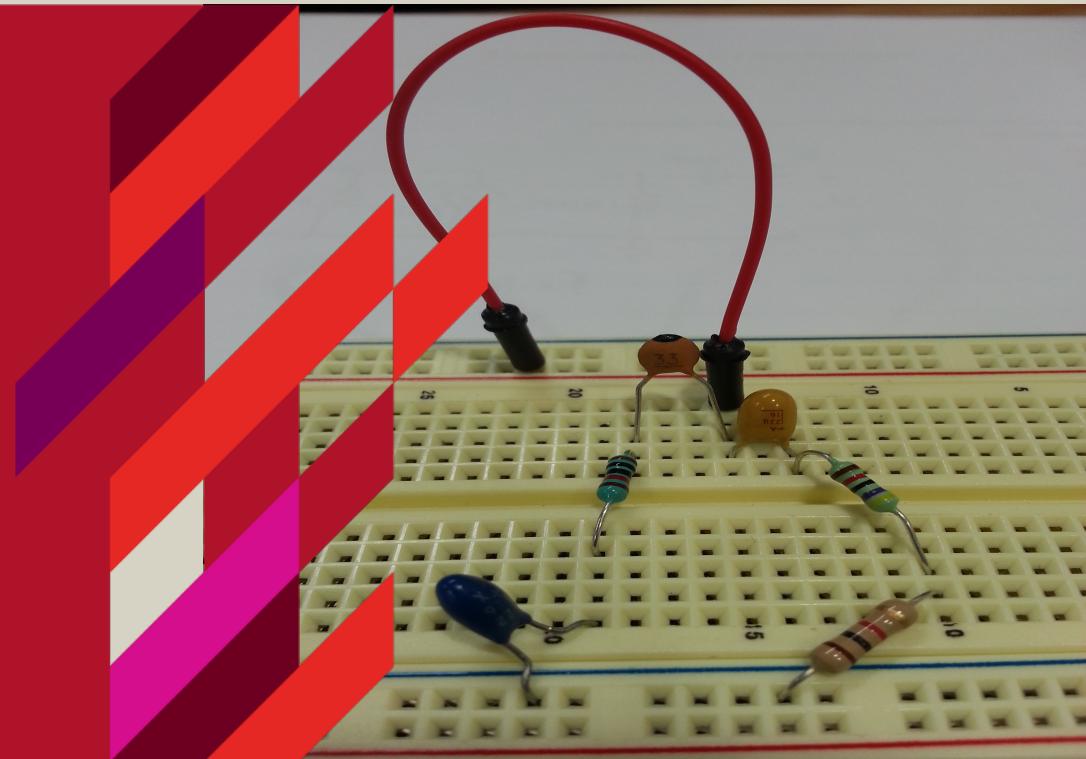




# ELEC2070 Circuits and Devices

Week 7: Frequency domain analysis for steady state sinusoidal inputs

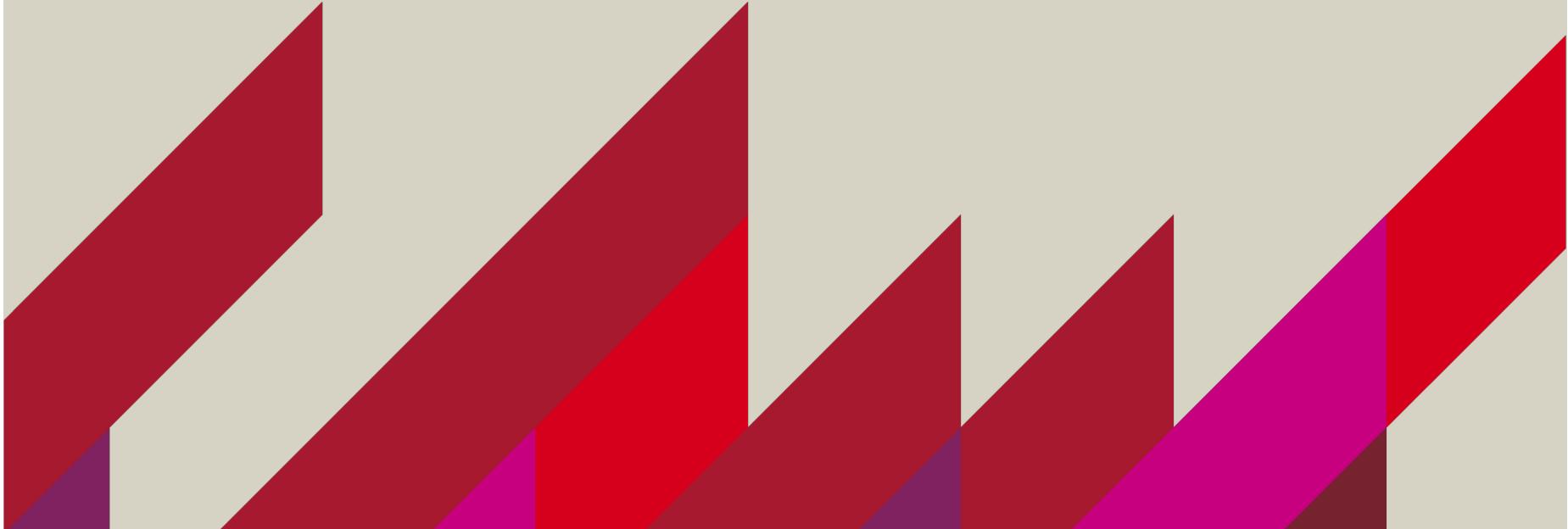
Stuart Jackson





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# Time domain vs frequency domain





# Frequency domain

Up to now we have considered:

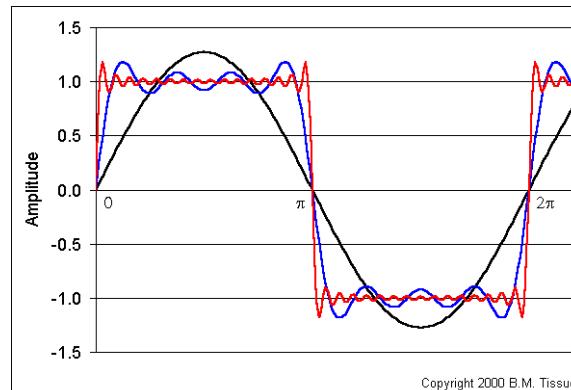
1. DC circuit analysis → mesh analysis, node equations, equivalent circuits, etc
2. Time domain analysis → energy storage, complete response of RL, RC and RLC circuits

Now we move to the **frequency domain**

FYI:

**Fourier Transform converts time-domain functions into frequency-domain functions (Chapter 15)**

**If the time-domain signal is periodic, the frequency-domain signal is a superposition of sinusoids.**





# Periodic signals

Typical periodic signals in time domain and frequency domain  
(you will test some in the laboratory sessions)

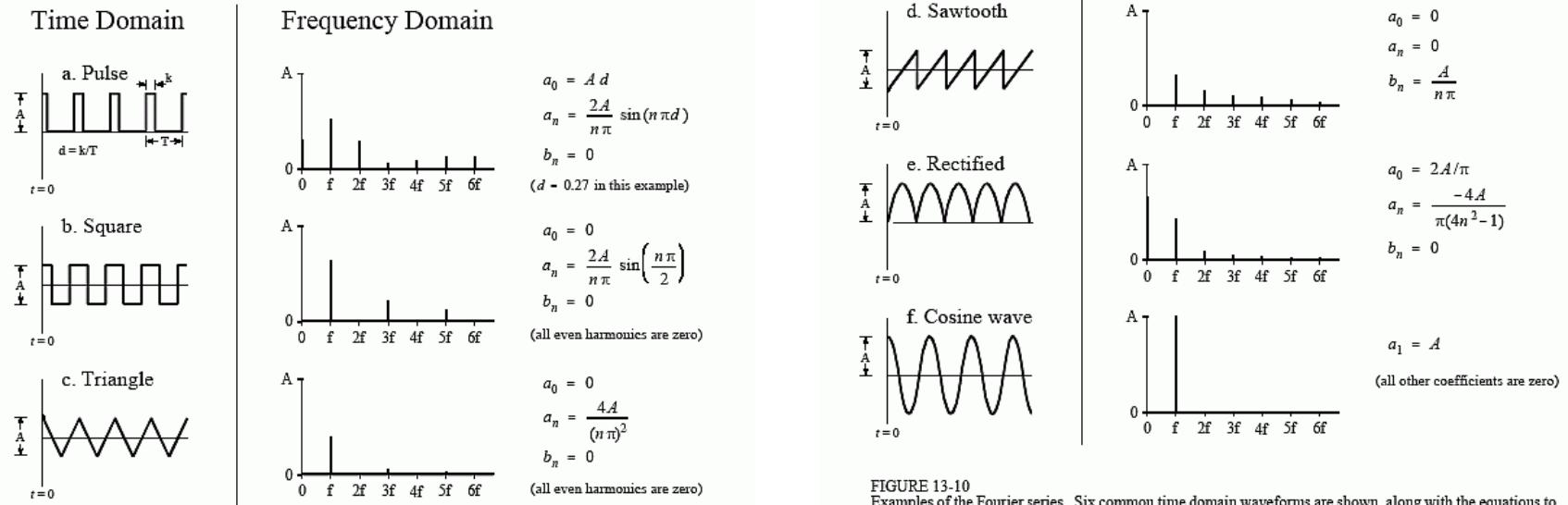


FIGURE 13-10  
Examples of the Fourier series. Six common time domain waveforms are shown, along with the equations to calculate their "a" and "b" coefficients.

