

Monday 2021-10-18

①

Pset #3 Due Wednesday

Last time

• $\tilde{F}(\omega) = \int_{-\infty}^{\infty} f(t) e^{-2\pi i \omega t} dt$ forward $t \rightarrow \omega$

$f(t) = \int_{-\infty}^{\infty} \tilde{F}(\omega) e^{2\pi i \omega t} d\omega$ inverse $\omega \rightarrow t$

• Euler's Formula

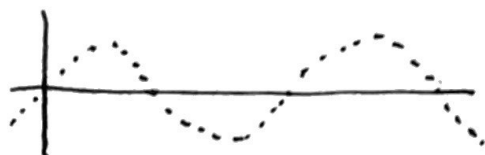
$$e^{2\pi i \theta} = \cos(2\pi \theta) + i \sin(2\pi \theta)$$

• FT \rightarrow function of freq.; complex number

▣ Amplitude (power @ given freq.)

▣ phase (L/R shift)

• REAL DATA is Discrete



\nwarrow a vector of numbers.

$$\vec{y} = f(\vec{x})$$

▣ CAN FIT W/ LEAST SQUARES

$$C = (G' * G) \setminus (f * G)'$$

f is data
 G is matrix
of models g

$$G = [g_1 \ g_2 \ \dots \ g_m]$$

$$g = e^{i\omega_0 m n \Delta t}$$

$$) \ \omega_0 = \frac{2\pi}{N \Delta t}$$

m functions
 n data points

$$m = 1, 2, \dots$$

$$n = [0 : N-1]$$

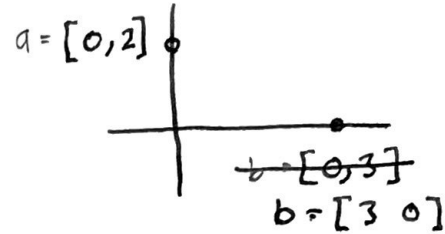
o ORTHOGONALITY

means $\vec{a} \perp \vec{b}$

we can use Dot Product (inner product) = 0 to show orthogonality.

ie. $\vec{a} \cdot \vec{b}^T = 0$

$$\begin{bmatrix} 0 & 2 \end{bmatrix} \begin{bmatrix} 3 \\ 0 \end{bmatrix} = (0 \cdot 3) + (2 \cdot 0) = 0$$



□ Orthonormal special case

$$|a| = |b| = 1$$

□ $\sin\left(\frac{2\pi n}{N}\right)$ & $\cos\left(\frac{2\pi n}{N}\right)$ are Orthogonal.

means all non diagonals elements are 0

we worked through algebra to get to

$$\tilde{C}_m = \frac{1}{N} \sum_{n=1}^N f(t) e^{(-i\omega_m n)} \leftarrow \text{DFT}$$

o Today.

DFT Properties

$$\textcircled{1} \tilde{F}(ax + by) = a\tilde{F}(x) + b\tilde{F}(y)$$

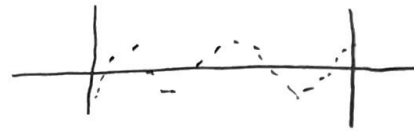
LINEARITY!

$$\textcircled{2} \tilde{X}_0 = \sum_{n=0}^{N-1} x_n \quad \dots \quad C_0 = \frac{1}{N} \sum_{n=0}^{N-1} f(t) e^{0} = 1$$

- 0 frequency (DC) measures the mean
 \wedge fixed component.

③ Linear phase shift
if X_{n-k} is substituted for X_n , $\frac{2\pi k m}{N}$
radians are subtracted from X_m

④ Values beyond the edge of vector
are assumed to be 0



- Corollary. Edge effects!
windowing is crucial.

