

**SIGNALS AND SYSTEMS USING MATLAB**  
**Chapter 1 — Continuous-time Signals**

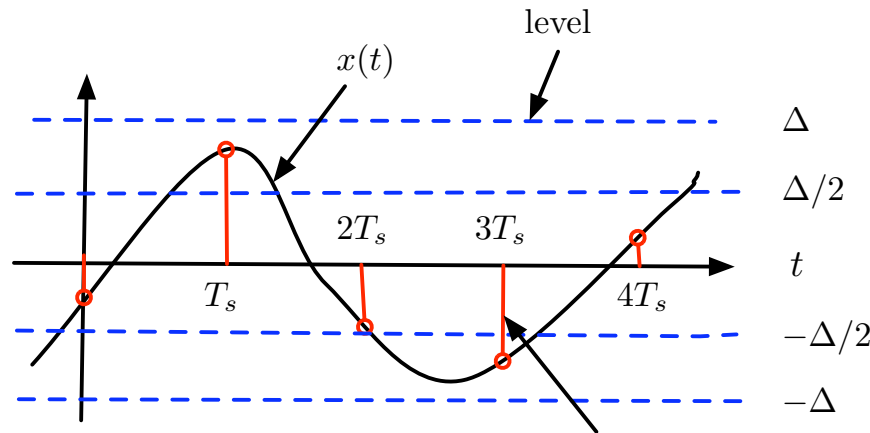
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# Classification of time-dependent signals

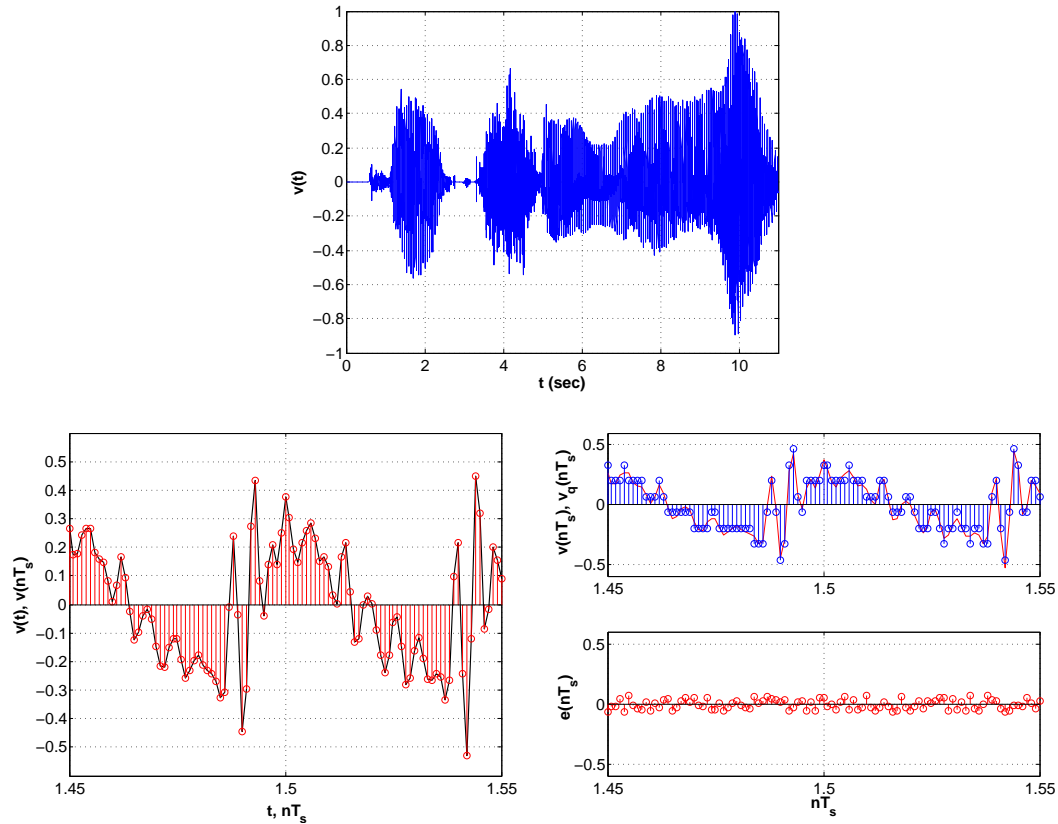
- *Predictability*: random or deterministic
- *Variation of time and amplitude*: continuous-time, discrete-time, or digital
- *Energy/power*: finite or infinite energy/power
- *Repetitive behavior*: periodic or aperiodic
- *Symmetry with respect to time origin*: even or odd
- *Support*: Finite or infinite support (outside support signal is always zero)

