5-9 二阶电路

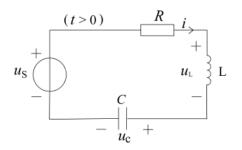
1、二阶电路的方程

RLC 电路,应用 KVL和VCR得:

$$Ri + u_L + u_C = u_S(t)$$
$$i = C \frac{du_C}{dt}$$

$$u_L = L\frac{di}{dt} = LC\frac{d^2u_C}{dt^2}$$

$$LC\frac{d^2u_C}{dt^2} + RC\frac{du_C}{dt} + u_C = u_S(t)$$



含有二个动态元件的线性电路, 称为二阶电路。

其电路方程为二阶线性常微 分方程。

当 $u_s(t)=0$ 时,二阶电路方程是一个齐次方程; 当 $u_s(t)\neq 0$ 时,二阶电路方程是一个非齐次方程。

2、二阶电路的零输入响应

已知

$$u_C(0_+) = U_0 \quad i(0_+) = 0$$

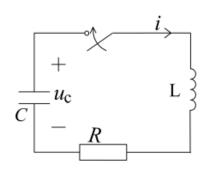
电路方程:

$$Ri + u_L - u_C = 0$$

$$i = -C\frac{du_C}{dt} \qquad u_L = L\frac{di}{dt}$$

以电容电压为变量得:

$$LC\frac{d^2u_C}{dt^2} + RC\frac{du_C}{dt} + u_C = 0$$



这是一个齐次方程。

由于输入激励项为0,所以 u_c 以及其他电路变量称为零输入响应。

以电容电压为变量时的初始条件:

$$u_C(0_+) = U_0 \quad i(0_+) = 0 \qquad \Longrightarrow \qquad \frac{du_C}{dt} \mid_{t=0_+} = 0$$

特征方程:

$$LCp^2 + RCp + 1 = 0$$

特征根:

$$p = \frac{-R \pm \sqrt{R^2 - 4L/C}}{2L} = -\frac{R}{2L} \pm \sqrt{(\frac{R}{2L})^2 - \frac{1}{LC}}$$

2. 零状态响应的三种情况

$$p = \frac{-R \pm \sqrt{R^2 - 4L/C}}{2L} = -\frac{R}{2L} \pm \sqrt{(\frac{R}{2L})^2 - \frac{1}{LC}}$$

$$R > 2\sqrt{\frac{L}{C}} \qquad \text{二个不等负实根} \qquad \text{过阻尼}$$

$$R = 2\sqrt{\frac{L}{C}} \qquad \text{二个相等负实根} \qquad \text{临界阻尼}$$

$$R < 2\sqrt{\frac{L}{C}} \qquad \text{二个共轭复根} \qquad \text{欠阻尼}$$

(1)
$$R > 2\sqrt{\frac{L}{C}}$$

$$u_C = A_1 e^{p_1 t} + A_2 e^{p_2 t}$$

由两个初始条件确定 A_1 和 A_2 。

$$u_C(0_+) = U_0 \to A_1 + A_2 = U_0$$

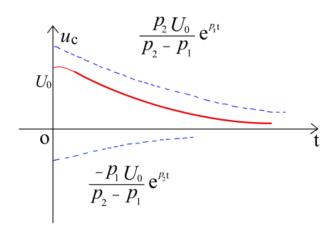
$$\frac{du_C}{dt} \mid_{(0_+)} = 0 \to p_1 A_1 + p_2 A_2 = 0$$

$$\begin{cases} A_1 = \frac{p_2}{p_2 - p_1} U_0 \\ A_2 = \frac{-p_1}{p_2 - p_1} U_0 \end{cases}$$

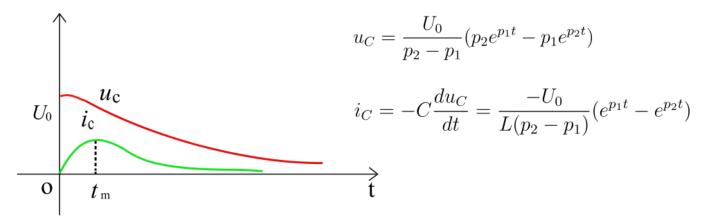
$$u_C = \frac{U_0}{p_2 - p_1} (p_2 e^{p_1 t} - p_1 e^{p_2 t})$$

