



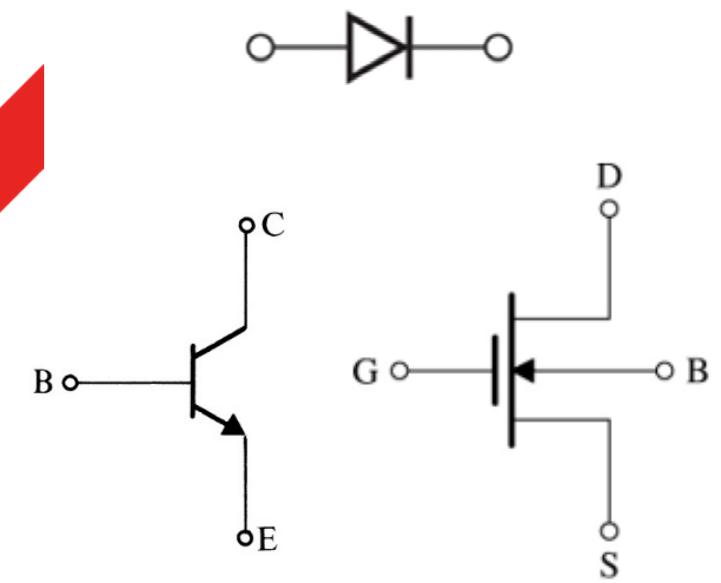
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# ELEC2005

# Electrical and Electronic Systems

MOSFETS – PART 2

DAVID PAYNE





# Lecture 6

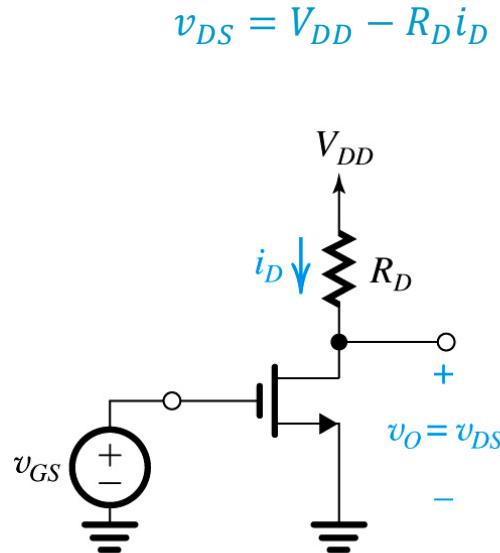
- 1. MOSFET Amplifiers**
- 2. MOSFET Small Signals**
- 3. Examples**

# MOSFET

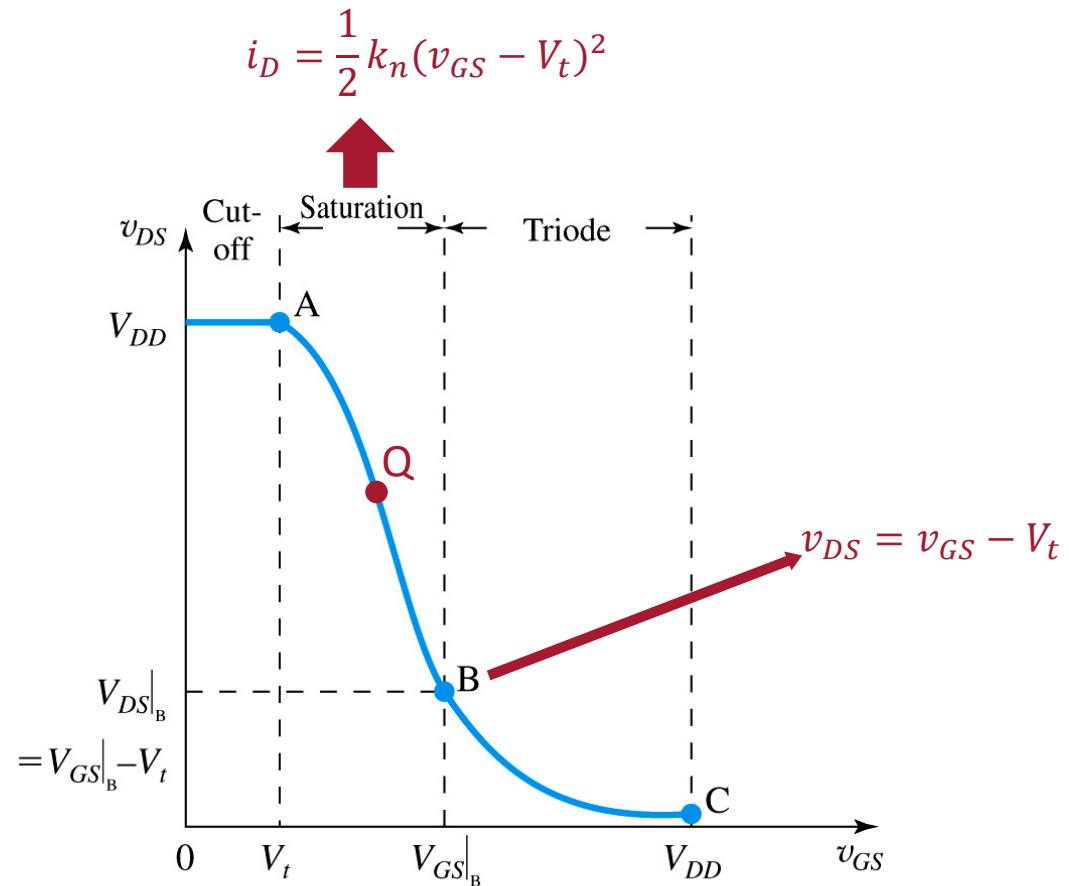


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## HOW TO MAKE AN AMPLIFIER OR SWITCH



