

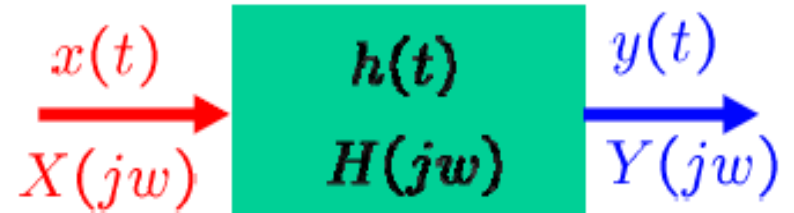
# **Introduction to Signals and Systems: V216**

## **Lecture #9**

### **Chapter 4: The Continuous-Time Fourier Transform**

# Convolution Property

- Convolution Property:



$$y(t) = x(t) * h(t) \xleftrightarrow{\mathcal{F}} Y(j\omega) = X(j\omega)H(j\omega)$$

$$= \int_{-\infty}^{\infty} x(\tau)h(t - \tau)d\tau$$

# Convolution Property

- From Convolution Integral:

$$y(t) = \int_{-\infty}^{+\infty} x(\tau) h(t - \tau) d\tau$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega$$

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$x(t - t_0) \xleftrightarrow{\mathcal{F}} e^{-j\omega t_0} X(j\omega)$$

$$\Rightarrow Y(j\omega) = \mathcal{F}\{y(t)\} = \int_{-\infty}^{+\infty} \left[ \int_{-\infty}^{+\infty} x(\tau) h(t - \tau) d\tau \right] e^{-j\omega t} dt$$

$$= \int_{-\infty}^{+\infty} x(\tau) \left[ \int_{-\infty}^{+\infty} h(t - \tau) e^{-j\omega t} dt \right] d\tau$$

$$= \int_{-\infty}^{+\infty} x(\tau) \left[ e^{-j\omega \tau} H(j\omega) \right] d\tau$$

$$= H(j\omega) \int_{-\infty}^{+\infty} x(\tau) e^{-j\omega \tau} d\tau$$

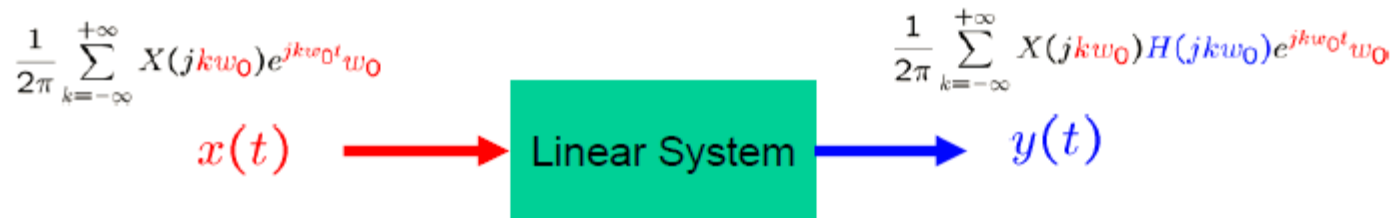
$$\rightarrow Y(j\omega) = H(j\omega) X(j\omega)$$

# Convolution Property

- From Superposition (or Linearity):

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega$$

$$= \lim_{\omega_0 \rightarrow 0} \frac{1}{2\pi} \sum_{k=-\infty}^{+\infty} X(jk\omega_0) e^{jk\omega_0 t} \omega_0$$



$$H(jk\omega_0) = \int_{-\infty}^{\infty} h(t) e^{-jk\omega_0 t} dt$$

$$y(t) = \lim_{\omega_0 \rightarrow 0} \frac{1}{2\pi} \sum_{k=-\infty}^{+\infty} X(jk\omega_0) H(jk\omega_0) e^{jk\omega_0 t} \omega_0$$

$$= \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) H(j\omega) e^{j\omega t} d\omega$$

# Convolution Property

- From Superposition (or Linearity):

$$\frac{1}{2\pi} \sum_{k=-\infty}^{+\infty} X(jkw_0) e^{jkw_0 t} \longrightarrow \frac{1}{2\pi} \sum_{k=-\infty}^{+\infty} X(jkw_0) H(jkw_0) e^{jkw_0 t}$$

$$y(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(jw) H(jw) e^{jw t} dw$$

