

EE430 - HW1
Solutions

1)
a) $x[n] = x_c(nT)$

$$= 4 \cdot \sin\left(\frac{20000\pi \cdot n}{3 \times 10^3} + \frac{\pi}{13}\right)$$

$$= 4 \cdot \sin\left(\frac{20\pi n}{3} + \frac{\pi}{13}\right) = 4 \cdot \sin\left(\frac{2\pi n}{3} + \frac{\pi}{13}\right)$$

$$x[n] = 4 \sin\left(\frac{2\pi n}{3} + \frac{\pi}{13} + 2\pi nk\right) \quad k = \dots, -1, 0, 1, \dots$$

$$\hat{x}(t) = 4 \sin\left(\frac{2\pi \cdot t}{3T} + \frac{\pi}{13} + 2\pi k \cdot \frac{t}{T}\right) \quad k = \dots, -1, 0, 1, \dots$$

$$= 4 \cdot \sin\left(2\pi \times 10^3 t + \frac{\pi}{13} + 6\pi \times 10^3 t \cdot k\right) \quad k = \dots, -1, 0, 1, \dots$$

$$= \boxed{4 \sin\left(2\pi \times 10^3 t (1+3k) + \frac{\pi}{13}\right)} \quad k = \dots, -1, 0, 1, \dots$$

b) $x[n] = 4 \cdot \sin\left(\frac{2\pi n}{3} + \frac{\pi}{13}\right) = 4 \sin\left(\frac{20000\pi n T}{3} + \frac{\pi}{13}\right)$

$$\frac{2\pi n}{3} + 2\pi nk = 2\pi \times 10^4 nT \Rightarrow \frac{1}{3} + k = 10^4 \cdot T$$

$$\frac{1+3k}{3 \cdot 10^4} = T \Rightarrow f_s = \frac{3 \times 10^4}{1+3k}$$

for $k = \dots, -1, 0, 1, \dots$

- 2) $\sin(1,74\pi(n+N) + 3,1) \stackrel{?}{=} \sin(1,74\pi n + 3,1)$
- $$\sin(1,74\pi n + \underbrace{1,74\pi \cdot N}_{2\pi k} + 3,1) \Rightarrow 2\pi k = 1,74\pi \cdot T$$
- $$\frac{2k}{1,74} = N \Rightarrow \frac{k}{0,87} = N \rightarrow \text{integer}$$
- $N_{\text{fund}} = 100$
- $\sin(1,74\pi n + 3,1\pi) \rightarrow N_{\text{fund}} = 100$
 - $\cos(15,74\pi n + \frac{3\pi}{8}) \rightarrow \frac{2\pi}{15,74\pi} k = T \rightarrow \text{integer}$
 - $\cos(\sqrt{\pi}n) \rightarrow \frac{2\pi}{\sqrt{\pi}} k = N \rightarrow \boxed{\text{not periodic}}$
 - $\cos(\pi\sqrt{\pi}n) \rightarrow \frac{2\pi}{\pi\sqrt{\pi}} k = N \rightarrow \boxed{\text{not periodic}}$
 - $\cos(\pi\sqrt{2}n) \rightarrow \frac{2\pi}{\pi\sqrt{2}} k = N \rightarrow \boxed{\text{not periodic}}$
- there is no integer value k , to obtain integer N .

3) $x[n] = \sum_{k=-\infty}^{+\infty} x[k], s[n-k] \xrightarrow{\text{linear}} y[n] = \sum_{k=-\infty}^{+\infty} x[k] h_k[n]$

$$y[n] = \sum_{k=-\infty}^{+\infty} x[k](n-k) u[n-k]$$

for $x[n-m]$, we want $y_1[n-m] = \sum_{k=-\infty}^{+\infty} x[k-m](n-k) u[n-k]$

but we have $y_2[n-m] = \sum_{k=-\infty}^{+\infty} x[k](n-k-m) u[n-k-m]$

$$y_1[n-m] = \sum_{k=-\infty}^n x[k-m] \cdot (n-k) \longrightarrow y_1[n-m] = y_2[n-m]$$

↑ time-invariant system.

$$y_2[n-m] = \sum_{k=-\infty}^{n-m} x[k] (n-k-m) = \sum_{k=0}^n x[k-m] (n-k)$$

4) $a x_1[n] + b x_2[n] \xrightarrow{\text{Linearity}} a y_1[n] + b y_2[n]$

$$a \cdot y_1[n] = \begin{cases} a \cdot x_1\left[\frac{n}{2}\right], & n \text{ is even} \\ \frac{a x_1\left[\frac{n-1}{2}\right] + a x_1\left[\frac{n+1}{2}\right]}{2}, & n \text{ is odd} \end{cases}$$

$a x_1[n] \xrightarrow{\quad} a y_1[n] \checkmark$

$$b \cdot y_2[n] = \begin{cases} b \cdot x_2\left[\frac{n}{2}\right], & n \text{ is even} \\ \frac{b x_2\left[\frac{n-1}{2}\right] + b x_2\left[\frac{n+1}{2}\right]}{2}, & n \text{ is odd} \end{cases}$$

$b x_2[n] \xrightarrow{\quad} b y_2[n] \checkmark$

$$\begin{aligned} a y_1[n] \\ + \\ b y_2[n] \end{aligned} = \begin{cases} a \cdot x_1\left[\frac{n}{2}\right] + b x_2\left[\frac{n}{2}\right], & n \text{ is even} \\ \frac{a x_1\left[\frac{n-1}{2}\right] + a x_1\left[\frac{n+1}{2}\right] + b x_2\left[\frac{n-1}{2}\right] + b x_2\left[\frac{n+1}{2}\right]}{2}, & n \text{ is odd} \end{cases}$$

when we take " $a x_1[n] + b x_2[n]$ " as a input of the system,
the same result can be obtained for " $a y_1[n] + b y_2[n]$ ".

