# Supplementary for "Adaptive Seasonal-Trend

- Decomposition for Streaming Time Series Data
- <sup>3</sup> with Transitions and Fluctuations in Seasonality"
  - Thanapol Phungtua-eng \( \sime \) and Yoshitaka Yamamoto
- 5 Department of Informatics, Shizuoka University, Shizuoka, Japan
- thanapol@yy-lab.info, yyamamoto@inf.shizuoka.ac.jp

## $_{ ext{r}}$ 1 Sliding Discrete Fourier transform (SDFT)

- 8 Here, this is proof for SDFT eq. (as shown in Eq. (2) in the main paper) [4,5].
- Given an input sequence with a length of at least (N+q+1), where q denotes
- the starting index of the DFT window, we consider a DFT of length N for the
- window  $(x_q, x_{q+1}, \dots, x_{q+N-1})$ :

$$X_q = \sum_{n=0}^{N-1} x_{n+q} e^{-j2\pi nk/N} \tag{1}$$

- Then, sliding to the next window with the starting point at the (q+1)-th posi-
- tion, we compute the DFT of length N for this new window  $(x_{q+1}, x_{q+2}, \dots, x_{q+N})$ ,
- dynamically tracking changes in the frequency domain as the window advances:

$$X_{q+1} = \sum_{n=0}^{N-1} x_{n+q+1} e^{-j2\pi nk/N}$$
 (2)

Substituting p = n + 1 for the range 1 to N, we have:

$$X_{q+1} = \sum_{p=1}^{N} x_{p+q} e^{-j2\pi(p-1)k/N}$$
(3)

Adjusting for the N-th term by subtracting and adding the p=0 case:

$$X_{q+1} = \sum_{p=0}^{N-1} x_{p+q} e^{-j2\pi(p-1)k/N} + x_{q+N} e^{-j2\pi(N-1)k/N} - x_q e^{j2\pi k/N}$$
 (4)

The exponential terms can be factored as follows:

$$X_{q+1} = e^{j2\pi k/N} \left[ \sum_{n=0}^{N-1} x_{p+q} e^{-j2\pi pk/N} + x_{q+N} e^{-j2\pi Nk/N} - x_q \right]$$
 (5)

The  $e^{-j2\pi Nk/N}$  term simplifies to 1+j0 for k is always integer values, since  $e^{-j2\pi Nk/N} = 1$ , leading to:

$$X_{q+1} = e^{j2\pi k/N} \left[ \sum_{p=0}^{N-1} x_{p+q} e^{-j2\pi pk/N} + x_{q+N} - x_q \right]$$

$$= e^{j2\pi k/N} \left[ \sum_{n=0}^{N-1} x_{n+q} e^{-j2\pi nk/N} + x_{q+N} - x_q \right]$$

$$= e^{j2\pi k/N} \left[ X_q + x_{q+N} - x_q \right]$$
(6a)
$$= e^{j2\pi k/N} \left[ X_q + x_{q+N} - x_q \right]$$
(6b)

$$= e^{j2\pi k/N} \left[ \sum_{n=0}^{N-1} x_{n+q} e^{-j2\pi nk/N} + x_{q+N} - x_q \right]$$
 (6b)

$$= e^{j2\pi k/N} [X_q + x_{q+N} - x_q]$$
 (6c)

Note that the summation enclosed in square brackets in Eq. (6a) represents 20 the DFT calculated for the kth component, using p as the indexing variable 21 rather than n. For the latest timestamp t, the DFT results from the current slid-22 ing window  $(x_{t-N+1},\ldots,x_t)$  and the previous sliding window  $(x_{t-N},\ldots,x_{t-1})$ are denoted as  $\mathcal{F}_t$  and  $\mathcal{F}_{t-1}$ , respectively. This notation allows us to succinctly express the DFT update formula, transitioning from  $\mathcal{F}_{t-1}$  to  $\mathcal{F}_t$  as follows:

$$\mathcal{F}_t(k) = e^{j2\pi k/N} [\mathcal{F}_{t-1}(k) + x_t - x_{t-N}]$$
(7)

#### Spectral Peak Location Estimation 2

Spectral peak location estimation interpolates index  $k_{peak}$ , which corresponds 27 to the largest power in the Fourier transform result without an increase in N28 [6]. The  $k_{peak}$  is determined by  $k_{peak} = k + \delta$ , where k denotes the index of 29 the peak location from  $\mathcal{P}(k)$ , and  $\delta$  denotes the residual frequency that can be positive or negative, as shown in Fig. 1. Spectral peak location estimation base 31 on the cuvre-fiting technique. This supplementary report presents a comparison between the non-estimator and the hybrid Aboutanios-Mulgrew and q-shift 33 estimator (HAQSE), which is utilized in our ASTD.

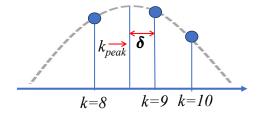


Fig. 1: Example of DFT coefficients resolution issue.

