

Introduction: Complementary Filters at KAGRA

- ▶ The pre-isolators of Type-A and Type-B suspensions in KAGRA are equipped with relative sensors and inertial sensors.
- ▶ A pair of complementary filters (low-pass and high-pass) can be used to combine the sensors into a virtual “super-sensor” that has superior noise properties.
- ▶ In this work, we discuss a scheme that uses \mathcal{H}_∞ methods to synthesize complementary filters that optimally blend the sensors according to the sensor noises.

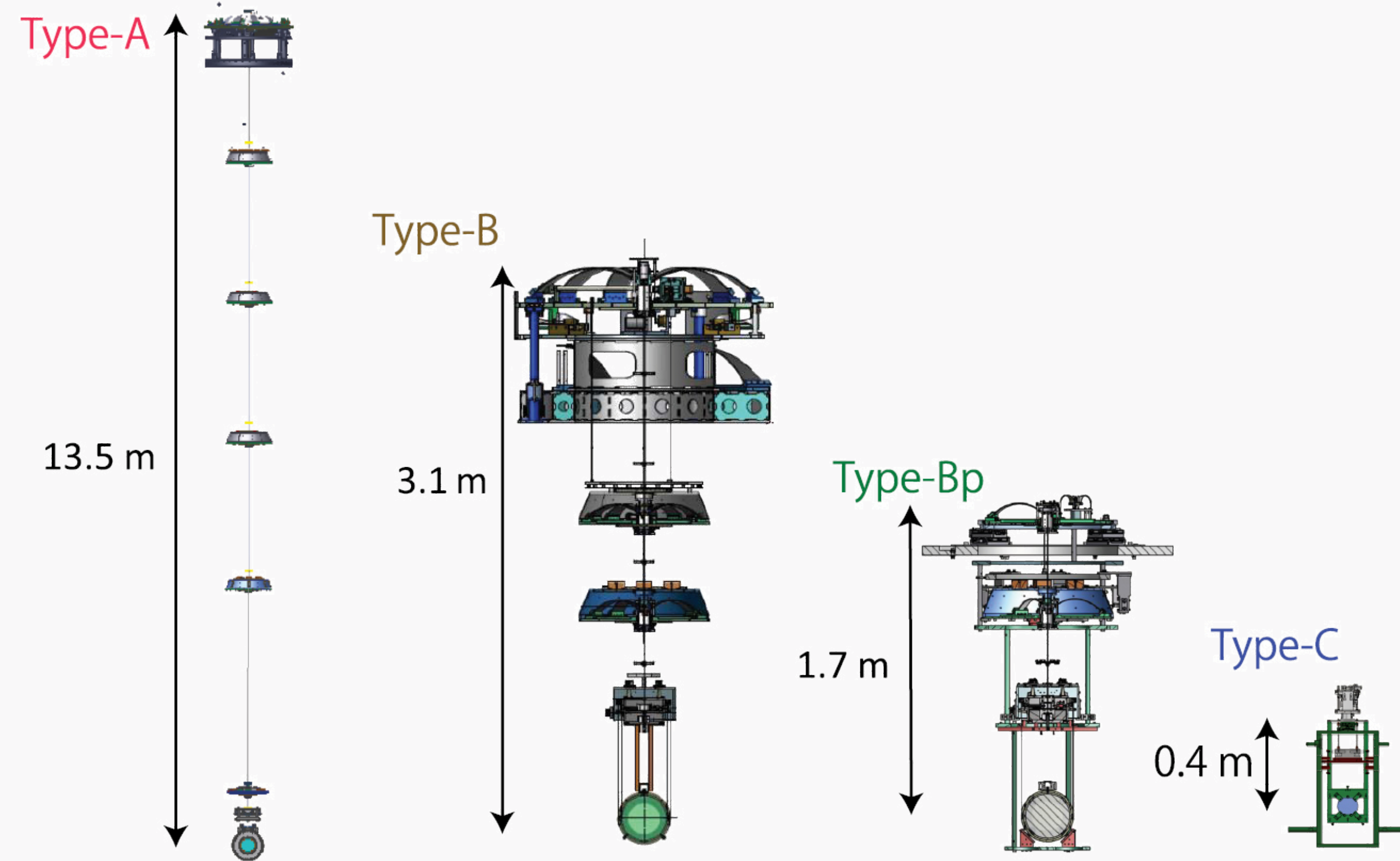


Figure 1: Type-A suspensions: input/end test masses, Type-B suspensions: beamsplitter and signal-recycling mirrors, Type-Bp suspensions: power-recycling mirrors, and Type-C suspensions: input/output mode cleaners [1].

Methodology: Complementary Filter Problem as an \mathcal{H}_∞ Problem

\mathcal{H}_∞ method in a nutshell:

1. As shown in Fig. 2, define signals w , z , u , and v , and hence,
2. derive a generalized plant $\mathbf{P}(s)$.
3. \mathcal{H}_∞ synthesis gives an \mathcal{H}_∞ optimal controller $\mathbf{K}_\infty(s)$ that minimizes the \mathcal{H}_∞ -norm of the closed-loop plant $\|\mathbf{G}(s; \mathbf{K}, \mathbf{P})\|_\infty$.

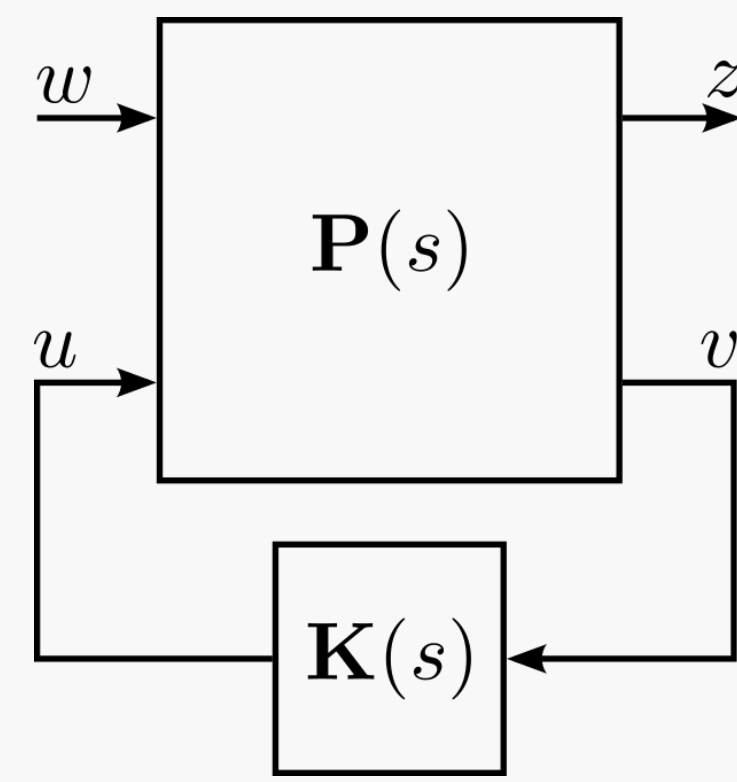


Figure 2: Generalized Plant Representation.

- ▶ The close-loop transfer function $\mathbf{G}(s)$ is defined such that $z = \mathbf{G}(s)w$ and $u = \mathbf{K}(s)v$, i.e.

$$\mathbf{G}(s) = P_{11}(s) + P_{12}(s)\mathbf{K}(s)[\mathbf{I} - P_{22}(s)\mathbf{K}(s)]^{-1}P_{21}(s). \quad (1)$$

- ▶ The \mathcal{H}_∞ -norm is defined as

$$\|\mathbf{G}(s)\|_\infty = \sup_{\omega} \bar{\sigma}(\mathbf{G}(j\omega)), \quad (2)$$

where $\bar{\sigma}$ denotes the maximum singular value and ω is the angular frequency.

