

$$H(z) = \frac{1 + 2z^{-2}}{1 + \frac{3}{4}z^{-1} + \frac{1}{8}z^{-2}}$$

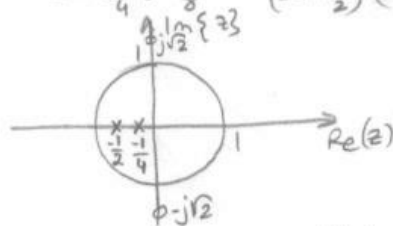
b) Plot the approximate magnitude characteristics of $H(e^{j\omega})$ by indicating $\omega=0$ and $\omega=\pi$ values. Determine the filter type.

d) $H(z)$ is cascade connected with a suitable filter $H_2(z)$ and $G(z)=H(z)H_2(z)$ is obtained as a FIR generalized linear-phase Type III filter. Find the minimum order $G(z)$ and plot its pole-zero diagram.

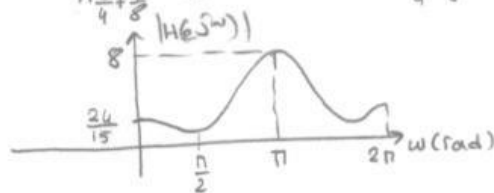
a) $H(z) = \frac{z^2 + 2}{z^2 + \frac{3}{4}z + \frac{1}{8}}$

Zeros: $z_1 = j\sqrt{2}$, $z_2 = -j\sqrt{2}$

Poles: $z_3 = -\frac{1}{2}$, $z_4 = -\frac{1}{4}$

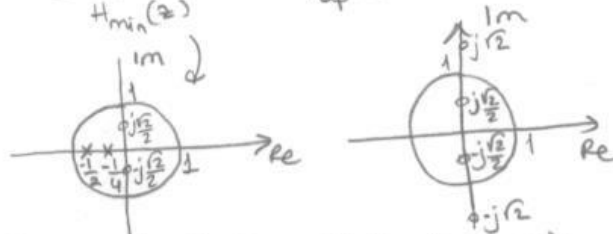


b) $H(e^{j0}) = \frac{3}{1 + \frac{3}{4} + \frac{1}{8}} = \frac{24}{15}$ $H(e^{j\pi}) = \frac{1+2}{1 - \frac{3}{4} + \frac{1}{8}} = 8$



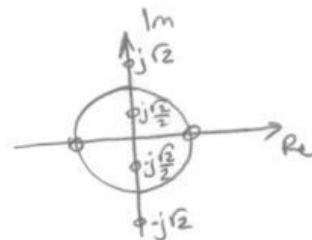
This is a highpass filter.

$$c) \quad H(z) = \frac{1+2z^{-2}}{1+\frac{3}{4}z^{-1}+\frac{1}{8}z^{-2}} = \underbrace{\frac{1+2z^{-2}}{1+\frac{3}{4}z^{-1}+\frac{1}{8}z^{-2}}}_{H_{mi}(z)} \cdot \underbrace{\frac{1+2z^{-2}}{1+2z^{-2}}}_{H_{ap}(z)}$$



$$d) \quad G(z) = \frac{1+2z^{-2}}{1+\frac{3}{4}z^{-1}+\frac{1}{8}z^{-2}} \cdot (1+\frac{3}{4}z^{-1}+\frac{1}{8}z^{-2}) \cdot (2+z^{-2})(1+z^{-1})(1-z^{-1})$$

$$G(z) = (1 + 2z^{-1})(2 + z^{-2})(1 + z^{-1})(1 - z^{-1})$$



Q3)

