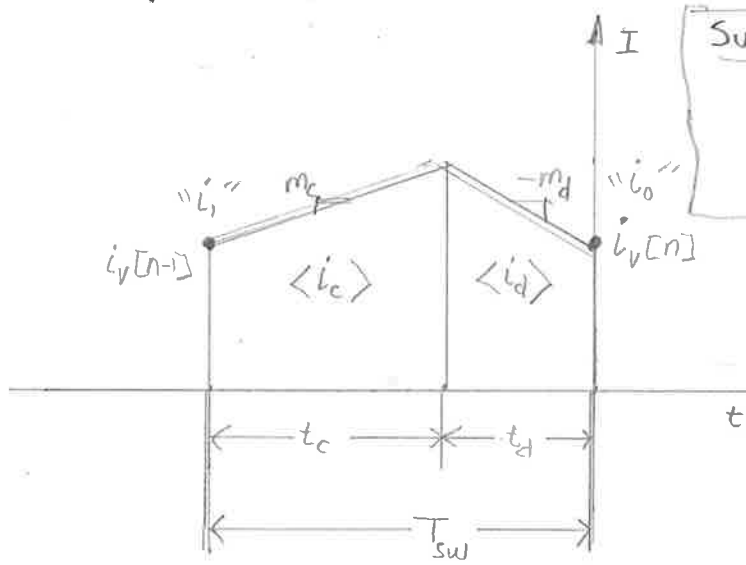


IMPULSE RESPONSE OF SWITCH-MODE INDUCTOR AS FIR FILTER

①



SUBSTITUTIONS,

$$a = \frac{1}{m_c + m_d}$$

$$b = T_{sw}(1 - a m_d)$$

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05 JULY, 2019

① VALLEY CURRENT " $i_v[n]$ " IS CONSIDERED AS AN IMPULSE TRAIN

② (THIS DERIVATION) AVERAGE DISCHARGE CURRENT IS EXPRESSED AS AN FIR FILTER CONVOLVED WITH VALLEY CURRENTS, e.g.,

$$\langle i_d \rangle[n] = A_0 i_v[n] + A_1 i_v[n-1]$$

STEP 1: FIND AVERAGE DISCHARGE CURRENT $\langle i_d \rangle$

AS A FUNCTION OF VALLEY CURRENTS, " i_v " AND " i_v "

$$t_d = T_{sw} - t_c$$

$$i_0 = i_1 + m_c t_c - m_d t_d$$

$$= i_1 + m_c t_c - m_d T_{sw} + m_d t_c$$

$$i_0 - i_1 + m_d T_{sw} = t_c (m_d + m_c)$$

$$t_c = \underbrace{\left[\frac{1}{(m_c + m_d)} \right]}_a [i_0 - i_1 + m_d T_{sw}] = a i_0 - a i_1 + a m_d T_{sw}$$

$$t_d = T_{sw} - t_c = T_{sw} + a i_1 - a i_0 - a m_d T_{sw}$$

$$\underbrace{T_{sw}(1 - a m_d)}_b + a i_1 - a i_0 = a i_1 - a i_0 + b$$

$$\langle i_d \rangle = \left(\frac{1}{T} \right) \left(\frac{1}{2} \right) [i_0 + i_0 + m_d t_d] t_d = \frac{1}{2T} (2 t_d i_0 + m_d t_d^2)$$

$$= \frac{1}{T} \left(t_d i_0 + \frac{1}{2} m_d t_d^2 \right)$$

$$t_d = a i_1 - a i_0 + b$$

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$$t_d^2 = (a i_1 - a i_0 + b)(a i_1 - a i_0 + b)$$

$$= \underbrace{a^2 i_1^2}_{\text{mmmm}} - \underbrace{a^2 i_0 i_1}_{\text{mmmm}} + \underbrace{a^2 i_0^2}_{\text{mmmm}} - \underbrace{a b i_0}_{\text{mmmm}} + \underbrace{a b i_1}_{\text{mmmm}} - \underbrace{a b i_0}_{\text{mmmm}} + b^2$$

