

DFT: DISCRETE FOURIER TRANSFORM

DFS: DISCRETE FOURIER SERIES

DFT: DISCRETE FOURIER TRANSFORM

IDFT: INVERSE DFT

THE SIGNIFICANCE OF DFT AND DFS: SAMPLING THE DTFT

PROPERTIES OF DFS AND DFT

- 1) Linearity
- 2) Time Shift Property (DFS) – Circular Time Shift Property (DFT)
- 3) Multiplication by a complex exponential
- 4) Duality
- 5) Symmetry Properties Real Sequences
- 6) Convolution Property
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- 7) Sampling the DTFT
- 8) Multiplication in Time Domain

IMPLEMENTING LTI SYSTEMS USING DFT

Overlap-Add

Overlap-Save

LINEAR CONVOLUTION AND CIRCULAR CONVOLUTION

DFT: DISCRETE FOURIER TRANSFORM

DTFT, $X(e^{j\omega})$, is a function of a continuous variable, ω .

However, if $x[n]$ is of finite length N , it can be considered as one period of its periodic extension

$$\tilde{x}[n] = \sum_{k=-\infty}^{\infty} x[n - kN].$$

Then, Fourier series representation of $\tilde{x}[n]$ can be used to represent $x[n]$ as well.

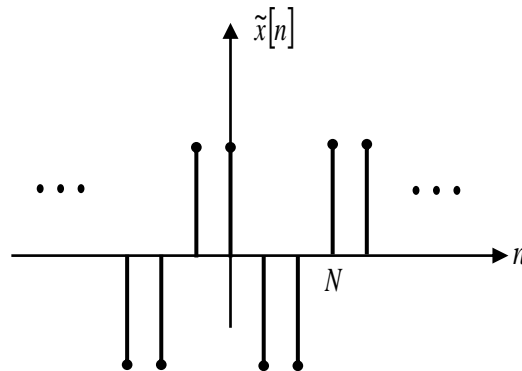
Note that the DFS representation of a periodic sequence of period N has N coefficients.

Therefore, instead of an infinite set of numbers as required by DTFT, a finite length sequence of length a finite length sequence, $x(n)$, of length N can be represented by N complex values (Fourier series coefficients).

Now, let's review Fourier series representation of periodic sequences.

DFS: DISCRETE FOURIER SERIES

Let $\tilde{x}[n]$ be a periodic sequence with fundamental period N ;



Its DFS representation is

$$\begin{aligned}\tilde{x}[n] &= \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{jk \frac{2\pi}{N} n} \\ &= \frac{1}{N} \left(\tilde{X}[0] + \tilde{X}[1] e^{j \frac{2\pi}{N} n} + \tilde{X}[2] e^{j 2 \frac{2\pi}{N} n} + \dots + \tilde{X}[N-1] e^{j(N-1) \frac{2\pi}{N} n} \right)\end{aligned}$$

This is a representation in terms of sinusoidal sequences at the fundamental frequency and its multiples (harmonic components).

$$\left\{ 1, e^{j\frac{2\pi}{N}n}, e^{j\frac{2\pi}{N}2n}, \dots, e^{j\frac{2\pi}{N}(N-1)n} \right\} = \left\{ 1, \underbrace{e^{j\omega_0 n}}_{\text{fundamental}}, \underbrace{e^{j2\omega_0 n}, \dots, e^{j(N-1)\omega_0 n}}_{\text{harmonics}} \right\} \quad \omega_0 = \frac{2\pi}{N}$$

Note that the number of frequency components depends on signal period, N .

Note also that, in this set $e^{jk\frac{2\pi}{N}n}$ and $e^{j(N-k)\frac{2\pi}{N}n}$ are complex conjugates of each other, i.e.

$$e^{jk\frac{2\pi}{N}n} = e^{-j(N-k)\frac{2\pi}{N}n}$$

The DFS coefficients $\tilde{X}[k], k=0,1,\dots,N-1$, are obtained as

$$\begin{aligned}\tilde{X}[k] &= \sum_{n=0}^{N-1} \tilde{x}[n] e^{-jk \frac{2\pi}{N} n} \\ &= \tilde{x}[0] + \tilde{x}[1] e^{-jk \frac{2\pi}{N}} + \tilde{x}[2] e^{-jk \frac{2\pi}{N} 2} + \dots + \tilde{x}[N-1] e^{-jk \frac{2\pi}{N} (N-1)}\end{aligned}$$

To obtain this expression multiply

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] e^{jk \frac{2\pi}{N} n}$$

by

$$e^{-jm \frac{2\pi}{N} n}$$

and sum over $n=0,1,2,\dots,N-1$.

$$\begin{aligned}& \sum_{n=0}^{N-1} \tilde{x}[n] e^{-jm \frac{2\pi}{N} n} \\ &= \frac{1}{N} \sum_{n=0}^{N-1} \sum_{k=0}^{N-1} \tilde{X}[k] e^{jk \frac{2\pi}{N} n} e^{-jm \frac{2\pi}{N} n} \\ &= \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] \underbrace{\sum_{n=0}^{N-1} e^{j(k-m) \frac{2\pi}{N} n}}_{=\begin{cases} N & \text{if } k-m=0 \\ 0 & \text{if } k-m \neq 0 \end{cases} \text{ or an integer multiple of } 2\pi}\end{aligned}$$

$$\Rightarrow \sum_{n=0}^{N-1} \tilde{x}[n] e^{-j \frac{2\pi}{N} mn} = \tilde{X}[m]$$

Note that $\tilde{X}[k]$ is periodic with N since

$$\tilde{X}[k + rN] = \sum_{n=0}^{N-1} \tilde{x}[n] e^{-j\frac{2\pi}{N}kn} \underbrace{e^{-j\frac{2\pi}{N}rNn}}_1 = \tilde{X}[k]$$

So it is sufficient to keep N values.

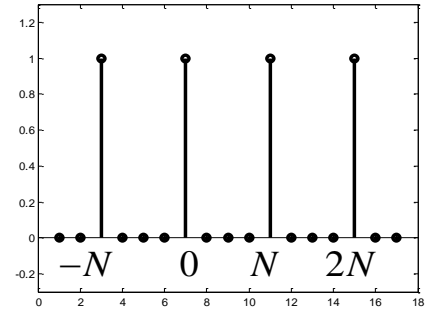
Notation: For convenience define $W_N \triangleq e^{-j\frac{2\pi}{N}}$

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] W_N^{-kn} \qquad \tilde{X}[k] = \sum_{n=0}^{N-1} \tilde{x}[n] W_N^{kn}$$

$W_N^{-kn} = e^{j\frac{2\pi}{N}kn}$ is the k^{th} harmonic.

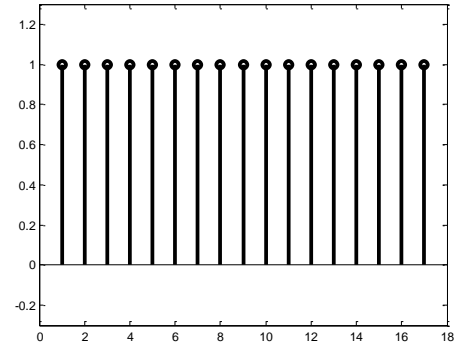
Ex:

$$\tilde{x}[n] = \sum_{r=-\infty}^{\infty} \delta[n - rN]$$



DFS coefficients of $\tilde{x}[n]$

$$\tilde{X}[k] = \sum_{n=0}^{N-1} \delta[n] W_N^{kn} = 1 \quad (\text{independent of } N)$$



DFS representation of $\tilde{x}[n]$

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} W_N^{-kn} = \frac{1}{N} \sum_{k=0}^{N-1} e^{j\frac{2\pi}{N}kn} = \frac{1}{N} N \sum_{k=0}^{N-1} \delta[n - kN] = \sum_{k=0}^{N-1} \delta[n - kN]$$

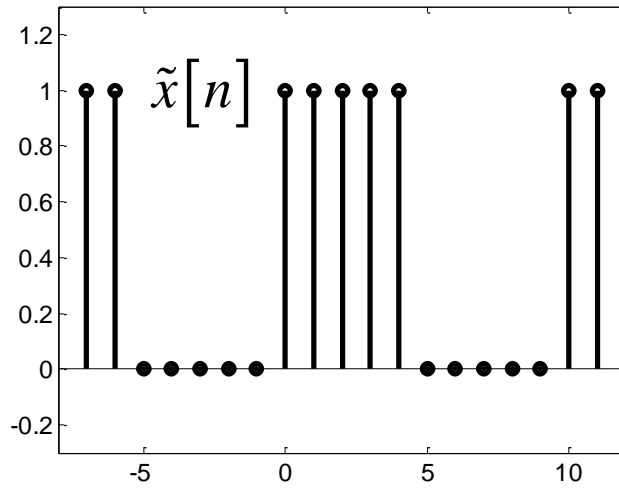
For $N = 2$

$$\tilde{x}[n] = \frac{1}{2} \sum_{k=0}^1 W_2^{-kn} = \frac{1}{2} (1 + e^{j\pi n}) = \begin{cases} 1 & \text{if } n \text{ is even} \\ 0 & \text{if } n \text{ is odd} \end{cases}$$

For $N = 3$

$$\tilde{x}[n] = \frac{1}{3} \sum_{k=0}^2 W_3^{-kn} = \frac{1}{3} \left(1 + e^{j\frac{2\pi}{3}n} + e^{j\frac{4\pi}{3}n} \right) = \begin{cases} 1 & \text{if } n \text{ is a multiple of 3} \\ 0 & \text{otherwise} \end{cases}$$

Ex:

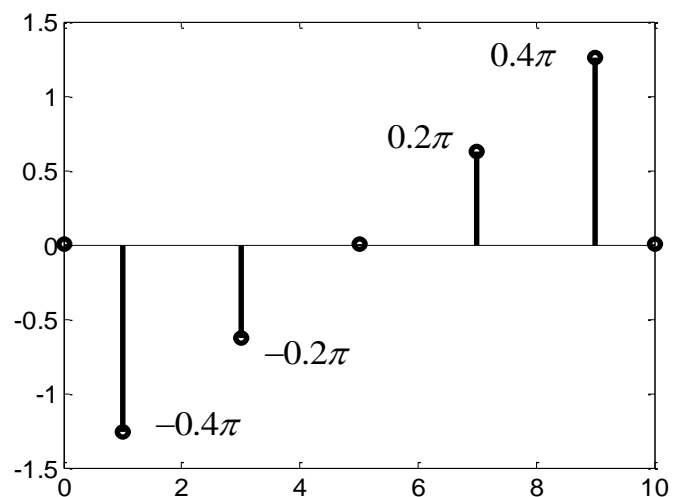
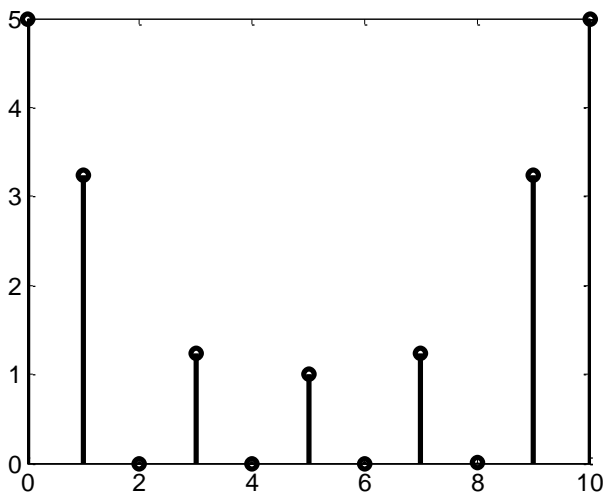