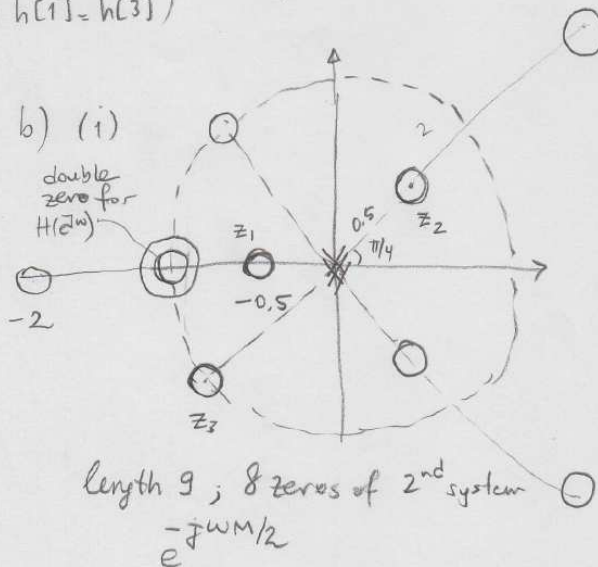
- a) The impulse response of a system, $h[n]$ is given as $h[1]=1, h[2]=2, h[3]=1$ and 0 for all other values of $h[n]$. Find and plot the magnitude and phase response ($\text{ARG}[\cdot]$) for this system. Comment on the type of this filter.
- b) Assume the system in part-a) is cascaded by a real Type-I generalized linear-phase system with length 9, whose zeros are equal to $z_1=-0.5, z_2=0.5e^{j\pi/4}, z_3=e^{-j3\pi/4}$.
- Find the overall pole-zero plot of the cascaded system
 - Find the group delay of the cascaded system.
- c) For the system in part-a), select $h[2]$ and $h[3]$ (while remaining all other values unchanged), in such a way that the system is still a generalized linear-phase system and $H(e^{j0})=0$.

a) $H(e^{j\omega}) = \sum_{n=-\infty}^{\infty} h[n] e^{-j\omega n} = e^{-j\omega} + 2e^{-j2\omega} + e^{-j3\omega} = e^{-j2\omega} (e^{+j\omega} + e^{-j\omega}) + 2e^{-j2\omega}$

$|H(e^{j\omega})| = (2\cos(\omega) + 2) e^{-j2\omega}$

$h[n]$ is a M even symmetric \Rightarrow Type I
 $\begin{pmatrix} h[0]=h[4] \\ h[1]=h[3] \end{pmatrix} \quad M=4$

$H(z) = z^{-1} + 2z^{-2} + z^{-3}$
 $= (z^2 + 2z + 1)/z^3$
 $= (z+1)^2/z^3$
 double zero at $z=-1$



(ii) Group delay = $\text{Grd}[H(e^{j\omega})] + \text{Grd}[G(e^{j\omega})]$
 $= 2 + 4$
 $= 6$

c) To guarantee $H(e^{j0})=0$ system, assume a Type III system (M even)

\Rightarrow

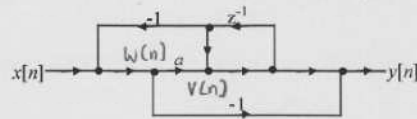
$h[n]$

$1 \quad ? \quad ?$

$0 \quad 1 \quad 2 \quad 3 \quad 4$

$\Rightarrow h[1]=-h[3] \quad \& \quad h[2]=0$
 $h[3]=-1$

Q4) The following signal flow graph representation is given for a system whose transfer function is $H(z)$.



a) Determine $H(z)$ in terms of the given system parameters, indicating its pole and zeros on the z -plane. Comment on the type of this IIR filter.

b) Obtain and draw the transpose of the above representation.

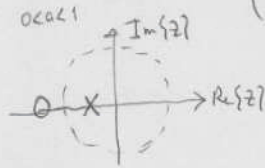
c) Let $G(z)$ be equal to $G(z) = \frac{1-z^{-1}}{1+0.5z^{-1}}$ and assume $G(z)$ is cascaded by itself. Draw the resulting

signal flow graph representation in direct form-I and modify it in such a way that 3 delay elements are utilized in this cascaded system.

$$\begin{aligned} a) \quad \left. \begin{aligned} v[n] &= a w[n] + v[n-1] \\ w[n] &= x[n] - v[n-1] \\ y[n] &= v[n] - w[n] \end{aligned} \right\} \begin{aligned} V(z) &= a W(z) + z^{-1} V(z) \Rightarrow V(z)(1 - z^{-1}) = a W(z) \\ W(z) &= X(z) - z^{-1} V(z) \Rightarrow W(z) = X(z) - \frac{a z^{-1}}{1 - z^{-1}} W(z) \\ Y(z) &= V(z) - W(z) \end{aligned} \end{aligned}$$

$$W(z) \left(\frac{1 - z^{-1} + a z^{-1}}{1 - z^{-1}} \right) = X(z)$$

$$\Rightarrow Y(z) = \left(\frac{a - (1 - z^{-1})}{1 - z^{-1}} \right) W(z) \Rightarrow Y(z) = \frac{(a-1) + z^{-1}}{1 - (a-1)z^{-1}} X(z)$$



All-pass system

