hwang. Introduction to Random Signals and Applied Kalman Filtering: With MATLAB Exercises and Solutions. 1997.
- [12] Charles Casimiro. Lecture notes in Statistical Signal Processing. 2019.
- [13] Rama Chellappa and Sergios Theodoridis. Signal Processing Theory and Machine Learning. Academic Press, 2014. ISBN: 0-12-396502-0.
- [14] David Keun Cheng. Field and Wave Electromagnetics. Pearson Education India, 1989.
- [15] Jon Dattorro. Convex Optimization & Euclidean Distance Geometry. Lulu. com, 2010. ISBN: 0-615-19368-4.
- [16] Paulo SR Diniz. Adaptive Filtering: Algorithms and Practical Implementation. Nowell, MA: Kluwer Academic Publishers, 2002.

- [17] Paulo SR Diniz, Eduardo AB Da Silva, and Sergio L Netto. *Digital Signal Processing: System Analysis and Design*. Cambridge University Press, 2010. ISBN: 1-139-49157-1.
- [18] Example Wikipedia Page. URL: https://en.wikipedia.org/wiki/ Gauss%27s\_law#Equation\_involving\_the\_D\_field.
- [19] Example Wikipedia Page. URL: https://en.wikipedia.org/wiki/Flux# Electric\_flux.
- [20] Andrea Goldsmith. Wireless Communications. Cambridge university press, 2005. ISBN: 0-521-83716-2.
- [21] Gene H Golub and Charles F Van Loan. Matrix Computations. JHU press, 2013. ISBN: 1-4214-0859-7.
- [22] Ian Goodfellow, Yoshua Bengio, and Aaron Courville. *Deep Learning*. Illustrated edição. Cambridge, Massachusetts: The MIT Press, Nov. 18, 2016. ISBN: 978-0-262-03561-3.
- [23] Ronald L Graham et al. "Concrete Mathematics: A Foundation for Computer Science". In: *Computers in Physics* 3.5 (1989), pp. 106–107. ISSN: 0894-1866.
- [24] Simon Haykin. Neural Networks and Learning Machines, 3/E. Pearson Education India, 2009. ISBN: 93-325-8625-X.
- [25] Simon S Haykin. *Adaptive Filter Theory*. Pearson Education India, 2002. ISBN: 81-317-0869-1.
- [26] Vinay K Ingle and John G Proakis. *Digital Signal Processing Using MAT-LAB*. Cole Publishing Company, 2000.
- [27] Steven M. Kay. Fundamentals of Statistical Processing, Volume 2: Detection Theory. Pearson Education India, 2009.
- [28] Basil Kouvaritakis and Mark Cannon. "Model Predictive Control". In: Switzerland: Springer International Publishing 38 (2016).
- [29] Erwin Kreyszig, K Stroud, and G Stephenson. Advanced Engineering Mathematics. Vol. 9. John Wiley & Sons, Inc. 9 th edition, 2006 Page 2 of 6 Teaching methods ..., 2008.
- [30] Alberto Leon-Garcia. Probability, Statistics, and Random Processes for Electrical Engineering. 3rd ed. edição. Upper Saddle River, NJ: Prentice Hall, 2007. ISBN: 978-0-13-147122-1.
- [31] Josef Nossek. Adaptive and Array Signal Processing. 2015.
- [32] Alan V. Oppenheim and Ronald W. Schafer. Discrete-Time Signal Processing: International Edition. 3ª edição. Upper Saddle River Munich: Pearson, Nov. 12, 2009. ISBN: 978-0-13-206709-6.
- [33] Kaare Brandt Petersen and Michael Syskind Pedersen. "The Matrix Cookbook". In: *Technical University of Denmark* 7.15 (2008), p. 510.
- [34] John Proakis and Masoud Salehi. *Digital Communications*. 5th ed. edição. Boston: Mc Graw Hill, Jan. 1, 2007. ISBN: 978-0-07-295716-7.

- [35] Simon Ramo, John R Whinnery, and Theodore Van Duzer. Fields and Waves in Communication Electronics. John Wiley & Sons, 1994. ISBN: 81-265-1525-2.
- [36] Kenneth H Rosen. "Discrete Mathematics and Its Applications (7Th Editio)". In: William C Brown Pub (2011).
- [37] Shayle R Searle and Andre I Khuri. *Matrix Algebra Useful for Statistics*. John Wiley & Sons, 2017. ISBN: 1-118-93514-4.
- [38] James Stewart. Calculus. Cengage Learning, 2011. ISBN: 1-133-17069-2.
- [39] Gilbert Strang et al. Introduction to Linear Algebra. Vol. 3. Wellesley-Cambridge Press Wellesley, MA, 1993.
- [40] Sergios Theodoridis. *Machine Learning: A Bayesian and Optimization Perspective*. 2nd ed. Academic Pr, 2020. ISBN: 978-0-12-818803-3.
- [41] Harry L Van Trees. Optimum Array Processing: Part IV of Detection, Estimation, and Modulation Theory. John Wiley & Sons, 2002. ISBN: 0-471-09390-4.
- [42] Harry L. Van Trees. Detection, Estimation, and Modulation Theory, Part I: Detection, Estimation, and Linear Modulation Theory. John Wiley & Sons, 2004.