· Mechanical Translational Systems:

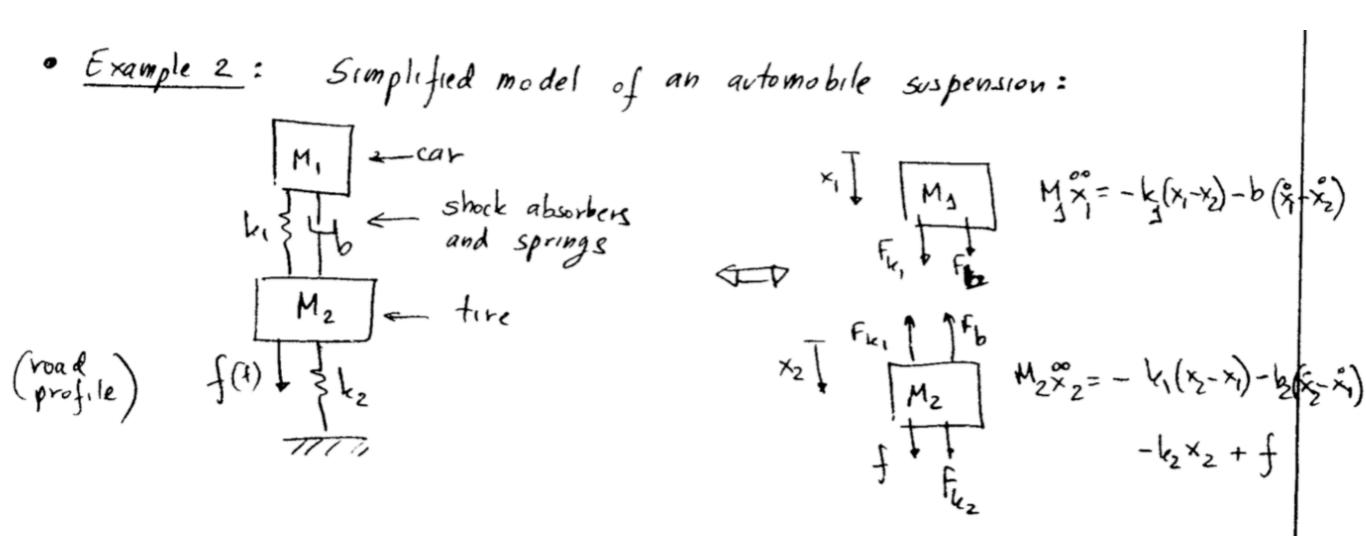
Elements:

a) Spring
$$F = -kx$$

b) Viscois damping $F = -bx$ (always apposes motion)

$$M_{x}^{\infty} = -kx - bx$$

$$M_{x}^{\circ} + b_{x}^{\circ} + k_{x} = 0$$



Example 2: Simplified model of an automobile suspension:

$$M_1 = -car$$
 $k_1 = k_2$
 $M_2 = tire$

Fig. 1 | Fib. |

 $M_2 = k_2$
 $M_3 = k_2$
 $M_4 = k_2$
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· Mechanical Rotational Systems

Basic Law: Newton's equation for rotational systems: JO = & Torques

Elements: moment of mertia (similar to mass)
friction
torsion

Torsion spring: 2 0,

$$7 = k(o_2 - o_1)$$

$$Z = B(\hat{o_2} - \hat{o_1})$$

Satellite attitude control: (with torque applied by

$$\frac{Jd^20}{dt^2} = Z$$

(or, in Laplace domain: $Js^20=Z \Rightarrow 0 = \frac{1}{Js^2} 7$:
essentially a dable integrator)

1) Electrical equation:
$$e_{a} = Rm I_{a} + I_{m} \frac{d i_{a}}{dt} + e_{m} \implies E_{a}(s) = (s I_{m} + R_{m}) I_{(s)} + E_{m}(s)$$

2) Back emf: $e_{m} = k_{m} \circ \implies E_{m}(s) = s k_{m} \circ (s)$

3) Mechanical equation: $Z = k_{z} I_{a}$

4) Newton's second equation: $J_{z} \frac{d \circ}{dt^{z}} + \beta_{z} \circ = Z \implies (s^{z} J_{z} + s \beta_{z}) \circ = Z(s)$

Rithing everything together yields the following block diagram:

$$E_{a} = \sum_{k=1}^{\infty} \frac{1}{s I_{m} + R_{m}} \frac{1}{s I_{m} + S \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z} + s \beta_{z}} = \sum_{k=1}^{\infty} \frac{1}{s^{z} J_{z} + s \beta_{z}} \frac{1}{s^{z} J_{z}$$

Surprise! The system has built-in feedback (through the back emf)

- φ : How do we find the transfer function from $E_a(i)$ to O(i)?
- A: We could try solving the 4 simultaneous equations (messy) or applying Mason's formula to the loop above. The latter approach yields:

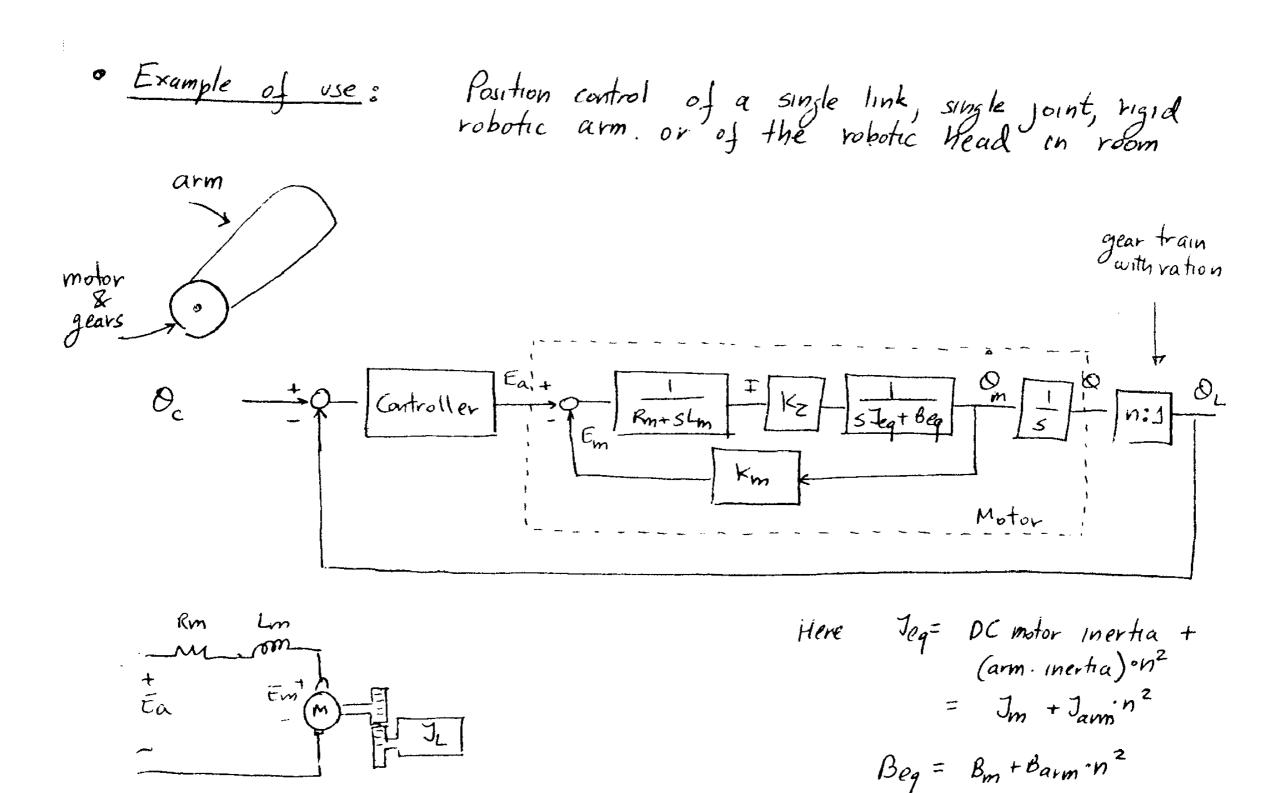
$$G(S) = G_1(S)$$

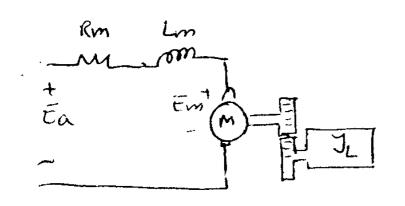
$$4 + 6 k_m G_1(S)$$
where
$$G_1(S) = \frac{K_Z}{(S L_m + R_m)(S^2 J_L + S B_L)}$$

so in principle we get a third order system.

Common simplifying assumption: neglect
$$L_m$$
 ($SL_m \ge 0$) \Rightarrow

$$G(s) = \frac{O(s)}{E_a(s)} = \frac{k_z}{R_m s} \frac{1}{(sJ_z + B_L)} = \frac{k_z}{R_m s} \frac{1}{(sJ_z + B_L)} \frac{1}{R_m s} \frac{1}{(sJ_z + B_L)} = \frac{K_z}{R_m s} \frac{1}{(sJ_z + B_L)} \frac{1}{R_m s} \frac{1}{(sJ_z + B_L)} = \frac{K_z}{S(sJ_z + B_L)} \frac{1}{S(sJ_z + B_L)}$$
(looks like the cascade of a pure integrator and a first order lag)





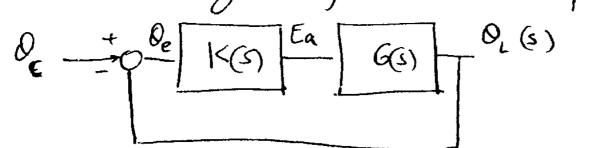
Here
$$J_{eq} = DC motor Inertia + (arm · Inertia) · n^2$$

$$= J_m + J_{arm} · n^2$$

$$Beg = B_m + Barm · n^2$$

Again, you get a third order system where you neglect Lm

The block diagram of the closed-loop system is given by:



ahere k(s) is the transfer function of the controller and 6(s) is the T.F. of the arm (including) reduction gears)

