

# Stochastic Processes

## Session 9 — Lecture

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# Outline for Session 9 — Lecture

Linear time-invariant system with WSS input

1st and 2nd order characterization

Summary of results

Examples

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Linear time-invariant system with WSS input

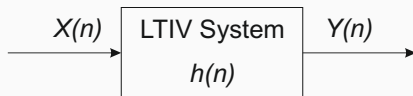
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## Linear time-invariant system with WSS input

We consider a *Linear time-invariant (LTIV) system* with input  $\{X(n)\}$  and output  $\{Y(n)\}$ :



From linear system theory, we know that the input-output relation of a LTIV system is in the form of a *convolution* with an *impulse response*  $h(n)$ :

$$Y(n) = \sum_{k=-\infty}^{\infty} h(k)X(n-k) = h(n) \star X(n)$$

We consider for a weak sense stationary (WSS)  $\{X(n)\}$  with mean  $\mu_X$  and autocorrelation  $R_X(k)$ , the questions:

- ▶ How is the mean of  $\{Y(n)\}$  related to properties of  $\{X(n)\}$ ?
- ▶ What is the autocorrelation of  $\{Y(n)\}$ — how does it relate to properties of  $\{X(n)\}$ ?
- ▶ Is  $\{Y(n)\}$  WSS?
- ▶ If so, what can we say about the PSD?

# Convolution and Fourier Relations

Knowledge of linear systems and Fourier transforms are prerequisites for the course. In this lecture, we make use of the following:

*Convolution* is a linear operator defined as:

$$g(n) \star h(n) := \sum_{k=-\infty}^{\infty} g(k)h(n-k) = \sum_{i=-\infty}^{\infty} g(n-i)h(i) = h(n) \star g(n)$$

*Discrete Fourier transform* is a linear operator defined as:

$$\mathcal{F}\{g(n)\}(f) := \sum_{n=-\infty}^{+\infty} g(n)e^{-j2\pi nf} = G(f).$$

Fourier transforms of convolutions simplify due to the *Convolution Theorem*:

$$\mathcal{F}\{g(n) \star h(n)\}(f) = G(f) \cdot H(f)$$

Fourier transforms of signals flipped in time (*Time Reversal Identity*):

$$\mathcal{F}\{g(-n)\}(f) = G^*(f)$$

where  $*$  denotes complex conjugation.

In combination, the convolution theorem and the time reversal identity, give

$$\mathcal{F}\{g(n) \star g(-n)\}(f) = G(f) \cdot G^*(f) = |G(f)|^2$$

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## Mean function of $\{Y(n)\}$

We compute the mean function of the output  $\{Y(n)\}$  as<sup>1</sup>

$$\begin{aligned}\mu_Y &:= \mathbb{E}[Y(n)] = \mathbb{E}[h(n) \star X(n)] \\ &= \mathbb{E}\left[\sum_{j=-\infty}^{\infty} h(j)X(n-j)\right] \\ &= \sum_j h(j)\mathbb{E}[X(n-j)] = \mu_X \sum_j h(j).\end{aligned}$$

The mean of  $\{Y(n)\}$  is constant and can be related to the mean of  $\{X(n)\}$  and the impulse response  $h(n)$ .

It is sometimes more convenient to write this relation in terms of the systems *transfer function*  $H(f) = \mathcal{F}\{h(n)\}(f)$ :

$$\mu_Y = H(0) \cdot \mu_X$$

*Check this relation!*

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<sup>1</sup>If not otherwise mentioned, the summations are from  $-\infty$  to  $+\infty$ .

## Autocorrelation function for $\{Y(n)\}$ — Part 1 of 3

The derivation of ACF the output process is tedious but straightforward: Insert  $\{Y(n)\}$  in the definition of autocorrelation and then simplify (carefully).

Since  $\{Y(n)\}$  could be nonstationary, we use the general form of the ACF:

$$\begin{aligned} R_Y(n, n+k) &= \mathbb{E}[Y(n)Y(n+k)] \\ &= \mathbb{E}[Y(n) \sum_j h(j)X(n+k-j)] \\ &= \sum_j h(j) \underbrace{\mathbb{E}[Y(n)X(n+k-j)]}_{=: R_{YX}(n, n+k-j)} \\ &= \sum_j h(j) R_{YX}(n, n+k-j) = h(k) \star R_{YX}(n, n+k) \end{aligned}$$

where it still remains to derive the *cross-correlation function*  $R_{YX}(n, n+k)$ .