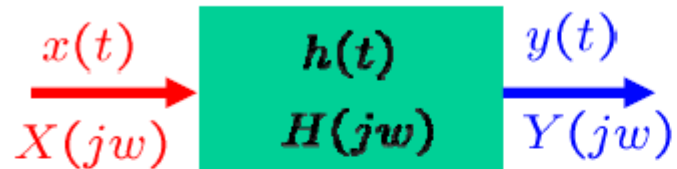
$$\text{Since } y(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} Y(jw) e^{jw t} dw$$

$$\Rightarrow Y(jw) = X(jw) H(jw)$$

$$y(t) = x(t) * h(t) \xleftrightarrow{\mathcal{F}} Y(jw) = X(jw) H(jw)$$

# Convolution Property

## ▪ Equivalent LTI Systems:



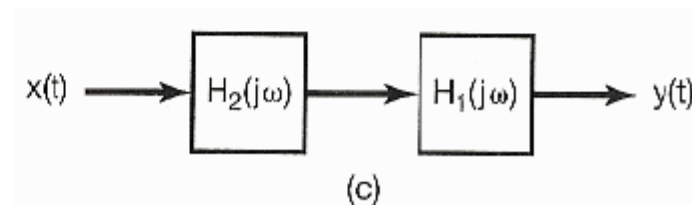
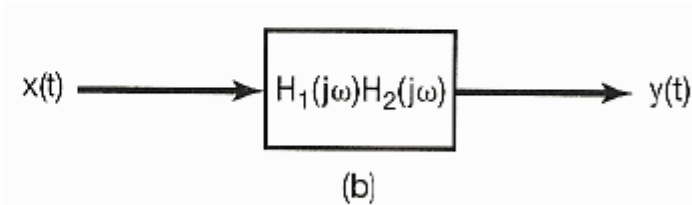
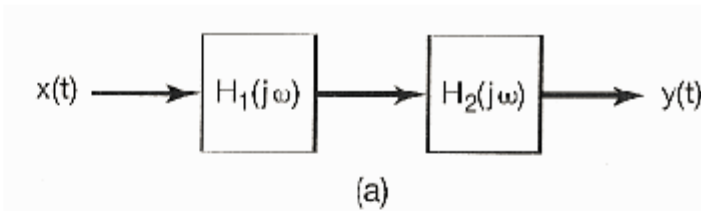
$$h(t) \xleftrightarrow{\mathcal{F}} H(j\omega)$$

impulse  
response

frequency  
response

$$y(t) = x(t) * h(t)$$

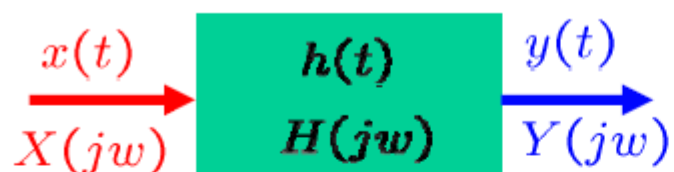
$$Y(j\omega) = X(j\omega)H(j\omega)$$



$$\Rightarrow Y(j\omega) = H_1(j\omega)H_2(j\omega)X(j\omega)$$

# Convolution Property

## ■ Example 4.15: Time Shift



$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(jw) e^{jw t} dw$$

$$X(jw) = \int_{-\infty}^{\infty} x(t) e^{-jw t} dt$$

$$x(t - t_0) \xleftrightarrow{\mathcal{F}} e^{-jw t_0} X(jw)$$

$$h(t) = \delta(t - t_0)$$

$$\Rightarrow H(jw) = e^{-jw t_0}$$

$$Y(jw) = H(jw) X(jw)$$

$$= e^{-jw t_0} X(jw)$$

$$\Rightarrow y(t) = x(t - t_0)$$

# Convolution Property

- Examples 4.16 & 17: Differentiator & Integrator

$$y(t) = \frac{d}{dt}x(t) \quad \Rightarrow \quad Y(j\omega) = j\omega X(j\omega)$$

$$\Rightarrow H(j\omega) = j\omega$$

$$y(t) = \int_{-\infty}^t x(\tau) d\tau \quad \Rightarrow \quad h(t) = u(t) \quad \text{impulse response}$$

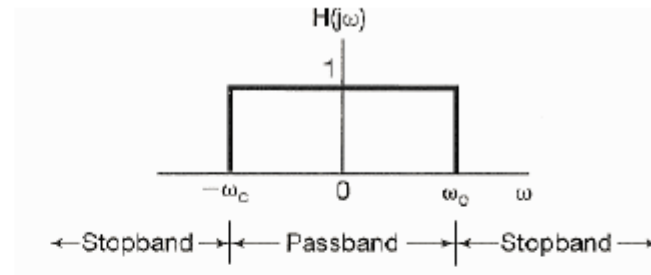
$$\Rightarrow H(j\omega) = \frac{1}{j\omega} + \pi\delta(\omega)$$



