

# FREQUENCY DOMAIN REPRESENTATION OF LTI SYSTEMS

## DICRETE TIME FOURIER TRANSFORM (DTFT)

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**DTFT OF WINDOWED SINUSOID** (ESTIMATING THE FREQUENCY, EFFECT OF NOISE)

## DICRETE TIME FOURIER TRANSFORM (DTFT)

The Fourier transform of a sequence  $x[n]$  is defined as

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

If the FT exists (summation converges) the sequence can be obtained from its FT as

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

What is this?

Fourier Transform is periodic with  $2\pi$ .

## LTI SYSTEMS

The frequency response function of a LTI system,

$$H(e^{j\omega}),$$

is the FT of its impulse response

$$h[n].$$

## EXISTENCE

FT of a sequence  $x[n]$  exists, i.e.,

$$\sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

converges to a continuous function of  $\omega$ , if  $x[n]$  is absolutely summable.

(sufficient condition)

Proof: Exercise

$$\begin{aligned}\sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n} &= \sum_{n=-\infty}^{\infty} x[n](\cos(\omega n) - j \sin(\omega n)) \\ &= \sum_{n=-\infty}^{\infty} x[n] \cos(\omega n) - j \sum_{n=-\infty}^{\infty} x[n] \sin(\omega n)\end{aligned}$$

Both sums have to converge

→ All stable LTI systems have frequency response functions.



**Ex**: DTFT of

$$x[n] = a^n u[n]$$

$$X(e^{j\omega}) = \sum_{n=0}^{\infty} a^n e^{-j\omega n}$$

$$= \sum_{n=0}^{\infty} (ae^{-j\omega})^n$$

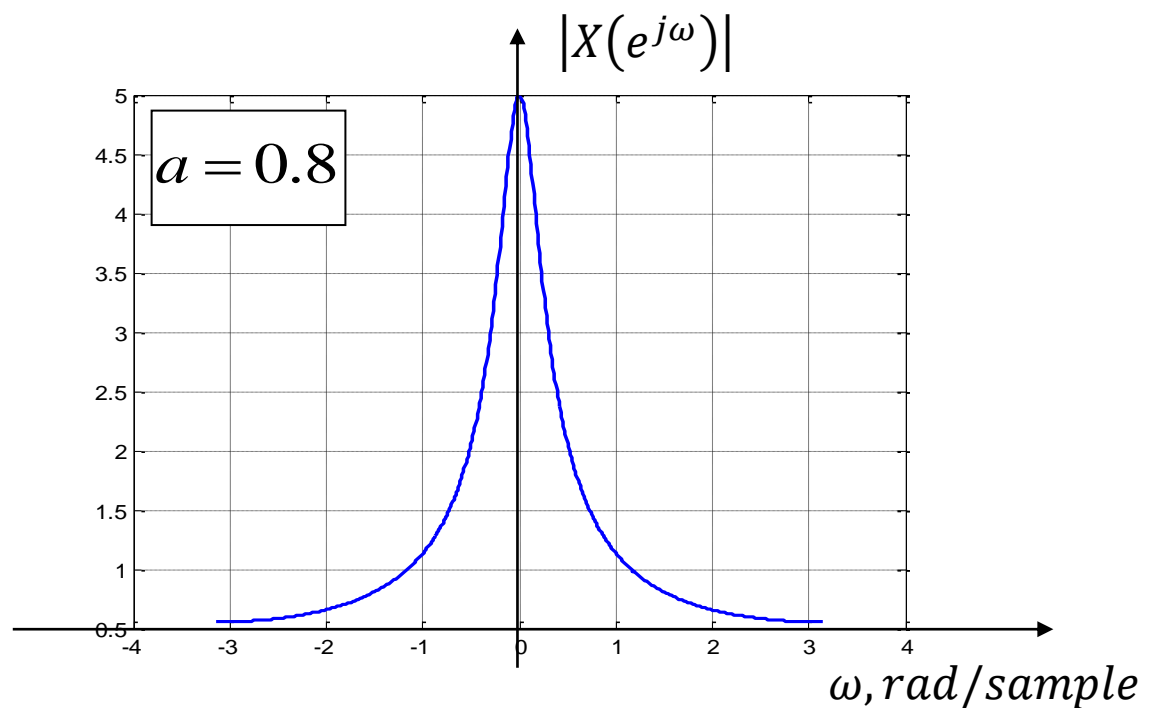
$$= \frac{1}{1 - ae^{-j\omega}} \quad \text{if } |ae^{-j\omega}| < 1 \quad \text{or} \quad |a| < 1$$

Magnitude of

$$X(e^{j\omega}) = \frac{1}{1 - ae^{-j\omega}}, \quad |a| < 1$$

$$\begin{aligned} |X(e^{j\omega})|^2 &= \frac{1}{|1 - ae^{-j\omega}|^2} \\ &= \frac{1}{|1 - a \cos(\omega) + ja \sin(\omega)|^2} \\ &= \frac{1}{1 + a^2 - 2a \cos(\omega)} \end{aligned}$$

