

Discrete-Time Signals

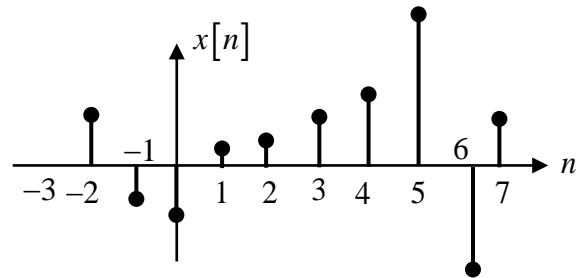
- Mathematical Representation of DT Signals
- Unit Sample Sequence
- Unit Step Sequence
- Exponential Sequences
- Sinusoidal Sequences
- Two Fundamental Properties of DT Sinusoidal Sequences

Discrete-Time Signals

A discrete-time signal is a sequence of numbers.

Its n^{th} element is $x[n]$, $n \in \mathbb{Z}$.

$x[n]$ may be real or complex.



Ex: Let $T = 0.001$ sec = 1 msec.

We do NOT write $\dots, x[-0.001], x[0], x[0.001], x[0.002] \dots!$

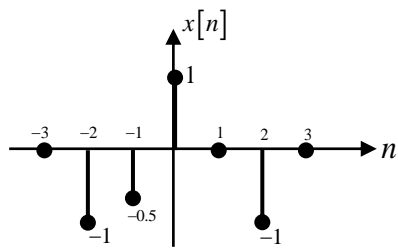
We write $\dots, x[-1], x[0], x[1], x[2] \dots$

$x[n]$ may have been obtained by sampling a continuous-time signal, i.e.,

$$x[n] = x_c(t)|_{t=nT}, \quad n \in \mathbb{Z}$$

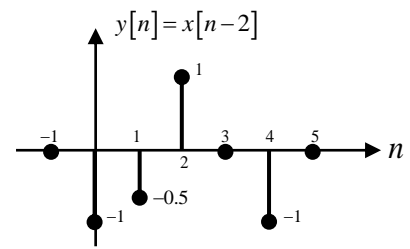
Time Shift of a Signal

Ex: Delay



$$y[n] = x[n - n_0]$$

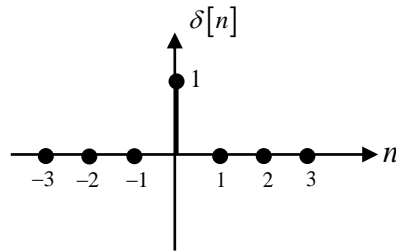
n_0 is always an integer!



We do NOT write sth. like $x[n - 2.15]$

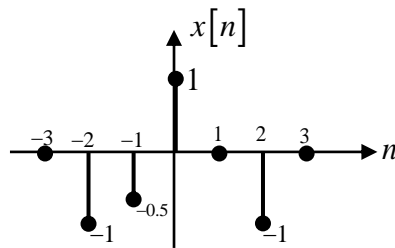
(continuous shift/interpolation...)

UNIT SAMPLE SEQUENCE



$$\delta[n] = \begin{cases} 1 & n = 0 \\ 0 & n \neq 0 \end{cases}$$

Ex: Let $x[n]$ be



Can be written as:

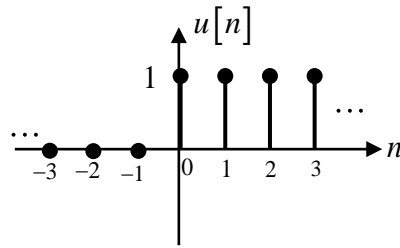
$$x[n] = -\delta[n+2] - 0.5\delta[n+1] + \delta[n] - \delta[n-2]$$

In general, any seq. can be written as

$$x[n] = \sum_{k=-\infty}^{\infty} x[k] \delta[n-k]$$

This is the fundamental expression in the derivation of the fact that the output of a LTI system is the convolution of the input and the system's impulse response.

UNIT STEP SEQUENCE



$$u[n] = \begin{cases} 1 & n \geq 0 \\ 0 & n < 0 \end{cases}$$

Ex: $u[n] = \sum_{k=0}^{\infty} \delta[n-k]$ (convolution of $u[n]$ and $\delta[n]$)

or

Ex: $u[n] = \sum_{k=-\infty}^n \delta[k]$ (like integration in cont. time)

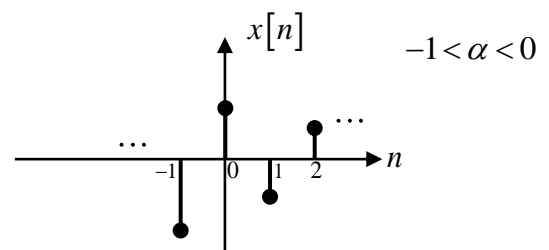
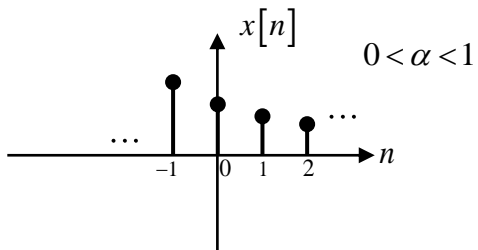
on the other hand

Ex: $\delta[n] = u[n] - u[n-1]$ (like differentiation in cont. time)

EXPONENTIAL SEQUENCES (real valued)

They appear in the solution and analysis of LTI systems.

