

GLIBv3 user manual

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GLIB team

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1.INTRODUCTION

The Gigabit Link Interface Board (GLIB) [1] is an FPGA-based system for users of high speed optical links in high energy physics experiments. The GLIB serves both as a platform for the evaluation of optical links in the laboratory as well as a triggering and/or data acquisition system in beam or irradiation tests of detector modules. The major hardware component of the platform is the GLIB Advanced Mezzanine Card (AMC) [2] that can be used either on a bench or in a μ TCA [3] crate. The GLIB AMC is based on a Xilinx Virtex-6 FPGA with Multi-Gigabit Transceivers (MGT) operating at rates of up to 5 Gb/s. This performance matches the specifications of the Gigabit Transceiver (GBT) [4] and Versatile Link [5] [6] projects with targeted data rate of 4.8 Gb/s.

Figure 1-1 illustrates the baseline configuration of a GBT - Versatile Link - GLIB system is shown at the top. Front-end (FE) ASICs are electrically connected to the GBT ASIC through e-links [7] while the GBT high-speed serial data-streams are converted to/from the optical domain through the Versatile Transceiver [8]. At the other end, the GLIB system converts data to/from the optical domain, implements the GBT data transmission protocol [9] and codes/decodes the user payload at the link back-end. An alternative configuration, useful for intermediate prototyping, is shown in Figure 1-2 with one GLIB interfacing to FE ASICs and VTRx, thus emulating the GBT, and a second GLIB at the back-end.

Figure 1-3 shows a picture of the production version of the GLIB AMC, highlighting the two high-pin count (HPC) FMC Mezzanine Card (FMC) [10] sockets. The presence of the HPC FMC sockets is a big advantage since they provide additional user-specific I/O, high-speed transceivers and clock lines that can be used to extend the I/O connectivity of the GLIB AMC. For that reason, most of the auxiliary boards developed for the GLIB platform adopt the FMC format. The purpose of the auxiliary boards is to enhance the GLIB AMC compatibility with legacy and future triggering and/or data acquisition interfaces as well as its I/O bandwidth when in bench-top operation.

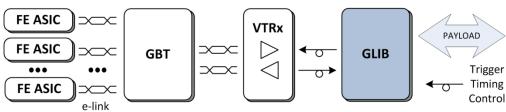


Figure 1: A GBT - Versatile Link system with the GLIB at the back-end.

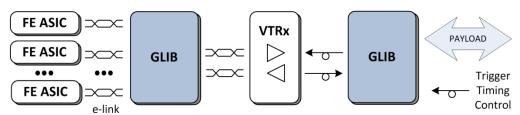


Figure 2: A GLIB interfacing to FE ASICs and VTRx with a second GLIB at the back-end.



Figure 3: Picture of the GLIB AMC, highlighting the two FMC sockets

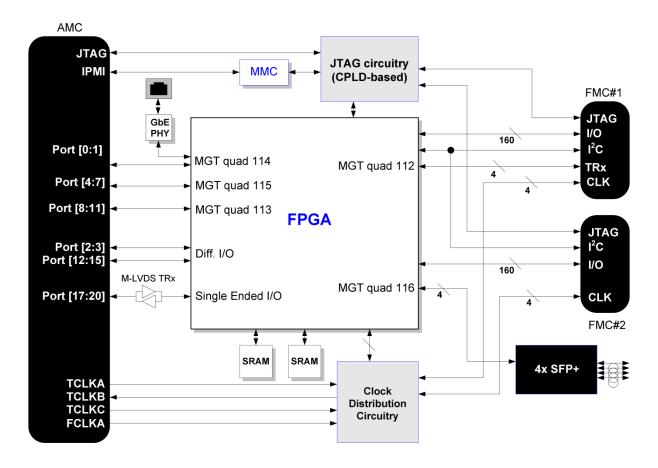


Figure 4: The block diagram of the GLIB card

Figure 1-5 shows a detailed diagram of the clock circuitry which is an essential part of the system. The MGT reference clocks and fabric clocks are shown in red and black, respectively.

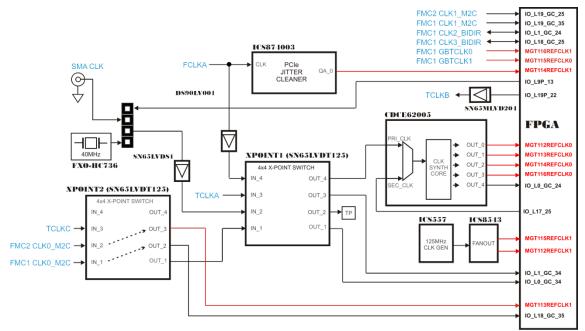
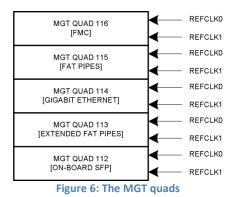


Figure 5: The clocking circuitry

It is important to mention that each MGT Reference clock (REFCLK) can be used to clock the MGT of its neighbouring MGT Quads (see Figure 1-6). For instance, the REFCLKO of the MGT 114 can also clock the MGT quads 113 and 115. Details about the Virtex-6 clocking resources can be found in Figure 1-6.



2.ARCHITECTURE

Figure 2-1 illustrates the FPGA firmware architecture of the GLIB that is organized in two main parts, the *system_core* and the *user_logic*.

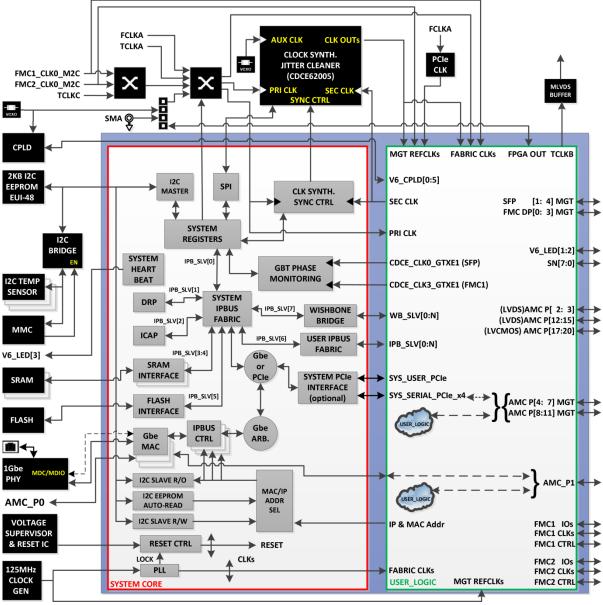


Figure 7: Firmware architecture.

The *system_core* firmware instantiates a simple IP-based control protocol (IPbus) designed for controlling xTCA-based hardware over Gigabit Ethernet that includes all basic transactions needed for this purpose (bitwise, single register and block transactions) [10]. The *system_core* also includes all interfaces to the on-board hardware e.g. I²C communication with the on-board temperature sensors and the serial memory, SPI communication with clock synthesizer, SRAM interface with the two (operating at up to 160MHz) etc.