$$|X(e^{j\omega})| = \frac{1}{\sqrt{1 + a^2 - 2a \cos(\omega)}}$$



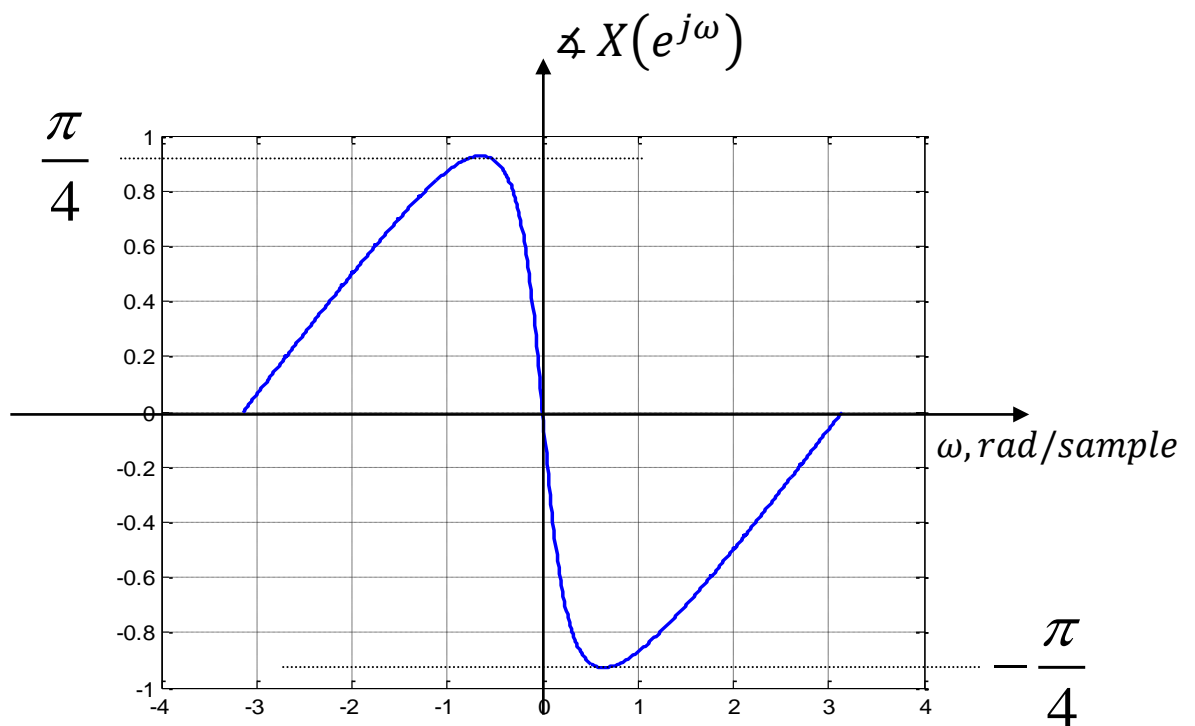
Phase of

$$X(e^{j\omega}) = \frac{1}{1 - ae^{-j\omega}}, \quad |a| < 1$$

$$\angle X(e^{j\omega}) = \angle 1 - \angle 1 - ae^{-j\omega}$$

$$= 0 - \angle(1 - a \cos(\omega) + ja \sin(\omega))$$

$$= -\tan^{-1}\left(\frac{a \sin(\omega)}{1 - a \cos(\omega)}\right)$$



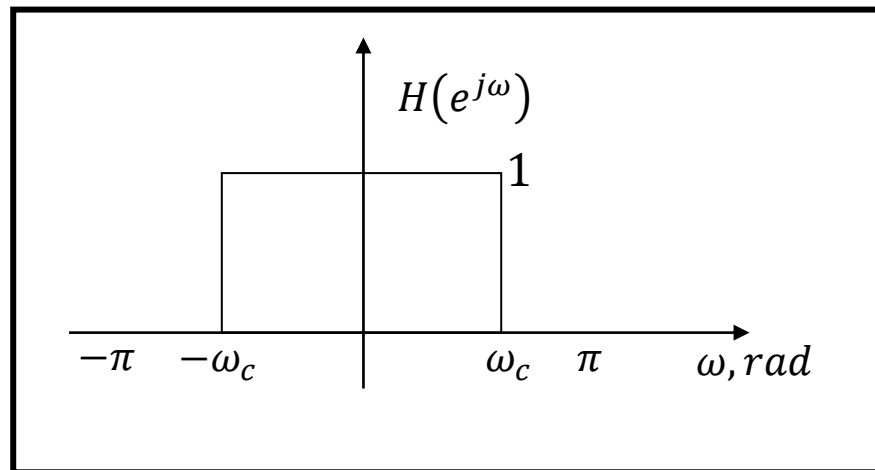
## MEAN SQUARE CONVERGENCE

Some sequences, which are not absolutely summable but square summable

$$\sum_{n=-\infty}^{\infty} |x[n]|^2 < \text{finite constant}$$

can still be represented by Fourier Transform, but...

**Ex:** Ideal lowpass filter.



$$H(e^{j\omega}) = \begin{cases} 1 & |\omega| < \omega_c \\ 0 & \omega_c < |\omega| \leq \pi \end{cases}$$

Let's find  $h[n]$

$$\begin{aligned}h[n] &= \frac{1}{2\pi} \int_{-\omega_c}^{\omega_c} e^{j\omega n} d\omega \\&= \frac{1}{j2\pi n} (e^{j\omega_c n} - e^{-j\omega_c n}) \\&= \frac{\sin(\omega_c n)}{\pi n}\end{aligned}$$

Note that,

$$h[n] = \frac{\sin(\omega_c n)}{\pi n}$$

is not absolutely summable!

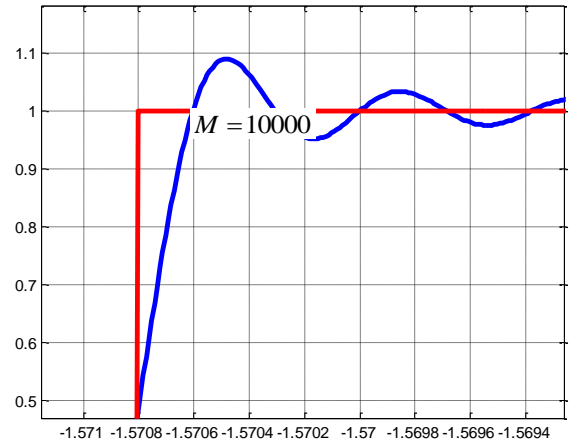
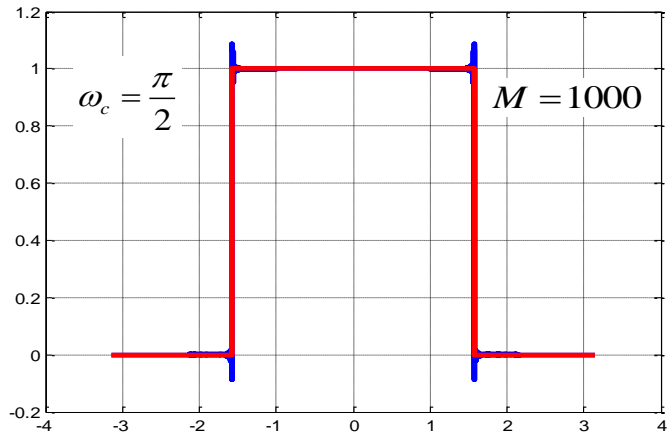
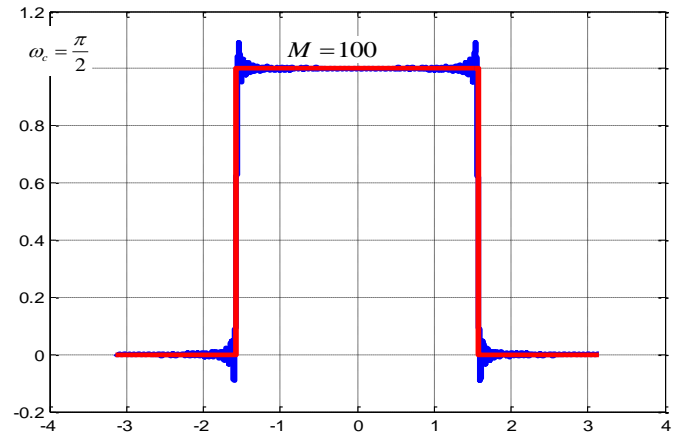
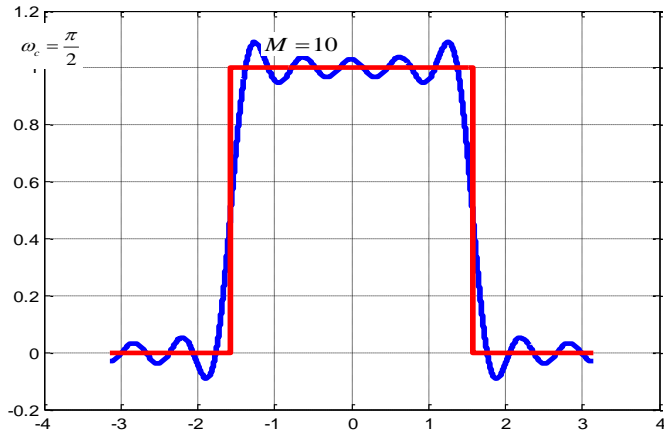
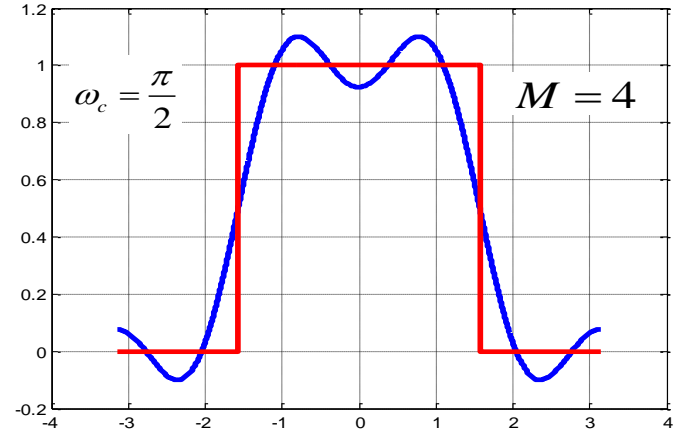
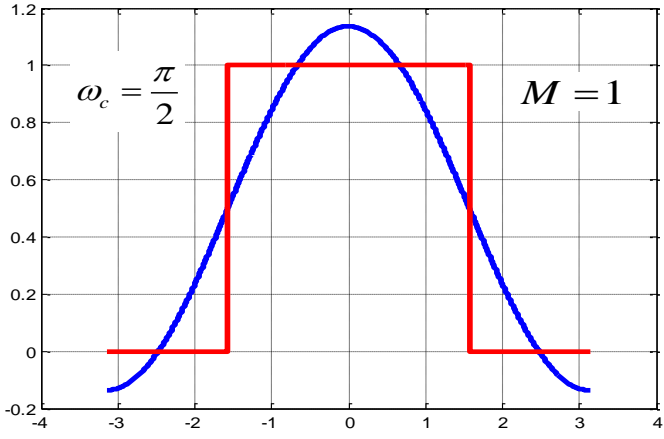
Then, one may question the Fourier transform of  $h[n]$ ,