Methodology: Complementary Filter Problem as an \mathcal{H}_∞ Problem

Formulating the complementary filter problem as an \mathcal{H}_∞ problem:

Complementary filter configuration:
 $N_1(s), N_2(s)$: sensor noises,
 $H_1(s), H_2(s)$: complementary filters,
 $N_{\text{super}}(s)$: super sensor noise.

Step 1:
Apply constraint $H_1(s) + H_2(s) = 1$.

Step 2:
Model the sensor noises with transfer functions $\hat{N}_1(s)$ and $\hat{N}_2(s)$.

Φ_1 and Φ_2 are white noise processes that have unit intensity.

Step 3:
Add weighting functions $W_1(s)$ and $W_2(s)$.
Define the generalized plant $\mathbf{P}(s)$ as Eqn. 3.

$W_1(s)$ and $W_2(s)$ are the inverse of the frequency dependent specification of the sensor noises [2].

Finally:
 \mathcal{H}_∞ synthesis gives optimal $H_1(s)$,
 and get $H_2(s)$ from $1 - H_1(s)$.

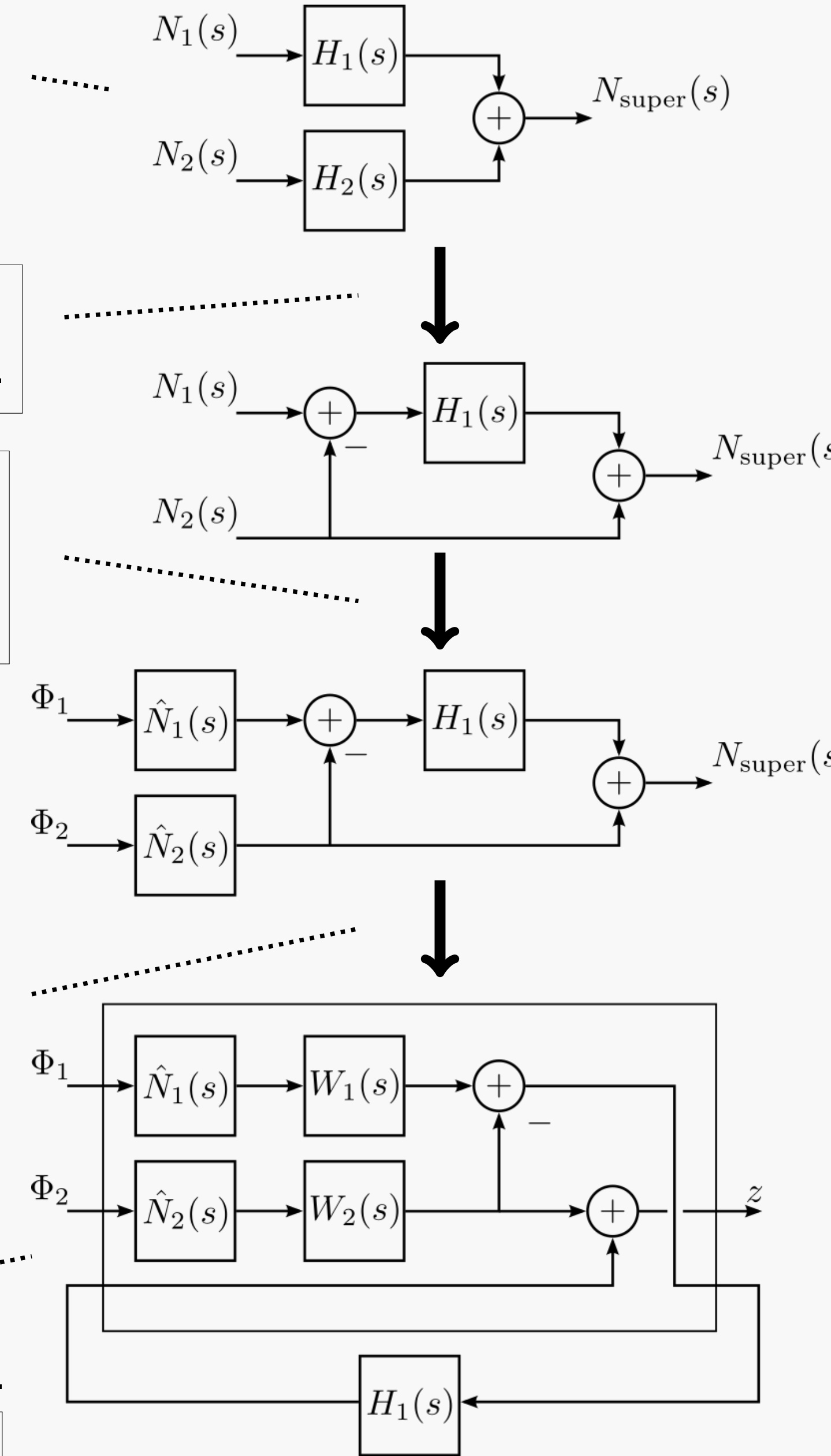


Figure 3: From a simple complementary filter configuration to generalized plant representation.

- ▶ The generalized plant is given by