2.1 System Core

2.1.1 PLL & Reset controller

The *system_core* contains a PLL which is clocked by the on-board 125MHz oscillator. The PLL provides a 62.5MHz clock which is used for the internal bus of the system (see §2.1.3) as well as a 125MHz clock for the Gigabit Ethernet MAC instantiations. The Reset controller generates a master reset pulse in the following cases:

- During power up (detected by the Voltage Supervisor)
- When the reset button is pressed (detected by the Voltage Supervisor).
- When the FPGA firmware is reloaded (internal logic).
- When the above mentioned PLL is not locked.

Both clocks (62.5MHz and 125MHz) as well as the reset pulse are forwarded to the user logic block.

2.1.2 **Gigabit Ethernet and IPbus**

For the Gigabit Ethernet links, MAC cores are instantiated. In the case of bench-top operation, the MAC core is configured as 1000Base-T in order to communicate with the external PHY. In the case of crate operation, the two MAC cores (AMC PO & P1¹) are configured as 1000Base-X for interfacing with the Gigabit Ethernet Switch carried on the crate's MCH.

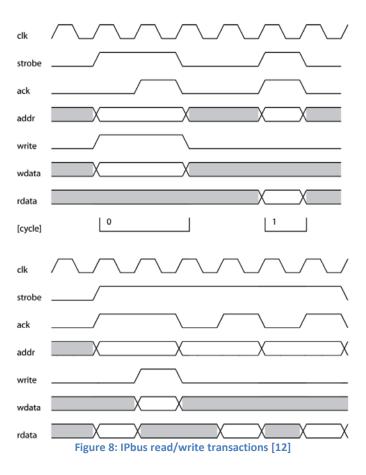
For every MAC core, an IPbus endpoint is also instantiated. The IPbus system allows the control of hardware via a 'virtual bus', using a standard IP-over-gigabit-Ethernet network connection. The IPbus specifies a simple transaction protocol between the hardware and a software controller, which assumes an A32/D32 connection to slave devices connected to the hardware endpoint. The current IPbus firmware implementation is using a UDP/IP protocol and a simple synchronous SoC bus [12]. This protocol is based upon the Wishbone SoC protocol [13], and is compatible with Wishbone cores. However, there are two important differences:

- The master is *not* required to explicitly deassert strobe between cycles. However, it is guaranteed to deassert strobe or begin the new cycle on the clock cycle following ack.
- Slaves are not allowed to tie ack high, and must deassert ack on the same clock cycle that strobe is deasserted. However, it is allowed to tie ack to strobe, if a zero-wait-state response is always possible.

Timing diagrams of read and write transactions for a slave with and without wait states are given below. The first diagram illustrates a write cycle to a slave with one wait state; the bus idle for two clock cycles; then a read to a slave with zero wait states. The second diagram illustrates a Read-Modify-Write transaction with a slave with zero wait states, followed immediately by two reads from a slave with one wait state. The reason of developing this custom SoC bus is to increase the bus efficiency by minimizing the dead-time.

IP & MAC addresses can be assigned by either the $user_logic$ block, the I2C EEPROM or by on-the-fly configuration via the MMC. When the provided IP is 0.0.0.0, the RARP mechanism of the IPbus is enabled, thus assigning IP addresses dynamically. The default IP address of the GLIB when used on a bench is 192.168.0.175. The default IP address when used in a μ TCA shelf ranges from 192.168.0.161 to 192.168.0.172 for AMC slots 1 to 12, respectively.

¹ The generation of AMC P1 MAC core when in crate mode is optional



2.1.3 **SoC bus**

As mentioned in §2.1.2, there is a case where more than one MAC/IPbus cores are instantiated. For deciding which Gigabit Ethernet link will take over the bus, an Arbitration module is instantiated between the IPbus cores and the bus fabric. The bus fabric redirects the bus to only one of the slave devices instantiated by decoding the address (based on the memory map of Figure 2-3). The *system_core* instantiates various system slaves e.g. the system registers (base address = 0x00000000), SRAM1 (base address = 0x02000000) and SRAM2 (base address = 0x04000000). Additionally, the *system_core* allocates a large memory space for user slaves; IPbus slaves (base address = 0x40000000) and Wishbone slaves (base address = 0x80000000) that can be added to the *user_logic*.

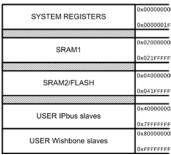


Figure 9: Memory map

2.1.4 <u>System Registers</u>

Table 2-1 shows the 16 registers that are currently implemented into the *System Registers* HDL block providing their address, a short description of their functionality and their type (Read-Write or Read-only). The 4-byte *Board_ID* register, as its name suggests, provides an identifier of the GLIB board (0x474C4942). When the 4 bytes of the identifier are represented in ASCII characters, the identifier corresponds to the word "GLIB". Table 2-2 shows how the register is organized. The 4-byte *System_ID* when represented in ASCII characters corresponds to "ip2x" declaring that the second version of IPbus firmware is used. Table 2-3 shows how the register is organized. The *Firmware_ID* register contains the date (*YY/MM/DD*) and the version number of the firmware (*major.minor.build*). The register *TestReg* is used only for testing the read/write transactions since it is not connected to the *system_core* logic. The *Ctrl* register (Table 2-5) is used to configure the clock circuitry (Figure 1-5). The register *Ctrl2* (Table 2-7) is used for loading firmware from the Platform Flash on demand.

Table 2-1: System Registers

Addr	Name	Description	Туре
0x00	Board_ID	The board identifier code	RO
0x01	System_ID	The system identifier code	RO
0x02	Firmware_ID	The firmware date and version number	RO
0x03	TestReg	Register for test purposes only	RW
0x04	Ctrl	Controls the external clocking circuitry	RW
0x05	Ctrl2	Flash control	RW
0x06	Status	Status from various external components	RO
0x07	Status2	Currently not used	RO
0x08	Ctrl_SRAM	SRAM interface: Control	RW
0x09	Status_SRAM	SRAM interface: Status	RO
0x0A	SPI_txdata	SPI interface: data from FPGA to clock synthesizer	RW
0x0B	SPI_command	SPI interface: configuration (polarity, phase, frequency etc.)	RW
0x0C	SPI_rxdata	SPI interface: data from clock synthesizer to FPGA	RO
0x0D	I2C_settings	I2C interface: configuration (bus select, frequency etc.)	RW
0x0E	I2C_command	I2C interface: transaction parameters (slave address, data to slave etc.)	RW
0x0F	I2C_reply	I2C interface: transaction reply (transaction status, data from slave etc.)	RO

Table 2-2: Board ID Register

Table 2 2. Dodra_D Register			
bit(s)	Name	Description	
[7:0]	board_id_char4	Board ID 4th character (ASCII code)	
[15:8]	board_id_char3	Board ID 3rd character (ASCII code)	
[23:16]	board_id_char2	Board ID 2nd character (ASCII code)	
[31:24]	board_id_char1	Board ID 1st character (ASCII code)	

Table 2-3: System_ID Register

bit(s)	Name	Description
[7:0]	system_id_char4	System_ID 4th character (ASCII code)
[15:8]	system_id_char3	System_ID 3rd character (ASCII code)
[23:16]	system_id_char2	System_ID 2nd character (ASCII code)
[31:24]	system_id_char1	System_ID 1st character (ASCII code)

The *Status* register (Table 2-8) is providing status information from various external components. The register *Status2* is reserved for future use.

