1

Digital Signal Processing Assignment 1

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CONTENTS

1	Software Installation	1
2	Digital Filter	1
3	Difference Equation	2
4	Z-transform	3
5	Impulse Response	6
6	DFT and FFT	10
7	Exercises	10

Abstract—This manual is the soution of the assignment of the course EE3900 assignment 1.

1 Software Installation

Run the following commands

sudo apt-get update sudo apt-get install libffi-dev libsndfile1 python3 -scipy python3-numpy python3-matplotlib sudo pip install cffi pysoundfile

2 Digital Filter

2.1 Download the sound file from

wget https://raw.githubusercontent.com/ gadepall/ EE1310/master/filter/codes/Sound Noise.way

2.2 You will find a spectrogram at https: //academo.org/demos/spectrum-analyzer. Upload the sound file that you downloaded in Problem 2.1 in the spectrogram and play. Observe the spectrogram. What do you find?

Solution: There are a lot of yellow lines between 440 Hz to 5.1 KHz. These represent the synthesizer key tones. Also, the key strokes are audible along with background noise.

2.3 Write the python code for removal of out of band noise and execute the code.

```
Solution:
import soundfile as sf
from scipy import signal
#read .wav file
input signal,fs = sf.read('Sound Noise.wav'
#sampling frequency of Input signal
sampl freq=fs
#order of the filter
order=4
#cutoff frquency 4kHz
cutoff freq=4000.0
#digital frequency
Wn=2*cutoff freq/sampl freq
# b and a are numerator and denominator
   polynomials respectively
b, a = signal.butter(order, Wn, 'low')
#filter the input signal with butterworth filter
output signal = signal.filtfilt(b, a,
   input signal)
#output \ signal = signal.lfilter(b, a,
   input signal)
#write the output signal into .wav file
sf.write('Sound With ReducedNoise.wav',
   output signal, fs)
```

2.4 The output of the python script Problem 2.3 in is the audio file Sound With ReducedNoise.wav. Play the file in the spectrogram in Problem 2.2. What do you observe?

Solution: The key strokes as well as background noise is subdued in the audio. Also, the signal is blank for frequencies above 5.1 kHz.

3 DIFFERENCE EQUATION

3.1 Let

$$x(n) = \left\{ \begin{array}{l} 1, 2, 3, 4, 2, 1 \\ \uparrow \end{array} \right\} \tag{3.1}$$

Sketch x(n).

Solution:

import numpy as np **import** matplotlib.pyplot as plt #If using termux #import subprocess #import shlex #end if x=np.array([1.0,2.0,3.0,4.0,2.0,1.0])k = 20y = np.zeros(20)#subplots plt.subplot(2, 1, 1)plt.stem(range(0,6),x) plt.title('Digital_Filter_Input-Output') plt.ylabel('\$x(n)\$')plt.grid()# minor plt.show()

3.2 Let

$$y(n) + \frac{1}{2}y(n-1) = x(n) + x(n-2),$$

$$y(n) = 0, n < 0 \quad (3.2)$$

Sketch y(n).

Solution: The following code yields Fig. 5.3.

```
import numpy as np
import matplotlib.pyplot as plt
#If using termux
#import subprocess
```

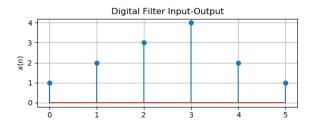


Fig. 3.1

```
#import shlex
#end if
x=np.array([1.0,2.0,3.0,4.0,2.0,1.0])
k = 20
y = np.zeros(20)
y[0] = x[0]
y[1] = -0.5*y[0]+x[1]
for n in range(2,k-1):
        if n < 6:
                 y[n] = -0.5*y[n-1]+x[n]+x
                     [n-2]
        elif n > 5 and n < 8:
                 y[n] = -0.5*y[n-1]+x[n-2]
        else:
                 y[n] = -0.5*y[n-1]
print(y)
#subplots
plt.subplot(2, 1, 1)
plt.stem(range(0,6),x)
plt.title('Digital_Filter_Input-Output')
plt.ylabel('$x(n)$')
plt.grid()# minor
plt.subplot(2, 1, 2)
```

```
plt.stem(range(0,k),y)
plt.xlabel('$n$')
plt.ylabel('$y(n)$')
plt.grid()# minor

#If using termux

#else
plt.show()
```