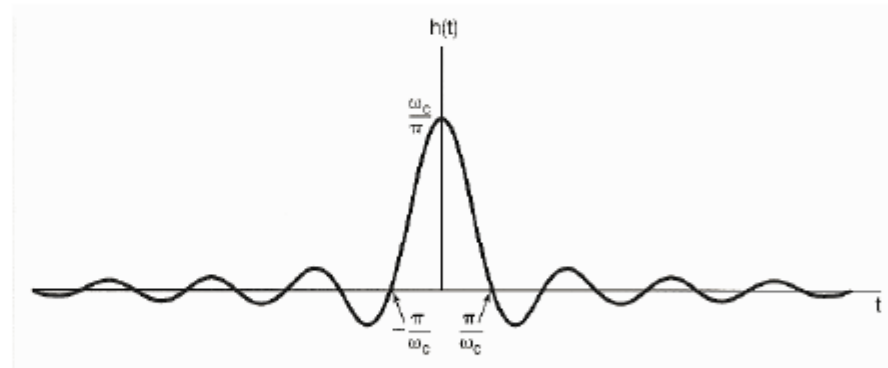
# Convolution Property

## ■ Example 4.18: Ideal Lowpass Filter

$$H(j\omega) = \begin{cases} 1, & |\omega| < \omega_c \\ 0, & |\omega| > \omega_c \end{cases}$$



$$\begin{aligned} \Rightarrow h(t) &= \frac{1}{2\pi} \int_{-\omega_c}^{+\omega_c} e^{j\omega t} d\omega \\ &= \frac{\sin(\omega_c t)}{\pi t} \end{aligned}$$




$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega$$

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

# Convolution Property

- Filter Design:

$$H(j\omega) = \int_{-\infty}^{\infty} h(t) e^{-j\omega t} dt$$


$y(t) = h(t) * x(t)$   
 $= \int_{-\infty}^{\infty} h(t - \tau) x(\tau) d\tau$

$$H(j\omega) = \int_{-\infty}^{\infty} h(t) e^{-j\omega t} dt$$

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

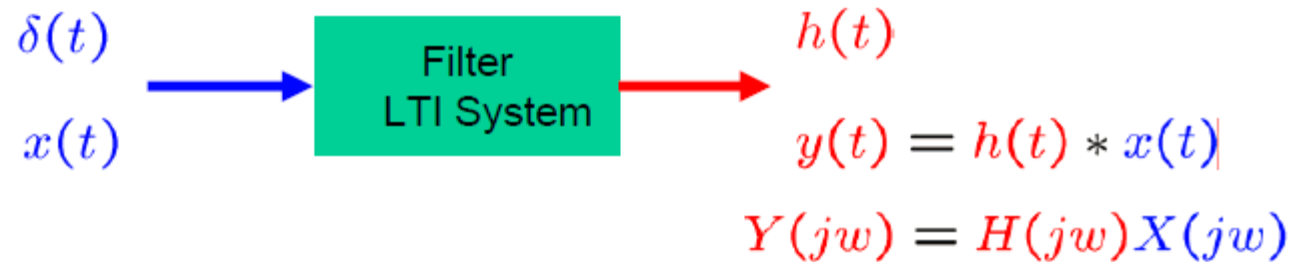
$$\Rightarrow Y(j\omega) = H(j\omega) X(j\omega)$$

$$\Rightarrow y(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} Y(j\omega) e^{j\omega t} d\omega$$

# Convolution Property

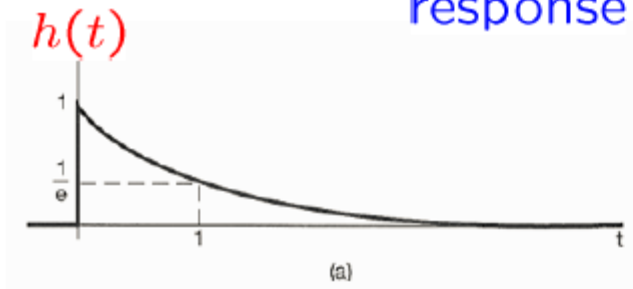
## Filter Design:

$$H(j\omega) = \int_{-\infty}^{\infty} h(t) e^{-j\omega t} dt$$

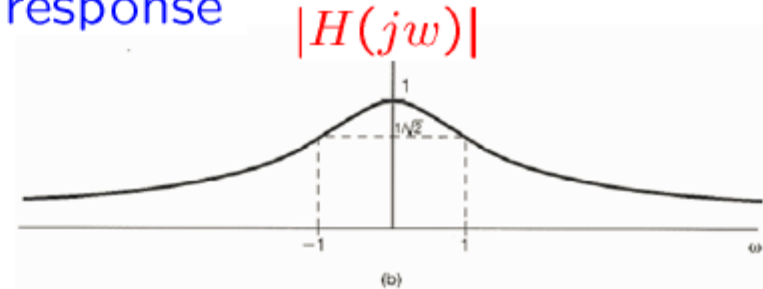


$$h(t) = e^{-t} u(t) \xleftrightarrow{\mathcal{F}} H(j\omega) = \frac{1}{j\omega + 1}$$

impulse  
response

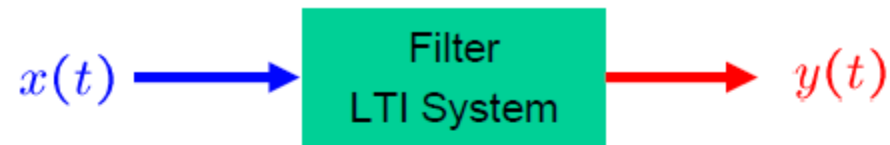


frequency  
response



# Convolution Property

- Example 4.19:



$$h(t) = e^{-at}u(t), \quad a > 0 \quad \Rightarrow \quad H(j\omega) = \frac{1}{a + j\omega}$$

$$x(t) = e^{-bt}u(t), \quad b > 0 \quad \Rightarrow \quad X(j\omega) = \frac{1}{b + j\omega}$$

$$\Rightarrow Y(j\omega) = H(j\omega)X(j\omega) = \frac{1}{a + j\omega} \frac{1}{b + j\omega}$$

$$\text{if } a \neq b \quad = \frac{1}{b - a} \left[ \frac{1}{a + j\omega} - \frac{1}{b + j\omega} \right]$$

# Convolution Property

- Example 4.19:

$$\text{if } a \neq b \quad Y(j\omega) = \frac{1}{b-a} \left[ \frac{1}{a+j\omega} - \frac{1}{b+j\omega} \right]$$
$$\Rightarrow y(t) = \frac{1}{b-a} \left[ e^{-at}u(t) - e^{-bt}u(t) \right]$$

$$\text{if } a = b \quad Y(j\omega) = \frac{1}{(a+j\omega)^2}$$

$$\text{since } e^{-at}u(t) \xleftrightarrow{\mathcal{F}} \frac{1}{a+j\omega}$$

$$\text{and } t e^{-at}u(t) \xleftrightarrow{\mathcal{F}} j \frac{d}{d\omega} \left[ \frac{1}{a+j\omega} \right] = \frac{1}{(a+j\omega)^2}$$

$$\Rightarrow y(t) = t e^{-at}u(t)$$

# Convolution Property

■ Example 4.20:

$$H(j\omega) = \int_{-\infty}^{\infty} h(t)e^{-j\omega t} dt$$

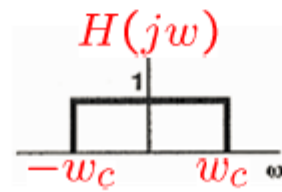
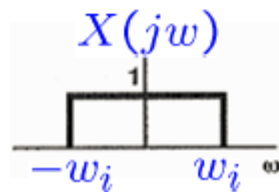


$$h(t) = \frac{\sin(\omega_c t)}{\pi t}$$

$$\Rightarrow Y(j\omega) = H(j\omega)X(j\omega)$$

$$\Rightarrow X(j\omega) = \begin{cases} 1, & |\omega| \leq \omega_i \\ 0, & \text{otherwise} \end{cases}$$

$$\Rightarrow H(j\omega) = \begin{cases} 1, & |\omega| \leq \omega_c \\ 0, & \text{otherwise} \end{cases}$$