## Autocorrelation function for $\{Y(n)\}$ — Part 2 of 3

We now derive an expression for the *cross-correlation function*  $R_{YX}(n, n+k)$ :

$$\begin{aligned} R_{YX}(n, n+k) &= \mathbb{E}[Y(n)X(n+k)] \\ &= \mathbb{E}\left[\sum_j h(j)X(n-j)X(n+k)\right] \\ &= \sum_j h(j) \underbrace{\mathbb{E}[X(n-j)X(n+k)]}_{=R_X(k+j)} \\ &= \sum_j h(j)R_X(k+j) \end{aligned}$$

This is a function of  $k$  only, and thus we write  $R_{YX}(n, n+k) = R_{YX}(k)$ .  
Then the cross-correlation can be written in the form

$$R_{YX}(k) = h(-k) \star R_X(k)$$

## Autocorrelation function for $\{Y(n)\}$ — Part 3 of 3

Combining the pieces of the derivation of the ACF we finally achieve:

$$\begin{aligned}R_Y(n, n+k) &= h(k) \star R_{YX}(n, n+k) \\&= h(k) \star R_{YX}(k) \\&= h(k) \star h(-k) \star R_X(k).\end{aligned}$$

Since it has constant mean and its ACF is a function of  $k$  only, *the output process is WSS* with autocorrelation

$$R_Y(k) = h(k) \star h(-k) \star R_X(k).$$

**Remark:** The symmetric and positive semidefinite function  $h(k) \star h(-k)$  is some-times called the “autocorrelation of the impulse response  $h(k)$ ” and denoted by  $R_h(k)$ . This is because

$$h(k) \star h(-k) = \sum_j h(j)h(j+k).$$

The term is used for historical reasons, but is strictly speaking a misnomer, since it is no in accordance with the definition of ACF.

## Power spectral density of the output process

Since the output process is WSS, its power spectral density is well defined. Starting from the autocorrelation of  $\{Y(n)\}$

$$R_Y(k) = h(k) \star h(-k) \star R_X(k)$$

we achieve by Fourier transforming and using the convolution theorem and time reversal identity the relation:

$$S_Y(f) = H(f) \cdot H^*(f) \cdot S_X(f) = |H(f)|^2 S_X(f).$$

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## Summary of results: 1st and 2nd order input-output relation

	Input	Output
Signal	$X(n)$	$Y(n) = h(n) \star X(n)$
Mean	$\mu_X$	$\mu_Y = \mu_X \sum_n h(n) = H(0)\mu_X$
ACF	$R_X(k)$	$R_Y(k) = h(k) \star h(-k) \star R_X(k)$
WSS	Yes	Yes
PSD	$S_X(f)$	$S_Y(f) =  H(f) ^2 S_X(f)$

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## Example: MA(1) process with zero mean

As a first example, we analyse the MA(1) process defined as

$$Y(n) = X(n) + X(n-1), \quad \{X(n)\} \stackrel{iid.}{\sim} \mathcal{N}(0, 3)$$

Thus,  $X(n)$  is a white Gaussian process with ACF  $R_X(k) = 3\delta(k)$ .

We can write the output process as  $Y(n) = h(n) \star X(n)$  with impulse response  $h(n) = \delta(n) + \delta(n-1)$ .

**Mean:**  $\mathbb{E}[Y(n)] = \mathbb{E}[X(n)] \cdot (h(0) + h(1)) = 0 \cdot 2 = 0$

**ACF:** By noting that  $h(k) \star h(-k) = \delta(k+1) + 2\delta(k) + 1\delta(k-1)$  we have the ACF:

$$R_Y(k) = [\delta(k+1) + 2\delta(k) + \delta(k-1)] \cdot 3 = 3\delta(k+1) + 6\delta(k) + 3\delta(k-1)$$

**PSD :** Note that  $X(f)$  has PSD  $S_X(f) = 3$ . The transfer function reads  $H(f) = 1 + \exp(-j2\pi f)$  and thus the PSD of  $Y(n)$  is

$$S_Y(f) = |1 + \exp(-j2\pi f)|^2 3 = 6 \cdot (1 + \cos(2\pi f))$$

## Example: MA(1) process with non-zero mean

As a second example, we analyse the MA(1) process defined as

$$Y(n) = X(n) + X(n-1), \quad \{X(n)\} \stackrel{iid.}{\sim} \mathcal{N}(3, 15)$$

Here, the ACF of  $X(n)$  is  $R_X(k) = 9 + 15\delta(k)$  and we observe that  $Y(n) = h(n) \star X(n)$  for  $h(n) = \delta(n) + \delta(n-1)$ .

**Mean:**  $\mathbb{E}[Y(n)] = \mathbb{E}[X(n)] \cdot (h(0) + h(1)) = 3 \cdot 2 = 6$

**ACF:** Since  $h(k) \star h(-k) = \delta(k+1) + 2\delta(k) + 1\delta(k-1)$ , we have

$$\begin{aligned} R_Y(k) &= (9 + 15\delta(k+1)) + 2 \cdot (9 + 15\delta(k)) + (9 + 15\delta(k-1)) \\ &= 36 + 15\delta(k+1) + 30\delta(k) + 15\delta(k-1). \end{aligned}$$

**PSD:** The PSD for  $X(f)$  is  $S_X(f) = 15 + 9\delta(f)$  and the transfer function reads  $H(f) = 1 + \exp(-j2\pi f)$ , thus the PSD for  $Y(n)$  is <sup>2</sup>

$$\begin{aligned} S_Y(f) &= |1 + \exp(-j\pi f)|^2 (15 + 9\delta(f)) \\ &= 2 \cdot (1 + \cos(2\pi f)) \cdot (15 + 9\delta(f)) \\ &= 30 \cdot (1 + \cos(2\pi f)) + 36\delta(f) \end{aligned}$$

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<sup>2</sup>The Dirac impulse  $36\delta(f)$  is due to the non-zero mean of  $Y(n)$ . The coefficient 36 equals the squared mean.