$N = 10$

DFT coefficients of $\tilde{x}[n]$

$$\begin{aligned}\tilde{X}[k] &= \sum_{n=0}^{4} W_{10}^{kn} = \frac{1 - W_{10}^{5k}}{1 - W_{10}^k} = \frac{1 - e^{-j\pi k}}{1 - e^{-j\frac{\pi}{5}k}} \\ &= \frac{e^{-j\frac{\pi}{2}k}}{e^{-j\frac{\pi}{10}k}} \frac{e^{j\frac{\pi}{2}k} - e^{-j\frac{\pi}{2}k}}{e^{j\frac{\pi}{10}k} - e^{-j\frac{\pi}{10}k}} = e^{-j\frac{4\pi}{10}k} \frac{\sin \frac{\pi}{2}k}{\sin \frac{\pi}{10}k}\end{aligned}$$

$\tilde{X}[k]$ is periodic with $N = 10$.

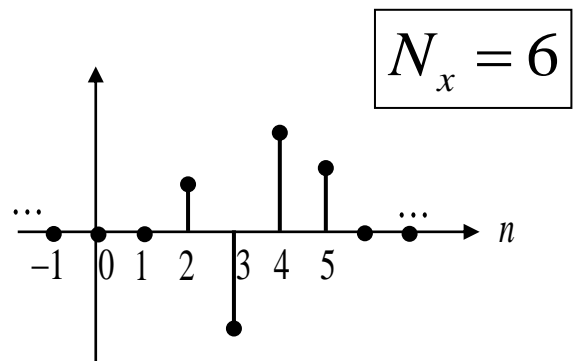
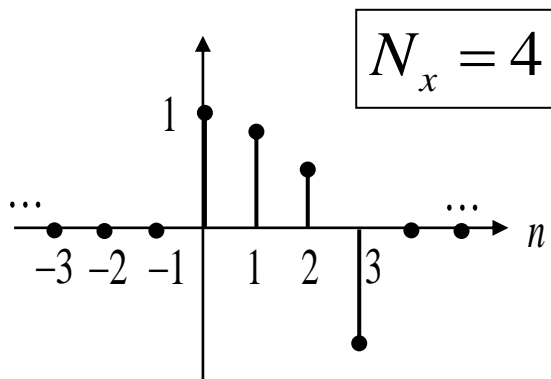


DFT: DISCRETE FOURIER TRANSFORM

Let $x[n]$ be a finite length (length = N_x) sequence such that

$$x[n] = 0 \quad \text{for} \quad n < 0 \quad \text{and} \quad n > N_x - 1$$

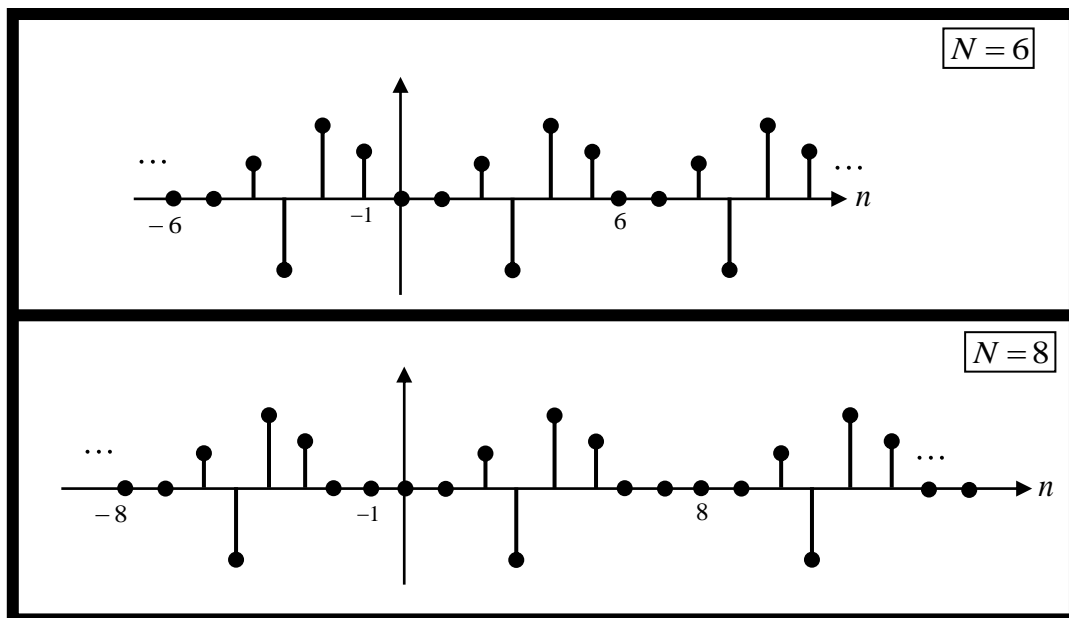
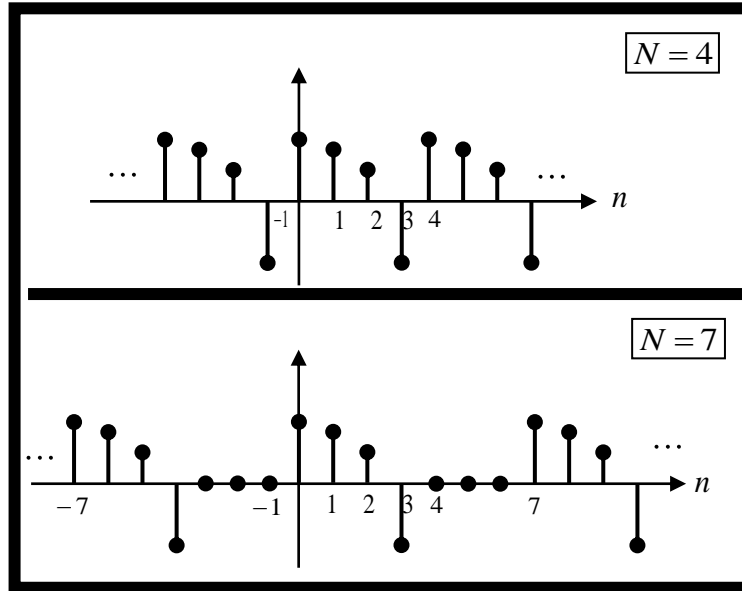
Ex:



Let

$$\tilde{x}[n] = \sum_{p=-\infty}^{\infty} x[n - pN], \quad N \geq N_x,$$

be its periodic extension with period N :



Let

$$\tilde{X}[k], k \in Z$$

be the DFS coefficients of

$$\tilde{x}[n].$$

Then, **N -point DFT** of $x[n]$ is defined as

$$X[k] = \begin{cases} \sum_{n=0}^{N-1} x[n] e^{j\frac{2\pi}{N}kn} & k = 0, 1, \dots, N-1 \\ 0 & \text{o.w.} \end{cases}$$

or

$$X[k] = \begin{cases} \tilde{X}[k] & k = 0, 1, \dots, N-1 \\ 0 & \text{o.w.} \end{cases}$$

IDFT: INVERSE DFT

$$x[n] = \begin{cases} \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j\frac{2\pi}{N}kn} & n = 0, 1, \dots, N-1 \\ 0 & n \neq 0, 1, \dots, N-1 \end{cases}$$

Notation: Modulo

Define

$$((n))_N \triangleq (n) \bmod N$$

then

$$\tilde{X}[k] = X[((k))_N]$$

and

$$\tilde{x}[n] = x[((n))_N]$$

```
>> x=1:5      x =   1   2   3   4   5
```

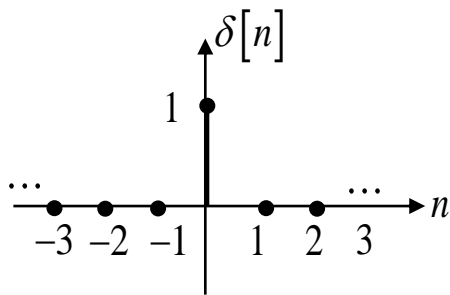
```
>> for i=1:5
```

```
    y(i) = x(mod(-(i-1),5)+1);      (since vector indices start from 1 in MATLAB)
```

```
end
```

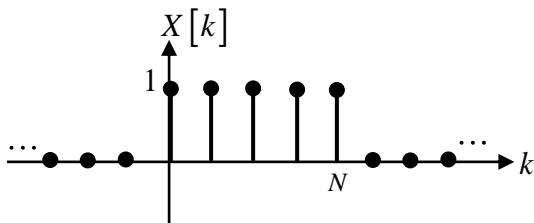
```
>> y y =   1   5   4   3   2
```

Ex: $x[n] = \delta[n]$

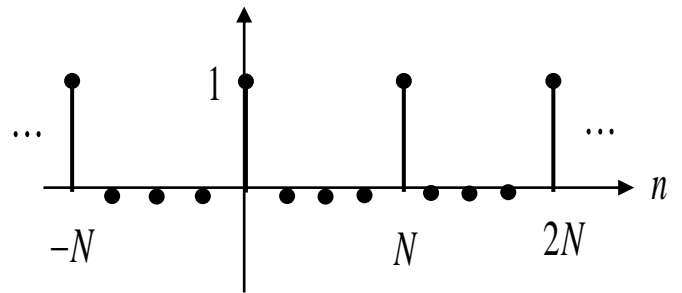


$$N_x = 1$$

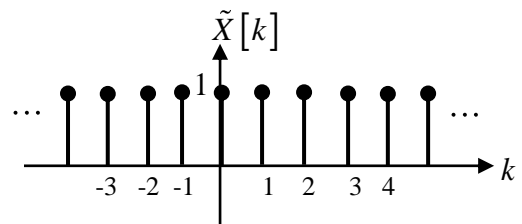
$$X[k] = \begin{cases} \tilde{X}[k] & k = 0, 1, 2, \dots, N-1 \\ 0 & \text{ow.} \end{cases}$$



$$\tilde{x}[n] = \sum_{p=-\infty}^{\infty} \delta[n - pN]$$



$$\tilde{X}[k] = \sum_{n=0}^{N-1} \delta[n] e^{-j\frac{2\pi}{N}kn} = 1 \quad k \in \mathbb{Z}$$