(a)



(b)

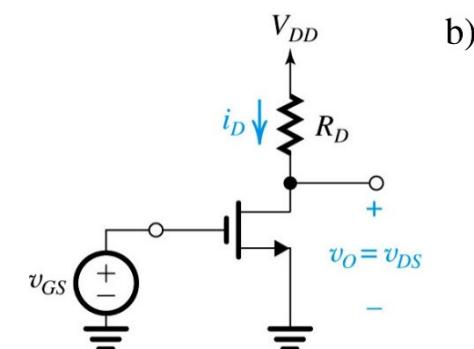
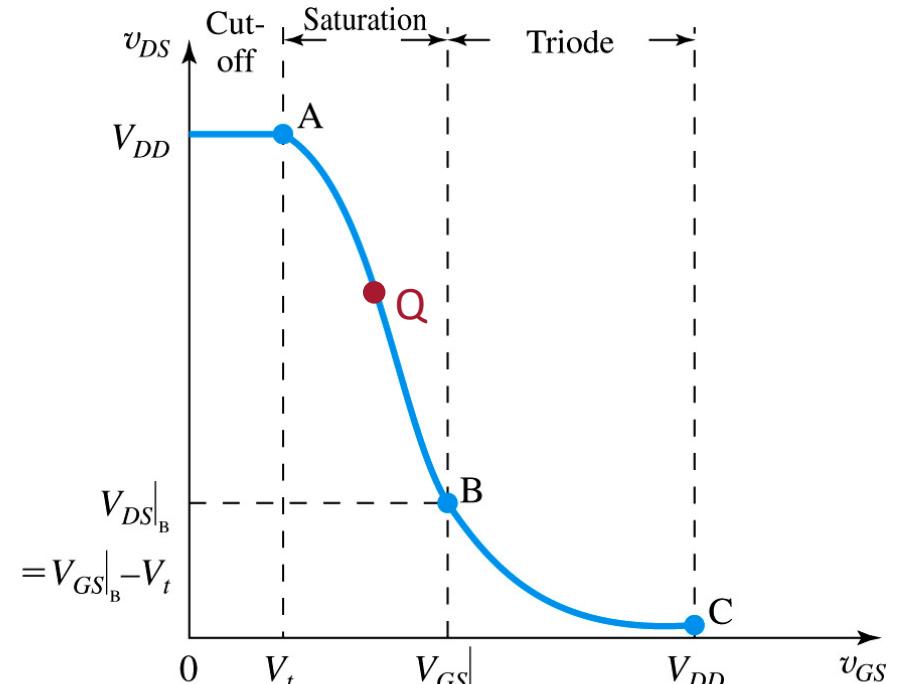
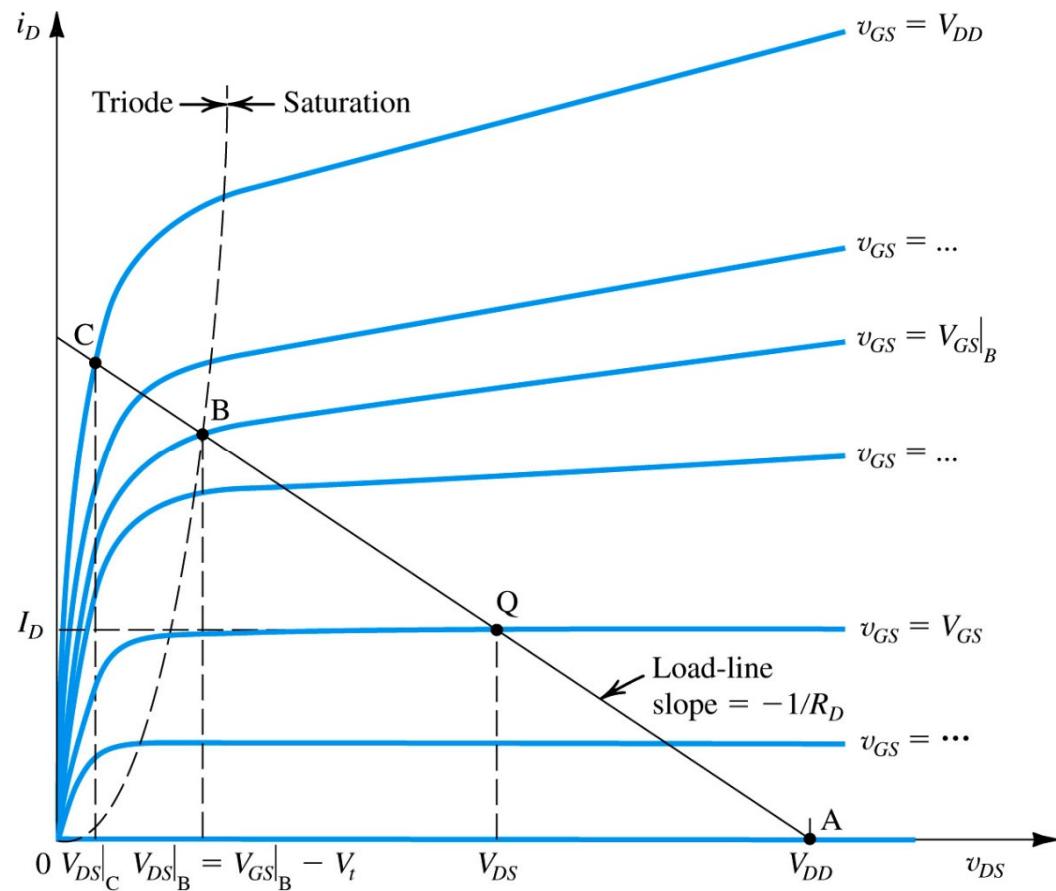
- (a) Simple MOSFET amplifier with input  $v_{GS}$  and output  $v_{DS}$ . (b) The voltage transfer characteristic (VTC) of the amplifier in (a). The three segments of the VTC correspond to the three regions of operation of the MOSFET.

