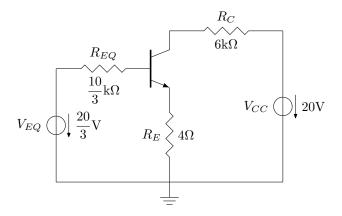
VE311 Homework 4

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Problem 1.

- (a) C_1 is a coupling capacitors which couples v_1 , C_2 is a by-pass capacitor, C_3 is a coupling capacitors which couples v_0 .
- (b) The signal voltage at the emitter of Q_1 is zero.
- (c) It is a PNP BJT transistor.

Problem 2.

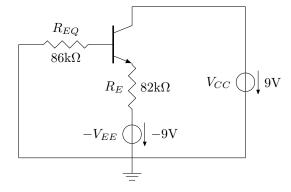


Suppose $V_{BE} = 0.7V$,

$$\begin{split} I_C &= \frac{V_{EQ} - V_{BE}}{\frac{R_{EQ}}{\beta_F} + \frac{\beta_F + 1}{\beta_F} R_E} = \frac{20/3 \, \mathrm{V} - 0.7 \, \mathrm{V}}{\frac{10/3 \, \mathrm{k}\Omega}{75} + \frac{75 + 1}{75} \cdot 4 \, \mathrm{k}\Omega} \approx 1.456 \, \mathrm{mA} \\ I_E &= \frac{V_{EQ} - V_{BE}}{\frac{R_{EQ}}{\beta_F + 1} + R_E} = \frac{20/3 \, \mathrm{V} - 0.7 \, \mathrm{V}}{\frac{10/3 \, \mathrm{k}\Omega}{75 + 1} + 4 \, \mathrm{k}\Omega} \approx 1.475 \, \mathrm{mA} \end{split}$$

 $V_{CE} = V_{CC} - I_C R_C - I_E R_E = 20 \,\text{V} - 1.456 \,\text{mA} \cdot 6 \,\text{k}\Omega - 1.475 \,\text{mA} \cdot 4 \,\text{k}\Omega = 5.364 \,\text{V}$ So the Q point is $(1.456 \,\text{mA}, \, 5.364 \,\text{V})$.

Problem 3.

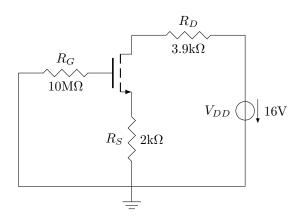


Suppose $V_{BE} = 0.7V$,

$$\begin{split} I_E &= \frac{V_{EE} - V_{BE}}{R_E} = \frac{9\,\mathrm{V} - 0.7\,\mathrm{V}}{82\,\mathrm{k}\Omega} \approx 101\,\mu\mathrm{A} \\ \\ I_C &= \frac{\beta_F}{\beta_F + 1}I_E = \frac{100}{100 + 1} \cdot 100\,\mu\mathrm{A} \\ \\ V_{CE} &= V_{CC} - I_C R_C - (-V_{BE}) = 9\,\mathrm{V} + 0.7\,\mathrm{V} = 9.7\,\mathrm{V} \end{split}$$

So the Q point is $(100 \,\mu\text{A}, 9.7 \,\text{V})$.

Problem 4.



According to the equations,

$$I_D = \frac{K_n}{2} (V_{GS} - V_{TN})^2$$
$$V_{GS} + I_D R_s = 0$$

We can get

$$V_{GS} + \frac{K_n R_s}{2} (V_{GS} - V_{TN})^2 = 0$$

$$V_{GS} = V_{TN} = \frac{1}{K_n R_s} \left(\sqrt{1 - 2K_n R_s V_{TN}} - 1 \right)$$

$$I_{D} = \frac{1}{2K_{n}R_{s}^{2}} \left(\sqrt{1 - 2K_{n}R_{s}V_{TN}} - 1 \right)^{2}$$

$$= \frac{1}{2 \cdot 400 \,\mu\text{A/V}^{2} \cdot (2 \,\text{k}\Omega)^{2}} \left(\sqrt{1 - 2 \cdot 400 \,\mu\text{A/V}^{2} \cdot 2 \,\text{k}\Omega \cdot -5 \,\text{V}} - 1 \right)^{2}$$