Using 1-point DFT, $x[n] = \delta[n]$ can be obtained by IDFT as,

$$x[n] = \begin{cases} \frac{1}{1} (X[0] W_1^{-0n}) = \frac{1}{1} (1 \times 1) = 1 & n = 0 \\ 0 & o.w. \end{cases}$$

Using 2-point DFT, $x[n] = \delta[n]$ can be obtained by IDFT as,

$$x[n] = \begin{cases} \frac{1}{2} (X[0] W_2^{-0n} + X[1] W_2^{-1n}) & n = 0,1 \\ 0 & o.w. \end{cases}$$

$$x[n] = \begin{cases} \frac{1}{2} (X[0] \times 1 + X[1] e^{j\pi n}) & n = 0,1 \\ 0 & o.w. \end{cases}$$

$$x[0] = \frac{1}{2} (1 + 1) = 1 \quad x[1] = \frac{1}{2} (1 - 1) = 0$$

Using 3-point DFT, $x[n] = \delta[n]$ can be obtained by IDFT as,

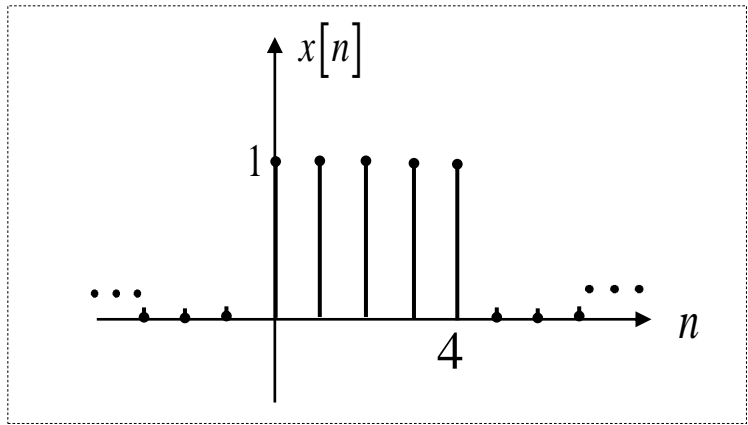
$$x[n] = \begin{cases} \frac{1}{3} (X[0] W_3^{-0n} + X[1] W_3^{-1n} + X[2] W_3^{-2n}) & n = 0,1,2 \\ 0 & o.w. \end{cases}$$

$$x[n] = \begin{cases} \frac{1}{3} \left(X[0] \times 1 + X[1] e^{j\frac{2\pi}{3}n} + X[2] e^{j2\frac{2\pi}{3}n} \right) & n = 0,1,2 \\ 0 & o.w. \end{cases}$$

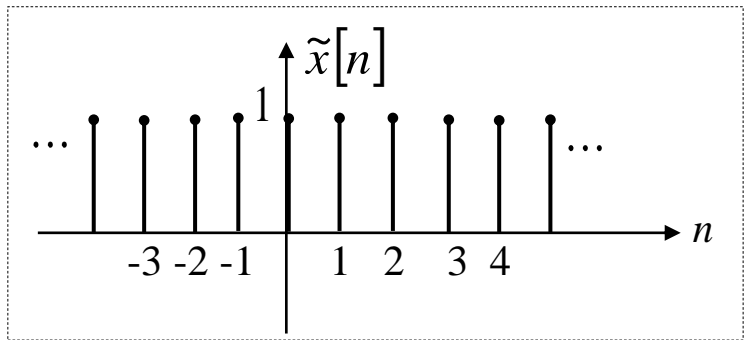
$$x[0] = \frac{1}{3} (1 + 1 + 1) = 1 \quad x[1] = \frac{1}{3} \left(1 + e^{j\frac{2\pi}{3}} + e^{j2\frac{2\pi}{3}} \right) = 0 \quad x[2] = \frac{1}{3} \left(1 + e^{j\frac{2\pi}{3}2} + e^{j2\frac{2\pi}{3}2} \right) = 0$$

Ex: $x[n]: \dots 0 \ 0 \ \underset{\substack{\uparrow \\ n=0}}{1} \ 1 \ 1 \ 1 \ 1 \ 0 \ 0 \dots$

Length of $x[n]$ is 5.



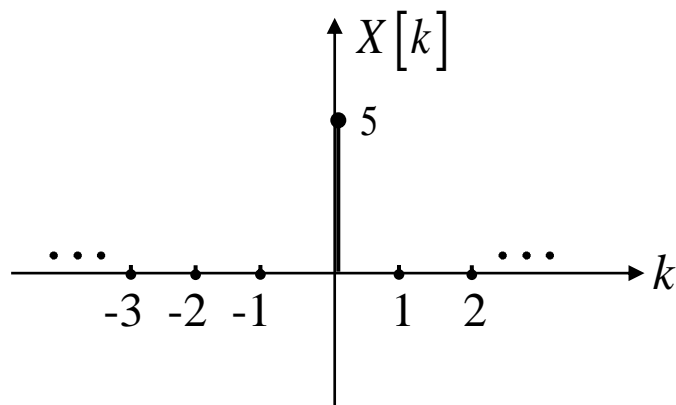
$$\tilde{x}[n] = \sum_{p=-\infty}^{\infty} x[n-5p]$$



Let's consider 5-point DFT ($N = 5$), so

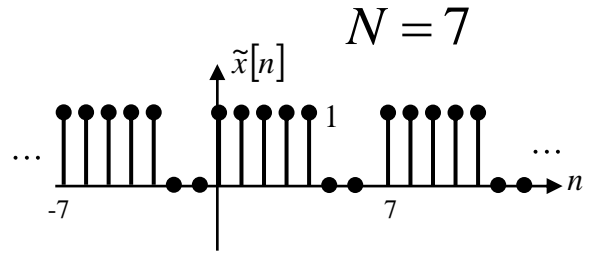
$$X[k] = \begin{cases} \sum_{n=0}^4 x[n] W_5^{kn} & k = 0, 1, \dots, 4 \\ 0 & \text{o.w.} \end{cases}$$

$$\sum_{n=0}^4 x[n] W_5^{kn} = \frac{1 - W_5^{k5}}{1 - W_5^k} = \frac{1 - e^{-j2\pi k}}{1 - e^{-j\frac{2\pi}{5}k}} = \begin{cases} 5 & k = 0 \\ 0 & k = 1, 2, 3, 4 \end{cases} = 5\delta[k]$$



Ex: Continued

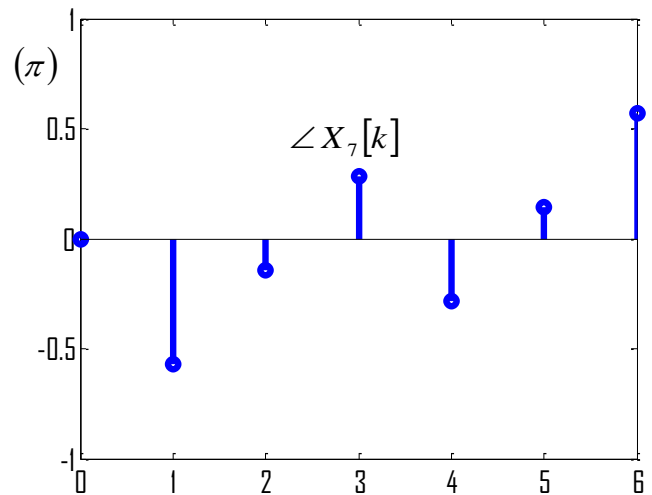
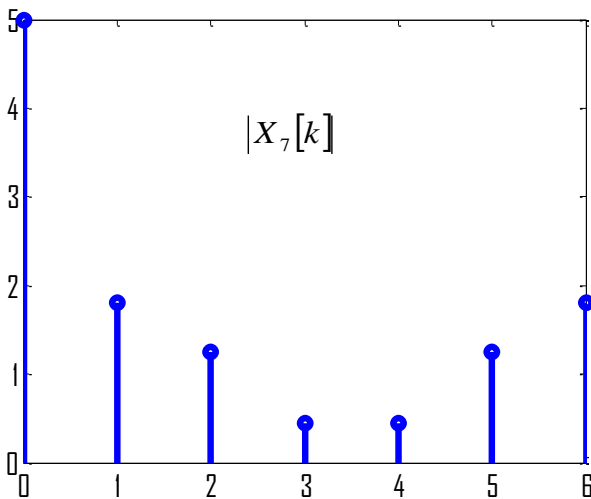
$$\tilde{x}[n] = \sum_{p=-\infty}^{\infty} x[n-7p]$$



Now consider 7-point DFT ($N = 7$), so

$$X[k] = \begin{cases} \sum_{n=0}^6 x[n] W_7^{kn} & k = 0, 1, \dots, 6 \\ 0 & \text{ow.} \end{cases}$$

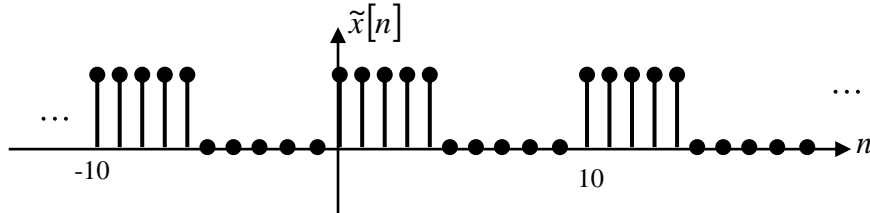
$$\sum_{n=0}^6 x[n] W_7^{kn} = \frac{1 - W_7^{k7}}{1 - W_7^k} = \frac{1 - e^{-j\frac{10\pi}{7}k}}{1 - e^{-j\frac{2\pi}{7}k}} = \frac{e^{-j\frac{5\pi}{7}k} (e^{j\frac{5\pi}{7}k} - e^{-j\frac{5\pi}{7}k})}{e^{-j\frac{\pi}{7}k} (e^{j\frac{\pi}{7}k} - e^{-j\frac{\pi}{7}k})} = e^{-j\frac{4\pi}{7}k} \frac{\sin(\frac{5\pi}{7}k)}{\sin(\frac{\pi}{7}k)}$$



Now consider 10-point DFT ($N = 10$)