Therefore, the system is linear.

For a system $x[n] \xrightarrow{\quad} x\left[\frac{n}{2}\right]$

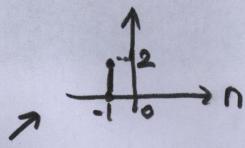
<u>input</u>	<u>output</u>
$x[n-m] \rightarrow x\left[\frac{n}{2}-m\right] = y_1[n-m]$	
$x[n-m] \rightarrow x\left[\frac{n-m}{2}\right] = y_2[n-m]$	

Also, the value of n (even or odd) has

an effect on the response \rightarrow time-varying system

$y_1[n-m] \neq y_2[n-m]$
the system is NOT time-invariant

$$5) y[n] = 2^{\delta[n+1]} + x[n-3]$$



* For bounded input, $x[n-3]$ is bounded and $2^{\delta[n+1]}$ is always bounded, so the output is bounded \rightarrow stable system

* We can easily see that the output depends on previous values of input \rightarrow causal system

$$y[n] = \begin{cases} y[-\delta[n-1]] + x[n-3] & n > 0 \\ 2^n x[n-3] & n \leq 0 \end{cases}$$

* $y[0] = 1 \cdot x[-3] \rightarrow y[0]$ is bounded for bounded input.

$$y[1] = y[-1] + x[-2] \rightarrow y[1]$$
 is bounded

$$y[-1] = \frac{1}{2} x[-4] \rightarrow y[-1]$$
 is bounded

If $n \neq 1$, $y[n] = y[0] + x[n-3] \Rightarrow y[n]$ is bounded
 $n > 0$

\rightarrow a constant

If $n \leq 0$, $0 < 2^n < 1$ and $|x[n]| < M \Rightarrow y[n]$ is bounded

* The system is stable

* The system is causal (The system output $y[n]$ depends on previous values of $x[n]$)

$$6) y[n] = \sum_{k=-\infty}^{+\infty} x[k] \cdot x[n-k] = \sum_{k=-6}^{24} x[k] \cdot x[n-k]$$

The first term $\rightarrow -3 s[n+6] \rightarrow$

The last term $\rightarrow -4 s[n-24]$

$-3 s[n+6] * x[n] \Rightarrow$ shift to the left (6) and multiply with -3

The first non-zero term $9 s[n+12]$

$-4 s[n-24] * x[n] \Rightarrow$ shift to the right (24) and mult. with -4

The last non-zero term $16 s[n-48]$

$$7) x[n] = u[n] = s[n] + s[n-1] + \dots \rightarrow h[n] \rightarrow h[n] + h[n-1] + \dots$$

$$y[n] = \sum_{k=0}^{\infty} h[n-k] = \sum_{k=0}^{\infty} 3\left(\frac{1}{2}\right)^{n-k} u[n-k] - 2\left(\frac{1}{3}\right)^{n-k-1} u[n-k]$$

$$= \sum_{k=0}^n \left(3\left(\frac{1}{2}\right)^{n-k} - 2\left(\frac{1}{3}\right)^{n-k-1} \right)$$

$$y[n] = \left(\frac{1}{3}\right)^{n-1} - 3\left(\frac{1}{2}\right)^n - 3 = 3 \cdot \left(\frac{1}{2}\right)^n \cdot \sum_{k=0}^n \left(2^k - 2\left(\frac{1}{3}\right)^{n-1-k} \right)$$

$$= 3 \left(\frac{1}{2}\right)^n \cdot \left(2^{n+1-1} - 2\left(\frac{1}{3}\right)^{n-1} \left(\frac{3^{n+1}-1}{2} \right) \right)$$

$$= 6 - 3\left(\frac{1}{2}\right)^n - 9 + \left(\frac{1}{3}\right)^{n-1} = \boxed{\left(\frac{1}{3}\right)^{n-1} - 3\left(\frac{1}{2}\right)^n - 3}$$

as $n \rightarrow \infty$

$$\boxed{y[n] = -3}$$

$$8) h[n] - \frac{1}{2} h[n-1] = s[n] - s[n-1] + s[n-2]$$

$$h[0] - \frac{1}{2} h[-1] = 1$$

$$h[-1] - \frac{1}{2} h[-2] = 0$$

$$h[1] - \frac{1}{2} h[0] = -1$$

$$h[2] - \frac{1}{2} h[1] = 1$$

$$h[3] - \frac{1}{2} h[2] = 0$$

For a causal LTI system, $h[n]$ should be zero for $n < 0$.