# Analog to digital and digital to analog conversion

- Analog to digital converter (ADC or A/D converter): converts analog signals into digital signals
- Digital to analog converter (DAC or D/A converter): converts digital to analog signals



*Discretization in time and in amplitude of analog signal  $x(t)$  using sampling period  $T_s$  and quantization level  $\Delta$ . In time, samples are taken at uniform times  $\{nT_s\}$ , and in amplitude the range of amplitudes is divided into a finite number of levels so that each sample value is approximated by one of them*



Segment of voice signal on top is sampled and quantized. Bottom left: voice segment (continuous line) and the sampled signal (vertical samples) using a sampling period  $T_s = 10^{-3}$  sec. Bottom-right: sampled and quantized signal at the top, and quantization error, difference between the sampled and the quantized signals, at the bottom.

$$\begin{array}{l} x(.) : \mathcal{R} \rightarrow \mathcal{R} \ (\mathcal{C}) \\ t \rightarrow x(t) \end{array}$$

Example: complex signal  $y(t) = (1 + j)e^{j\pi t/2}$ ,  $0 \leq t \leq 10$ , 0 otherwise

$$\begin{aligned} y(t) &= \sqrt{2}e^{j(\pi t/2 + \pi/4)} \\ &= \begin{cases} \sqrt{2}[\cos(\pi t/2 + \pi/4) + j \sin(\pi t/2 + \pi/4)], & 0 \leq t \leq 10, \\ 0, & \text{otherwise} \end{cases} \end{aligned}$$

If  $x(t) = \sqrt{2} \cos(\pi t/2 + \pi/4)$ ,  $-\infty < t < \infty$   
 $p(t) = 1$ ,  $0 \leq t \leq 10$ , 0 otherwise then

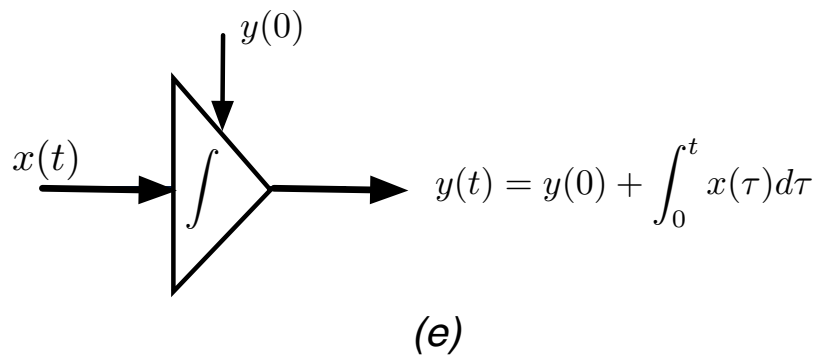
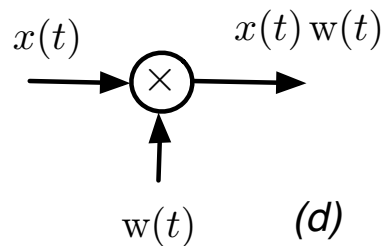
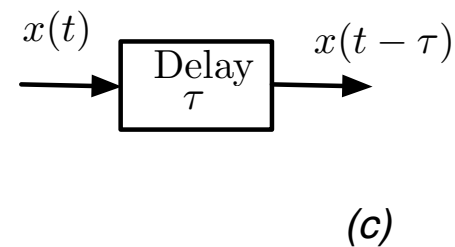
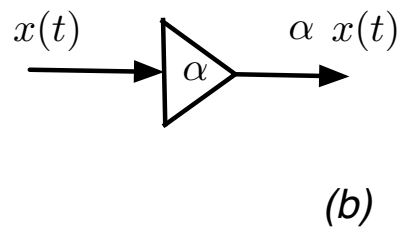
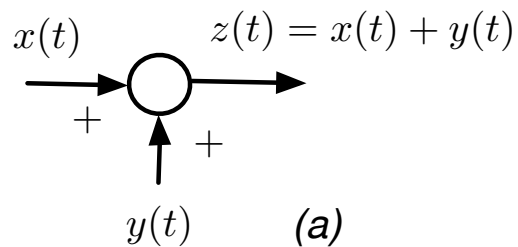
$$y(t) = [x(t) + jx(t - 1)]p(t)$$