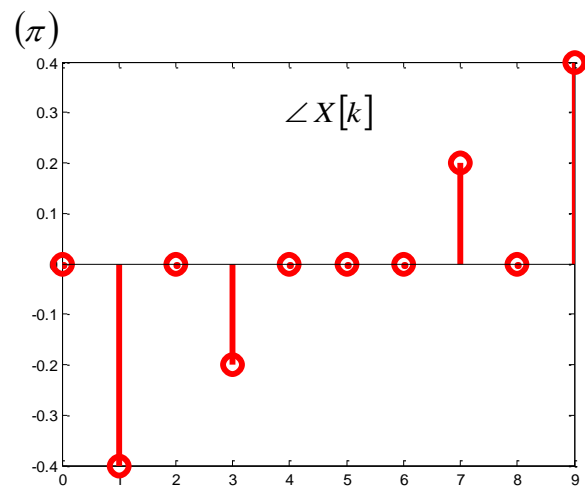
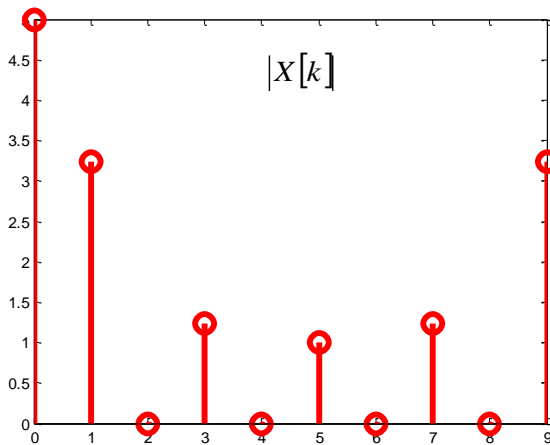
$$\tilde{x}[n] = \sum_{p=-\infty}^{\infty} x[n-10p] \quad N=10$$

So



$$X[k] = \begin{cases} \sum_{n=0}^4 x[n] W_{10}^{kn} & k = 0, 1, \dots, 9 \\ 0 & \text{ow.} \end{cases}$$

$$\sum_{n=0}^4 x[n] W_{10}^{kn} = \frac{1 - W_{10}^{k5}}{1 - W_{10}^k} = \frac{1 - e^{-j\pi k}}{1 - e^{-j\frac{\pi k}{5}}} = \frac{e^{-j\frac{\pi k}{2}} (e^{j\frac{\pi k}{2}} - e^{-j\frac{\pi k}{2}})}{e^{-j\frac{\pi k}{10}} (e^{j\frac{\pi k}{10}} - e^{-j\frac{\pi k}{10}})} = e^{-j\frac{4\pi}{10}k} \frac{\sin(\frac{\pi k}{2})}{\sin(\frac{\pi}{10}k)}$$



Note that all of the above DFTs can be used to get $x[n]$ back!

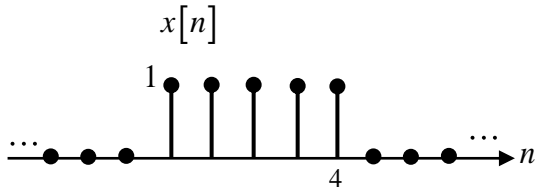
THE SIGNIFICANCE OF DFT AND DFS: SAMPLING THE DTFT

$\tilde{X}[k]$ are N uniformly spaced samples of $X(e^{j\omega})$.

Therefore, also are $X[k]$.

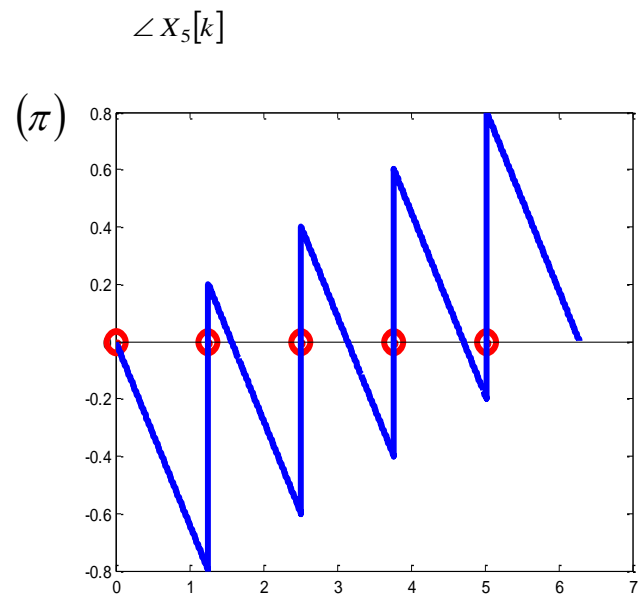
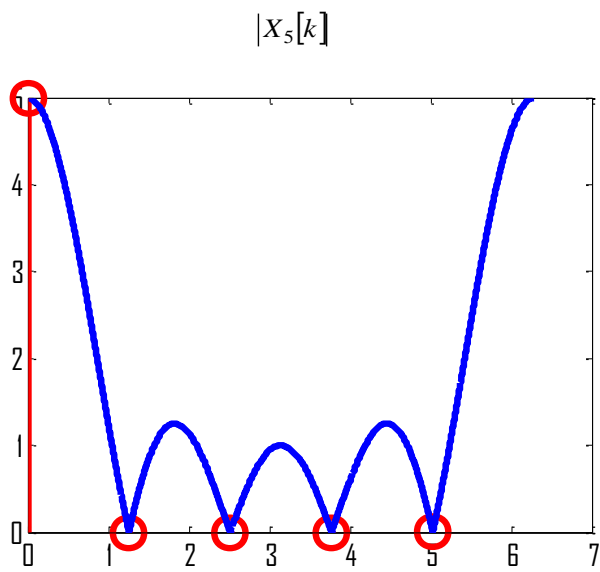
$$\begin{aligned} X[k] &= X(e^{j\omega}) \Big|_{\omega=\frac{2\pi}{N}k} \\ &= \sum_{n=0}^{N-1} x[n] e^{-j\left(\frac{2\pi}{N}k\right)n} \quad k = 0, 1, \dots, N-1 \end{aligned}$$

Ex: Continued

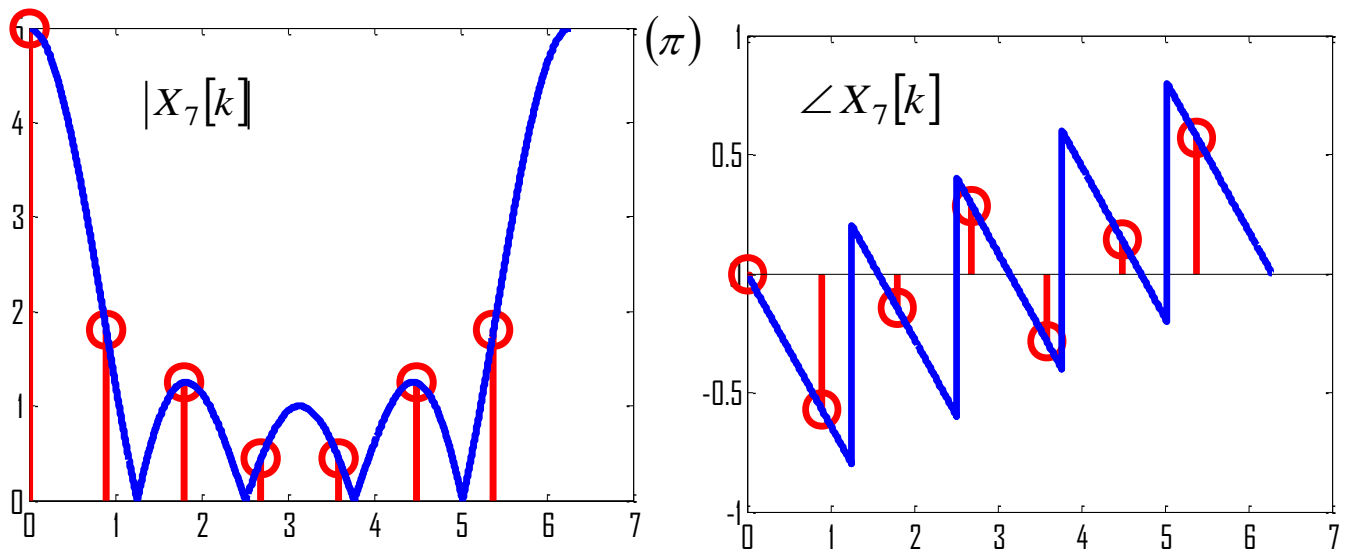


$$X(e^{j\omega}) = \sum_{n=0}^4 e^{-j\omega n} = e^{-j2\omega} \frac{\sin\left(\frac{5\omega}{2}\right)}{\sin\left(\frac{\omega}{2}\right)}$$

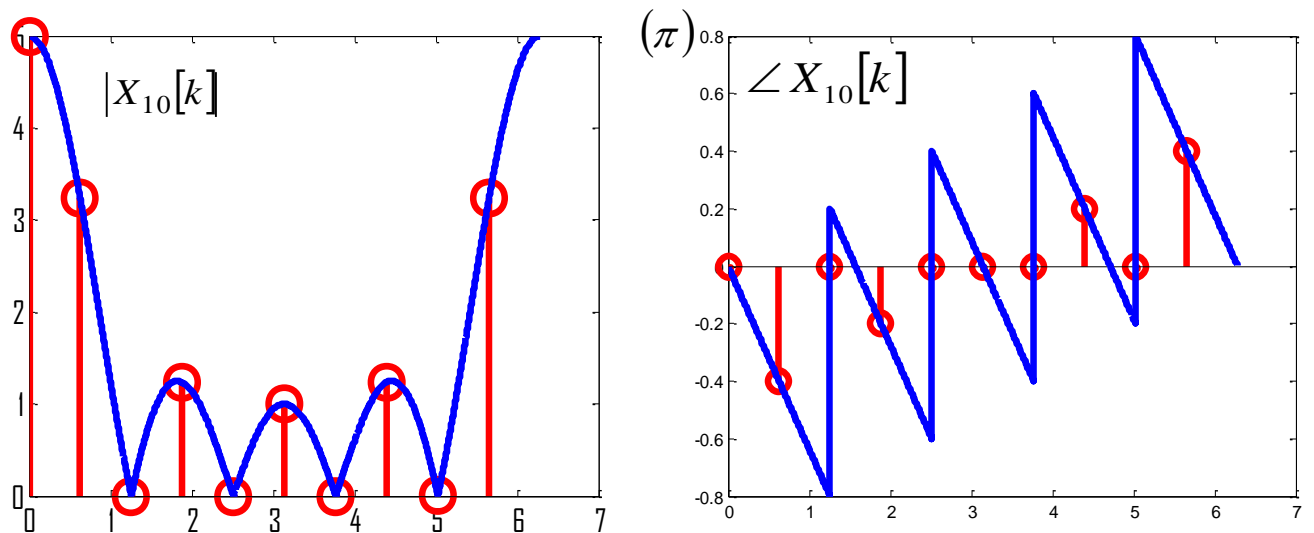
5-point DFT



7-point DFT



10-point DFT



HORIZONTAL SCALE !

