## Course 5

The theory of electric circuits - Second part of the course Cap. 1. Basic notions

The electric cincuits can be passive ( no sources) plangest transversal dimention of

Penetration depth (conductivity)
(adancime de patrundere) magnetic
permeability Lumped circuit conditions (circuite ou elemente concentrate)

Rength of circuit l ZZ N = C- light speed

xavelength & frequency

(g=50 Hz => N = 6000 km => l ≤ 60 km)

1.1. Ideal circuit elements (Elmente de circuit ideale)

1) Ideal resistor

Ub Paraday law:

ep = dosp

Pezistenta ideala = s R = 0, L=0, C=0

inductivity capacity (inductivitate) (capacitate)

, \$ Sp = L.i = 0 = s ep = 0

Ohm's law: ep = ug - ub

e et = R.i (tensiumea de -a lungul finului conducta)

tensilunea la borne

2) Ideal inductor

 $P = \mu_b, i = R. i^2 > 0$  (Youle law)

Tpower (putere) L
Simbol bobina (cail): \_m\_ or \_\_\_\_

 $e_{\Gamma} = -\frac{d\phi_{S\Gamma}}{dt}$ ,  $\phi_{S\Gamma} = L \cdot i = , e_{\Gamma} = -L \cdot \frac{di}{dt}$ 

 $e_{\Gamma} = u_{g} - u_{b}$   $\int_{0}^{\infty} = 0 - L \cdot \frac{di}{dt} = -u_{b} = 0 \quad \text{if } = L \cdot \frac{di}{dt}$ Mg = R. i = 0

P = d Wmg = d (Lil 20

Wing - energie magne tica (aush)

L = 0, R = 0, C= 0

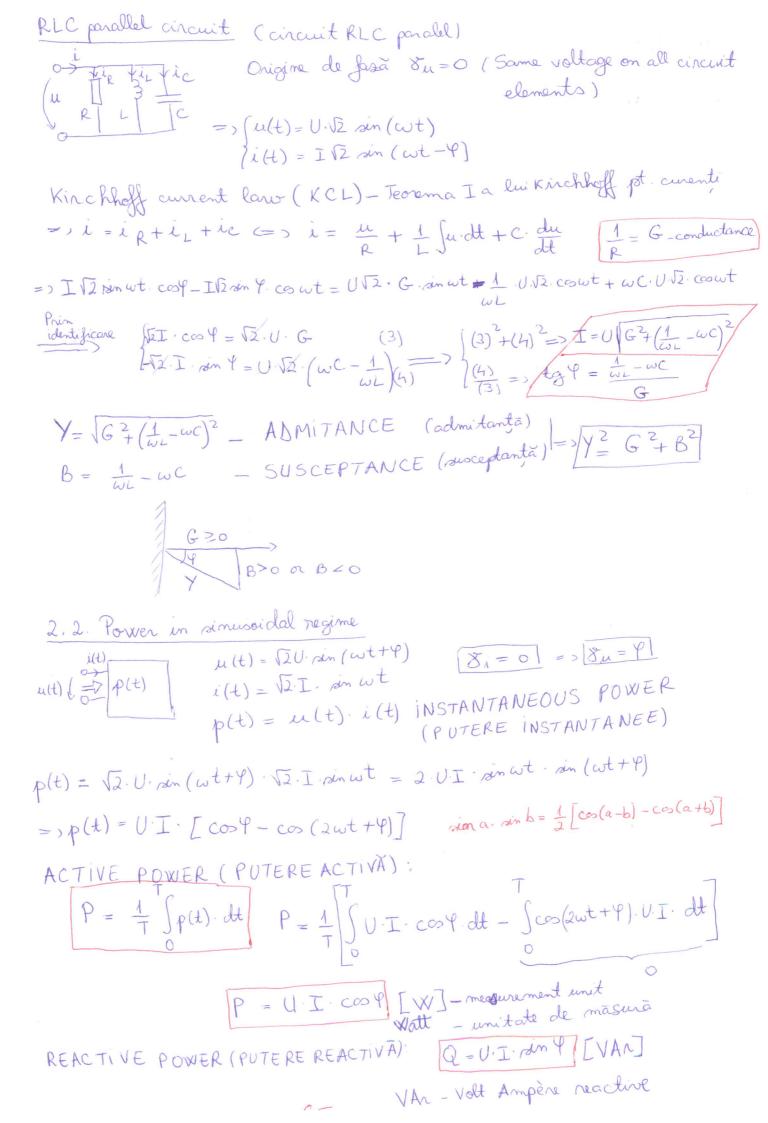
3. Ideal capacitor er = - desp ; der = L.1 = 0 = s er = 0 Mb ( The C) Mc et= me+mc-mp => mc=mp  $i = \frac{2}{4} = \lambda u_c = \frac{2}{c}$   $i = \frac{2}{dt} = \lambda 2(t) = \int i(t) dt$ 0 R.i. C +0, R=0, L=0  $= \lambda u_b = \frac{2}{c} = \frac{1}{c} \int i(t) \cdot dt$ =  $p = \frac{dWd}{dt} = \frac{d}{dt} \left(\frac{c \cdot u^2}{2}\right) \ge 0$ 4. Ideal voltage source (pursa de tensiune) E-constant, indiffrent de ideal characteristic

