

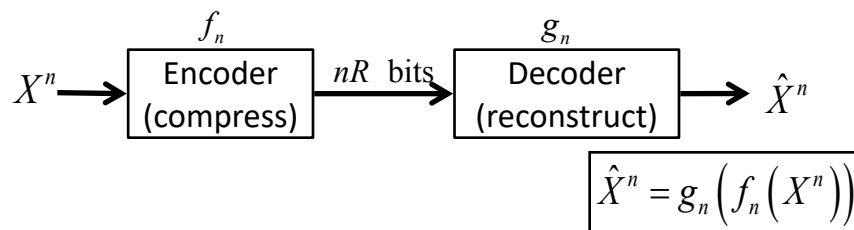
EE231A Information Theory  
Lecture 13:  
The Rate Distortion Function

- A. Rate vs Distortion for Lossy Compression
- B. Calculation of the Rate vs. Distortion Function
- C.  $R(D)$  for a Gaussian source

Part 13A:  
Rate vs. Distortion  
for  
Lossy Compression

## Lossy Compression

- $X_1, X_2, \dots, X_n$  are i.i.d. from alphabet  $\mathcal{X}$ ,



- If  $R < H(X)$ , such a scheme cannot guarantee that  $\hat{X}^n = X^n$ . There is a loss in fidelity.

## Measuring Distortion

- Distortion Metrics  $d: \mathcal{X} \times \hat{\mathcal{X}} \rightarrow \mathbb{R}^+$
- Maximum Distortion  $d_{\max} = \max_{x \in \mathcal{X}, \hat{x} \in \hat{\mathcal{X}}} d(x, \hat{x})$
- Bounded distortion  $d_{\max} < \infty$

## Hamming Distortion

$$d(x, \hat{x}) = \begin{cases} 0 & \hat{x} = x \\ 1 & \hat{x} \neq x \end{cases}$$

- For Hamming distortion  $E[d(x, \hat{x})] = P(\hat{X} \neq X)$

## Squared-Error Distortion

$$d(x, \hat{x}) = (x - \hat{x})^2$$

## Distortion between sequences

- Distortion between sequences

$$d(x^n, \hat{x}^n) = \frac{1}{n} \sum_{i=1}^n d(x_i, \hat{x}_i)$$

- Other definitions are possible...

## Distortion of block-based lossy compression

$$E\left[d\left(X^n, g_n\left(f_n(X^n)\right)\right)\right] = \sum_{x^n} p(x^n) d\left(X^n, g_n\left(f_n(X^n)\right)\right)$$

## Rate Distortion Pair

- A rate, distortion pair  $(R, D)$  is *achievable* if there is a sequence of  $(2^{nR}, n)$  lossy compression codes  $(f_n, g_n)$  with

$$\lim_{n \rightarrow \infty} E \left[ d \left( X^n, g_n \left( f_n(X^n) \right) \right) \right] \leq D \quad .$$

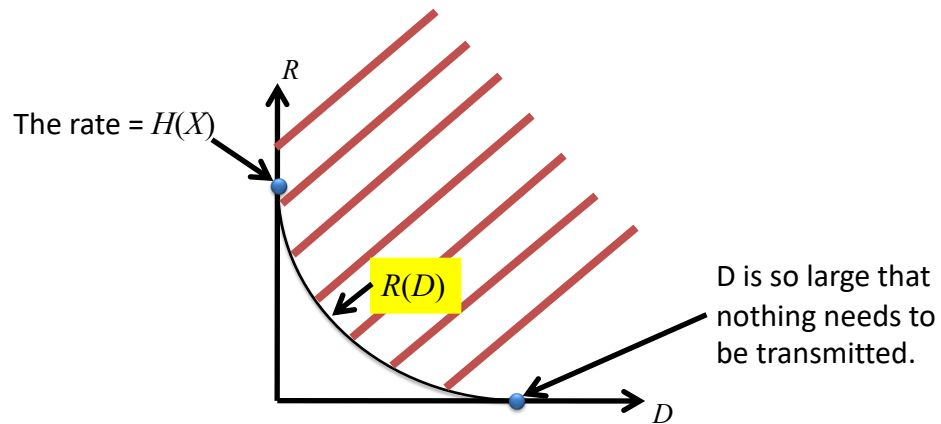
## Rate-Distortion function

- For  $X$  i.i.d.  $\sim p(x)$  and  $d(x, \hat{x})$  bounded

$$R(D) = \min_{p(\hat{x}|x), E[d] \leq D} I(X; \hat{X})$$

## Achievable Pairs $R, D$

- Rate distortion function bounds achievable pairs  $R, D$



## Part 13B: Calculation of the Rate Distortion Function

## Rate-Distortion function

- For  $X$  i.i.d.  $\sim p(x)$  and  $d(x, \hat{x})$  bounded

$$R(D) = \min_{p(\hat{x}|x), E[d] \leq D} I(X; \hat{X})$$

$$\begin{aligned} E[d] &= \sum_{x, \hat{x}} p(x, \hat{x}) d(x, \hat{x}) \\ &= \sum_{x, \hat{x}} p(x) \underbrace{p(\hat{x} | x)}_{\text{This is what we control.}} d(x, \hat{x}) \end{aligned}$$

## Proof will come later

- Proof that  $R(D)$  is achievable is in a future lecture.
- Today, we focus on how to compute  $R(D)$ .