PROPERTIES OF DFS AND DFT

Effectively, properties of DFT and DFS are the same.

We just need to fit the notation!

1) Linearity

DFS

Let $\tilde{x}_1[n]$ and $\tilde{x}_2[n]$ be periodic sequences with the same period N .

$$\tilde{x}_1[n] \xleftrightarrow{\text{DFS}} \tilde{X}_1[k]$$

$$\tilde{x}_2[n] \xleftrightarrow{\text{DFS}} \tilde{X}_2[k]$$

Then

$$\tilde{x}_3[n] = a\tilde{x}_1[n] + b\tilde{x}_2[n] \xleftrightarrow{\text{DFS}} \tilde{X}_3[k] = a\tilde{X}_1[k] + b\tilde{X}_2[k]$$

DFT

Let $x_1[n]$ and $x_2[n]$ be finite-length sequences

$$x_1[n] \xleftrightarrow{N\text{-point DFT}} X_1[k]$$

$$x_2[n] \xleftrightarrow{N\text{-point DFT}} X_2[k]$$

Then

$$x_3[n] = ax_1[n] + bx_2[n] \xleftrightarrow{N\text{-point DFT}} X_3[k] = aX_1[k] + bX_2[k]$$

Ex: Let

$$x_1[n] = \delta[n] + \delta[n-1] \xleftrightarrow{\text{4-point DFT}} X_1[k] = 1 + e^{-j\frac{2\pi}{4}k}$$

(note that $x_1[n]$ is a 2-point sequence)

and

$$\begin{aligned} x_2[n] = \delta[n] + \delta[n-1] + \delta[n-2] + \delta[n-3] &\xleftrightarrow{\text{4-point DFT}} X_2[k] = 1 + e^{-j\frac{2\pi}{4}k} + e^{-j\frac{2\pi}{4}k2} + e^{-j\frac{2\pi}{4}k3} \\ &= 1 + e^{-j\frac{\pi}{2}k} + e^{-j\pi k} + e^{-j\frac{3\pi}{2}k} \end{aligned}$$

$$\begin{aligned} x_3[n] = x_1[n] + x_2[n] &= 2\delta[n] + 2\delta[n-1] + \delta[n-2] + \delta[n-3] \\ &\xleftrightarrow{\text{4-point DFT}} X_3[k] = 2 + 2e^{-j\frac{\pi}{2}k} + e^{-j\pi k} + e^{-j\frac{3\pi}{2}k} = X_1[k] + X_2[k] \end{aligned}$$

2) Time Shift Property (DFS) – Circular Time Shift Property (DFT)

Shift Property (DFS)

$$\tilde{x}[n] \xleftrightarrow{\text{DFS}} \tilde{X}[k] \Rightarrow \tilde{x}[n-\Delta] \xleftrightarrow{\text{DFS}} e^{-j\frac{2\pi}{N}k\Delta} \tilde{X}[k] = W_N^{k\Delta} \tilde{X}[k]$$

Proof:

$$\tilde{y}[n] = \tilde{x}[n-\Delta]$$

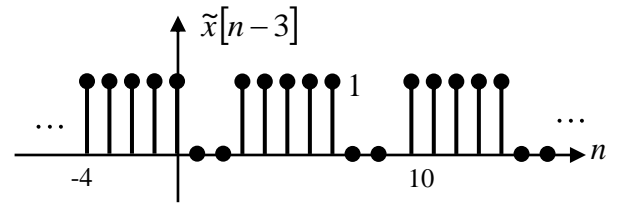
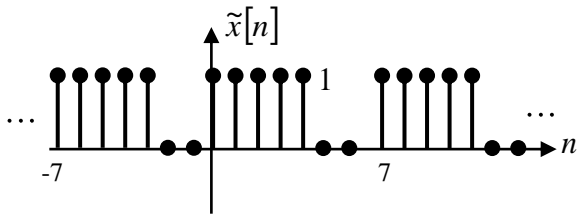
$$\tilde{Y}[k] = \sum_{n=0}^{N-1} \tilde{x}[n-\Delta] e^{-jk\frac{2\pi}{N}n} \quad \text{let } m = n - \Delta$$

$$\tilde{Y}[k] = e^{-j\frac{2\pi}{N}\Delta} \sum_{m=-\Delta}^{N-1-\Delta} \tilde{x}[m] e^{-j\frac{2\pi}{N}m}$$

Since $\tilde{x}[m]$ and $e^{-j\frac{2\pi}{N}m}$ are periodic with N and the summation is over N consecutive values

$$\tilde{Y}[k] = e^{-j\frac{2\pi}{N}\Delta} \sum_{m=0}^{N-1} \tilde{x}[m] e^{-j\frac{2\pi}{N}m} = e^{-j\frac{2\pi}{N}\Delta} \tilde{X}[k]$$

Ex:



$$\tilde{X}[k] = e^{-j\frac{4\pi}{7}k} \frac{\sin\left(\frac{5\pi}{7}k\right)}{\sin\left(\frac{\pi}{7}k\right)}$$

$$\tilde{x}[n-3] \xleftrightarrow{DFS} W_7^{3k} \tilde{X}[k]$$

$$= \underbrace{W_7^{3k} e^{-j\frac{4\pi}{7}k}}_{e^{-j\frac{2\pi}{7}5k}} \frac{\sin\left(\frac{5\pi}{7}k\right)}{\sin\left(\frac{\pi}{7}k\right)}$$

Circular Shift Property (DFT)

$$x[n] \xleftrightarrow{\text{DFT}} X[k]$$
$$a[n] = ? \xleftrightarrow{\text{DFT}} W_N^{k\Delta} X[k]$$

Since

$$x[n] \triangleq \begin{cases} \tilde{x}[n] & n = 0, 1, \dots, N-1 \\ 0 & n \neq 0, 1, \dots, N-1 \end{cases}$$

$$\Rightarrow a[n] \triangleq \begin{cases} \tilde{x}[n-\Delta] & n = 0, 1, \dots, N-1 \\ 0 & n \neq 0, 1, \dots, N-1 \end{cases}$$

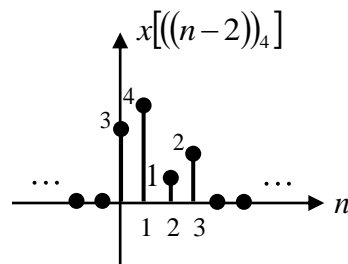
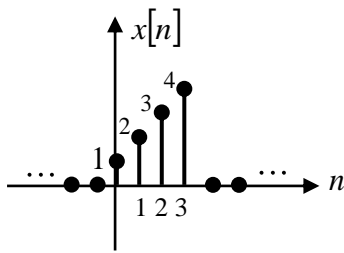
Or can be written as

$$a[n] = \begin{cases} x[((n-\Delta))_N] & n = 0, 1, \dots, N-1 \\ 0 & o.w. \end{cases}$$

Ex: If “signal length = DFT length”.

$$x[n] \xleftrightarrow{\text{4-point DFT}} X[k]$$

$$x[((n-2))_4] \xleftrightarrow{\text{4-point DFT}} W_4^{k2} X[k]$$



Distorted

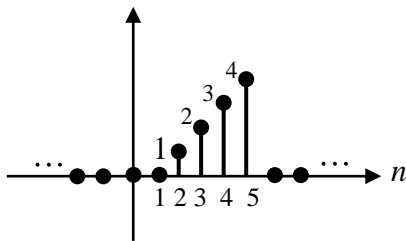
Ex: continued, “signal length < DFT length”.

If 6-point DFT is used $x[n] \xleftrightarrow{\text{6-point DFT}} X[k]$

$$x[(n-2)_6] \xleftrightarrow{\text{6-point DFT}} W_6^{2k} X[k]$$

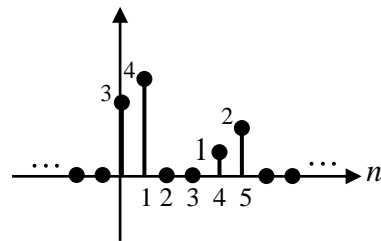
$$x[((n+2))_6] \xleftrightarrow{\text{6-point DFT}} W_6^{-k2} X[k]$$

$$x[((n-2))_6]$$



Undistorted

$$x[((n-4))_6] = x[((n+2))_6]$$



Distorted

3) Multiplication by a complex exponential

DFS

For a periodic sequence with period N

$$e^{j\frac{\Delta 2\pi}{N}n} \tilde{x}[n] = W_N^{-\Delta n} \tilde{x}[n] \quad \leftrightarrow \quad \tilde{X}[k - \Delta]$$

DFT

$$e^{j\Delta \frac{2\pi}{N}n} x[n] = W_N^{-\Delta n} x[n] \xleftrightarrow{N\text{-point DFT}} X[\left((k - \Delta)\right)_N] \quad k = 0, 1, \dots, N-1$$

Note that the length, N_x , of $x[n]$ has to satisfy $N \geq N_x$.

Ex: Let $x[n]$ be of length 7.

Find N and Δ so that

$$e^{j\frac{2\pi}{3}n}x[n] = W_N^{-\Delta n}x[n] \xleftrightarrow{N\text{-pointDFT}} X[((k - \Delta))_N] \quad k = 0, 1, \dots, N-1$$

For $N = 9$ and $\Delta = 3$

$$e^{j3\frac{2\pi}{9}n}x[n] = W_9^{-3n}x[n] \xleftrightarrow{9\text{-pointDFT}} X[((k - 3))_9] \quad k = 0, 1, \dots, N-1$$

where $X[k]$ is the 9-point DFT of $x[n]$.

