Chapter 5

Time Response



Outline

- 1 Poles, Zeros, and System Response
- 2 First order systems
- 3 Second order systems
- 4 Time domain Solution of State equations



Poles of a Transfer Function

- The values of the Laplace transform variable,s, that cause the transfer function to infinite.
- Any roots of the denominator of transfer function that are common to roots of the numerator.



Zeros of a Transfer Function

- The values of the Laplace transform variable,s, that cause the transfer function to zero.
- Any roots of the numerator of transfer function that are common to roots of the denominator.



Unit step response of systems

$$R(s) = \frac{1}{s}$$

$$S + 2$$

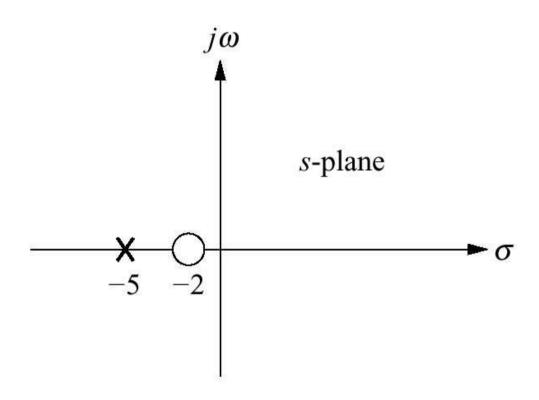
$$S + 5$$

$$C(s)$$

$$C(s) = \frac{(s+6)}{s(s+6)}$$



Pole-Zero plot of the system on s-plane





$$C(s) = \frac{(s+\omega)}{s(s+\alpha)} = \frac{A}{s} + \frac{B}{s+\alpha}$$

By using Partial Fraction Expansion

$$A = \left[\frac{(s+\omega)}{(s+\omega)} \right]_{s \to 0} = \frac{\omega}{\varepsilon}$$

$$B = \left[\frac{\left(s + \wp \right)}{s} \right]_{s \to \mathscr{E}} = \frac{\varpi}{\mathscr{E}}$$



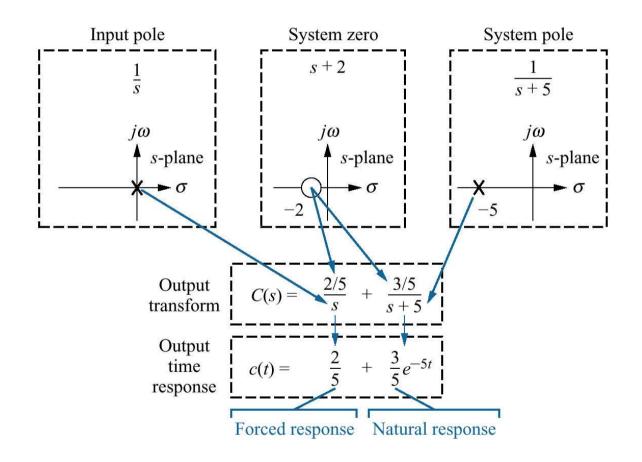
$$C(s) = \frac{60/6}{s} + \frac{60/6}{s+6}$$

By Inverse Laplace transform

$$c(t) = \frac{6}{6} + \frac{6}{6}e^{-6t}$$

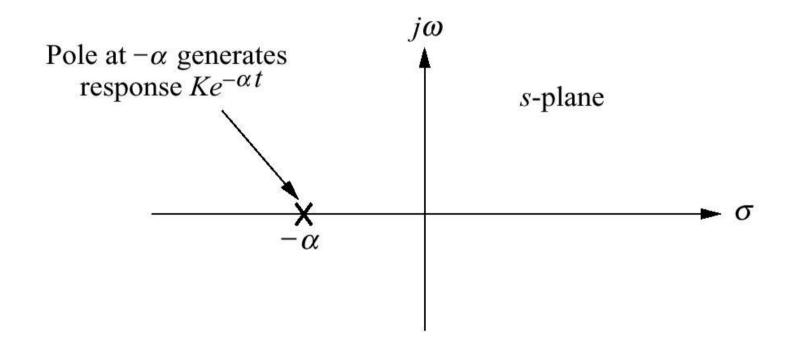


System response and response component generated by the pole or zero.



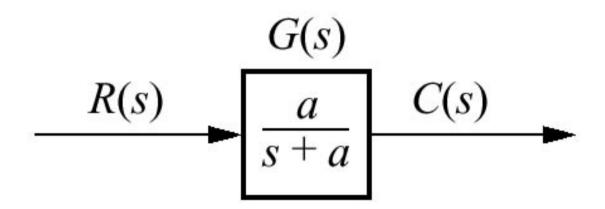


Effect of a real-axis pole upon transient response



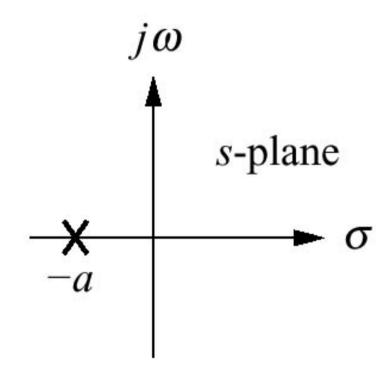


First-Order Systems





Pole plot of first order system





Unit-step response of First-Order System

$$C(s)=R(s)G(s)=\frac{a}{s(s+a)}$$

By Inverse Laplace Transform

$$c(t) = c_f(t) + c_n(t) = o - e^{-at}$$



Unit-step response of First-Order System

$$c(t) = \omega - e^{-at}$$

At
$$t = \omega/a$$
, ω/a and ω/a

$$c(t) = 0.5\sigma$$
, 0.65 and 0.66



Time response definition

Rise time is define as the time for the waveform to go from o.o to o.o of its final value.