# MOSFET



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## GRAPHICAL METHOD TO FIND OP MODE



# MOSFET

## VOLTAGE GAIN



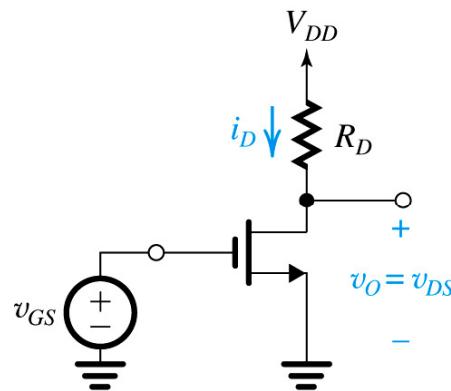
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$$v_{DS} = V_{DD} - i_D R_D$$

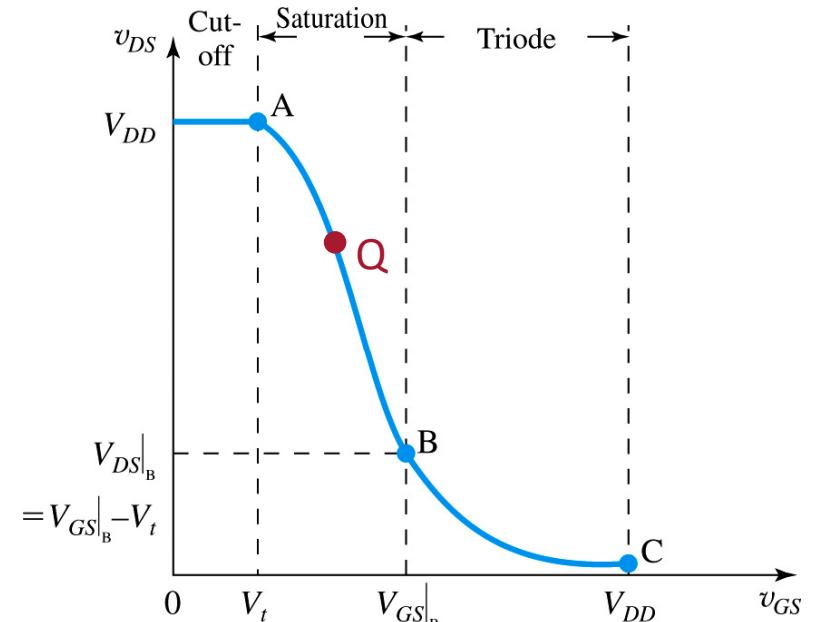
In saturation (Q):

$$i_D = \frac{1}{2} k_n (v_{GS} - V_t)^2$$

$$v_{DS} = V_{DD} - \frac{1}{2} R_D k_n (v_{GS} - V_t)^2$$



(a)



(b)

Gain:

$$A_v = \frac{\partial v_{DS}}{\partial v_{GS}} = -k_n R_D (v_{GS} - V_t) = -k_n R_D V_{OV}$$

$$i_D = \frac{1}{2} k_n V_{OV}^2 \rightarrow A_v = -\frac{i_D R_D}{V_{OV}/2}$$

as max  $i_D = \frac{V_{DD}}{R_D}$

$$|A_{v,\max}| = \frac{V_{DD}}{V_{OV}/2} \quad \text{typically } \sim 10$$

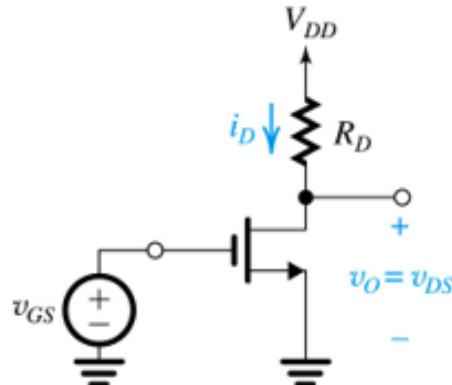
# Class Exercise



A MOSFET amplifier's supply voltage is 7V. If  $V_{GS\ max}=2V$  and  $V_{tn}=1V$ , what is the maximum voltage gain?