①电容电压:
$$u_C = \frac{U_0}{p_2 - p_1} (p_2 e^{p_1 t} - p_1 e^{p_2 t})$$

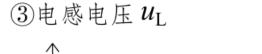
设
$$|p_2| > |p_1|$$

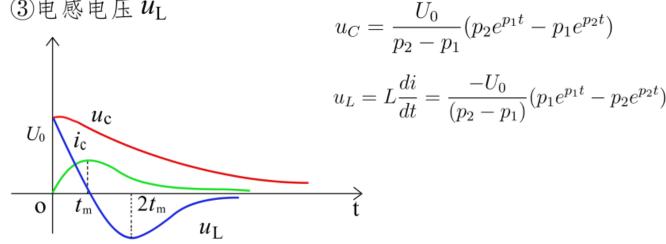


②电容和电感电流



$$t=0_+$$
 $i_c=0_\circ$ $t=\infty$, $i_C=0$ $i_c>0$ $t=t_m$ 时 i_c 最大





$$t=0$$
 , $u_L=U_0$ $t=\infty$, $u_L=0$ $0 < t < t_m$, i 增加 $t > t_m$, i 減小 , $u_L < 0$ $t=2t_m$ 时 $|u_L|$ 最大

$$u_L = L\frac{di}{dt} = \frac{-U_0}{(p_2 - p_1)}(p_1 e^{p_1 t} - p_2 e^{p_2 t})$$

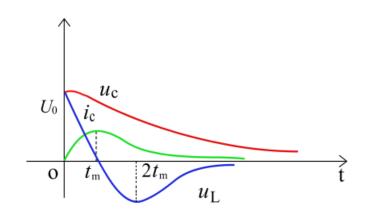
 $i_C=i$ 为极值时,即 $u_L=0$ 时的 t_m 计算如下:

$$(p_1 e^{p_1 t} - p_2 e^{p_2 t}) = 0 \quad \Longrightarrow \quad \frac{p_2}{p_1} = \frac{e^{p_1 t_m}}{e^{p_2 t_m}} \quad \Longrightarrow \quad t_m = \frac{\ln(\frac{p_2}{p_1})}{p_1 - p_2}$$

由 du_L/dt 可确定 u_L 为极小时的t。

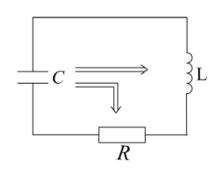
$$(p_1^2 e^{p_1 t} - p_2^2 e^{p_2 t}) = 0 \quad \Longrightarrow \quad t = \frac{2 \ln(\frac{p_2}{p_1})}{p_1 - p_2} \quad \Longrightarrow \quad t = 2t_{\rm m}$$

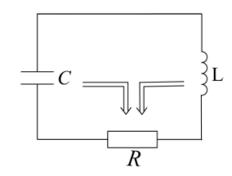
④能量转换关系



 $0 < t < t_{\rm m}$, $u_{\rm C}$ 減小, i 增加。

 $t > t_{\rm m}$, $u_{\rm C}$ 減小, i 減小。





$$(2)R < 2\sqrt{\frac{L}{C}}$$
 $p_{1,2} = -\frac{R}{2L} \pm \sqrt{(\frac{R}{2L})^2 - \frac{1}{LC}}$ 共轭复根

令:
$$\delta = \frac{R}{2L}$$
 (衰减系数), $\omega_0 = \sqrt{\frac{1}{LC}}$ (谐振角频率)

$$\omega = \sqrt{\omega_0^2 - \delta^2}$$
 (固有振荡角频率) $p = -\delta \pm j\omega$

$$u_c$$
的解答形式: $u_C = A_1 e^{p_1 t} + A_2 e^{p_2 t} = e^{-\delta t} (A_1 e^{j\omega t} + A_2 e^{-j\omega t})$