$$T_r = \omega.\omega \omega/a - o.\omega\omega/a$$

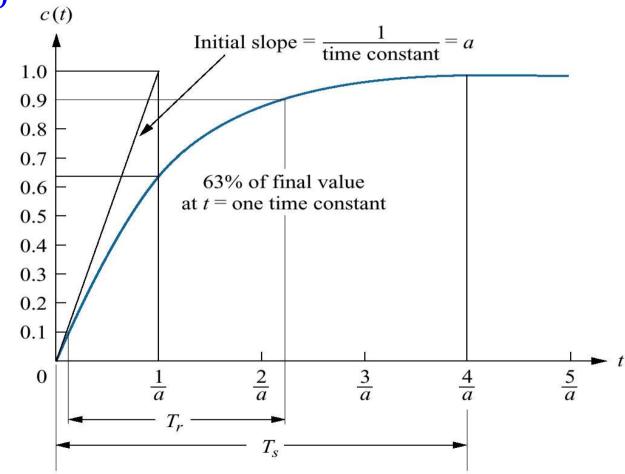
Settling time is the time for the response to reach, and stay within \omega_{\pi} of its final value.

$$T_s = \epsilon/a$$



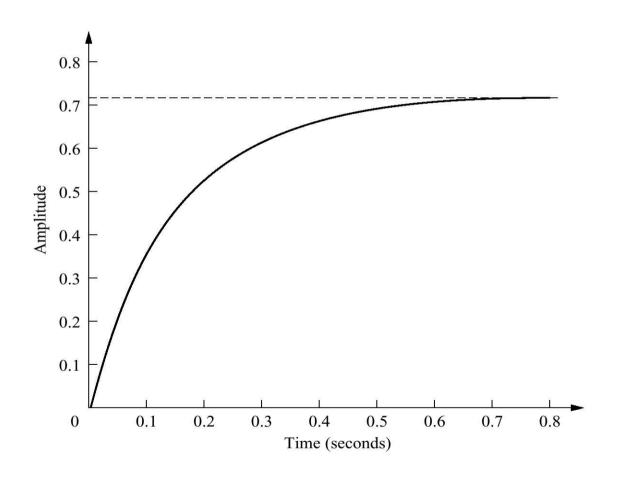
First-order system response to a unit

step



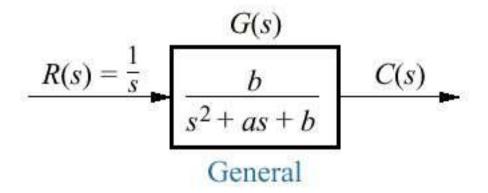


Laboratory results of a system unit step response test



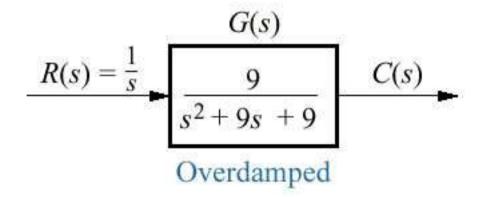


Second-Order Systems



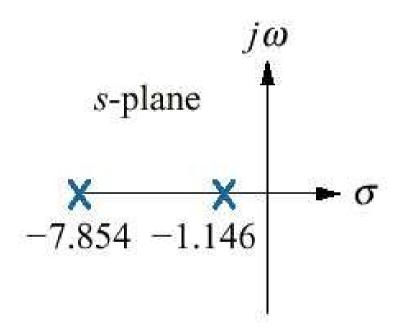


Overdamped Response





Poles plot of Overdamped System





Overdamped Response

$$C(s) = \frac{\alpha'}{s(s'+\alpha's+\alpha')}$$

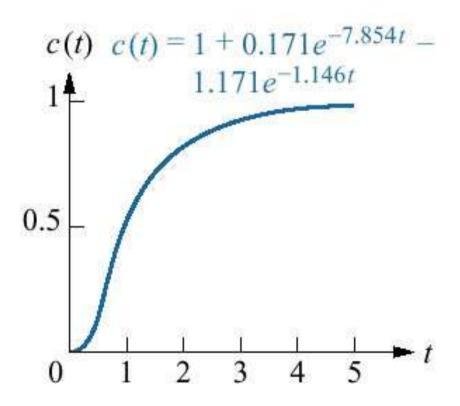
$$=\frac{\alpha}{s(s+\omega.\alpha\alpha)(s+\omega.\omega\alpha)}$$

By Inverse Laplace Transform

$$c(t) = K_{0} + K_{0}e^{-\alpha \cdot \alpha \cdot \alpha \cdot d} + K_{0}e^{-\alpha \cdot \alpha \cdot \alpha \cdot d}$$

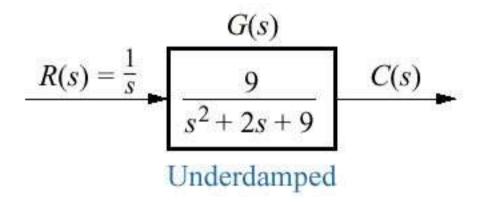


Overdamped Response



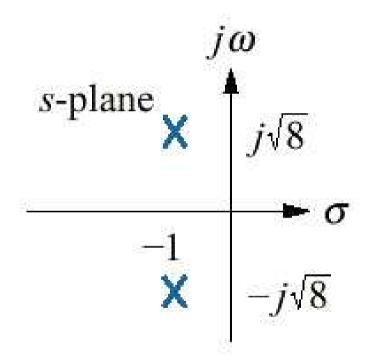


Underdamped Response





Poles plot of Underdamped Response





Underdamped Response

$$C(s) = \frac{\alpha}{s(s + \omega s + \alpha)} \qquad s = -\omega \pm j\sqrt{\alpha}$$

$$C(s) = \frac{K_{\odot}}{s} + \frac{K_{\odot}}{s - (\omega + j\sqrt{\omega})} + \frac{K_{\odot}}{s - (\omega - j\sqrt{\omega})}$$