When poll is active, respond at [pollev.com/davidpayne187](http://pollev.com/davidpayne187)

What is the maximum voltage gain



14

10

1

Insufficient Data

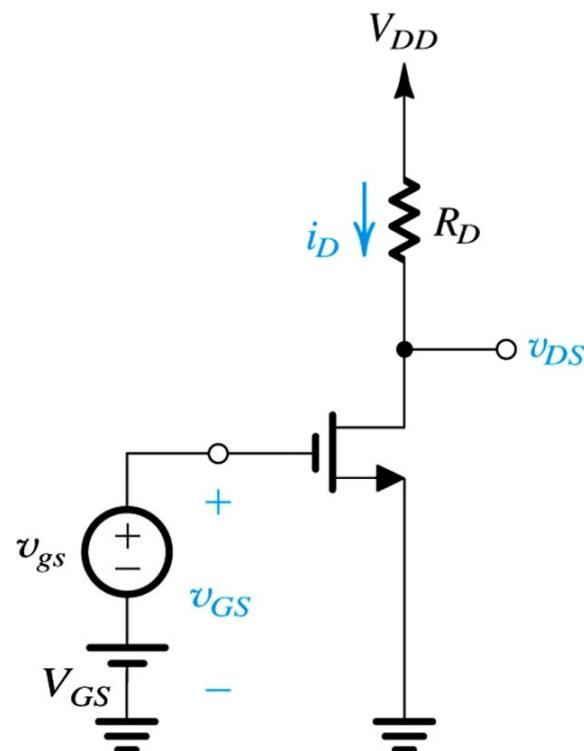
# MOSFET

## SMALL SIGNALS

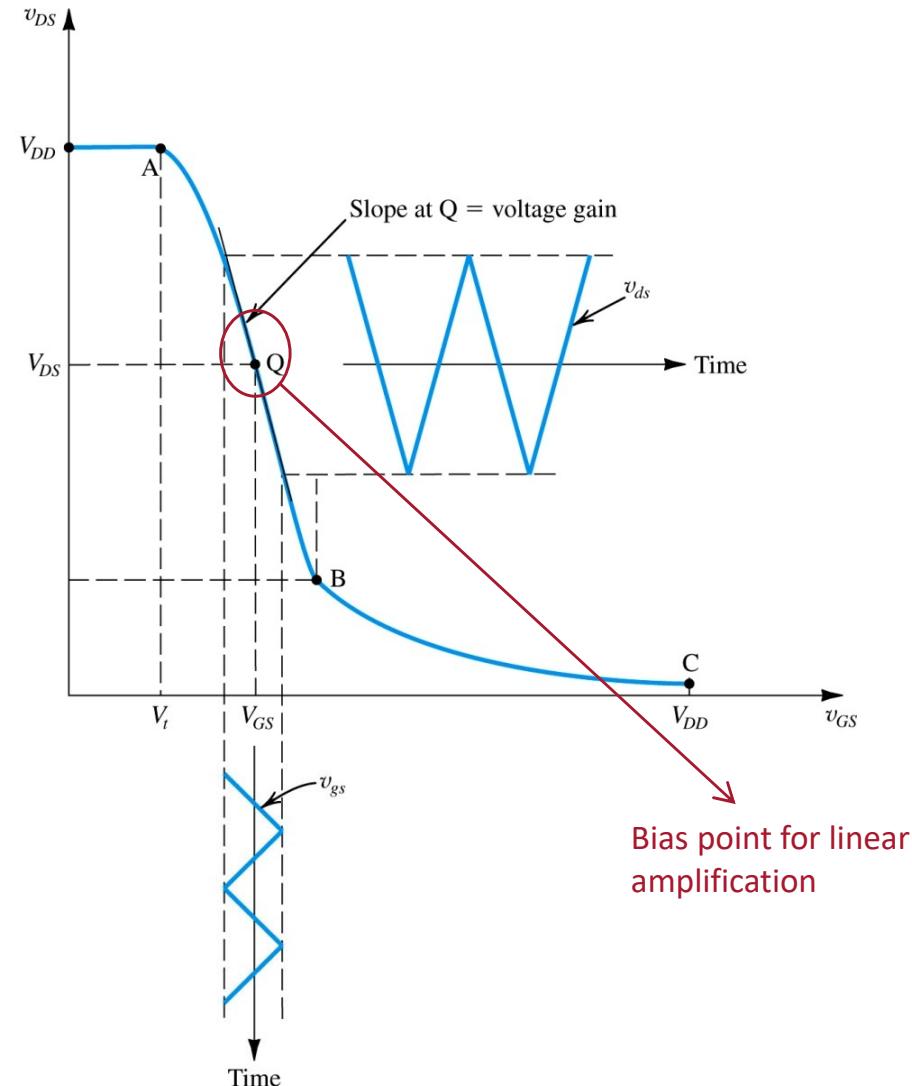


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DC voltages ( $V_{GS}$  and  $V_{DD}$ ) set the bias point (Q) and AC signal  $v_{gs}$  is amplified.



(a)



(b)