<http://www.dspguide.com/ch13/4.htm>

# Analysing circuits in the frequency domain

---

For the next few lectures we will be looking at:

1. Steady state solutions with sinusoidal sources (i.e., **sinusoidal forcing functions**).
2. Forced response only - transients are not calculated.
3. Determine response from AC forcing functions - one frequency at a time.
4. Superposition is used to find the total frequency response.
5. Differential equations are not needed, **complex algebra** is used.



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# AC circuits and sinusoidal functions

Chapter 10 of Dorf and Svoboda





# AC circuits

---

Consider a circuit having a sinusoidal input. The complete response will be:

$$v(t) = v_n(t) + v_f(t)$$

The natural response [ $v_n(t)$ ] will die out leaving the forced response  $v_f(t)$ . When all the inputs have the **same** frequency the forced response will have the **same** frequency.

Overall when the following 2 conditions are fulfilled:

1. All of the inputs are sinusoidal and have the same frequency and
2. the circuit is at steady state,

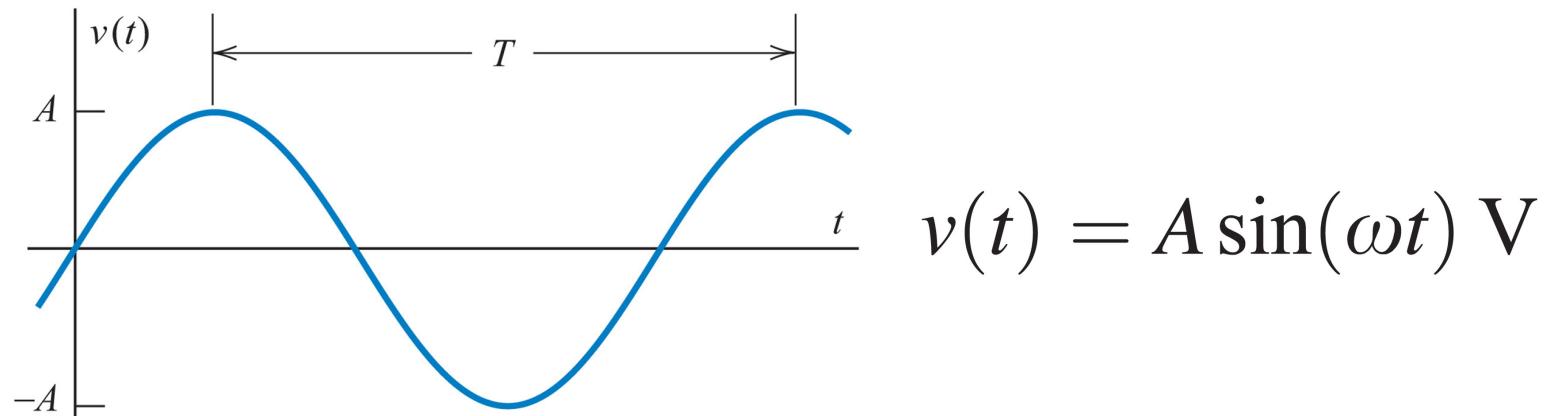
Then **all of the currents and voltages are sinusoidal and have the same frequency as the inputs.**

Traditionally, sinusoidal currents have been called alternating currents (AC) and circuits that satisfy the above conditions are called AC circuits



# Sinusoidal sources

We are going to examine all what we have covered so far in the unit when the independent forcing function is a **sinusoidal** function - AC circuit analysis (**always in the steady state**)



From RL, RC and RLC circuit analysis, we know that the forced response part of the complete response will be a sinusoid with the same frequency!

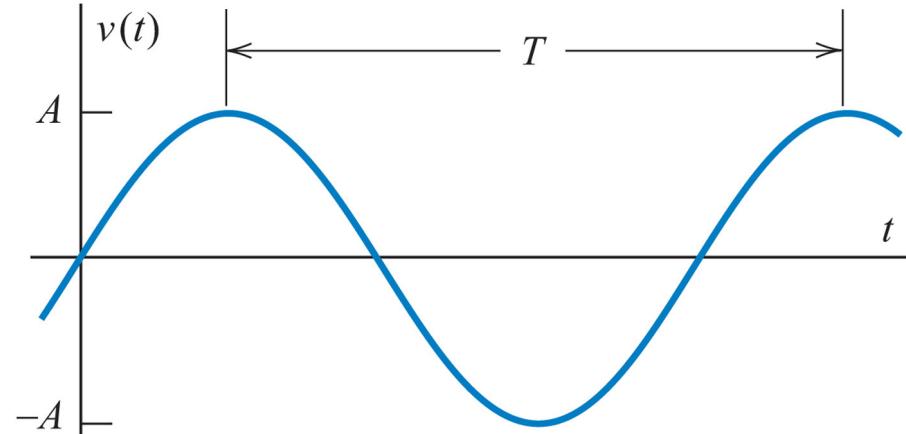
Forced responses with sinusoidal input (forcing functions) are easier dealt with in the **frequency domain**.

# The important parameters relating to a sinusoidal signal

A sinusoid is a periodic function, i.e.:

$$v(t) = v(t + T)$$

Period (s)



The general functional form for a sinusoid is:  $v(t) = A \sin(\omega t)$  (V)

Angular frequency (in radians/second)

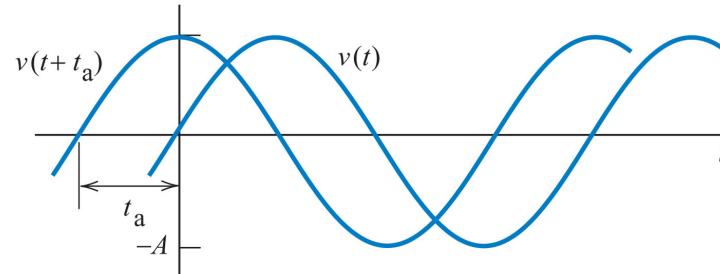
The angular frequency (rad/s) is defined as:  $\omega = 2\pi f$

Where the frequency (in  $s^{-1}$ ) is given by:  $f = \frac{1}{T} = \frac{\omega}{2\pi}$  (Hz)

There are  $2\pi$  radians in a “wavelength”.

# Time advance, delay and phase angle

Consider the effect of replacing  $t$  by  $t + t_a$



$$v(t + t_a) = A \sin[\omega(t + t_a)] = A \sin[\omega t + \omega t_a] = A \sin[\omega t + \theta]$$

↓  
Time  
advance (s)

(in rad)

↑  
Phase angle (rad)

The phase angle (radians) is related to the time  $t_a$ :  $\omega t_a = \theta = \frac{2\pi t_a}{T}$

# Phase angle

---

Similarly, replacing  $t$  by  $t - t_d$  produces a sinusoid that is identical to  $v(t)$  except that  $v(t - t_d)$  is delayed from  $v(t)$  by time  $t_d$ .

We have  $v(t - t_d) = A \sin[\omega(t - t_d)] = A \sin[\omega t - \omega t_d] = A \sin[\omega t + \theta]$

The phase angle here is:  $\theta = -\frac{2\pi t_d}{T}$

NOTE:

1. An advance or delay of a full period leaves a sinusoid unchanged, that is  $v(t \pm T) = v(t)$ . Consequently, an **advance** by time  $t_a$  is equivalent to a **delay** by time  $T - t_a$ .
2. Similarly, a **delay** by time  $t_d$  is equivalent to an **advance** by time  $T - t_d$



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# Phasors and phasor arithmetic





# Phasor notation

---

ANY current or voltage in an AC circuit is a sinusoid at the SAME input frequency. Such a current or voltage is characterised by its amplitude and phase angle.