$$\mathbf{P}(s) = \begin{bmatrix} 0 & \hat{N}_2(s)W_2(s) & 1 \\ \hat{N}_1(s)W_1(s) & -\hat{N}_2(s)W_2(s) & 0 \end{bmatrix}. \quad (3)$$

- ▶ The closed-loop plant is given by

$$\mathbf{G}(s) = [H_1(s)\hat{N}_1(s)W_1(s) \ H_2(s)\hat{N}_2(s)W_2(s)]. \quad (4)$$

- ▶ If we set $W_1(s) = 1/\hat{N}_2(s)$ and $W_2(s) = 1/\hat{N}_1(s)$, then the target specification of $N_1(s)$ is set to $N_2(s)$ when $|N_1(s)| \gg |N_2(s)|$, and vice versa.
- ▶ Minimizing $\|\mathbf{G}(s)\|_\infty$ gives optimal complementary filters that minimizes the maximum difference between the super sensor noise and the lower bound of the sensor noise in logarithmic scale.
- ▶ This is equivalent to minimizing the cost function

$$J = \sup_{\omega} (\log |N_{\text{super}}(j\omega)| - \log \min(|N_1(j\omega)|, |N_2(j\omega)|)). \quad (5)$$

- ▶ \mathcal{H}_∞ solvers are readily available in packages such as MATLAB and python-control.
- ▶ Complementary filters can be synthesized using the kontrol python package, which is developed for KAGRA's control systems.