$$= 1.25 \,\text{mA}$$

$$V_{DS} = V_{DD} - I_{D}(R_{D} + R_{S}) = 16 \,\text{V} - 1.25 \,\text{mA} \cdot (3.9 \,\text{k}\Omega + 2 \,\text{k}\Omega) = 8.625 \,\text{V}$$

$$V_{GS} - V_{TN} = \frac{1}{K_{n}R_{s}} \left(\sqrt{1 - 2K_{n}R_{s}V_{TN}} - 1 \right)$$

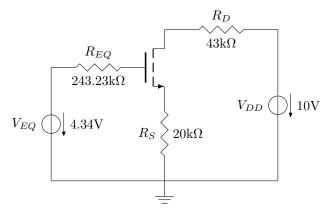
$$= \frac{1}{2 \cdot 400 \,\mu\text{A/V}^{2} \cdot 2 \,\text{k}\Omega} \left(\sqrt{1 - 2 \cdot 400 \,\mu\text{A/V}^{2} \cdot 2 \,\text{k}\Omega \cdot -5 \,\text{V}} - 1 \right)$$

$$= 1.25 \,\text{V} < V_{DS}$$

So the Q point is $(1.25 \,\mathrm{mA},\ 8.625 \,\mathrm{V})$, and it is in the saturated region.

Problem 5.

First, we draw the dc equivalent circuit:



According to the equations,

$$I_D = \frac{K_n}{2} (V_{GS} - V_{TN})^2$$
$$V_{GS} + I_D R_s = V_{EQ}$$

We can get

$$V_{GS} + \frac{K_n R_s}{2} (V_{GS} - V_{TN})^2 = V_{EQ}$$

$$V_{GS} = V_{TN} + \frac{1}{K_n R_s} \left(\sqrt{1 + 2K_n R_s (V_{EQ} - V_{TN})} - 1 \right)$$

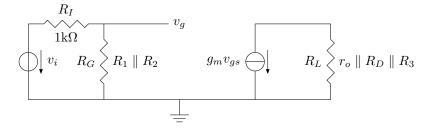
$$I_D = \frac{1}{2K_n R_s^2} \left(\sqrt{1 + 2K_n R_s (V_{EQ} - V_{TN})} - 1 \right)^2$$

$$= \frac{1}{2 \cdot 0.5 \,\text{mA/V}^2 \cdot (20 \,\text{k}\Omega)^2} \left(\sqrt{1 + 2 \cdot 0.5 \,\text{mA/V}^2 \cdot 20 \,\text{k}\Omega \cdot (4.34 \,\text{V} - 1 \,\text{V})} - 1 \right)^2$$

$$\approx 131 \,\mu\text{A}$$

$$\begin{split} V_{DS} &= V_{DD} - I_D(R_D + R_S) = 10 \, \text{V} - 131 \, \mu \text{A} \cdot (43 \, \text{k}\Omega + 20 \, \text{k}\Omega) \approx 1.75 \, \text{V} \\ V_{GS} - V_{TN} &= \frac{1}{K_n R_s} \left(\sqrt{1 + 2 K_n R_s (V_{EQ} - V_{TN})} - 1 \right) \\ &= \frac{1}{0.5 \, \text{mA/V}^2 \cdot 20 \, \text{k}\Omega} \left(\sqrt{1 + 2 \cdot 0.5 \, \text{mA/V}^2 \cdot 20 \, \text{k}\Omega} \cdot (4.34 \, \text{V} - 1 \, \text{V}) - 1 \right) \\ &\approx 0.717 \, \text{V} < V_{DS} \end{split}$$

