

Problem	1	2	3	4	Total
Maximum score	25	25	25	25	100
Course learning outcome	2	1	2	2	1, 2

Problem 1)

Consider the circuit shown in Figure P1. Assume that there is no energy stored in the circuit at $t = 0$. Write your answers in the table shown below.

- Find the transfer function $H(s) = V_L(s) / V_G(s)$. (15 points)
- Find $v_L(t)$ if $v_G(t) = 2u(t)$, where $u(t)$ is the unit-step signal. (10 points)

	Answer
$H(s)$	
$v_L(t)$	

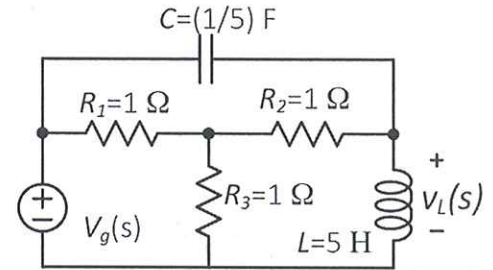


Figure P1

Problem 2)

Assume that the op amp shown in Figure P2 is ideal and signal voltages are provided in terms of differential mode and common mode voltages, v_{dm} and v_{cm} , respectively.

$$v_A = v_{cm} - \frac{1}{2}v_{dm}$$

$$v_B = v_{cm} + \frac{1}{2}v_{dm}$$

$$R_1 = 1 \text{ k}\Omega, R_2 = 10 \text{ k}\Omega$$

$$R_3 = 1 \text{ k}\Omega, R_4 = 1 \text{ k}\Omega$$

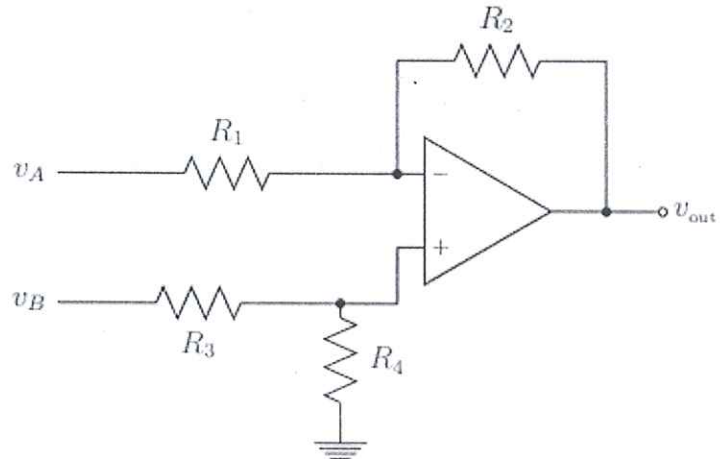


Figure P2

- Derive an expression for the output voltage in terms of v_{dm} and v_{cm} . (15 points)
- Calculate the common-mode gain. (Specify your units in either V/V or dB). (3 points)
- Calculate the differential-mode gain. (Specify your units in either V/V or dB). (3 points)
- Calculate the CMRR. (4 points)

Problem 3)

Consider the circuit shown in Figure P3.

- Derive the transfer function, $H(s) = V_o(s)/V_s(s)$, of the circuit shown in Figure P3. **(10 points)**
- Determine the type of the filter shown in Figure P3. **(5 points)**
- Calculate the cutoff frequency(ies) of the filter shown in Figure P3. **(5 points)**
- Find the output response of the filter shown in Figure P3 when $v_s(t) = \cos(2t + 30^\circ)$. **(5 points)**

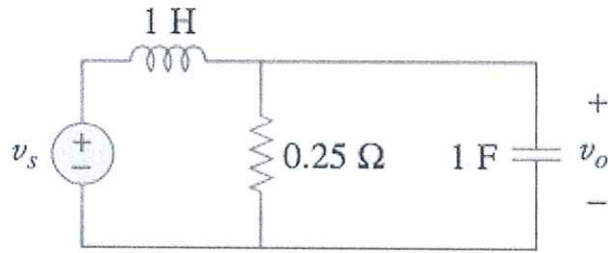


Figure P3

Problem 4)

A series RLC bandpass passive filter has cutoff frequencies at 1 kHz and 10 kHz. The input impedance at resonance is 6Ω .