## Computing $R(D)$

- *First example:*  $R(D)$  for a binary source with Hamming distortion:  $P(x=1) = p \leq \frac{1}{2}$
- One way to find  $R(D)$  is to find a lower bound on  $I(X; \hat{X})$  and then achieve it.
- We have the constraint  $E[d] = P(\hat{X} \neq X) \leq D$

## Lower bound on $I(X; \hat{X})$

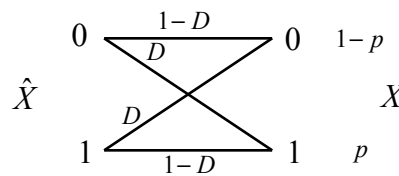
$$\begin{aligned}
 I(X; \hat{X}) &= H(X) - H(X | \hat{X}) \\
 &= H(p) - H(X \oplus \hat{X} | \hat{X}) \\
 &\geq H(p) - H(X \oplus \hat{X}) \\
 &= H(p) - H(P(X \neq \hat{X})) \\
 &\geq H(p) - H(D) \quad \text{for } D \leq \frac{1}{2}
 \end{aligned}$$



## Achievability of the lower bound

- So  $R(D) \geq H(p) - H(D)$
- Can we achieve  $I(X; \hat{X}) = H(p) - H(D)$  with  $E(d) \leq D$  ?
- We need to find a  $p(x, \hat{x})$  that does that.

## The Test Channel



- $H(p) - H(D)$  is the  $I(X; \hat{X})$  for a BSC with transition probability  $D$  and output distribution  $p, 1-p$ .
- Can we find an input distribution  $p(\hat{x})$  to make it work?

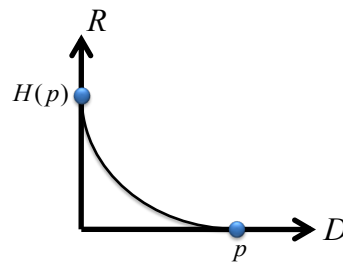
### Achievability of the lower bound (cont.)

$$\begin{aligned} P(X=0) &= (1-D)P(\hat{X}=0) + DP(\hat{X}=1) \\ &= 1-p \end{aligned}$$

$$\begin{aligned} P(X=1) &= (1-D)P(\hat{X}=1) + DP(\hat{X}=0) \\ &= p \end{aligned}$$

$$\begin{aligned} p(\hat{x}=0) &= \frac{1-p-D}{1-2D} \\ p(\hat{x}=1) &= \frac{p-D}{1-2D} \end{aligned}$$

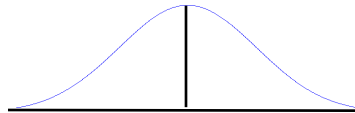
- So  $R(D) = H(p) - H(D)$ .



Part 13C:  
 $R(D)$  for a Gaussian source

## One bit quantization of Gaussian

- $R(D) = \min_{p(\hat{x}|x), E[d] \leq D} I(X; \hat{X})$
- Consider one bit quantization of a Gaussian.



- Optimal 1-bit quantization

$$S_1 = \{x : x > 0\} \quad \hat{x}_1 = \sqrt{2/\pi}\sigma \quad \text{(Conditional mean)}$$

$$S_2 = \{x : x \leq 0\} \quad \hat{x}_2 = -\sqrt{2/\pi}\sigma$$

$$E[(x - \hat{x}_i(x))^2] = \frac{\pi - 2}{\pi} \sigma^2 = 0.36\sigma^2$$

## $R(D)$ for Gaussian source

- $R(D)$  for  $X_1, X_2, \dots$  i.i.d.  $N(0, \sigma^2)$
- Find a lower bound on  $I(X; \hat{X})$  and achieve it.

$$\begin{aligned} I(X; \hat{X}) &= h(X) - h(X | \hat{X}) \\ &= \frac{1}{2} \log(2\pi e \sigma^2) - h(X - \hat{X} | \hat{X}) \\ &\geq \frac{1}{2} \log(2\pi e \sigma^2) - h(X - \hat{X}) \\ &\geq \frac{1}{2} \log(2\pi e \sigma^2) - h(\mathcal{N}(0, E(X - \hat{X})^2)) \\ &\geq \frac{1}{2} \log(2\pi e \sigma^2) - h(\mathcal{N}(0, D)) \end{aligned}$$

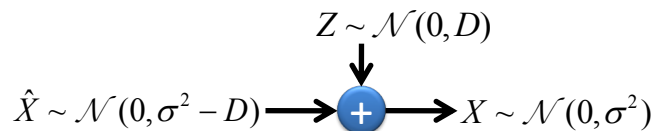
## Gaussian test channel

$$I(X; \hat{X}) \geq \frac{1}{2} \log(2\pi e \sigma^2) - h(\mathcal{N}(0, D))$$

$$= \frac{1}{2} \log\left(\frac{\sigma^2}{D}\right)$$

$$= \frac{1}{2} \log\left(1 + \frac{\sigma^2 - D}{D}\right)$$

$$R(D) = \frac{1}{2} \log\left(\frac{\sigma^2}{D}\right)$$



## $R(D)$ for Gaussian channel (cont.)

- So we can achieve the lower bound.

$$\begin{aligned} R(D) \Big|_{D=0.36\sigma^2} &= \frac{1}{2} \log\left(\frac{1}{0.36}\right) \\ &= 0.737 \text{ bits} \end{aligned}$$

$$\text{or } R(0.25\sigma^2) = 1 \text{ bit}$$

So Information Theory indicates we can do a better job than optimal one-bit quantization by considering multiple symbols at a time.