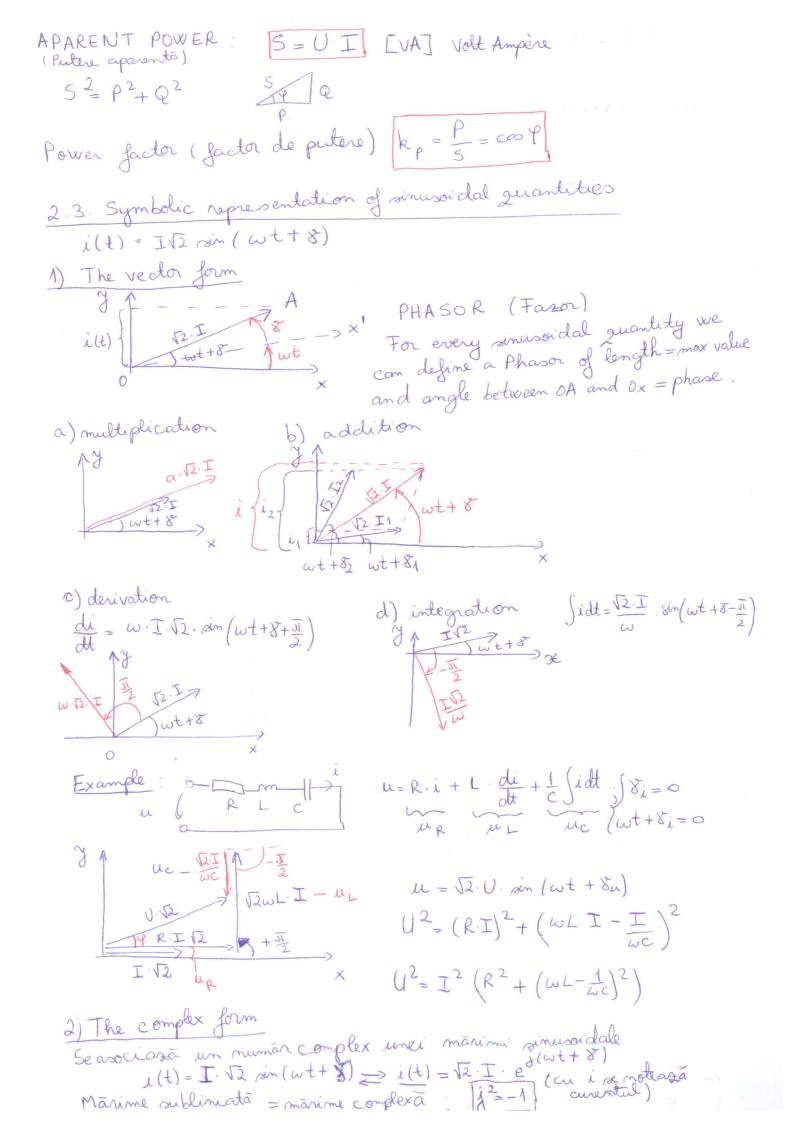
Treal characteristic valoarea curentului Se cunoaste: valoarea tensiuni furnisate de surse Necunoscut: curentul ce circulo prin sursa (depinde de configuratia circuitului din care face parte sursa) 5. Ideal current source (sursa de curent) I - constant, indifférent de valoares lens real | ideal Simboluri : We know the value of the current given by the source Unknown: the voltage drop on the current source (depends on the circuit's configuration) lapl. Sinusoidal quantities Definitions. Characteristic values. Let i(t) be a periodic function: i(t) = i(t+T) T - perioada semnalului;  $g = \frac{1}{T} - fuguency (fucuență)$  $\omega = 2 \text{ Ji } f = \frac{2 \text{ Ji}}{T} = 2 \text{ Ji} \quad \omega = 2 \text{ Ji} \quad \omega = \text{angular frequency (pulsative)}$ 