# MOSFET

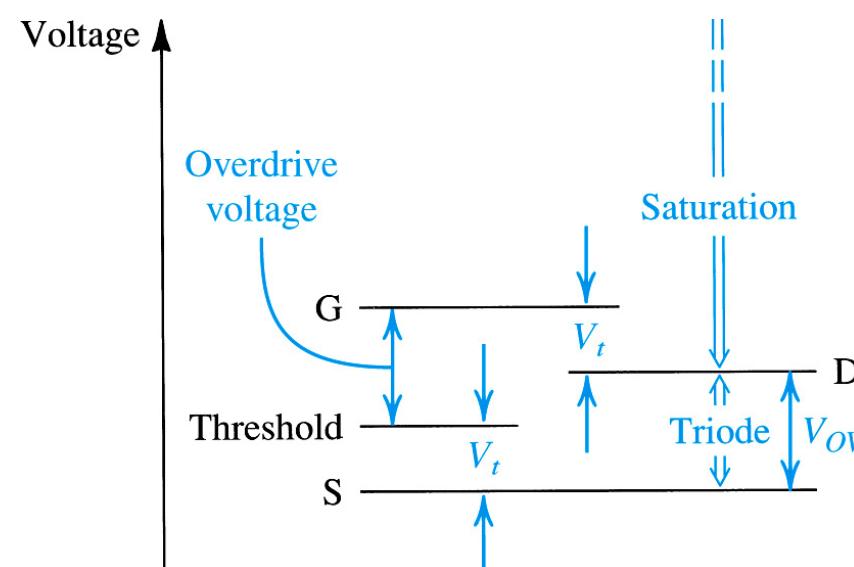
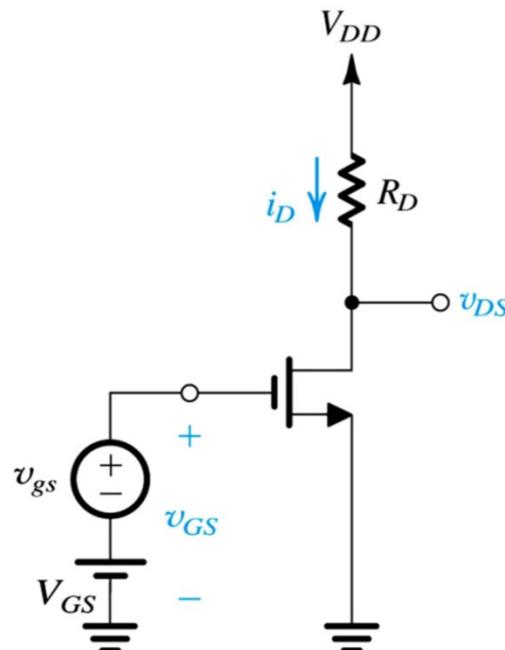


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## EXAMPLE-01

Consider the amplifier circuit shown below. The transistor is specified to have  $V_t = 0.4V$ ,  $k'_n = 0.4$  mA/V<sub>2</sub>,  $W/L = 10$  and  $\lambda = 0$ . Also let  $V_{DD} = 1.8V$ ,  $R_D = 17.5k\Omega$ , and  $V_{GS} = 0.6V$

- a) For  $v_{gs} = 0$  (and hence  $v_{ds} = 0$ ) find  $V_{OV}$ ,  $I_D$ ,  $V_{DS}$  and  $A_v$ ,
- b) DC bias point: what mode is the transistor in?



# MOSFET

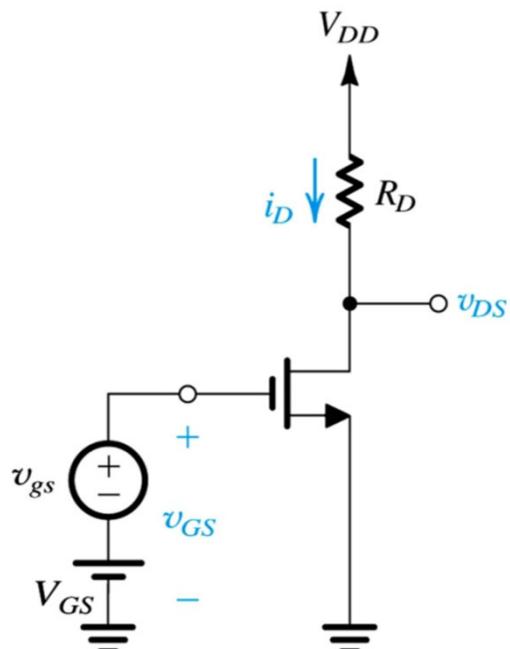


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## EXAMPLE-01

Consider the amplifier circuit shown below. The transistor is specified to have  $V_t = 0.4V$ ,  $k'_n = 0.4 \text{ mA/V}_2$ ,  $W/L = 10$  and  $\lambda = 0$ . Also let  $V_{DD} = 1.8V$ ,  $R_D = 17.5\text{k}\Omega$ , and  $V_{GS} = 0.6V$

- a) For  $v_{gs} = 0$  (and hence  $v_{ds} = 0$ ) find  $V_{ov}$ ,  $I_D$ ,  $V_{DS}$  and  $A_v$ ,
- b) DC bias point: what mode is the transistor in?



a) no small signal

$$V_{ov} = V_{GS} - V_t = 0.6 - 0.4 = \underline{\underline{0.2}}$$

$$I_D = \frac{1}{2} k_n (V_{ov})^2 = \frac{0.4 \times 10 \times 0.2^2}{2} = \underline{\underline{0.08 \text{ mA}}}$$

$$V_{DS} = V_{DD} - I_D R_D = 1.8 - 1.4 = 0.4V$$

$$A_v = -k_n R_D V_{ov} = -4 \times 17.5 \times 0.2 = -14$$

b) ↑ more than  
 $V_{ov}$   
∴ saturation

Saturation, with  $V_{ov}=0.2 \text{ V}$  and  $I_D=0.08 \text{ mA}$ ,  $V_{DS}=0.4 \text{ V}$