⑤ Differentiation

$$\frac{d\tilde{F}(x)}{dx} = \frac{d}{dx} \tilde{F}(x) = \frac{d}{dx} \int_{-\infty}^{\infty} f(x) e^{-i\omega t} d\omega = \int_{-\infty}^{\infty} f(x) d\omega \int_{-\infty}^{\infty} \frac{d}{dx} e^{-i\omega t} d\omega$$

$$\frac{d\tilde{F}(x)}{dx} = \int_{-\infty}^{\infty} f(x) d\omega \cdot -i\omega e^{-i\omega t} = -i\omega \underbrace{\int_{-\infty}^{\infty} f(x) e^{-i\omega t} d\omega}_{\tilde{F}(x)}$$

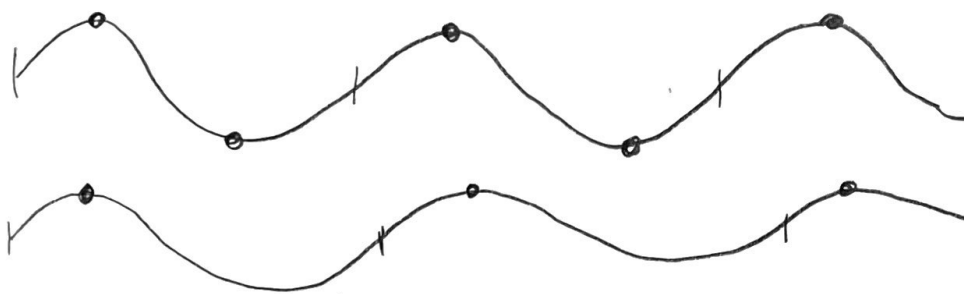
$$\boxed{\frac{d\tilde{F}(x)}{dx} = -i\omega \tilde{F}(x)}$$

⑥ Parseval's Theorem

$$\sum_{n=0}^{N-1} X_n^2 = \frac{1}{N} \sum_{n=0}^{N-1} \tilde{X}_m^2$$

power is preserved.

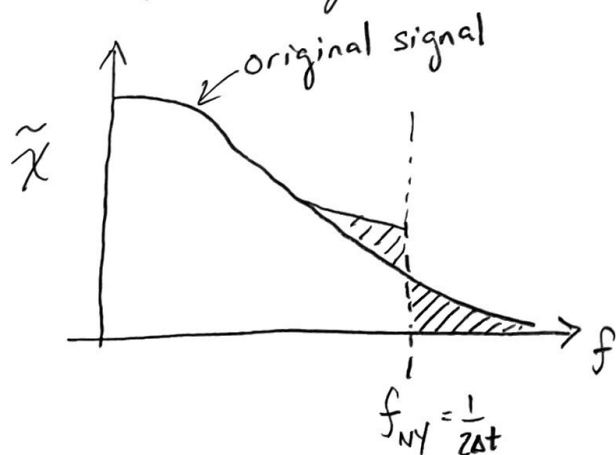
if a continuous signal is sampled at greater than twice its highest freq. component, then it is possible to recover the signal from the samples



OK

NOT
OK

If a signal is under sampled aliasing will occur.
maps higher f to lower f



$$\text{Nyquist freq.} = \frac{1}{2\Delta t}$$

↑
samp. interval

Therefore the range of independent freq

$$\vec{f} = \begin{bmatrix} 0 & \frac{1}{N\Delta t} & - & - & - & - & \frac{1}{2\Delta t} \end{bmatrix}$$

\uparrow DC \uparrow duration of signal $\uparrow f_{NY}$

$$\uparrow f_{NY} = \frac{N}{2(Nat)}$$

" $\frac{1}{2}$ of freq)

$$\vec{f} = \frac{[0; \frac{N}{2}]}{Nat}$$

Implication: all actual data is
band limited at discrete
frequencies

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Other Properties of DFT

① DFT has N pos. freq. comps. ($m=N$)

$$\begin{aligned}\tilde{X}_{m+N} &= \sum_{n=0}^{N-1} x_n e^{-2\pi i(m+N)n/N} \\ &= \sum_{n=0}^{N-1} x_n e^{-2\pi i m n/N} e^{-2\pi i N n/N} = \sum_{n=0}^{N-1} x_n e^{-2\pi i m n/N} \underbrace{e^{-2\pi i n}}_{\substack{\uparrow \\ 1 \text{ for any integer } n}}\end{aligned}$$

So freq. beyond N are redundant.

$$\begin{aligned}\tilde{X}_{m-N} &= \sum_{n=0}^{N-1} x_n e^{-2\pi i(m-N)n/N} \\ &= \sum_{n=0}^{N-1} x_n e^{-(-2\pi i m n/N)} \underbrace{e^{-2\pi i n N/N}}_{K=1}\end{aligned}$$

conjugate of (X_m)

$$\overline{X_m}$$

$$\text{so } X_{m-N} = \overline{X_m}$$

therefore we take $M=N$