# Basic signal operations

Given signals  $x(t)$ ,  $y(t)$ , constants  $\alpha$  and  $\tau$ , and function  $w(t)$ :

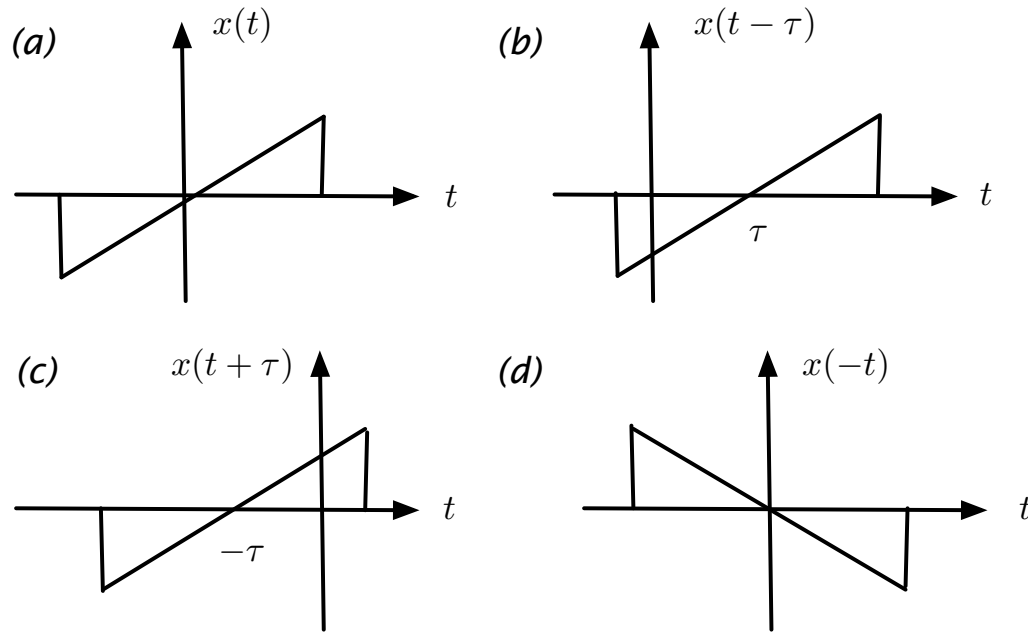
- *Signal addition/subtraction:*  $x(t) + y(t)$ ,  $x(t) - y(t)$
- *Constant multiplication:*  $\alpha x(t)$
- *Time shifting*
  - $x(t - \tau)$  is  $x(t)$  *delayed* by  $\tau$
  - $x(t + \tau)$  is  $x(t)$  *advanced* by  $\tau$
- *Time scaling*  $x(\alpha t)$ 
  - $\alpha = -1$ ,  $x(-t)$  reversed in time or *reflected*
  - $\alpha > 1$ ,  $x(\alpha t)$  is  $x(t)$  *compressed*
  - $\alpha < 1$ ,  $x(\alpha t)$  is  $x(t)$  *expanded*
- *Time windowing*  $x(t)w(t)$ ,  $w(t)$  *window*
- *Integration*

$$y(t) = \int_{t_0}^t x(\tau) d\tau + y(t_0)$$



*Basic signal operations: (a) adder, (b) constant multiplier, (c) delay, (d) time-windowing, (e) integrator*

## Delayed, advanced and reflected signals



*Continuous-time signal (a), and its delayed (b), advanced (c), and reflected (d) versions.*



## Example

$$x(t) = \begin{cases} t & -1 \leq t \leq 1 \\ 0 & \text{otherwise} \end{cases}$$

$$\text{delayed by 1: } x(t-1) = \begin{cases} t-1 & 0 \leq t \leq 2 \\ 0 & \text{otherwise} \end{cases}$$

$$\text{advanced by 1: } x(t+1) = \begin{cases} t+1 & -2 \leq t \leq 0 \\ 0 & \text{otherwise} \end{cases}$$

$$\text{reflected: } x(-t) = \begin{cases} -t & -1 \leq t \leq 1 \\ 0 & \text{otherwise} \end{cases}$$

