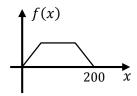
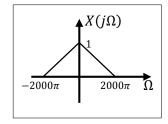
EE 430 Digital Signal Processing - 2014 Fall

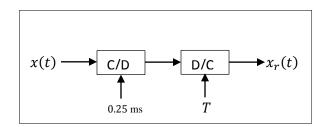
Submit for problems 1,2,3,5,7,8

1) A function f(x) is shown in the figure. Plot $g(x) = A \sum_{k=-\infty}^{\infty} f(Ax - k8A)$ for A = 50 and A = 20. Find the condition on A so that f(Ax - k8A), $k \in \mathbb{Z}$, do not overlap.



- **2)** x(t) is a complex valued signal whose CTFT is zero for $\Omega \le -100\pi$ and $\Omega \ge 1000\pi$. Find the minimum sampling frequency in Hz so that x(t) can be recovered from its samples, x[n].
- **3)** x(t) has a hypothetical CTFT, $X(j\Omega)$, as shown in the figure. x(t) is applied to the system shown below.





Find and plot $X_r(j\Omega)$ for

- a) T = 0.75 ms
- b) T = 0.05 ms
- **4)** A discrete-time signal, x[n], is obtained by sampling a continuous-time signal, $x_c(t)$, bandlimited to $\frac{\pi}{T}$. Sampling period is T. x[n] is applied to a causal LTI system whose impulse response is

$$h_D[n] = \begin{cases} \frac{\sin\left(\frac{\pi}{T}\left(\frac{T}{5} + (n-D)T\right)\right)}{\frac{\pi}{T}\left(\frac{T}{5} + (n-D)T\right)} & n = 0,1,...,2D \end{cases}$$

$$0 & o.w.$$

Discuss the relationship of the output, y[n], of this system to $x_c(t)$. May your discussion have any practical implications?

5) It is desired to have an ideal bandpass filter with the system below.

$$x(t) \xrightarrow{C/D} \xrightarrow{H(e^{j\omega})} \xrightarrow{D/C} x_r(t)$$

Specifically, it is desired to have

$$H_{eff}(j\Omega) = \frac{X_r(j\Omega)}{X(j\Omega)} = \begin{cases} 1 & 9000\pi \le |\Omega| \le 10000\pi \\ o.w. \end{cases}$$

It is known that $H(e^{j\omega})$ is a highpass filter.

- a) Find T, specify $H(e^{j\omega})$; are any constraints/conditions required to make the system act as the specified bandpass filter?
- b) Answer part (a) to have

$$H_{eff}(j\Omega) = \frac{X_r(j\Omega)}{X(j\Omega)} = \begin{cases} 1 & 900\pi \le |\Omega| \le 1000\pi \\ 0 & o.w. \end{cases}$$

6) Consider the following system. x(t) is bandlimited to $\frac{\pi}{r}$. Black box contains upsamplers, downsaplers, ideal lowpass filters and delay elements.

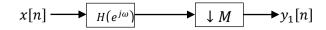
$$x(t) \longrightarrow C/D \longrightarrow y[n]$$

- $x(t) \xrightarrow{\text{C/D}} y[n]$ a) It is desired to have $y[n] = x\left(t-\frac{7}{13}\right)_{t=n\frac{T}{3}}$. Design the system in the black box.
- b) How would you answer if it is desired to have $y[n] = x \left(t \frac{7}{13}\right)_{t=n/3}$?
- 7) Consider the following two systems:

$$x[n] \longrightarrow \underbrace{\downarrow 5} \underbrace{z_2[n]} \longrightarrow \underbrace{\uparrow 2} \longrightarrow y_2[n]$$

- a) Express $Z_1(e^{j\omega})$ and $Y_1(e^{j\omega})$ in terms of $X(e^{j\omega})$.
- b) Express $Z_2(e^{j\omega})$ and $Y_2(e^{j\omega})$ in terms of $X(e^{j\omega})$.
- c) Are $y_1[n]$ and $y_2[n]$ the same? Prove your claim.
- d) Does your answer to part-(c) change if the downsampling factor is 4?

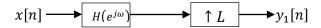
8) Consider the two systems:



$$x[n] \longrightarrow A \longrightarrow G(e^{j\omega}) \longrightarrow y_2[n]$$

Assuming that x[n] is bandlimited to $\frac{\pi}{M}$, can you find a relationship between $H(e^{j\omega})$ and $G(e^{j\omega})$ so that the two systems are equivalent, i.e., $y_1[n] = y_2[n]$? Prove your claim.