Other solutions: $N = 12 \quad \Delta = 4,$
 $N = 15 \quad \Delta = 5, \dots$

In general $e^{j\frac{2\pi}{3}n} = e^{j\Delta\frac{2\pi}{N}n \left(\frac{N}{3\Delta}\right)}$

Therefore

$$N = \text{multiple of } 3, \quad N > 7, \quad \Delta = \frac{N}{3}$$

4) Duality

DFS

$$\tilde{x}[n] \xleftrightarrow{DFS} \tilde{X}[k] \Leftrightarrow \tilde{X}[n] \xleftrightarrow{DFS} N \tilde{x}[-k]$$

Proof:

$$\tilde{x}[n] = \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}[k] W_N^{-kn}$$



compare

$$\tilde{X}[k] = \sum_{n=0}^{N-1} \tilde{x}[n] W_N^{kn}$$

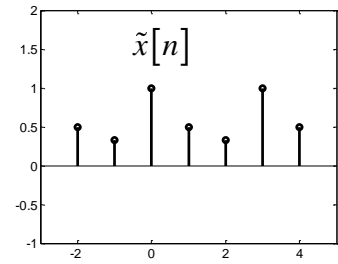
$$\underbrace{N \tilde{x}[-n]}_{\text{DFS coefficients of the sequence } \tilde{X}[k]} = \sum_{k=0}^{N-1} \tilde{X}[k] W_N^{kn}$$

$$\Rightarrow \tilde{X}[n] \xleftrightarrow{DFS} N \tilde{x}[-k]$$

DFT

$$x[n] \xleftrightarrow{DFT} X[k] \Leftrightarrow X[n] \xleftrightarrow{DFT} N x[((-k))_N]$$

Ex: Let $\tilde{x}[n] = \begin{bmatrix} 1 & \frac{1}{2} & \frac{1}{3} \end{bmatrix}$ periodic with 3



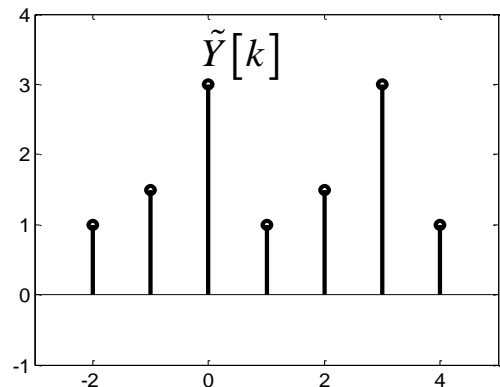
$$\begin{aligned} \tilde{X}[k] &= 1 + \frac{1}{2}e^{-j\frac{2\pi}{3}k} + \frac{1}{3}e^{-j\frac{2\pi}{3}2k} \\ &= \begin{bmatrix} \frac{11}{6} & \frac{7-j\sqrt{3}}{12} & \frac{7+j\sqrt{3}}{12} \end{bmatrix} \text{ periodic with 3} \end{aligned}$$

Then the DFS coefficients of the periodic sequence

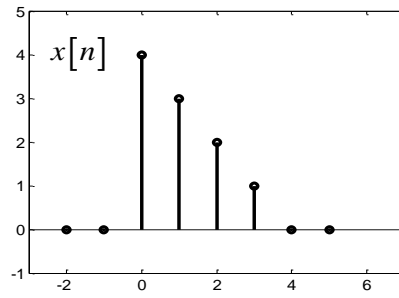
$$\tilde{y}[n] = \tilde{X}[n] = \begin{bmatrix} \frac{11}{6} & \frac{7-j\sqrt{3}}{12} & \frac{7+j\sqrt{3}}{12} \end{bmatrix} \text{ periodic with 3}$$

are

$$\begin{aligned} Y[k] &= 3\tilde{x}[-k] \\ &= 3 \begin{bmatrix} 1 & \frac{1}{3} & \frac{1}{2} \end{bmatrix} \end{aligned}$$



Ex: Let $x[n]$ be

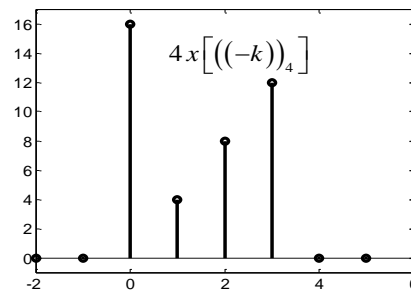


Using 4-point DFT,

$$4x\left[\left((-k)\right)_4\right]$$

is the DFT of

$$X[n] = 4 + 3e^{-j\frac{2\pi}{4}n} + 2e^{-j\frac{2\pi}{4}2n} + e^{-j\frac{2\pi}{4}3n}$$



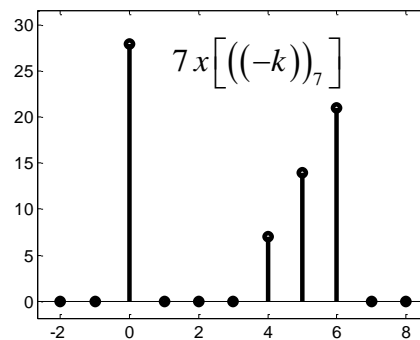
However, if, for example, 7-point DFT is used

Then

$$7x\left[\left((-k)\right)_7\right]$$

is the DFT of

$$X[n] = 4 + 3e^{-j\frac{2\pi}{7}n} + 2e^{-j\frac{2\pi}{7}2n} + e^{-j\frac{2\pi}{7}3n}$$



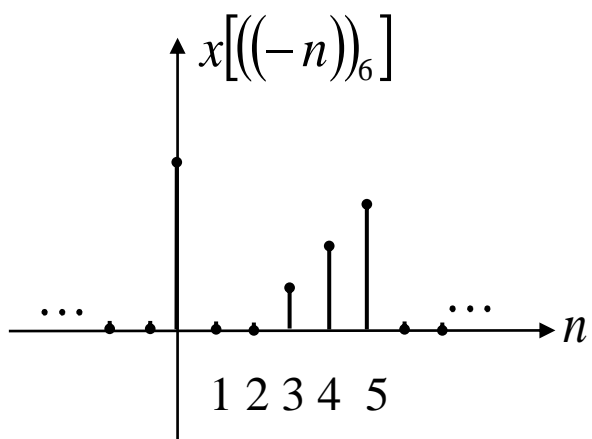
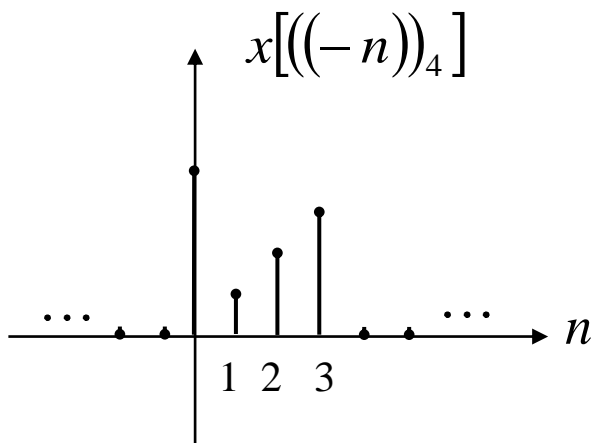
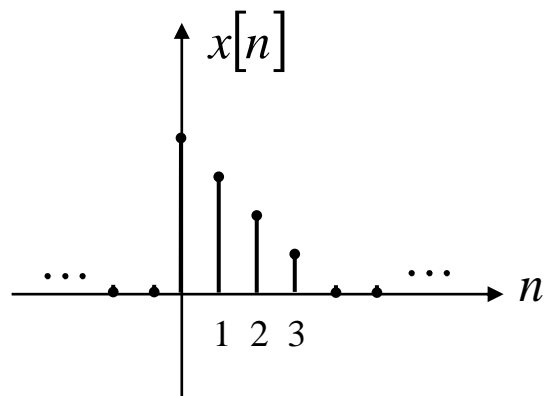
5) Symmetry Properties

Symmetry Properties of DFT are strictly related to those of DTFT

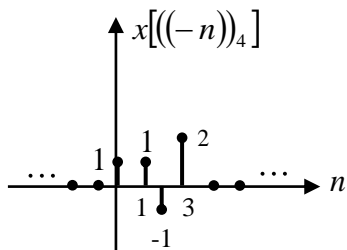
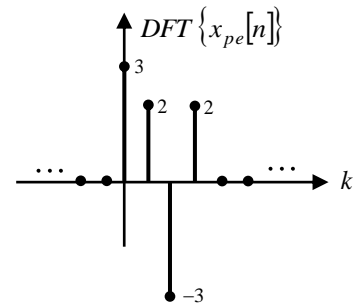
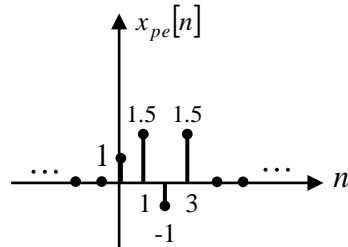
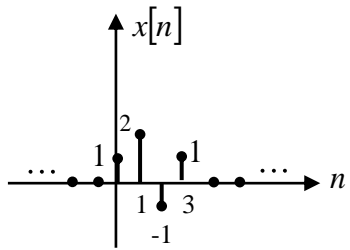
$$X[k] = \underbrace{X(e^{j\omega})}_{DTFT} \Big|_{\omega = \frac{2\pi}{N}k}$$

DFS	DFT
	$n = 0, 1, \dots, N-1$ $k = 0, 1, \dots, N-1$
$\tilde{x}[n] \leftrightarrow \tilde{X}[k]$	$x[n] \leftrightarrow X[k]$
$\tilde{x}^*[n] \leftrightarrow \tilde{X}^*[-k]$	$x^*[n] \leftrightarrow X^*[((-k))_N]$
$\tilde{x}[-n] \leftrightarrow \tilde{X}[-k]$	$x[((-n))_N] \leftrightarrow X[((-k))_N]$
$\text{Re}\{\tilde{x}[n]\} \leftrightarrow \tilde{X}_e[k] = \frac{1}{2}(\tilde{X}[k] + \tilde{X}^*[-k])$ conjugate symmetric part	$\text{Re}\{x[n]\} \leftrightarrow X_{pe}[k] = \frac{1}{2}(X[k] + X^*[((-k))_N])$ periodic conjugate symmetric part
$j \text{Im}\{\tilde{x}[n]\} \leftrightarrow \tilde{X}_o[k] = \frac{1}{2}(\tilde{X}[k] - \tilde{X}^*[-k])$ conjugate antisymmetric part	$j \text{Im}\{x[n]\} \leftrightarrow X_{po}[k] = \frac{1}{2}(X[k] - X^*[((-k))_N])$ periodic conjugate antisymmetric part
$\tilde{x}_e[n] = \frac{1}{2}(\tilde{x}[n] + \tilde{x}^*[-n]) \leftrightarrow \text{Re}\{\tilde{X}[k]\}$ conjugate symmetric part	$x_{pe}[n] = \frac{1}{2}(x[n] + x^*[((-n))_N]) \leftrightarrow \text{Re}\{X[k]\}$ periodic conjugate symmetric part
$\tilde{x}_o[n] = \frac{1}{2}(\tilde{x}[n] - \tilde{x}^*[-n]) \leftrightarrow j \text{Im}\{\tilde{X}[k]\}$ conjugate antisymmetric part	$x_{po}[n] = \frac{1}{2}(x[n] - x^*[((-n))_N]) \leftrightarrow j \text{Im}\{X[k]\}$ periodic conjugate antisymmetric part

Ex: What is $x[((-n))_N]$?