$$\sum_{n=-\infty}^{\infty} \frac{\sin(\omega_c n)}{\pi n} e^{-j\omega n} = ?$$

Define

$$H_M = \sum_{n=-M}^M \frac{\sin(\omega_c n)}{\pi n} e^{-j\omega n}$$





detail

Even if you take  $M \rightarrow \infty$  oscillations do not die to zero.

However

$$\lim_{M \rightarrow \infty} \int_{-\pi}^{\pi} |H(e^{j\omega}) - H_M(e^{j\omega})|^2 d\omega = 0.$$

This is called “mean square” convergence.

The oscillatory behavior around  $\omega = \omega_c$  is called the Gibbs phenomenon.

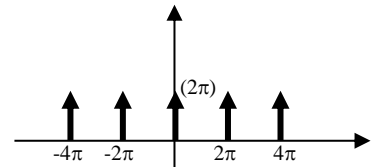
## MATLAB code

```
clear all; close all;
precision = 0.0001
w = [-pi:precision:pi];
ideal = zeros(1,length(w));
wc = pi/2;
orta = round(length(w)/2);
ideal((orta-round(wc/precision)):(orta+round(wc/precision)))=1;
M = 10000;
H = 0;
for n = -M:-1
    H = H+(sin(wc*n)/(pi*n))*exp(-i*w*n);
end
for n = 1:M
    H = H+(sin(wc*n)/(pi*n))*exp(-i*w*n);
end
H = H+(wc/pi);
plot(w,H); hold on;
plot(w,ideal,'r')
grid
```

## FOURIER TRANSFORM OF A CONSTANT SEQUENCE

$$x[n] = 1 \quad \leftrightarrow \quad X(e^{j\omega}) = 2\pi \sum_{r=-\infty}^{\infty} \delta(\omega + 2\pi r)$$

not absolutely  
summable



or we can write as

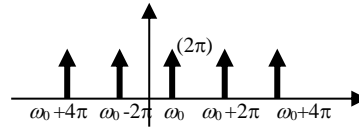
$$X(e^{j\omega}) = 2\pi\delta(\omega) \quad 0 \leq \omega < 2\pi$$

keeping in mind that FT is periodic with  $2\pi$ .

## FOURIER TRANSFORM OF A COMPLEX EXPONENTIAL SEQUENCE

$$x[n] = e^{j\omega_0 n} \quad \leftrightarrow \quad X(e^{j\omega}) = 2\pi \sum_{r=-\infty}^{\infty} \delta(\omega - \omega_0 + 2\pi r)$$

not absolutely  
summable



or we can write as

$$X(e^{j\omega}) = 2\pi\delta(\omega - \omega_0) \quad 0 \leq \omega \leq 2\pi$$

keeping in mind that FT is periodic with  $2\pi$ .

## FOURIER TRANSFORM OF A SINUSOIDAL SEQUENCE

$$\begin{aligned}x[n] &= \cos(\omega_0 n) \\&= \frac{1}{2}(e^{j\omega_0 n} + e^{-j\omega_0 n})\end{aligned}$$

Note that it is not absolutely summable.

$$X(e^{j\omega}) = \pi \left( \sum_{r=-\infty}^{\infty} \delta(\omega + \omega_0 + 2r\pi) + \delta(\omega - \omega_0 + 2r\pi) \right)$$

## FOURIER TRANSFORM OF UNIT STEP SEQUENCE

$$x[n] = u[n] \quad \leftrightarrow \quad X(e^{j\omega}) = \frac{1}{1 - e^{-j\omega}} + \pi \sum_{r=-\infty}^{\infty} \delta(\omega + 2\pi r)$$

not absolutely summable

# SYMMETRY PROPERTIES OF FOURIER TRANSFORM

Definitions:

Conjugate symmetric (CS) sequence.

$$x[n] = x^*[-n]$$

Conjugate antisymmetric (CaS) sequence.

$$x[n] = -x^*[-n]$$



Using the above definitions, any sequence can be written as

$$x[n] = x_e[n] + x_o[n]$$

where

$$x_e[n] = \frac{1}{2}(x[n] + x^*[-n]) \text{ is the CS part}$$

and

$$x_o[n] = \frac{1}{2}(x[n] - x^*[-n]) \text{ is the CaS part.}$$

# SYMMETRY PROPERTIES

## Fundamental relations

Let  $x[n] \leftrightarrow X(e^{j\omega})$  be a FT pair. Then, the following hold:

$$x^*[n] \leftrightarrow X^*(e^{-j\omega}) \quad \text{since} \quad X^*(e^{-j\omega}) = \sum_{n=-\infty}^{\infty} x^*[n]e^{-j\omega n}$$

and

$$x[-n] \leftrightarrow X(e^{-j\omega}) \quad \text{since} \quad X(e^{-j\omega}) = \sum_{n=-\infty}^{\infty} x[-n]e^{-j\omega n}$$

Above yields

$$x^*[-n] \leftrightarrow X^*(e^{j\omega})$$

The two relations above also yield:

1)

$$\operatorname{Re}\{x[n]\} = \frac{x[n] + x^*[n]}{2} \quad \leftrightarrow \quad \underbrace{X_e(e^{j\omega}) = \frac{X(e^{j\omega}) + X^*(e^{-j\omega})}{2}}_{\text{CS part of } X(e^{j\omega})}$$

2)

$$j\operatorname{Im}\{x[n]\} = \frac{x[n] - x^*[n]}{2} \quad \leftrightarrow \quad \underbrace{X_o(e^{j\omega}) = \frac{X(e^{j\omega}) - X^*(e^{-j\omega})}{2}}_{\text{CaS part of } X(e^{j\omega})}$$

3)

$$x_e[n] = \frac{x[n] + x^*[-n]}{2} \quad \leftrightarrow \quad \text{Re}\{X(e^{j\omega})\} = \frac{X(e^{j\omega}) + X^*(e^{j\omega})}{2}$$

Therefore FT of an even sequence is real!