$$h[0] = 1 \quad h[1] = -\frac{1}{2} \quad h[2] = \frac{3}{4}$$

$$h[n] = \frac{3}{4} \left(\frac{1}{2}\right)^{n-2} \text{ for } n \geq 2$$

$$h[n] = \frac{3}{4} \left(\frac{1}{2}\right)^{n-2}, u[n] - 2s[n-1] - 2s[n]$$

$$b) y(e^{j\omega}) - \frac{1}{2} y(e^{j\omega}) \cdot e^{-j\omega} = X(e^{j\omega}) [1 - e^{-j\omega} + e^{+j2\omega}]$$

$$\frac{y(e^{j\omega})}{X(e^{j\omega})} = \frac{1 - e^{-j\omega} + e^{+j2\omega}}{1 - \frac{1}{2} e^{-j\omega}} \quad (h[n] = \left(\frac{1}{2}\right)^n u[n] - \left(\frac{1}{2}\right)^{n-1} u[n-1] + \left(\frac{1}{2}\right)^{n-2} u[n-2])$$

$$d) H(e^{j\frac{\pi}{3}}) = \frac{1 - e^{-j\frac{\pi}{3}} + e^{-j\frac{2\pi}{3}}}{1 - \frac{1}{2} e^{-j\frac{\pi}{3}}}$$

$$= \frac{1 - \frac{1}{2}e^{j\frac{\pi}{3}} + \frac{1}{2}e^{j\frac{2\pi}{3}}}{1 - \frac{1}{2}e^{-j\frac{\pi}{3}}} = 0$$

$$H(e^{j\frac{\pi}{2}}) = \frac{1 - e^{-j\frac{\pi}{2}} + e^{-j\pi}}{1 - \frac{1}{2} e^{-j\frac{\pi}{2}}}$$

$$= \frac{j}{1 + \frac{1}{2}j} = |H(e^{j\frac{\pi}{2}})| \cdot e^{j \angle H(e^{j\frac{\pi}{2}})}$$

$$y[n] = |H(e^{j\frac{\pi}{2}})| \sin\left(\frac{\pi}{2}n + \frac{\pi}{4} + \angle H(e^{j\frac{\pi}{2}})\right)$$

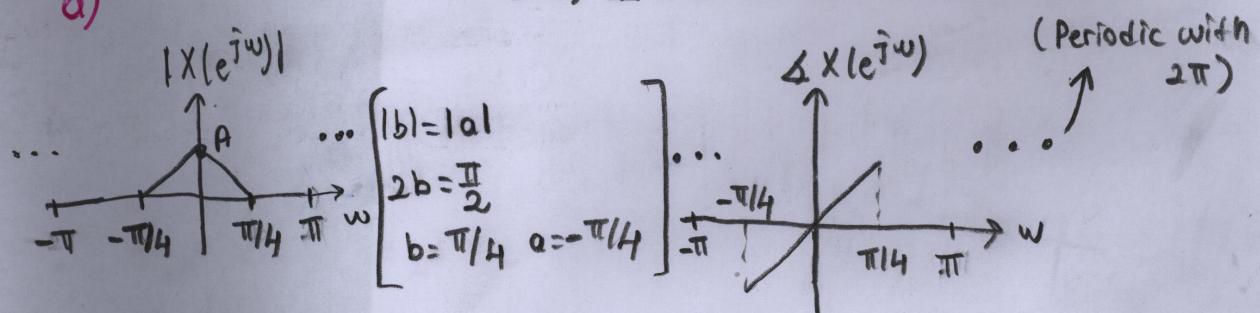
e) $H^*(e^{j(2\pi-\omega)}) = H^*(e^{-j\omega}) \rightarrow$ periodic with 2π
 $= F\{h^*[n]\}$

$$h[n] = h^*[n] \rightarrow \text{for real valued } h[n]$$

$$(H^*(e^{j(2\pi-\omega)}) = \frac{1-e^{-j\omega}+e^{-j2\omega}}{(1-\frac{1}{2}e^{-j\omega})})$$

g) For real valued $x[n] \rightarrow |X(e^{j\omega})|$ is an even function.

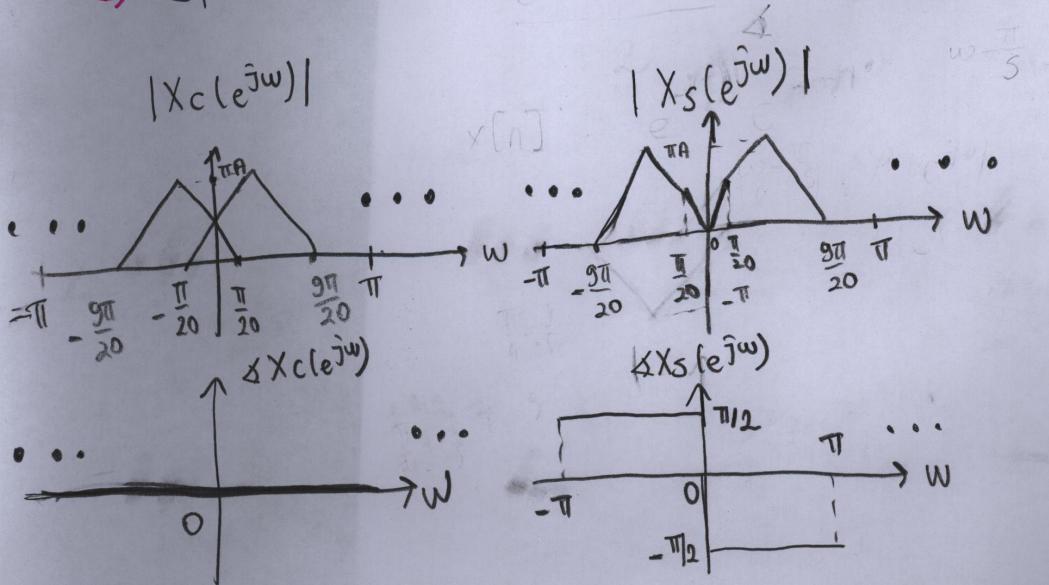
a) $\Delta X(e^{j\omega})$ is an odd function



b) $X_c(e^{j\omega}) = \pi [X(e^{j(\omega - \frac{\pi}{5})}) + X(e^{j(\omega + \frac{\pi}{5})})]$

$$X_S(e^{j\omega}) = \frac{\pi}{j} [X(e^{j(\omega - \frac{\pi}{5})}) - X(e^{j(\omega + \frac{\pi}{5})})]$$

c) If $X(e^{j\omega})$ is real valued $\rightarrow \Delta X(e^{j\omega}) = 0$



10) 2.55)

$$X(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} x[n] \cdot e^{-jn\omega}$$

$$\bullet X(e^{j0}) = \sum_{n=-\infty}^{+\infty} x[n] = 8 \quad \bullet X(e^{j\pi}) = \sum_{n=-\infty}^{+\infty} x[n] (-1)^n = 2 - (6) = -4$$

