## FREQUENCY DOMAIN REPRESENTATION OF LTI SYSTEMS

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## DICRETE TIME FOURIER TRANSFORM (DTFT)

The Fourier transform of a sequence x[n] is defined as

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

If the FT exists (summation converges) the sequence can be obtained from its FT as

$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

Fourier Transform is periodic with  $2\pi$ .

# LTI SYSTEMS

The frequency response function,  $H\!\left(e^{j\omega}\right)$ 

is the FT of the impulse response

h[n]

#### **EXISTENCE**

FT of a sequence x[n] exists, i.e.,

$$\sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$

converges to a continuous function of  $\omega$ ,

if x[n] is absolutely summable.

(sufficient condition)

**Proof: Exercise** 

$$\sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n} = \sum_{n=-\infty}^{\infty} x[n](\cos(\omega n) - j\sin(\omega n))$$
$$= \sum_{n=-\infty}^{\infty} x[n]\cos(\omega n) - j\sum_{n=-\infty}^{\infty} x[n]\sin(\omega n)$$

Both sums have to converge

ightarrow All stable LTI systems have frequency response functions.

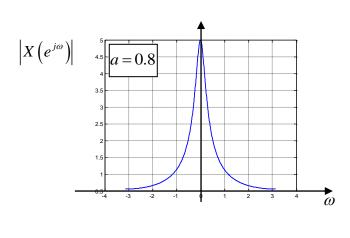
$$\underline{\mathbf{Ex}} : x[n] = a^n \ u[n]$$

$$X\left(e^{j\omega}\right) = \sum_{n=0}^{\infty} a^n e^{-j\omega n} = \sum_{n=0}^{\infty} \left(ae^{-j\omega}\right)^n = \frac{1}{1 - ae^{-j\omega}} \quad \text{if} \quad \left|ae^{-j\omega}\right| < 1 \quad \text{or} \quad \left|a\right| < 1$$

$$\left|X\left(e^{j\omega}\right)\right|^{2} = \frac{1}{\left|1 - ae^{-j\omega}\right|^{2}}$$

$$= \frac{1}{\left|1 - a\cos\omega + j\sin\omega\right|^{2}}$$

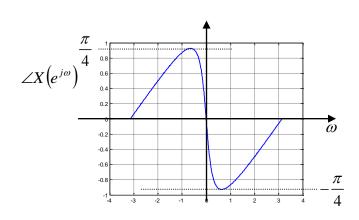
$$= \frac{1}{1 + a^{2} - 2a\cos\omega}$$



$$\angle X(e^{j\omega}) = \angle 1 - \angle (1 - ae^{-j\omega})$$

$$= 0 - \angle (1 - a\cos\omega + ja\sin\omega)$$

$$= -\tan^{-1} \left(\frac{a\sin\omega}{1 - a\cos\omega}\right)$$



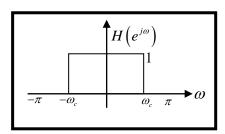
## MEAN SQUARE CONVERGENCE

Some sequences, which are not absolutely summable but square summable

$$\sum_{n=-\infty}^{\infty} \left| x[n] \right|^2 < \infty$$

can still be represented by Fourier Transform, but...

Ex: Ideal lowpass filter.



$$H\left(e^{j\omega}\right) = \begin{cases} 1 & |\omega| < \omega_c \\ 0 & \omega_c < |\omega| \le \pi \end{cases}$$

Let's find h[n]!

$$h[n] = \frac{1}{2\pi} \int_{-\omega_c}^{\omega_c} e^{j\omega n} d\omega$$
$$= \frac{1}{j2\pi n} \left( e^{j\omega_c n} - e^{-j\omega_c n} \right)$$
$$= \frac{\sin(\omega_c n)}{\pi n}$$

Note that,

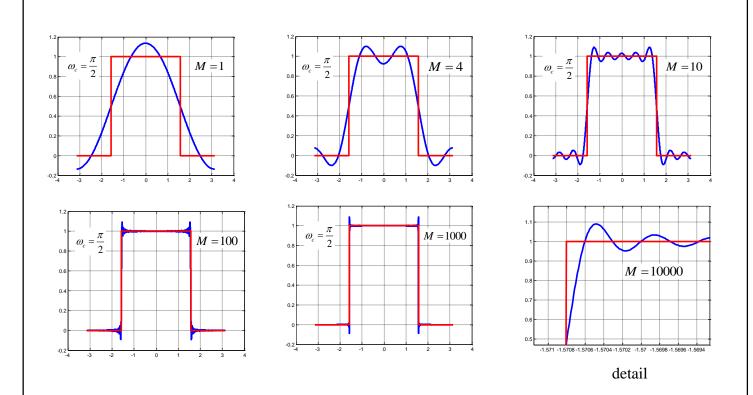
$$h[n] = \frac{\sin(\omega_c n)}{\pi n}$$

is not absolutely summable!