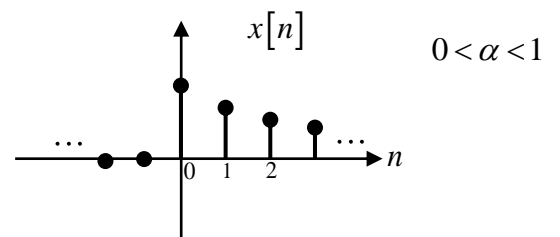
$$x[n] = A\alpha^n$$



if $|\alpha| > 1$ then $|x[n]|$ grows as $n \rightarrow \infty$

TRUNCATED EXPONENTIAL SEQUENCE:

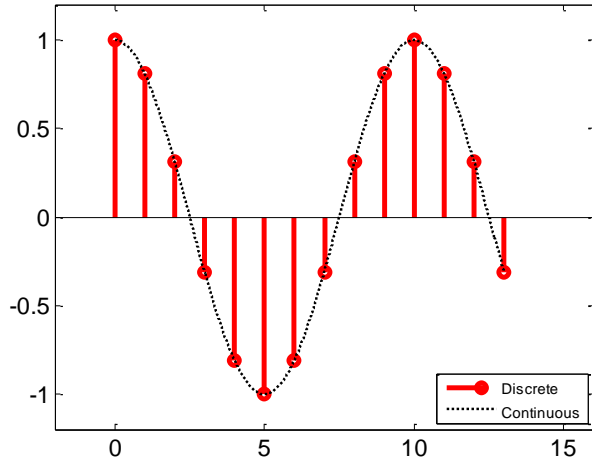
$$x[n] = \begin{cases} A\alpha^n & n \geq 0 \\ 0 & n < 0 \end{cases} = A\alpha^n u[n]$$



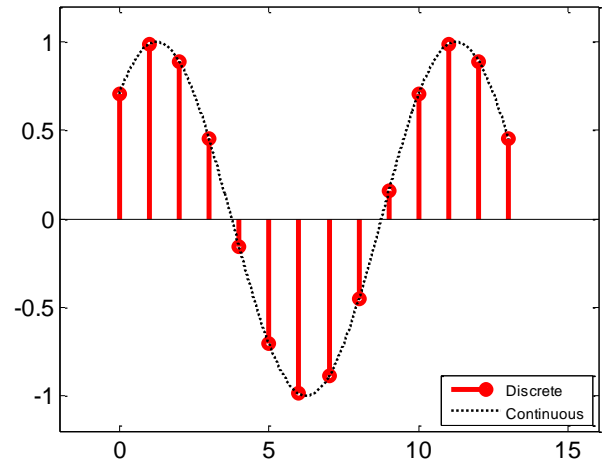
SINUSOIDAL SEQUENCES

$$x[n] = A \cos(\omega_0 n + \phi)$$

$$\omega_0 = \frac{2\pi}{10} \quad \phi = 0$$



$$\omega_0 = \frac{2\pi}{10} \quad \phi = \frac{\pi}{4}$$



Note that, $x_1[n]$ and $x_2[n]$ cannot be related by a simple time shift.

~~$$x_2[n] = x_1\left[n - \frac{5}{4}\right]$$~~

~~$$x_2[n] = x_1\left[n - \frac{5}{4}\right]$$~~

code_1.m

```
clear all
close all

N = 10;
w0 = 2*pi / N ;
alfa = 1;
phase = 0;  %pi/4;

M = N + 3;
n = [0:0.01:M];
y = alfa.^n.*cos(w0*n-phase);

nn = [0:M];
x = alfa.^nn.*cos(w0*nn-phase);

stem(nn,x,'r','LineWidth',3)
hold
plot(n,y,'k:','LineWidth',2)
v = axis;
dV = v(4)-v(3);
v = [v(1)-2 v(2)+2 v(3)-0.1*dV v(4)+0.1*dV];
axis(v)

set(gca,'fontsize',14)
hleg = legend('Discrete','Continuous','location','southeast');
set(hleg,'fontsize', 9)
```