• $x[n+2] \Rightarrow$ real and even function

$$X(e^{j\omega}) \cdot e^{j2\omega} \Rightarrow \text{real valued} \quad \boxed{\Delta X(e^{j\omega}) = -2\omega}$$

$$\bullet x[n] = \frac{1}{2\pi} \int_{-2\pi}^{2\pi} X(\omega) \cdot e^{j\omega n} d\omega \Rightarrow x[0] = \frac{1}{2\pi} \int_{-2\pi}^{2\pi} X(\omega) d\omega$$

$$\boxed{\int_{-2\pi}^{2\pi} X(\omega) d\omega = 2\pi}$$

$$\bullet X(e^{-j\omega}) \xrightarrow{F^{-1}} \boxed{x[-n]}$$

$$\bullet \operatorname{Re}\{X(e^{j\omega})\} = \frac{X(e^{j\omega}) + X^*(e^{j\omega})}{2}$$

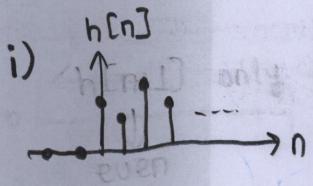
$$\xrightarrow{F^{-1}} \frac{x[n] + x^*[-n]}{2} = \boxed{\frac{x[n] + x[-n]}{2}}$$

11) 2.60) a) ii) $H(e^{j\omega}) = H^*(e^{-j\omega}) \Rightarrow h[n] = h^*[n]$
(real valued)

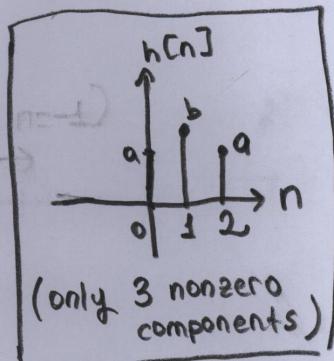
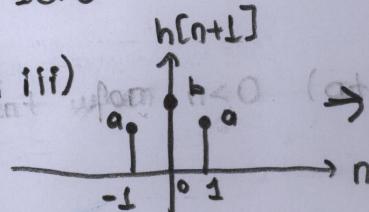
iii) $H(e^{j\omega}) \cdot e^{j\omega}$ is real $\Rightarrow \Delta H(e^{j\omega}) = -\omega$

$h[n+1] \Rightarrow$ real and even

i) for $n < 0$, $h[n]$ is zero



i) and iii)



11) 2.60) b)

iv) $\frac{1}{2\pi} \int_{-\pi}^{\pi} H(e^{j\omega}) = \boxed{h[0] = 2}$

\uparrow
from (10)

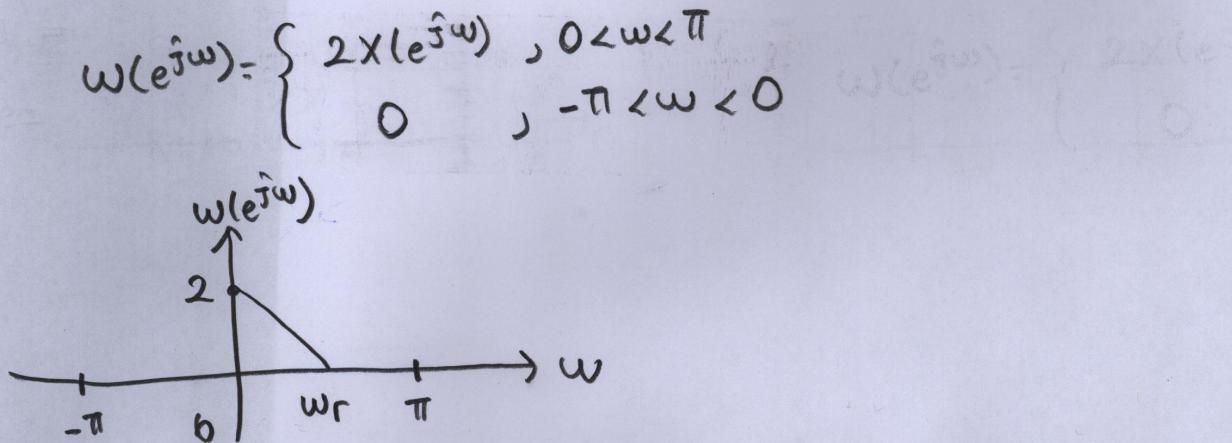
$\boxed{h[2]} = h[2]$

v) $H(e^{j\pi}) = \sum_{n=0}^2 h[n](-1)^n = 0$
 \downarrow
 $\boxed{h[1]} = 4$

$$\begin{aligned} &= h[0] - h[1] + h[2] = 0 \\ &= 4 - h[1] = 0 \end{aligned}$$

$$12) 2.65) \quad y(e^{j\omega}) = \begin{cases} -jx(e^{j\omega}), & 0 < \omega < \pi \\ +jx(e^{j\omega}), & -\pi < \omega < 0 \end{cases}$$