Then, one may question the Fourier transform of h[n],

$$\sum_{n=-\infty}^{\infty} \frac{\sin(\omega_c n)}{\pi n} e^{-j\omega n} = ?$$

# Define $H_M = \sum_{n=-M}^{M} \frac{\sin(\omega_c n)}{\pi n} e^{-j\omega n}$



Even if you take  $M \to \infty$  oscillations do not die to zero.

However  $\lim_{M\to\infty}\int\limits_{-\pi}^{\pi}\left|H\left(e^{j\omega}\right)-H_{M}\left(e^{j\omega}\right)\right|^{2}d\omega=0$ . This is called "mean square" convergence.

The oscillatory behavior around  $\omega = \omega_c$  is called the Gibbs phenomenon.

#### MATLAB code

```
clear all; close all;
precision =0.0001
w = [-pi:precision:pi];
ideaL = zeros(1,length(w));
wc = pi/2;
orta = round(length(w)/2);
ideaL((orta-round(wc/precision)):(orta+round(wc/precision)))=1;
M = 10000;
H = 0;
for n = -M:-1
  H = H+(\sin(wc^*n)/(pi^*n))^*\exp(-i^*w^*n);
end
for n = 1:M
  H = H + (\sin(wc*n)/(pi*n))*\exp(-i*w*n);
  H = H+(wc/pi);
plot(w,H); hold on;
plot(w,idea,'r')
grid
```

## FOURIER TRANSFORM OF A CONSTANT SEQUENCE

$$x[n] = 1 \qquad \Longleftrightarrow \qquad X\left(e^{j\omega}\right) = 2\pi \sum_{r=-\infty}^{\infty} \delta\left(\omega + 2\pi r\right)$$
 not absolutely summable 
$$\underbrace{ \left(e^{j\omega}\right)}_{-4\pi} = 2\pi \sum_{r=-\infty}^{\infty} \delta\left(\omega + 2\pi r\right)$$

or we can write as

$$X(e^{j\omega}) = 2\pi \delta(\omega)$$
  $0 \le \omega < 2\pi$ 

keeping in mind that FT is periodic with  $2\pi$ .

#### FOURIER TRANSFORM OF A COMPLEX EXPONENTIAL SEQUENCE

$$x[n] = e^{j\omega_0 n} \qquad \longleftrightarrow \qquad X\left(e^{j\omega}\right) = 2\pi \sum_{r=-\infty}^{\infty} \delta\left(\omega - \omega_0 + 2\pi r\right) \qquad \underbrace{\uparrow}_{\omega_0 + 4\pi} \underbrace{\uparrow}_{\omega_0 - 2\pi} \underbrace{\uparrow}_{\omega_0 - \omega_0 + 2\pi} \underbrace{\uparrow}_{\omega_0 + 4\pi} \underbrace{\uparrow}_{\omega_0 + 2\pi} \underbrace{\uparrow}_{\omega_0 + 4\pi} \underbrace{\downarrow}_{\omega_0 + 2\pi} \underbrace{\uparrow}_{\omega_0 + 4\pi} \underbrace{\downarrow}_{\omega_0 + 2\pi} \underbrace{\uparrow}_{\omega_0 + 4\pi} \underbrace{\downarrow}_{\omega_0 + 2\pi} \underbrace{\downarrow}_{\omega_0$$

or we can write as

$$X(e^{j\omega}) = 2\pi\delta(\omega - \omega_0)$$
  $0 \le \omega < 2\pi$ 

keeping in mind that FT is periodic with  $2\pi\,$ 

#### FOURIER TRANSFORM OF A SINUSOIDAL SEQUENCE

$$x[n] = \cos(\omega_0 n) = \frac{1}{2} \left( e^{j\omega_0 n} + e^{-j\omega_0 n} \right) \quad \Longleftrightarrow \quad X(e^{j\omega}) = \pi \left( \sum_{r=-\infty}^{\infty} \delta(\omega + \omega_0 + 2r\pi) + \delta(\omega - \omega_0 + 2r\pi) \right)$$
not absolutely summable

or  $X(e^{j\omega}) = \pi(\delta(\omega + \omega_0) + \delta(\omega - \omega_0))$   $0 \le \omega < 2\pi$  since FT is periodic with  $2\pi$ 