A **phasor** is a complex number that is used to represent the amplitude and phase angle of a sinusoid. The relationship between the sinusoid and the phasor is described by:

$$A \cos(\omega t + \theta) \quad \leftrightarrow \quad A \angle \theta$$

Note:

1. The sinusoid is represented using the **cosine** rather than the sine function. This is a convention.
2. The phasor is a complex number represented here in **polar form**.

The magnitude of the phasor is equal to the amplitude of the sinusoid, and the angle of the phasor is equal to the phase angle of the sinusoid.

Why phasors? Since all the elements in the circuit have a steady state response with an identical frequency to the source (forcing function) frequency, **only the amplitude and phase of the currents and voltages in every other element in the circuit will be different** – hence we use an abbreviation that just includes amplitude and phase angle!



# Phasors

---

Amplitudes and angles depend on the frequency (even if they are not always shown this way)

$$A(\omega) \cos[\omega t + \theta(\omega)] \leftrightarrow A(\omega) \angle \theta(\omega)$$

Frequency  
Domain !

Lowercase  $i$  and  $v$  is used to indicate a sinusoidal current or voltage (often with a subscript) that is a sinusoidal function of time. Sometimes we will write  $i_1$  instead of  $i_1(t)$ .

We will use bold uppercase  $\mathbf{I}$  and  $\mathbf{V}$  to indicate the corresponding phasor current or voltage with the same subscript.

The phasors are functions of the input frequency. In an AC circuit, the input frequency is fixed and we often shorten  $I_1(\omega)$  to  $I_1$ .

**The bold capital notation of phasor means it is a vector quantity:**

$$v(t) = A \cos[\omega t + \theta] \leftrightarrow \mathbf{V}(\omega) = \mathbf{V} = A \angle \theta$$



## Example 10.3-1

---

Determine the phasors corresponding to the sinusoids:

$$i_1(t) = 120 \cos(400t + 60^\circ) \text{ mA}$$

$$i_2(t) = 100 \sin(400t - 75^\circ) \text{ mA}$$

Solution:

$$\mathbf{I}_1(\omega) = 120\angle 60^\circ \text{ mA}$$

$\sin(-\theta)$	$=$	$-\sin \theta$
$\cos(-\theta)$	$=$	$\cos \theta$
$\tan(-\theta)$	$=$	$-\tan \theta$

And

Need to convert to cosine first

$$i_2(t) = 100 \sin(400t - 75^\circ) = 100 \cos(400t - 75^\circ - 90^\circ)$$

Hence

$$\mathbf{I}_2(\omega) = 100\angle -165^\circ \text{ mA}$$



# Important trig functions

---

$$\sin(-\theta) = -\sin \theta$$

$$\cos(-\theta) = \cos \theta$$

$$\tan(-\theta) = -\tan \theta$$

$$\cos \theta = \sin\left(\frac{\pi}{2} - \theta\right)$$

$$\sin \theta = \cos\left(\frac{\pi}{2} - \theta\right)$$

$$\cot \theta = \tan\left(\frac{\pi}{2} - \theta\right)$$

$$\tan \theta = \cot\left(\frac{\pi}{2} - \theta\right)$$

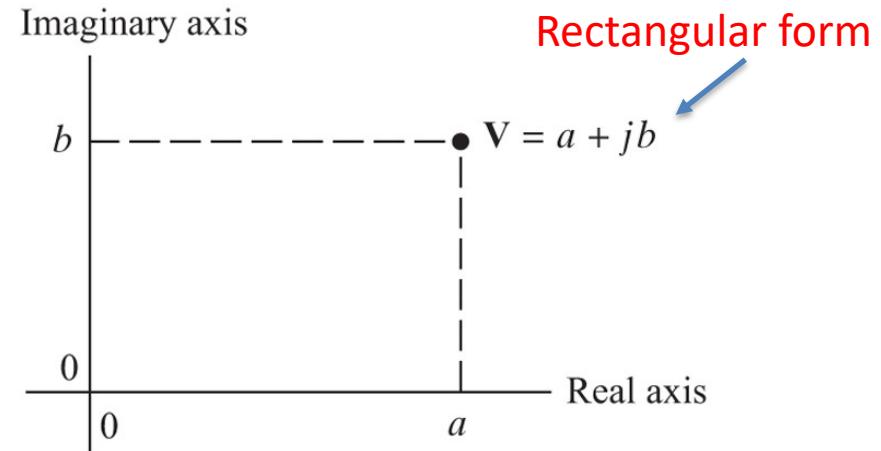
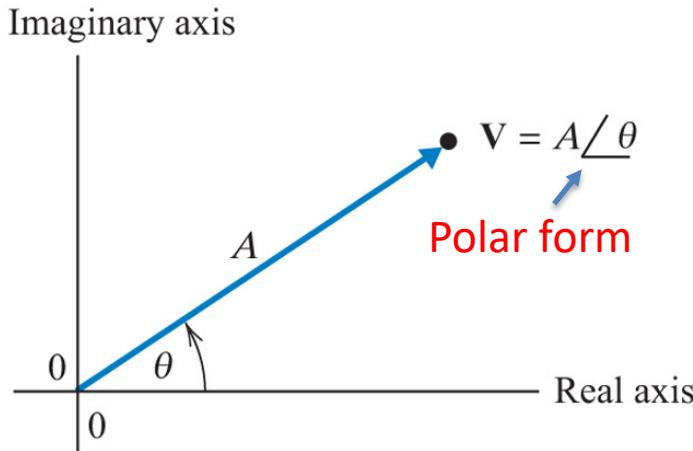
$$\csc \theta = \sec\left(\frac{\pi}{2} - \theta\right)$$

$$\sec \theta = \csc\left(\frac{\pi}{2} - \theta\right)$$

# Defining phasors as complex numbers



We can represent our phasor in the complex plane as: Which has complex coordinates:



$A$  is the magnitude and  $\theta$  is the angle of the phasor:  $A = |\mathbf{V}|$  and  $\theta = \angle \mathbf{V}$

The phasor can be represented as a complex (with a real and imaginary components) number:

$$a = \operatorname{Re}\{\mathbf{V}\} \text{ and } b = \operatorname{Im}\{\mathbf{V}\}$$

Hence we have:  $a + jb = \mathbf{V} = A \angle \theta$

$$\text{Remember } j = \sqrt{-1}$$



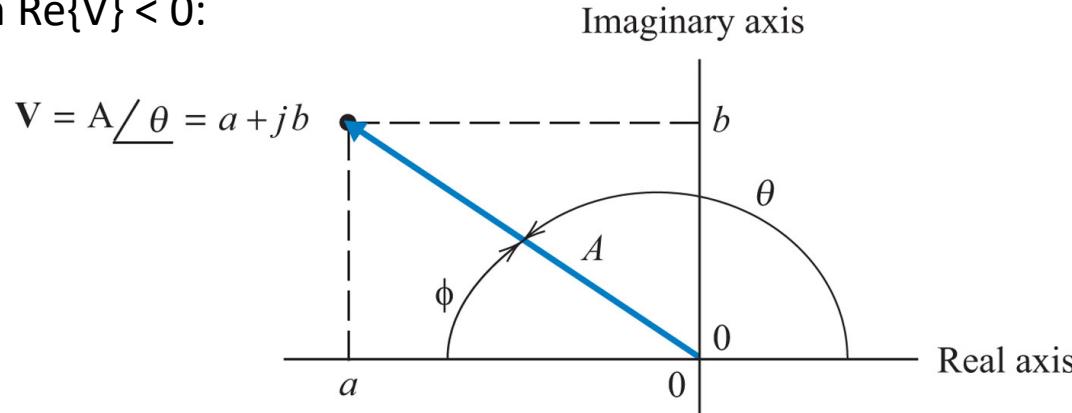
# Why the complex plane?

---

Phasor arithmetic can be easily accomplished using the two forms of the phasor. For the arithmetic operations addition and subtraction we use the rectangular form of the phasor  $\mathbf{V}$  and for division and multiplication we use the polar form for the phasor  $\mathbf{V}$ .

