PROBLEM #1 (30pts)

WE WANT TO CONTROL THE DINAMICS OF X WITH A SIMPLE FIRST-ORDER SYSTEM OF S

SIMPLIFY EXPRESSION USING Y AND Q

WE WANT V < 0 TO HELP SATISFY BARBALATS LEMMA ASSURING S-> O

$$\times + a_1 \times^2 \cos(x) + a_2 \sin(2x) = u$$

a, Laz ARE UNKNOWN CONSTANTS, U. IS CONTROL IN

a) DESIGN ANADAPTINE CONTROLLER FOR THE SYSTEM TO TRACK A DISTRED TRAJECTORY, XI (+)

$$S = (\overrightarrow{A} + \lambda)^{n-1} \times n = 3$$

$$= \overset{\circ}{\times} - \overset{\circ}{\times}_{d} + 2\lambda \overset{\circ}{\times} + \lambda^{2} \overset{\circ}{\times}$$

$$= \overset{\circ}{\times} - \overset{\circ}{\times}_{r} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times} - \overset{\circ}{\times}_{s}$$

$$= \overset{\circ}{\times} - \overset{\circ}{\times}_{r} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{r} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{d} - 2\lambda \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_{s} = \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} - 2\lambda \overset{\circ}{\times}_{s} \qquad \overset{\circ}{\times}_$$

EMPLOY LYAPUNOU THEORY TO ESTABLISH A CONTROL LAW CAPABLE OF DRIVING S-DO AND, THEREFORE; X -> 0

CANDIDATE LYAPUNOV FUNCTION

$$V = \frac{1}{2} \mathcal{R}_0^{1} S^2$$

$$\mathring{V} = S \mathring{S}_0^{1} S^2$$

$$= S (u - Y \tilde{a} - \ddot{x}_r)$$

ASSUME a IS KNOWN, CHOOSE U= 8a-ks + Xr THIS WILL MAKE V = SOMETHING VERY SIMPLE

HOW DO WE ASSURE S -DO ? LETS USE BARBALATS LEAMA

IF (1) V IS LOWER-BOUNDED LD 52 20 b/c 5= x-xn 1 AND @ V IS NEGATIVE SEMI-DEFINITE 1 k>0 50 - k52 50

AND 3 V IS UNIFORMLY CONTINUOUS → VOIS FINITE B/C 52 IS BOUNDED V

THEN V > 0 SON S > 0 > \$ >0

PROBLEM #1

CONTROL LAW

Q = PARAM, EST. ERROR

A TO TAKE DERIVATIVE AND KEEP SOME THING FROM THE AWDED TREM, à MUSTBE TIME VARYING (à (a)

ADAPTATION LAW

HOWEVER, WE DO NOT KNOW a. THE BEST WE CAN DO IS GUESS AT VALUES FOR OUR CONTROLLAW

k>0

NOW,

$$\dot{V} = S(8\hat{a} - ks + \ddot{x}_r - 8\bar{a} - \ddot{x}_r)$$

$$= -ks^2 + S(8\hat{a}) \qquad \tilde{a} = \hat{a} - \bar{a}$$

$$= WANT TO GET RID OF$$

THIS SO I CAN USE BARBALATS

TRY ADDING A TERM TO LYAPUNOV

$$V = \frac{1}{2}S^{2} + \frac{1}{2}\tilde{\alpha}^{T}\tilde{p}^{T}\tilde{\alpha} \qquad \hat{\alpha} \neq 0$$

$$V = S\tilde{S} + \hat{\alpha}^{T}\tilde{p}^{T}\tilde{\alpha} \qquad \hat{\alpha} \neq 0$$

$$= -k\tilde{S} + S\tilde{S}\tilde{\alpha} + \hat{\alpha}^{T}\tilde{p}^{T}\tilde{\alpha} \qquad \hat{\alpha} \neq 0$$

$$= -k\tilde{S} + (S\tilde{S} + \hat{\alpha}^{T}\tilde{p}^{T})\tilde{\alpha} \qquad \hat{\alpha} \neq 0$$

TO ELIMINATE EFFECT OF UNKNOWN PARAMETERS (SX+OTF) = 0

b) SIMULATE THE RESPONSES OF THE SYSTEM USING THE ADARTIVE CONTROLLER DESIGNED IN (a) TO TRACK Xa (+) CHOOSE VALUES FOR P, X, AND K TO ACHIEVE GOALS