EXPONENTIAL SEQUENCES (complex valued)

$$x[n] = A \alpha^n \quad A, \alpha \in \mathbb{C}$$

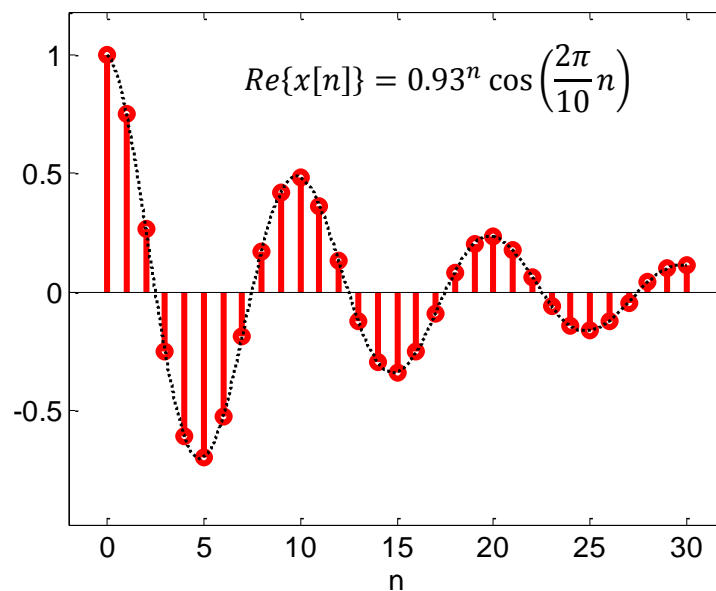
$$x[n] = 0.93^n \cos\left(\frac{2\pi}{10}n\right)$$

$$A = |A|e^{j\phi} \quad \alpha = |\alpha|e^{j\omega_0}$$

$$\Rightarrow x[n] = |A||\alpha|^n \cos(\omega_0 n + \phi) + j|A||\alpha|^n \sin(\omega_0 n + \phi)$$

For example,

$$A = 1 \quad \alpha = 0.93 \quad \phi = 0 \quad \omega_0 = \frac{2\pi}{10}$$



COMPLEX EXPONENTIAL SEQUENCES:

Let

$$|\alpha| = 1$$

in

$$x[n] = A \alpha^n \quad A, \alpha \in \mathbb{C}$$

Then,

$$|A|e^{j(\omega_0 n + \phi)}$$

is called a complex exponential sequence.

$$\Rightarrow Ae^{j\omega_0 n} = |A|\cos(\omega_0 n + \phi) + j|A|\sin(\omega_0 n + \phi)$$

A sinusoidal sequence can be expressed in terms of a complex exponential sequence.

$$M \cos(\omega_0 n + \phi) = \operatorname{Re}\{Ae^{j\omega_0 n}\} = \frac{1}{2}(Ae^{j\omega_0 n} + A^*e^{-j\omega_0 n}); \quad A = Me^{j\phi}, \quad M \in \mathbb{R}$$

$$M \sin(\omega_0 n + \phi) = \operatorname{Im}\{Ae^{j\omega_0 n}\} = \frac{1}{2j}(Ae^{j\omega_0 n} - A^*e^{-j\omega_0 n}); \quad A = Me^{j\phi}, \quad M \in \mathbb{R}$$

ω_0 : frequency (radians/sample or, shortly, radians, NOT rad/sec!)

ϕ : phase shift (radians)

TWO FUNDAMENTAL PROPERTIES OF COMPLEX EXPONENTIAL (SINUSOIDAL) DISCRETE-TIME SEQUENCES

FIRST: For any frequency value ω_0 , $\omega_0 + k2\pi$ (k : integer) is an equivalent frequency value, i.e.,

if $x[n] = Ae^{j\omega_0 n}$ and $y[n] = Ae^{j(\omega_0 + k2\pi)n}$

then $x[n] = y[n] \quad \forall n \in \mathbb{Z}$

$$\cos(\omega_0 n) = \cos(\omega_0 n + k2\pi n)$$

$$\sin(\omega_0 n) = \sin(\omega_0 n + k2\pi n)$$

In other words, the elements of the set $\{\omega | \omega = \omega_0 + k2\pi, \omega_0 \in \mathbb{R}, k \in \mathbb{Z}\}$ are equivalent if they are considered as the frequencies of discrete-time complex exponentials/sinusoids.

$$\dots = \cos\left(-\frac{9\pi}{5}n\right) = \cos\left(\frac{\pi}{5}n\right) = \cos\left(\frac{11\pi}{5}n\right) = \cos\left(\frac{21\pi}{5}n\right) = \dots$$

$$\dots = e^{-j\frac{9\pi}{5}n} = e^{j\frac{\pi}{5}n} = e^{j\frac{11\pi}{5}n} = e^{j\frac{21\pi}{5}n} = \dots$$

Therefore an interval of 2π covers all distinct frequencies.

(effectively an interval of π ! Why?)