$$= a^2 i_0^2 + a^2 i_1^2 - 2 a^2 i_0 i_1 - 2 a b i_0 + 2 a b i_1 + b^2$$

$$t_d i_0 = i_0 (a i_1 - a i_0 + b) = a i_0 i_1 - a i_0^2 + b i_0$$

$$\begin{aligned} \frac{1}{2} m_d t_d^2 &= \frac{1}{2} m_d a^2 i_0^2 + \frac{1}{2} m_d a^2 i_1^2 - m_d a^2 i_0 i_1 - m_d a b i_0 + m_d a b i_1 + \frac{1}{2} m_d b^2 \\ &+ t_d i_0 - a i_0^2 + a i_0 i_1 + b i_0 \end{aligned}$$

$$\frac{(t_d i_0 + \frac{1}{2} m_d t_d^2)}{T} = \underbrace{\left(\frac{\frac{1}{2} m_d a^2 - a}{T} \right) i_0^2}_c + \underbrace{\frac{\frac{1}{2} m_d a^2}{T} i_1^2}_d + \underbrace{\left(a - \frac{m_d a^2}{T} \right) i_0 i_1}_e + \underbrace{\left(\frac{b - m_d a b}{T} \right) i_0}_f + \underbrace{\frac{m_d a b}{T} i_1}_g + \underbrace{\frac{\frac{1}{2} m_d b^2}{T}}_h$$

$$\langle i_d \rangle = c i_0^2 + d i_1^2 + e i_0 i_1 + f i_0 + g i_1 + h$$

STEP 2: PERTURB $\langle i_d \rangle$

FIND $\Delta \langle i_d \rangle$ AS A FUNCTION OF Δi_0 and Δi_1 (SMALL-SIGNAL GAIN)

$$\langle i_d \rangle = \langle I_d \rangle + \langle \Delta_d \rangle$$

$$i_0 = I_0 + \Delta_0$$

$$i_1 = I_1 + \Delta_1$$

$$I_0 = I_1 = i_0 = i_1 \equiv i_{\infty}, \text{ steady-state}$$

FOR SMALL SIGNALS,

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$$\langle I_d + \Delta_d \rangle \approx \left. \frac{\partial}{\partial i_0} c i_0^2 \right|_{i_0 = i_{00}} \times \Delta_0 = \Delta_0 2c i_0 / i_{00} = 2c i_{00} \Delta_0$$

$$+ \left. \frac{\partial}{\partial i_1} d i_1^2 \right|_{i_1 = i_{10}} \times \Delta_1 = \Delta_1 2d i_1 / i_{10} = 2d i_{10} \Delta_1$$

$$+ \left. \frac{\partial}{\partial i_0} e i_0 i_1 \right|_{i_1 = i_{10}} \times \Delta_0 + \left. \frac{\partial}{\partial i_1} e i_0 i_1 \right|_{i_0 = i_{00}} \times \Delta_1 = e i_1 / i_{10} \Delta_0 + e i_0 / i_{00} \Delta_1 = e i_{10} \Delta_0 + e i_{00} \Delta_1$$

$$+ \left. \frac{\partial}{\partial i_0} f i_0 \right|_{i_0 = i_{00}} \Delta_0 = f \Delta_0$$

$$+ \left. \frac{\partial}{\partial i_1} g i_1 \right|_{i_1 = i_{10}} \Delta_1 = g \Delta_1$$

+ h CONST, NOT PART OF AC GAIN

$$\begin{aligned} \langle i_d \rangle &= 2c i_{00} \Delta_0 + 2d i_{10} \Delta_1 + e i_{10} \Delta_0 + e i_{00} \Delta_1 + f \Delta_0 + g \Delta_1 + h \\ &= \underbrace{(2c i_{00} + e i_{10} + f) \Delta_0}_{A_0} + \underbrace{(2d i_{10} + e i_{00} + g) \Delta_1}_{A_1} + h \end{aligned}$$

RE-INTERPRET GEOMETRIC MODEL IN TERMS OF VALLEY-CURRENT
IMPULSE TRAIN: $\Delta_0 = i_v[n]$, $\Delta_1 = i_v[n-1]$, $\langle i_d \rangle = \langle i_d \rangle[n]$