# The phasor angle is positive in the counter clockwise direction

A phasor  $V$  with  $\text{Re}\{V\} < 0$ :



Two important observations

$$a = a \angle 0^\circ$$

$$jb = b \angle 90^\circ$$

Because  $\phi \neq \theta$  we reference the phasor angle **ALWAYS** from the **POSITIVE** real axis

To convert from phasor to complex notation:  $a = A \cos(\theta)$ ,  $b = A \sin(\theta)$ ,  $A = \sqrt{a^2 + b^2}$

with: 
$$\theta = \begin{cases} \tan^{-1}\left(\frac{b}{a}\right) & a > 0 \\ 180^\circ - \tan^{-1}\left(\frac{b}{-a}\right) & a < 0 \end{cases}$$

$$1 = 1 \angle 0^\circ, j = 1 \angle 90^\circ, -1 = 1 \angle \pm 180^\circ \text{ and } -j = 1 \angle -90^\circ = 1 \angle 270^\circ$$



# Complex algebra

---

$$\mathbf{V}_1 + \mathbf{V}_2 = (a + j b) + (c + j d) = (a + c) + j (b + d)$$

For addition and subtraction use rectangular notation

$$\mathbf{V}_1 - \mathbf{V}_2 = (a + j b) - (c + j d) = (a - c) + j (b - d)$$

$$\mathbf{V}_1 \cdot \mathbf{V}_2 = (E \angle \theta) (F \angle \phi) = EF \angle (\theta + \phi)$$

For multiplication and division use phasor notation

$$\frac{\mathbf{V}_1}{\mathbf{V}_2} = \frac{A \angle \theta}{B \angle \phi} = \frac{A}{B} \angle (\theta - \phi)$$

The complex conjugate

$$\begin{aligned}\mathbf{V}_1^* &= (a + jb)^* = a - jb \\ &= (E \angle \theta)^* = E \angle -\theta\end{aligned}$$



# The exponential form of a phasor

Euler's formula is given by:

$$e^{j\theta} = \cos \theta + j \sin \theta \quad \cos \theta = \frac{e^{j\theta} + e^{-j\theta}}{2} \quad \sin \theta = \frac{e^{j\theta} - e^{-j\theta}}{2j}$$

Svoboda & Dorf: Appendix B

If

$$Ae^{j\phi} = A \cos \phi + j A \sin \phi$$

and

$$A\langle\phi = A \cos \phi + j A \sin \phi$$

then

$$A e^{j\phi} = A \angle \phi$$

The exponential  
form of a phasor

The conversion between polar and exponential forms of phasors is immediate.

# Forms of a complex voltage or current

$$A e^{j(\omega t + \theta)} = A \cos(\omega t + \theta) + j A \sin(\omega t + \theta)$$

$$A \cos(\omega t + \theta) = \operatorname{Re} \left\{ A e^{j(\omega t + \theta)} \right\}$$

POLAR FORM:  $\mathbf{V} = M \angle \theta$

RECTANGULAR FORM:  $\mathbf{V} = a + jb$

EXPONENTIAL FORM:  $\mathbf{V} = M e^{j\theta}$

# KVL and KCL with complex values

---

KVL:

If a set of sinusoidal voltages  $v_i(t)$  satisfy KVL for an ac circuit, the corresponding phasor voltages  $V_i(\omega)$  satisfy the same KVL equation.

KCL:

If a set of sinusoidal currents  $i_i(t)$  satisfy KCL for an AC circuit, the corresponding phasor currents  $I_i(\omega)$  satisfy the same KCL equation.



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# Circuit elements in the time and frequency domains

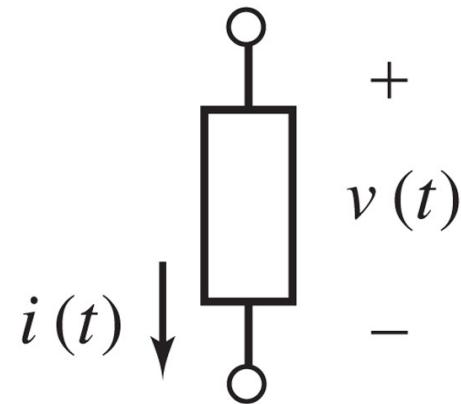


# Voltage and current in the time and frequency domains

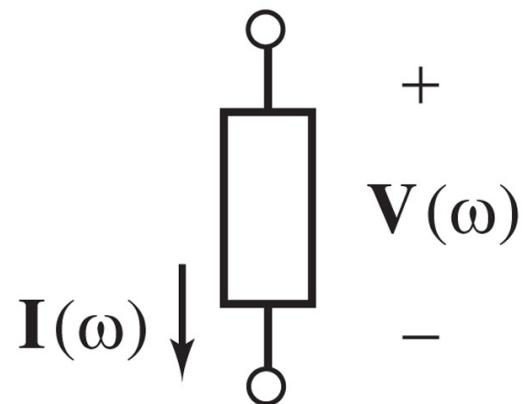


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Time domain



Frequency domain



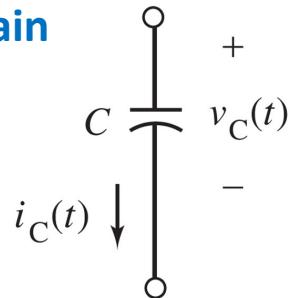
**VOLTAGE:**  $v(t) = V_m \cos[\omega t + \theta] \leftrightarrow V(\omega) = V_m \angle \theta$

**CURRENT:**  $i(t) = I_m \cos[\omega t + \phi] \leftrightarrow I(\omega) = I_m \angle \phi$

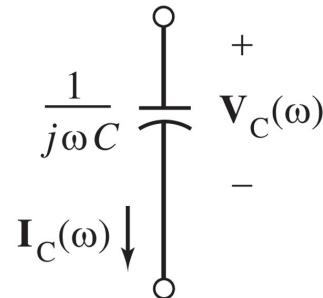
# Capacitors in the time and frequency domains



Time domain



Frequency domain



**VOLTAGE:**  $v_C(t) = V_{Cm} \cos[\omega t + \theta] \leftrightarrow V_C(\omega) = V_{Cm} \angle \theta$

Voltage and current are not in phase

**CURRENT:**  $i_C(t) = C \frac{dv_C(t)}{dt} = -C\omega V_{Cm} \sin[\omega t + \theta] = C\omega V_{Cm} \cos[\omega t + \theta + 90^\circ]$

CURRENT LEADS THE VOLTAGE BY  $90^\circ$

$$I_C(\omega) = C\omega V_{Cm} \angle (\theta + 90^\circ) = jC\omega V_C(\omega)$$

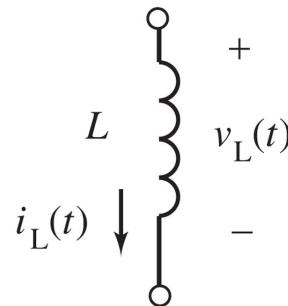
$$I_C(\omega) = j\omega C V_C(\omega)$$

Remember:  $j = 1 \angle 90^\circ$

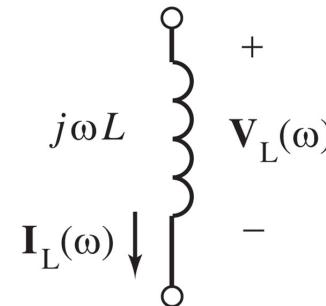
# Inductors in the time and frequency domains



Time domain



Frequency domain



CURRENT:  $i_L(t) = I_{Lm} \cos[\omega t + \phi] \leftrightarrow \mathbf{I}_L(\omega) = I_{Lm} \angle \phi$

Voltage and current are not in phase

VOLTAGE:  $v_L(t) = L \frac{di_L(t)}{dt} = -L\omega I_{Lm} \sin[\omega t + \phi] = L\omega I_{Lm} \cos[\omega t + \phi + 90^\circ]$

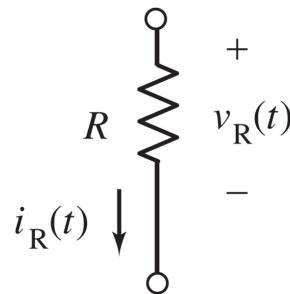
VOLTAGE LEADS THE CURRENT BY  $90^\circ$

$$v_L(t) = L \frac{di_L(t)}{dt} \leftrightarrow \mathbf{V}_L(\omega) = L\omega I_{Lm} \angle (\phi + 90^\circ) = jL\omega \mathbf{I}_L(\omega)$$