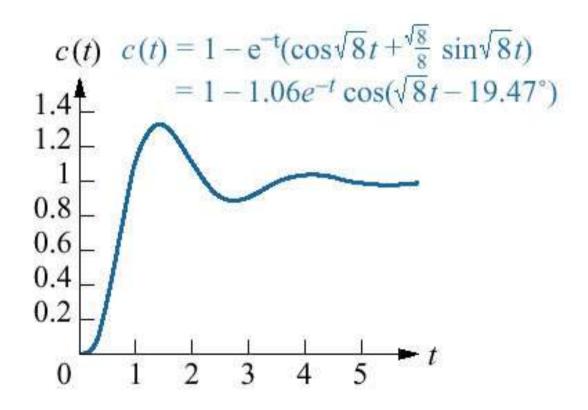
By Inverse Laplace Transform

$$c(t) = e^{-t} \left(\cos \sqrt{\epsilon} t + \frac{\sqrt{\epsilon}}{\epsilon} \sin \sqrt{\epsilon} t\right)$$

$$c(t) = \omega - \omega \cos^{-t} \cos(\sqrt{\kappa}t - \omega \kappa \cdot \omega)$$

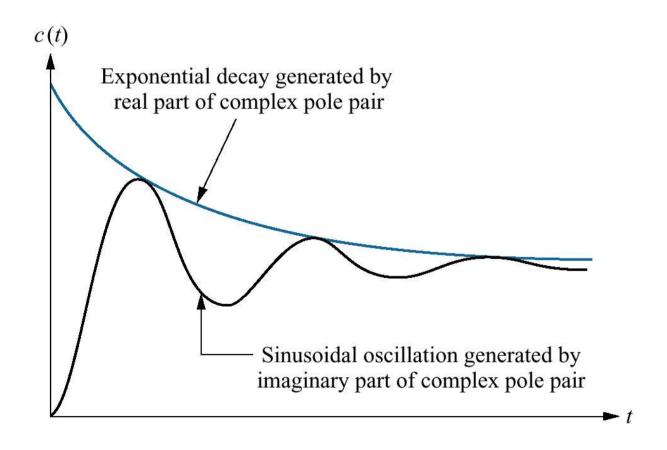


Underdamped Response



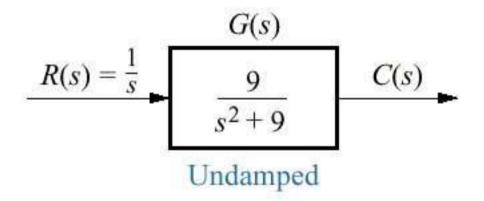


Second-order unit step response components generated by complex poles



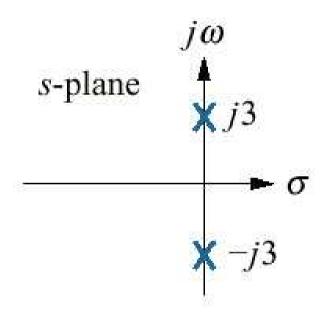


Undamped Response





Poles plot of Undamped Response





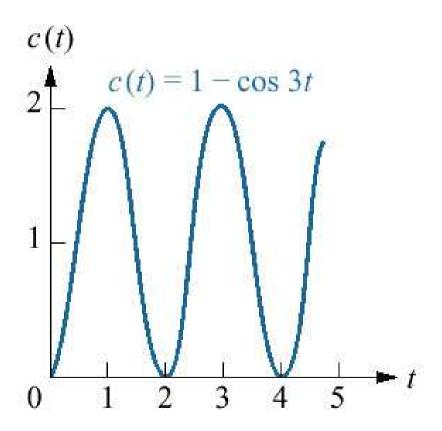
Undamped Response

$$C(s) = \frac{\alpha'}{s(s^{6} + \alpha')} \qquad s = \pm j\omega$$

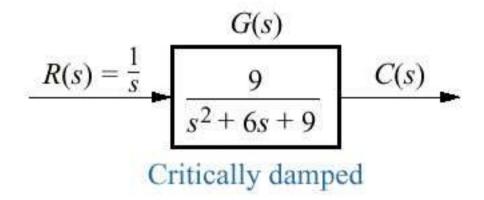
$$c(t) = K_{o} + K_{d} \cos(\omega t - \phi)$$



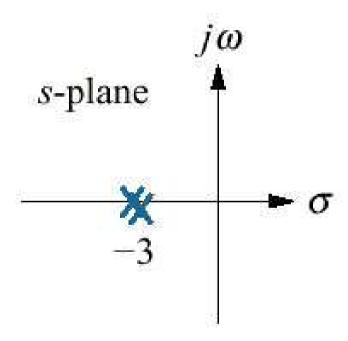
Undamped Response













$$C(s) = \frac{\alpha}{s(s^{6} + bs + \alpha)}$$

$$S = - c$$

$$c(t) = K_{\odot} + K_{\odot} e^{-\omega t} + K_{\varpi} t e^{-\omega t}$$

$$K_{\circ} = 0$$



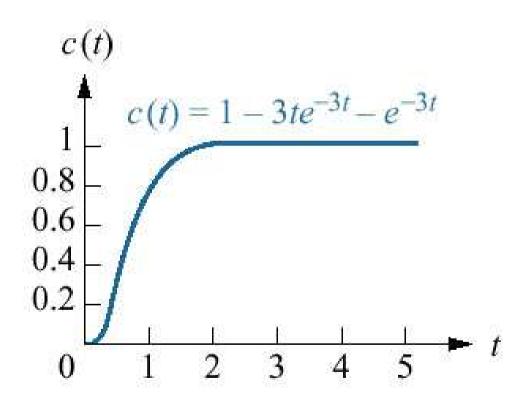
Two real poles at same point

$$c(t) = K_{\scriptscriptstyle o} e^{-\sigma_{\scriptscriptstyle o} t} + K_{\scriptscriptstyle b} t e^{-\sigma_{\scriptscriptstyle b} t}$$