FOR GOOD TRACKING BEHAVIOR,

$$P=I$$
 $\lambda=2$ $k=1$

C) SIMULATE THE FOLLOWING CASES AND COMMENT ON EACH PARAMKTER'S IMPACT ON TRACKING PERFORMANCE, PARAMETER CONVERGENCE, AND MAGNITUDE OF CONTROL INPUT

	TRACKING	PHRANKTKX CONVERGENCE	MAGNITUDE OF CONTROL INPUT
INCREASE P	CLOSER	FASTER	GREATER
INCREASE	CLOSER	FASTER	GREATER
INCREASE	CLOSKR	SLOWER	GREATER

(D)

PROBLEM #2

$$\ddot{x} + \alpha_1 \dot{x}^2 \cos(x) + \alpha_2 \sin(2x) = u + da$$

WHERE $d(x) \leq D = const$

a) DESIGN AN ADAPTIVE TRACKING CONTROLLER ROBUST TO THE SPECIFIED DISTURBANCE

$$S = \left(\frac{d}{dt} + \lambda\right)^{n-1} \tilde{\chi} \qquad n = 3$$

$$= \tilde{\chi}' - \tilde{\chi}'_{r} \qquad \tilde{\chi}_{r} = \tilde{\chi}_{d} - 2\chi \tilde{\chi} - \chi^{2} \tilde{\chi}$$

$$\tilde{S} = \tilde{\chi}' - \tilde{\chi}'_{r}$$

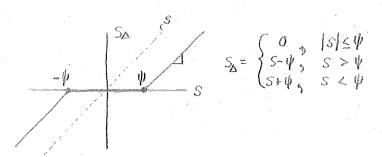
$$= \left(-\alpha_{1} \tilde{\chi}^{2} \cos(x) - \alpha_{2} \sin(2x) + u + d\right) - \tilde{\chi}'_{r}$$

$$= u + d - \chi \tilde{\alpha} - \tilde{\chi}'_{r}$$

$$\downarrow \chi = \left[\tilde{\chi}^{2} \cos(x) \sin(2x)\right] \tilde{\alpha} = \left[\frac{\alpha_{1}}{\alpha_{2}}\right]$$

WE LIGULD LIKE TO S-PO, BUT THE UNKNOWN DISTURBANCE WILL PREVENT THIS FROM HAPPENING. WE WOULD LIKE TO ESTABLISH A BOUNDARY LAYER ABOUT S=0 WHERE WE BELIEVE THE TRACKING ERROR IS DOMINATED BY THE UNKNOWN DISTURBANCE BECAUSE WE HAVE FOUND VALUES FOR Y & WHICH GIVE THE BEST ESTIMATE POSSIBLE. CHOOSE, THEN

SO THE CONTROL AND ADAPTATIONS NOW HAVE A DEAD BAND"



EMPLOY LYAPUNOU TO DEVELOP A CONTROL LAW TO DRIVE S-DO AND, THEREFORE, X < E.

CANDIDATE LYAPUNOV:
$$V = \frac{1}{2} \vec{A}_0 \vec{S}_{\Delta}^2 + \frac{1}{2} \vec{\alpha}^T \vec{P} \vec{\alpha}$$

$$\dot{V} = S_{\Delta} \vec{S} + \hat{\alpha}^T \vec{P} \vec{\alpha}$$

$$= S_{\Delta} (u + d - V \vec{\alpha} - \ddot{X}_r) + \hat{\alpha}^T \vec{P} \vec{\alpha}$$

CHOOSE A CONTROLLAW, DON'T USE -KS, B/C, I THINK, THIS WILL MAKE THE BL BIGGER. INSTRAD, USE -KSat (&) AS WE HAVE BEFORE

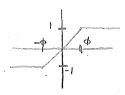
USE ADDITIONAL TERM FROM ADAPTATION EFFORT IN THE LAST PROBLEM

SUB S

(2)

PROBLEM #2

Sat
$$\begin{pmatrix} s \\ \phi \end{pmatrix} = \begin{cases} 1, & s \ge \phi \\ -1, & s \le -\phi \\ \frac{s}{4}, & |s| \le \phi \end{cases}$$



$$\hat{a} = \hat{a} - \bar{a}$$

ADAPTATION LAW

USE BARBALATS LEMMA TO SHOW

V -> O SIMILAR TO PREVIOUS

(BARBALAT OR INVSETTHM?