4)

$$x_o[n] = \frac{x[n] - x^*[-n]}{2} \quad \leftrightarrow \quad j\text{Im}\{X(e^{j\omega})\} = \frac{X(e^{j\omega}) - X^*(e^{j\omega})}{2}$$

Therefore FT of an odd sequence is purely imaginary!

**Ex:** Let  $a[n]$  and  $b[n]$  be two real sequences with their DTFTs  $A(e^{j\omega})$  and  $B(e^{j\omega})$ , respectively.

Let

$$x[n] = a[n] + jb[n]$$

Then,

$$X(e^{j\omega}) = A(e^{j\omega}) + jB(e^{j\omega})$$

Note that  $A(e^{j\omega})$  is NOT the real part of  $X(e^{j\omega})$ .

However,

$$A(e^{j\omega}) = \frac{X(e^{j\omega}) + X^*(e^{-j\omega})}{2}$$

i.e. conjugate symmetric part of  $X(e^{j\omega})$ ,

since

$$X^*(e^{-j\omega}) = \underbrace{A^*(e^{-j\omega})}_{A(e^{j\omega})} - j \underbrace{B^*(e^{-j\omega})}_{B(e^{j\omega})} .$$

Similarly,

$$jB(e^{j\omega}) = \frac{X(e^{j\omega}) - X^*(e^{-j\omega})}{2}$$

i.e. conjugate antisymmetric part of  $X(e^{j\omega})$ .

**Ex:** (cont'd)

$$a[n]: [-1 \ 1]$$

$$b[n]: [1 \ 1]$$

$$x[n]: [-1 + j \ 1 + j]$$

$$A(e^{j\omega}) = -1 + e^{-j\omega}$$

$$B(e^{j\omega}) = 1 + e^{-j\omega}$$

$$X(e^{j\omega}) = -1 + e^{-j\omega} + j + je^{-j\omega}$$

$$X^*(e^{-j\omega}) = -1 + e^{-j\omega} - j - je^{-j\omega}$$

$$\frac{X(e^{j\omega}) + X^*(e^{-j\omega})}{2} = -1 + e^{-j\omega}$$

$$\frac{X(e^{j\omega}) - X^*(e^{-j\omega})}{2} = j + je^{-j\omega}$$

## REAL SEQUENCES

Based on the above relations, for real sequences ( $x[n] = x^*[n]$ ):

$$X(e^{j\omega}) = X^*(e^{-j\omega}) \quad (\text{conjugate symmetry})$$

which implies

Magnitude is even.....  $|X(e^{j\omega})| = |X(e^{-j\omega})|$

Phase is odd.....  $\angle X(e^{j\omega}) = -\angle X(e^{-j\omega})$

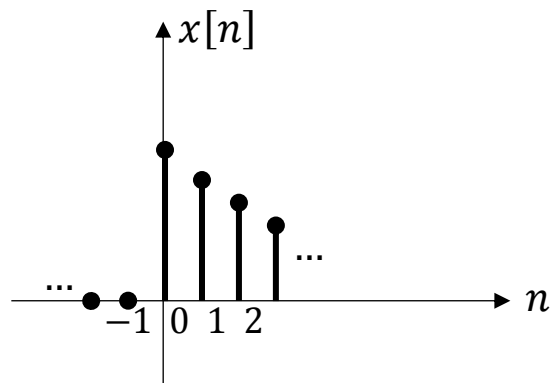
Real part is even.....  $\text{Re}\{X(e^{j\omega})\} = \text{Re}\{X(e^{-j\omega})\}$

Imaginary part is odd.....  $\text{Im}\{X(e^{j\omega})\} = -\text{Im}\{X(e^{-j\omega})\}$

## Verification by an example

$$x[n] = a^n u[n] \quad \leftrightarrow \quad X(e^{j\omega}) = \frac{1}{1 - ae^{-j\omega}} \quad |a| < 1$$

$$0 < a < 1$$





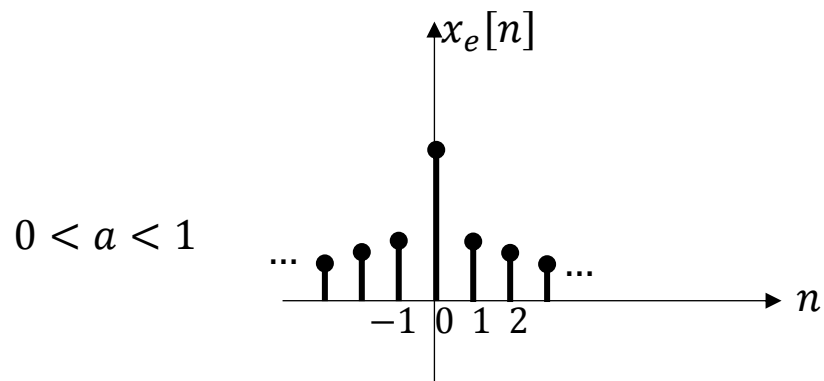
a) FT is conjugate symmetric:

$$X(e^{j\omega}) = X^*(e^{-j\omega}) = \frac{1}{1 - ae^{-j\omega}}$$

b) Real part of FT is an even function:

$$\operatorname{Re}\{X(e^{j\omega})\} = \operatorname{Re}\{X(e^{-j\omega})\} = \frac{1 - a \cos(\omega)}{1 + a^2 - 2a \cos(\omega)}$$

c)  $\text{Re}\{X(e^{j\omega})\}$  is the FT of  $x_e[n]$ :



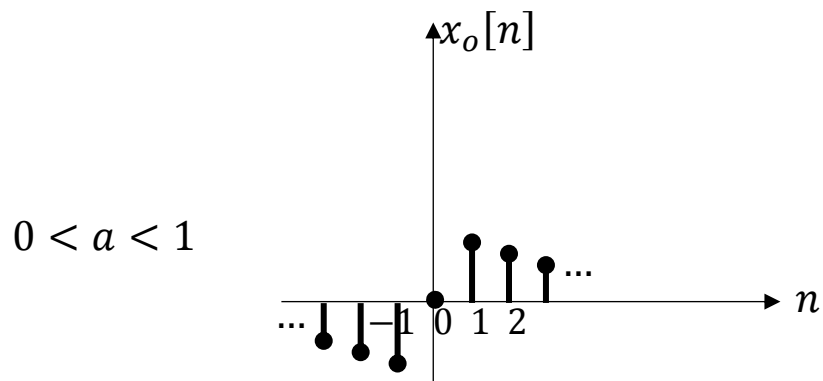
$$\text{Re}\{X(e^{j\omega})\} = \frac{1 - a \cos(\omega)}{1 + a^2 - 2a \cos(\omega)}$$

$$x_e[n] = \frac{1}{2}(a^n u[n] + a^{-n} u[-n])$$

d) Imaginary part of FT is an odd function.