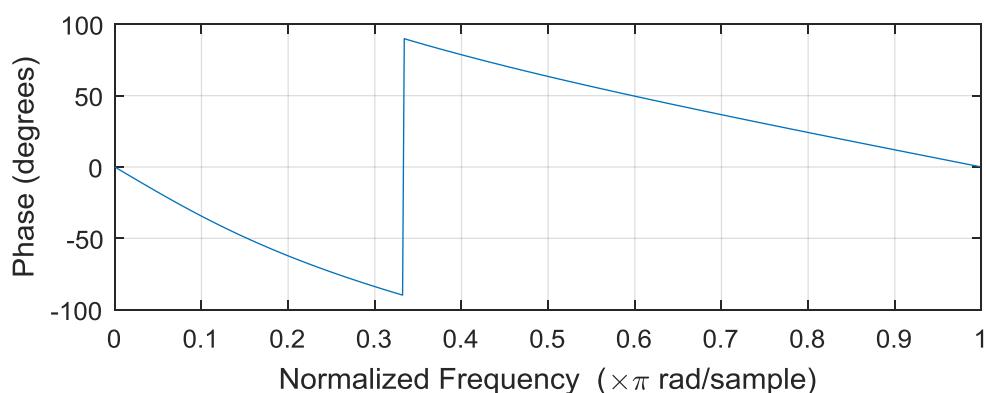
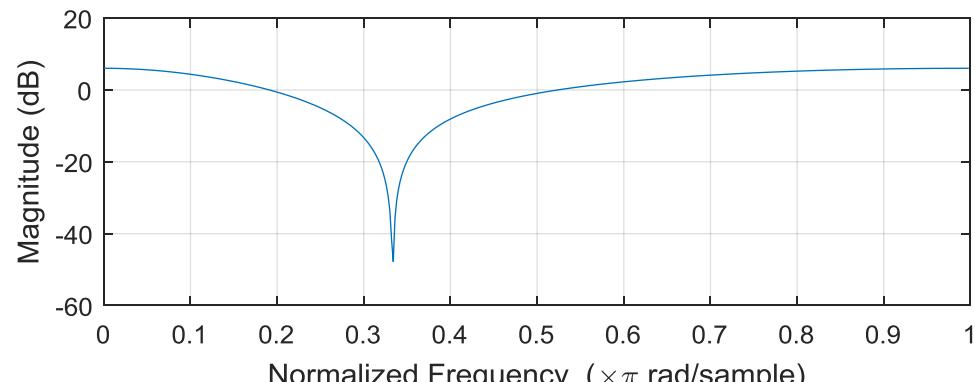
$$\begin{aligned} w(e^{j\omega}) &= x(e^{j\omega}) + j \underbrace{y(e^{j\omega})}_{\begin{cases} x(e^{j\omega}), & 0 < \omega < \pi \\ -x(e^{j\omega}), & -\pi < \omega < 0 \end{cases}} \\ &= \begin{cases} 2x(e^{j\omega}), & 0 < \omega < \pi \\ 0, & -\pi < \omega < 0 \end{cases} \end{aligned}$$



MATLAB PART

8) c)

```
>> freqz([1 -1 1], [1 -1/2])
```

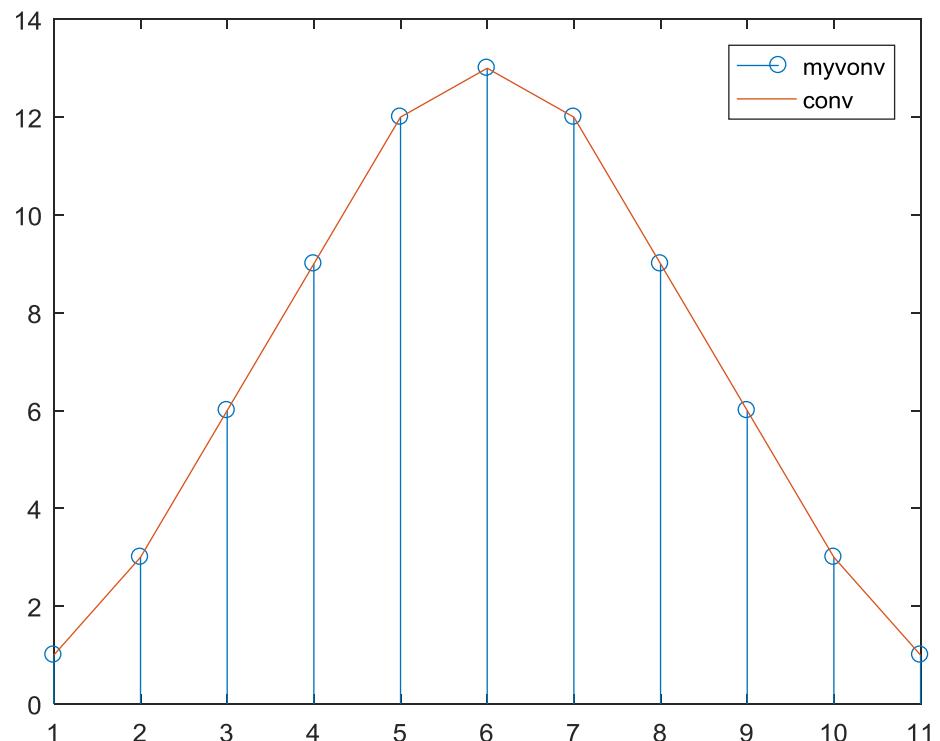


13)

a)