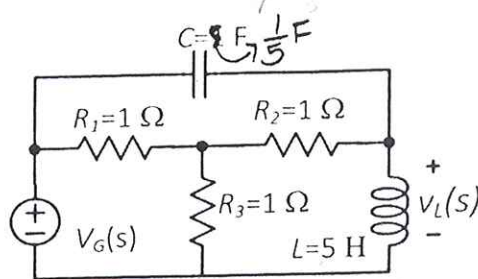
- Calculate the bandwidth, the center frequency and the quality factor of the filter. **(5 points)**
- Calculate the values of circuit components L , C and R . Draw the circuit indicating component values and input and output configuration. **(10 points)**
- Derive the transfer function of the bandpass filter. **(10 points)**

EEEN 202 Midterm Exam Solution Key

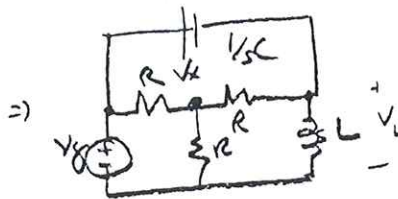
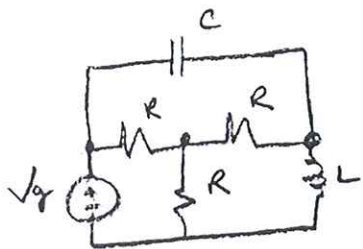
Problem 1)

1) Consider the circuit below. Assume there is no energy stored in the circuit at $t = 0$. Write your answers in the table below:

- Find the transfer function $H(s) = V_L(s) / V_G(s)$.
- Find $v_L(t)$ if $v_G(t) = 2u(t)$.



	Answer
$H(s)$	$= \frac{s(5+3s)}{3s^2+10s+3}$
$v_L(t)$	$= (e^{-3t} + e^{-\frac{1}{3}t})u(t)$



$$I \quad \frac{V_x}{R} + \frac{V_x - V_G}{R} + \frac{V_x - V_L}{R} = 0$$

$$II \quad (V_L - V_G) sC + \frac{V_L}{sL} + \frac{V_L - V_x}{R} = 0$$

From I $\frac{3V_x}{R} = \frac{V_G}{R} + \frac{V_L}{R} \Rightarrow V_x = \frac{V_G + V_L}{3}$

In II $V_L \left(sC + \frac{1}{sL} + \frac{1}{R} \right) = V_G sC + \frac{V_x}{R} = V_G sC + \frac{V_G + V_L}{3R}$

$$V_L \left(sC + \frac{1}{sL} + \frac{1}{R} - \frac{1}{3R} \right) = V_G \left(sC + \frac{1}{3R} \right)$$

$$V_L \left(\frac{s^2 L C (3R) + 3R + 2sL}{3sRL} \right) = V_G \frac{1+3sRC}{3R}$$

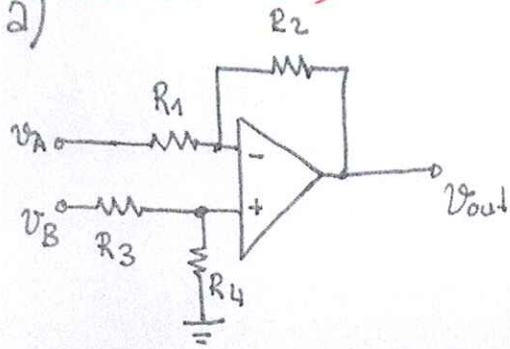
$$\Rightarrow \frac{V_L}{V_G} = \frac{s(1+3RCs)L}{3RLC s^2 + 2sL + 3R} \Rightarrow \text{with the values} \Rightarrow \frac{V_L}{V_G} = \frac{5s(1+\frac{3}{5}s)}{3s^2 + 10s + 3} = \frac{s(5+3s)}{3s^2 + 10s + 3}$$

$$V_L(s) = \frac{2}{5} \frac{s(5+3s)}{3s^2 + 10s + 3} = \frac{2(5+3s)}{(s+3)(3s+1)} = \frac{A}{s+3} + \frac{B}{3s+1}$$

$$\begin{cases} 3A+B=6 \\ A+3B=10 \end{cases} \Rightarrow \begin{cases} A=1 \\ B=3 \end{cases}$$

$$V_L(s) = \frac{1}{s+3} + \frac{3}{3s+1} \Rightarrow v_L(t) = (e^{-3t} + e^{-\frac{1}{3}t})u(t)$$

