Assignment 2 (ML for TS) - MVA

Thomas Gravier thomas.gravier@gadz.org Richard Goudelin rgoudelin@gmail.com

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1 Introduction

Objective. The goal is to better understand the properties of AR and MA processes and do signal denoising with sparse coding.

Warning and advice.

- Use code from the tutorials as well as from other sources. Do not code yourself well-known procedures (e.g., cross-validation or k-means); use an existing implementation.
- The associated notebook contains some hints and several helper functions.
- Be concise. Answers are not expected to be longer than a few sentences (omitting calculations).

Instructions.

- Fill in your names and emails at the top of the document.
- Hand in your report (one per pair of students) by Monday 2nd December 11:59 PM.
- Rename your report and notebook as follows:
 FirstnameLastname1_FirstnameLastname1.pdf and
 FirstnameLastname2_FirstnameLastname2.ipynb.
 For instance, LaurentOudre_CharlesTruong.pdf.
- Upload your report (PDF file) and notebook (IPYNB file) using this link: https://docs.google.com/forms/d/e/1FAIpQLSfCqMXSDU9jZJbYUMmeLCXbVeckZYNiDpPl4hRUwcJ2

2 General questions

A time series $\{y_t\}_t$ is a single realisation of a random process $\{Y_t\}_t$ defined on the probability space (Ω, \mathcal{F}, P) , i.e. $y_t = Y_t(w)$ for a given $w \in \Omega$. In classical statistics, several independent realizations are often needed to obtain a "good" estimate (meaning consistent) of the parameters of the process. However, thanks to a stationarity hypothesis and a "short-memory" hypothesis, it is still possible to make "good" estimates. The following question illustrates this fact.

Question 1

An estimator $\hat{\theta}_n$ is consistent if it converges in probability when the number n of samples grows to ∞ to the true value $\theta \in \mathbb{R}$ of a parameter, i.e. $\hat{\theta}_n \stackrel{\mathcal{D}}{\longrightarrow} \theta$.

- Recall the rate of convergence of the sample mean for i.i.d. random variables with finite variance.
- Let $\{Y_t\}_{t\geq 1}$ a wide-sense stationary process such that $\sum_k |\gamma(k)| < +\infty$. Show that the sample mean $\bar{Y}_n = (Y_1 + \cdots + Y_n)/n$ is consistent and enjoys the same rate of convergence as the i.i.d. case. (Hint: bound $\mathbb{E}[(\bar{Y}_n \mu)^2]$ with the $\gamma(k)$ and recall that convergence in L_2 implies convergence in probability.)

Answer 1

Firstly:

The rate of convergence of the sample mean for i.i.d. random variables with finite variance is derived from the central limit theorem. Let $X_1, X_2, ..., X_n$ be i.i.d. random variables with mean θ and finite variance σ^2 . The sample mean is given by:

$$\hat{\theta}_n = \frac{1}{n} \sum_{i=1}^n X_i.$$

According to the central limit theorem, the scaled deviation of $\hat{\theta}_n$ from θ converges in distribution to a normal random variable:

$$\sqrt{n}(\hat{\theta}_n - \theta) \xrightarrow{D} \mathcal{N}(0, \sigma^2),$$

where \xrightarrow{D} denotes convergence in distribution. This indicates that the error in the estimator $\hat{\theta}_n$ decreases at the rate of $O\left(\frac{1}{\sqrt{n}}\right)$.

Secondly:

By applying the provided hint, we derive:

$$\mathbb{E}\left[(\overline{Y}_n - \mu)^2\right] = \mathbb{E}\left[\overline{Y}_n^2 - 2\overline{Y}_n\mu + \mu^2\right] = \mathbb{E}\left[\overline{Y}_n^2\right] - \mu^2 \le \mathbb{E}\left[\overline{Y}_n^2\right].$$