# FOURIER TRANSFORM OF UNIT STEP SEQUENCE

$$x[n] = u[n]$$
  $\longleftrightarrow$   $X(e^{j\omega}) = \frac{1}{1 - e^{-j\omega}} + \pi \sum_{r = -\infty}^{\infty} \delta(\omega + 2\pi r)$ 

not absolutely summable

#### SYMMETRY PROPERTIES OF FOURIER TRANSFORM

**Definitions:** 

Conjugate symmetric (CS) sequence.

$$x[n] = x^*[-n]$$

Conjugate antisymmetric (CaS) sequence.

$$x[n] = -x^*[-n]$$

Using the above definitions, any sequence can be written as

$$x[n] = x_e[n] + x_o[n]$$

where

$$x_e[n] = \frac{1}{2}(x[n] + x^*[-n])$$
 is the CS part

and

$$x_o[n] = \frac{1}{2}(x[n] - x^*[-n])$$
 is the CaS part.

#### SYMMETRY PROPERTIES

## Fundamental relations

Let  $x[n] \leftrightarrow X(e^{j\omega})$  be a FT pair. Then, the following hold:

$$x^*[n] \longleftrightarrow X^*(e^{-j\omega})$$

$$x^*[n] \leftrightarrow X^*(e^{-j\omega})$$
 since  $X^*(e^{-j\omega}) = \sum_{n=-\infty}^{\infty} x^*[n]e^{-j\omega n}$ 

$$x[-n] \leftrightarrow X(e^{-j\omega})$$

since 
$$X(e^{-j\omega}) = \sum_{n=-\infty}^{\infty} x[-n]e^{-j\omega n}$$

Above yields

$$x^*[-n] \longleftrightarrow X^*(e^{j\omega})$$

The two relations above also yield:

1) 
$$\operatorname{Re}\left\{x[n]\right\} = \frac{x[n] + x^*[n]}{2} \iff X_e\left(e^{j\omega}\right) = \frac{X\left(e^{j\omega}\right) + X^*\left(e^{-j\omega}\right)}{2}$$
 (CS part of  $X\left(e^{j\omega}\right)$ )

2) 
$$j\operatorname{Im}[x[n]] = \frac{x[n] - x^*[n]}{2} \leftrightarrow X_o(e^{j\omega}) = \frac{X(e^{j\omega}) - X^*(e^{-j\omega})}{2}$$
 (CaS part of  $X(e^{j\omega})$ )

3) 
$$x_e[n] = \frac{x[n] + x^*[-n]}{2} \iff \text{Re}\left\{X\left(e^{j\omega}\right)\right\} = \frac{X\left(e^{j\omega}\right) + X^*\left(e^{j\omega}\right)}{2}$$
 (real part of  $X\left(e^{j\omega}\right)$ )

Therefore FT of an even seq. is real!

4) 
$$x_o[n] = \frac{x[n] - x^*[-n]}{2} \iff j \operatorname{Im}\left\{X\left(e^{j\omega}\right)\right\} = \frac{X\left(e^{j\omega}\right) - X^*\left(e^{j\omega}\right)}{2}$$
 (imag. part of  $X\left(e^{j\omega}\right)$ )

Therefore FT of an odd seq. is purely imaginary!

**Ex:** Let a[n] and b[n] be two real sequences with their DTFTs  $A(e^{j\omega})$  and

 $B(e^{j\omega})$ , respectively.

Let

$$x[n] = a[n] + jb[n]$$

Then,

$$X(e^{j\omega}) = A(e^{j\omega}) + jB(e^{j\omega})$$

Note that  $A(e^{j\omega})$  is NOT the real part of  $X(e^{j\omega})$ .