Table 2-4: Firmware_ID Register

bit(s)	Name	Description
[31:28]	firmware_id_VER_MAJOR	Firmware version (major)
[27:24]	firmware_id_VER_MINOR	Firmware version (minor)
[23:16]	firmware_id_VER_BUILD	Firmware version (build)
[15:9]	firmware_id_YY	Firmware year (0-99)
[8:5]	firmware_id_MM	Firmware month
[4:0]	firmware_id_DD	Firmware day

Table 2-5: Ctrl Register

bit(s)	Name	Description
[0]	pcie_clk_fsel	ICS874003 output multiplication factor. 0 -> OUT = 2.5 x IN. 1 -> OUT =1.25 x IN.
[1]	pcie_clk_mr	ICS874003 master reset. 1 -> reset. 0 -> normal operation.
[2]	pcie_clk_oe	ICS874003 output enable. 1 -> outputs enabled. 0 -> outputs disabled.
[4]	cdce_powerup	CDCE62005 Control: power up of. 0 -> power down. 1 -> power up.
[5]	cdce_refsel	CDCE62005 Control: Clock input selection. 1 -> CLK1. 0 -> CLK2.
[6]	cdce_sync	CDCE62005 Control: synchronization. A transition from 0 to 1 is needed to resync.
[7]	cdce_ctrl_sel	Select who drives the control of the CDCE62005. 0 -> System_core. 1 -> user logic.
[8]	tclkb_dr_en	Enables the TCLKB output towards the backplane. 0 -> off. 1 -> on
[12]	xpoint2_s10	Input select for IC23 (SN65LVDT125) OUT1. See Table 2-6
[13]	xpoint2_s11	Input select for IC23 (SN65LVDT125) OUT1. See Table 2-6
[16]	xpoint1_s10	Input select for IC28 (SN65LVDT125) OUT1. See Table 2-6
[17]	xpoint1_s11	Input select for IC28 (SN65LVDT125) OUT1. See Table 2-6
[18]	xpoint1_s20	Input select for IC28 (SN65LVDT125) OUT2. See Table 2-6
[19]	xpoint1_s21	Input select for IC28 (SN65LVDT125) OUT2. See Table 2-6
[20]	xpoint1_s30	Input select for IC28 (SN65LVDT125) OUT3. See Table 2-6
[21]	xpoint1_s31	Input select for IC28 (SN65LVDT125) OUT3. See Table 2-6
[22]	xpoint1_s40	Input select for IC28 (SN65LVDT125) OUT4. See Table 2-6
[23]	xpoint1_s41	Input select for IC28 (SN65LVDT125) OUT4. See Table 2-6

Table 2-6: xpoint1 input select

Sx0	Sx1	Select Input	Example
0	0	IN_1 (see Figure 1-5)	When S40 = 0 & S41 = 0 then IN_1 -> OUT_4
0	1	IN_2 (see Figure 1-5)	When S40 = 0 & S41 = 1 then IN_2 -> OUT_4
1	0	IN_3 (see Figure 1-5)	When S40 = 1 & S41 = 0 then IN_3 -> OUT_4
1	1	IN_4 (see Figure 1-5)	When S40 = 1 & S41 = 1 then IN_4 -> OUT_4

Table 2-7: Ctrl2 Register

bit(s)	Name	Description
[1:0]	flash_firmware_page	Selects one of the 4 possible pages of the Platform Flash XL
[4]	Load_flash_firmware	Loads the firmware from the selected page

Table 2-8: Status Register

Table 2-6. Status Register			
bit(s)	Name	Description	
[2:0]	glib_sfp1_status[2:0]	on-board SFP1 status. bit[0] -> Mod_abs. bit [1] -> RxLOS. bit [2] -> TxFault.	
[6:4]	glib_sfp2_status[2:0]	on-board SFP2 status. bit [0] -> Mod_abs. bit [1] -> RxLOS. bit [2] -> TxFault.	
[10:8]	glib_sfp3_status[2:0]	on-board SFP3 status. bit [0] -> Mod_abs. bit [1] -> RxLOS. bit [2] -> TxFault.	
[14:12]	glib_sfp4_status[2:0]	on-board SFP4 status. bit [0] -> Mod_abs. bit [1] -> RxLOS. bit [2] -> TxFault.	
[16]	gbe_int	Interrupt request from GbE PHY. 0 -> no interrupt. 1 -> interrupt	
[17]	fmc1_presence	Presence of FMC1. 1-> yes, 0-> no	
[18]	fmc2_presence	Presence of FMC2. 1-> yes, 0-> no	
[19]	fpga_reset	state of the fpga_reset line (driven by the CPLD)	
[25:20]	v6_cpld	state of the 6-bit bus between the FPGA and the CPLD	
[28]	cdce_lock	Status of CDCE62005. 1 -> locked. 0 -> unlocked	

The registers Ctrl_SRAM and Status_SRAM are presented in §2.2.1.

The registers I2C_settings, I2C_command and I2C_reply are controlling the operation of the dual-bus I2C master controller that can be used to access the on-board I2C devices when in bench-top mode (in crate mode, the MMC is responsible for that task). Table 2-9 presents the I2C devices and their corresponding I2C 7-bit slave addresses. The operation of the above mentioned registers is the following:

- The *I2C_settings* register used for configuring the controller (Table 2-10).
- The I2C command for setting the I2C transaction parameters (Table 2-11).
- The *I2C_reply* provides the transaction status as well as data sent by the slave in case of read transactions (Table 2-12).

Table 2-13 gives an example list of transactions needed for reading the temperature of the FPGA with an I2C bus frequency of 62.5kHz.

Table 2-9: The I2C devices and their 7-bit slave addresses

I2C Device	Description	I2C Address
24AA025E48 EEPROM	256x8bit EEPROM with unique 48-bit serial number (EUI-48)	1010110
LM82 Temperature sensor #1	Provides the temperature of the FPGA's die at reg1 [addr:0x01]	0101010
LM82 Temperature sensor #2	Provides the pcb temperature at the front at reg0 [addr:0x00]	0011010
LM82 Temperature sensor #3	Measures the pcb temperature at the rear at reg0 [addr:0x00]	1001110
I2C Slave RW	Configures the IP and/or MAC address on-the-fly	1111010
I2C Slave RO	Reports the current IP & MAC address	1111110
FMC#1	The I2C devices hosted on the FMC#1	XXXXX00
FMC#2	The I2C devices hosted on the FMC#2	XXXXX11

Table 2-10: "I2C_settings" register

bit(s)	Name	Description
[9:0]	i2c_prescaler	I2C clock prescaler. I2C clk (kHz)=62500/i2c_prescaler
[10]	i2c_bus_select	Select I2C bus. 1 -> I2C for PHY. 0 -> I2C for all other devices.
[11]	i2c_enable	Enable I2C controller. 1 -> enabled. 0 -> disabled.
[12]	reserved	Reserved. Keep it always at 0

Table 2-11: "I2C_command" register

bit(s)	Name	Description
[7:0]	Wrdata	Byte to write to the i2c slave.
[15:8]	reserved	Reserved. Keep always to 0.
[22:16]	slv_addr	The 7-bit slave address
[23]	wr_en	Write enable. 0 -> Read transaction. 1 -> Write transaction.
[24]	reserved	Reserved. Keep always to 0.
[25]	mode16b	16bit operation (for PHY only). 0 -> standard 8-bit mode. 1 -> 16bit mode
[31]	i2c_strobe	Execute Transaction strobe signal. It clears automatically. Keep it always to 1.