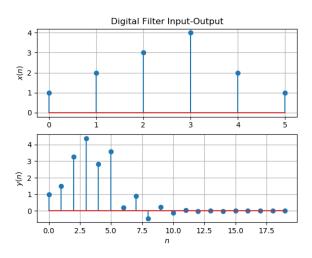


Fig. 3.2

3.3 Repeat the above exercise using a C code. **Solution:**

```
#include <stdio.h>
// A function which calculates y[i] takings
    array x and and i as parameters
float calculatey(int* p1,int n){
        int y=0;
         if(n==0){y = *p1;}
         else if(n==1){y=*(p1+1)+calculatey(
             p1,0);
         else if(n<6)\{y = *(p1+n-1) + *(p1+n-1)\}
             n-3) - calculatey(p1,n-1)/2.0;}
    else \{y = -\text{calculatey}(p_1, n-1)/2.0;\}
         return y;
int main(){
         FILE* fp;
    const int number Of Points=20;
    int x[] = \{1,2,3,4,2,1\};
    float y[number Of Points];
```

```
int i;
for(int i=0;i<number_Of_Points;i++){
        y[i]=calculatey(x,i);}

        // opening file
    fp = fopen("y(n).dat","w");

        // writing to the file
    for(i=0;i<number_Of_Points;i++){
        fprintf(fp,"%f\n",y[i]);
    }

        // closing file
    fclose(fp);
    return 0;
}</pre>
```

For plotting the data we can use the python code below

```
import matplotlib.pyplot as plt
import numpy as np
y = np.loadtxt("y(n).dat",dtype = "double")
x = np.array([1,2,3,4,2,1])
# ploting graphs
plt.subplot(211)
plt.stem(np.arange(len(x)),x)
plt.xlabel("n")
plt.ylabel("$x(n)$")
plt.grid()
plt.subplot(212)
plt.stem(np.arange(len(y)),y)
plt.xlabel("n")
plt.ylabel("$y(n)$")
plt.grid()
plt.show()
```

4 Z-TRANSFORM

4.1 The Z-transform of x(n) is defined as

$$X(z) = \mathcal{Z}\{x(n)\} = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
 (4.1)

Show that

$$Z{x(n-1)} = z^{-1}X(z)$$
 (4.2)

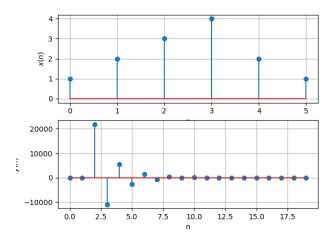


Fig. 3.3

and find

$$\mathcal{Z}\{x(n-k)\}\tag{4.3}$$

Solution: From (4.1),

$$\mathcal{Z}\{x(n-k)\} = \sum_{n=-\infty}^{\infty} x(n-1)z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} x(n)z^{-n-1} = z^{-1} \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$
(4.4)

resulting in (4.2). Similarly, it can be shown that

$$\mathcal{Z}\{x(n-k)\} = z^{-k}X(z) \tag{4.6}$$

4.2 Obtain X(z) for x(n) defined in problem 3.1. **Solution:**

Given that

$$x(n) = \left\{ 1, 2, 3, 4, 2, 1 \right\} \tag{4.7}$$

$$X(z) = \sum_{n=-\infty}^{\infty} x(n)z^{-n}$$

$$= z^{-0} + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + 1z^{-5}$$

$$= 1 + 2z^{-1} + 3z^{-2} + 4z^{-3} + 2z^{-4} + 1z^{-5}$$

$$(4.10)$$

4.3 Find

$$H(z) = \frac{Y(z)}{X(z)}$$
 (4.11)

from (3.2) assuming that the Z-transform is a linear operation.