However,

$$A(e^{j\omega}) = \frac{X(e^{j\omega}) + X^*(e^{-j\omega})}{2}$$

Since

$$X^*(e^{-j\omega}) = \underbrace{A^*(e^{-j\omega})}_{A(e^{j\omega})} - j\underbrace{B^*(e^{-j\omega})}_{B(e^{j\omega})}$$

Similarly,

$$jB(e^{j\omega}) = \frac{X(e^{j\omega}) - X^*(e^{-j\omega})}{2}$$

Ex: (cont'd)

a[n]:  $[-1 \ 1]$ 

b[n]: [1 1]

 $x[n]: [-1+j \ 1+j]$ 

$$A(e^{j\omega}) = -1 + e^{-j\omega}$$

$$B\!\left(e^{j\omega}\right)=1+e^{-j\omega}$$

$$X(e^{j\omega}) = -1 + e^{-j\omega} + j + je^{-j\omega}$$

$$X^*(e^{-j\omega}) = -1 + e^{-j\omega} - j - je^{-j\omega}$$

$$\frac{X(e^{j\omega}) + X^*(e^{-j\omega})}{2} = -1 + e^{-j\omega}$$

$$\frac{X(e^{j\omega}) - X^*(e^{-j\omega})}{2} = j + je^{-j\omega}$$

#### **REAL SEQUENCES**

Based on the above relations, for real sequences ( $x[n] = x^*[n]$ ):

 $X(e^{j\omega}) = X^*(e^{-j\omega})$ , conjugate symmetry

which implies

Magnitude is even.....  $|X(e^{j\omega})| = |X(e^{-j\omega})|$ 

## Verification by an example

$$\underline{\mathbf{Ex}} \colon x[n] = a^n \ u[n] \quad \Longleftrightarrow \quad X\left(e^{j\omega}\right) = \frac{1}{1 - ae^{-j\omega}} \qquad |a| < 1$$

$$0 < a \le 1$$

$$\cdots$$

$$-1 \qquad 0 \qquad 1 \qquad 2$$

a) FT is conjugate symmetric:

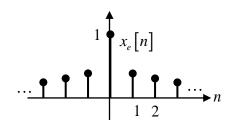
$$X(e^{j\omega}) = X^*(e^{-j\omega}) = \frac{1}{1 - ae^{-j\omega}} \qquad |a| < 1$$

b) Real part of FT is an even function:

$$\operatorname{Re}\left\{X\left(e^{j\omega}\right)\right\} = X_{R}\left(e^{j\omega}\right) = \frac{1 - a\cos\omega}{1 + a^{2} - 2a\cos\omega} = X_{R}\left(e^{-j\omega}\right)$$

c)  $X_R(e^{j\omega})$  is the FT of  $x_e[n]$ :

$$X_{R}\left(e^{j\omega}\right) = \frac{1 - a\cos\omega}{1 + a^{2} - 2a\cos\omega}$$
$$X_{e}\left[n\right] = \frac{1}{2}\left(a^{n} u\left[n\right] + a^{-n} u\left[-n\right]\right)$$



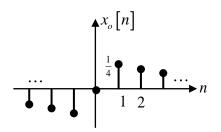
d) Imaginary part of Ft is an odd function.

$$\operatorname{Im}\left\{X\left(e^{j\omega}\right)\right\} = X_{I}\left(e^{j\omega}\right) = \frac{-a\sin\omega}{1 + a^{2} - 2a\cos\omega} = -X_{I}\left(e^{-j\omega}\right)$$

e)  $X_{I}(e^{j\omega})$  is the FT of  $x_{o}[n]$ :

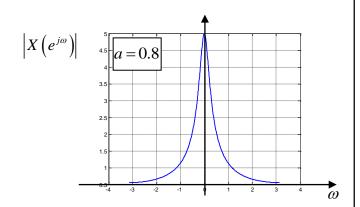
$$X_{I}\left(e^{j\omega}\right) = \frac{-a\sin\omega}{1 + a^{2} - 2a\cos\omega}$$

$$x_o[n] = \frac{1}{2}(a^n u[n] - a^{-n} u[-n])$$



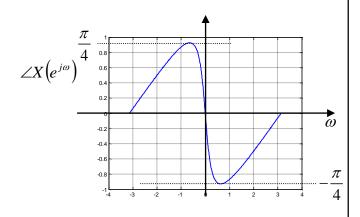
f) Magnitude of FT is an even function:

$$\left| X\left(e^{j\omega}\right) \right| = \frac{1}{\sqrt{1+a^2 - 2a\cos\omega}} = \left| X\left(e^{-j\omega}\right) \right|$$



g) Phase of FT is an odd function

$$\angle X(e^{j\omega}) = -\tan^{-1}\left(\frac{a\sin\omega}{1-a\cos\omega}\right) = -\angle X(e^{-j\omega})$$



#### FOURIER TRANSFORM THEOREMS

$$x[n] \stackrel{F}{\longleftrightarrow} X(e^{j\omega}) \qquad \left(x[n] = F\left\{X(e^{j\omega})\right\}, \quad X(e^{j\omega}) = F^{-1}\left\{x[n]\right\}\right)$$