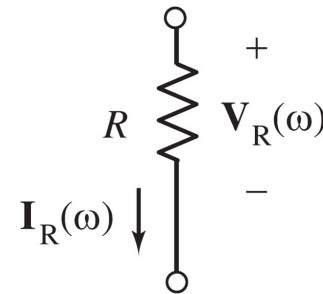
$$\mathbf{V}_L(\omega) = jL\omega \mathbf{I}_L(\omega)$$

# Resistors in the time and frequency domains

Time domain



Frequency domain



**CURRENT:**  $i_R(t) = I_{Rm} \cos[\omega t + \phi] \leftrightarrow \mathbf{I}_R(\omega) = I_{Rm} \angle \phi$

**VOLTAGE:**  $v_R(t) = R i_R(t) = R I_{Rm} \cos[\omega t + \phi] \leftrightarrow \mathbf{V}_R(\omega) = R I_{Rm} \angle \phi = R \mathbf{I}_L(\omega)$

$$\mathbf{V}_R(\omega) = R \mathbf{I}_L(\omega)$$

Voltage and  
current are in  
phase

## Voltage and Current Relationships In Time Domain

	Resistor	Capacitor	Inductor
Voltage	$R i_R$	$v_C$ (no instantaneous change in time)	$L \frac{di_L}{dt}$
Current	$i_R$	$C \frac{dv_C}{dt}$	$i_L$ (no instantaneous change in time)

## Voltage and Current Relationships using s-operators (In Time Domain)

	Resistor	Capacitor	Inductor
Voltage	$R i_R$	$\frac{1}{sC} i_C$	$sL i_L$
Current	$i_R$	$i_C$	$i_L$

## Voltage and Current Relationships In Frequency Domain

	Resistor	Capacitor	Inductor
Voltage	$R I_R$	$\frac{1}{j\omega C} I_C$	$j\omega L I_L$
Current	$I_R$	$I_C$	$I_L$



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# Impedance and Admittance





# Ohm's Law in the frequency domain

Impedance is similar to the resistance in the time domain

A sinusoidal voltage through  
a circuit element:

$$v(t) = V_m \cos[\omega t + \theta] \leftrightarrow V(\omega) = V_m \angle \theta$$

Note same frequency      Different phase angle

The corresponding current:

$$i(t) = I_m \cos[\omega t + \phi] \leftrightarrow I(\omega) = I_m \angle \phi$$

We define:

**Impedance:**  $Z(\omega) = \frac{V(\omega)}{I(\omega)} = \frac{V_m \angle \theta}{I_m \angle \phi} = \frac{V_m}{I_m} \angle (\theta - \phi)$

**Ohm's Law:**  $V(\omega) = Z(\omega) \cdot I(\omega)$

Impedance  
unit is  $\Omega$  (ohm)

Impedance vocabulary:

(Since impedance is complex)

$$Z(\omega) = R(\omega) + jX(\omega)$$

Resistance (real part)      Reactance (imaginary part)

# Ohm's Law

In terms of admittance (similar to conductance in time domain)

---

Admittance:

$$Y(\omega) = \frac{1}{Z(\omega)} = \frac{I(\omega)}{V(\omega)}$$

Admittance unit is  $1/\Omega$  (mho)  
or Siemens (S)

For an AC circuit in the frequency domain, a RLC circuit can be analysed using the impedances since the system is steady state. Remember in DC circuits, capacitors and inductors act like open and short circuits.



# Impedances of circuit elements

---

Capacitor

$$\mathbf{Z}_C(\omega) = \frac{\mathbf{V}_C(\omega)}{\mathbf{I}_C(\omega)} = \frac{1}{j\omega C} \quad \Omega$$

Inductor

$$\mathbf{Z}_L(\omega) = \frac{\mathbf{V}_L(\omega)}{\mathbf{I}_L(\omega)} = j\omega L \quad \Omega$$

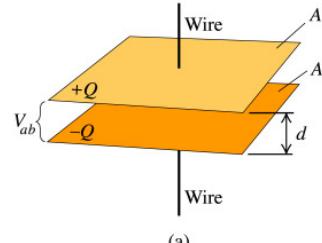
Resistor

$$\mathbf{Z}_R(\omega) = \frac{\mathbf{V}_R(\omega)}{\mathbf{I}_R(\omega)} = R \quad \Omega$$

This means that the value of the impedance **CHANGES** with the frequency. For every source frequency applied to a circuit, the impedances of **EVERY** element in that circuit will change.

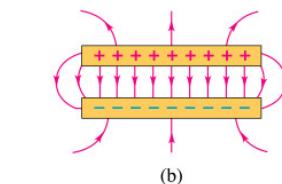
# What does impedance mean?

It is a combination of electric and magnetic fields and resistance



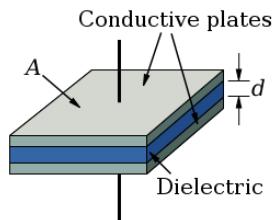
$$C = \frac{Q}{V}$$

Electric charge →  
Voltage →

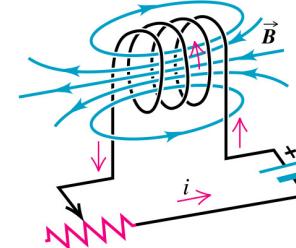


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<http://www.physics.sjsu.edu/becker/physics51/capacitors.htm>



$$C = \frac{\epsilon_r \epsilon_0 A}{d}$$



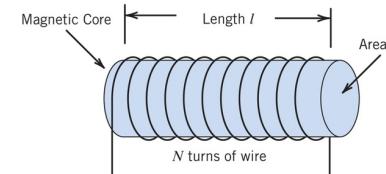
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$$B = \mu \frac{N}{l} I$$

Current →

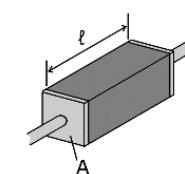
Magnetic field

<http://www.physics.sjsu.edu/becker/physics51/induction.htm>



$$L = \frac{\mu_r \mu_0 N^2 A}{l}$$

- L, C, and R values don't change with frequency
- But impedance does!



$$R = \frac{\rho l}{A}$$



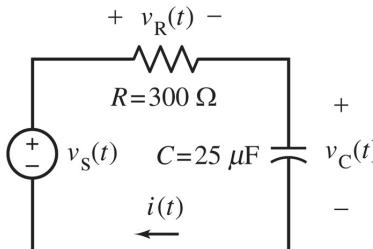
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# Kirchhoff's Laws using phasors



# Kirchhoff's Laws

Apply in the frequency domain



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Example 10.3-4 (Dorf 9<sup>th</sup> ed.)

Kirchhoff's laws apply to all time domain voltages and currents.

Therefore they work for sinusoidal voltages and currents.

Hence they work in frequency domain too.

Source voltage:  $v_s(t) = 25 \cos(100t + 15^\circ)$  V      In phasor form:  $\mathbf{V}_s(100) = 25\angle 15^\circ$  V

KVL:  $\mathbf{V}_s(\omega) = \mathbf{Z}_R(\omega) \mathbf{I}(\omega) + \mathbf{Z}_C(\omega) \mathbf{I}(\omega)$

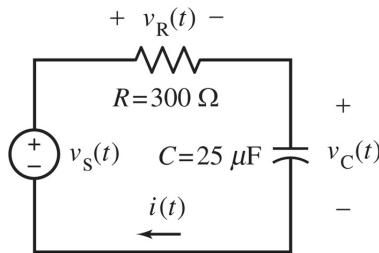
CAPACITOR:  $\mathbf{Z}_C(100) = \frac{1}{j100C} = \frac{1}{j100 \times 25 \times 10^{-6}} = -j400 \Omega$       RESISTOR:  $\mathbf{Z}_R(100) = 300 \Omega$

$$\mathbf{I}(100) = \frac{\mathbf{V}_s(100)}{\mathbf{Z}_R(100) + \mathbf{Z}_C(100)} = \frac{25\angle 15^\circ}{300 - j400} = \frac{25\angle 15^\circ}{500\angle(-53.1^\circ)} = 0.05\angle 68.1^\circ \text{ A}$$

$$i(t) = 0.05 \cos(100t + 68.1^\circ) \text{ A}$$

# Voltage division

Applies in the frequency domain



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Example 10.3-4

Source voltage:  $v_s(t) = 25 \cos(100t + 15^\circ)$  V

In phasor form:  $\mathbf{V}_s(100) = 25\angle 15^\circ$  V

$$\text{Voltage division: } \mathbf{V}_R(\omega) = \frac{\mathbf{Z}_R(\omega)}{\mathbf{Z}_C(\omega) + \mathbf{Z}_R(\omega)} \mathbf{V}_S(\omega)$$

$$\mathbf{V}_R(100) = \frac{300}{300 - j400} 25\angle 15^\circ = \frac{3}{3 - j4} 25\angle 15^\circ = \frac{3}{5\angle(-53.1^\circ)} 25\angle 15^\circ = 15\angle 68.1^\circ$$

$$v_R(t) = 15 \cos(100t + 68.1^\circ) \text{ V}$$

Remember to convert to radians  
before calculating cosine!