Solution: Applying (4.6) in (3.2),

$$Y(z) + \frac{1}{2}z^{-1}Y(z) = X(z) + z^{-2}X(z)$$
 (4.12)

$$\implies \frac{Y(z)}{X(z)} = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}} \tag{4.13}$$

4.4 Find the Z transform of

$$\delta(n) = \begin{cases} 1 & n = 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.14)

and show that the Z-transform of

$$u(n) = \begin{cases} 1 & n \ge 0 \\ 0 & \text{otherwise} \end{cases}$$
 (4.15)

is

$$U(z) = \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.16}$$

Solution: It is easy to show that

$$\delta(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} 1 \tag{4.17}$$

and from (4.15),

$$U(z) = \sum_{n=0}^{\infty} z^{-n}$$
 (4.18)

$$= \frac{1}{1 - z^{-1}}, \quad |z| > 1 \tag{4.19}$$

using the fomula for the sum of an infinite geometric progression.

4.5 Show that

$$a^n u(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} \frac{1}{1 - az^{-1}} \quad |z| > |a|$$
 (4.20)

Solution:

Given that,

$$f(n) = a^n u(n) \tag{4.21}$$

and u(n) is a unit step function.

Taking the Z transform of f(n) we get,

$$U(z) = \sum_{n = -\infty}^{\infty} a^n u(n) z^{-n}$$
 (4.22)

$$=\sum_{n=-\infty}^{\infty} a^n z^{-n} \tag{4.23}$$

$$= 1 + az^{-1} + a^2z^{-2} + \dots (4.24)$$

$$= \frac{1}{1 - az^{-1}}, \quad \left| az^{-1} \right| < 1 \tag{4.25}$$

$$= \frac{1}{1 - az^{-1}}, \quad |a| < |z| \tag{4.26}$$

4.6 Let

$$H(e^{j\omega}) = H(z = e^{j\omega}).$$
 (4.27)

Plot $|H(e^{j\omega})|$. Is it periodic? If so, find the period. $H(e^{j\omega})$ is known as the *Discret Time Fourier Transform* (DTFT) of x(n).

Solution: The following code plots Fig. 4.6.

import numpy as np

import matplotlib.pyplot as plt

#DTFT

def H(z):

num = np.polyval([1,0,1],z**(-1)) den = np.polyval([0.5,1],z**(-1))

H = num/den

return H

#Input and Output

omega = np.linspace(-3*np.pi,3*np.pi,1e2)

#subplots

plt.plot(omega, **abs**(H(np.exp(1j*omega)))) plt.title('Filter_Frequency_Response')

plt.xlabel('\$\omega\$')

 $plt.ylabel(`\$|H(e^{\star (\mbox{\mbox{$\mbox{$\mbox{$}}}})|_\$')$

plt.grid()

plt.show()

We know that

$$H(z) = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (4.28)

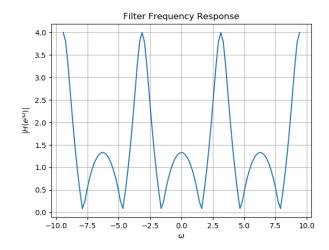


Fig. 4.6: $|H(e^{j\omega})|$

then,

$$H(e^{jw}) = \frac{1 + e^{-2jw}}{1 + \frac{1}{2}e^{-jw}}$$
(4.29)

$$=\frac{1+e^{2jw}}{e^{2jw}+\frac{1}{2}e^{jw}}\tag{4.30}$$

$$\implies |H(e^{jw})| = \left| \frac{1 + e^{2jw}}{e^{2jw} + \frac{1}{2}e^{jw}} \right|$$
 (4.31)

$$= \frac{\left|1 + e^{2jw}\right|}{\left|e^{2jw} + \frac{1}{2}e^{jw}\right|}$$

$$= \frac{\left|\cos 2w + 1 + 2j\sin w\right| * 2}{\left|e^{jw} + 1\right|}$$
(4.32)

$$=\frac{|cosw + jsinw| * 4cosw}{|2cosw + 1 + 2jsinw|}$$
(4.34)