Table 2-12: "I2C_reply" register

bit(s)	Name	Description
[7:0]	rddata_lo	Byte read from the i2c slave. Low byte in case of 16-bit mode
[15:8]	rddata_hi	High byte read from the i2c slave in case of 16-bit mode
[27:26]	i2c_status[1:0]	Transaction status. 01 -> succeed. 11 -> failed. 00 -> pending. 10 -> pending.

Table 2-13: Example transaction list for reading the temperature of the FPGA's die

Register	Access	Description	Value (hex)
I2C_settings	WRITE	i2c_enable -> 1, i2c_bus_sel -> 0, i2c_prescaler -> 1000	0x00000BE8
I2C_command	WRITE	wr -> 1, mode16b -> 0, slv_addr -> 0x2A wrdata -> 0x01 (select LM82's register#1)	0x80AA0001
I2C_reply	READ	Verify the transaction status	
I2C_command	WRITE	wr -> 0, mode16b -> 0, slv_addr -> 0x2A, wrdata -> don't care	0x802A0000
I2C_reply READ		Verify the transaction status. If successful, the temperature will be available in rddata_lo	
I2C_settings	WRITE	<pre>i2c_enable -> 0, i2c_bus_sel -> don't care, i2c_prescaler -> don't care</pre>	0x00000000

The registers *SPI_txdata*, *SPI_command* and *SPI_rxdata* are controlling the operation the SPI master controller that is used for the configuration of the CDCE62005 clock synthesizer. The operation of the above mentioned registers is the following:

- The SPI_txdata where the 32-bit word to be transmitted to the SPI slave is defined.
- The SPI_rxdata where the 32-bit word transmitted by the SPI slave is presented.
- The *SPI_command* where the SPI master's configuration is defined as well as the transaction parameters (Table 2-14).

Table 2-15 shows an example list of transactions for reading one of CDCE62005's registers. Table 2-16 shows an example of writing one of CDCE62005's registers.

Table 2-14: The SPI_command register

bit(s)	Name	Description
[11:0]	spi_prescaler	SPI clock prescaler. SPI clk (MHz)=62.5/spi_prescaler. Suggested value: 0x014
[27:12]	reserved	Reserved. Keep always to 0xFA38.
[31]	spi_strobe	Execute Transaction strobe signal. It clears automatically. Keep it always to 1.

Table 2-15: Transaction list for reading CDCE62005's Register 8

Register	Access	Description	Value (hex)
SPI_txdata	WRITE	Set the value that corresponds to the reading register 8 command of the CDCE62005	0x0000008E
SPI_command	WRITE	Execute the transaction	0x8FA38014
SPI_txdata	WRITE	Set a dummy value that does not corresponds to any CDCE62005 command	ОхАААААААА
SPI_command	WRITE	Execute the transaction	0x8FA38014
SPI_rxdata	READ	The contents of CDCE62005's Register 8 appear here	

Table 2-16: Example transaction list for writing CDCE62005's Register 0

Register	Access	Description	Value (hex)
SPI_txdata	WRITE	Set the value to write to Register 0. Note that bits[3:0] must be zero for addressing the register correctly	0xEB840720
SPI_command WRITE Exe		Execute the transaction	0x8FA38014

2.2 User Logic

From the *user_logic* point of view, there are two types of interfaces, direct connections with FPGA pins and interfaces with firmware blocks instantiated in the *system_core*. The use of the pins directly connected to the FPGA is straight forward (just if the *system_core* and the top level did not exist). This section focuses in the use of the SRAM, FMC I/O and Wishbone bus interfaces as well as in the instantiation of GBT-based links.

2.2.1 **SRAM**

The two single-port 2Mx36 SRAM devices that are available on board are accessible both by the *system_core* and the *user_logic*. The register *Ctrl_SRAM* (Table 2-17) is used to select which of the two blocks takes control over the SRAM, the mode of operation (normal or Built-In Self Test (BIST)). Additionally, since the SRAM2 and the FLASH memory share the same bus, it selects which of the two memories is addressed (please note that the access of the FLASH is not covered in this section). The register *Status_SRAM* (Table 2-18) is used only to report the results of the BIST. Logic Analyzer Waveforms captured with Chipscope for single write, single read, block write and block read SRAM transactions from the *user_logic* are shown in Figure 2-4, Figure 2-5, Figure 2-6 and Figure 2-7, respectively. Figure 2-8 shows an example use of an SRAM in the *user_logic* block.

Table 2-17: "Ctrl_SRAM" register

bit(s)	Name	Description	
[0]	user_ctrl_sram1	Selects who is taking over the SRAM1 bus. 0 -> system_core. 1 -> user_logic.	
[1]	bist_sram1	Built-in Self Test (BIST) Enable for SRAM1. 0 -> normal operation. 1 -> run the BIST.	
[16]	user_ctrl_sram2	Selects who is taking over the SRAM2 bus. 0 -> system_core. 1 -> user_logic.	
[17]b	bist_sram2	Built-in Self Test (BIST) Enable for SRAM2. 0 -> normal operation. 1 -> run the BIST.	
[20]	flash_select	Selects between SRAM2 & FLASH. 0 -> SRAM2. 1 -> FLASH.	

Table 2-18: "Status SRAM" register

bit(s)	Name	Description	
[0]	bist_done_sram1	SRAM1 BIST completed flag. 0 -> test pending. 1 -> test completed.	
[1]	bist_status_sram1	SRAM1 BIST status flag. 0 -> test failed. 1 -> test passed.	
[16]	bist_done_sram2	SRAM2 BIST completed flag. 0 -> test pending. 1 -> test completed.	
[17]	bist_status_sram2	SRAM2 BIST status flag. 0 -> test failed. 1 -> test passed.	