Problem 2)



Assume opamp is ideal $v_p = v_n$

$$v_p = \frac{v_B \cdot R_4}{(R_3 + R_4)}$$

$$\frac{v_A - v_n}{R_1} = \frac{v_n - v_{out}}{R_2}$$

$$\frac{v_A - \frac{v_B \cdot R_4}{(R_3 + R_4)}}{\left(\frac{R_1}{R_2}\right)} = \frac{\frac{v_B \cdot R_4}{(R_3 + R_4)} - v_{out}}{\frac{R_2}{R_2}} \Rightarrow \boxed{v_{out} = \frac{v_B \cdot R_4 \cdot (1 + \frac{R_2}{R_1})}{(R_3 + R_4)} - v_A \cdot \frac{R_2}{R_1}} \quad (1)$$

$v_A, v_B = f(v_{cm}, v_{dm})$, rewrite (1)

$$v_{out} = v_{cm} \left[\left(\frac{R_4}{(R_3 + R_4)} \cdot \left(1 + \frac{R_2}{R_1} \right) \right) - \frac{R_2}{R_1} \right] + \frac{1}{2} v_{dm} \left[\left(\frac{R_4}{(R_3 + R_4)} \cdot \left(1 + \frac{R_2}{R_1} \right) \right) + \frac{R_2}{R_1} \right]$$

$$R_4 = 1 \text{ k}\Omega, R_2 = 10 \text{ k}\Omega, R_3 = 1 \text{ k}\Omega, R_1 = 1 \text{ k}\Omega$$

$$v_{out} = v_{cm} \left[\left(\frac{1}{2} \times 11 \right) - 10 \right] + v_{dm} \times \frac{1}{2} \left[\frac{11}{2} + 10 \right]$$

$$v_{out} = v_{cm} \times (-4.5) + v_{dm} \cdot (7.75)$$

b) $A_{cm} = -4.5 \text{ V/V}$

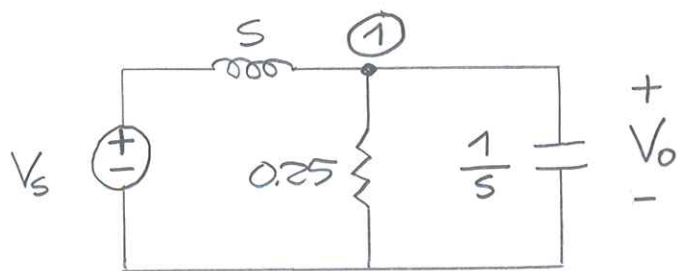
c) $A_{dm} = 7.75 \text{ V/V}$

d) $\text{CMRR} = 20 \log_{10} \left| \left(\frac{A_{dm}}{A_{cm}} \right) \right|$

$$\text{CMRR} = 20 \log_{10}^{1.72} = 4.72 \text{ dB}$$

Problem 3)

a. We consider the s-domain equivalent circuit



KCL at node ① :

$$\frac{V_o - V_s}{s} + \frac{V_o}{0.25} + \frac{V_o}{1/s} = 0$$

$$\Rightarrow \left(\frac{1}{s} + 4 + s \right) V_o = \frac{V_s}{s} \Rightarrow V_s = (s^2 + 4s + 1) V_o$$

$$\Rightarrow H(s) = \frac{V_o}{V_s} = \frac{1}{s^2 + 4s + 1}$$

b. $s = j\omega \Rightarrow H(j\omega) = \frac{1}{(j\omega)^2 + j4\omega + 1}$

$$|H(j\omega)| = \frac{1}{\sqrt{(1-\omega^2)^2 + 16\omega^2}}, \quad \angle H(j\omega) = 0 - \tan^{-1} \frac{4\omega}{1-\omega^2} = -\tan^{-1} \frac{4\omega}{1-\omega^2}$$

$\omega = 0$:

$$|H(j0)| = \frac{1}{\sqrt{(1-0)^2 + 0^2}} = 1, \quad \angle H(j0) = 0^\circ$$

$\omega \rightarrow \infty$:

$$\lim_{\omega \rightarrow \infty} |H(j\omega)| = \lim_{\omega \rightarrow \infty} \frac{1}{\sqrt{(1-\omega^2)^2 + 16\omega^2}} = \lim_{\omega \rightarrow \infty} \frac{1}{\sqrt{\omega^4 + 16\omega^2}} = 0$$

$$\lim_{\omega \rightarrow \infty} \angle H(j\omega) = \lim_{\omega \rightarrow \infty} -\tan^{-1} \frac{4\omega}{1-\omega^2} = \lim_{\omega \rightarrow \infty} -\tan^{-1} \frac{4}{-\omega} = 0^\circ$$