(4.33)

$$= \frac{4\cos w}{\sqrt{(2\cos w + 1)^2 + 4\sin^2 w}}$$
(4.35)

$$=\frac{4\cos w}{\sqrt{5+4\cos w}}\tag{4.36}$$

Therefore

$$H(e^{jw}) = \frac{4\cos w}{\sqrt{5 + 4\cos w}} \tag{4.37}$$

Consider

$$f(x) = \frac{4\cos x}{\sqrt{5 + 4\cos x}} \tag{4.38}$$

then

$$f(x+t) = \frac{4\cos(x+t)}{\sqrt{5+4\cos(x+t)}}$$
 (4.39)

$$f(x+t) = \frac{4\cos(x+t)}{\sqrt{5+4\cos(x+t)}}$$
 (4.40)

$$f(x+t) = \frac{4\cos(x+t)}{\sqrt{5+4\cos(x+t)}}$$
(4.40)
=
$$\frac{4(\cos x \cos t - \sin x \sin t)}{\sqrt{5+4(\cos x \cos t - \sin x \sin t)}}$$
(4.41)

By comparing (4.38) and (4.39), we get cost=1 and sint=0

This is true for $t = 2k\pi$. This implies that the principal period of this function is 2π .

4.7 Express h(n) in terms of $H(e^{j\omega})$.

Solution: We know that

$$H(e^{j\omega}) = \sum_{n=-\infty}^{\infty} h(n) e^{-jn\omega}$$

$$\Rightarrow (e^{j\omega}) * e^{jk\omega} = \sum_{n=-\infty}^{\infty} h(n) e^{-jn\omega} e^{jk\omega}$$

$$(4.43)$$

$$\Rightarrow \int_{-\pi}^{\pi} H(e^{j\omega}) * e^{jk\omega} d\omega = \int_{-\pi}^{\pi} \sum_{n=-\infty}^{\infty} h(n) e^{-jn\omega} e^{jk\omega}$$

$$(4.44)$$

$$\Rightarrow \int_{-\pi}^{\pi} H(e^{j\omega}) * e^{jk\omega} d\omega = \sum_{n=-\infty}^{\infty} h(n) \int_{-\pi}^{\pi} e^{-jn\omega} e^{jk\omega}$$

$$(4.45)$$

NOTE: We know that,

$$\int_{-\pi}^{\pi} e^{-jn\omega} e^{jk\omega} = \begin{cases} 2\pi & n = k \\ 0 & \text{otherwise} \end{cases}$$
 (4.46)

Hence,

$$\int_{-\pi}^{\pi} H\left(e^{j\omega}\right) * e^{jk\omega} d\omega = \int_{-\pi}^{\pi} \sum_{n=-\infty}^{\infty} h\left(n\right) e^{-jn\omega} e^{jk\omega}$$

$$\iff \int_{-\pi}^{\pi} H\left(e^{j\omega}\right) * e^{jk\omega} d\omega = 2\pi h\left(n\right)$$

$$\iff \frac{1}{2\pi} \int_{-\pi}^{\pi} H\left(e^{j\omega}\right) * e^{jk\omega} d\omega = h\left(n\right)$$

$$\tag{4.48}$$

$$\iff \frac{1}{2\pi} \int_{-\pi}^{\pi} H\left(e^{j\omega}\right) * e^{jk\omega} d\omega = h\left(n\right)$$

$$\tag{4.49}$$

Therefore,

$$h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} H\left(e^{j\omega}\right) e^{j\omega n} d\omega \qquad (4.50)$$

5 IMPULSE RESPONSE

5.1 Using long division, find

$$h(n), \quad n < 5$$
 (5.1)

for H(z) in (4.13).