$$A_v = -k_n R_D V_{ov} = -k'_n \frac{W}{L} R_D V_{ov} = -14$$

# MOSFET



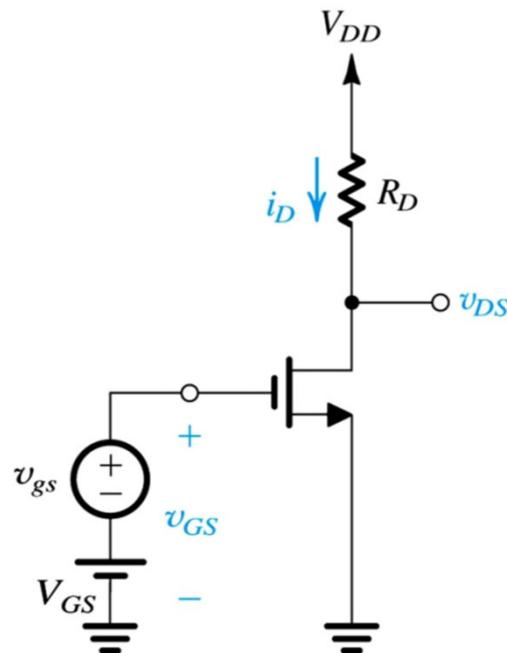
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## EXAMPLE-01

Consider the amplifier circuit shown below. The transistor is specified to have  $V_t = 0.4V$ ,  $k'_n = 0.4 \text{ mA/V}_2$ ,  $W/L = 10$  and  $\lambda = 0$ . Also let  $V_{DD} = 1.8V$ ,  $R_D = 17.5\text{k}\Omega$ , and  $V_{GS} = 0.6V$

c) What is the maximum symmetrical signal swing allowed at the drain? Hence find the maximum allowable amplitude of a sinusoidal  $v_{gs}$ .

d) What is the criterion for our gain calculation to be applicable?



c) We need to stay in saturation  
so  $V_{DS} > V_{OV}$  and  $V_{GS} > V_t$   
 $V_{DS} - V_{OV} = 0.2V \leftarrow \text{Max output swing}$   
- Divide by gain to get input swing

$$\frac{0.2V}{|A_v|} = 14.2mV$$

$v_{DS} > V_{OV}$ , hence the allowed swing at  $V_{DS}$  is 0.2 V swing.

$$\hat{v}_{ds} < 0.2 \text{ V}$$

At first glance:

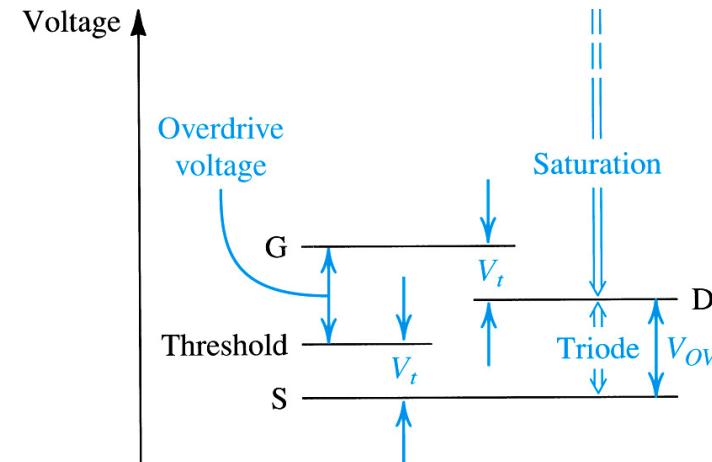
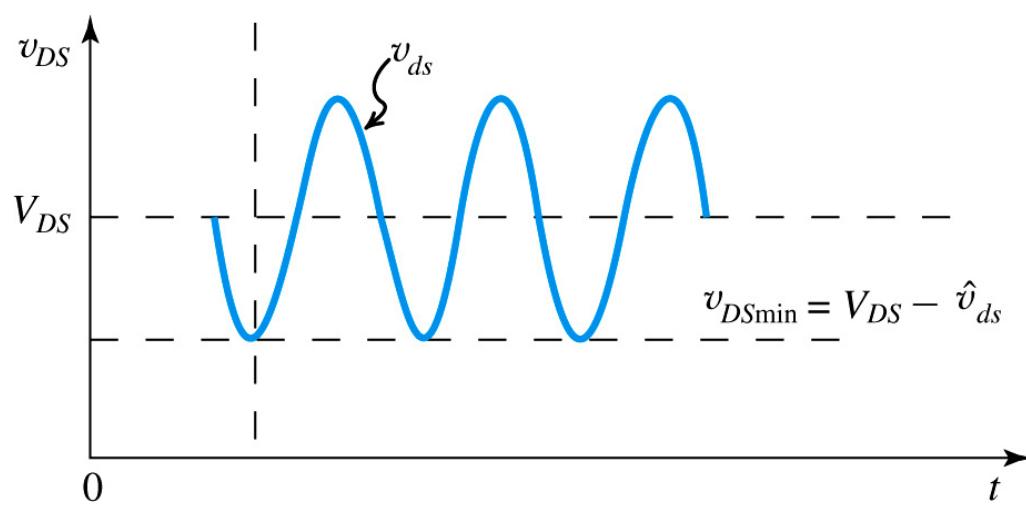
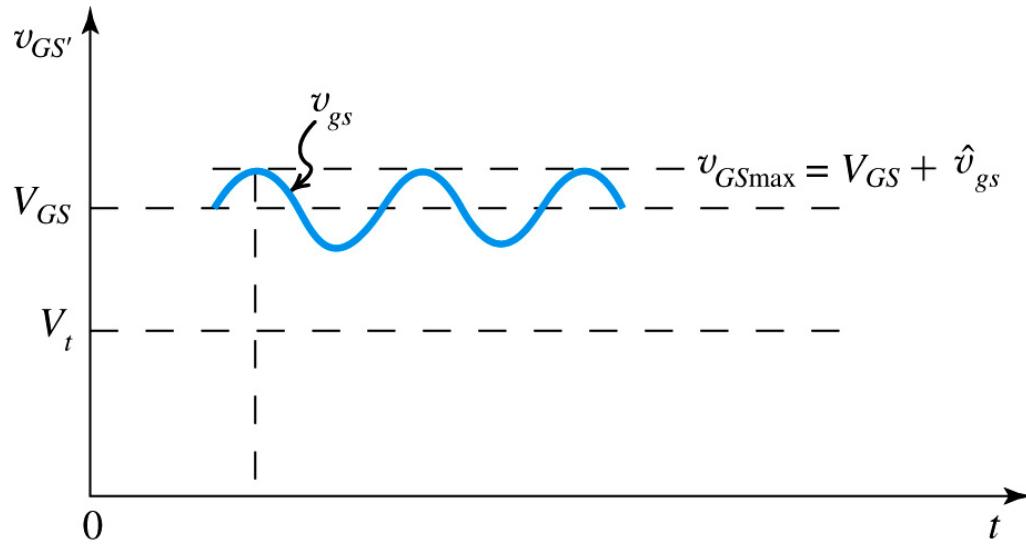
$$v_{gs} \leq \frac{0.2}{|A_v|} = 14.2mV$$