To find the closed-loop transfer function $\frac{\partial L}{\partial c}$ we could, for instance write down the equations:

$$\frac{\partial L}{\partial c}$$
 we could, for

$$\mathcal{O}_{e} = \mathcal{Q}_{c} - \mathcal{Q}_{L}, \qquad \mathcal{O}_{L} = 60 \text{ K(s) } \mathcal{O}_{e}$$

Eliminating
$$Q_e$$
 yields: $6 \cdot K(Q_c - Q_L) = Q_L$ / $Q_c = \frac{G \cdot K}{1 + G \cdot K}$

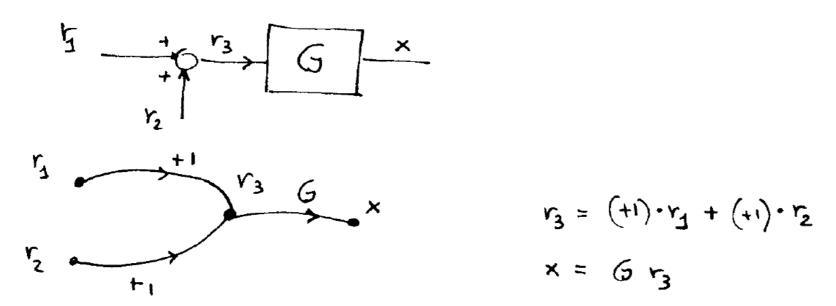
This is a special case of Meson's formula:

They provide an alternative representation of Transfer Function relationships and an alternative (often simpler) to Cramer's rule or block diagram manipulations for computing T.F.

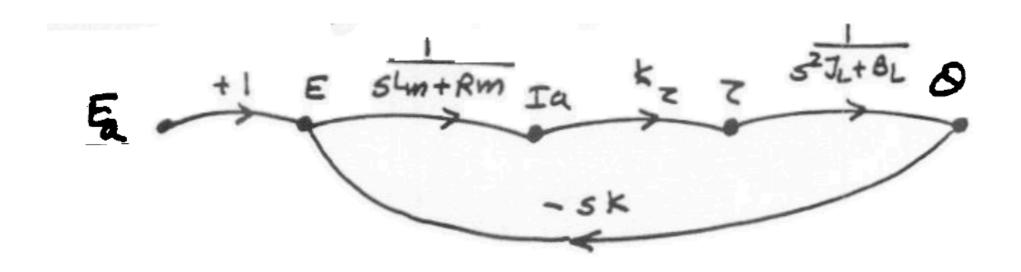
Rules:

- · Each signal is represented by a node
- · Each transfer function is represented by a branch (arrow)

· Summing junctions are represented implicitly: all the inputs converging to a node are added together:



Question: Find the signal flow graph representation of the DC motor:



Some Termmology:

source node: A node that has all synals Howing away from it.

sink node: A node with incoming squals

Path:

Continuous connection of branches between 2 nodes (directed)

Loop: Closed path (i.e starting node = finishing mode)

 \Leftrightarrow

Path (100p) 6ain: froduct of all T.F. of all the branches

Non Touching loops: Loops that do not have any nodes in common.

• Mason's Formula (section 2.4) Browndes an alternative to Cramer's rule or elimination for finding Transfer Functions

$$T_{CR} = \frac{1}{\Lambda} \sum_{k=1}^{P} M_k \Delta_k = \frac{1}{\Lambda} (M_3 \Delta_3^{-1} + M_p \Delta_p)$$

• Mason's Formula (section 2.4) Browides an alternative to Cramer's rule or elimination for finding Transfer Functions

$$\frac{P}{CR} = \frac{1}{\Delta} \sum_{k=1}^{P} M_k \Delta_k = \frac{1}{\Delta} \left(\frac{M_1 \Delta_1}{\Delta_1} + \frac{M_2 \Delta_p}{\Delta_p} \right)$$

Where:

•
$$\Delta = \Delta - (Z | loops) + Z (von-touching | loops)$$

- $Z (products of triplets of non-touching | loops)$

• $M_k = Gain of the k^{th} path between R and C$

• $\Delta_k = Value of \Delta$ when the nodes in the path M_k are removed from the graph.

Example 1: DC motor:

Synal flow graph:

$$+1 = R + LS = KZ = JS^2 + BS + 1$$

$$-5 Km$$

1 loop: L3 = - kz km s (R+Ls)(JS+B)s

$$\frac{M_{\Delta}}{M_{\Delta}} = \frac{Kz}{(R+Ls)(3s+8)s}$$

$$\Delta_{\perp} = \Delta$$

 $T_{0E_{\alpha}} = \frac{1}{\Delta} \cdot M_{1} \Delta_{\Delta} =$

$$= \frac{(R+LS)(JS+B)S}{\Delta + \frac{k_E k_M S}{(R+LS)(JS+B)S}}$$

only & path from Ea to 10:

 $= \frac{kz}{(R+Ls)(Js+B)s + k_m k_z s} #$