1) 
$$ax[n] + by[n] \qquad \stackrel{F}{\leftrightarrow} \qquad aX(e^{j\omega}) + bY(e^{j\omega})$$
 Linearity

2) 
$$x[n-n_0] \longleftrightarrow e^{-j\omega n_0} X(e^{j\omega})$$
 Time-shift

3) 
$$e^{j\omega_0 n}x[n]$$
  $\stackrel{F}{\longleftrightarrow}$   $X\left(e^{j(\omega-\omega_0)}\right)$  Freq. shift

4) 
$$x[-n] \longleftrightarrow X(e^{-j\omega})$$
 Time reversal

5) 
$$nx[n] \leftrightarrow j\frac{dX(e^{j\omega})}{d\omega}$$
 Differentiation in frequency domain

6) 
$$x[n] * y[n] \qquad \stackrel{F}{\leftrightarrow} \qquad X(e^{j\omega})Y(e^{j\omega})$$
 Convolution

7) 
$$y[n] = x[n]w[n] \leftrightarrow Y(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\phi})W(e^{j\omega-\phi}) d\phi$$
 Modulation, windowing

Parseval's theorem (prove as an exercise)

8) 
$$\sum_{n=-\infty}^{\infty} |x[n]|^2 = \frac{1}{2\pi} \int_{0}^{\pi} |X(e^{j\omega})|^2 d\omega$$
 energy of the signal

 $\left|X\left(e^{j\omega}\right)\right|^2$  is called the "energy density spectrum".

9) 
$$\sum_{n=-\infty}^{\infty} x[n] y^*[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) Y^*(e^{j\omega}) d\omega$$

Proof of (6):

$$w[n] \stackrel{\text{def}}{=} \sum_{k=-\infty}^{\infty} x[k]y[n-k] = x[n] * y[n]$$

$$W(e^{j\omega}) = ?$$

$$x(e^{j\omega}) - \sum_{k=-\infty}^{\infty} \left(\sum_{k=-\infty}^{\infty} x[k]y[n-k]\right) e^{-j\omega n}$$

$$W(e^{j\omega}) = \sum_{n=-\infty}^{\infty} \left(\sum_{k=-\infty}^{\infty} x[k]y[n-k]\right) e^{-j\omega n}$$

$$= \sum_{k=-\infty}^{\infty} x[k] \left(\sum_{n=-\infty}^{\infty} y[n-k]e^{-j\omega n}\right)$$

$$= \sum_{k=-\infty}^{\infty} x[k]e^{-j\omega k} \left(\sum_{m=-\infty}^{\infty} y[m]e^{-j\omega m}\right)$$

$$= X(e^{j\omega})Y(e^{j\omega})$$

Proof of (8) using (6):

Let

$$w[n] \stackrel{\text{def}}{=} \sum_{k=-\infty}^{\infty} x[k]x^*[k-n] = x[n] * x^*[-n]$$

$$W(e^{j\omega}) = X(e^{j\omega})X^*(e^{j\omega}) = |X(e^{j\omega})|^2$$

$$w[0] = \sum_{k=-\infty}^{\infty} x[k]x^*[k] = \sum_{k=-\infty}^{\infty} |x[k]|^2$$

$$w[0] = \frac{1}{2\pi} \int_{-\pi}^{\pi} W(e^{j\omega}) d\omega = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X(e^{j\omega})|^2 d\omega$$

## Proof of (9):

$$\sum_{n=-\infty}^{\infty} x[n]y^*[n]$$

$$= \sum_{n=-\infty}^{\infty} \left(\frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega})e^{j\omega n} d\omega\right) y^*[n]$$