\therefore It is a low pass filter!

c. $|H(j\omega)| = \frac{1}{\sqrt{2}} \cdot \underbrace{H_{max}}_{=1} \Rightarrow \omega \equiv \omega_c$

$$\Rightarrow \frac{1}{\sqrt{(1-\omega_c^2)^2 + 16\omega_c^2}} = \frac{1}{\sqrt{2}} \cdot 1 \Rightarrow (1-\omega_c^2)^2 + 16\omega_c^2 = 2$$

$$\Rightarrow \omega_c^4 + 14\omega_c^2 + 1 = 2 \Rightarrow \omega_c^4 + 14\omega_c^2 - 1 = 0$$

$$\Rightarrow \omega_c^2 = \frac{-14 + \sqrt{14^2 + 4}}{2} = -7 + 5\sqrt{2}$$

$$\Rightarrow \omega_c = \sqrt{5\sqrt{2} - 7} = 0.2666 \text{ rad/sec}$$

(05)

d. The steady-state output response is calculated as

$$v_{o,ss}(t) = |H(j2)| \cos(2t + 30^\circ + \angle H(j2))$$

where

$$|H(j2)| = \frac{1}{\sqrt{(1-4)^2 + 16.4}} = \frac{1}{\sqrt{73}} = 0.117$$

$$\angle H(j2) = -\tan^{-1} \frac{4 \cdot 2}{1-4} = \tan^{-1} \frac{8}{3} = 69.444^\circ$$

(05)

Hence ;

$$\begin{aligned} v_{o,ss}(t) &= 0.117 \cos(2t + 30^\circ + 69.444^\circ) \\ &= 0.117 \cos(2t + 99.444^\circ) \text{ V} \end{aligned}$$

Problem 4)

a) At resonance input impedance of series RLC circuit is equal to R . Hence $R = 6\Omega$

$$\beta = 2\pi(f_{c2} - f_{c1}) = 2\pi \cdot 9 \times 10^3 = 56.52 \text{ krad/s}$$

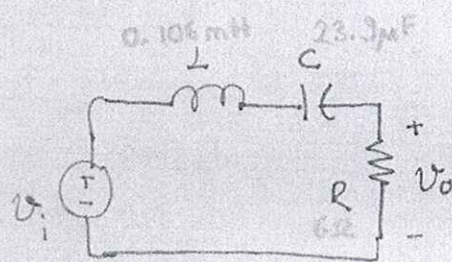
$$\omega_0^2 = \omega_{c1} \cdot \omega_{c2} \Rightarrow \omega_0 = \sqrt{4\pi^2 \times 10 \times 10^6} = 19.85 \text{ krad/s}$$

$$Q = \frac{\omega_0}{\beta} = \frac{19.85 \text{ krad/s}}{56.52 \text{ krad/s}} = 0.35$$

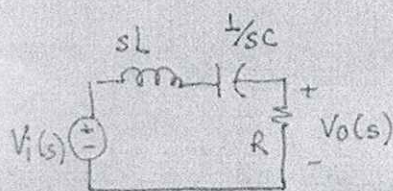
b) $R = 6\Omega$

$$\beta = \frac{R}{L} \Rightarrow L = \frac{6}{56.52 \times 10^3} = 0.106 \text{ mH}$$

$$\omega_0^2 = \frac{1}{LC} \Rightarrow C = \frac{1}{0.106 \times 10^{-3} \times (19.85 \times 10^3)^2} = 23.9 \mu\text{F}$$



c) s-domain



$$\frac{V_o(s)}{V_i(s)} = \frac{R}{sL + \frac{1}{sC} + R} = \frac{sRC \times \frac{1}{LC}}{(s^2 LC + sRC + 1) \times \frac{1}{LC}}$$

$$\boxed{\frac{V_o(s)}{V_i(s)} = \frac{sR/L}{s^2 + sR/L + \frac{1}{LC}}}$$

where $\beta = R/L$

$$\omega_0^2 = 1/LC$$