Convention for the Reference Frequency Interval

Since frequencies are equivalent when multiples of 2π is added/subtracted, convention is to use either of the following as the basic interval

$$-\pi \leq \omega < \pi$$

$$0 \leq \omega < 2\pi$$



Ex:

$$\text{i) } \cos(416.31\pi n) = \cos(208.155(2\pi n)) = \cos(0.155(2\pi n)) = \cos(0.31\pi n)$$

$$\text{ii) } \sin(416.31\pi n) = \sin(208.155(2\pi n)) = \sin(0.155(2\pi n)) = \sin(0.31\pi n)$$

Ex:

$$\text{i) } \cos(417.31\pi n) = \cos(208.655(2\pi n)) = \cos(0.655(2\pi n)) = \cos(1.31\pi n) = \cos(0.69\pi n)$$

$$\text{ii) } \sin(417.31\pi n) = \sin(208.655(2\pi n)) = \sin(0.655(2\pi n)) = \sin(1.31\pi n) = -\sin(0.69\pi n) \quad (\text{minus sign !!!})$$

Practically, it is sufficient consider

$$\cos(f\pi n), \sin(f\pi n) \text{ for } 0 \leq f \leq 1$$

since

$$\cos(f\pi n) = \cos((2 - f)\pi n)$$

and

$$\sin(f\pi n) = -\sin((2 - f)\pi n)$$

Ex:

A 100 MHz signal $x_c(t) = \cos(2 \times 10^8 \pi t)$ is sampled at a rate of 250 MHz

(i.e. sampling period is $T_s = \frac{1}{250\,000\,000} = 4$ pico sec.)

$$x[n] = \cos(0.8\pi n)$$

Find another continuous-time (CT) sinusoid that would yield the same discrete-time sinusoid (i.e., $x[n]$) at this sampling frequency.

How many other CT sinusoids would yield the same DT sequence?

Answer:

$$\cos(\omega_0 n) = \cos((\omega_0 + k2\pi) n)$$

$$= \cos\left(\frac{(\omega_0 + k2\pi)}{T_s} T_s n\right)$$

$$\rightarrow \cos((2\pi f_0 f_s + k2\pi f_s) t)$$

where

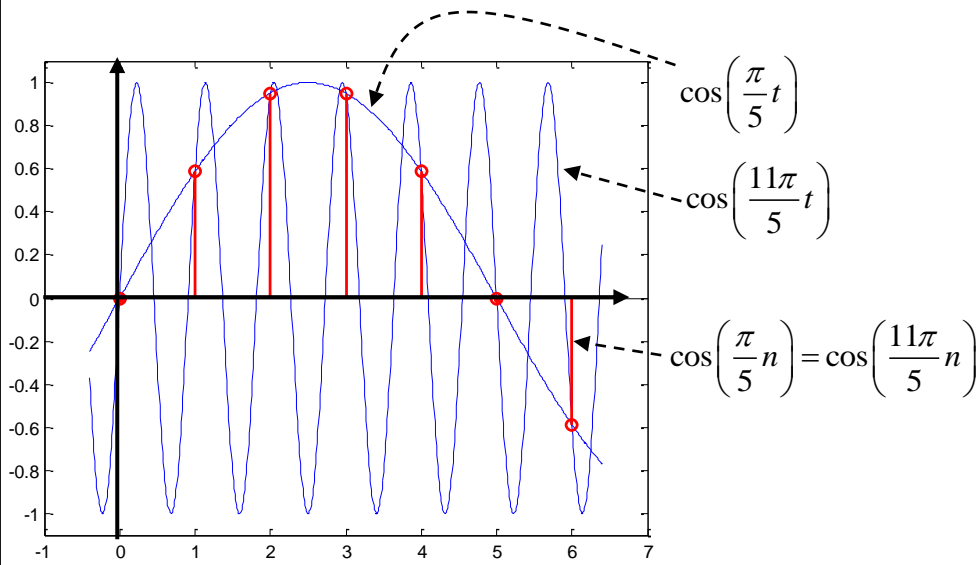
$$f_0 = 0.4 \qquad f_s = 250 \text{ MHz}$$

Therefore all CT sinusoids at frequencies 100 MHz, 350 MHz, 600 MHz, 850 MHz, ... yield the same DT sinusoid for this sampling frequency.

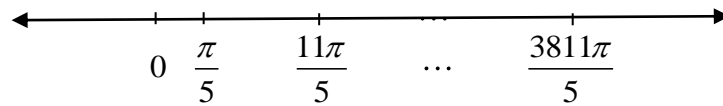
Their frequencies can be expressed as

$$\frac{(k + 0.4)}{T_s} = (k + 0.4) 250 \text{ MHz}$$

Ex:



```
clear all
close all
n = [-0.4:0.01:6.4];
nn = [0:6];
x = sin(n*pi/5);
y = sin(n*11*pi/5);
z = sin(nn*11*pi/5);
plot(n,x);
hold
plot(n,y)
stem(nn,z,'r','LineWidth',2)
v=axis;
v=[v(1) v(2) -1.1 1.1];
axis(v)
```



SECOND:

A DT sinusoidal ($\cos(\omega_0 n + \phi)$) or complex exponential signal $e^{j(\omega_0 n + \phi)}$ is not necessarily periodic!