Results: Complementary Filters for LVDTs and geophones

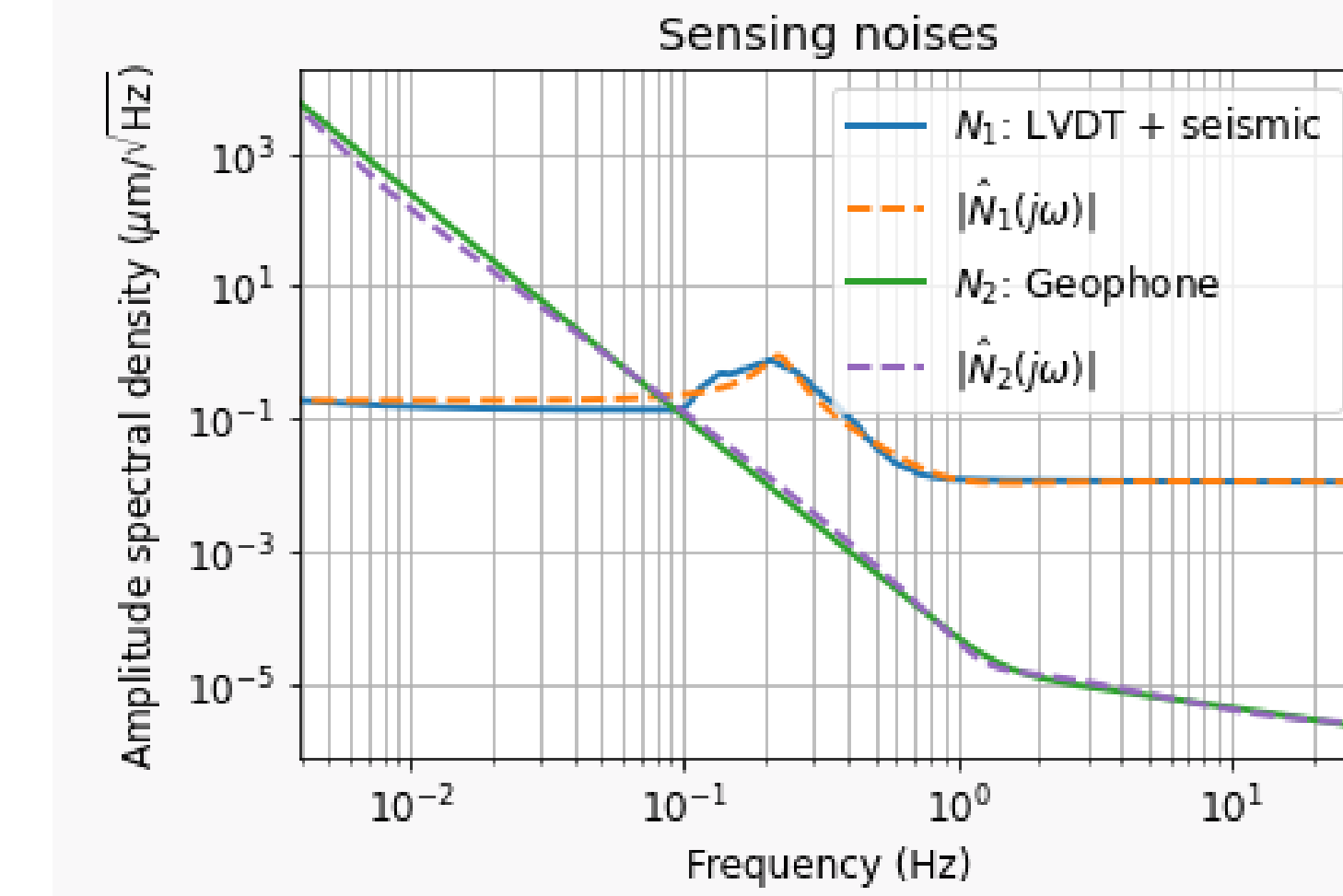


Figure 4: SRM sensor noises.