Let's develop $\mathbb{E}(\overline{Y}_n^2)$:

$$\mathbb{E}(\overline{Y}_n^2) = \mathbb{E}\left(\frac{1}{n^2} \sum_{i=1}^n \sum_{j=1}^n Y_i Y_j\right)$$
$$= \frac{1}{n^2} \sum_{i=1}^n \sum_{j=1}^n \mathbb{E}(Y_i Y_j)$$
$$= \frac{1}{n^2} \sum_{i=1}^n \sum_{j=1}^n \gamma[i-j]$$

$$\leq \frac{1}{n^2} \sum_{i=1}^n \sum_{j=1}^n |\gamma[i-j]|$$

$$\leq \frac{1}{n^2} \sum_{i=1}^n \sum_{j=1}^n 2|(\gamma[|i-j|])|$$

$$\leq \frac{2}{n^2} \sum_{i=1}^n \sum_{k=0}^{n-1} |(\gamma[k])|$$

We know that:

$$\sum_{k=0}^{\infty} |(\gamma[k])| < \infty \implies \exists C \in \mathbb{R}, \quad \sum_{k=0}^{n-1} |(\gamma[k])| \le C.$$

Thus:

$$\mathbb{E}(\overline{Y}_n^2) \le \frac{2C}{n^2}.$$

Therefore:

$$\lim_{n\to\infty} \mathbb{E}\left((\overline{Y}_n - \mu)^2\right) = 0.$$

It follows that:

$$\mathbb{E}\left[(\overline{Y}_n - \mu)^2\right] \to 0 \quad \Longrightarrow \quad \overline{Y}_n \xrightarrow{L^2} \mu \quad \Longrightarrow \quad \overline{Y}_n \xrightarrow{p} \mu.$$

Thus, the sample mean is consistent.

Given:

$$\mathbb{E}\left[(\bar{Y}_n - \mu)^2\right] \leq \frac{1}{n} \sum_{k} |\gamma(k)|,$$

where $\bar{Y}_n = \frac{1}{n} \sum_{t=1}^n Y_t$ is the sample mean and $\gamma(k)$ is the autocovariance function, if $\sum_k |\gamma(k)| < \infty$, then:

$$\mathbb{E}\left[(\bar{Y}_n - \mu)^2\right] \le \frac{C}{n}, \quad C = \sum_{k} |\gamma(k)|.$$

Thus, the variance decreases as $\mathcal{O}(1/n)$ and the standard deviation as $\mathcal{O}(1/\sqrt{n})$. The sample mean therefore converges to μ with a rate of $\mathcal{O}(1/\sqrt{n})$.

3 AR and MA processes

Question 2 *Infinite order moving average MA*(∞)

Let $\{Y_t\}_{t\geq 0}$ be a random process defined by

$$Y_t = \varepsilon_t + \psi_1 \varepsilon_{t-1} + \psi_2 \varepsilon_{t-2} + \dots = \sum_{k=0}^{\infty} \psi_k \varepsilon_{t-k}$$
 (1)

where $(\psi_k)_{k\geq 0} \subset \mathbb{R}$ ($\psi=1$) are square summable, i.e. $\sum_k \psi_k^2 < \infty$ and $\{\varepsilon_t\}_t$ is a zero mean white noise of variance σ_{ε}^2 . (Here, the infinite sum of random variables is the limit in L_2 of the partial sums.)

- Derive $\mathbb{E}(Y_t)$ and $\mathbb{E}(Y_tY_{t-k})$. Is this process weakly stationary?
- Show that the power spectrum of $\{Y_t\}_t$ is $S(f) = \sigma_{\varepsilon}^2 |\phi(e^{-2\pi i f})|^2$ where $\phi(z) = \sum_j \psi_j z^j$. (Assume a sampling frequency of 1 Hz.)

The process $\{Y_t\}_t$ is a moving average of infinite order. Wold's theorem states that any weakly stationary process can be written as the sum of the deterministic process and a stochastic process which has the form (1).