Solution:

$$H(z) = \frac{1 + z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.2)

Let $z^{-1} = x$, then, by polynomial long division

$$\frac{2x-4}{x^2+1}$$

$$\frac{-x^2-2x}{-2x+1}$$

$$\frac{2x+4}{5}$$

$$\implies (1+z^{-2}) = (\frac{1}{2}z^{-1}+1)(2z^{-1}-4)+5$$

$$\implies \frac{(1+z^{-2})}{\frac{1}{2}z^{-1}+1} = (2z^{-1}-4)+\frac{5}{\frac{1}{2}z^{-1}+1}$$

$$(5.4)$$

$$\implies H(z) = (2z^{-1}-4)+\frac{5}{\frac{1}{2}z^{-1}+1}$$

$$(5.5)$$

Now, consider $\frac{5}{\frac{1}{2}z^{-1}+1}$

The denominator $\frac{1}{2}z^{-1} + 1$ can be expressed as sum of an infinite geometric progression, which as its first term equal to 1 and common ratio $\frac{-1}{2}z^{-1}$

we can write $\frac{5}{\frac{1}{3}z^{-1}+1}$ $5\left(1+\left(\frac{-1}{2}z^{-1}\right)+\left(\frac{-1}{2}z^{-1}\right)^{2}+\left(\frac{-1}{2}z^{-1}\right)^{3}+\left(\frac{-1}{2}z^{-1}\right)^{4}+\ldots\right)$ herefore, H(z) can be given by

$$H(z) = (2z^{-1} - 4) + \frac{5}{\frac{1}{2}z^{-1} + 1}$$
 (5.6)

$$= 2z^{-1} - 4 + 5 + \frac{-5}{2}z^{-1} + \frac{5}{4}z^{-2} + \frac{-5}{8}z^{-3} + \frac{5}{16}z^{-4} + .$$

$$\implies H(z) = 1z^{0} + \frac{-1}{2}z^{-1} + \frac{5}{4}z^{-2} + \frac{-5}{8}z^{-3} + \frac{5}{16}z^{-4} + .$$
(5.9)

Comparing the above expression to (4.1) we get h(n) for n<5 as,

$$h(0) = 1 \tag{5.10}$$

$$h(1) = \frac{-1}{2} \tag{5.11}$$

$$h(2) = \frac{5}{4} \tag{5.12}$$

$$h(3) = \frac{-5}{8} \tag{5.13}$$

$$h(4) = \frac{5}{16} \tag{5.14}$$

5.2 Find an expression for h(n) using H(z), given that

$$h(n) \stackrel{\mathcal{Z}}{\rightleftharpoons} H(z) \tag{5.15}$$

and there is a one to one relationship between h(n) and H(z). h(n) is known as the *impulse response* of the system defined by (3.2).

Solution: From (4.13),

$$H(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{z^{-2}}{1 + \frac{1}{2}z^{-1}}$$
 (5.16)

$$\implies h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2)$$
(5.17)

using (4.20) and (4.6).

5.3 Sketch h(n). Is it bounded? Justify theoretically.

Solution: The following code plots Fig. 5.3.

import numpy as np
import matplotlib.pyplot as plt

n = np.arange(10)

$$fn=(-1/2)**n$$

hn1=np.pad(fn, (0,2), 'constant', constant values=(0))

hn2=np.pad(fn, (2,0), 'constant',

constant_values=(0))
plt.stem(np.arange(12), hn1+hn2)
plt.title('Filter_Impulse_Response')
plt.xlabel('\$n\$')
plt.ylabel('\$h(n)\$')
plt.grid()# minor

plt.show()

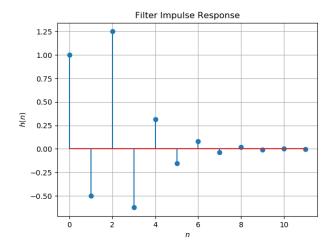


Fig. 5.3

From (5.17) we know that

$$h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2) \quad (5.18)$$

Implies we can write that

$$h(n) = \begin{cases} 0 & , n < 0 \\ \left(\frac{-1}{2}\right)^n & , 0 \le n < 2 \\ 5\left(\frac{-1}{2}\right)^n & , n \ge 2 \end{cases}$$
 (5.19)