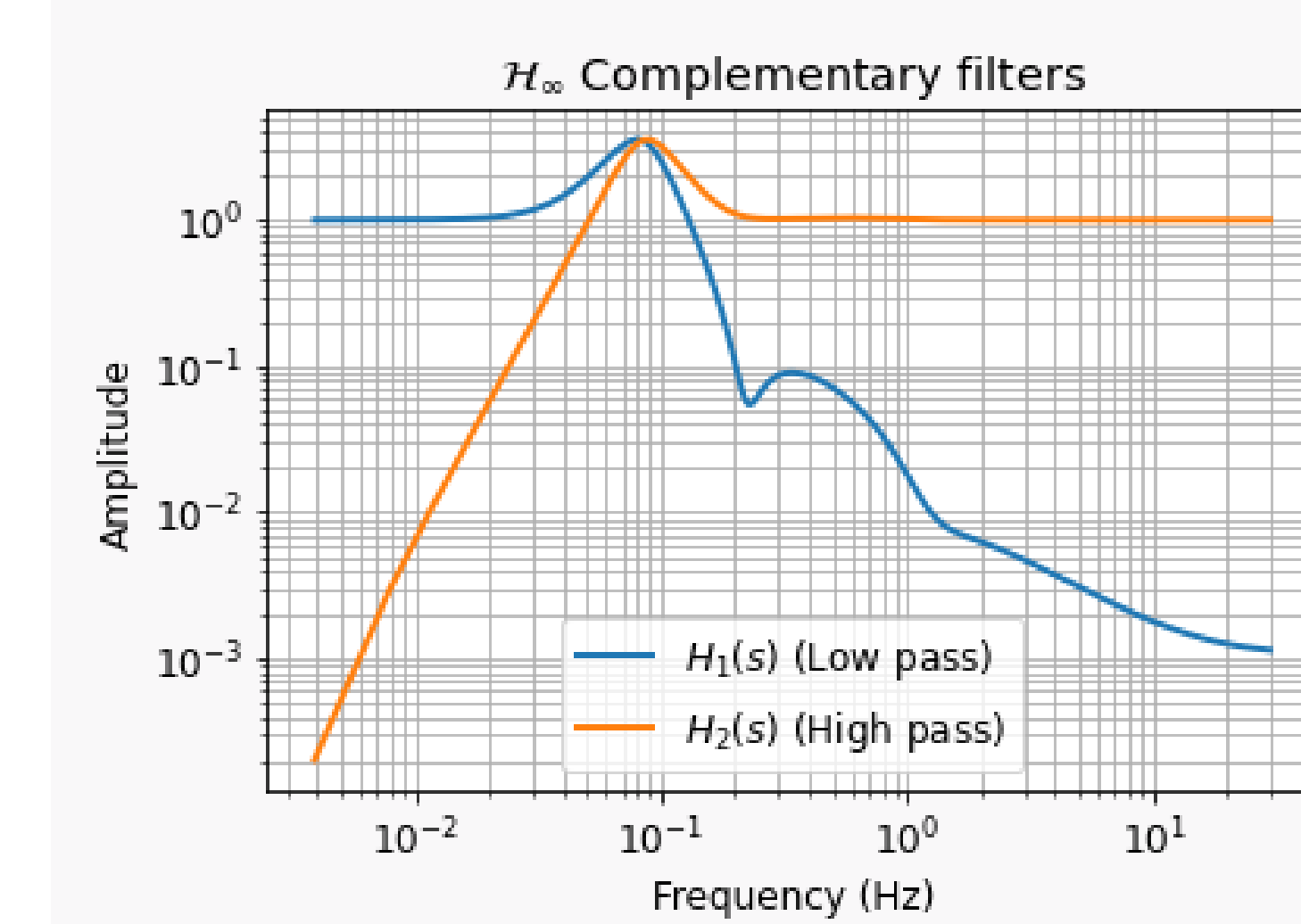


Figure 5: Filters (using proposed method)

- ▶ Fig. 4: the empirical model of the seismic noise-coupled LVDT and geophone readout noises N_1 and N_2 and the transfer function models $\hat{N}_1(s)$ and $\hat{N}_2(s)$
- ▶ Fig. 5: The complementary filters synthesized using \mathcal{H}_∞ method.
- ▶ Fig. 6: The predicted super sensor noise using optimal complementary filters $H_1(s)$ and $H_2(s)$.