Then

$$c(t) = 0 - cte^{-ct} - e^{-ct}$$







Natural responses of second order system

Overdamped response

Poles: two real at
$$-\sigma_{_{0}}$$
, $\sigma_{_{k_{0}}}$

$$c(t) = K_{\scriptscriptstyle o} e^{-p_{\scriptscriptstyle o} t} + K_{\scriptscriptstyle b} e^{-p_{\scriptscriptstyle o} t}$$



b. Underdamped response

Poles : two real at $-\sigma_{_{\rm d}} \pm {\rm j}\omega_{_{\rm d}}$

$$c(t) = K_{o} e^{-\sigma_{d} t} \cos(\omega_{d} t - \phi)$$



ග. Undamped response

Poles: two real at $\pm j\omega_{r}$

$$c(t) = A\cos(\omega_{o}t - \phi)$$



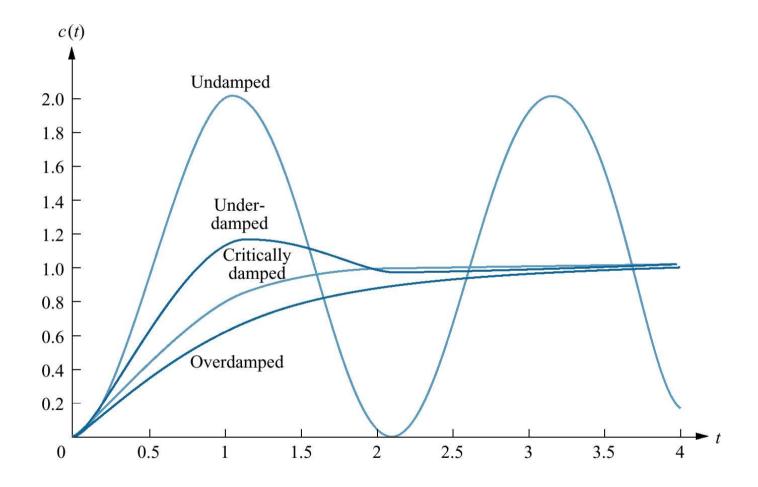
a. Critically damped responses

Poles: two real at $-\sigma_{_{\odot}}$

$$c(t) = K_{\scriptscriptstyle o} e^{-p_{\scriptscriptstyle o} t} + K_{\scriptscriptstyle b} t e^{-p_{\scriptscriptstyle o} t}$$



Step responses for second-order system damping cases





The General Second-Order Systems

Transfer function

$$G(s) = \frac{\omega_n^{\circ}}{s^{\circ} + \omega \zeta \omega_n s + \omega_n^{\circ}}$$

 W_n - Natural Frequency

ζ - Damping Ratio



Poles of the Transfer function

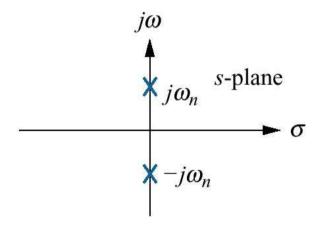
$$S_{0,0} = -\zeta \omega_n \pm \omega_n \sqrt{\zeta^0 - \omega}$$



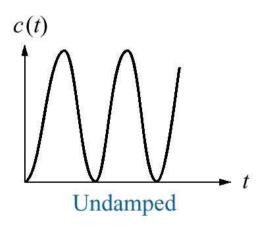
The General Second-Order Systems

Second-order response as a function of damping ratio.