A sequence is said to be bounded when

$$|x_n| \le M, \forall n \in \mathcal{N} \tag{5.20}$$

Now consider (5.19),

For n < 0,

$$|h(n)| \le 0 \tag{5.21}$$

For $0 \le n < 2$,

$$|h(n)| = (\frac{1}{2})^n$$
 (5.22)

$$\implies |h(n)| \le 1 \tag{5.23}$$

For $n \geq 2$,

$$|h(n)| = 5(\frac{1}{2})^n$$
 (5.24)

$$\implies |h(n)| \le 5 \tag{5.25}$$

From above we can say that,

$$M = \max\{0, 1, 5\} \tag{5.26}$$

$$= 5 \tag{5.27}$$

Therefore since M exists and is a real value, we can say that h(n) is bounded.

5.4 Convergent? Justify using the ratio test.

Solution:

We can check if a sequence os convergent by ratio test. From the defination of ratio test we can say that a sequence is convergent if

$$\lim_{n \to \infty} \left| \frac{x_{n+1}}{x_n} \right| < 1 \tag{5.28}$$

Here, applying ratio test on (5.17) is same as applying ratio test on (5.19)

$$\lim_{n \to \infty} \left| \frac{h(n+1)}{h(n)} \right| = \lim_{n \to \infty} \left| \frac{5(\frac{-1}{2})^{n+1}}{5(\frac{-1}{2})^n} \right|$$
 (5.29)

$$= \left| \frac{-1}{2} \right| \tag{5.30}$$

$$=\frac{1}{2}$$
 (5.31)

From (5.31) we can clearly say that the sequence h(n) is convergent.

5.5 The system with h(n) is defined to be stable if

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.32}$$

Is the system defined by (3.2) stable for the impulse response in (5.15)?

Solution:

From (5.17) we know that,

$$h(n) = \left(-\frac{1}{2}\right)^n u(n) + \left(-\frac{1}{2}\right)^{n-2} u(n-2) \quad (5.33)$$

Given that, a system is stable when

$$\sum_{n=-\infty}^{\infty} h(n) < \infty \tag{5.34}$$

$$\implies \sum_{n=-\infty}^{\infty} h(n) = \sum_{n=-\infty}^{\infty} \left(\left(-\frac{1}{2} \right)^n u(n) + \left(-\frac{1}{2} \right)^{n-2} u(n-2) \right)$$

$$= 2 * \sum_{n=-\infty}^{\infty} \left(-\frac{1}{2} \right)^n u(n)$$
(5.36)

$$\implies \sum_{n=-\infty}^{\infty} h(n) = 2 * \left(\frac{1}{1 + \frac{1}{2}}\right) \qquad (5.37)$$
$$= \frac{4}{3} < \infty \qquad (5.38)$$

Hence, the system is stable.

5.6 Verify the above result using a python code. **Solution:**

$$\begin{array}{l} \mbox{def } u(n): \\ & \mbox{if } n <\! 0: \\ & \mbox{return } 0.0 \\ & \mbox{else :} \\ & \mbox{return } 1.0 \\ \mbox{def } h(n): \\ & \mbox{return } u(n)*(-1.0/2)**n + u(n-2) \\ & \mbox{} *(-1.0/2)**(n-2) \\ \mbox{sum } = 0 \\ \mbox{for i in } \mbox{range}(200000): \\ & \mbox{sum } +=h(i) \\ \mbox{print(sum)} \end{array}$$

5.7 Compute and sketch h(n) using

$$h(n) + \frac{1}{2}h(n-1) = \delta(n) + \delta(n-2), \quad (5.39)$$

This is the definition of h(n).

Solution: The following code plots Fig. 5.7. Note that this is the same as Fig. ??.

import numpy as np
import matplotlib.pyplot as plt

k = 12
h = np.zeros(k)
h[0] = 1
h[1] = -0.5*h[0]
h[2] = -0.5*h[1] + 1

for n in range(3,k-1):

h[n] = -0.5*h[n-1]

```
#subplots
plt.stem(range(0,k),h)
plt.title('Impulse_Response_Definition')
plt.xlabel('$n$')
plt.ylabel('$h(n)$')
plt.grid()# minor

plt.show()
```

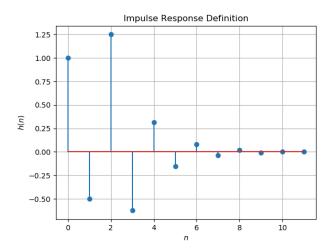


Fig. 5.7: h(n) from the definition

5.8 Compute

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.40)

Comment. The operation in (5.40) is known as *convolution*.