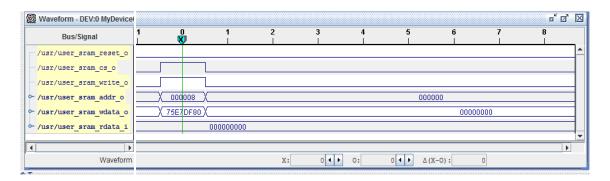


Figure 10: User_logic SRAM single write

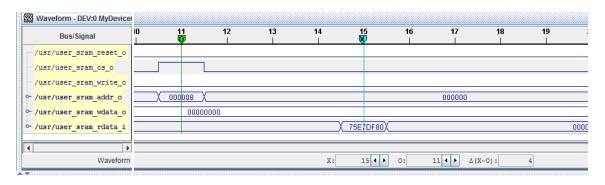


Figure 11: User_logic SRAM single read

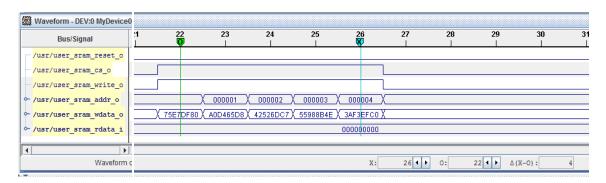


Figure 12: User_logic SRAM block write

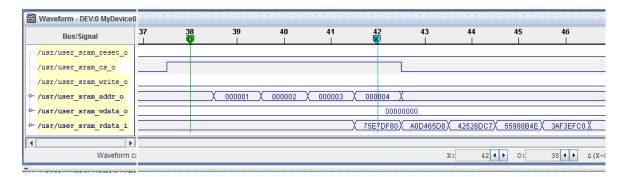


Figure 13: User_logic SRAM block read

```
architecture my_arch of user_logic is
constant my_sram1 : integer:= 1;
begin my_arch of user_logic
 ._______
                                       <= '0';
user_sram_control_o(my_sram1).reset
                                                           -- active high
user_sram_control_o(my_sram1).clk <= my_clk;
user_sram_control_o(my_sram1).cs <= '1':</pre>
                                                          -- rising edge
user_sram_control_o(my_sram1).cs
                                       <= '1';
user_sram_control_o(my_sram1).writeEnable <= '1';
                                                          -- active high
                                      <= x"075E7DF80";
                                                          -- 36 bit data
user_sram_wdata_o (my_sram1)
user sram addr o (my sram1)
                                        <= "0" & x"000008"; -- 25 bit addr
```

Figure 14: Example use of the SRAM in the user_logic file

2.2.2 FMC I/O

All FMC I/Os are directly connected to the *user_logic* block and can be configured independently as LVCMOS 2.5V or LVDS inputs or outputs by simply instantiating the corresponding Xilinx primitive buffer such as IBUF, OBUF, IOBUF, IBUFDS, OBUFDS etc. (see Figure 2-9).

Figure 15: Example I/O buffer instantiation in the user_logic for the LA15_P line of FMC1 using Xilinx primitives

2.2.3 Wishbone bus

The user has the possibility to instantiate wishbone-compatible slaves inside the *user_logic* as well as to attach them on the wishbone bus available. Figure 2-10 shows an example of how to declare the total number of wishbone slaves (and their symbolic names). Figure 2-11 shows the address mapping of the wishbone slaves. Figure 2-12 illustrates an example of a wishbone slave instantiation in the *user_logic*. Example wishbone slaves can be found in the example projects available.

Figure 16: Wishbone slave declaration (user_package)

Figure 17: Example wishbone slave address mapping (user_addr_decode.vhd)

```
wb_timer: entity work.wb_tics
______
generic map (g_period => 62500) -- 1ms
port map
10
   ----- master out/slave in (MOSI) ------
              => wb_mosi_i(user_wb_timer).wb_rst,
   rst i
                       => wb_mosi_i(user_wb_timer).wb_clk,
   clk sys i
                       => wb_mosi_i(user_wb_timer).wb_adr,
   wb addr i
   wb data i
                       => wb_mosi_i(user_wb_timer).wb_dat,
                       => wb_mosi_i(user_wb_timer).wb_cyc,
   wb cyc i
                       => wb_mosi_i(user_wb_timer).wb_sel,
   wb sel i
   wb_stb_i
                       => wb_mosi_i(user_wb_timer).wb_stb,
                        => wb_mosi_i(user_wb_timer).wb_we,
   wb_we_i
   ----- master in/slave out (MISO) ------
   wb_data_o
                        => wb_miso_o(user_wb_timer).wb_dat,
                        => wb miso o(user wb timer).wb ack,
   wb ack o
                         => wb_miso_o(user_wb_timer).wb_err
   wb_err_o
```

Figure 18: Example instantiation of a wishbone slave in the user_logic file

2.2.4 **GBT-based links**

As mentioned in the introduction, one of the principle use cases of the GLIB is to be part of a GBT – Versatile Link system for evaluation purposes. For that reason, the GLIB firmware release includes the example project <code>glib_v3_gbt_fpga</code> that instantiates inside the <code>user_logic</code> block the GBT-FPGA reference design for Virtex-6 as delivered by the GBT-FPGA team. The source files and associated documentation can be found in [16].

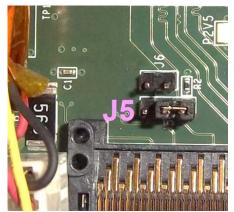
3. How to use the GLIB

3.1 Hardware

3.1.1. <u>Jumpers</u>

The operation mode of the GLIB is set up through two jumpers (J5 and J10). The position of the jumpers for the two modes of operation is shown below.

Bench-top operation



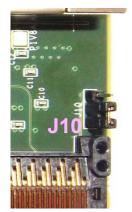


Figure 3-19: Jumpers J5 and J10 set up for bench-top operation.

Crate operation



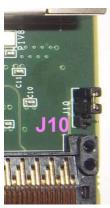


Figure 3-20: Jumpers J5 and J10 set up for crate operation.

External Clocking

As previously illustrated in Figure 1-5 and Figure 2-1, the GLIB features an SMA connector for external clocking. The connection scheme of the SMA connector (SMA1) and the 40MHz crystal oscillator (QZ2) of the GLIB is set up through the jumper J14. The default configuration sets the SMA connector as input/output of the FPGA and the 40Mhz crystal oscillator as input for the cross point switch 1 (IC28).

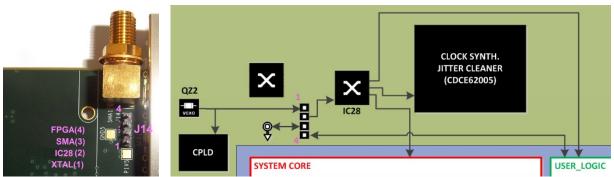


Figure 3-21: SMA/Jumper J14 picture (left). SMA and 40 MHz crystal oscillator (QZ2) setup block diagram (right)

3.1.2. Switches

The default position of the DIP switches is as illustrated below. All other combinations are reserved.





Figure 3-22: Default switches positions.

3.1.3. **Powering**

For bench-top operation, the GLIB can be powered through a socket compatible with the 12V 4-pin connector that is typically available in ATX compatible power supplies (Figure 3-5 left and right, respectively). In case of ATX power supply use, the PS-ON pin of the 20-pin ATX connector must be connected to ground as shown in Figure 3-6 (bottom) in order to start-up the power supply.





Figure 3-23: Powering a GLIB with an ATX power supply.

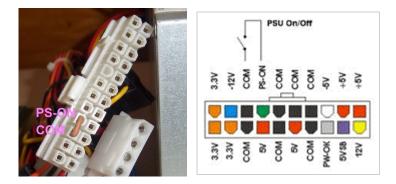


Figure 3-24: How to start-up an ATX Power supply outside a computer.