So the Q point is $(131 \,\mu\text{A}, 1.75 \,\text{V})$, and it is in the saturated region. Then we can draw the ac equivalent circuit:



$$g_m = \frac{2I_D}{V_{GS} - V_{TN}} = \frac{2 \cdot 131 \,\mu\text{A}}{0.717 \,\text{V}} \approx 3.65 \times 10^{-4} \,\Omega^{-1}$$

$$r_o = \frac{\frac{1}{\lambda} + V_{DS}}{I_D} = \frac{\frac{1}{0.0133 \,\text{V}^{-1}} + 1.75 \,\text{V}}{131 \,\mu\text{A}} \approx 587 \,\text{k}\Omega$$

$$R_G = R_1 \parallel R_2 \approx 243.23 \,\text{k}\Omega$$

$$R_L = r_o \parallel R_D \parallel R_3 \approx 28.6 \,\text{k}\Omega$$

The voltage gain is

$$\begin{split} A_v^{CS} &= -g_m R_L \left(\frac{R_G}{R_G + R_I} \right) \\ &= -3.65 \times 10^{-4} \, \Omega^{-1} \cdot 28.6 \, \text{k}\Omega \cdot \left(\frac{243.23 \, \text{k}\Omega}{243.23 \, \text{k}\Omega + 1 \, \text{k}\Omega} \right) \\ &\approx -10.4 \end{split}$$

Problem 6.

(a)
$$g_d = \frac{I_S}{V_T} \exp\left(\frac{V_D}{V_T}\right) = \frac{8 \text{ fA}}{0.025 \text{ V}} \exp\left(\frac{0.6 \text{ V}}{0.025 \text{ V}}\right) \approx 8.48 \times 10^{-3} \,\Omega^{-1}$$
$$r_d = \frac{1}{g_d} = \frac{1}{8.48 \times 10^{-3} \,\Omega^{-1}} \approx 118 \,\Omega$$

(b)
$$g_d = \frac{I_S}{V_T} \exp\left(\frac{V_D}{V_T}\right) = \frac{8 \text{ fA}}{0.025 \text{ V}} \exp\left(\frac{0 \text{ V}}{0.025 \text{ V}}\right) \approx 3.2 \times 10^{-13} \,\Omega^{-1}$$
$$r_d = \frac{1}{q_d} = \frac{1}{3.2 \times 10^{-13} \,\Omega^{-1}} \approx 3.125 \times 10^{12} \,\Omega$$

(c)
$$g_d = \frac{I_S}{V_T} \exp\left(\frac{V_D}{V_T}\right) = \frac{1}{r_d}$$

$$V_D = V_T \ln\left(\frac{V_T}{I_S r_d}\right) = 0.025 \, \mathrm{V} \ln\left(\frac{0.025 \, \mathrm{V}}{8 \, \mathrm{fA} \cdot 10^{12} \, \Omega}\right) \approx 2.85 \times 10^{-2} \, \mathrm{V}$$

Problem 7.

$$\begin{split} r_\pi &= \frac{\beta_o V_T}{I_C} = \frac{100 \cdot 0.025 \, \mathrm{V}}{40 \, \mu \mathrm{A}} = 62.5 \, \mathrm{k}\Omega \\ g_m &= \frac{I_C}{V_T} = \frac{40 \, \mu \mathrm{A}}{0.025} = 1.6 \times 10^{-3} \, \Omega^{-1} \\ r_o &= \frac{1}{g_o} = \frac{V_A + V_{CE}}{I_C} = \frac{75 \, \mathrm{V} + 10 \, \mathrm{V}}{40 \, \mu \mathrm{A}} = 2125 \, \mathrm{k}\Omega \end{split}$$

$$R_L = r_o \parallel R_C \parallel R_3 \approx 48.85 \,\mathrm{k}\Omega$$

The input resistance is

$$R_B \parallel r_\pi \approx 38.46 \,\mathrm{k}\Omega$$

The voltage gain is

$$A_v^{CE} = -g_m R_L \left[\frac{R_B \parallel r_\pi}{R_I + (R_B \parallel r_\pi)} \right]$$
$$= -1.6 \times 10^{-3} \,\Omega^{-1} \cdot 48.85 \,\mathrm{k}\Omega \cdot \left(\frac{38.46 \,\mathrm{k}\Omega}{750 \,\Omega \cdot 38.46 \,\mathrm{k}\Omega} \right)$$
$$\approx -76.7$$