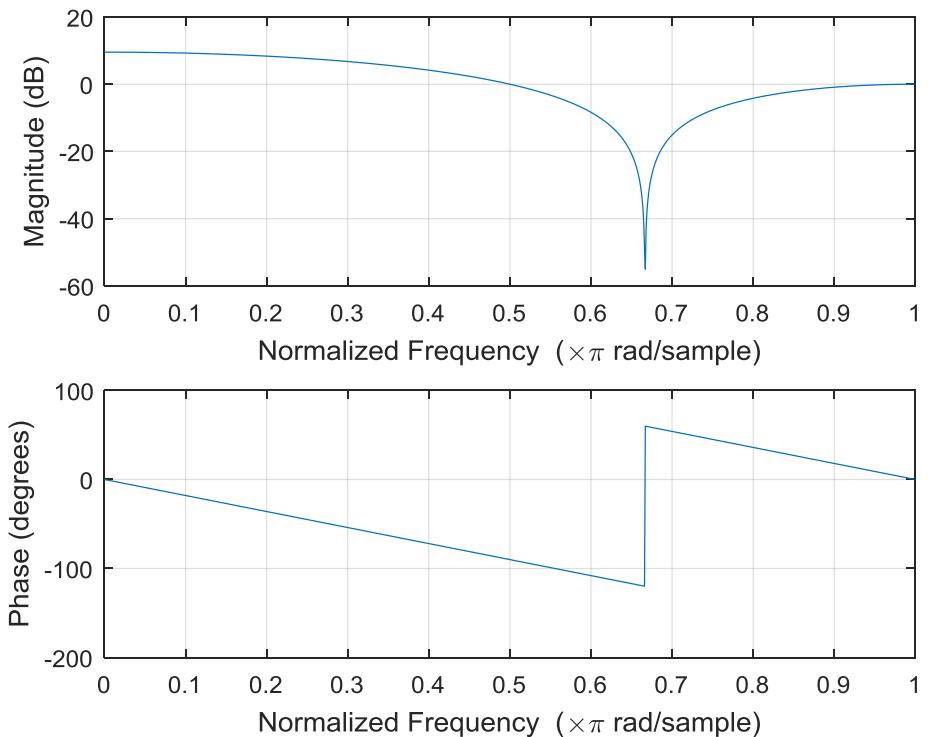
```
1 function y=myconv(x,h)
2 y=zeros(1,length(h)+length(x)-1);
3 for kk=1:length(h)
4     y(kk:kk+length(x)-1)=y(kk:kk+length(x)-1)+h(kk)*x;
5 end
6 end
7
```

b)