经常写为:
$$u_C = Ae^{-\delta t}\sin(\omega t + \beta)$$

由初始条件
$$\begin{cases} u_C(0^+) = U_0 \to A \sin \beta = U_0 \\ \frac{du_C}{dt}(0^+) = 0 \to A(-\delta) \sin \beta + A\omega \cos \beta = 0 \end{cases}$$

$$\sin \beta = \frac{\omega}{\omega_0}$$

$$A = \frac{\omega_0}{\omega} U_0 \qquad \Longrightarrow \qquad \delta$$

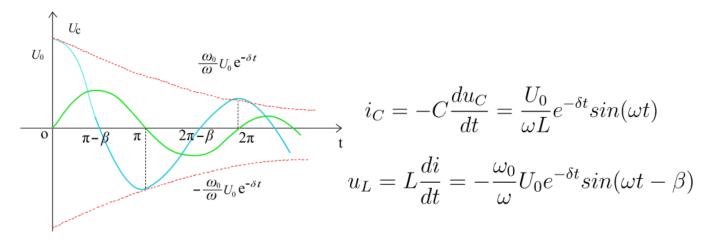
$$\omega = \sqrt{\omega_0^2 - \delta^2} \qquad \delta$$

所以,
$$u_C = \frac{\omega_0}{\omega} U_0 e^{-\delta t} \sin(\omega t + \beta)$$

$$u_C = \frac{\omega_0}{\omega} U_0 e^{-\delta t} \sin(\omega t + \beta)$$

 u_c 是振幅以 $\pm \frac{\omega_0}{\omega} U_0$ 为包络线依指数衰减的正弦函数。

$$t=0$$
 时 $u_{C}=U_{0}$ $u_{C}=0:\omega t=\pi-\beta,\ 2\pi-\beta...n\pi-\beta$



$$(3)R = 2\sqrt{\frac{L}{C}} \qquad p_1 = p_2 = -\frac{R}{2L} = -\delta$$

$$u_C = A_1 e^{-\delta t} + A_2 t e^{-\delta t}$$
相等负实根

由初始条件
$$\begin{cases} u_C(0^+) = U_0 \to A_1 = U_0 \\ \frac{du_C}{dt}(0^+) = 0 \to A_1(-\delta) + A_2 = 0 \end{cases}$$

$$\begin{cases} A_1 = U_0 \\ A_2 = U_0 \delta \end{cases}$$

小结:
$$R > 2\sqrt{\frac{L}{C}}$$
 过阻尼,非振荡放电 $u_C = A_1 e^{p_1 t} + A_2 e^{p_2 t}$ 临界阻尼,非振荡放电 $u_C = A_1 e^{-\delta t} + A_2 t e^{-\delta t}$

$$R < 2\sqrt{\frac{L}{C}}$$
 欠阻尼,振荡放电
$$u_C = Ae^{-\delta t}sin(\omega t + \beta)$$

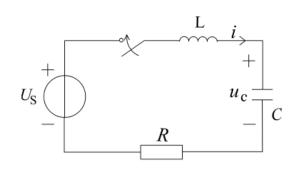
由初始条件
$$\begin{cases} u_C(0_+) \\ \frac{\mathrm{d}u_C}{\mathrm{d}t}(0_+) \end{cases}$$
 定常数

3、二阶电路零状态响应和全响应

微分方程为

$$LC\frac{d^2u_C}{dt^2} + RC\frac{du_C}{dt} + u_C = U_S \qquad U_S$$

$$u_C = u_C' + u_C''$$



特征方程为

$$LCp^2 + RCp + 1 = 0$$

特解
$$u'_C = U_S$$

uc解答形式为

 $u_C = U_S + A_1 e^{p_1 t} + A_2 e^{p_2 t} (p_1 \neq p_2)$ 过阻尼非振荡充电或放电 $u_C = U_S + A_1 e^{-\delta t} + A_2 t e^{-\delta t} (p_1 = p_2 = -\delta)$ 临界阻尼非振荡充电或放电 $u_C = U_S + A e^{-\delta t} \sin(\omega t + \beta) (p_{1,2} = -\delta \pm j\omega)$ 欠阻尼振荡充电或放电 由初值 $u_C(0^+)$, $\frac{du_C}{dt}(0^+)$ 确定两个常数

当 $u_{\rm C}(0_+)$ 和 $i_{\rm L}(0_+)$ 均为零时,则为零状态响应; 当 $u_{\rm C}(0_+)$ 和 $i_{\rm L}(0_+)$ 至少有一个不为零,则为全响应。