$$\text{Im}\{X(e^{j\omega})\} = -\text{Im}\{X(e^{-j\omega})\} = \frac{-a \sin(\omega)}{1 + a^2 - 2a \cos(\omega)}$$

e)  $\text{Im}\{X(e^{j\omega})\}$  is the FT of  $x_o[n]$ :

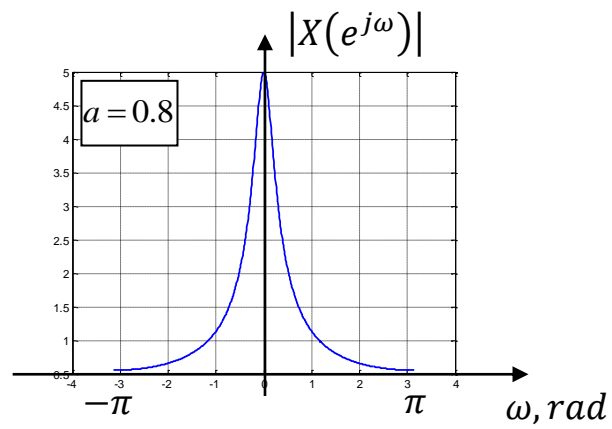


$$\text{Im}\{X(e^{j\omega})\} = \frac{-a \sin(\omega)}{1 + a^2 - 2a \cos(\omega)}$$

$$x_o[n] = \frac{1}{2} (a^n u[n] - a^{-n} u[-n])$$

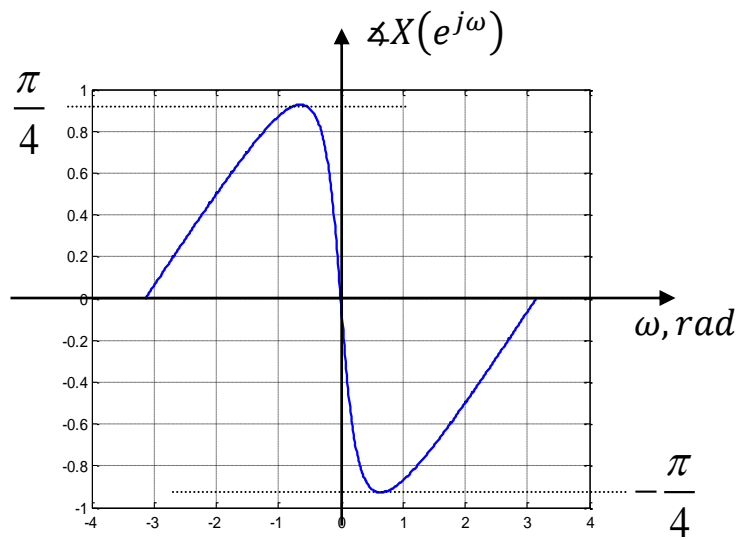
f) Magnitude of FT is an even function:

$$|X(e^{j\omega})| = |X(e^{-j\omega})| = \frac{1}{\sqrt{1 + a^2 - 2a \cos(\omega)}}$$



g) Phase of FT is an odd function

$$\angle X(e^{j\omega}) = -\angle X(e^{j\omega}) = -\tan^{-1}\left(\frac{a \sin(\omega)}{1 - a \cos(\omega)}\right)$$



## FOURIER TRANSFORM THEOREMS

$$x[n] \leftrightarrow X(e^{j\omega})$$

1)  $ax[n] + by[n] \leftrightarrow aX(e^{j\omega}) + bY(e^{j\omega})$  linearity

2)  $x[n - n_0] \leftrightarrow e^{-j\omega n_0} X(e^{j\omega})$  time-shift

3)  $e^{j\omega_0 n} x[n] \leftrightarrow X(e^{j(\omega - \omega_0)})$  freq.-shift

4)  $x[-n] \leftrightarrow X(e^{-j\omega})$  time reversal

5)  $nx[n] \leftrightarrow j \frac{dX(e^{j\omega})}{d\omega}$  differentiation in freq. domain

6)  $x[n] * y[n] \leftrightarrow X(e^{j\omega})Y(e^{j\omega})$  convolution (proof below)

7)

$$x[n]y[n] \leftrightarrow \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\theta})Y(e^{j(\omega - \theta)})d\theta$$

modulation, windowing

8) Parseval's theorem (prove as an exercise)

$$\sum_{n=-\infty}^{\infty} |x[n]|^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega$$

Note that  $|X(e^{j\omega})|^2$  is called the “**energy density spectrum**”.

9)

$$\sum_{n=-\infty}^{\infty} x[n]y^*[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega})Y^*(e^{j\omega})d\omega$$



Proof of (6):

Let

$$w[n] \stackrel{\text{def}}{=} \sum_{k=-\infty}^{\infty} x[k]y[n-k] = x[n] * y[n]$$

$$W(e^{j\omega}) = ?$$

$$\begin{aligned} W(e^{j\omega}) &= \sum_{n=-\infty}^{\infty} \left( \sum_{k=-\infty}^{\infty} x[k]y[n-k] \right) e^{-j\omega n} \\ &= \sum_{k=-\infty}^{\infty} x[k] \left( \sum_{n=-\infty}^{\infty} y[n-k] e^{-j\omega n} \right) \\ &= \sum_{k=-\infty}^{\infty} x[k] e^{-j\omega k} \left( \sum_{m=-\infty}^{\infty} y[m] e^{-j\omega m} \right) \\ &= X(e^{j\omega}) Y(e^{j\omega}) \end{aligned}$$

Proof of (8) using (6):

Let

$$w[n] \stackrel{\text{def}}{=} \sum_{k=-\infty}^{\infty} x[k]x^*[k-n] = x[n] * x^*[-n]$$

$$W(e^{j\omega}) = X(e^{j\omega})X^*(e^{j\omega}) = |X(e^{j\omega})|^2$$

$$w[0] = \sum_{k=-\infty}^{\infty} x[k]x^*[k] = \sum_{k=-\infty}^{\infty} |x[k]|^2$$

$$w[0] = \frac{1}{2\pi} \int_{-\pi}^{\pi} W(e^{j\omega}) d\omega = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega$$

Proof of (9):

$$\begin{aligned} & \sum_{n=-\infty}^{\infty} x[n]y^*[n] \\ &= \sum_{n=-\infty}^{\infty} \left( \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega \right) y^*[n] \\ &= \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) \underbrace{\left( \sum_{n=-\infty}^{\infty} y^*[n] e^{j\omega n} \right)}_{Y^*(e^{j\omega})} d\omega \\ &= \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) Y^*(e^{j\omega}) d\omega \end{aligned}$$