$$\text{reflected and delayed by 1: } x(-t+1) = \begin{cases} -t+1 & 0 \leq t \leq 2 \\ 0 & \text{otherwise} \end{cases}$$

$$\text{reflected and advanced by 1: } x(-t-1) = \begin{cases} -t-1 & -2 \leq t \leq 0 \\ 0 & \text{otherwise} \end{cases}$$

$$\text{compressed by 2: } x(2t) = \begin{cases} 2t & -\frac{1}{2} \leq t \leq \frac{1}{2} \\ 0 & \text{otherwise} \end{cases}$$

$$\text{expanded by 2: } x(t/2) = \begin{cases} t/2 & -2 \leq t \leq 2 \\ 0 & \text{otherwise} \end{cases}$$

## Even and odd signals

$$\begin{array}{ll} x(t) & \text{even : } x(t) = x(-t) \\ x(t) & \text{odd : } x(t) = -x(-t) \end{array}$$

- *Even and odd decomposition:* For any signal  $y(t)$

$$y(t) = y_e(t) + y_o(t)$$

$$y_e(t) = 0.5 [y(t) + y(-t)] \quad \text{even component}$$

$$y_o(t) = 0.5 [y(t) - y(-t)] \quad \text{odd component}$$

**Example**  $x(t) = \cos(2\pi t + \theta)$ ,  $-\infty < t < \infty$

$$\text{even } x(t) = x(-t) \rightarrow \cos(2\pi t + \theta) = \cos(-2\pi t + \theta) = \cos(2\pi t - \theta)$$

$$\theta = -\theta, \text{ or } \theta = 0, \pi$$

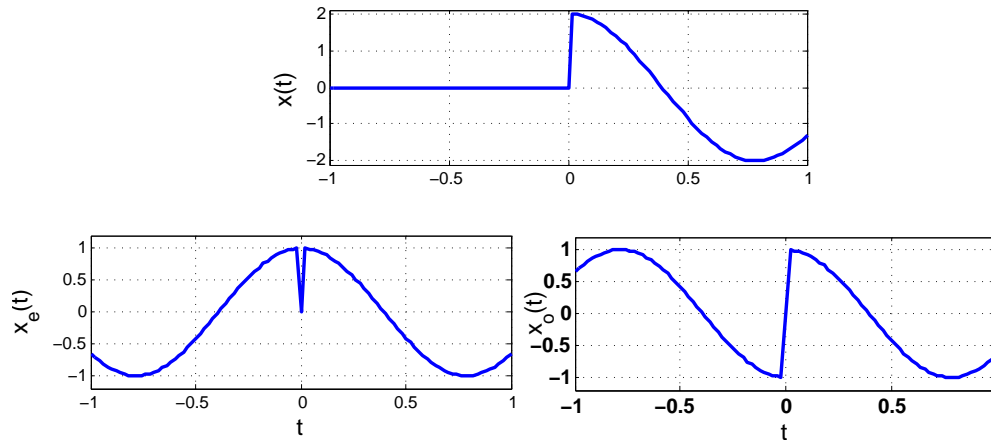
$$\begin{aligned} \text{odd } x(t) = -x(-t) &\rightarrow \cos(2\pi t + \theta) = -\cos(-2\pi t + \theta) = \cos(-2\pi t + \theta \pm \pi) \\ &= \cos(2\pi t - \theta \mp \pi) \end{aligned}$$

$$\theta = -\theta \mp \pi, \text{ or } \theta = \mp \pi/2$$

Example Given

$$x(t) = \begin{cases} 2 \cos(4t) & t > 0 \\ 0 & \text{otherwise} \end{cases}$$

not even or odd, its even and odd components are



If signal is 2 at  $t = 0$

$$x_1(t) = \begin{cases} 2 \cos(4t) & t \geq 0 \\ 0 & \text{otherwise} \end{cases}$$

the odd component is same as before, and the even component is 2 at  $t = 0$  and same as before otherwise

$x(t)$  is periodic if

(i)  $x(t)$  defined in  $-\infty < t < \infty$ , and

(ii) there is  $T_0 > 0$ , the fundamental period of  $x(t)$ ,  
such that  $x(t + kT_0) = x(t)$ , integer  $k$