Ex: The DFT of a periodic even sequence is real valued.



Real Sequences

$$x[n] = x^*[n] \quad \Rightarrow \quad X(e^{j\omega}) = X^*(e^{-j\omega})$$

We know that DTFT is also conjugate symmetric wrt π .

Then, since DFT is obtained by uniformly sampling DTFT,

$X[k]$ is conjugate symmetric over $k = 1, 2, \dots, N-1$.

DFS	DFT
$\tilde{X}[k] = \tilde{X}^*[-k]$	or $X[k] = X^*[((-k))_N]$ $X[k] = X[N-k] \quad k = 1, 2, \dots, N-1$
$\text{Re}\{\tilde{X}[k]\} = \text{Re}\{\tilde{X}[-k]\}$	$\text{Re}\{X[k]\} = \text{Re}\{X[N-k]\}$ $k = 1, 2, \dots, N-1$
$\text{Im}\{\tilde{X}[k]\} = -\text{Im}\{\tilde{X}[-k]\}$	$\text{Im}\{X[k]\} = -\text{Im}\{X[N-k]\}$ $k = 1, 2, \dots, N-1$
$ \tilde{X}[k] = \tilde{X}[-k] $	$ X[k] = X[N-k] $ $k = 1, 2, \dots, N-1$
$\angle \tilde{X}[k] = -\angle \tilde{X}[-k]$	$\angle X[k] = -\angle X[N-k]$ $k = 1, 2, \dots, N-1$

Ex: $x[n] = \delta[n] + \delta[n-1] \quad \rightarrow \quad X(e^{j\omega}) = 1 + e^{j\omega} = e^{-j\frac{\omega}{2}} \cos\left(\frac{\omega}{2}\right)$

For 8-point DFT

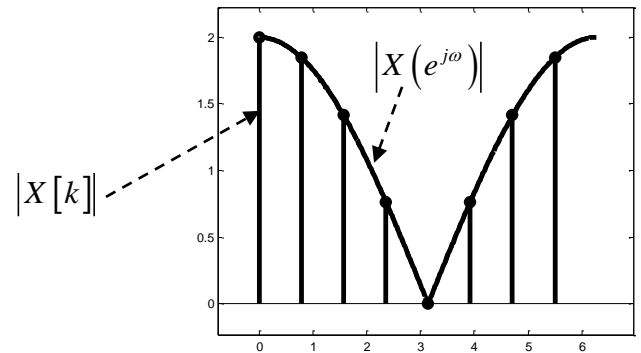
$$X[k] = 1 + e^{j\frac{2\pi}{8}k}$$

$$|X[k]| = |X[((-k))_8]|$$

$$|X[1]| = |X[7]|$$

$$|X[2]| = |X[6]|$$

$$|X[3]| = |X[5]|$$

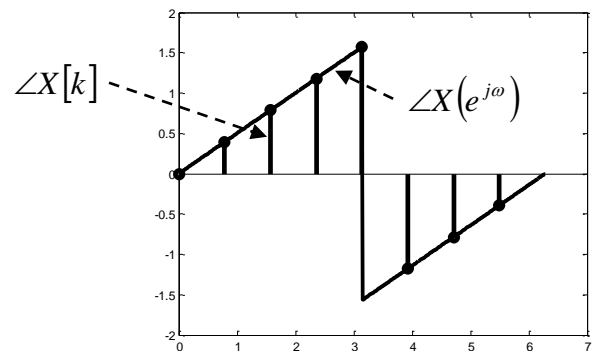


$$\angle X[k] = -\angle X[((-k))_8]$$

$$\angle X[1] = -\angle X[7]$$

$$\angle X[2] = -\angle X[6]$$

$$\angle X[3] = -\angle X[5]$$



6) Convolution Property

DFS: Periodic Convolution

DFT: Circular Convolution

$$x[n] * y[n] \leftrightarrow X(e^{j\omega})Y(e^{j\omega})$$

$$\tilde{x}_1[n] \leftrightarrow \tilde{X}_1[k] \quad \text{and} \quad \tilde{x}_2[n] \leftrightarrow \tilde{X}_2[k]$$

(same fund. period)

$$\tilde{X}_3[k] = \tilde{X}_1[k] \tilde{X}_2[k]$$

$$\tilde{x}_3[n] = ?$$

$$\begin{aligned} \tilde{x}_3[n] &= \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}_1[k] \tilde{X}_2[k] W_N^{-kn} \\ &= \frac{1}{N} \sum_{k=0}^{N-1} \tilde{X}_1[k] \left(\sum_{r=0}^{N-1} \tilde{x}_2[r] W_N^{kr} \right) W_N^{-kn} \\ &= \frac{1}{N} \sum_{r=0}^{N-1} \tilde{x}_2[r] \underbrace{\left(\sum_{k=0}^{N-1} \tilde{X}_1[k] W_N^{k(r-n)} \right)}_{N\tilde{x}_1[n-r]} \end{aligned}$$

herefore

$$\tilde{x}_3[n] = \sum_{r=0}^{N-1} \tilde{x}_1[n-r] \tilde{x}_2[r]$$

periodic convolution of $\tilde{x}_1[n]$ and $\tilde{x}_2[n]$.

Ex:

```
x=1:5;
```

```
y=[1 0 0 -2 0];
```

```
X=fft(x);
```

```
Y=fft(y);
```

```
Z=X.*Y;
```

```
z=ifft(Z)      (z = [-5 -6 -7 2 1])
```

```
stem(z)
```

DFT: Circular Convolution

$$x_1[n] \xleftrightarrow{N\text{-point DFT}} X_1[k] \quad \text{and} \quad x_2[n] \xleftrightarrow{N\text{-point DFT}} X_2[k],$$

$$X_3[k] = X_1[k] X_2[k]$$

$$x_3[n] = ?$$

$$x_3[n] = \begin{cases} \tilde{x}_3[n] & n = 0, 1, \dots, N-1 \\ 0 & \text{o.w.} \end{cases}$$

Using the result from DFS

$$x_3[n] = \begin{cases} \sum_{r=0}^{N-1} x_1[((n-r))_N] x_2[r] & n = 0, 1, \dots, N-1 \\ 0 & \text{o.w.} \end{cases}$$

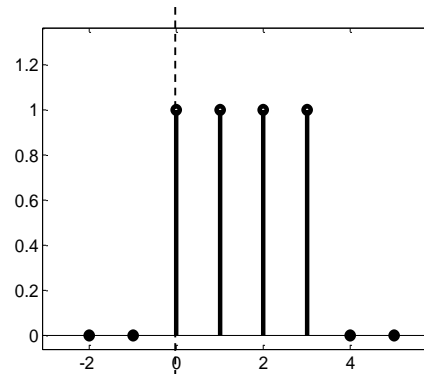
This is called “**N-point circular convolution**” of $x_1[n]$ and $x_2[n]$

$$x_3[n] = x_1[n] \circledast_N x_2[n]$$

Ex: Linear convolution

$$x[n]: [\dots 0 \ 0 \ 1 \ 1 \ 1 \ 1 \ 0 \ 0 \dots]$$

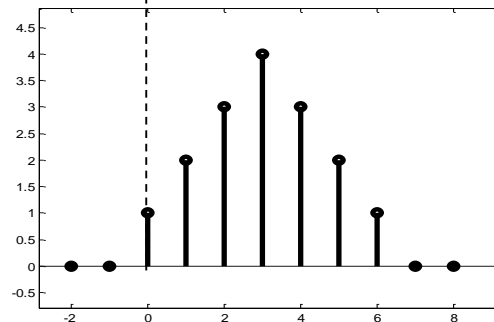
\uparrow
 $n = 0$



The linear convolution of $x[n]$ with itself is

$$y[n] = x[n] * x[n] \qquad y[n]: [\dots 0 \ 0 \ 1 \ 2 \ 3 \ 4 \ 3 \ 2 \ 1 \ 0 \ 0 \dots]$$

\uparrow
 $n = 0$



Note that the length of $y[n]$ is 7 (= 4+4-1)

Ex: Continued: 4-point circular convolution of $x[n]$ with itself

Let

$$W[k] = X_4[k]X_4[k]$$

and $w[n]$ be the IDFT of $W[k]$, then

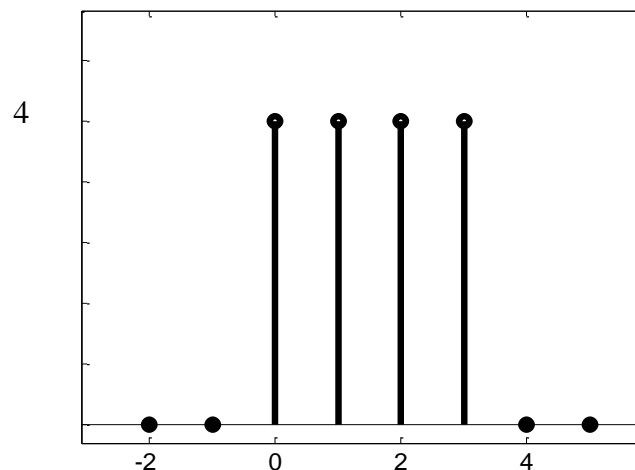
$$\begin{aligned} w[n] &= x[n] \circledast_4 x[n] \\ &= \sum_{r=0}^3 x[(n-r)_4] x[r] \end{aligned}$$

To compute one needs

$$x[(-r)_4], x[(1-r)_4], x[(2-r)_4], x[(3-r)_4]$$

Then

$$w[n] = \sum_{r=0}^3 1 = 4, \quad n = 0, 1, 2, 3$$



This is not equal to the result of linear convolution!