$$\zeta = 0$$

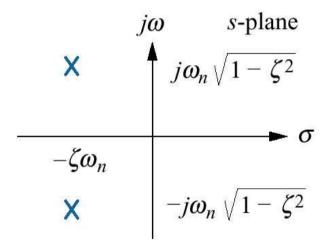


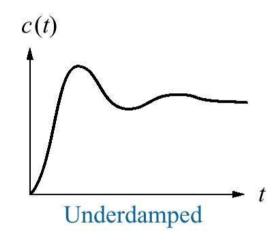
Poles





$$_{0}$$
< ζ < $_{0}$

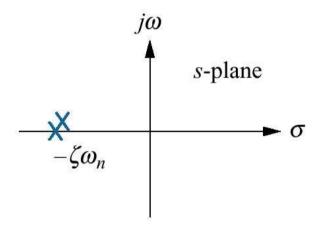


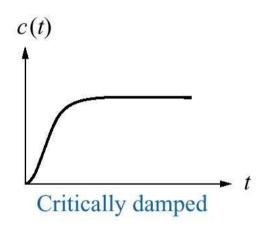


Poles



$$\zeta = 0$$

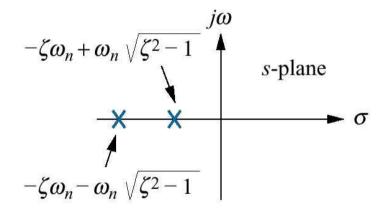


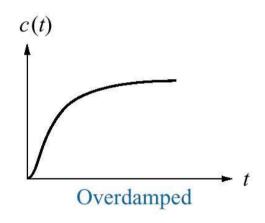


Poles



$$m{\zeta}\!>_{\scriptscriptstyle{0}}$$



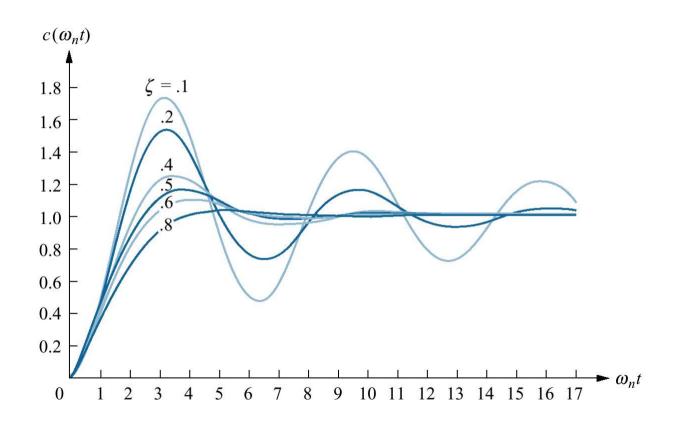


Poles



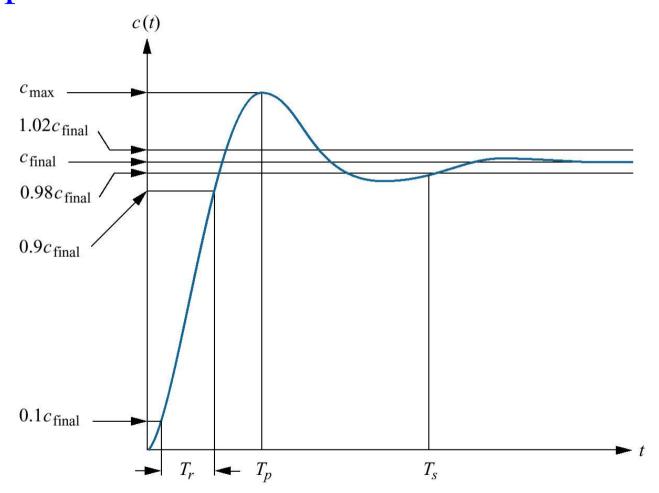
Underdamped Second-Order Systems

Second-order underdamped responses for damping ratio values





Second-order underdamped response specifications





Natural Frequency of a second order system is the frequency of oscillation of the system without damp ing.

Damping ratio is the proportion between exponential decay frequency and natural frequency (rad/second)

$$\zeta = \frac{Exponential\ decay\ frequency}{Natural\ frequency\left(rad\ /\ second\right)}$$

$$= \frac{1}{10\pi} \frac{\text{natural period (seconds)}}{\text{Exponential time constant}}$$



Percent overshoot is a ratio between the different between system time response and final value.

$$OS = \frac{c_{max} - c_{final}}{c_{final}} \times 000$$

Delay time is the time require for the system time response y(t) to reach &0% of final value.

Peak time is the time require to reach the first, or maximum, peak.



Time constant is the time require for step response to 50% of its final value.

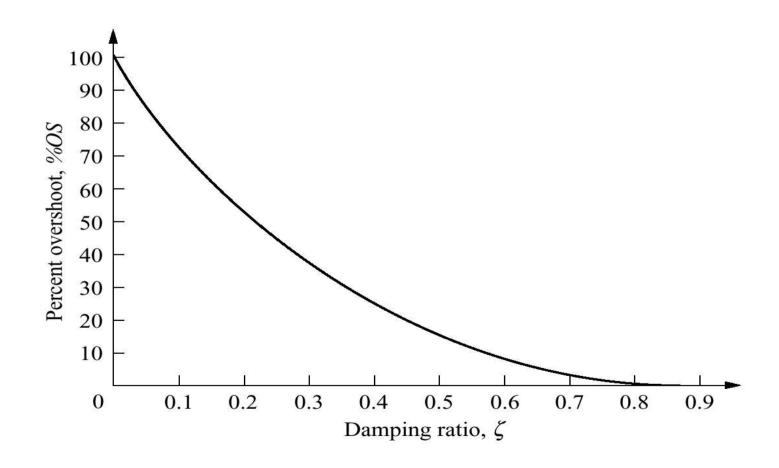
Time constant has the units (\circ seconds), or frequency, thus we call the parameter a the exponential frequency.