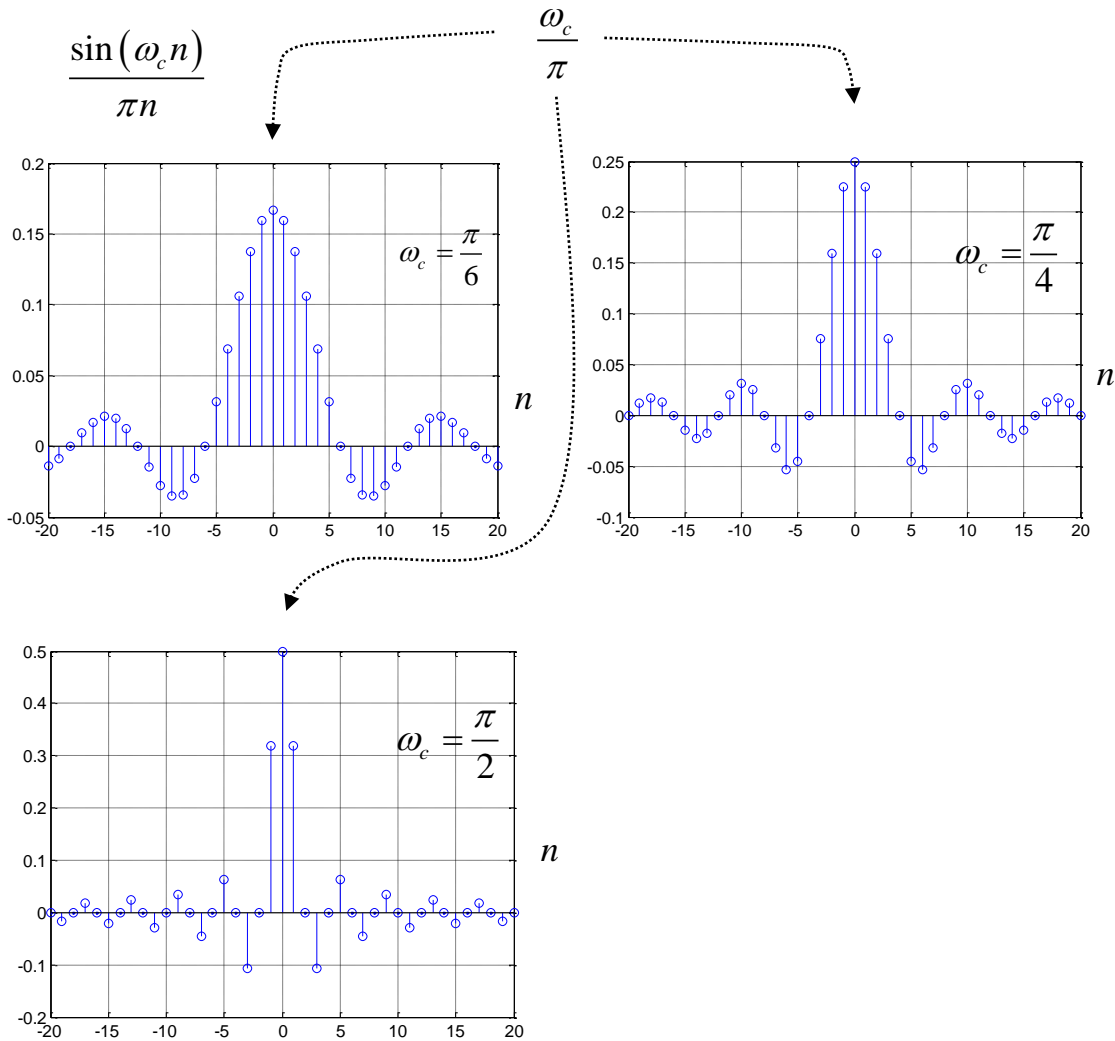
# FOURIER TRANSFORM PAIRS

$\delta[n]$	$1$
$\delta[n - n_0]$	$e^{-j\omega n_0}$
$1 \quad -\infty < n < \infty$	$X(e^{j\omega}) = 2\pi \sum_{k=-\infty}^{\infty} \delta[\omega + 2\pi k]$
$a^n u[n] \quad  a  < 1$	$\frac{1}{1 - ae^{-j\omega}}$
$u[n]$	$\frac{1}{1 - ae^{-j\omega}} + \pi \sum_{k=-\infty}^{\infty} \delta(\omega + 2\pi k)$
$na^n u[n] \quad  a  < 1$	$\frac{ae^{-j\omega}}{(1 - ae^{-j\omega})^2}$
$(n + 1)a^n u[n]$ $= (a^n u[n]) * (a^n u[n])$ $ a  < 1$	$\frac{1}{(1 - ae^{-j\omega})^2}$
$\frac{1}{2}(n + 2)(n + 1)a^n u[n]$	$\frac{1}{(1 - ae^{-j\omega})^3}$
$\frac{(n + k - 1)!}{(k - 1)! n!} a^n u[n]$	$\frac{1}{(1 - ae^{-j\omega})^k}$
$\frac{1}{\sin(\omega_0)} r^n \sin(\omega_0(n + 1)) u[n]$ $ r  < 1$	$\frac{1}{1 - 2r \cos(\omega_0) e^{-j\omega} + r^2 e^{-j2\omega}}$ show using the DTFT of $a^n u[n]$

$\frac{\sin(\omega_c n)}{\pi n}$	$X(e^{j\omega}) = \begin{cases} 1 &  \omega  < \omega_c \\ 0 & \omega_c <  \omega  \leq \pi \end{cases}$
$x[n] = \begin{cases} 1 & 0 \leq n \leq M \\ 0 & \text{otherwise} \end{cases}$	$\frac{\sin\left(\omega \frac{(M+1)}{2}\right)}{\sin\left(\frac{\omega}{2}\right)} e^{-j\omega \frac{M}{2}}$
$e^{j\omega_0 n}$	$2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - \omega_0 + 2\pi k)$
$\cos(\omega_0 n + \phi)$	$\pi e^{j\phi} \sum_{k=-\infty}^{\infty} \delta(\omega - \omega_0 + 2\pi k) + \pi e^{-j\phi} \sum_{k=-\infty}^{\infty} \delta(\omega + \omega_0 + 2\pi k)$
$\sin(\omega_0 n + \phi)$	$-j\pi e^{j\phi} \sum_{k=-\infty}^{\infty} \delta(\omega - \omega_0 + 2\pi k) + j\pi e^{-j\phi} \sum_{k=-\infty}^{\infty} \delta(\omega + \omega_0 + 2\pi k)$

## Ex: IDEAL LOWPASS FILTER (IMPULSE RESPONSE)

Plots of  $\sin \frac{(\omega_c n)}{\pi n}$  sequence

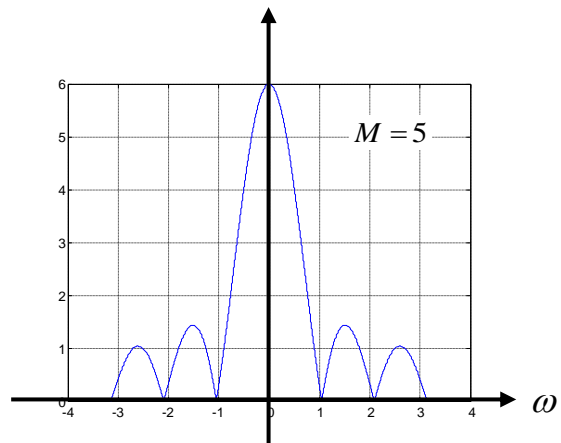
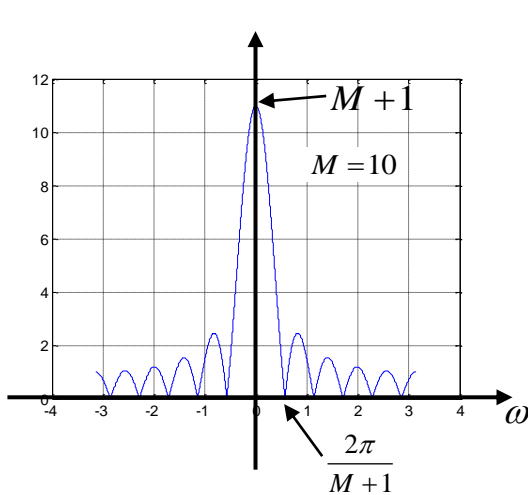