3.1.4. Configuration

JTAG connectors

The GLIB features two JTAG connectors. J13 configures the CPLD & J12 configures the FPGA.

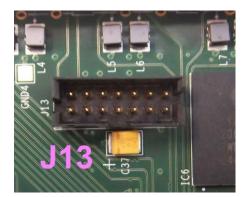




Figure 3-25: JTAG connectors for configuring the CPLD (left) and FPGA (right).

FPGA configuration scheme

For advanced FPGA configuration schemes please see Appendix B.

3.1.5. <u>Reset</u>

As mentioned in §2.1.1, the GLIB board features a button for resetting the FPGA logic. The button (SW1) is located in between the two FMC sockets and it is connected to a TLC7725 power supervisor (IC14). When the button is pressed, the power supervisor generates a pulse that is used to reset the *system_core*. Additionally, the reset signal is forward to the *user_logic* block where it can be freely used.

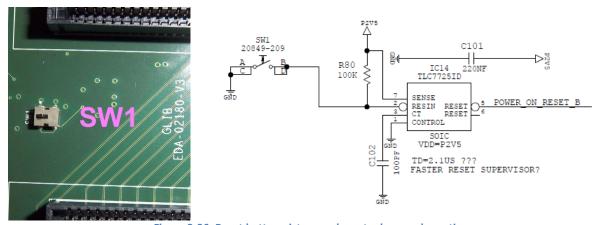


Figure 3-26: Reset button picture and reset scheme schematic.

3.1.6. **RI45 socket**

The RJ45 socket (and associated circuitry) carried by the GLIB board provides a Gigabit Ethernet connection in case of bench-top operation. Please note that since the GLIB supports only Ethernet at 1Gbps (and not 10/100Mbps), please connect it using standard (not crossover) network cables either directly to a PC with a Gigabit Ethernet card or to a Gigabit Ethernet switch.

3.1.7. <u>Serial Number</u>

As previously mentioned in Table 2-9, the GLIB carries a serial memory with a 48-bit unique serial number that could be used to identify the GLIB boards. Furthermore, the GLIB offers the possibility to the user of an additional 8-bit serial number that is directly accessible from the *user_logic* block and it can be hard-coded by placing up to 8 resistors on the pads shown in Figure 3-9.

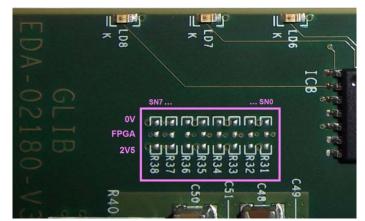


Figure 3-27: Resistors pads for optional hard-coding of 8-bit serial number.

3.1.8. LEDs

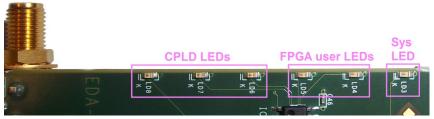


Figure 3-28: CPLD, FPGA user and system LEDs

System LED

The system core blinks the system LED (LD3) to indicate that the GLIB firmware has been loaded correctly.

FPGA user LEDs

The FPGA user LEDs (LD4 and LD5) can be directly controlled by the logic placed on the user core.

CPLD LEDs

The LEDs LD6, LD7, LD8 are connected to the CPLD. The meaning of these LED depends on the CPLD firmware version.

MMC LEDs

When in crate operation, the LEDs MMC_RED_LED (LD1), MMC_GREEN_LED (LD2) and MMC_BLUE_LED (LD9) are controlled by the MMC and indicate the state of the GLIB.

Table 19: MMC LEDs and GLIB state relationship

	(GLIB state
MMC LEDs	Inactive	Active
MMC_RED_LED	ON	OFF
MMC_BLUE_LED	ON	OFF
MMC_GREEN_LED	OFF	ON





Figure 3-29: Inactive GLIB (left) and active GLIB (right)

3.2 Firmware/Software

Firmware:

In order to access all files required, you will need to do the following:

- Checkout amc_glib repository [https://svn.cern.ch/reps/ph-ese/be/amc_glib] to a folder e.g. dev.
- Checkout the ipbus_2_0_v1 tag [https://svn.cern.ch/reps/cactus/tags/ipbus_fw/ipbus_2_0_v1] into dev/cactus/tags/.
- Checkout the *gbt-fpga* repository [https://svn.cern.ch/reps/ph-ese/be/gbt_fpga] into *dev*.
- Checkout the *mmc* repository [https://svn.cern.ch/reps/ph-ese/be/mmc] into dev.

The folder tree should be as shown in Figure 3-12



Figure 3-30: Folder tree

The GLIB team recommends using Xilinx ISE v14.5 for firmware development and implementation.

For using the existing example code as is, simply compile the *.xise project files of the *latest* "amc_glib" tag available under \tags\[latest tag]\glib_v3\fw\fpga\prj\]. For developing your own code, please keep in mind that in order to receive support from the GLIB team, the files under [..]\glib_v3\fw\src\system MUST remain unchanged. On the other hand, the files under [..]\glib_v3\fw\src\user\cap can be freely modified according to the user needs. Concerning the CPLD & MMC firmware, we provide the source files for your reference only and we strongly suggest not modifying them.

IMPORTANT NOTICE: The GLIB boards are delivered with the FLASH memory empty.

Software:

For software development, it is strongly recommended to use the µHAL [17] software distribution, also available in the CACTUS SVN repository. For hardware verification purposes only, we also provide a few Python scripts performing basic test functions that must be used *as is*. Please note that the scripts are using a Python library namely "Pychips" that is not supported any longer by its developers, therefore you should avoid in any cost to base your developments on that platform. In order to execute the hardware verification scripts, please make sure that Python 2.7.x is installed. PyChips is NOT compatible with Python 3.x. For the correct operation of the scripts, some variables have to be defined.

Under Windows, define the following environmental variables:

- PYTHONPATH with value the path pointing to the [..]\glib_v3\sw\PyChips\src folder.
- PATH with value the path pointing to the Python installation folder e.g. "C:\Python27"

Under Linux, use the export command to define the PYTHONPATH variable. All hardware verification scripts as well as the GLIB IP address and register address table declarations are located at [..]\glib_v3\sw\PyChips\scripts

Additionally, in order to communicate successfully with the GLIB, the IP address of the Ethernet port has to be set in order to be in the same subnet e.g. for the default GLIB addresses the configuration could be 192.168.0.10 as IP address and 255.255.255.0 as subnet mask.