SO, THE CONTROL LAW IS CHOSEN TO BR

SUCH THAT

$$V = -S_{\Delta}k \operatorname{sat}(x) + S_{\Delta}(d-y) + S_{\Delta} \operatorname{Ya} + \hat{\alpha}^{T} P^{T} \hat{\alpha}$$

$$\Rightarrow \operatorname{IF} \quad \varphi = \psi, \quad \operatorname{THEN} \quad -S_{\Delta}k \operatorname{sat}(x) = -k|S_{\Delta}|$$

$$\Rightarrow \operatorname{IF} \quad \operatorname{CHOOSE}(k=D+T), \quad \operatorname{THEN} \quad -k|S_{\Delta}| + S_{\Delta}(d-y) \leq 0$$

$$V = -k|S_{\Delta}| + S_{\Delta}(d-y) + (S_{\Delta} \operatorname{Y} + \hat{\alpha}^{T} P^{T}) \hat{\alpha}$$

$$\operatorname{MAKE} \quad \operatorname{THIS} = 0$$

NOTICE, B/C V -> O, SA IS "SQUEEZED" TO ZERO

b) SIMULATE RESPONSE TO TRACKING Xd = sin(0,84)

$$a_1 = 2$$
 $a_2 = 5$ $P = 5I$ $\lambda = 3$ $\gamma = 1$
 $D = [0.1 \ 0.5 \ 1.0]$

DESCRIBE EFFECT OF CHANGING D

INCREASE D

- CLOSER TRACKING
- VARIED EFFECT ON PARAMETER ESTIMATION. STEADY-STATE ERROR IS NOT MONOTONIC.
- APPEARS TO INCREASE CONTROL INPUT AT THE START, BUT LESS AS TIME PROGRESSES

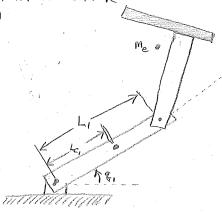
(D) DIRECTLY AFFRCTS (R) ACTING AS A GAIN ON THE CONTROL OUTSIDE OF THE BL. MORE GAIN MEANS FASTER REACHING TIME, BUTTHE POSSIBLE OVERSHOPT IS OBVIATED BY THE BL CONDITION. IT AFFECTS PARAMETER ESTIMATION IN A COMPLEX MANNER BECAUSE IT IMPACTS THE TIME THE STATE IS OUTSIDE THE BL AND, THEREFORE, THE TIME & CONDITIONS ADAPTATION LAW IS APPLIED. HIGHER GAIN MEANS MORE CONTROL EFFORT, CAUSING FASTER CONVENCENCE TO WHERE LESS CONTROL IS REQUIRED.

OBSERVATIONS

EXPLANATIONS

* PROBLEM # 3

m= Kg. I= Kg·m² TUO-LINK MANIPULATOR (EXAMPLE 9,1)



$$\begin{bmatrix} H_{11} & H_{12} \\ H_{21} & H_{22} \end{bmatrix} \begin{bmatrix} \ddot{g}_1 \\ \ddot{g}_2 \end{bmatrix} + \begin{bmatrix} -h\dot{g}_2 & -h(\dot{g}_1 + \dot{g}_2) \\ h\dot{g}_1 & 0 \end{bmatrix} \begin{bmatrix} \ddot{g}_1 \\ \ddot{g}_2 \end{bmatrix} = \begin{bmatrix} \mathcal{T}_1 \\ \mathcal{T}_2 \end{bmatrix}$$

$$H_{11} = a_1 + 2a_3 \cos(g_1) + 2a_4 \sin(g_2)$$
 $H_{12} = H_{21} = a_2 + a_1 \cos(g_2) + a_4 \sin(g_2)$
 $H_{12} = a_2$
 $h = a_2 \sin(g_1) - a_4 \cos(g_2)$
 $a_1 = I_1 + M_1 L_{11}^2 + I_2 + M_2 L_{12}^2$
 $a_2 = I_2 + M_2 L_{12}^2$
 $a_3 = M_2 L_{12} L_{12} L_{12}^2$
 $a_4 = M_2 L_{12}^2$
 $a_4 = M_2 L_{12}^2$
 $a_5 = M_2 L_{12}^2$
 $a_6 = M_2 L_{12}^2$
 $a_6 = M_2 L_{12}^2$
 $a_7 = M_2 L_{12}^2$
 $a_8 = M$