# MOSFET

## EXAMPLE CONT'D



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$$v_{DS\min} \geq v_{GS\max} - V_t$$

$$0.4 \text{ V} - \hat{v}_{ds} \geq 0.6 \text{ V} + \hat{v}_{gs} - 0.4 \text{ V}$$

express in terms of just  $v_{gs}$ :

$$0.4 - |A_v| v_{gs} \geq 0.2 + v_{gs}$$

$$v_{gs} (|A_v| + 1) \leq 0.2$$

$$v_{gs} \leq \frac{0.2}{|A_v| + 1} = 13.3 \text{ mV}$$



# Lecture 6

1. MOSFET Amplifiers
2. **MOSFET Small Signals**
3. Examples

# MOSFET



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## SMALL SIGNAL APPROXIMATION – SATURATION REGION

Similar technique for linearization as for the exponential diode and the BJT, but no need for Taylor series

$$v_{GS} = V_{GS} + v_{gs}$$

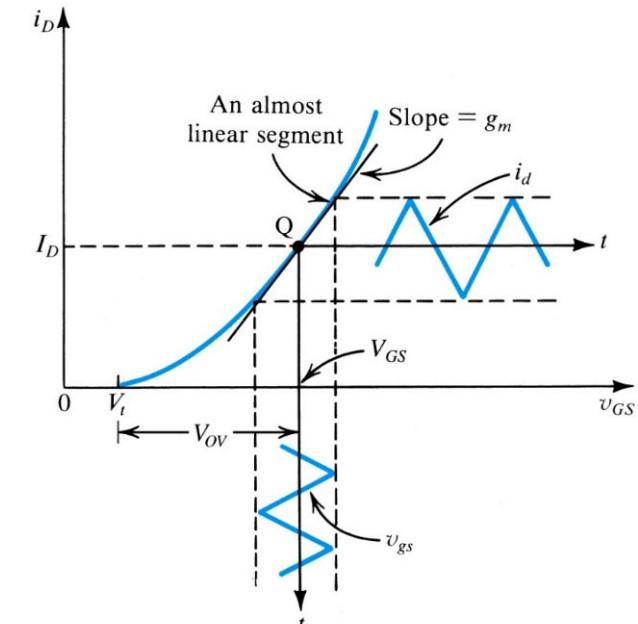
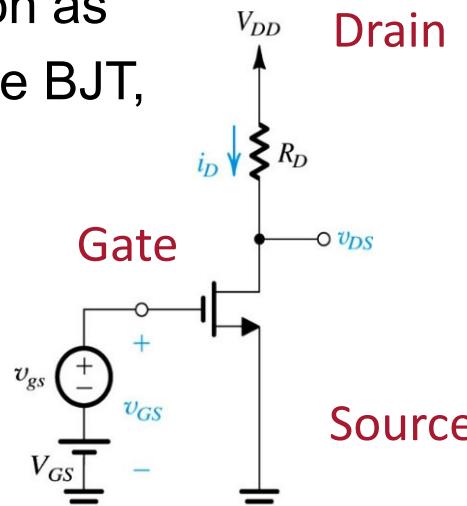
$$i_D = I_D + i_d$$

$$i_D = \frac{1}{2} k_n (v_{GS} - V_t)^2$$

$$= \frac{1}{2} k_n (V_{GS} - V_t)^2 + k_n (V_{GS} - V_t) v_{gs} + \frac{1}{2} k_n v_{gs}^2$$

if  $v_{gs} \ll 2(V_{GS} - V_t)$

$$i_D \approx \underbrace{\frac{1}{2} k_n (V_{GS} - V_t)^2}_{I_D} + \underbrace{k_n (V_{GS} - V_t) v_{gs}}_{i_d}$$



Transconductance

$$i_d = g_m v_{gs} \quad g_m = k_n (V_{GS} - V_t)$$

# MOSFET



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## SMALL SIGNAL APPROX – SATURATION REGION

$$i_D \approx \frac{1}{2} k_n (V_{GS} - V_t)^2 + \underbrace{k_n (V_{GS} - V_t) v_{gs}}_{i_d}$$