$$c(t) = _{\odot} - e^{-at}$$

$$t = \frac{6}{a}$$

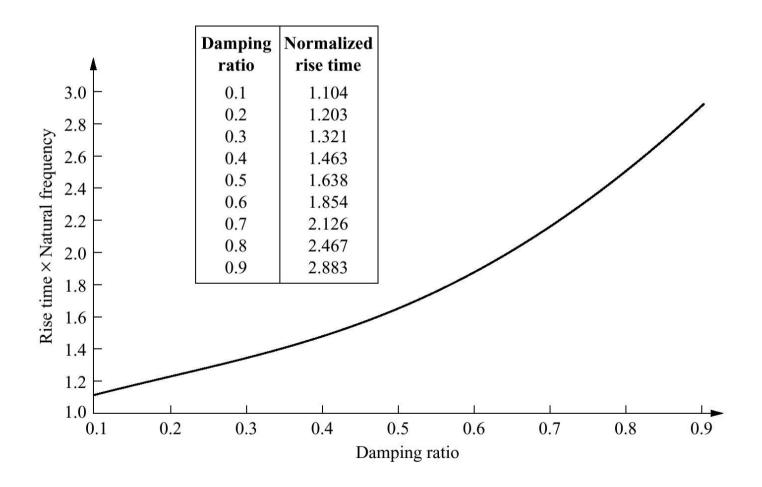


Percent overshoot vs. damping ratio





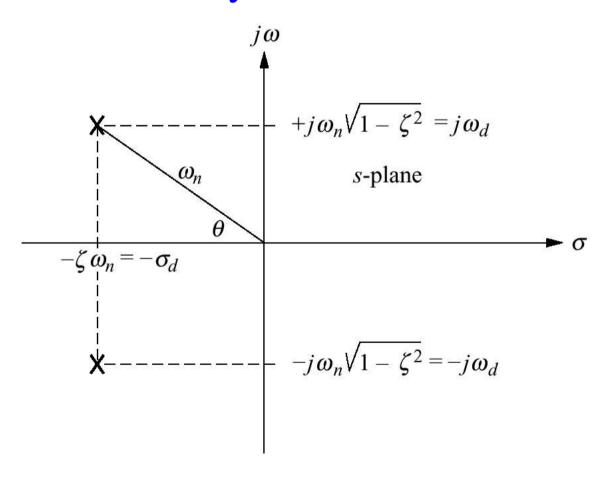
Normalized risetime vs. damping ratio for a second-order underdamped response





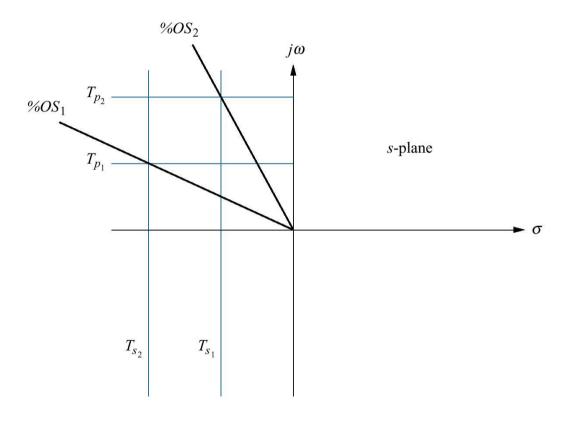
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Pole plot for an underdamped second-order system





Lines of constant peak time, T_p , settling time, T_s , and percent overshoot, %OS

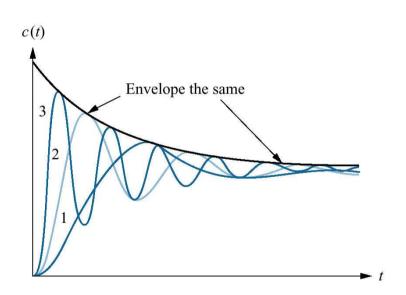


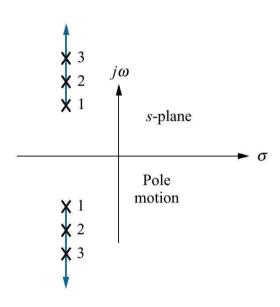
Note:
$$T_s < T_s$$
; $T_p < T_p$; $\%OS_o < \%OS_b$



Step responses of second-order underdamped systems as poles move

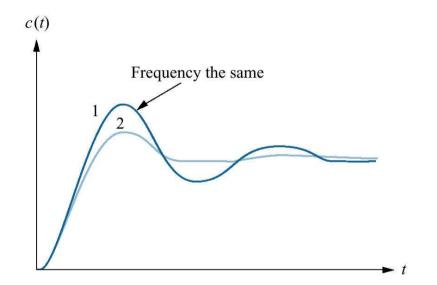
Constant real part

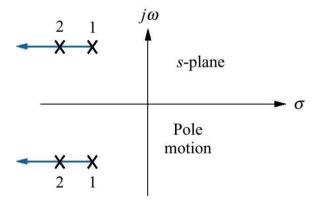






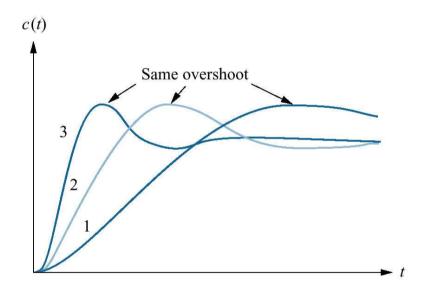
Constant imaginary part

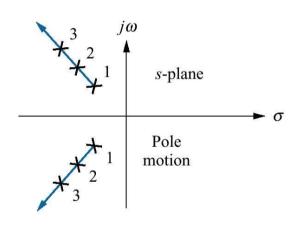






Constant damping ratio







System Response with Additional Poles

Component responses of a three-pole system :

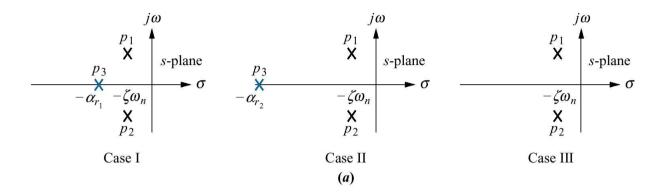
Case I : non-dominant pole is near dominant

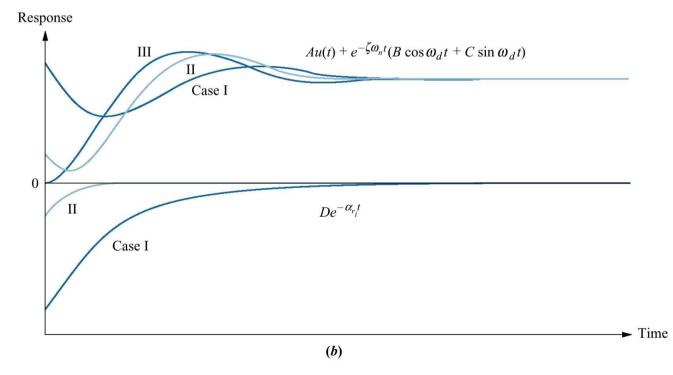
second-order poles pair.