$$m_e = 1$$
 $l_i = 1$ $m_e = 2$ $\delta_e = 30^\circ$ $I_i = 0.12$ $l_{c_i} = 0.5$ $I_e = 0.25$ $l_{c_i} = 0.6$
 $0 t = 0$, $g_1 = g_2 = \hat{g}_1 = \hat{g}_2 = 0$ $g_{d_1} = 60^\circ$ $g_{d_2} = 90^\circ$
 $K_0 = |00 I$ $K_p = 20 K_0$
 $\overline{T} = -K_p \tilde{g}_1 - K_p \tilde{g}_2$

PROBLEM #3a

SIMULATE THE RESPONSE OF THE MANIPULATOR (9,(+) 92(+))
USING A PD-CONTROLLER STARTING FROM ZERO ICS AND
SEEKING THE DESIRED FINAL POSITIONS, Bd

$$M_1 = 1$$
 $l_1 = 1$ $m_e = 2$ $S_e = 30^{\circ}$ $I_1 = 0.12$ $l_{c_1} = 0.5$

$$I_e = 0.25$$
 $l_{ce} = 0.6$ $K_p = 100I$ $K_p = 20K_0$ $\overline{g}_d = \begin{bmatrix} 1 \\ 2 \end{bmatrix}$

IN DESIGNING A PD-CONTROLLER, UE MAY TAKE FOR GRANTED THE CONTROL WILL HAVE THE FORM

MAE 5803

NOTE, THERE IS NOT YET UNCERTAINTY IN THE PROBLEM

AS SUCH, THE SET-UP FOR SIMULATION IS TRIVIAL AFTER TRANSFORMING TO A SYSTEM OF FIRST-ORDER ODES AS DONE ON THE PREVIOUS PAGE

ANALYSIS

THE CONTROL TORQUE IS LARGE, INITIALLY, THIS IS A CONSEQUEIVER OF A LARGE PROPORTIONAL FEEDBACK GAIN

THE RESULTS OF SIMULATION ARE AS EXPRETED.

PROBLEM. #36

X

DESIGN AN ADAPTIVE CONTROLLER FOR THE MANIPULATOR
ASSUMING NO INITIAL KNOWLEDGE OF PARAMETERS, &.
EXAMPLE 9.3 GIVES VALUES FOR MATRIX [Y], SIMULATE
THE RESPONSE TO DESIRED TRAJECTORY

$$\begin{cases} \overline{q}_{d_1} = \left[1 - \overline{e}^{t} \cdot 2(1 - \overline{e}^{t})\right]^T \\ \overline{q}_{d_2} = \left[1 - \cos(2\pi t) \quad 2(1 - \cos(2\pi t))\right]^T \end{cases}$$

$$\begin{cases} P_1 = \text{diag} \left([0.6, 0.1 \ 0.1 \ 0.00] \right) \\ P_2 = 200 \ P_1 \\ P_3 = 0.1 \ P_1 \end{cases}$$

$$V = 50I$$
 $K^b = 100I$

DISCUSS TRACKING PERFORMANCE AND PARAMETER CONVERGENCE FOR EACH CASE.

FIND DEFINITIONS FOR CONTROL LAW AND AVAPTATION LAW USING LYAPUNOV THEORY. FIRST SIMPLIFY TO FIRST-ORDER SYSTEM BY INTRODUSING FINERMEDIATE PARAMETER,

$$\vec{S} = (\vec{A} + \Lambda)^{n-1} \vec{q} \qquad n=2$$

$$= \vec{q} + \Lambda \vec{q}$$

$$= \vec{q} - \vec{q}r \qquad (\vec{q}r = \vec{q}d - \Lambda \vec{q})$$

$$\vec{S} = \vec{q} - \vec{q}r$$

CANDIDATE LYAPUNOV FUNCTION

$$V = \frac{1}{2} \vec{S} T \vec{H} \vec{S} + \frac{1}{2} \vec{\alpha} T \vec{P} \vec{\alpha}$$

$$V = \vec{S} T \vec{H} \vec{S} + \frac{1}{2} \vec{S} T \vec{H} \vec{S} + \hat{\alpha} T \vec{P} \vec{\alpha}$$