They are infinitely long sequences in  $-\infty < n < \infty$   
 Plots are arbitrarily in  $-20 < n < 20$

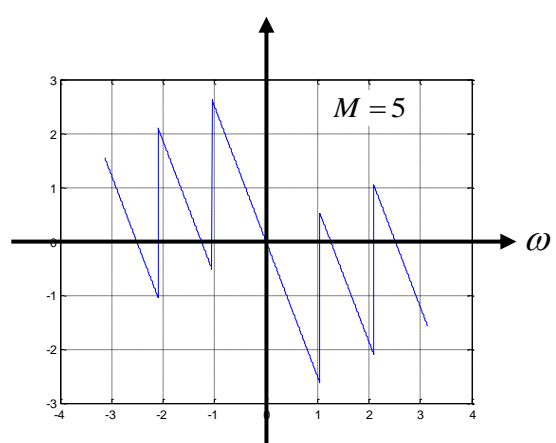
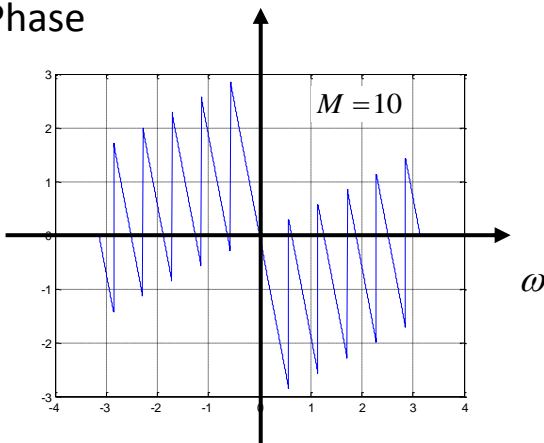
## Ex: MOVING AVERAGE FILTER

Magnitude

$$\frac{\sin\left(\omega \frac{(M+1)}{2}\right)}{\sin\left(\frac{\omega}{2}\right)} e^{-j\omega \frac{M}{2}}$$



Phase



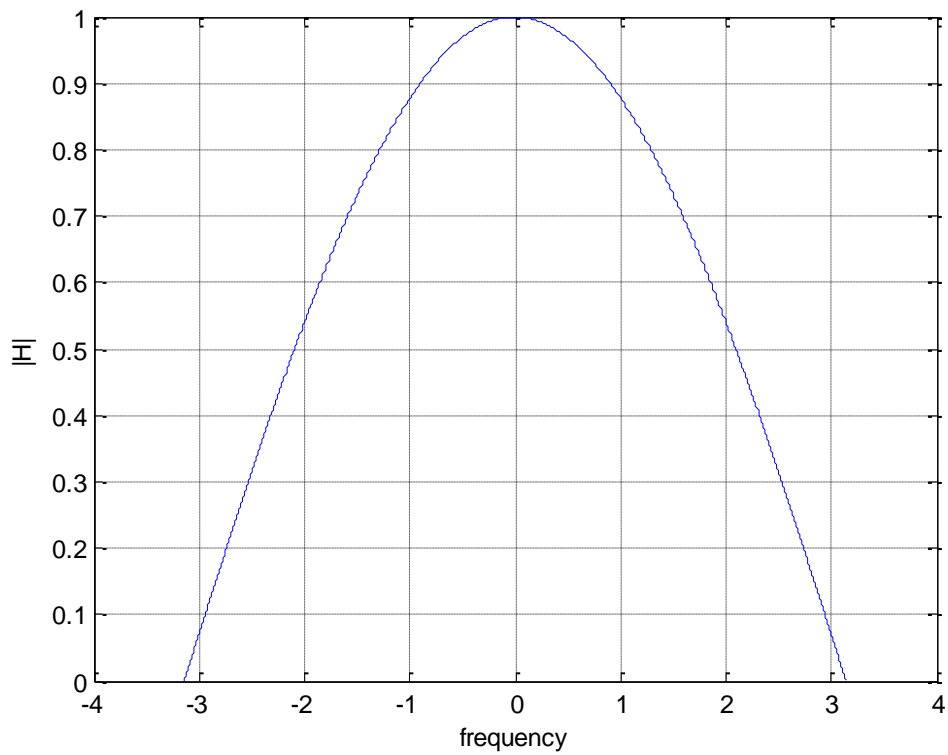
DTFT functions are plotted in  $-\pi \leq \omega < \pi$  or in  $0 \leq \omega < 2\pi$

$$M = 1$$

$$h[n] = \frac{1}{2}(\delta[n] + \delta[n-1]) \leftrightarrow H(e^{j\omega}) = \frac{1}{2}(1 + e^{-j\omega})$$

$$= e^{-j\frac{\omega}{2}} \cos\left(\frac{\omega}{2}\right)$$

$$y[n] = \frac{1}{2}(x[n] + x[n-1])$$





$$M = 4$$

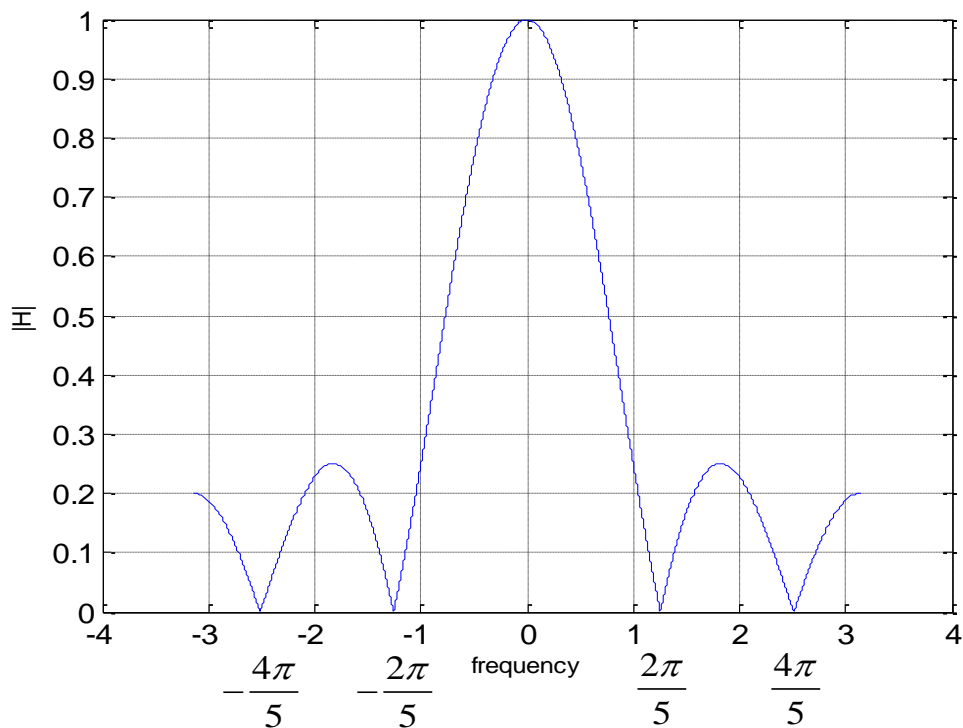
$$h[n] = \frac{1}{5} (\delta[n] + \delta[n-1] + \delta[n-2] + \delta[n-3] + \delta[n-4]) \leftrightarrow H(e^{j\omega})y[n]$$