Case II: non-dominant pole is far from the pair.

Case III: non-dominant pole is at infinity.









Example ๔.๘ (Nise)

$$T_{\circ}(s) = \frac{\text{bd.6db}}{s^{\circ} + ds + \text{bd.6db}}$$

$$T_{\text{b}}(s) = \frac{\text{bd.&db}}{(s+\text{oo})(s^{\text{b}}+\text{ds}+\text{bd.&db})}$$

$$T_{\circ}(s) = \frac{\text{bd.edb}}{(s+c)(s^{b}+ds+\text{bd.edb})}$$



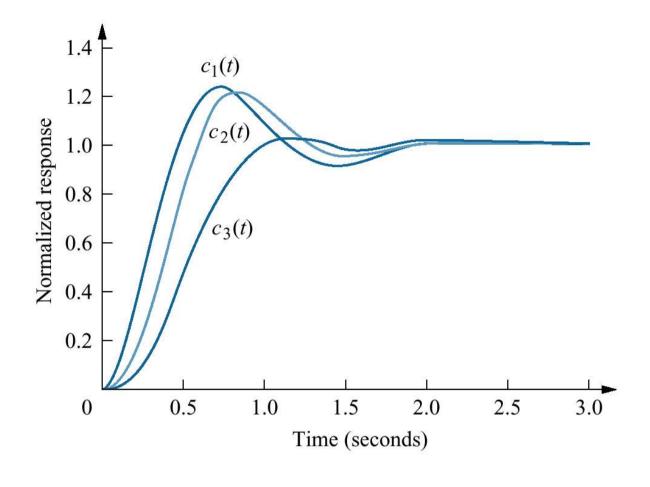
$$c_{o}(t) = o - o \cdot o e^{-b \cdot t} \cos(\epsilon \cdot \epsilon \cdot \delta)$$

$$c_{\scriptscriptstyle (0)}(t) = 0 - 0.0 \, \text{cm}^{-0.0t} - 0.0 \, \text{cm}^{-0.0t} \cos(\alpha.60 \, \text{cm}^{-0.0t} - \alpha.0 \, \text{cm}^{-0.0t})$$

$$c_{\scriptscriptstyle \mathrm{G}}(t) = 0 - 0.0 \, \mathrm{de}^{-ct} - 0.00 \, \mathrm{ord}^{-bt} \cos\left(\mathrm{d.dcht} + \mathrm{old.bc}\right)$$



Step responses of system $T_{_{\odot}}(s)$, system $T_{_{\square}}(s)$, and system $T_{_{\square}}(s)$





System Response with Zeros

$$T(s) = \frac{(s+a)}{(s+b)(s+c)} = \frac{A}{s+b} + \frac{B}{s+c}$$

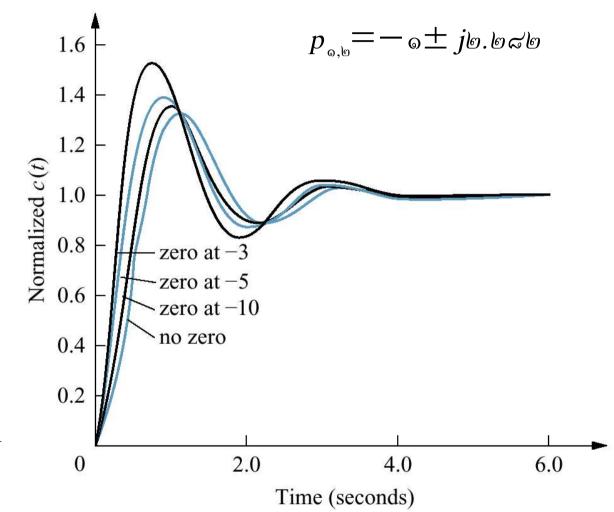
$$T(s) = \frac{(-b+a)/(-b+c)}{s+b} + \frac{(-c+a)/(-c+b)}{s+c}$$

If zero is far from poles, then a is large compared to b and c.