Ex: Continued

6-point circular convolution of $x[n]$ with itself

Let

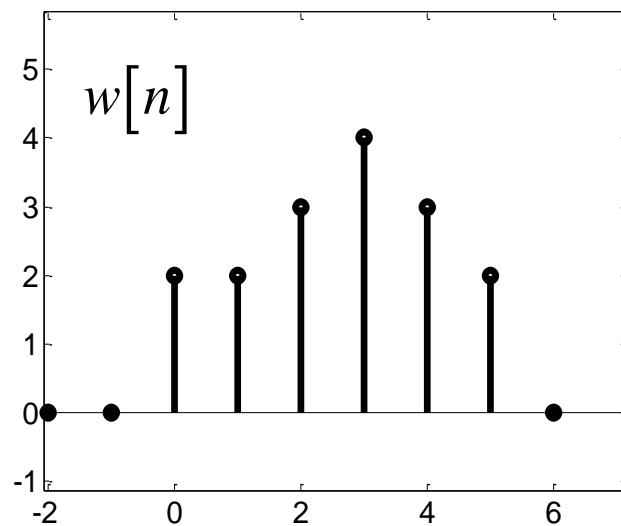
$$W[k] = X_6[k]X_6[k]$$

and $w[n]$ be the IDFT of $W[k]$, then

$$w[n] = x[n] \circledast_6 x[n]$$

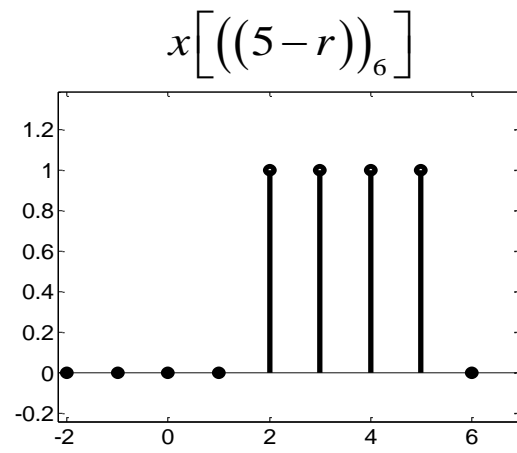
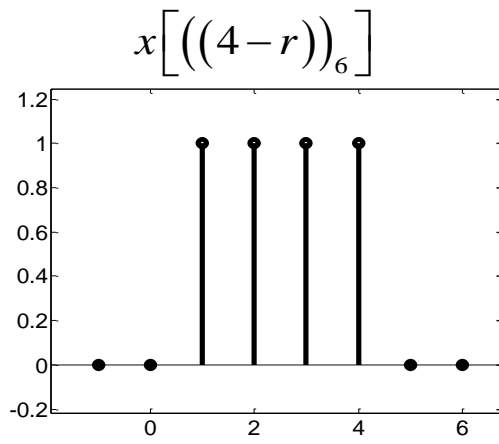
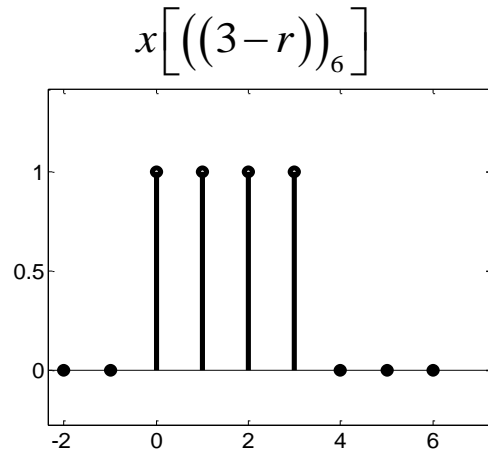
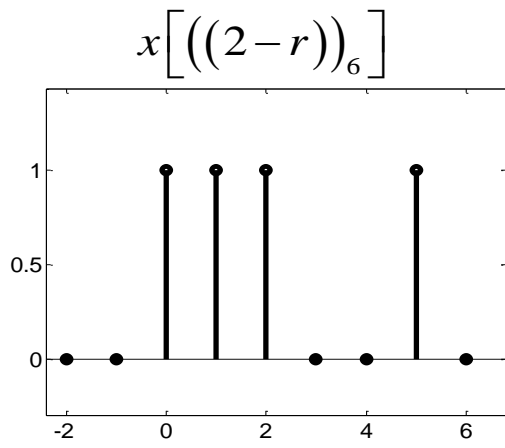
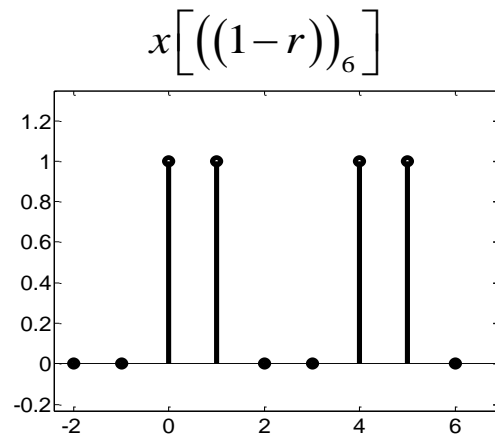
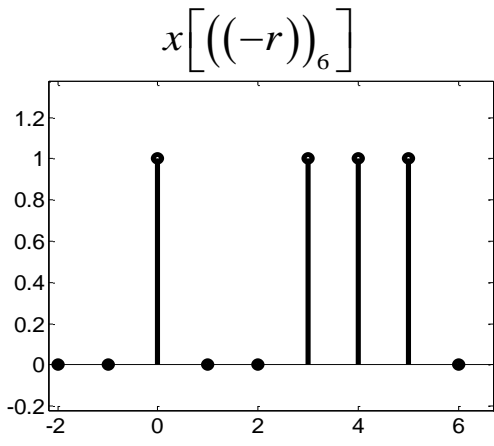
$$= \sum_{r=0}^5 x[(n-r)_6]x[r]$$

$$w[n] = \left[\dots 0 \quad 0 \quad \underbrace{2}_{n=0} \quad 2 \quad 3 \quad 4 \quad 3 \quad 2 \quad 0 \quad 0 \dots \right]$$



This is also not the same as the result of linear convolution!

However, partially correct!



Ex: Continued

7-point circular convolution of $x[n]$ with itself

$$W[k] = X_7[k]X_7[k]$$

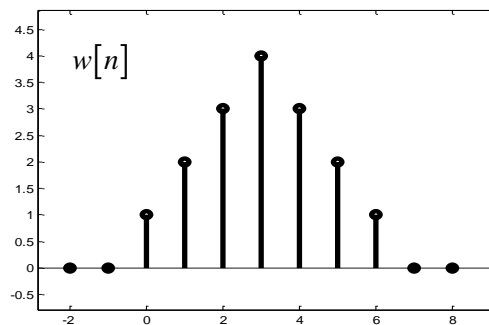
$$w[n] = x[n] \circledast_7 x[n]$$

$$= \sum_{r=0}^6 x[(n-r)_7] x[r]$$

$$w[n] = \left[\dots 0 \quad 0 \quad \underbrace{2}_{n=0} \quad 2 \quad 3 \quad 4 \quad 3 \quad 2 \quad 0 \quad 0 \dots \right]$$

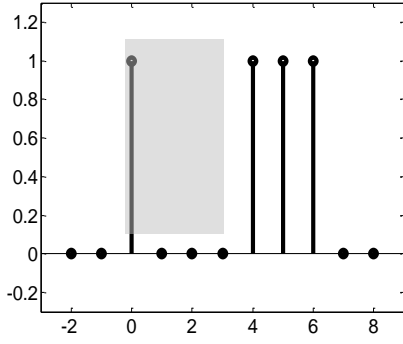
$w[n]$ is the IDFT of $X_7[k]X_7[k]$, $X_7[k]$: 7-point DFT

$$w[n] = \left[\dots 0 \quad 0 \quad \underbrace{1}_{n=0} \quad 2 \quad 3 \quad 4 \quad 3 \quad 2 \quad 1 \quad 0 \quad 0 \dots \right]$$

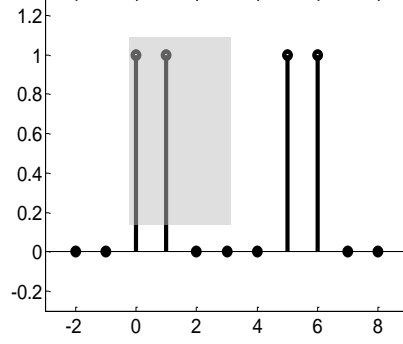


In this case the result is the same as that of linear convolution.

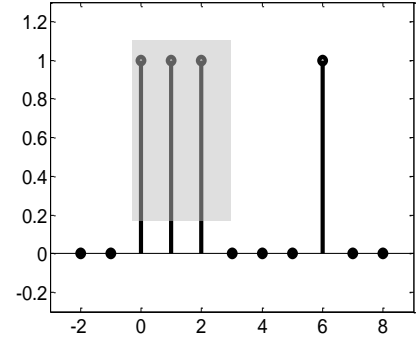
$$x\left[\left((-r)\right)_7\right]$$



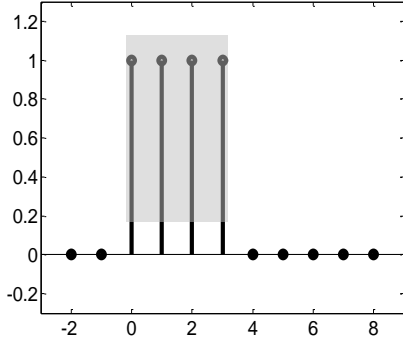
$$x\left[\left((1-r)\right)_7\right]$$



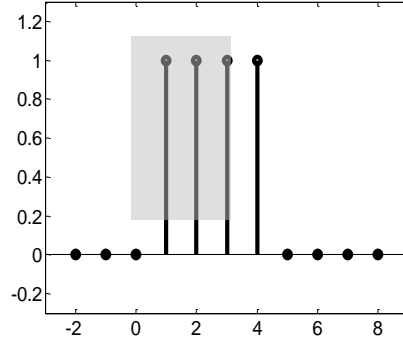
$$x\left[\left((2-r)\right)_7\right]$$



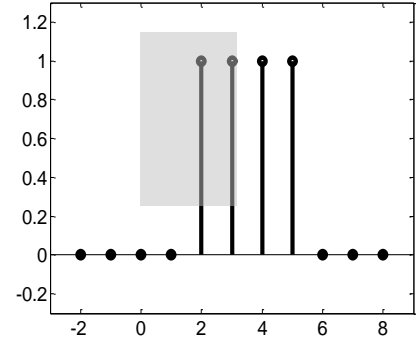
$$x\left[\left((3-r)\right)_7\right]$$



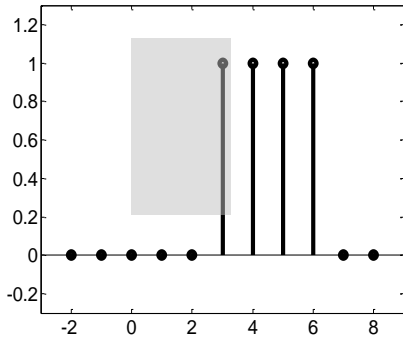
$$x\left[\left((4-r)\right)_7\right]$$



$$x\left[\left((5-r)\right)_7\right]$$



$$x\left[\left((6-r)\right)_7\right]$$



Nonzero values of $x[r]$ are over the shaded region.