Solution: The following code plots Fig. 5.8. Note that this is the same as y(n) in Fig. 5.3.

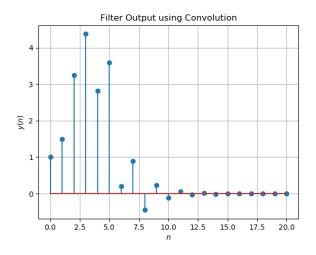


Fig. 5.8: y(n) from the definition of convolution

5.9 Express the above convolution using a Teoplitz matrix.

Solution:

And this is what we got in (5.40)

5.10 Show that

$$y(n) = \sum_{n = -\infty}^{\infty} x(n - k)h(k)$$
 (5.41)

Solution:

From the defination of convolution given in (5.40), we know that,

$$y(n) = x(n) * h(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k)$$
 (5.42)

$$\implies y(n) = \sum_{k=-\infty}^{\infty} x(k)h(n-k) \qquad (5.43)$$

Replace k with n-k. Now since n varies from $n=-\infty$ to $n=\infty$, n-k will also vary from $-\infty$ to ∞ . Therefore, we get,

$$y(n) = \sum_{n-k=-\infty}^{\infty} x(n-k)h(k)$$
 (5.44)

$$=\sum_{k=-\infty}^{\infty}x(n-k)h(k)$$
 (5.45)

6 DFT AND FFT

6.1 Compute

$$X(k) \stackrel{\triangle}{=} \sum_{n=0}^{N-1} x(n)e^{-j2\pi kn/N}, \quad k = 0, 1, \dots, N-1$$
(6.1)

and H(k) using h(n).

6.2 Compute

$$Y(k) = X(k)H(k) \tag{6.2}$$

6.3 Compute

$$y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) \cdot e^{j2\pi kn/N}, \quad n = 0, 1, \dots, N-1$$
(6.3)

Solution: The following code plots Fig. 5.8. Note that this is the same as y(n) in Fig. 5.3.

wget https://raw.githubusercontent.com/ gadepall/EE1310/master/**filter**/codes/yndft. py

- 6.4 Repeat the previous exercise by computing X(k), H(k) and y(n) through FFT and IFFT.
- 6.5 Wherever possible, express all the above equations as matrix equations.
- 6.6 Verify the above equations by generating the DFT matrix in python.

7 Exercises

Answer the following questions by looking at the python code in Problem 2.3.

7.1 The command

output_signal = signal.lfilter(b, a, input_signal)

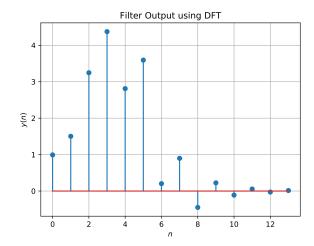


Fig. 6.3: y(n) from the DFT

in Problem 2.3 is executed through the following difference equation

$$\sum_{m=0}^{M} a(m) y(n-m) = \sum_{k=0}^{N} b(k) x(n-k) \quad (7.1)$$

where the input signal is x(n) and the output signal is y(n) with initial values all 0. Replace **signal.filtfilt** with your own routine and verify.

- 7.2 Repeat all the exercises in the previous sections for the above a and b.
- 7.3 What is the sampling frequency of the input signal?

Solution: Sampling frequency(fs)=44.1kHZ.

7.4 What is type, order and cutoff-frequency of the above butterworth filter

Solution: The given butterworth filter is low pass with order=2 and cutoff-frequency=4kHz.

7.5 Modifying the code with different input parameters and to get the best possible output.