$$= \vec{S} T (\vec{H} \vec{g} - \vec{H} \vec{g}_r) + \frac{1}{2} \vec{S} T \vec{H} \vec{S} + \hat{\alpha} T \vec{P} \vec{\alpha}$$

$$= \vec{S} T (\vec{\tau} - \vec{C} \vec{g} - \vec{H} \vec{g}_r) + \frac{1}{2} \vec{S} T \vec{H} \vec{S} + \hat{\alpha} T \vec{P} \vec{\alpha}$$

$$= \vec{S} T (\vec{\tau} - \vec{H} \vec{g}_r - \vec{C} \vec{g}_r) + \frac{1}{2} \vec{S} T \vec{H} \vec{S} + \hat{\alpha} T \vec{P} \vec{\alpha}$$

$$= \vec{S} T (\vec{\tau} - \vec{H} \vec{g}_r - \vec{C} \vec{g}_r) + \frac{1}{2} \vec{S} T (\vec{H} - \vec{G} \vec{g}_r) + \hat{\alpha} T \vec{P} \vec{\alpha}$$

$$= \vec{S} T (\vec{\tau} - \vec{Y} \vec{a}) + \hat{\alpha} T \vec{P} \vec{\alpha}$$

*ASSUMING C IS PROPER"

(Co IS UNIQUE)

COULDFIND Y USING

Ya=Hq-Cq, IF IT WAS NOT GIVEN

PROBLEM #36

CHOOSE T TO MAKE CONTROL LAW (WEUANT V<0)

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 $K_0 = K_0^T > 0$

 $\tilde{\alpha} = \hat{\alpha} - \bar{a}$

50 V = -5 K, 5 + 5 Y & + & T F &

I WANT V < O SO GET RID OF) BY ADDING TO LYAP FUNC. AND CHOOSING &, YIELDING ADAPTATION LAL

NOW, $V = -\overline{S}^T K_D \overline{S} < O$ SO, USING BARBALAT'S LEMMA, 06×3 6 065 6 06 N

SIMULATION IS ACCOMPLISHED IN A SIMILAR MANNERAS BEFORE, IN PART 1 & 2 WHERE QUIS INTEGRATED ALONG WITH THE DYNAMICS AS THOUGH THEY ARE ADDITIONAL STATES.

DISCUSS TRACKING PERFORMANCE AND PARAMETER CONVERGENCE

DESCRIBE EACH DESIRED TRAJECTORY AND EFFECTS

GIG, HAS ONLY EXPONENTIAL TERMS. AS SUCH, THE ACTUATORS MUST WORK HARD TO ADJUST TO PAST POSITION CHANGE, BUT LATER PORS NOT HAVE TO MOVE NUCH.

gd2 HAS ONLY SINUSOIDAL TERMS, AS SUCH, THE DESIRED CHANGE SPEED IS PERSISTENTLY CYCLIC, SO EXPRICT CONTROL EFFORT TO BE CYCLIC

DESCRIBE EFFECT OF MAGNITUDE OF [P]

PJ CONTAINS THE GAINS ACTING ON THE DYNAMICS OF THE PARAMETER ESTIMATION PROGRESSION HIGHER GAINS CAUSE SHORTER REACHING TIME U/ POSSIBILITY FOR OVERSHOOT (ON PARAM ESTIMATION)

OBSELVATIONS OF SYMULATION RESULTS

· EACH CASE WITH Gd, SETTLES TO ZERO POSITION ERROR * EACH CASE WITH Bd, HAS PERSISTENT, CYCLIC ERROR & TORQUE FOR [P] AT SMALLER VALUES, THE COSITION ERROR IS CONTROLLED MORK QUICKLY AND ACCURATICLY, THIS IS BECAUSE THE (E) WALUE PREDICTION DOES NOT OVERSHOOT SO ETREAKLY AS SEEN EN

PLOTS OF (a) FOR [P2].

FOR Gd, SINUSOLDAL > . FOR [P] AT LARGE VALUES, THE SINUSOLDAL Gd2 IS CONTROLLED MORE QUICKLY AND THE RESULTING REDUCTION IN POSITION ERROR CAUSES THE PROPORTIONAL-GAIN CONTROL TORQUE TO BE LOURL OUR TIME, HOLKUER, THE INITIAL RESPONSE CHATTERS FROM LARGE OVERSHOOT,

FOR gd, , KYONKINTIAL

CTOPS