To be periodic,

ω_0 must be a *rational* multiple of π ,

i.e.,

$$\omega_0 = \frac{p}{q}\pi, \quad p, q \in \mathbb{Z}$$

Proof:

$$A \cos(\omega_0 n + \phi) \stackrel{?}{=} A \cos(\omega_0 (n + N) + \phi)$$

$$A \cos(\omega_0 (n + N) + \phi) = A \cos(\omega_0 n + \omega_0 N + \phi)$$

$$\text{For periodicity } \omega_0 N = k 2\pi \Rightarrow \omega_0 = \frac{k}{N} 2\pi \quad \text{or} \quad \frac{\omega_0}{2\pi} = \frac{k}{N} \quad k \in \mathbb{Z}$$

has to be satisfied.

Ex: $\cos(5n)$ $\omega_0 = 5$ $\frac{\omega_0}{2\pi} = \frac{5}{2\pi}$ is not rational so it is not periodic.

FUNDAMENTAL PERIOD, N , IS NOT NECESSARILY EQUAL TO $\frac{2\pi}{\omega_0}$

Since, for periodic sinusoids,

$$\omega_0 N = k2\pi$$

i.e.,

$$N = \frac{k2\pi}{\omega_0},$$

fundamental period, N , is not necessarily equal to $\frac{2\pi}{\omega_0}$.

Finding the Fundamental Period of a Sinusoid

Find the smallest k , k_{min} , so that $k_{min} \frac{2\pi}{\omega_0}$ is an integer.

Then, the fundamental period is

$$N = k_{min} \frac{2\pi}{\omega_0}.$$

$$\underline{\text{Ex:}} \cos\left(\frac{\pi}{5}n\right) \quad \omega_0 = \frac{\pi}{5} \quad \frac{\omega_0}{2\pi} = \frac{1}{10} \quad N = k \frac{2\pi}{\omega_0} = k \frac{2\pi}{\frac{\pi}{5}} = 10 \quad (k=1)$$

$$\underline{\text{Ex:}} \cos\left(\frac{5\pi}{17}n\right) \quad \omega_0 = \frac{5\pi}{17} \quad \frac{\omega_0}{2\pi} = \frac{5}{34} \quad N = k \frac{34}{5} = 34 \quad (k=5)$$

$$\underline{\text{Ex:}} \cos\left(\frac{6\pi}{5}n\right) \quad \omega_0 = \frac{6\pi}{5} \quad \frac{\omega_0}{2\pi} = \frac{3}{5} \quad N = k \frac{5}{3} = 5 \quad (k=3)$$

Ex: Let $x_1[n] = \cos(\omega_1 n)$ and $x_2[n] = \cos(\omega_2 n)$. Find two “frequencies” ω_1 and ω_2 such that $\omega_1 \neq \omega_2 + k2\pi$ for any integer k , and $x_1[n]$ and $x_2[n]$ are both periodic with fundamental period $N=13$.