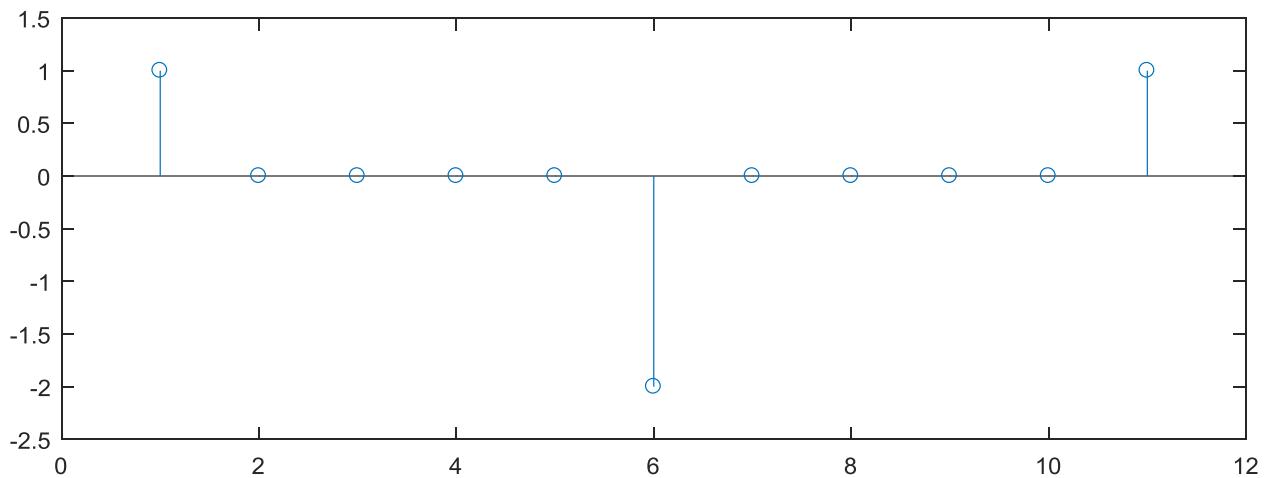
c)

```
>> freqz(h, 1, 1024)
```

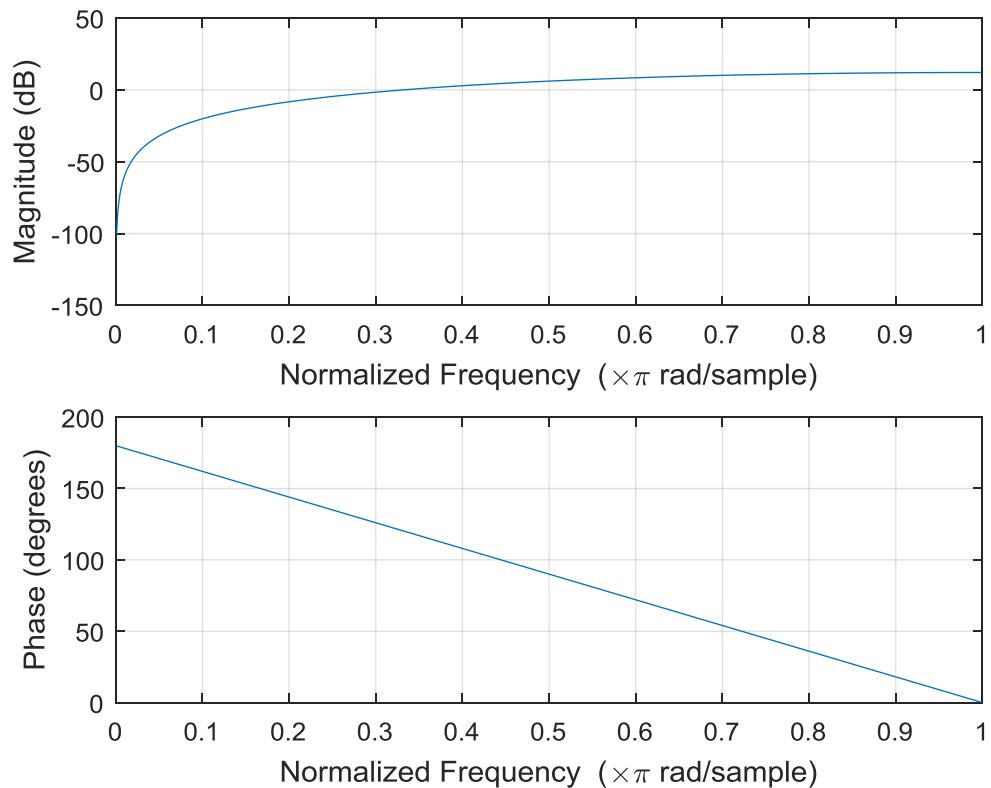


It is a lowpass filter and it is a moving average filter. Hence, it smoothes the original signal.

d)

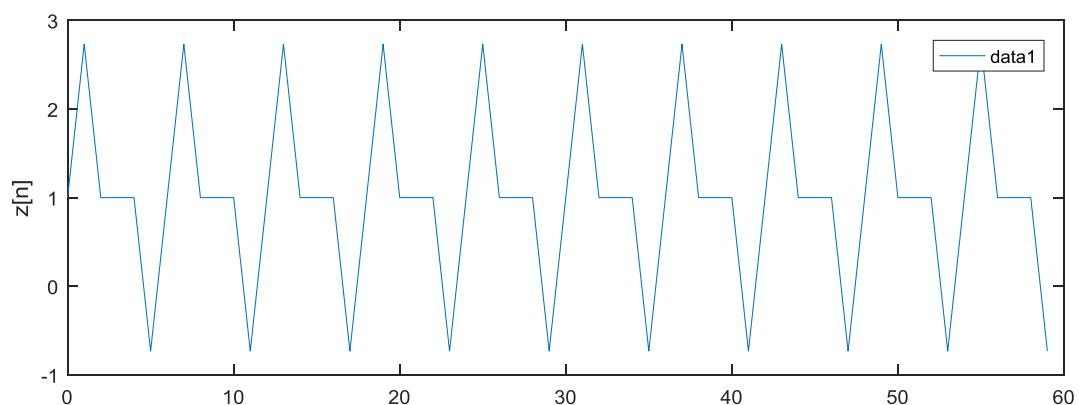


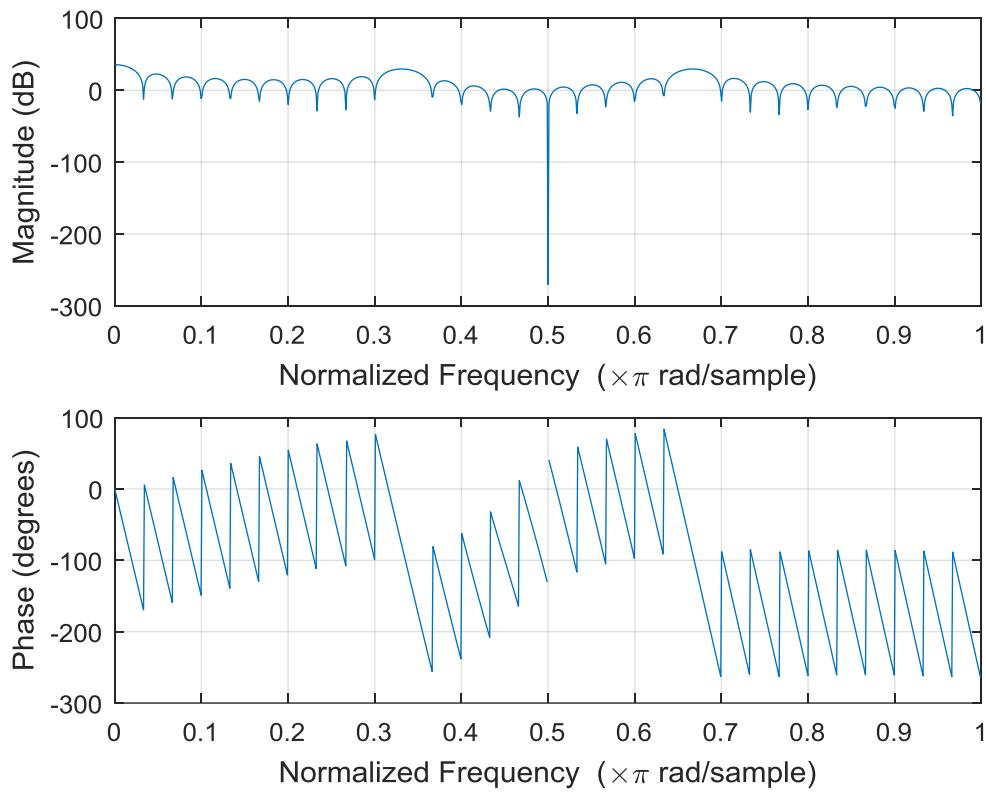
e)



It is a highpass filter and the filter extracts the high frequency component of the input signal. Notice the sharp changes in the output signal.