# Current and voltage division



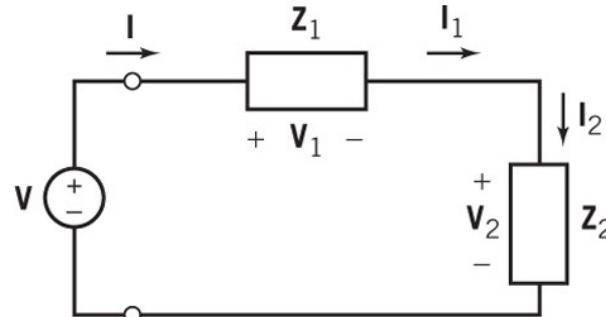
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Also applies in the frequency domain

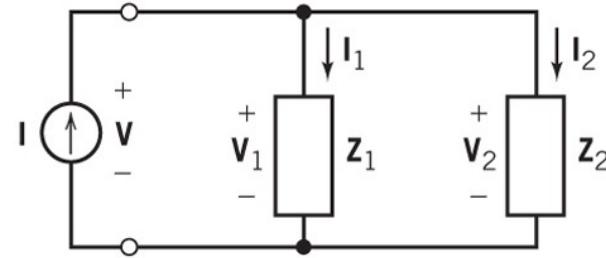
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CIRCUIT

Voltage division



Current division



EQUATIONS

$$I_1 = I_2 = I$$

$$V_1 = \frac{Z_1}{Z_1 + Z_2} V$$

$$V_2 = \frac{Z_2}{Z_1 + Z_2} V$$

$$V_1 = V_2 = V$$

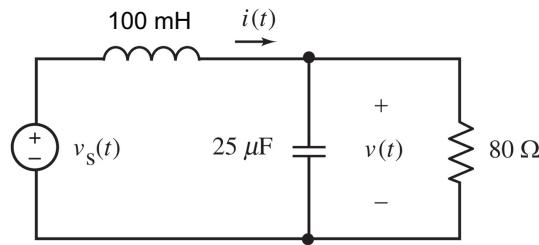
$$I_1 = \frac{Z_2}{Z_1 + Z_2} I$$

$$I_2 = \frac{Z_1}{Z_1 + Z_2} I$$

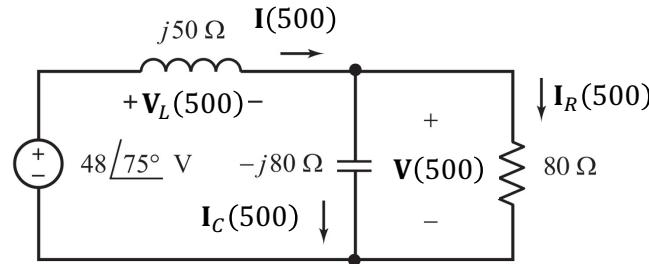
# Circuits in the frequency domain

Example 10.4-2

Time Domain



Frequency Domain



$$v_s(t) = 48 \cos(500t + 75^\circ) \text{ V}$$

$$\omega = 500 \text{ rad/s}$$

$$V_s(500) = 48 \angle 75^\circ \text{ V}$$

$$Z_R = 80 \Omega$$

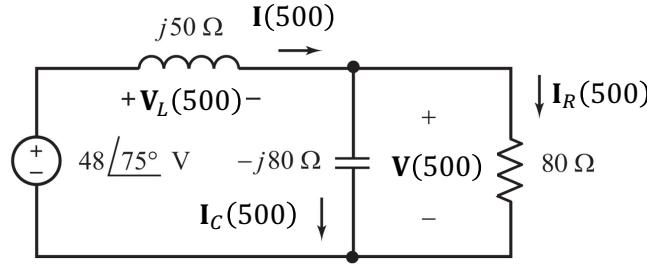
$$Z_C(500) = \frac{1}{j500 C} = -j80 \Omega$$

$$Z_L(500) = j500 L = j50 \Omega$$



# Solution to Example 10.4-2

## Frequency Domain



$$\text{KVL: } \mathbf{V}_s(500) = \mathbf{V}_L(500) + \mathbf{V}(500)$$

$$\text{KCL: } \mathbf{I}(500) = \mathbf{I}_C(500) + \mathbf{I}_R(500)$$

$$\text{OL: } \mathbf{V}_L(500) = j50 \cdot \mathbf{I}(500)$$

$$\mathbf{V}_s(500) = 48 \angle 75^\circ \text{ V}$$

$$\mathbf{I}_C(500) = \frac{\mathbf{V}(500)}{-j80}$$

$$\mathbf{I}_R(500) = \frac{\mathbf{V}(500)}{80}$$

$$\text{KVL now becomes: } 48 \angle 75^\circ = j50 \left( \frac{\mathbf{v}}{-j80} + \frac{\mathbf{v}}{80} \right) + \mathbf{v}$$

Short notation to reduce complexity of the calculation:

$$48 \angle 75^\circ = (-0.625 + j0.625 + 1)\mathbf{v} = (0.375 + j0.625)\mathbf{v}$$

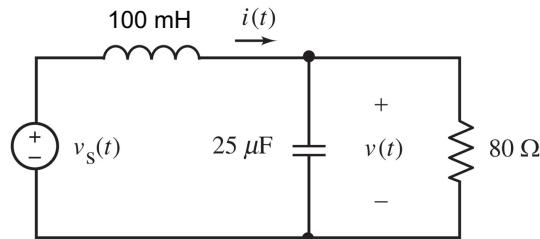
$$\mathbf{v} = \frac{48 \angle 75^\circ}{0.375 + j0.625} = 65.9 \angle 16^\circ$$

# Solution in both domains



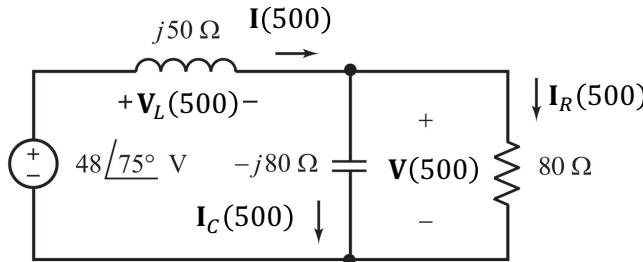
Example 10.4-2

## Time Domain



$$v_s(t) = 48 \cos(500t + 75^\circ) \text{ V}$$

## Frequency Domain



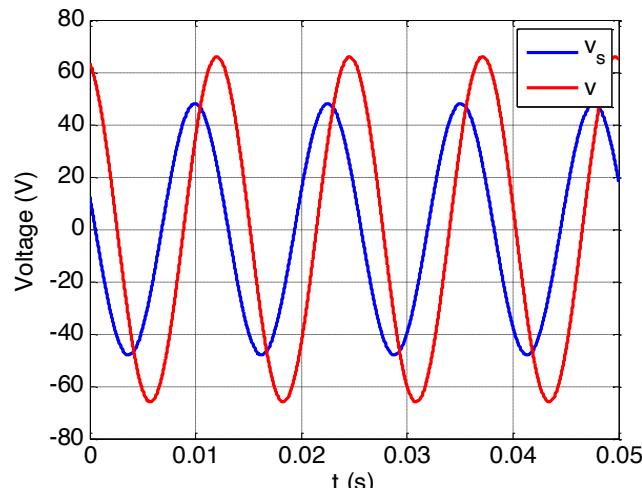
$$V_s(500) = 48 \angle 75^\circ \text{ V}$$