Answer 2

First we derive $\mathbb{E}(Y_t)$:

$$\mathbb{E}(Y_t) = \mathbb{E}\left(\sum_{r=0}^{\infty} \psi_r \epsilon_{t-r}\right) = \sum_{r=0}^{\infty} \psi_r \mathbb{E}(\epsilon_{t-r}) = 0.$$

Second we derive $\mathbb{E}(Y_t Y_{t-k})$:

$$\mathbb{E}(Y_t Y_{t-k}) = \mathbb{E}\left[\left(\sum_{i=0}^{\infty} \psi_i \varepsilon_{t-i}\right) \left(\sum_{j=0}^{\infty} \psi_j \varepsilon_{t-k-j}\right)\right]$$

$$= \mathbb{E}\sum_{i=0}^{\infty} \sum_{j=0}^{\infty} \psi_i \psi_j \varepsilon_{t-i} \varepsilon_{t-k-j}$$

$$= \sum_{i=0}^{\infty} \sum_{j=0}^{\infty} \psi_i \psi_j \mathbb{E}(\varepsilon_{t-i} \varepsilon_{t-k-j}).$$

Separate the cases:

- When i = j + k, the noise terms are correlated.
- When $i \neq j + k$, the noise terms are independent.

$$= \sum_{i=0}^{\infty} \psi_i \psi_{i-k} \mathbb{E}(\epsilon_{t-i}^2) + \sum_{i=0}^{\infty} \sum_{j \neq i-k} \psi_i \psi_j \mathbb{E}(\epsilon_{t-i}) \mathbb{E}(\epsilon_{t-k-j}).$$

Using the properties of white noise ($\mathbb{E}(\epsilon_t^2) = \sigma_\epsilon^2$ and $\mathbb{E}(\epsilon_t) = 0$:

$$=\sum_{i=k}^{\infty}\psi_{i}\psi_{i-k}\sigma_{\epsilon}^{2}+0$$

$$= \sigma_{\epsilon}^2 \sum_{i=k}^{\infty} \psi_i \psi_{i-k}.$$

This value do not depend on t, this imply that the autocovariance depends only on lag

Finally to show that a process is weakly stationary we need to show that the variance is constant $Var(Y_t)$:

$$\operatorname{Var}(Y_t) = \mathbb{E}(Y_t^2) - [\mathbb{E}(Y_t)]^2$$

Since $\mathbb{E}(Y_t) = 0$:

$$Var(Y_t) = \mathbb{E}(Y_t^2)$$

$$=\mathbb{E}\left[\left(\sum_{i=0}^{\infty}\psi_{i}\epsilon_{t-i}
ight)^{2}
ight]$$

$$= \mathbb{E}\left[\sum_{i=0}^{\infty}\sum_{j=0}^{\infty}\psi_{i}\psi_{j}\epsilon_{t-i}\epsilon_{t-j}\right]$$

$$=\sum_{i=0}^{\infty}\psi_i^2\sigma_{\epsilon}^2$$

$$= \sigma_{\epsilon}^2 \sum_{i=0}^{\infty} \psi_i^2.$$

and we know that $\sum_k \psi_k^2 < \infty$.

To conclude we got a constant variance, autocorrelation that depend only on lag and an average that is egual to 0 thus this process is weakly stationnary.

Now we can compute

$$\left|\sum_{j=0}^N \psi_j e^{-2i\pi f j}\right|^2, \quad \text{for } N \in \mathbb{N}:$$

$$\left|\sum_{j=0}^{N} \psi_j e^{-2i\pi f j}\right|^2 = \left(\sum_{j=0}^{N} \psi_j e^{-2i\pi f j}\right) \left(\sum_{l=0}^{N} \psi_l e^{2i\pi f l}\right)$$

$$= \sum_{j=0}^{N} \sum_{l=0}^{N} \psi_{j} \psi_{l} e^{-2i\pi f(j-l)}$$

$$= \sum_{\tau=-N+1}^{N-1} \sum_{n=0}^{N-\tau-1} \psi_{n} \psi_{n+\tau} e^{-2i\pi f\tau} \quad \text{(e.g Assignment 1)}.$$

We make $N \to \infty$, and we finally get:

$$\left|\phi\left(e^{-2i\pi f}\right)\right|^2 = \sum_{\tau=-\infty}^{\infty} \sum_{n=0}^{\infty} \psi_n \psi_{n+\tau} e^{-2i\pi f \tau}.$$