**Example**  $x(t) = e^{j2t}$  and  $y(t) = e^{j\pi t}$

- $x(t) = \cos(2t) + j \sin(2t)$  periodic with  $T_0 = 2\pi/2 = \pi$
- $y(t) = \cos(\pi t) + j \sin(\pi t)$  periodic with  $T_1 = 2\pi/\pi = 2$
- $z(t) = x(t) + y(t)$  is not periodic as  $T_0/T_1 \neq M/N$  where  $M, N$  integers
- $w(t) = x(t)y(t) = e^{j(2+\pi)t} = \cos(\Omega_2 t) + j \sin(\Omega_2 t)$ ,  $\Omega_2 = 2 + \pi \rightarrow w(t)$  periodic with  $T_2 = 2\pi/(2 + \pi)$
- $p(t) = (1 + x(t))(1 + y(t)) = 1 + x(t) + y(t) + x(t)y(t)$  not periodic

$$\text{Energy of } x(t) : E_x = \int_{-\infty}^{\infty} |x(t)|^2 dt,$$

$$\text{Power of } x(t) : P_x = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T |x(t)|^2 dt$$

- $x(t)$  is *finite-energy, or square integrable*, if  $E_x < \infty$
- $x(t)$  is *finite-power* if  $P_x < \infty$

## Example

- $x(t) = e^{-at}$ ,  $a > 0$ ,  $t \geq 0$  and 0 otherwise is finite energy and zero power
- $y(t) = (1 + j)e^{j\pi t/2}$ ,  $0 \leq t \leq 10$ , and 0 otherwise is finite energy and zero power

$$E_y = \int_0^{10} |(1 + j)e^{j\pi t/2}|^2 dt = 2 \int_0^{10} dt = 20$$

## Power of periodic signal

$x(t)$  period of fundamental period  $T_0$  is

$$P_x = \frac{1}{T_0} \int_{t_0}^{t_0+T_0} x^2(t) dt$$

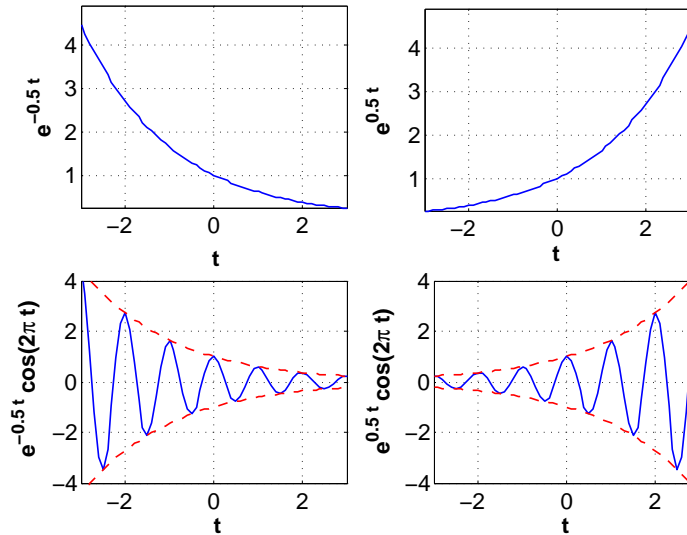
for any  $t_0$ , i.e., the average energy in a period of the signal Let  $T = NT_0$ , integer  $N > 0$ :

$$\begin{aligned} P_x &= \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T x^2(t) dt = \lim_{N \rightarrow \infty} \frac{1}{2NT_0} \int_{-NT_0}^{NT_0} x^2(t) dt \\ &= \lim_{N \rightarrow \infty} \frac{1}{2NT_0} \left[ N \int_{-T_0}^{T_0} x^2(t) dt \right] = \frac{1}{2T_0} \int_{-T_0}^{T_0} x^2(t) dt \\ &= \frac{1}{T_0} \int_{t_0}^{t_0+T_0} x^2(t) dt \end{aligned}$$

# Basic signals

- Complex exponential

$$\begin{aligned}x(t) &= Ae^{at} = |A|e^{j\theta}e^{(r+j\Omega_0)t} \\ &= |A|e^{rt} [\cos(\Omega_0 t + \theta) + j \sin(\Omega_0 t + \theta)] \quad -\infty < t < \infty\end{aligned}$$



*Analog exponentials: decaying exponential (top left), growing exponential (top right), modulated exponential decaying and growing (bottom left and right).*

- Sinusoid

$$A \cos(\Omega_0 t + \theta) = A \sin(\Omega_0 t + \theta + \pi/2) \quad -\infty < t < \infty$$