$$v(t) = 65.9 \cos(500t + 16^\circ) \text{ V}$$

$$f = \frac{\omega}{2\pi} = \frac{500}{2\pi} = 79.6 \text{ Hz}$$

$$T = \frac{1}{f} = 12.6 \text{ ms}$$

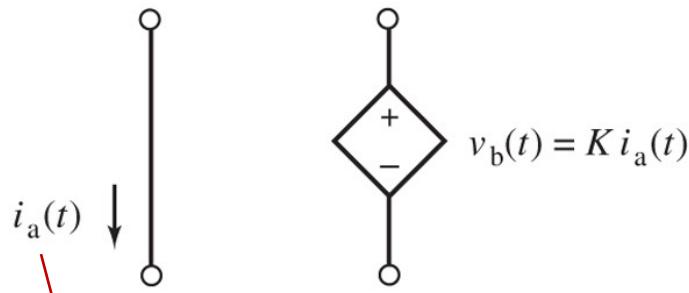
$$V(500) = 65.9 \angle 16^\circ$$



# Dependent sources in the time and frequency domains



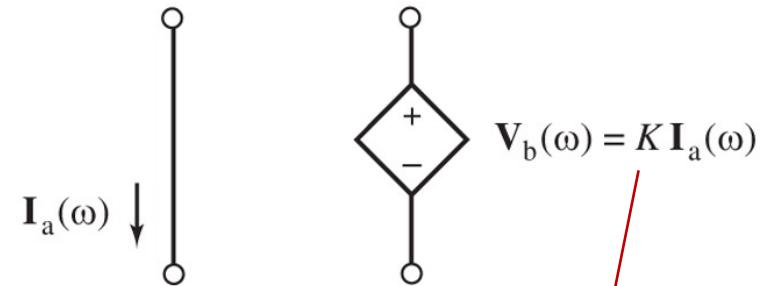
Time Domain



$$v_b(t) = K i_a(t)$$

Controlling current:  $i_a(t) = A \cos(\omega t + \theta)$

Frequency Domain



$$\mathbf{V}_b(\omega) = K \mathbf{I}_a(\omega)$$

$K$  can be a complex number

The current controlled voltage is:  $v_b(t) = KA \cos(\omega t + \theta)$

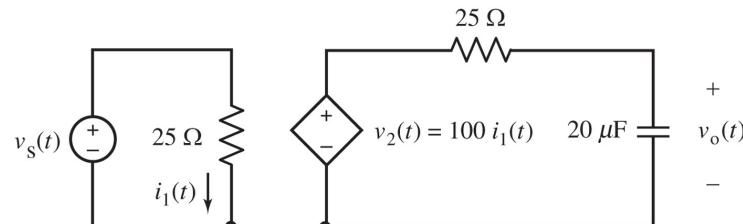
Using phasor notation:  $\mathbf{I}_a(\omega) = A\langle\theta\rangle$  and  $\mathbf{V}_b(\omega) = KA\langle\theta\rangle$ , hence  $\mathbf{V}_b(\omega) = K\mathbf{I}_a(\omega)$

# Circuit with a dependent source



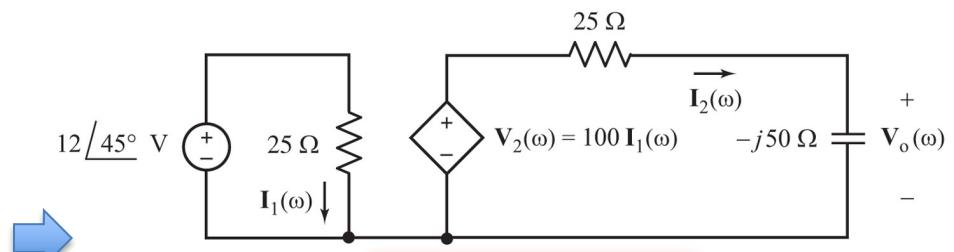
Example 10.4-3

**Time Domain**



$$v_s(t) = 12 \cos(1000t + 45^\circ) \text{ V}$$

**Frequency Domain**



$$\omega = 1000 \text{ rad/s}$$

$$V_s = 12\angle 45^\circ \text{ V}$$

$$Z_C = \frac{1}{j1000C} = \frac{1}{j1000 \times 20 \times 10^{-6}} = -j50 \Omega$$

$$I_1 = \frac{12\angle 45^\circ}{25} = 0.48\angle 45^\circ \text{ A} = \text{controlling current}$$

$$V_2 = 100I_1 = 48\angle 45^\circ \text{ A}$$

$$V_o = \frac{-j50}{25 - j50} V_2 = \frac{-j2}{1 - j2} V_2 = \frac{2\angle(-90^\circ)}{2.24\angle(-63.4^\circ)} 48\angle 45^\circ = 42.93\angle 18.4^\circ \text{ V}$$



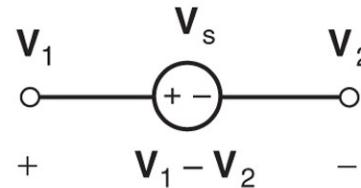
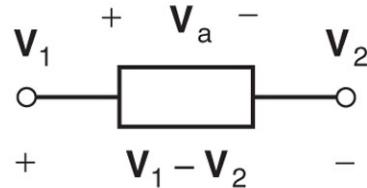
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# Node voltage and mesh current analysis using phasors

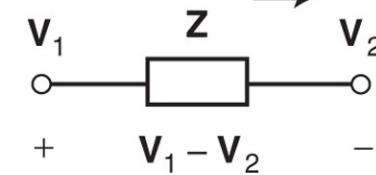


# Node analysis in the frequency domain

- Express element voltages and currents in terms of node voltages.
- Same as the DC case, but using phasors.
- Apply Kirchhoff's Current Law at each node.
- Solve simultaneous equations to find each node voltage.



$$I = \frac{V_1 - V_2}{Z}$$



(a)

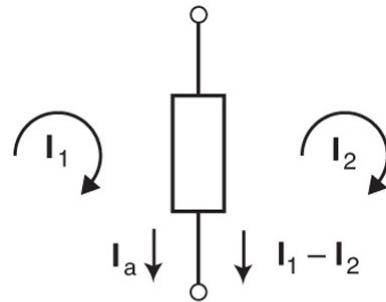
(b)

(c)

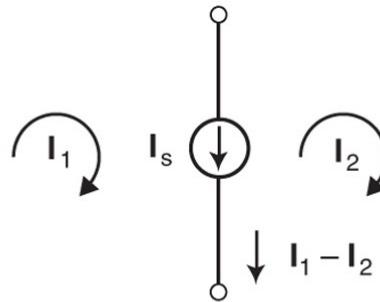
# Mesh equations in the frequency domain



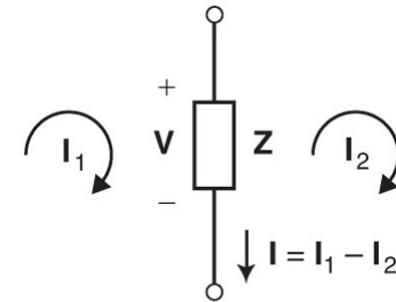
- Express element voltages and currents in terms of mesh currents.
- Same as the DC case, but using phasors.
- Apply Kirchhoff's Voltage Law for each mesh.
- Solve simultaneous equations to find each mesh current.



(a)



(b)

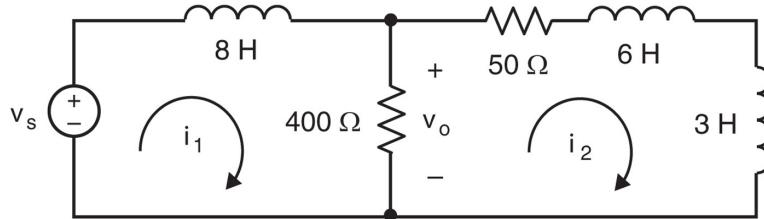


(c)

# Problem 10.6-8 – Mesh analysis



Find  $i_1$ ,  $i_2$  and  $v_o$



$$v_s(t) = 25 \cos(40t + 45^\circ) \text{ V}$$



$$\omega = 40 \text{ rad/s}$$

Convert to frequency domain (phasors)