Let's compute the power spectrum, let *f*:

$$S(f) = \sum_{\tau = -\infty}^{\infty} \gamma(\tau) e^{-2i\pi f \tau}, \quad \text{with } f_s = 1 \text{ Hz.}$$

$$S(f) = \sigma_{\epsilon}^2 \sum_{\tau = -\infty}^{\infty} \sum_{n=0}^{\infty} \psi_n \psi_{n+\tau} e^{-2i\pi f \tau}.$$

$$S(f) = \sigma_{\epsilon}^2 \left| \phi \left(e^{-2i\pi f} \right) \right|^2.$$

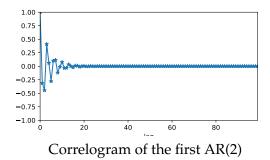
Question 3 *AR*(2) *process*

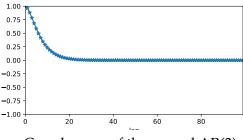
Let $\{Y_t\}_{t\geq 1}$ be an AR(2) process, i.e.

$$Y_t = \phi_1 Y_{t-1} + \phi_2 Y_{t-2} + \varepsilon_t \tag{2}$$

with $\phi_1, \phi_2 \in \mathbb{R}$. The associated characteristic polynomial is $\phi(z) := 1 - \phi_1 z - \phi_2 z^2$. Assume that ϕ has two distinct roots (possibly complex) r_1 and r_2 such that $|r_i| > 1$. Properties on the roots of this polynomial drive the behavior of this process.

- Express the autocovariance coefficients $\gamma(\tau)$ using the roots r_1 and r_2 .
- Figure 1 shows the correlograms of two different AR(2) processes. Can you tell which one has complex roots and which one has real roots?
- Express the power spectrum S(f) (assume the sampling frequency is 1 Hz) using $\phi(\cdot)$.
- Choose ϕ_1 and ϕ_2 such that the characteristic polynomial has two complex conjugate roots of norm r = 1.05 and phase $\theta = 2\pi/6$. Simulate the process $\{Y_t\}_t$ (with n = 2000) and display the signal and the periodogram (use a smooth estimator) on Figure 2. What do you observe?





Correlogram of the second AR(2)

Figure 1: Two AR(2) processes

Answer 3

1. Expression of Autocovariance Coefficients $\gamma(\tau)$ Using Roots r_1 and r_2 :

To express Y(z) using r_1 and r_2 , we do the same logic for the parameter estimation of an AR model in the course:

$$Y_{t} = \phi_{1}Y_{t-1} + \phi_{2}Y_{t-2} + \epsilon_{t}$$

$$\Leftrightarrow Y_{t} - \phi_{1}Y_{t-1} - \phi_{2}Y_{t-2} = \epsilon_{t}$$

$$\Rightarrow Y_{t}Y_{t-\tau} - \phi_{1}Y_{t-1}Y_{t-\tau} - \phi_{2}Y_{t-2}Y_{t-\tau} = \epsilon_{t}Y_{t-\tau}$$

$$\Rightarrow \mathbb{E}(Y_{t}Y_{t-\tau}) - \phi_{1}\mathbb{E}(Y_{t-1}Y_{t-\tau}) - \phi_{2}\mathbb{E}(Y_{t-2}Y_{t-\tau}) = \mathbb{E}(\epsilon_{t}Y_{t-\tau})$$

$$= \mathbb{E}(\epsilon_t) \cdot \mathbb{E}(Y_{t-\tau}) = 0$$
 (by independence)

$$\Rightarrow \gamma(\tau) = \phi_1 \gamma(\tau - 1) + \phi_2 \gamma(\tau - 2)$$

So here it is a recurrent series of order 2, with the following polynomial characteristic:

$$z^2 - \phi_1 z - \phi_2 = P(z).$$

So we can see that

$$P(z) = z^2 \Phi(1/z)$$

we deduce that the roots of P are the inverse of the roots of Φ

- Cas 1: $\Phi(z)$ admet deux racines réelles distinctes $\frac{1}{r_1}$ et $\frac{1}{r_2}$ $(r_1 \neq r_2)$

$$\exists (\lambda, \mu) \in \mathbb{R}^2 \text{ tel que } \forall \tau \in \mathbb{N}, \quad \gamma(\tau) = \lambda \left(\frac{1}{r_1}\right)^{\tau} + \mu \left(\frac{1}{r_2}\right)^{\tau}.$$