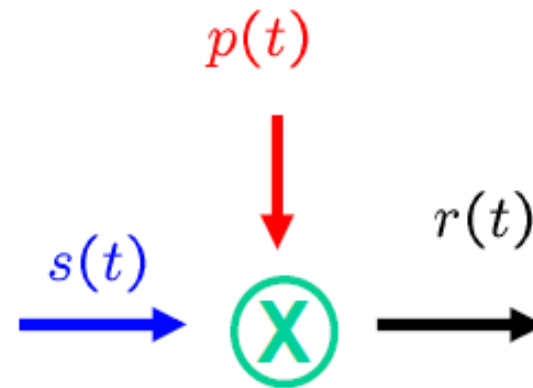
$$\omega_0 = \min(\omega_c, \omega_i)$$

$$\Rightarrow Y(j\omega) = \begin{cases} 1, & |\omega| \leq \omega_0 \\ 0, & \text{otherwise} \end{cases}$$

$$\Rightarrow y(t) = \begin{cases} \frac{\sin(\omega_c t)}{\pi t}, & \omega_c \leq \omega_i \\ \frac{\sin(\omega_i t)}{\pi t}, & \omega_c \geq \omega_i \end{cases}$$

# Multiplication Property

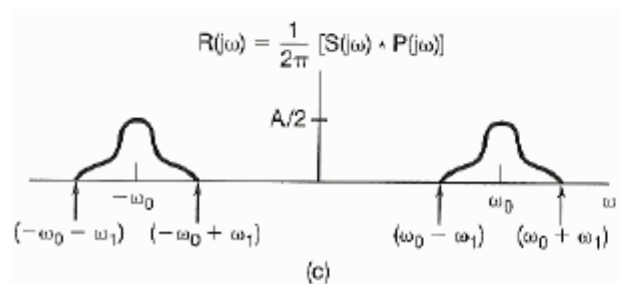
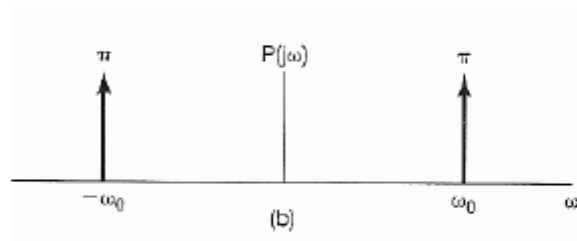
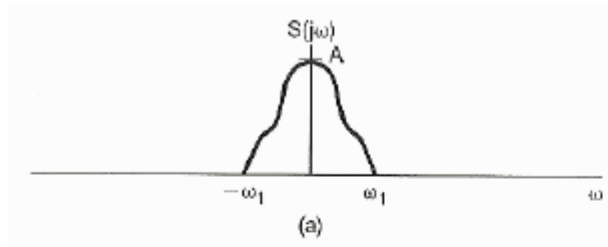
- Multiplication Property:



$$r(t) = s(t)p(t) \xleftrightarrow{\mathcal{F}} R(jw) = \frac{1}{2\pi} \int_{-\infty}^{\infty} S(j\theta) P(j(w - \theta)) d\theta$$

# Multiplication Property

## ■ Example 4.21:



$$r(t) = s(t)p(t)$$

$$s(t) \xleftrightarrow{\mathcal{F}} S(j\omega)$$

$$p(t) \xleftrightarrow{\mathcal{F}} P(j\omega)$$

$$p(t) = \cos(\omega_0 t)$$

$$P(j\omega) = \pi\delta(\omega - \omega_0) + \pi\delta(\omega + \omega_0)$$

$$R(j\omega) = \frac{1}{2\pi} [S(j\omega) * P(j\omega)]$$

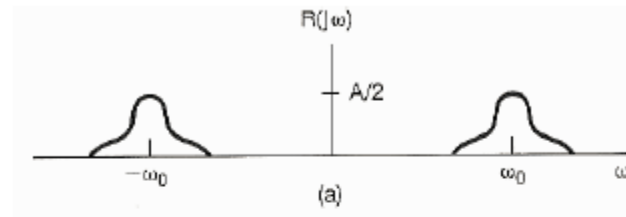
$$= \frac{1}{2\pi} \int_{-\infty}^{\infty} S(j\theta) P(j(\omega - \theta)) d\theta$$