$$N = 13 = k \frac{2\pi}{\omega}, k: \text{integer}$$

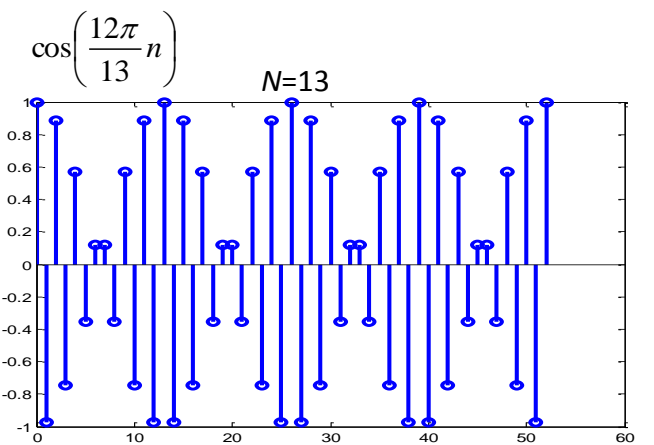
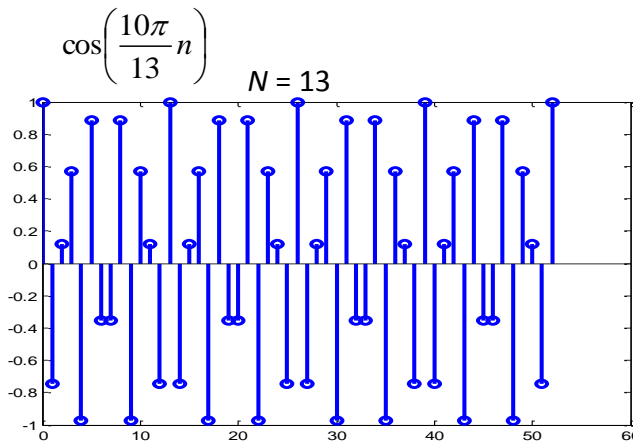
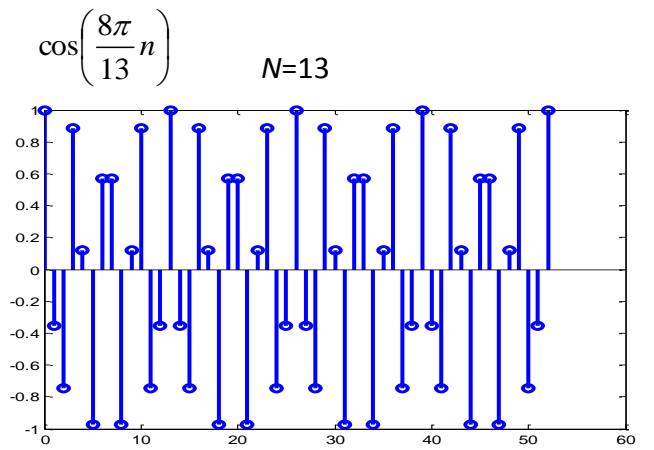
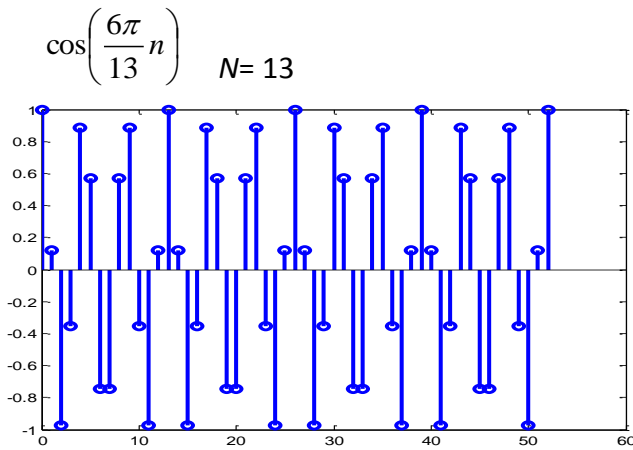
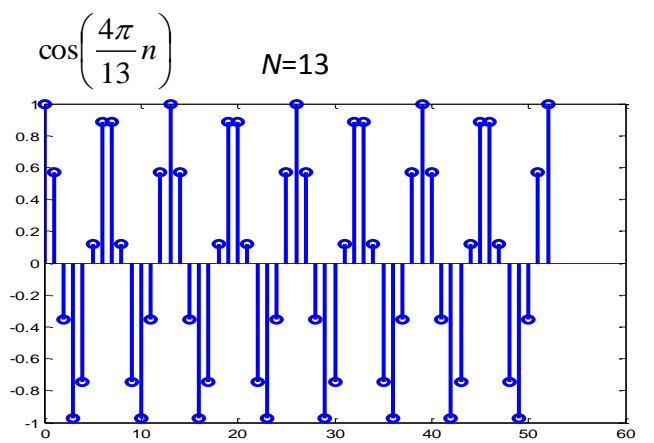
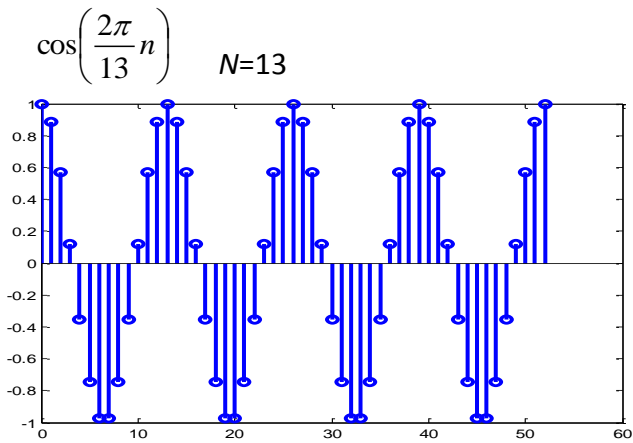
$$\Rightarrow \omega = k \frac{2\pi}{13}$$

Choose, for example, $k=1$ and $k=2 \quad \Rightarrow \omega_1 = \frac{2\pi}{13}, \omega_2 = \frac{4\pi}{13}$

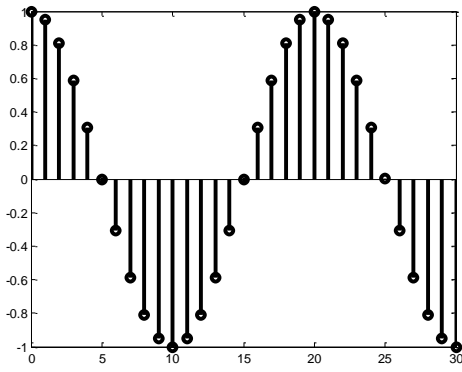
**Therefore,
DT sinusoids may have different “frequencies”
although
their fundamental periods are the same!**

What do the discrete-time sinusoids look like?