## **2.1 HAQSE**

- In this study, we utilize the HAQSE to determine  $k_{peak}$  from  $\mathcal{P}(k)$  [8]. HAQSE is iterative estimator that operates at a computational cost of O(N). The HAQSE algorithm processes the results of  $\mathcal{P}(k)$  as follows:
- <sup>39</sup> 1. **Peak Identification**: The peak index  $(\hat{k})$  from  $\mathcal{P}(k)$  is identified using  $\hat{k} = \operatorname{argmax}_k(\mathcal{P}(k))$ .
- 2. Initial  $\delta_{\alpha}$  Calculation: The initial value of  $\delta_{\alpha}$  is calculated according to:

$$\delta_{\alpha} = \frac{N}{2\pi} \arcsin\left(\sin\left(\frac{\pi}{N}\right) \Re\left\{\frac{X_{\hat{k}+0.5} + X_{\hat{k}-0.5}}{X_{\hat{k}+0.5} - X_{\hat{k}-0.5}}\right\}\right),\tag{8}$$

- where  $\Re\{\cdot\}$  denotes the real part, and  $X_k$  denotes Fourier transfrom for non-integer k values to accommodate fine-grained frequency estimation.
- 3. Final  $\delta$  Estimation: The final  $\delta$  is estimated using:

$$\delta = \frac{1}{c(q)} \left( \Re \left\{ \frac{X_{\hat{k} + \delta_{\alpha} + q} + X_{\hat{k} + \delta_{\alpha} - q}}{X_{\hat{k} + \delta_{\alpha} + q} - X_{\hat{k} + \delta_{\alpha} - q}} \right\} \right) + \delta_{\alpha}, \tag{9}$$

- with  $q = \frac{1}{\sqrt[3]{N}}$  and  $c(q) = \frac{1 \pi q \cot(\pi q)}{q \cos^2(\pi q)}$
- 4. Frequency Estimation:  $k_{peak}$  is estimated as  $\hat{k} + \delta$ . The actual frequency, which is the peak value of  $\mathcal{P}(k)$ , is computed as  $f_{peak} = k_{peak}/N$ .
- Here,  $X_k = \sum_{n=0}^{N-1} x_n \exp(-j2\pi nk/N)$  extends k to non-integer values, including  $\hat{k} \pm 0.5$ , and  $\hat{k} + \delta_{\alpha} \pm q$ , enabling HAQSE to estimate the actual frequency with higher resolution than possible with DFT's integer frequency bins. This method efficiently identifies  $k_{peak}$  by leveraging HAQSE's computational advantages, notably its O(N) computation cost, which is attributed to the transformation from the time domain to the frequency domain using the twiddle factor in steps 2 and 3.

### 55 2.2 Comparison between None-estimator and HAQSE

- We conducted an evaluation with a synthetic data set consisting of a sine wave with a season length of 50 instances and a total length of 500 instances. The evaluation process began with calculating  $\mathcal{P}(k)$  using the data within a window, followed by identifying  $\hat{k} = \operatorname{argmax}_k(\mathcal{P}(k))$ . This evaluation structured the analysis into two distinct groups.
  - 1. Non-estimator: we take the reciprocal with  $\hat{k}/N$  to get the season length.
- 62 2. **HAQSE estimator**: we find  $k_{peak}$  using HAQSE estimator. Then, we take the reciprocal with  $k_{peak}/N$  to get the season length.

65

66

67

68

70

71

72

73

The results are shown in Fig. 2, where the x-axis represents the window sizes and the y-axis represents the season length results provided by the estimator. Notably, N is considered to be the optimal window size for accurately determining  $k_{peak}$  if N divided by a positive integer  $k_{peak}$  equals 50. This condition ensures the most accurate determination of the peak frequency without the estimator.

The season length determined by the non-estimator is unstable and often diverges significantly from the ground truth. However, using the optimal window size can provide the correct season length, which exactly matches the ground truth. HAQSE exhibited stable results without the influence of the window size. Therefore, we used HAQSE to avoid the problem of the influence of the window

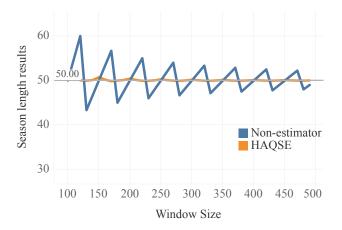


Fig. 2: Utilizing HAQSE estimator

#### Experimental Metrics for Real-world Datasets 3

This section provides additional details on the metrics used in our experiment 76 on real-world datasets. 77

**Trend Smoothness:** Measures the smoothness of the trend component via the standard deviation of its first-order difference [7]. Given the trend component de-79 noted by  $T = (T_0, T_1, \dots, T_t)$ , the first-order difference of the trend component, 80  $(\Delta T_i)$ , is calculated as: 81

$$\Delta T_i = T_{i+1} - T_i \quad \text{for} \quad i = 0, 1, \dots, t - 1,$$
 (10)

where t denotes the latest timestamp, indicating the total length of the trend component data. The smoothness measure, denoted as  $(\sigma_{\Delta T})$ , is then the standard deviation of these first-order differences:

$$\sigma_{\Delta T} = \sqrt{\frac{1}{t - 1} \sum_{i=0}^{t-1} (\Delta T_i - \mu_{\Delta T})^2}$$
 (11)

where  $\mu_{\Delta T}$  is the mean of the first-order differences of trend component. Lower values of  $\sigma_{\Delta T}$  idenote smoother trends.