$$T(s) = \frac{a}{(s+b)(s+c)}$$

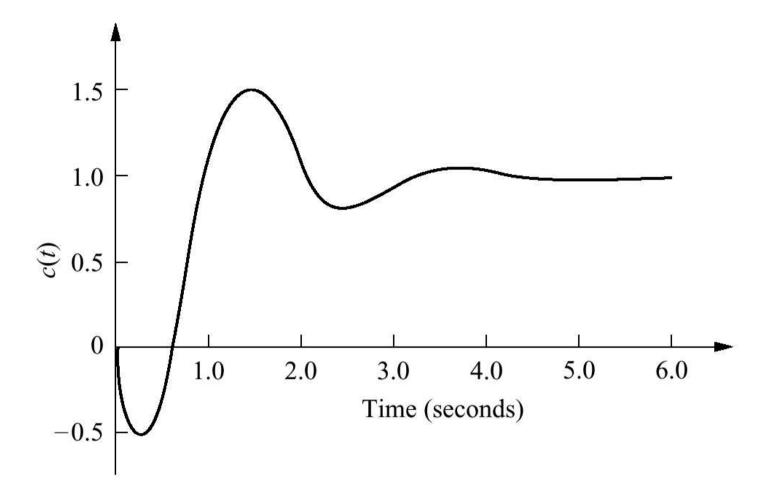


System Response with Zeros



Effect of adding a zero to a two-pole system



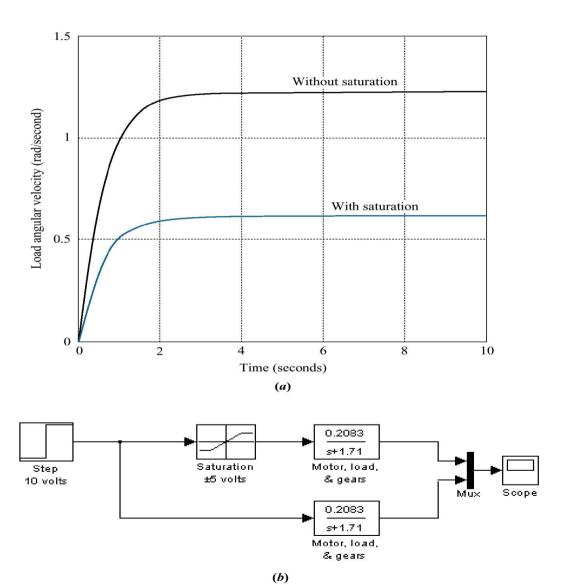


Step response of a nonminimum-phase system



Effects of Non-linearities Upon Time Response

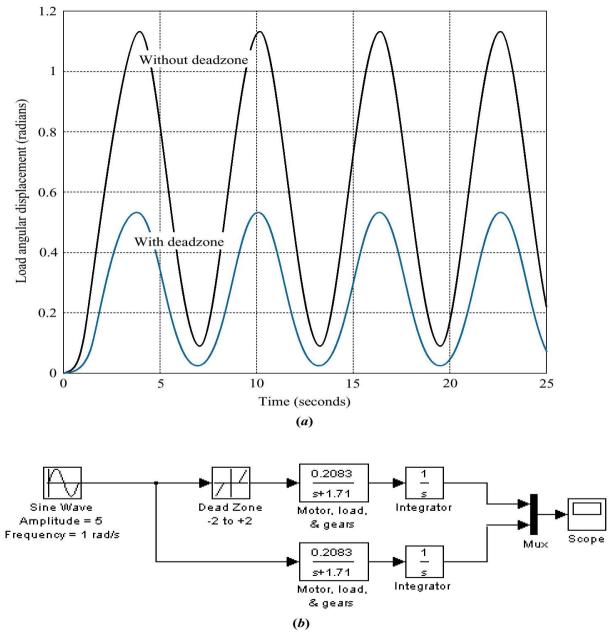
a. Effect of amplifier saturation on load angular velocity response; **b**. Simulink block diagram





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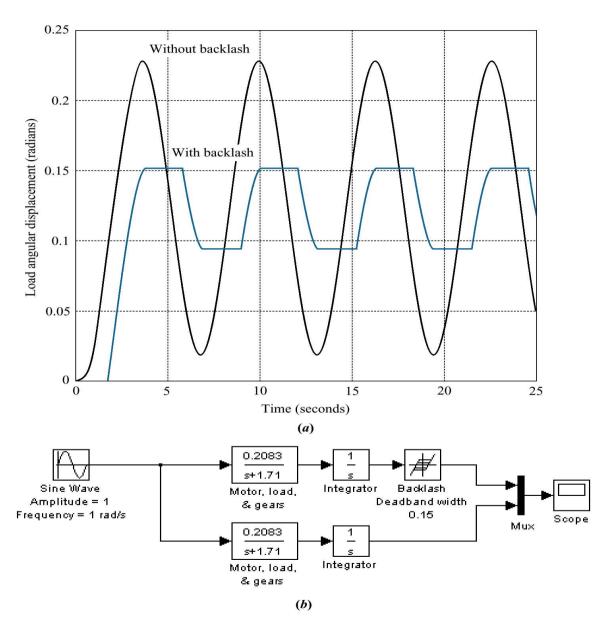
a. Effect of deadzone on load angular displacement response;b. Simulink block diagram





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a. Effect of backlash on load angular displacement response; **b**. Simulink block diagram





Time domain Solution of State equations

$$e^{at} = 0 + at + \frac{0}{0!} a^{0} t^{0} + ... + \frac{0}{k!} a^{k} t^{k} + ...$$

$$e^{At} = I + At + \frac{0}{10!} A^{10}t^{10} + ... + \frac{0}{k!} A^{k}t^{k} + ...$$

Scalar differential equation

$$\begin{array}{c}
\cdot \\
x = ax \longrightarrow x(t) = e^{at} x(o)
\end{array}$$

Vector-matrix differential equation

$$\overset{\cdot}{X} = AX \longrightarrow X(t) = e^{At} X(0)$$



$$x = Ax(t) + Bu(t)$$

Rearrange and multiply both sides by e^{-At}

$$e^{-At}[\dot{x}(t)-Ax(t)]=e^{-At}Bu(t)$$

Derivative of the product $e^{-At}x(t)$

$$\frac{d}{dt} \left[e^{-At} \mathbf{x}(t) \right] = e^{-At} \dot{\mathbf{x}}(t) - Ae^{-At} \mathbf{x}(t)$$



Integrating both sides yields

$$\left[e^{-At}x(t)\right]_{\circ}^{t}=e^{-At}x(t)-x(\circ)=\int_{0}^{t}e^{-A\tau}Bu(\tau)d\tau$$

We obtain

$$x(t) = e^{At} x(o) + \int_{o}^{t} e^{A(t-\tau)} Bu(\tau) d\tau$$
$$= \Phi(t) x(o) + \int_{o}^{t} \Phi(t-\tau) Bu(\tau) d\tau$$

$$\Phi(t) = e^{At}$$
 - state-transition matrix



 $(s \ I - A)^{-\circ}$ is the Laplace transform of the state-transtion matrix $\Phi(t)$

$$L^{-\circ}[(s I - A)^{-\circ}] = \Phi(t)$$

$$det(sI-A) = 0$$
 - characteristic equation