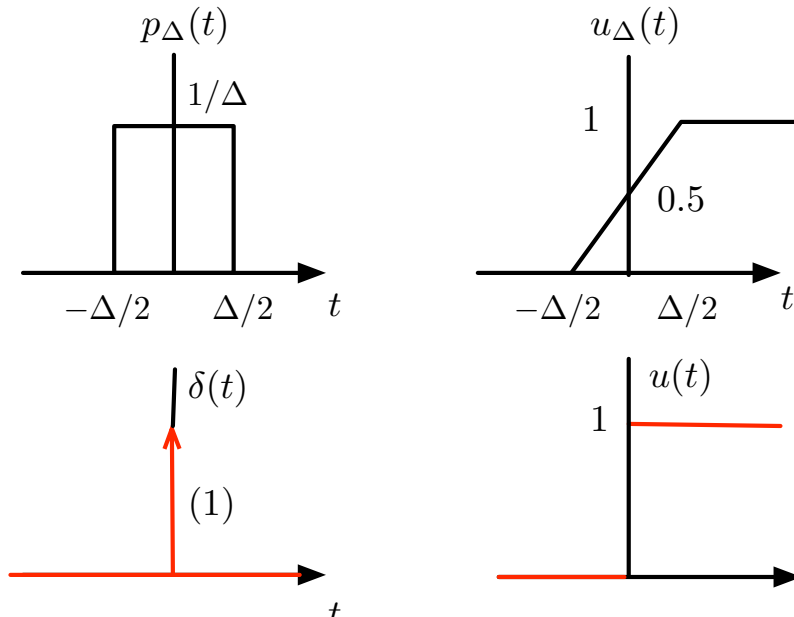
Modulation systems

$$A(t) \cos(\Omega(t)t + \theta(t))$$

- *Amplitude modulation or AM:*  $A(t)$  changes according to the message, frequency and phase constant,
- *Frequency modulation or FM:*  $\Omega(t)$  changes according to the message, amplitude and phase constant,
- *Phase modulation or PM:*  $\theta(t)$  changes according to the message, amplitude and frequency constant



- Unit-impulse signal



Unit-impulse  $\delta(t)$  and unit-step  $u(t)$  as  $\Delta \rightarrow 0$  in pulse  $p_\Delta(t)$  and its integral  $u_\Delta(t)$ .

### Unit-impulse

$$\delta(t) = \begin{cases} 0 & t \neq 0 \\ \text{undefined} & t = 0 \end{cases}$$

$$\int_{-\infty}^t \delta(\tau) d\tau = \begin{cases} 1 & t > 0 \\ 0 & \text{otherwise} \end{cases}$$

- Unit–step signal

$$u(t) = \begin{cases} 1 & t > 0 \\ 0 & t \leq 0 \end{cases}$$

- Ramp signal

$$r(t) = tu(t) = \begin{cases} t & t \geq 0 \\ 0 & t < 0 \end{cases}$$

- Relations

$$\frac{dr(t)}{dt} = u(t), \quad \frac{d^2r(t)}{dt^2} = \delta(t)$$

$$\frac{du(t)}{dt} = \delta(t)$$

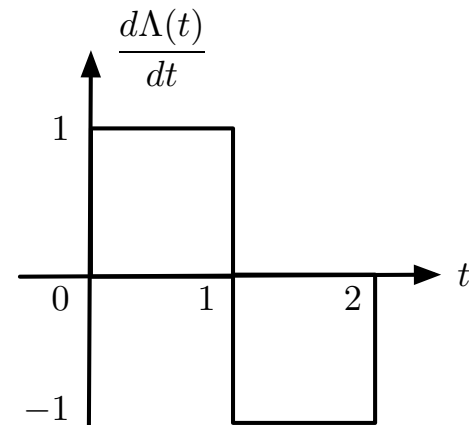
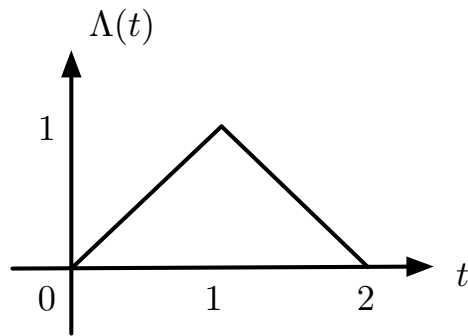
$$\int_{-\infty}^t \delta(\tau) d\tau = u(t), \quad \int_{-\infty}^t u(\tau) d\tau = r(t)$$