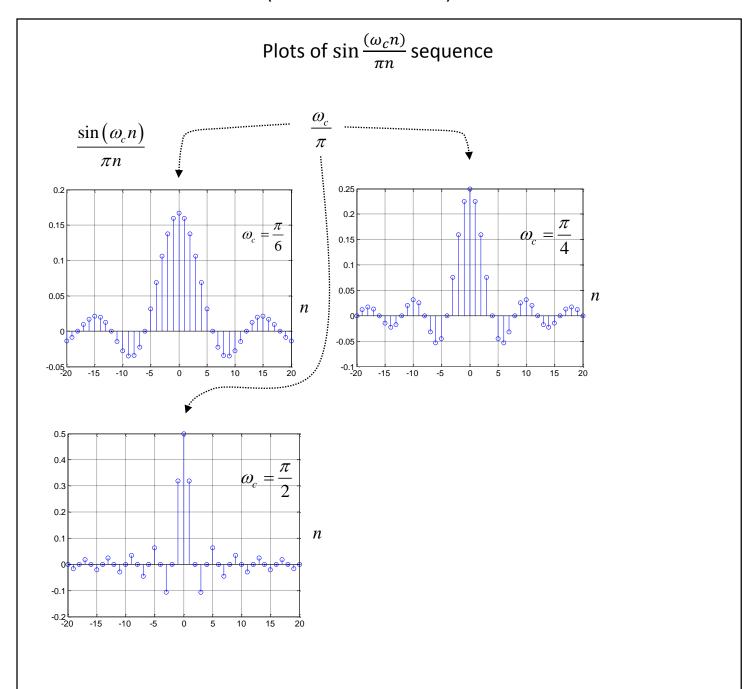
$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) \left(\sum_{n=-\infty}^{\infty} y^*[n]e^{j\omega n}\right) d\omega$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) Y^*(e^{j\omega}) d\omega$$

## FOURIER TRANSFORM PAIRS

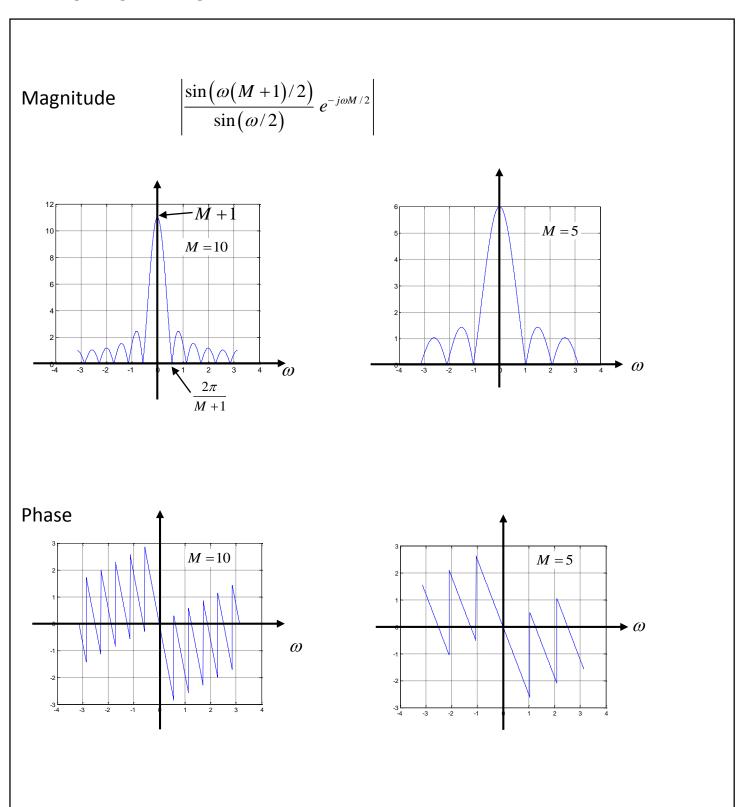
$\delta[n]$	1
$\frac{\delta[n]}{\delta[n-n_0]}$	$e^{-j\omega n_0}$
$\begin{array}{ c c c c c c c c c c c c c c c c c c c$	$X\left(e^{j\omega}\right) = 2\pi \sum_{n=-\infty}^{\infty} \delta(\omega + 2\pi r)$
$x[n] = a^n \ u[n] \qquad  a  < 1$	$\frac{1}{1-ae^{-j\omega}}$
x[n] = u[n]	$\frac{1}{1-e^{-j\omega}} + \sum_{k=-\infty}^{\infty} \pi \ \delta(\omega + 2\pi k)$
$na^n u[n]$ $ a  < 1$	$\frac{ae^{-j\omega}}{\left(1-ae^{-j\omega}\right)^2}$
$ \begin{array}{c c} (n+1)a^n & u[n] &  a  < 1 \\ = (a^n & u[n]) * (a^n & u[n]) \end{array} $	$\frac{1}{\left(1-ae^{-j\omega}\right)^2}$
$\frac{(n+2)(n+1)a^n u[n]}{2}$	$\frac{1}{\left(1-ae^{-j\omega}\right)^3}$
$\frac{1}{(k-1)!} \frac{(n+k-1)!}{n!} a^n u[n]$	$\frac{1}{\left(1-ae^{-j\omega}\right)^k}$
$\frac{1}{(k-1)!}(n+k-1)(n+k-2)(n+1) a^n u[n]$	
$\left  \frac{1}{\sin \omega_p} r^n \sin \left( \omega_p (n+1) \right) u[n] \right   r  < 1$	$\frac{1}{1-2r\cos(\omega_p)e^{-j\omega}+r^2e^{-j2\omega}}$ show using
	$x[n] = a^n \ u[n]$
$\frac{\sin(\omega_c n)}{\pi n}$	$X\left(e^{j\omega}\right) = \begin{cases} 1, &  \omega  < \omega_c \\ 0, & \omega_c <  \omega  \le \pi \end{cases}$
$x[n] = \begin{cases} 1, & 0 \le n \le M \\ 0, & \text{otherwise} \end{cases}$	$\frac{\sin(\omega(M+1)/2)}{\sin(\omega/2)}e^{-j\omega M/2}$
`	` /
$e^{ja_0n}$	$2\pi\sum_{r=-\infty}^{\infty}\delta(\omega-\omega_0+r2\pi)$
$\frac{e^{j\omega_0 n}}{\cos(\omega_0 n + \phi)}$	m

