

Chapter 2

Quantization

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1. Quantization process



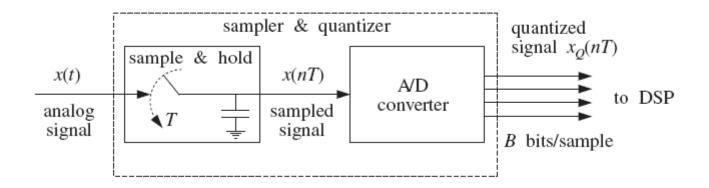


Fig: Analog to digital conversion

- * The quantized sample $x_Q(nT)$ is represented by **B** bit, which can take 2^B possible values.
- ❖ An A/D is characterized by a **full-scale range R** which is divided into 2^B quantization levels. Typical values of R in practice are between 1-10 volts.

1. Quantization process



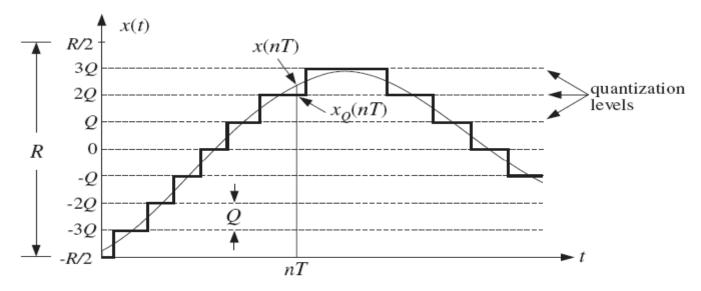


Fig: Signal quantization

- Quantizer resolution or quantization width $Q = \frac{R}{2^B}$
- A bipolar ADC $-\frac{R}{2} \le x_Q(nT) < \frac{R}{2}$
- A unipolar ADC $0 \le x_Q(nT) < R$

1. Quantization process –Quantization error



- * Quantization by rounding: replace each value x(nT) by the nearest quantization level.
- ❖ Quantization by truncation: replace each value x(nT) by its below quantization level.
- Quantization error: $e(nT) = x_Q(nT) x(nT)$
- ❖ Consider rounding quantization: $-\frac{Q}{2} \le e \le \frac{Q}{2}$

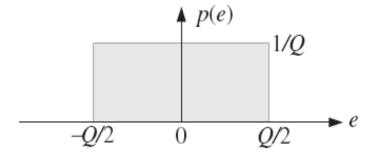


Fig: Uniform probability density of quantization error

1. Quantization process –Quantization error



- * The mean value of quantization error $\overline{e} = \int_{-Q/2}^{Q/2} ep(e)de = \int_{-Q/2}^{Q/2} e^{-Q/2} de = 0$
- The mean-square error (power) $\sigma^2 = \overline{e^2} = \int_{-Q/2}^{Q/2} e^2 p(e) de = \int_{-Q/2}^{Q/2} e^2 \frac{1}{Q} de = \frac{Q^2}{12}$
- Root-mean-square (rms) error: $e_{rms} = \sigma = \sqrt{\overline{e^2}} = \frac{Q}{\sqrt{12}}$
- * R and Q are the ranges of the signal and quantization noise, then the signal to noise ratio (SNR) or dynamic range of the quantizer is defined as

$$SNR_{dB} = 20\log_{10}\left(\frac{R}{Q}\right) = 20\log_{10}(2^{B}) = B\log_{10}(2) = 6B dB$$

which is referred to as 6 dB bit rule.

1. Quantization process –Example



❖ In a digital audio application, the signal is sampled at a rate of 44 KHz and each sample quantized using an A/D converter having a full-scale range of 10 volts. Determine the number of bits B if the rms quantizzation error mush be kept below 50 microvolts. Then, determine the actual rms error and the bit rate in bits per second.

2. Digital to Analog Converters (DACs)



❖ We begin with A/D converters, because they are used as the building blocks of successive approximation ADCs.

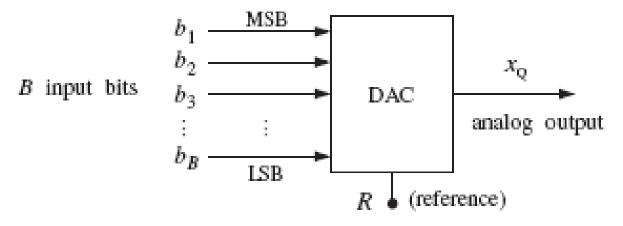


Fig: B-bit D/A converter

- ❖ Vector B input bits : $b=[b_1, b_2,...,b_B]$. Note that b_B is the least significant bit (LSB) while b_1 is the most significant bit (MSB).
- For unipolar signal, $x_Q \in [0, R)$; for bipolar $x_Q \in [-R/2, R/2)$.