MAKING SUBSTITUTIONS,

$$\langle i_d \rangle[n] = A_0 i_v[n] + A_1 i_v[n-1]$$

STEP 3: FIND I_{∞} , OPERATING POINT GAIN

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$$I_{\infty} = c i_{\infty}^2 + d i_{\infty}^2 + e i_{\infty}^2 + f i_{\infty} + g i_{\infty} + h, \quad i_{\infty} = i_0 = i_1 \text{ @ steady state}$$

$$I_{\infty} = (c+d+e) i_{\infty}^2 + (f+g) i_{\infty} + h$$

CAN BE RE-APPLIED AFTER
[STEP 5] NORMALIZATION

STEP 4: Z-TRANSFORM

$$\langle i_d \rangle[n] = A_0 i_v[n] + A_1 i_v[n-1]$$

$$\langle I_d \rangle(z) = A_0 I_v(z) + A_1 I_v z^{-1}$$

$$H(z) = \frac{\langle I_d \rangle(z)}{I_v(z)} = A_0 + A_1 z^{-1}$$

STEP 5: NORMALIZE DC GAIN FOR EASY COMPARISON WITH CONVENTIONAL RHPZ

$$\text{Observation: } z^{-1} = e^{-j\omega T} = 1 \text{ @ } \omega = 0$$

GAIN @ $\omega = 0$

$$H(z) \Big|_{\omega=0} = A_0 + A_1 e^0$$

A NORMALIZED TRANSFER FUNCTION IS AS FOLLOWS

$$H_N(z) = \frac{1}{A_0 + A_1} (A_0 + A_1 z^{-1}) = \frac{A_1}{A_0 + A_1} \left(\frac{A_0}{A_1} + z^{-1} \right)$$

IN APPENDIX A IT IS SHOWN HOW THIS CAN BE
USED TO OBTAIN THE 3dB CUT-OFF FREQUENCY

TO SEE FREQUENCY COMPONENT SEPARATELY FROM
DC GAIN, $H(z)$ CAN BE EXPRESSED AS FOLLOWING

$$H(z) = I_{\infty} H_N(z)$$

APPENDIX A: FIND 3-dB CUT-OFF FREQUENCY FOR RHPZ

$$H_N(z) = \frac{A_1}{A_0 + A_1} \left(\frac{A_0}{A_1} + z^{-1} \right)$$

$$= \frac{1}{\frac{A_0}{A_1} + 1} \left(\frac{A_0}{A_1} + z^{-1} \right)$$

$$= \frac{1}{\frac{1}{\omega_z} + 1} \left(\frac{1}{\omega_z} + z^{-1} \right),$$

z-domain root does not coincide w/ 3dB cut-off
 $\omega_z = \frac{A_1}{A_0}$

FIND ω in $z^{-1} = e^{-j\omega T}$ such that,
 $|H_N(z)| = z^{\pm \frac{1}{2}} = |H_C|$, (either $\sqrt{2}$ or $\frac{1}{\sqrt{2}}$)

$$H_C\left(\frac{1}{\omega_z} + 1\right) = \left| \frac{1}{\omega_z} + \cos(\omega_0) - j \sin(\omega_0) \right|$$

$$\sqrt{\left(\frac{1}{\omega_z} + \cos(\omega_0)\right)^2 + \sin^2(\omega_0)}$$

$$\left[H_C\left(\frac{1}{\omega_z} + 1\right) \right]^2 = \left(\frac{1}{\omega_z} + \cos(\omega_0)\right)^2 + \sin^2(\omega_0)$$

$$\left(\frac{1}{\omega_z}\right)^2 + \frac{2 \cos(\omega_0)}{\omega_z} + \underbrace{\cos^2(\omega_0) + \sin^2(\omega_0)}_1$$

$$\cos^{-1} \left[\frac{\omega_z \left(\left[H_C\left(\frac{1}{\omega_z} + 1\right) \right]^2 - 1 - \frac{1}{\omega_z^2} \right)}{2} \right] = \omega_0 = \omega_c T$$

$$\omega_c = \frac{1}{T} \cos^{-1} \left[\frac{\omega_z}{2} \left(\left(H_C\left(\frac{1}{\omega_z} + 1\right) \right)^2 - 1 - \frac{1}{\omega_z^2} \right) \right]$$