R.M.S. (Root mean square) value  $= \frac{1}{t_2 - t_1} \int_{t_1} i(t) \cdot dt = \frac{1}{t_1} \int_{t_1} i(t) \cdot dt - \frac{1}{t_1} \int_{t_1} i(t) \cdot d$ Now, let i(t) be a sinusoidal function:  $i(t) - I_m \cdot \sin(\omega t + \delta)$ = the phase of the quantity (current)
epoch angle or initial phase (fasa initials a semnalului) For a sinusoidal quantity:  $= \frac{1}{T} \int_{A} i(t) \cdot dt = \frac{1}{T} \int_{A} I_{m} \sin(\omega t + \delta) \cdot dt =$  $= -\frac{Im}{\omega T} \cdot \cos(\omega t + 8) \Big|_{t_1}^{t_1 + T} = \frac{Im}{\omega t} \Big[ -\cos(\omega t_1 + \omega T + 8) + \frac{1}{\omega t} \Big]_{t_1}^{t_1 + T} = \frac{Im}{\omega t} \Big[ -\cos(\omega t_1 + \omega T + 8) + \frac{1}{\omega t} \Big]_{t_1 + T}^{t_1 + T} = \frac{Im}{\omega t} \Big[ -\cos(\omega t_1 + \omega T + 8) + \frac{1}{\omega t} \Big]_{t_1 + T}^{t_1 + T} = \frac{Im}{\omega t} \Big[ -\cos(\omega t_1 + \omega T + 8) + \frac{1}{\omega t} \Big]_{t_1 + T}^{t_1 + T}$  $I' = \frac{1}{T} \int I_m^2 \sin^2(\omega t + 8) dt = \frac{1}{T} \cdot I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int I_m^2 \cdot \int [1 - \cos(2\omega t + 2\delta)] dt = \frac{1}{T} \cdot \int [1 - \cos(2\omega t +$  $= \frac{\prod_{n=1}^{\infty} \left[ \int_{0}^{\infty} dt + \int_{0}^{\infty} \cos(2\omega t + 2\delta) dt \right]}{2} = \frac{\prod_{n=1}^{\infty}}{2}$ val med pe o perioado = 0 =>  $i(t) = \sqrt{2.1} \cdot \sin(\omega t + 8)$  This form will be used from now on!

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3/12= 81-82 phase shift (defazaj inte i, sii)
                       igt)
                          re(t)
                                         is - lead (defasat mainte)
                                        iz - lags (defasat in wrma)
                                      in(t) = I1. 52. sin (wt + 51)
                                     iz(t) = Iz. V2. sin (wt+ 82)
                                    812 - the two currents are in phase (sunt in faxa)
                                  812 = ± JT - the two are in oposite phase
   Sin (wt+81)=0
                                                   ( sunt in antifază)
  =>wt+81=0
                                   8/12 = I - quadrature ( anadraturo)
    => 81 = -wt
    => for t 20=> 5/>0!
2.1 Mathematical operations with sinusoidal quantities
a) Multiplication by a scalar (insultire cu un scalar)
    a. ilt) = a. I. 12. sin (wt+8)
     i_1(t) = I_1 \sqrt{2} \sin (\omega t + \delta_1)

i_2(t) = I_2 \sqrt{2} \sin (\omega t + \delta_2) => i(t) = i_1 + i_2 = I. (2 \sin (\omega t + \delta))
b) Addition (adunare)
  with [ I = [I12+ I2+2 I1 I2cos (81-82)
              tg 8= I 1 sin 81 + I2 sin 82
I1. cos 81 + I2. cos 82
  c) berivation (derivare)
        di = d [I 12 sin(wt+8)] = wI. V2. eas (wt+8) = \( \tau \times \) with \( \times \)
         Los leads the initial value by I (accesto mainime e defasata ou II inaintea continue initial)
    d) Integration (integrare)
        Si(t) dt = \int I \sqrt{2} \sin(\omega t + \delta) dt = \frac{I \cdot \sqrt{2}}{\omega} \cdot \cot(\omega t + \delta) = \frac{I \sqrt{2}}{\omega} \cdot \sin(\omega t + \delta - \frac{JI}{2})
                                                                   -\cos d = \sin \left( \frac{1}{2} \right)
          the amplitude is a times smaller
           Lo lags behind the initial value by J
             ( defasat în urma cu JI)
    RLC revies circuit (circuit RLC revie)
                                  u(t) = uR + UL + UC - Ohm's Row
                                    u(t) = U\sqrt{2}\sin(\omega t + \delta_u)
                                     : (+) = I 52 sin (wt + 8i)
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$$i(t) = \sqrt{2} \cdot \text{I} \cdot \cos(\omega t + \delta) + j \cdot \sqrt{1} \cdot \text{I} \cdot \sin(\omega t + \delta) \quad \text{Euller} : e^{j \cdot \omega} = \cos(\omega t) \cdot \sin(\omega t)$$

$$= \text{i}(t) = \sqrt{1} \text{ in } \left\{ i(t) \right\} \quad \text{imaginary part}$$

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$$i(t) = \omega \cdot \text{i}(t) \Rightarrow i_1(t) + i_2(t)$$

$$i_1(t) + i_3(t) \Rightarrow i_1(t) \Rightarrow i_2(t) \Rightarrow i_2(t) \Rightarrow i_3(t) \Rightarrow i_3(t$$

3)  $u(t) = 3.\sqrt{2}$ ,  $aim(\omega t + J)$   $\longrightarrow U = 3.e^{3J} = -3$