

Dynamics and Control of Inverted Pendulum Systems

1 Standard Inverted Pendulum on a Cart

1.1 System Description

- Cart mass: M
- Pendulum mass: m
- Pendulum length to CoM: l
- Gravity: g
- Horizontal force input: F
- x : horizontal position of the cart
- θ : pendulum angle from vertical (clockwise positive)

1.2 Nonlinear Equations of Motion

$$(M + m)\ddot{x} + ml\ddot{\theta} \cos \theta - ml\dot{\theta}^2 \sin \theta = F \quad (1)$$

$$ml\ddot{x} \cos \theta + ml^2\ddot{\theta} - mgl \sin \theta = 0 \quad (2)$$

1.3 Nonlinear State-Space Model

Define state vector and input:

$$\mathbf{x} = \begin{bmatrix} x \\ \dot{x} \\ \theta \\ \dot{\theta} \end{bmatrix}, \quad u = F$$

Express the system as:

$$\dot{\mathbf{x}} = \begin{bmatrix} \dot{x} \\ \ddot{x} \\ \dot{\theta} \\ \ddot{\theta} \end{bmatrix} = f(\mathbf{x}, u)$$

To solve for \ddot{x} and $\ddot{\theta}$, rewrite (1), (2) in matrix form:

$$\underbrace{\begin{bmatrix} M + m & ml \cos \theta \\ ml \cos \theta & ml^2 \end{bmatrix}}_{D(\theta)} \begin{bmatrix} \ddot{x} \\ \ddot{\theta} \end{bmatrix} = \begin{bmatrix} u + ml\dot{\theta}^2 \sin \theta \\ mgl \sin \theta \end{bmatrix}$$

Then:

$$\begin{bmatrix} \ddot{x} \\ \ddot{\theta} \end{bmatrix} = D(\theta)^{-1} \begin{bmatrix} u + ml\dot{\theta}^2 \sin \theta \\ mgl \sin \theta \end{bmatrix}$$

Hence the nonlinear state-space model is:

$$\dot{\mathbf{x}} = \begin{bmatrix} \dot{x} \\ \ddot{x}(\mathbf{x}, u) \\ \dot{\theta} \\ \ddot{\theta}(\mathbf{x}, u) \end{bmatrix}$$

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2 Wheeled Inverted Pendulum (Self-Balancing Robot)

2.1 System Description

- Wheel mass: M , radius R , inertia I_w
- Pendulum mass: m , pendulum CoM length: l , inertia I_p
- Gravity: g
- Torque input: τ (applied to wheel)
- x : horizontal position of wheel
- θ : pendulum angle from vertical (clockwise positive)

2.2 Nonlinear Equations of Motion

$$\left(M + m + \frac{I_w}{R^2} \right) \ddot{x} + ml\ddot{\theta} \cos \theta - ml\dot{\theta}^2 \sin \theta = \frac{\tau}{R} \quad (3)$$

$$ml\ddot{x} \cos \theta + (I_p + ml^2) \ddot{\theta} - mgl \sin \theta = 0 \quad (4)$$

2.3 Nonlinear State-Space Model

Define

$$\mathbf{x} = \begin{bmatrix} x \\ \dot{x} \\ \theta \\ \dot{\theta} \end{bmatrix}, \quad u = \tau$$

Rewrite (3), (4) as:

$$D(\theta) = \begin{bmatrix} M + m + \frac{I_w}{R^2} & ml \cos \theta \\ ml \cos \theta & I_p + ml^2 \end{bmatrix}, \quad \mathbf{b} = \begin{bmatrix} \frac{u}{R} + ml\dot{\theta}^2 \sin \theta \\ mgl \sin \theta \end{bmatrix}$$

Then

$$\begin{bmatrix} \ddot{x} \\ \ddot{\theta} \end{bmatrix} = D(\theta)^{-1} \mathbf{b}$$

The nonlinear state-space form is

$$\dot{\mathbf{x}} = \begin{bmatrix} \dot{x} \\ \ddot{x}(\mathbf{x}, u) \\ \dot{\theta} \\ \ddot{\theta}(\mathbf{x}, u) \end{bmatrix}$$

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3 Linearized Models and State-Space Form

3.1 Linearized Dynamics for Cart Pendulum

Linearize around $\theta = 0$, $\dot{\theta} = 0$:

$$\sin \theta \approx \theta, \quad \cos \theta \approx 1$$

Linearized equations:

$$\begin{bmatrix} M + m & ml \\ ml & ml^2 \end{bmatrix} \begin{bmatrix} \ddot{x} \\ \ddot{\theta} \end{bmatrix} = \begin{bmatrix} u \\ mgl\theta \end{bmatrix}$$

Explicitly:

$$\begin{aligned} (M + m)\ddot{x} + ml\ddot{\theta} &= u \\ ml\ddot{x} + ml^2\ddot{\theta} - mgl\theta &= 0 \end{aligned}$$

3.2 Linearized State-Space Form for Cart Pendulum

Define state:

$$\mathbf{x} = \begin{bmatrix} x \\ \dot{x} \\ \theta \\ \dot{\theta} \end{bmatrix}, \quad u = F$$

The linear state-space model is:

$$\dot{\mathbf{x}} = A\mathbf{x} + Bu$$

Where

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & -\frac{mgl}{M} & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & \frac{(M+m)g}{Ml} & 0 \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ \frac{1}{M} \\ 0 \\ -\frac{1}{Ml} \end{bmatrix}$$

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3.3 Linearized Dynamics for Wheeled Inverted Pendulum

Linearize around $\theta = 0, \dot{\theta} = 0$:

$$\sin \theta \approx \theta, \quad \cos \theta \approx 1$$

$$D = \begin{bmatrix} \alpha & ml \\ ml & \beta \end{bmatrix}$$

with

$$\alpha = M + m + \frac{I_w}{R^2}, \quad \beta = I_p + ml^2$$

Linearized equations:

$$\begin{bmatrix} \alpha & ml \\ ml & \beta \end{bmatrix} \begin{bmatrix} \ddot{x} \\ \ddot{\theta} \end{bmatrix} = \begin{bmatrix} \frac{u}{R} \\ mgl\theta \end{bmatrix}$$

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3.4 Linearized State-Space Form for Wheeled Pendulum

State vector:

$$\mathbf{x} = \begin{bmatrix} x \\ \dot{x} \\ \theta \\ \dot{\theta} \end{bmatrix}, \quad u = \tau$$

State-space model:

$$\dot{\mathbf{x}} = A\mathbf{x} + Bu$$

with

$$A = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & -\frac{m^2 gl^2}{\Delta R} & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & \frac{mgl\alpha}{\Delta} & 0 \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ \frac{\beta}{\Delta R} \\ 0 \\ -\frac{ml}{\Delta} \end{bmatrix}$$

where the determinant

$$\Delta = \alpha\beta - (ml)^2$$

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4 Controller Design

4.1 PID Control

For both systems, a PID controller can be used to regulate the pendulum angle:

$$u(t) = K_p e(t) + K_i \int e(t) dt + K_d \frac{de(t)}{dt}, \quad e(t) = \theta_{\text{ref}} - \theta(t)$$

Tuning methods (Ziegler-Nichols, manual) apply.

4.2 LQR Control

Given linearized models:

$$\dot{\mathbf{x}} = A\mathbf{x} + Bu, \quad u = -K\mathbf{x}$$

The gain matrix K minimizes

$$J = \int_0^\infty (\mathbf{x}^T Q \mathbf{x} + u^T R u) dt$$

where Q and R are weighting matrices chosen by the designer. K is computed by solving the algebraic Riccati equation.

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