## Example Triangular pulse

$$\Lambda(t) = \begin{cases} t & 0 \leq t \leq 1 \\ -t + 2 & 1 < t \leq 2 \\ 0 & \text{otherwise} \end{cases}$$
$$= r(t) - 2r(t-1) + r(t-2)$$

Derivative

$$\frac{d\Lambda(t)}{dt} = \begin{cases} 1 & 0 \leq t \leq 1 \\ -1 & 1 < t \leq 2 \\ 0 & \text{otherwise} \end{cases}$$
$$= u(t) - 2u(t-1) + u(t-2)$$

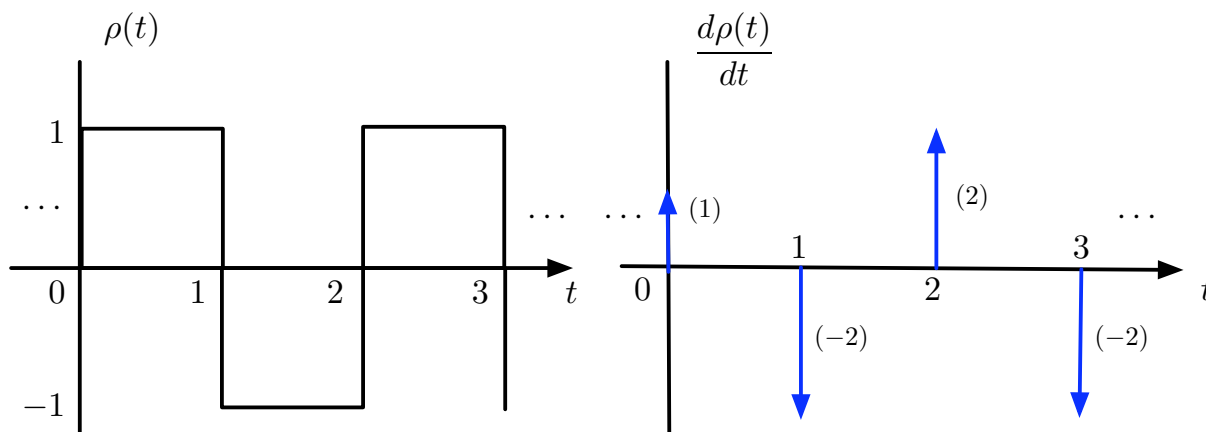


### Example Causal train of pulses

$$\rho(t) = \sum_{k=0}^{\infty} s(t - 2k), \quad s(t) = u(t) - 2u(t - 1) + u(t - 2)$$

Derivative

$$\frac{d\rho(t)}{dt} = \delta(t) + 2 \sum_{k=1}^{\infty} \delta(t - 2k) - 2 \sum_{k=1}^{\infty} \delta(t - 2k + 1)$$

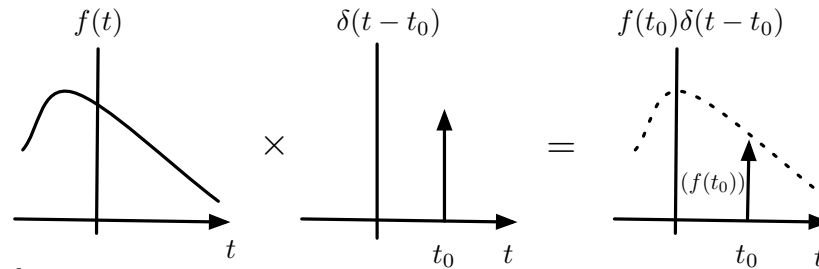


*The number in ( ) is area of the corresponding delta signal and it indicates the jump at the particular discontinuity, positive when increasing and negative when decreasing*

# Generic representation of signals

- Sifting property of  $\delta(t)$

$$\int_{-\infty}^{\infty} f(t)\delta(t - \tau)dt = \int_{-\infty}^{\infty} f(\tau)\delta(t - \tau)dt = f(\tau) \int_{-\infty}^{\infty} \delta(t - \tau)dt = f(\tau), \text{ for any } \tau$$



- Generic representation

$$x(t) = \int_{-\infty}^{\infty} x(\tau)\delta(t - \tau)d\tau$$