$$y[n] = \frac{1}{2} (x[n] + x[n-1] + x[n-2] + x[n-3] + x[n-4])$$

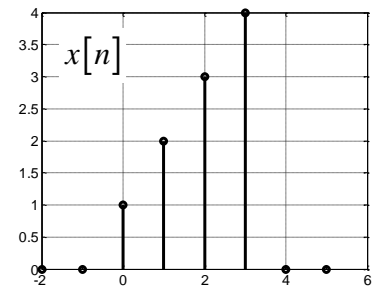
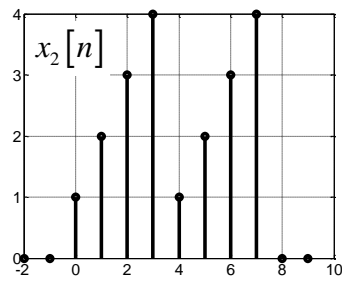
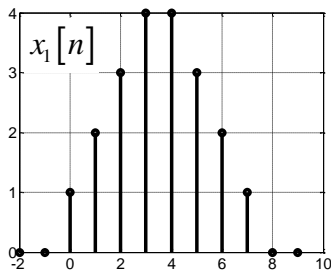
$$H(e^{j\omega}) = \frac{1}{5} (1 + e^{-j\omega} + e^{-j2\omega} + e^{-j3\omega} + e^{-j4\omega})$$

$$= \frac{1}{5} e^{-j2\omega} (1 + 2 \cos(\omega) + 2 \cos(2\omega))$$

$$= \frac{1}{5} e^{-j2\omega} \frac{\sin\left(\frac{5}{2}\omega\right)}{\sin\left(\frac{\omega}{2}\right)}$$



**Ex:** Express  $X_1(e^{j\omega})$ ,  $X_2(e^{j\omega})$ , in terms of  $X(e^{j\omega})$ , the DTFT of  $x[n]$ .



$$x_1[n] = x[n] + x[-(n - 7)]$$

$$X_1(e^{j\omega}) = X(e^{j\omega}) + e^{-j7\omega}X(e^{-j\omega})$$

$$x_2[n] = x[n] + x[n - 4]$$

$$X_2(e^{j\omega}) = X(e^{j\omega}) + e^{-j4\omega}X(e^{j\omega})$$

One can also write as

$$\begin{aligned}X_1(e^{j\omega}) &= X(e^{j\omega}) + e^{-j7\omega}X(e^{-j\omega}) \\&= X(e^{j\omega}) + e^{-j7\omega}X^*(e^{j\omega}) \quad \text{since } x[n] \text{ is real} \\&= |X(e^{j\omega})| \left( e^{j4X(e^{j\omega})} + e^{-j7\omega} e^{-j4X(e^{j\omega})} \right) \\&= |X(e^{j\omega})| e^{-j\frac{7}{2}\omega} \left( e^{j\frac{7}{2}\omega} e^{j4X(e^{j\omega})} + e^{-j\frac{7}{2}\omega} e^{-j4X(e^{j\omega})} \right) \\&= 2 \operatorname{Re} \left\{ e^{j(4X(e^{j\omega}) + \frac{7}{2}\omega)} \right\} |X(e^{j\omega})| e^{-j\frac{7}{2}\omega}\end{aligned}$$

$$\begin{aligned}X_2(e^{j\omega}) &= X(e^{j\omega})(1 + e^{-j4\omega}) \\&= X(e^{j\omega})e^{-j2\omega}(e^{j2\omega} + e^{-j2\omega}) \\&= X(e^{j\omega})e^{-j2\omega}2 \cos(2\omega) \\&= 2 \cos(2\omega) |X(e^{j\omega})| e^{-j2\omega} e^{j\angle X(e^{j\omega})}\end{aligned}$$

**Ex:** What is the inverse DTFT,  $y[n]$ , of

$$Y(e^{j\omega}) = \frac{2e^{-j3\omega}}{\left(1 - \frac{1}{8}e^{-j\omega}\right)^2} ?$$

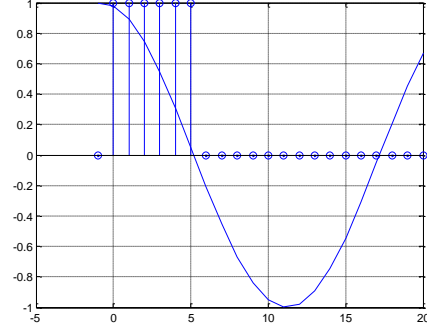
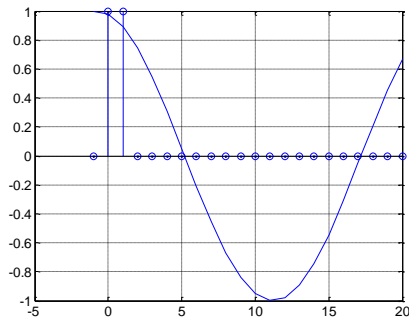
From the table

$$(n+1)a^n u[n] \leftrightarrow \frac{1}{(1 - ae^{-j\omega})^2} \quad |a| < 1$$

$$2(n+1)\left(\frac{1}{8}\right)^n u[n] \leftrightarrow \frac{2}{\left(1 - \frac{1}{8}e^{-j\omega}\right)^2}$$

$$2(n-2)\left(\frac{1}{8}\right)^{n-3} u[n-3] \leftrightarrow \frac{2e^{-j3\omega}}{\left(1 - \frac{1}{8}e^{-j\omega}\right)^2}$$

Why does the high frequency gain of MA filter “decrease” as M increases?



Comment on the above illustrations.

## LCCDEs AND FREQUENCY RESPONSE

$$y[n] = x[n] * h[n] \xleftrightarrow{FT} Y(e^{j\omega}) = X(e^{j\omega})H(e^{j\omega})$$

$$\Rightarrow H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})}$$

$$\begin{aligned} \sum_{k=0}^N a_k y[n-k] &= \sum_{k=0}^M b_k x[n-k] \xleftrightarrow{FT} \sum_{k=0}^N a_k e^{-jk\omega} Y(e^{j\omega}) \\ &= \sum_{k=0}^M b_k e^{-jk\omega} X(e^{j\omega}) \end{aligned}$$

$$\Rightarrow Y(e^{j\omega}) \sum_{k=0}^N a_k e^{-jk\omega} = X(e^{j\omega}) \sum_{k=0}^M b_k e^{-jk\omega}$$

$$\Rightarrow \frac{Y(e^{j\omega})}{X(e^{j\omega})} = H(e^{j\omega}) = \frac{\sum_{k=0}^M b_k e^{-jk\omega}}{\sum_{k=0}^N a_k e^{-jk\omega}}$$

$$H(e^{j\omega}) = \frac{b_0 + b_1 e^{-j\omega} + b_2 e^{-j2\omega} + \dots + b_M e^{-jM\omega}}{a_0 + a_1 e^{-j\omega} + a_2 e^{-j2\omega} + \dots + a_N e^{-jN\omega}}$$