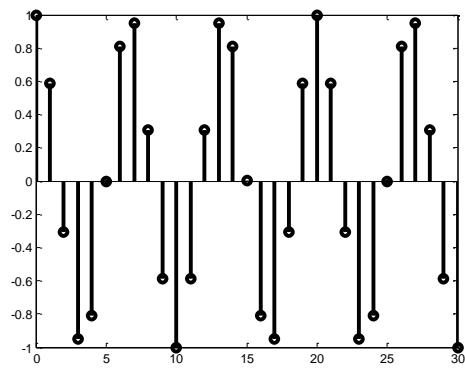
Some frequencies between 0 and π



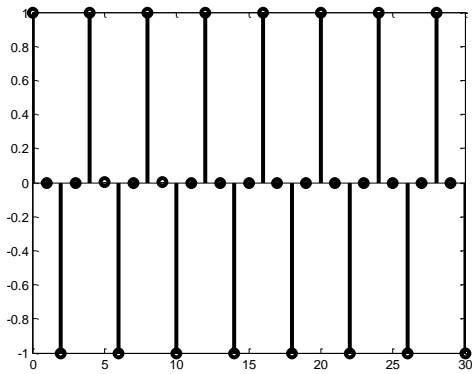
$\cos(0.1\pi n)$ $N=20$



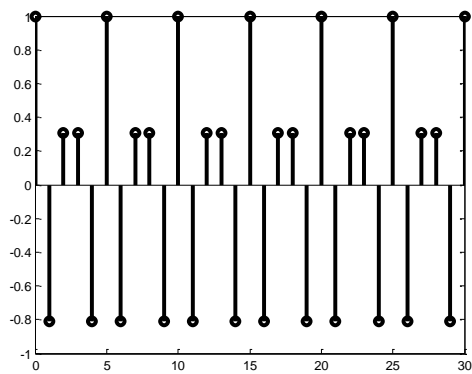
$\cos(0.3\pi n)$ $N=20$



$\cos(0.5\pi n)$ $N=4$

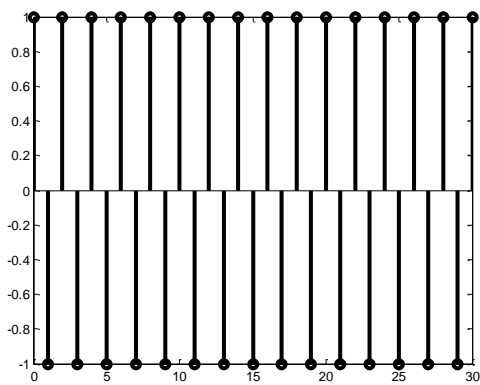


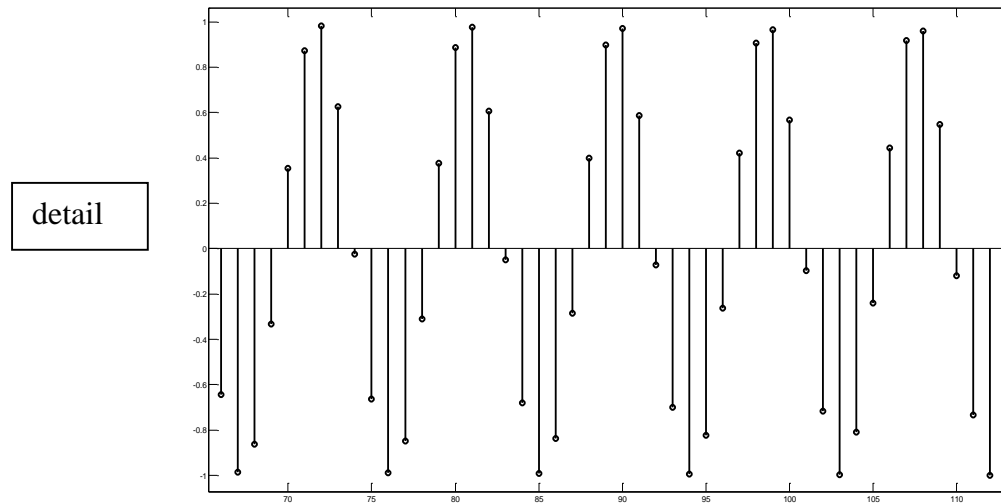
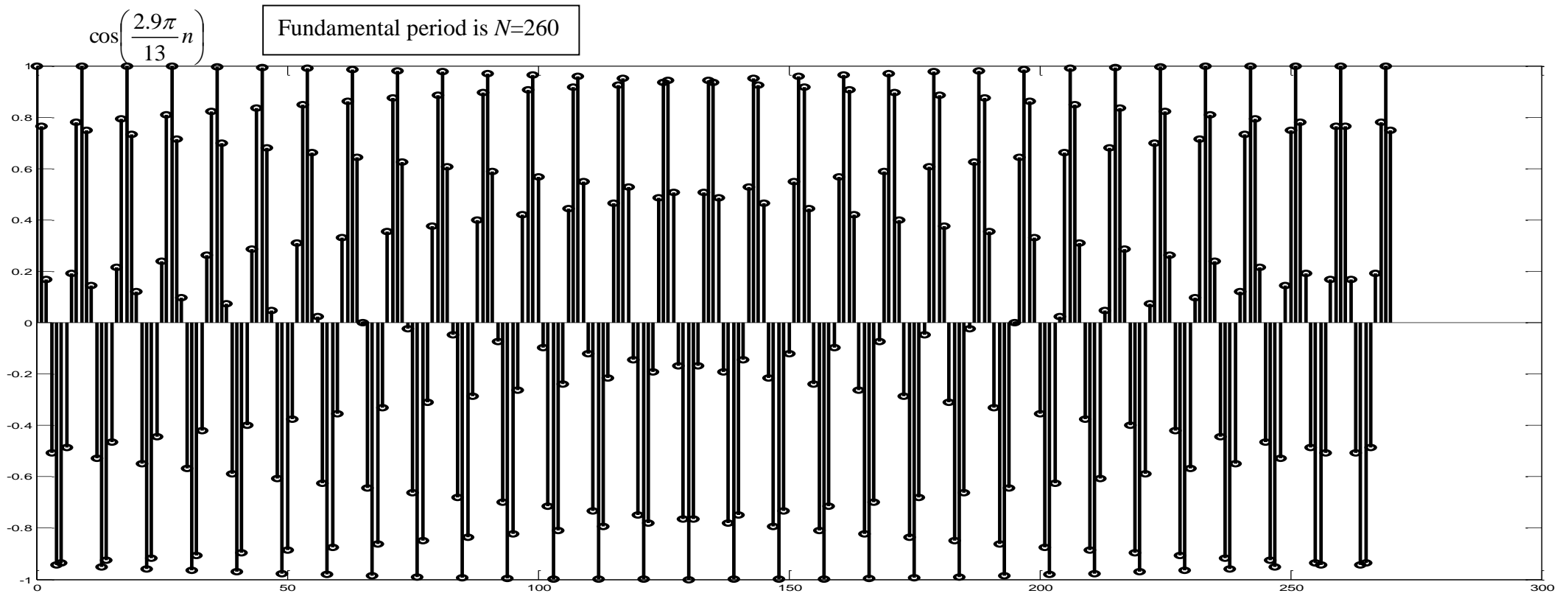
$\cos(0.8\pi n)$ $N=5$