$$= \frac{1}{2} S(j(\omega - \omega_0)) + \frac{1}{2} S(j(\omega + \omega_0))$$



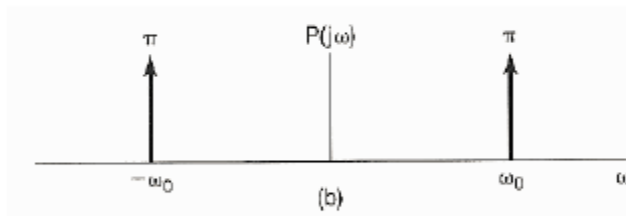
# Multiplication Property

## ■ Example 4.22:



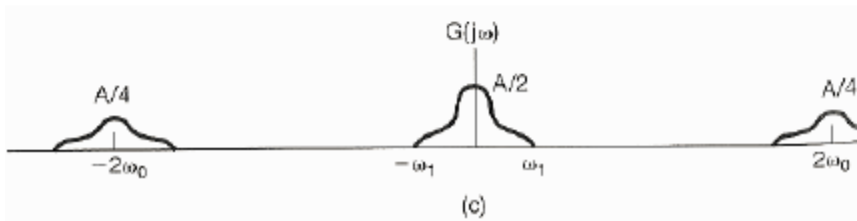
$$g(t) = r(t)p(t)$$

$$r(t) \xleftrightarrow{\mathcal{F}} R(j\omega)$$



$$p(t) \xleftrightarrow{\mathcal{F}} P(j\omega)$$

$$p(t) = \cos(\omega_0 t)$$



$$G(j\omega) = \frac{1}{2\pi} [R(j\omega) * P(j\omega)]$$

# Multiplication Property

## ■ Example 4.23:

$$x(t) = \frac{\sin(t) \sin(t/2)}{\pi t^2}$$

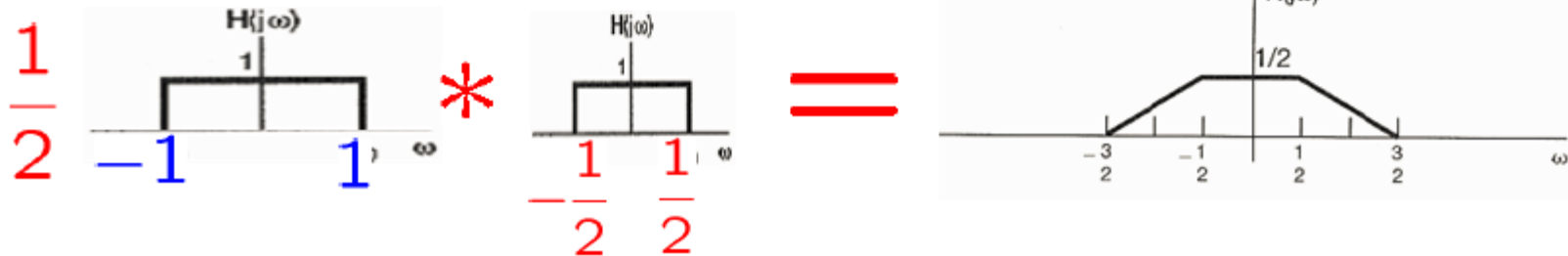
$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega$$

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$X(j\omega) = \int_{-\infty}^{\infty} \frac{\sin(t) \sin(t/2)}{\pi t^2} e^{-j\omega t} dt$$

$$= \pi \left( \frac{\sin(t)}{\pi t} \right) \left( \frac{\sin(t/2)}{\pi t} \right)$$