Seasonality Presence: Measures the presence of seasonality by applying the Kruskal-Wallis test to the seasonal component [2]. Given the seasonal component denoted by  $S = (S_0, S_1, \ldots, S_t)$ , the Kruskal-Wallis test statistic is calculated as:

$$W = \frac{12}{N(N+1)} \sum_{j=1}^{g} \frac{U_j^2}{n_j} - 3(N+1)$$
 (12)

where N denotes the length of S, g denotes the number of groups,  $n_j$  denotes the number of observations in the j-th group, and  $U_j$  denotes the sum of ranks in the j-th group. To determine the number of groups (g), it is set equal to the season length m, which reflects the position within the cycle [2]. For example, if we have monthly data spanning one year (with a season length of 12), we group the data into 12 groups, with each group corresponding to one month within the cycle. Therefore, we group the observations by month, starting with January as the first group, February as the second, and so on until December, which is the twelfth group. This aligns with the season length m.

To illustrate the calculation of the sum of ranks  $(U_j)$  within each group for the Kruskal-Wallis test, consider an example with three groups, resulting in each group having its unique set of data points:

91

92

93

100

107

111

115

Ranks are assigned to the original observations within each group, and  $U_j$ , the sum of ranks in the j-th group, is calculated:

```
\begin{array}{lll} & -U_1 \ {\rm for \ Group \ 1: 5+3+8=16} \\ & -U_2 \ {\rm for \ Group \ 2: 7+6+2=15} \\ & -U_3 \ {\rm for \ Group \ 3: \ 4+9+1=14} \end{array}
```

Thus,  $U_j$  denotes the sum of ranks within each group. The values are  $U_1 = 16$ ,  $U_2 = 15$ , and  $U_3 = 14$ . After calculating the Kruskal-Wallis test statistic W, it is compared against a chi-square distribution with g-1 degrees of freedom. The resulting p-value is used to determine the statistical significance of the observed test statistic. Lower values suggest stable repeating cycles, indicating consistent seasonality, whereas higher values may indicate inconsistency in the seasonal component.

123

124

126

127

128

Randomness: Measures randomness in the residual component by applying the Ljung-Box test to the residual component [3, 2]. Given the residual component denoted by  $R = (R_0, R_1, \ldots, R_t)$ , the Ljung-Box test statistic is calculated as:

$$Q = N(N+2) \sum_{k=1}^{h} \frac{\hat{\rho}_k^2}{N-k}$$
 (13)

where N denotes the length of R, h is the number of lags being tested, and  $\hat{\rho}_k$  is the autocorrelation at lag k.  $\hat{\rho}_k$  is calculated as:

$$\hat{\rho}_k = \frac{\sum_{i=0}^{N-1-k} (R_i - \mu_R)(R_{i+k} - \mu_R)}{\sum_{i=0}^{N-1} (R_i - \mu_R)^2}$$
(14)

where  $\mu_R$  denotes the the mean of the residual component (R). To determine h, it is set to  $\min(2m, N/5)$ , where m is the season length [3]. Lower values suggest that the residual component originates from independent and identically distributed (iid) data, indicating the successful extraction of the seasonal component.

Note that the source code for the Kruskal–Wallis and Ljung-Box tests is available in 'evaluation/02\_Real1\_dataset/' [1].

## References

- 131 1. Supplementary website, https://sites.google.com/view/astd-ecmlpkdd
- Bee Dagum, E., Bianconcini, S.: Linear Filters Seasonal Adjustment Methods: Census Method II and Its Variants, pp. 79–114. Springer International Publishing, Cham
   (2016)
- Hyndman, R., Athanasopoulos, G.: Forecasting: principles and practice, 2nd edition.
   OTexts (2018)
- 4. Jacobsen, E., Lyons, R.: The sliding DFT. IEEE Signal Processing Magazine 20(2),
   74–80 (2003)
- 5. Jacobsen, E., Lyons, R.: An update to the sliding DFT. IEEE Signal Processing
   Magazine 21(1), 110-111 (2004)
- 6. Jacobsen, E., Kootsookos, P.: Fast, accurate frequency estimators. IEEE Signal Processing Magazine 24(3), 123–125 (2007)
- 7. Mishra, A., Sriharsha, R., Zhong, S.: OnlineSTL: Scaling time series decomposition
   by 100x. VLDB 15(7), 1417–1425 (2022)
- 8. Serbes, A.: Fast and efficient sinusoidal frequency estimation by using the DFT coefficients. IEEE Transactions on Communications **67**(3), 2333–2342 (2019)