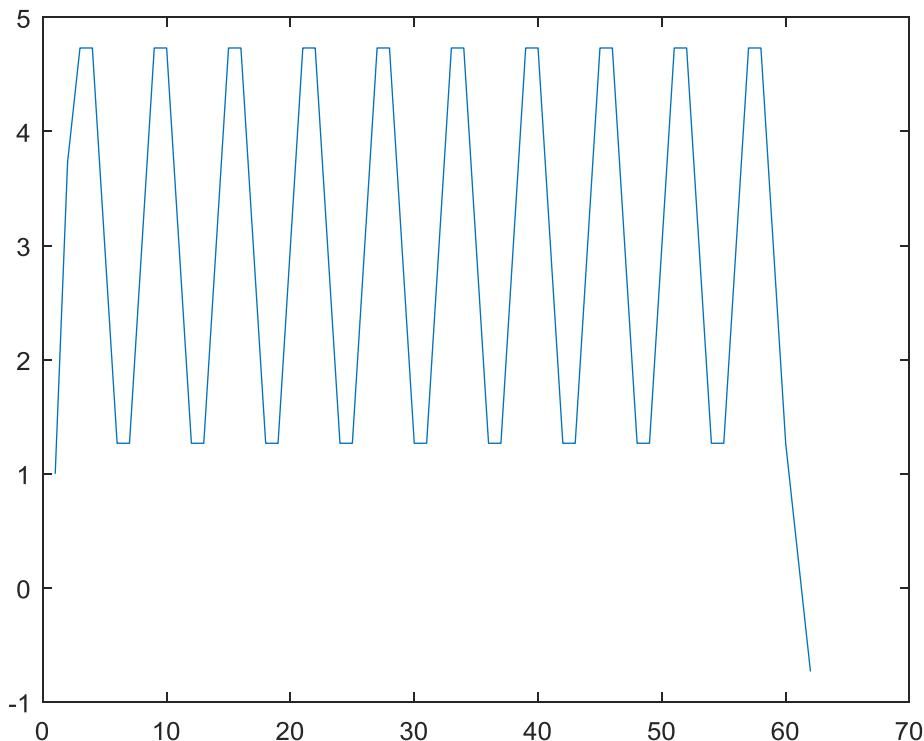
f)

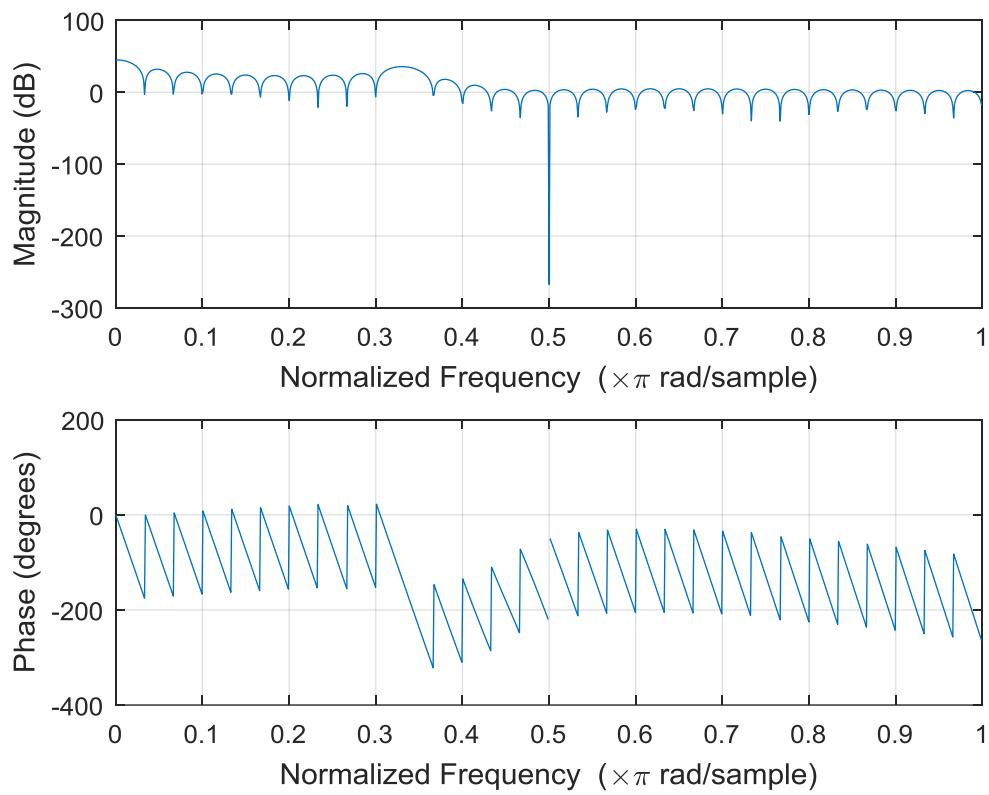




We observe peaks in the magnitude graph at the 0 , $\pi/3$, and $2\pi/3$ frequencies which are the frequencies of the three sinusoidal components of $z[n]$.

g)





Since the filter has a zero at $2\pi/3$ frequency, the correspondind sinusoidal component of $z[n]$ is eliminated.