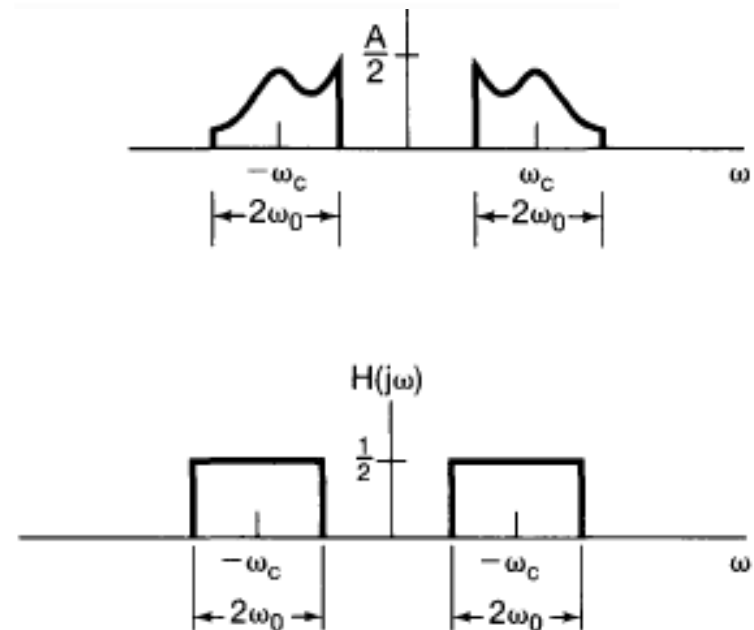
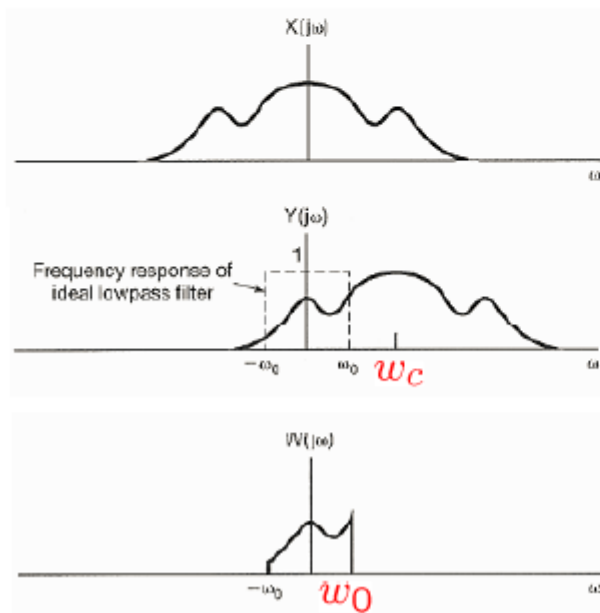
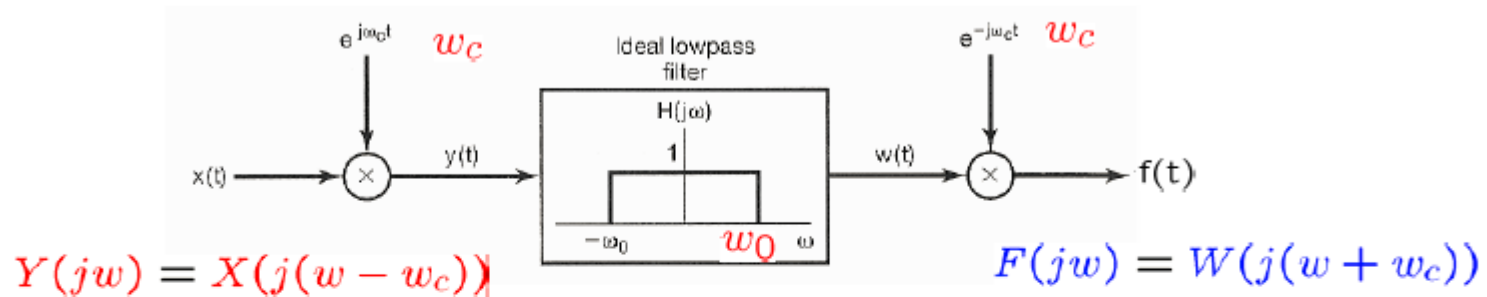
$$\Rightarrow X(j\omega) = \frac{1}{2} \mathcal{F} \left\{ \frac{\sin(t)}{\pi t} \right\} * \mathcal{F} \left\{ \frac{\sin(t/2)}{\pi t} \right\}$$



# Multiplication Property

## ■ Bandpass Filter Using Amplitude Modulation:

$$e^{j\omega_c t} \xleftrightarrow{\mathcal{F}} 2\pi\delta(\omega - \omega_c)$$

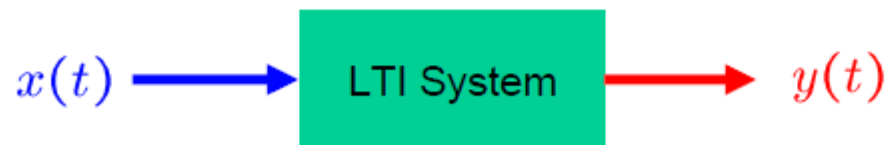


# Systems Characterized by Linear Constant Coefficient Differential Equations

- A useful class of CT LTI systems:

$$\begin{aligned} a_N \frac{d^N y(t)}{dt^N} + a_{N-1} \frac{d^{N-1} y(t)}{dt^{N-1}} + \dots + a_1 \frac{dy(t)}{dt} + a_0 y(t) \\ = b_M \frac{d^M x(t)}{dt^M} + b_{M-1} \frac{d^{M-1} x(t)}{dt^{M-1}} + \dots + b_1 \frac{dx(t)}{dt} + b_0 x(t) \end{aligned}$$

$$\sum_{k=0}^N a_k \frac{d^k y(t)}{dt^k} = \sum_{k=0}^M b_k \frac{d^k x(t)}{dt^k}$$



$$Y(j\omega) = X(j\omega)H(j\omega) \quad H(j\omega) = \frac{Y(j\omega)}{X(j\omega)}$$