- Cas 2 : $\Phi(z)$ admet une racine double $\frac{1}{r_1} = \frac{1}{r_2} = \frac{1}{r}$

$$\exists (\lambda,\mu) \in \mathbb{R}^2 \text{ tel que } \forall \tau \in \mathbb{N}, \quad \gamma(\tau) = \lambda \left(\frac{1}{r}\right)^{\tau} + \mu \tau \left(\frac{1}{r}\right)^{\tau}.$$

– Cas 3 : $\Phi(z)$ admet deux racines complexes conjuguées $\frac{1}{r_1}=\frac{1}{R}e^{-i\theta}$ et $\frac{1}{r_2}=\frac{1}{R}e^{i\theta}$

$$\exists (\lambda, \mu) \in \mathbb{R}^2 \text{ tel que } \forall \tau \in \mathbb{N}, \quad \gamma(\tau) = \left(\frac{1}{R}\right)^{\tau} \left(\lambda \cos(\tau\theta) + \mu \sin(\tau\theta)\right).$$

2. Process with complex roots:

Based on the observations above, we conclude that $r_1, r_2 \in \mathbb{C}$ correspond to an oscillatory system whose autocorrelation fades to 0 over lag (in fact $|r_i| > 1$). Thus, the correlogram on the left indicates the presence of complex roots. On the other hand, $r_1, r_2 \in \mathbb{R}$ correspond to a second-order system with monotonic decay to 0. Therefore, the correlogram on the right is associated with real roots.

3. Expression of the Power Spectrum S(f):

Let's introduce the lag operator L such that $LY_t = Y_{t-1}$. So by applying $\phi(L)$ over Y_t we have:

$$\phi(L)Y_t = (1 - \phi_1 L - \phi_2 L^2)Y_t = Y_t - \phi_1 Y_{t-1} - \phi_2 Y_{t-2} = \epsilon_t \quad (Y_t = \phi_1 Y_{t-1} + \phi_2 Y_{t-2} + \epsilon_t)$$

r1 and r2 being the roots of $\phi(z)$ we have:

$$-\phi_2(r_1 - L)(r_2 - L)Y_t = -\phi_2\left(1 - \frac{L}{r_1}\right)\left(1 - \frac{L}{r_2}\right)Y_t = \epsilon_t$$

equivalent to:

$$Y_t = rac{1}{-\phi_2 \left(1 - rac{L}{r_1}
ight) \left(1 - rac{L}{r_2}
ight)} \epsilon_t$$

We can remark a geometric sum, with |L/ri| < 1 cause |ri| > 1 that

$$\frac{1}{1 - \frac{L}{r_i}} = \sum_{k=0}^{\infty} \frac{L^k}{r_i^k}$$

Which led to:

$$Y_t = \frac{\epsilon_t}{-\phi_2} \sum_{k=0}^{\infty} \frac{L^k}{r_1^k} \sum_{l=0}^{\infty} \frac{L^l}{r_2^l}$$

$$Y_t = \frac{\epsilon_t}{-\phi_2} \sum_{k=0}^{\infty} \frac{L^k}{r_1^k} \sum_{l=0}^{\infty} \frac{L^l}{r_2^l}$$

$$Y_t = \frac{\epsilon_t}{-\phi_2} \sum_{k=0}^{\infty} \sum_{l=0}^{\infty} \frac{L^{k+l}}{r_1^k r_2^l}$$

$$Y_t = \frac{1}{-\phi_2} \sum_{k=0}^{\infty} \sum_{l=0}^{\infty} \frac{\epsilon_{t-(k+l)}}{r_1^k r_2^l}$$

$$Y_t = \frac{1}{-\phi_2} \sum_{j=0}^{\infty} \sum_{k=0}^{j} \frac{\epsilon_{t-j}}{r_1^k r_2^{j-k}}$$

$$Y_t = \frac{1}{-\phi_2} \sum_{j=0}^{\infty} c_j \epsilon_{t-j}, \quad \text{where } c_j = \sum_{k=0}^{j} \frac{1}{r_1^k r_2^{j-k}}.$$