# Ex: IDEAL LOWPASS FILTER (IMPULSE RESPONSE)



They are infinitely long sequences in  $-\infty < n < \infty$  Plots are arbitrarily in -20 < n < 20

## **Ex**: MOVING AVERAGE FILTER

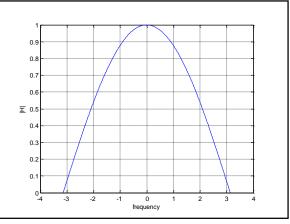


DTFT functions are plotted in  $-\pi \le \omega \le \pi$  or in  $0 \le \omega \le 2\pi$ 

$$\mathbf{M} = \mathbf{1}$$

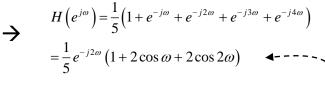
$$h[n] = \frac{1}{2} \left( \delta[n] + \delta[n-1] \right) \rightarrow H\left(e^{j\omega}\right) = \frac{1}{2} \left(1 + e^{-j\omega}\right) = e^{-j\frac{\omega}{2}} \cos\frac{\omega}{2}$$

$$y[n] = \frac{1}{2} \left(x[n] + x[n-1]\right)$$



M = 4

$$y[n] = \frac{1}{5} (x[n] + x[n-1] + x[n-2] + x[n-3] + x[n-4])$$
$$h[n] = \frac{1}{5} (\delta[n] + \delta[n-1] + \delta[n-2] + \delta[n-3] + \delta[n-4])$$

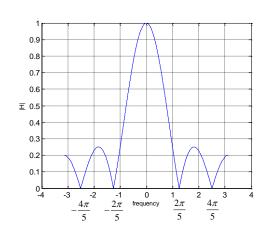


or

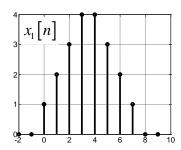
$$H(e^{j\omega}) = \frac{1}{5} \left( 1 + e^{-j\omega} + e^{-j2\omega} + e^{-j3\omega} + e^{-j4\omega} \right)$$

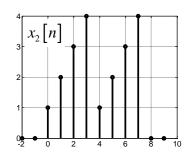
$$= \frac{1}{5} \frac{1 - e^{-j5\omega}}{1 - e^{-j\omega}} = \frac{1}{5} \frac{e^{-j\frac{5}{2}\omega} \left( e^{j\frac{5}{2}\omega} - e^{-j\frac{5}{2}\omega} \right)}{e^{-j\frac{\omega}{2}} \left( e^{j\frac{\omega}{2}} - e^{-j\frac{\omega}{2}} \right)}$$

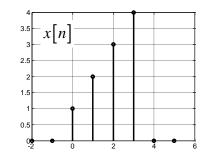
$$= \frac{1}{5} e^{-j2\omega} \frac{\sin\left(\frac{5}{2}\omega\right)}{\sin\left(\frac{\omega}{2}\right)}$$



**Ex**: Express  $X_1(e^{j\omega})$ ,  $X_2(e^{j\omega})$ , in terms of  $X(e^{j\omega})$ , the DTFT of x[n].







$$x_1[n] = x[n] + x[-(n-7)] = x[n] + x[-n+7] \quad \Rightarrow \quad X_1(e^{j\omega}) = X(e^{j\omega}) + e^{-j7\omega}X(e^{-j\omega})$$

$$\rightarrow X_1(e^{j\omega}) = X(e^{j\omega}) + e^{-j7\omega}X(e^{-j\omega})$$

$$x_2[n] = x[n] + x[n-4]$$

$$\rightarrow$$
  $X_2(e^{j\omega}) = X(e^{j\omega}) + e^{-j4\omega}X(e^{j\omega})$ 