9) Consider the two systems:



$$x[n] \longrightarrow f$$
 $L \longrightarrow G(e^{j\omega}) \longrightarrow y_2[n]$

Can you find a relationship between $H(e^{j\omega})$ and $G(e^{j\omega})$ so that the two systems are equivalent, i.e., $y_1[n]=y_2[n]$? Prove your claim.

10) (MATLAB) The distortion resulting from the sampling and reconstruction of a finite duration sinusoidal signal will be analyzed. Let x[n] be obtained by sampling $x_c(t) = \cos(2000\pi t)$ at $f_s = 3$ kHz. The sampling duration is 5 msec.

$$x[n] = \begin{cases} \cos\left(\frac{2\pi}{3}n\right) & n = 0,1,\dots,15 \\ 0 & o.w. \end{cases}$$

y(t) is obtained from x[n] by using an ideal C/D converter. Therefore

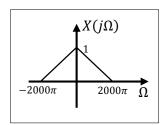
$$y(t) = \sum_{n=0}^{15} x[n] \frac{\sin\left(\frac{\pi(t-nT)}{T}\right)}{\frac{\pi(t-nT)}{T}}$$
$$= \sum_{n=0}^{15} \cos\left(\frac{2\pi}{3}n\right) \frac{\sin\left(3000\pi\left(t - \frac{n}{3000}\right)\right)}{3000\pi\left(t - \frac{n}{3000}\right)}$$

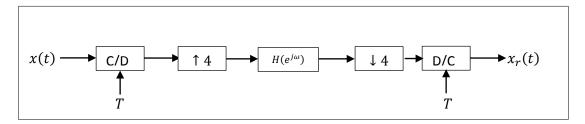
- a) Plot $x_c(t)$ and y(t) for -5 $msec \le t \le 10$ msec in the same figure.
- b) Plot the absolute error $e(t) = |x_c(t) y(t)|$ for -5 $msec \le t \le 10$ msec.
- c) How does the error behave in $0 \, msec \leq t \leq 5 \, msec$ interval? Zoom into this interval. Observe the peaky behavior. Where does the smallest peak of e(t) occur? What is this smallest peak value?
- d) Change sampling duration to 15 msec. Answer the questions in part-(c) using an observation interval of $-5 \, msec \le t \le 20 \, msec$. What is your conclusion?
- e) For the original sampling duration of 5 msec and observation interval of -5 msec $\leq t \leq 10$ msec, increase the sampling rate to $f_s = 3$ kHz. Answer the questions in part-(c). What is your conclusion?

You may use the code below:

```
% bandllimited interpolation error analysis
clear all
close all
DelT = 5;
                     % sampling duration of signal in msec
  = -5:0.001:10; % time interval in which we plot our results in msec
    = t * 0.001;
   = 1000;
                      % frequency of continuous time sinuoid to be sampled
Omega = 2 * pi * f ;
f_s = 3000
                        % sampling frequency
T_s = 1/f s
N = round(DelT*0.001/T s)
k
    = 0:N;
    = cos(Omega*k*T s); % samples of the continuous time sinuoid
% computing the sinc signals in the summation
for n = 0:N
    h r(n+1,:) = x(n+1) * sin(f s*pi*(t-n*T s))./(f s*pi*(t-n*T s));
    h r(n+1, find(isnan(h r(n+1,:)))) = x(n+1);
end
x_r = sum(h_r); % reconstructed cont-time signal
plot(t,cos(Omega*t),'k','LineWidth',3)
plot(t,x r,'r--','LineWidth',1.5)
figure
plot(t,abs(x_r - cos(Omega*t)),'r','LineWidth',1.5)
```

11) Consider the system below and the spectrum of its input signal:





- a) Is the discrete-time system including the upsampler, $H(e^{j\omega})$ and downsampler a LTI system? Hence, can one claim the existence of an $H_{eff}(j\Omega)$ for an arbitrary $H(e^{j\omega})$? Justify your answer.
- b) Assuming that T fulfills the Nyquist rate requirement, can one propose a constraint on $H(e^{j\omega})$ to have an $H_{eff}(j\Omega)$?
- c) Now, assume that $H(e^{j\omega}) = 0$, $\frac{\pi}{12} \le |\omega| \le \pi$. What is the set of T values for which one can find $H_{eff}(j\Omega)$?