2. DAC-Example DAC Circuit



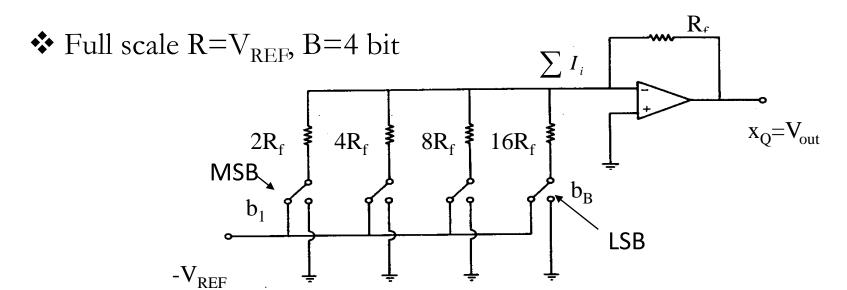


Fig: DAC using binary weighted resistor

$$\begin{split} \sum I = V_{REF} \left(\frac{b_1}{2R_f} + \frac{b_2}{4R_f} + \frac{b_3}{8R_f} + \frac{b_4}{16R_f} \right) \\ x_Q = V_{OUT} = \sum I \cdot R_f = V_{REF} \left(\frac{b_1}{2} + \frac{b_2}{4} + \frac{b_3}{8} + \frac{b_4}{16} \right) \\ x_Q = R2^{-4} \left(b_1 2^{-3} + b_2 2^{-2} + b_3 2^{-1} + b_4 2^{0} \right) = Q \left(b_1 2^{-3} + b_2 2^{-2} + b_3 2^{-1} + b_4 2^{0} \right) \end{split}$$

2. D/A Converters



\rightharpoonup Unipolar natural binary $x_Q = R(b_1 2^{-1} + b_2 2^{-2} + ... + b_B 2^{-B}) = Qm$

where m is the integer whose binary representation is $b=[b_1, b_2,...,b_B]$.

$$m = b_1 2^{B-1} + b_2 2^{B-2} + \dots + b_B 2^0$$

❖ Bipolar offset binary: obtained by shifting the x_Q of unipolar natural binary converter by half-scale R/2:

$$x_Q = R(b_1 2^{-1} + b_2 2^{-2} + \dots + b_B 2^{-B}) - \frac{R}{2} = Qm - \frac{R}{2}$$

* Two's complement code: obtained from the offset binary code by complementing the most significant bit, i.e., replacing b_1 by $\overline{b_1} = 1 - b_1$.

$$x_Q = R(\overline{b_1}2^{-1} + b_22^{-2} + \dots + b_B2^{-B}) - \frac{R}{2}$$

2. D/A Converters-Example



- ❖ A 4-bit D/A converter has a full-scale R=10 volts. Find the quantized analog values for the following cases?
- a) Natural binary with the input bits b=[1001]?
- b) Offset binary with the input bits b=[1011]?
- c) Two's complement binary with the input bits b=[1101]?



❖ A/D converters quantize an analog value x so that is is represented by B bits $b=[b_1, b_2,...,b_B]$.

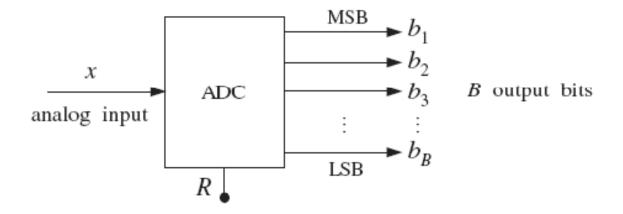


Fig: B-bit A/D converter



❖ One of the most popular converters is the successive approximation A/D converter

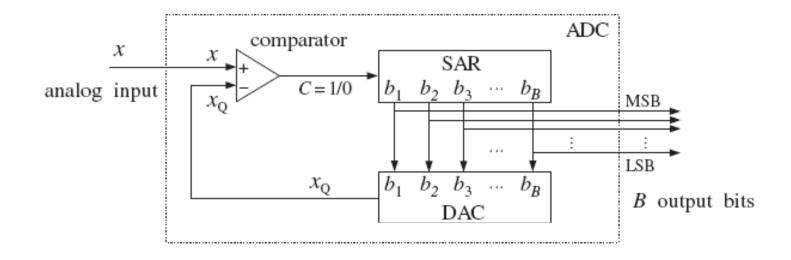


Fig: Successive approximation A/D converter

After B tests, the successive approximation register (SAR) will hold the correct bit vector b.



Successive approximation algorithm

for each
$$x$$
 to be converted, do:
initialize $\mathbf{b} = [0, 0, ..., 0]$
for $i = 1, 2, ..., B$ do:
 $b_i = 1$
 $x_Q = \text{dac}(\mathbf{b}, B, R)$
 $b_i = u(x - x_Q)$

where the unit-step function is defined by
$$u(x) = \begin{cases} 1 & \text{if } x \ge 0 \\ 0 & \text{if } x < 0 \end{cases}$$

This algorithm is applied for the <u>natural and offset</u> binary with truncation quantization.

3. A/D converter-Example



❖ Consider a 4-bit ADC with the full-scale R=10 volts. Using the successive approximation algorithm to find offset binary of truncation quantization for the analog values x=3.5 volts and x=-1.5 volts.



❖ For <u>rounding</u> quantization, we shift x by Q/2:

for each x to be converted, do:

$$y = x + Q/2$$

initialize $\mathbf{b} = [0, 0, ..., 0]$
for $i = 1, 2, ..., B$ do:
 $b_i = 1$
 $x_Q = \text{dac}(\mathbf{b}, B, R)$
 $b_i = u(y - x_Q)$

For the two's complement code, the sign bit b_1 is treated separately.

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for each x to be converted, do:

y = x + Q/2

initialize \mathbf{b} = [0, 0, ..., 0]

b_1 = 1 - u(y)

for i = 2, 3, ..., B do:

b_i = 1

x_Q = \text{dac}(\mathbf{b}, B, R)

b_i - u(y - x_Q)
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3. A/D converter-Example



❖ Consider a 4-bit ADC with the full-scale R=10 volts. Using the successive approximation algorithm to find offset and two's complement of rounding quantization for the analog values x=3.5 volts .

Homework



Problems 2.1, 2.2, 2.3, 2.5, 2.6