One can also write as

$$X_{1}(e^{j\omega}) = X(e^{j\omega}) + e^{-j7\omega}X(e^{-j\omega})$$

$$= X(e^{j\omega}) + e^{-j7\omega}X^{*}(e^{j\omega}) \quad \text{since } x[n] \text{ is real}$$

$$= |X(e^{j\omega})| \left( e^{j \angle X(e^{j\omega})} + e^{-j7\omega}e^{-j \angle X(e^{j\omega})} \right)$$

$$= |X(e^{j\omega})| e^{-j\frac{7}{2}\omega} \left( e^{j\frac{7}{2}\omega}e^{j \angle X(e^{j\omega})} + e^{-j\frac{7}{2}\omega}e^{-j \angle X(e^{j\omega})} \right)$$

$$= 2 \operatorname{Re} \left\{ e^{j\left( \angle X(e^{j\omega}) + \frac{7}{2}\omega \right)} \right\} |X(e^{j\omega})| e^{-j\frac{7}{2}\omega}$$

$$X_{2}(e^{j\omega}) = X(e^{j\omega}) (1 + e^{-j4\omega})$$

$$= X(e^{j\omega})e^{-j2\omega} (e^{j2\omega} + e^{-j2\omega})$$

$$= X(e^{j\omega})e^{-j2\omega} 2 \cos(2\omega)$$

$$= 2 \cos(2\omega) |X(e^{j\omega})| e^{-j2\omega}e^{j \angle X(e^{j\omega})}$$

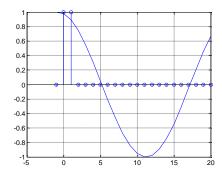
**Ex**: What is the inverse DTFT, y[n], of  $Y(e^{j\omega}) = \frac{2e^{-j3\omega}}{\left(1-\frac{1}{8}e^{-j\omega}\right)^2}$ ?

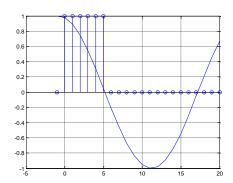
From the table  $(n+1)a^n u[n] \longleftrightarrow \frac{1}{(1-ae^{-j\omega})^2}$  for |a| < 1

$$2(n+1)\frac{1}{8}^{n} u[n] \qquad \Longleftrightarrow \qquad \frac{2}{\left(1-\frac{1}{8}e^{-j\omega}\right)^{2}}$$

$$2(n-2)\frac{1}{8}^{(n-3)}u[n-3] \qquad \Longleftrightarrow \qquad \frac{2e^{-j3\omega}}{\left(1-\frac{1}{8}e^{-j\omega}\right)^2}$$

Why does the high frequency gain of MA filter "decrease" as M increases?





Comment on the above illustrations.

#### LCCDES AND FREQUENCY RESPONSE

$$y[n] = x[n] * h[n] \stackrel{FT}{\longleftrightarrow} Y(e^{j\omega}) = X(e^{j\omega})H(e^{j\omega})$$

$$\Rightarrow H(e^{j\omega}) = \frac{Y(e^{j\omega})}{X(e^{j\omega})}$$

$$N$$

$$\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k] \stackrel{FT}{\longleftrightarrow} \sum_{k=0}^{N} a_k e^{-jk\omega} Y(e^{j\omega})$$
$$= \sum_{k=0}^{M} b_k e^{-jk\omega} X(e^{j\omega})$$

$$\Rightarrow Y(e^{j\omega})\sum_{k=0}^N a_k e^{-jk\omega} = X(e^{j\omega})\sum_{k=0}^M b_k e^{-jk\omega}$$

$$\Rightarrow \frac{Y(e^{j\omega})}{X(e^{j\omega})} = H(e^{j\omega}) = \frac{\sum_{k=0}^{M} b_k e^{-jk\omega}}{\sum_{k=0}^{N} a_k e^{-jk\omega}}$$

$$H(e^{j\omega}) = \frac{b_0 + b_1 e^{-j\omega} + b_2 e^{-j2\omega} + \dots + b_M e^{-jM\omega}}{a_0 + a e^{-j\omega} + a_2 e^{-j2\omega} + \dots + a_N e^{-jN\omega}}$$