# Systems Characterized by Linear Constant Coefficient Differential Equations

$$\mathcal{F} \left\{ \sum_{k=0}^N a_k \frac{d^k y(t)}{dt^k} \right\} = \mathcal{F} \left\{ \sum_{k=0}^M b_k \frac{d^k x(t)}{dt^k} \right\}$$

$$\sum_{k=0}^N a_k \mathcal{F} \left\{ \frac{d^k y(t)}{dt^k} \right\} = \sum_{k=0}^M b_k \mathcal{F} \left\{ \frac{d^k x(t)}{dt^k} \right\}$$

$$\sum_{k=0}^N a_k (j\omega)^k Y(j\omega) = \sum_{k=0}^M b_k (j\omega)^k X(j\omega)$$

$$Y(j\omega) \left[ \sum_{k=0}^N a_k (j\omega)^k \right] = X(j\omega) \left[ \sum_{k=0}^M b_k (j\omega)^k \right]$$

$$\Rightarrow H(j\omega) = \frac{Y(j\omega)}{X(j\omega)} = \frac{\sum_{k=0}^M b_k (j\omega)^k}{\sum_{k=0}^N a_k (j\omega)^k} = \frac{b_M (j\omega)^M + \dots + b_1 (j\omega) + b_0}{a_N (j\omega)^N + \dots + a_1 (j\omega) + a_0}$$

# Systems Characterized by Linear Constant Coefficient Differential Equations

- Examples 4.24 & 4.25:

$$H = \frac{Y}{X}$$

$$\frac{dy(t)}{dt} + ay(t) = x(t) \quad \Rightarrow \quad H(jw) = \frac{1}{jw + a}$$

$$(jw)Y(jw) + aY(jw) = X(jw) \quad \Rightarrow \quad h(t) = e^{-at}u(t)$$

$$\frac{d^2y(t)}{dt^2} + 4\frac{dy(t)}{dt} + 3y(t) = \frac{dx(t)}{dt} + 2x(t)$$

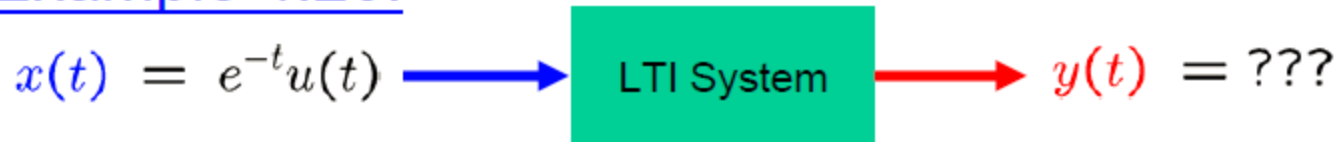
$$\Rightarrow H(jw) = \frac{(jw) + 2}{(jw)^2 + 4(jw) + 3} = \frac{(jw + 2)}{(jw + 1)(jw + 3)}$$

$$= \frac{1/2}{jw + 1} + \frac{1/2}{jw + 3}$$

$$\Rightarrow h(t) = \frac{1}{2}e^{-t}u(t) + \frac{1}{2}e^{-3t}u(t)$$

# Systems Characterized by Linear Constant Coefficient Differential Equations

- Example 4.26:



$$H(j\omega) = \frac{(j\omega + 2)}{(j\omega + 1)(j\omega + 3)}$$

$$\Rightarrow Y(j\omega) = X(j\omega)H(j\omega)$$

$$= \left[ \frac{1}{j\omega + 1} \right] \left[ \frac{j\omega + 2}{(j\omega + 1)(j\omega + 3)} \right]$$

$$= \frac{j\omega + 2}{(j\omega + 1)^2(j\omega + 3)}$$

$$= \frac{\frac{1}{4}}{j\omega + 1} + \frac{\frac{1}{2}}{(j\omega + 1)^2} - \frac{\frac{1}{4}}{j\omega + 3}$$

$$\Rightarrow y(t) = \left[ \frac{1}{4} e^{-t} + \frac{1}{2} t e^{-t} - \frac{1}{4} e^{-3t} \right] u(t)$$