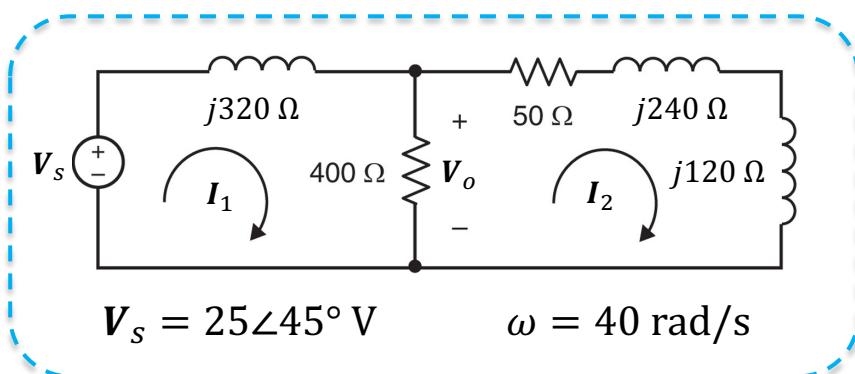
$$V_s = 25\angle 45^\circ \text{ V}$$

$$Z_L(\omega) = j\omega L$$

$$Z_{L8} = j40 \cdot 8 = j320 \Omega$$

$$Z_{L6} = j40 \cdot 6 = j240 \Omega$$

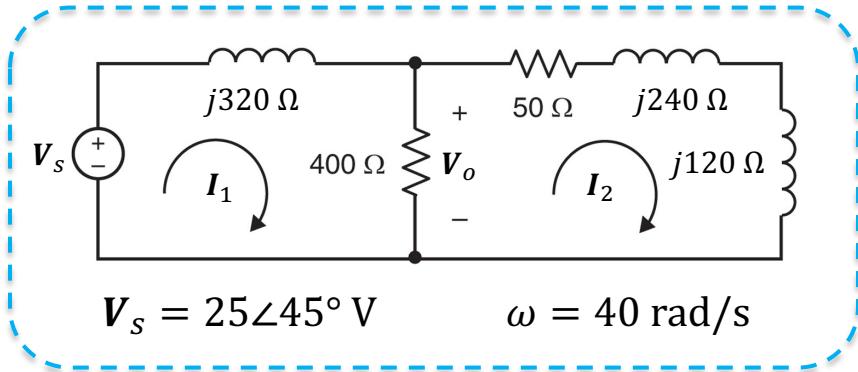
$$Z_{L3} = j40 \cdot 3 = j120 \Omega$$



# Problem 10.6-8



Find  $i_1$ ,  $i_2$  and  $v_o$



Two mesh currents makes two equations

$$(1) \quad 25\angle 45^\circ = j320I_1 + 400(I_1 - I_2)$$

$$25\angle 45^\circ = (400 + j320)I_1 - 400I_2$$

$$(2) \quad 0 = 400(I_2 - I_1) + 50I_2 + j240I_2 + j120I_2$$

$$0 = -400I_1 + (400 + 50 + j240 + j120)I_2$$

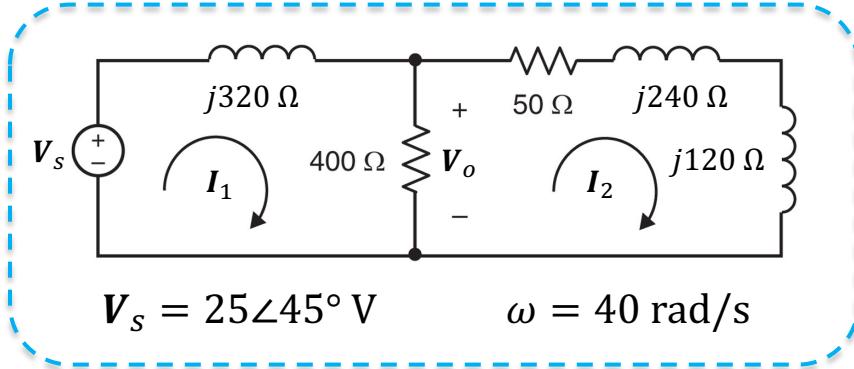
$$0 = -400I_1 + (450 + j360)I_2$$

Solve (1) and (2) simultaneously

# Problem 10.6-8



Find  $i_1$ ,  $i_2$  and  $v_o$



Solve (1) and (2) simultaneously

$$25\angle 45^\circ = (400 + j320)\mathbf{I}_1 - 400\mathbf{I}_2 \quad (1)$$

$$0 = -400\mathbf{I}_1 + (450 + j360)\mathbf{I}_2 \quad (2)$$

Using Matlab:

$$\begin{bmatrix} 25 \angle 45^\circ \\ 0 \end{bmatrix} = \begin{bmatrix} 400 + j320 & -400 \\ -400 & 450 + j360 \end{bmatrix} \cdot \begin{bmatrix} \mathbf{I}_1 \\ \mathbf{I}_2 \end{bmatrix}$$

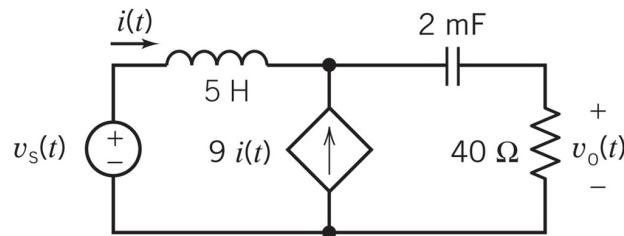
$$\begin{bmatrix} \mathbf{I}_1 \\ \mathbf{I}_2 \end{bmatrix} = \begin{bmatrix} 47.5 \angle -24.63^\circ \\ 33.3 \angle -63.29^\circ \end{bmatrix} \text{ mA}$$

Solution in frequency domain:  $V_o = 400(\mathbf{I}_1 - \mathbf{I}_2) = 11.98\angle 18.8^\circ \text{ V}$

In time domain:  $v_o(t) = 11.98 \cos(40t + 18.8^\circ) \text{ V}$

# Problem 10.6-14: find $v_o(t)$

Find



$$v_s(t) = 25 \cos(100t - 15^\circ) \text{ V}$$



$$\omega = 100 \text{ rad/s}$$

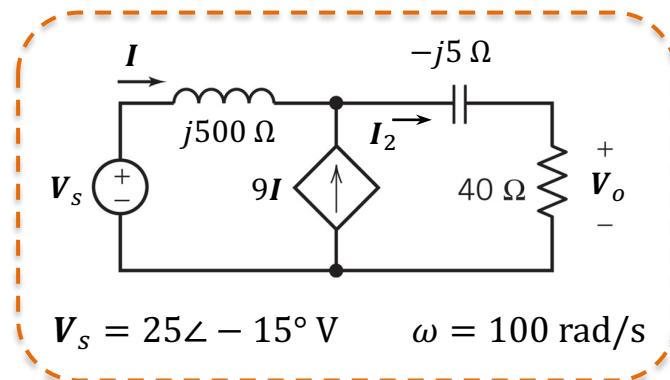
Convert to frequency domain



$$V_s = 25\angle -15^\circ \text{ V}$$

$$Z_L(\omega) = j\omega L$$

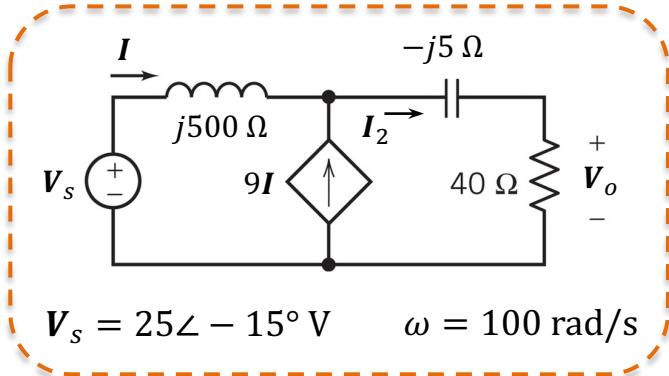
$$Z_C(\omega) = 1/j\omega C$$



$$Z_L = j100 \cdot 5 = j500 \Omega$$

$$Z_C = \frac{1}{j100 \times 2 \times 10^{-3}} = -j5 \Omega$$

# Problem 10.6-14



$$\text{KCL at top node: } I_2 = (9 + 1)I = 10I$$

Super-mesh equation:

$$25\angle -15^\circ = j500I + (-j5)10I + (40)10I$$

$$25\angle -15^\circ = j500I - j50I + 400I$$

$$25\angle -15^\circ = (j450 + 400)I$$

$$I = \frac{25\angle -15^\circ}{j450 + 400} = (41.52 \angle -63.37^\circ) \text{ mA}$$

Solution in frequency domain:  $V_o = (40)10I = (16.61 \angle -63.37^\circ) \text{ V}$

In time domain:  $v_o(t) = 16.61 \cos(100t - 63.37^\circ) \text{ V}$