Finally here we recognize that a AR(2) model is a MA(∞)] model thus from question two we deduce that it's power spectrum is :

$$S(f) = \frac{1}{\phi_2^2} \sigma_{\epsilon}^2 \left| \phi \left(e^{-2i\pi f} \right) \right|^2.$$

4. Simulation and Periodogram:

First, let's find the values of ϕ_1 and ϕ_2 : Knowing $r_1 = re^{i\theta}$ and $r_2 = re^{-i\theta}$, we have the following system of equations:

$$\begin{cases} 1 - \phi_1 r_1 - \phi_2 r_1^2 = 0, \\ 1 - \phi_1 r_2 - \phi_2 r_2^2 = 0. \end{cases}$$

Solving the system:

$$\phi_1 = \frac{1 - \phi_2 r_1^2}{r_1},$$

$$1 - \frac{1 - \phi_2 r_1^2}{r_1} r_2 - \phi_2 r_2^2 = 0,$$

$$1 - \frac{r_2}{r_1} + \phi_2 r_2 (r_1 - r_2) = 0,$$

$$\phi_2=-\frac{1}{r_1r_2}.$$

Substituting $r_1 = re^{i\theta}$ and $r_2 = re^{-i\theta}$:

$$\phi_2 = -\frac{1}{(re^{i\theta})(re^{-i\theta})} = -\frac{1}{r^2}.$$

$$\phi_1 = \frac{r_1 + r_2}{r_1 r_2} = \frac{re^{i\theta} + re^{-i\theta}}{r^2} = \frac{2r\cos\theta}{r^2} = \frac{2\cos\theta}{r}.$$

Final expressions:

$$\phi_1 = \frac{2\cos\theta}{r}, \quad \phi_2 = -\frac{1}{r^2}.$$

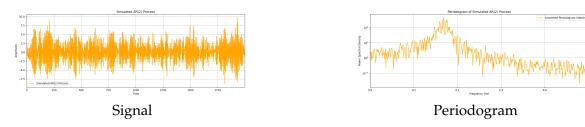


Figure 2: AR(2) process

A peak is observed near the frequency f=0.17, indicating a dominant frequency component in the time series. This observation aligns with $\theta=\frac{2\pi}{6}$, as it corresponds to $2\pi f=\theta$, leading to $f=\frac{1}{6}$.

4 Sparse coding

The modulated discrete cosine transform (MDCT) is a signal transformation often used in sound processing applications (for instance, to encode an MP3 file). A MDCT atom $\phi_{L,k}$ is defined for a length 2L and a frequency localisation k (k = 0, ..., L-1) by

$$\forall u = 0, \dots, 2L - 1, \quad \phi_{L,k}[u] = w_L[u] \sqrt{\frac{2}{L}} \cos\left[\frac{\pi}{L} \left(u + \frac{L+1}{2}\right) (k + \frac{1}{2})\right]$$
 (3)

where w_L is a modulating window given by

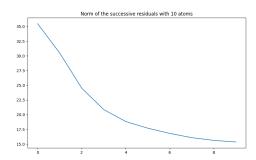
$$w_L[u] = \sin\left[\frac{\pi}{2L}\left(u + \frac{1}{2}\right)\right]. \tag{4}$$

Question 4 Sparse coding with OMP

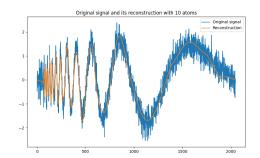
For the signal provided in the notebook, learn a sparse representation with MDCT atoms. The dictionary is defined as the concatenation of all shifted MDCDT atoms for scales L in [32,64,128,256,512,1024].

- For the sparse coding, implement the Orthogonal Matching Pursuit (OMP). (Use convolutions to compute the correlation coefficients.)
- Display the norm of the successive residuals and the reconstructed signal with 10 atoms.

Answer 4



Norms of the successive residuals



Reconstruction with 10 atoms

Figure 3: Question 4