Brief description of the hardware verification scripts:

glib_board_info.py: prints general board status information

glib sysreg test.py: example of r/w access of the system registers

glib cdce read.py: reads through SPI the contents of the CDCE62005 registers

glib cdce write.py: writes through SPI the GLIB default values to the CDCE62005 registers

qlib cdce write test only.py: writes through SPI non-sense values to the CDCE62005 registers

glib cdce powerup and sync.py: power-cycles and synchronizes the CDCE62005

glib sram1 read.py: reads the contents of the sram1, writes them to the tmp.txt and prints the first 32 values

<u>glib_sram1_write.py:</u> writes and incrementing by 1 value to all memory locations of sram1

glib sram1 clear.py: writes "zeros" to all memory locations of sram1

glib sram2 read.py: reads the contents of the sram1, writes them to the tmp.txt and prints the first 32 values

glib sram2 write.py: writes and incrementing by 1 value to all memory locations of sram2

glib sram2 clear.py: writes "zeros" to all memory locations of sram2

glib i2c eeprom read eui.py: reads the 48-bit serial number from the 24AA025E48 I2C EEPROM

<u>glib i2c temperature.py:</u> reads the three on-board I2C temperature sensors

<u>glib i2c mac ip eeprom.py:</u> programs the memory space of the I2C EEPROM allocated for the IP & MAC address assignment.

<u>glib i2c mac ip ctrl.py:</u> accesses the FPGA's "I2C slave R/W" in order to program on-the-fly the IP & MAC address (for test purposes only, this procedure is normally reserved for the MMC)

<u>glib i2c mac ip status.py:</u> accesses the FPGA's "I2C slave R/O" in order to read the actual IP & MAC address (for test purposes only, this procedure is normally reserved for the MMC)

glib_icap_interface_test.py: reads and writes several registers of the icap

<u>glib icap jump to image.py:</u> triggers the reconfiguration of the FPGA (used for Multiboot/Safe configuration) <u>glib flash golden prom rw.py:</u> writes the Golden image PROM file to the FLASH or reads the Golden image on FLASH and dumps it to a file

<u>glib flash user prom rw.py:</u> writes the User image PROM file to the FLASH or reads the User image on FLASH and dumps it to a file

glib flash corrupt user image.py: corrupts the User image stored on FLASH, for testing purposes only

4. REFERENCES

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- [17] CERN Cactus SVN/Wiki https://svnweb.cern.ch/trac/cactus/wiki

APPENDIX A: GLIB v3 FPGA configuration scheme

Introduction

In most FPGA applications, one bitstream (FPGA image) is programmed onto the FLASH memory and the FPGA is configured when the system is powered on. This is the simplest FPGA configuration scheme possible when using a FLASH memory for storing the configuration file (from now on it will be referred as **single image FPGA configuration scheme**). However, there is an intrinsic safety problem when using this scheme because either a bitstream corruption or an incorrect firmware may lead to a system crash (see Figure A 1).

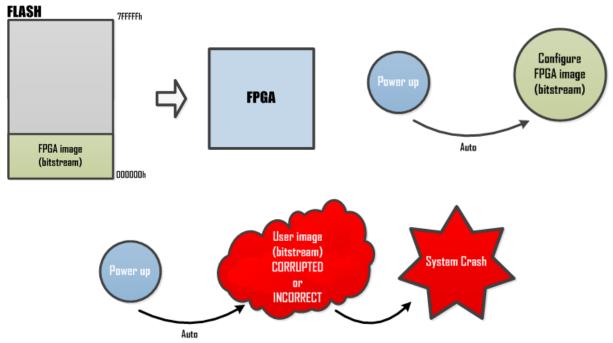


Figure A 1: Standard FPGA configuration scheme.

In order to provide a reliable FPGA configuration procedure, the GLIB board features the **Multiboot/Safe FPGA configuration scheme**. This scheme requires two different bitstreams stored on different memory positions of the FLASH (see Figure A 2). One of the bit streams, the "Golden image", is used as backup firmware enabling the possibility of remotely restoring/updating the system when needed. The other bitstream, the "User image", is the main firmware which comprises the actual system implemented by the user.

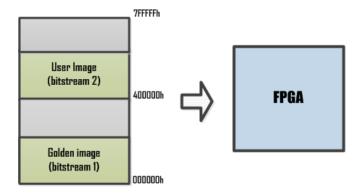


Figure A 2: Golden and User images on FLASH.

The handling of these bit streams by the GLIB system can be done through either the Internal Configuration Access Port (ICAP) or by pulling up the address 22 pin of the FLASH memory (A22).

Figure A 3 shows the **ICAP approach** (default setup), where the FPGA is initially loaded with the Golden image during power up. Once the FPGA is configured, the User image can be loaded ondemand through the ICAP interface module that is controlled via IPbus. If for any reason the User image is corrupted, the FPGA will automatically trigger a "Fallback configuration" after the failed configuration process, reverting to the Golden image. Moreover, if the User image is not corrupted but it does not behave as expected ("incorrect"), it is possible to revert to the Golden image by either another ICAP command or by power cycling (in case of a User image with non-functioning IPbus/ICAP interface). As mentioned before, once the Golden image is loaded, the system is again under control so a correct User image may be reprogrammed onto the FLASH and loaded to the FPGA on-demand (see Figure A 4).



Figure A 3: Multiboot/Safe FPGA configuration scheme (ICAP approach).

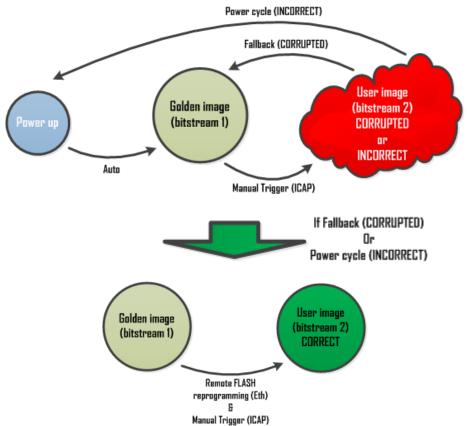


Figure A 4: Multiboot/Safe FPGA configuration system recovery (ICAP approach).

Although the ICAP is the recommended Multiboot/Safe FPGA configuration scheme approach due to the reliability provided by the Fallback and the Power cycle recovery procedures, there are some applications where the configuration time is a critical factor (such as PCle) so the ICAP approach may not meet the requirements of the system. For this kind of applications, the user has the possibility of using the A22 pin pull-up approach (see Figure A 5) which allows the shortest configuration times, even if loosing reliability. In this case, the A22 pin of the FLASH memory is pulled up during power up so the FPGA is directly configured with the User image, avoiding the extra time of the Golden image configuration and the manual triggering of the User image. However, as aforementioned, this approach is not as good as the previous (ICAP) in terms of reliability because it only allows system recovery when the bitstream is corrupted (Fallback configuration). In case of an "incorrect" User image (e.g. just and LED blinking and no ICAP instead of the user's system), the power cycle procedure will no longer be a solution because in this approach, the Golden image is not the bitstream loaded after power up (see Figure A 6).

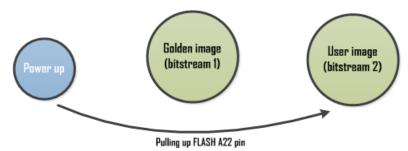


Figure A 5: Multiboot/Safe FPGA configuration scheme (A22 pin pull-up approach).