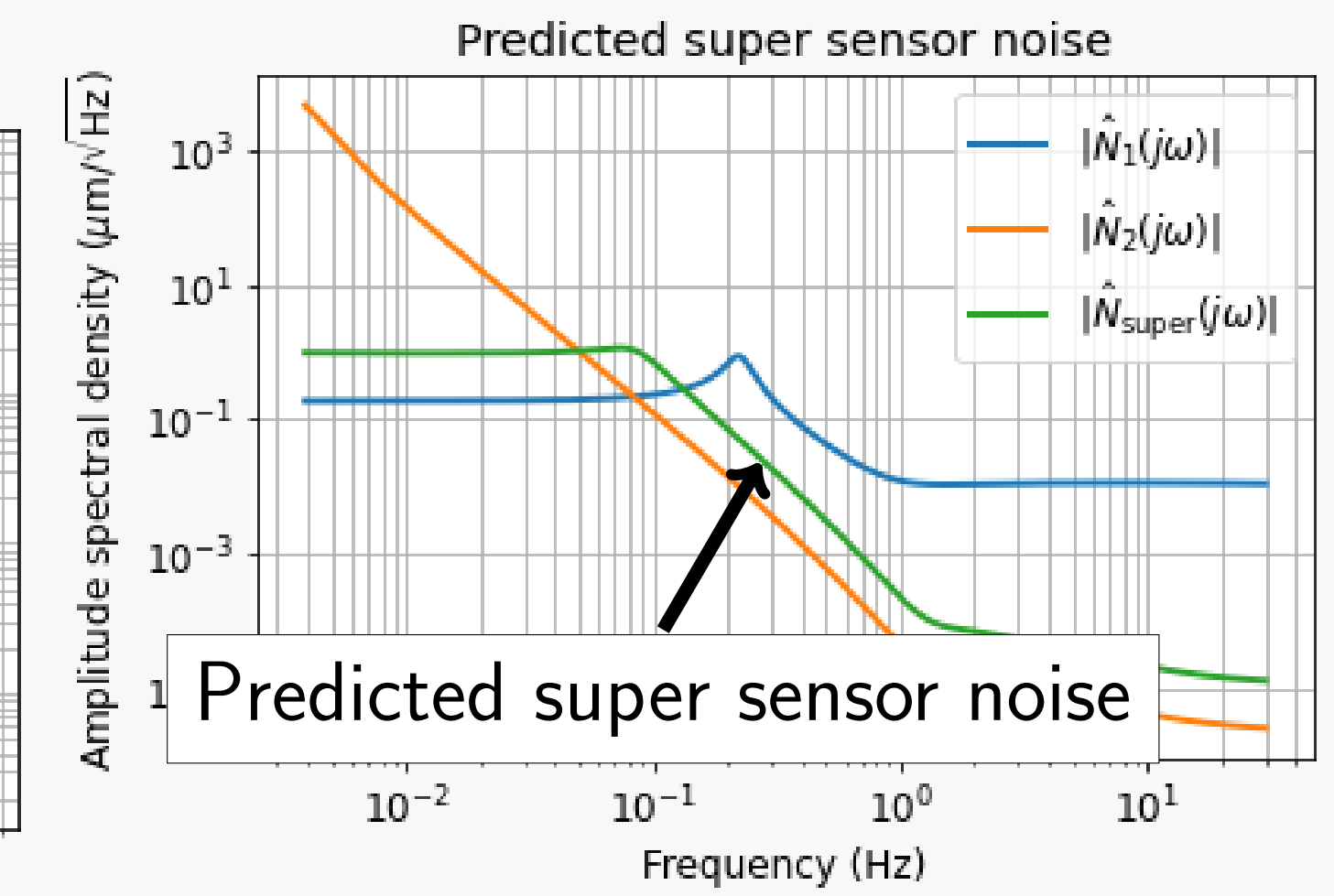


Figure 6: Predicted super sensor noise

- ▶ The predicted super sensor noise is defined by

$$|\hat{N}_{\text{super}}(j\omega)| = \left[|H_1(s)|^2 |\hat{N}_1(s)|^2 + |H_2(s)|^2 |\hat{N}_2(s)|^2 \right]^{1/2}. \quad (6)$$

- ▶ The super sensor noise is equally close to the lower bound at all frequencies (logarithmically).
- ▶ Preliminary implementation results can be found in Ref. [3].

Discussion and Future work.

- ▶ Unlike static filter designs, such as those in Refs. [4, 5, 6], the proposed method can generate filters that is optimal for any arbitrary sensor noises.
- ▶ Sensor correction and feedback-control filters can both be formulated into a complementary filter configuration. They can both be solved using the same method.

References

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