THE HIGHEST FREQUENCY SIGNAL IN DISCRETE-TIME

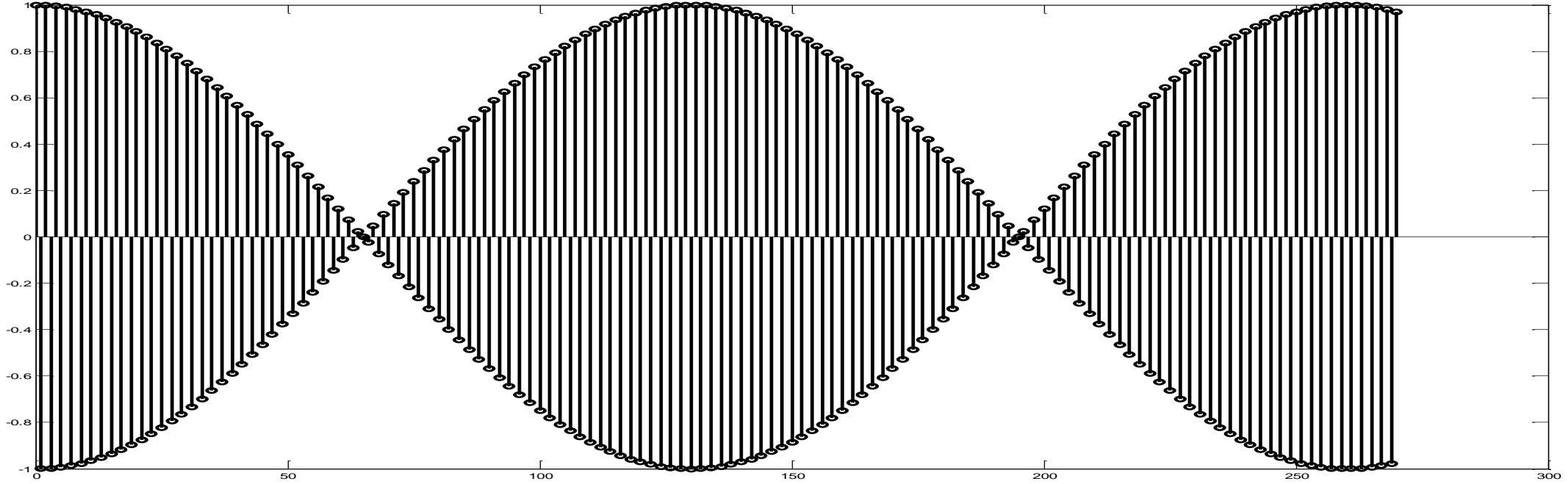
$\cos(\pi n)$ $N=2$





$$\cos\left(\frac{12.9\pi}{13}n\right)$$

Fundamental period is $N=260$



detail