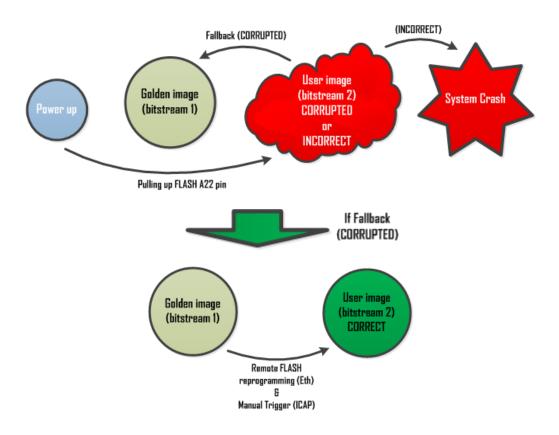


Figure A 6: Multiboot/Safe FPGA configuration system recovery (A22 pin pull-up approach).

Implementation on GLIB and utilisation

Figure A 7 shows the circuitry that implements the above mentioned configuration schemes where the CPLD serves as multiplexor between the FPGA and the FLASH memory. During the first configuration after power up, the A22 pin of the FLASH is pulled high or low, depending on the position of the switch #2 of the DIP switch SW2. If **pulled low** (SW2(2) = on) (default position), the first bitstream loaded by the FPGA is either the FPGA image (single image FPGA configuration scheme) or the Golden image (Multiboot/Safe FPGA configuration scheme with ICAP approach) depending on the FPGA configuration scheme chosen by the user. In case of the Golden image is loaded, the user may jump to the User image when needed by running the appropriate python script (see Table A-1). On the other hand, if the switch is **pulled high** (SW2(2) = off), the first bitstream loaded by the FPGA after power up is the User image (Multiboot/Safe FPGA configuration scheme with A22 pin pull-up approach).

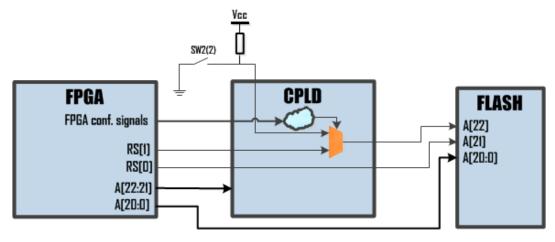


Figure A 7: GLIB configuration scheme.

Regarding to the programming of the FLASH memory, if the user decides to use the Multiboot/Safe FPGA configuration scheme, it is recommended to program the different images by utilising the python scripts provided for this purpose with the GLIB software. There is a programing script for the Golden image as well as other for the User image (see Table A-1). Both scripts use as PROM file a standard MSC file set up for BPI Flash/Single FPGA configuration (the same MCS file set up used for the single image FPGA configuration scheme) which can be generated using iMPACT (see Figure A 8). The user does not need to care about the memory position of the PROM file on the FLASH because it is automatically handled by the scripts.

Table A-1: Multiboot/Safe FPGA configuration main python scripts.

Description	Script name
Golden image programming script	glib_flash_golden_prom_rw.py
User image programming script	glib_flash_user_prom_rw.py
ICAP jump script	glib_icap_jump_to_image.py

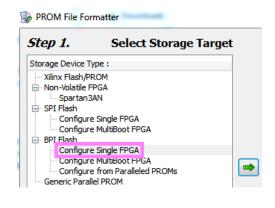


Figure A 8: iMPACT PROM File Formatter.

APPENDIX B: GLIB IP address assignment scheme in a shelf

As mentioned briefly in §2.1.2, the GLIB proposes a simplified scheme of IP address assignment in a shelf based on the slot number (firmware example available in the SVN repository) that is sufficient for most small-scale setups. However, in some complex system implementations, it is also required to have some additional options, such as:

- Arbitrary IP address assignment via IPMI
- IP address assignment via Reverse Address Resolution Protocol (RARP), a feature supported by IPbus.

Moreover, it is also required to be able to switch the IP assignment method dynamically between them without reprogramming. In order to satisfy the above requirements, special GLIB *system_core* firmware and MMC software have been developed.

Figure B 1 shows a diagram of the scheme which is based on the use of the two local I2C buses of the GLIB. Bus #1 (shown in green) was originally foreseen for the communication between the MMC and the three on-board temperature sensors while Bus #2 (shown in red) was intended to be used for the communication between the FPGA and the 24AA025E48 EEPROM for hardware identification reasons, since it contains both a unique 48-bit identifier hardcoded into the chip and the CERN assigned MAC address allocated for the GLIB (08:00:30:f1:00:xx).

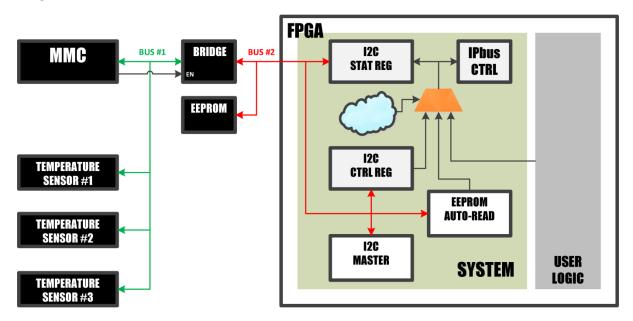


Figure B 1: The IP address assignment scheme

On the FPGA side, apart from the I2C master used for on-demand I2C transactions via IPbus, three additional I2C devices are implemented:

- The EEPROM auto-read: a secondary I2C master that takes over the bus only once after power-up/reset, reads the first 16 bytes of the memory and releases it automatically.
- An I2C slave that implements 16 control registers of 1 byte each.

- An I2C slave that implements 16 status registers of 1 byte each.

Finally, the last I2C component is the I2C bridge located between the two buses that allows the connection or separation of the two I2C buses. The *EN* pin of the bridge is controlled by the MMC.

The IP address assignment scheme functions as following: The MMC, after the handshake with the MCH and before enabling the payload power, ensures that the buses #1 & #2 are disconnected by forcing the *EN* line low, thus avoiding collisions from simultaneous transactions. When the FPGA is loaded, the EEPROM is read automatically and the bus #2 is released. If the contents are valid and the loading *mode* (2-bit setting) is set to "load MAC & IP"², the IP address is forwarded to the IPbus controller. In all other cases, the IPbus controller is using the IP address provided by the user logic. After the board initialization, a new IP address can be assigned via IPMI in two ways:

- By modifying the IP address stored in the EEPROM and powercycle the GLIB in order to load the new settings.
- By assigning an IP address on-the-fly by programming the I2C control registers (without altering the contents of the EEPROM). The settings in that case are volatile and are retained until the next powercycle.

Setting the IP address to 0.0.0.0 initiates the RARP mechanism that assigns IP addresses automatically.

In it important to mention that there is also the possibility for the user to read via IPMI the I2C status registers that report the current IP address loaded via this circuitry (this does not include any IP address loaded via RARP, since this is happening on a different level). Additionally, the circuitry also allows to identify the GLIB cards via IPMI by reading the unique 48-bit identifier hardcoded in the EEPROM as well as its CERN assigned MAC address.

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² By default, the EEPROM is set in "load MAC only" mode, that serves only for loading the pre-programmed MAC address to the IPbus controller, while the IP address is taken by the user logic