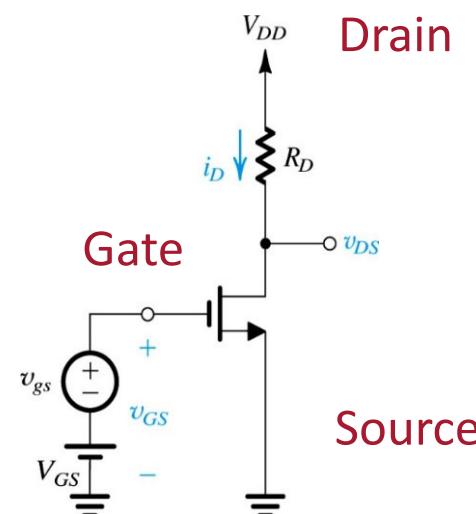
$$v_{GS} = V_{GS} + v_{gs}$$

$$i_D = I_D + i_d$$

$$I_D = \frac{1}{2} k_n (v_{GS} - V_t)^2$$

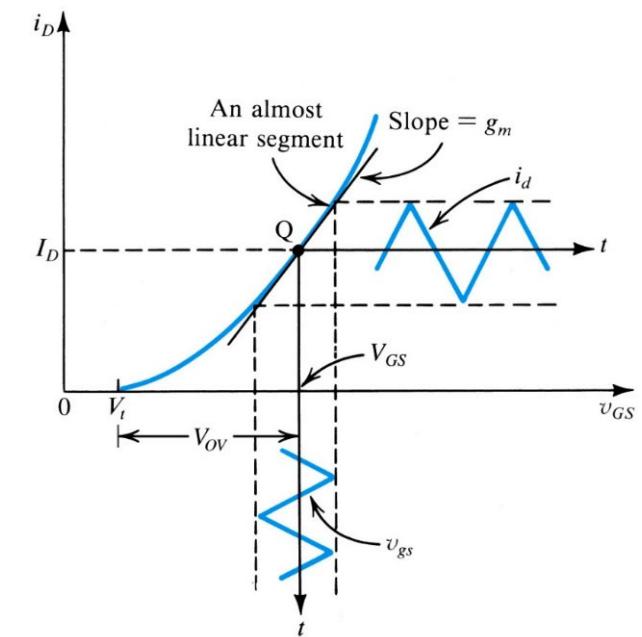
$$g_m = \left| \frac{\partial i_D}{\partial v_{GS}} \right|$$

Calculated at the DC operating point



Transconductance

$$i_d = g_m v_{gs} \quad g_m = k_n (V_{GS} - V_t)$$



# MOSFET



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## SMALL SIGNAL APPROX – SATURATION REGION

$$v_{GS} = V_{GS} + v_{gs}$$

$$i_D = I_D + i_d$$

$$i_D = \frac{1}{2} k_n (v_{GS} - V_t)^2$$

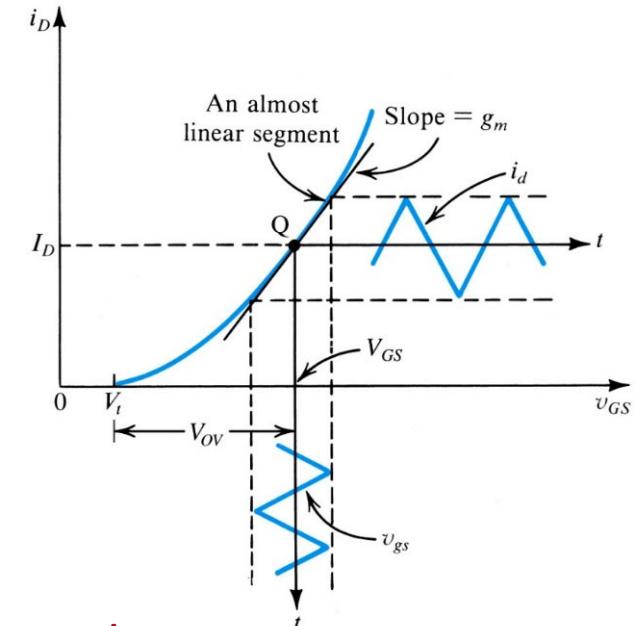
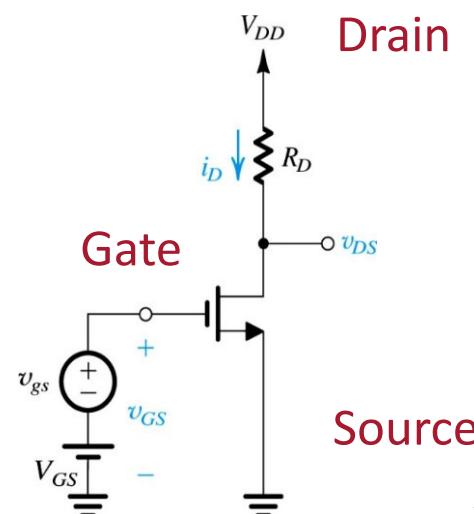
Transconductance

$$i_d = g_m v_{gs}$$

$$g_m = k_n (V_{GS} - V_t)$$

$$v_{DS} = V_{DD} - R_D i_D = V_{DD} - R_D (I_D + i_d) = \underbrace{V_{DS}}_{\text{Linear approx.}} - \underbrace{R_D i_d}_{\text{DC}} - \underbrace{R_D i_d}_{\text{AC}}$$

Linear  
approx.



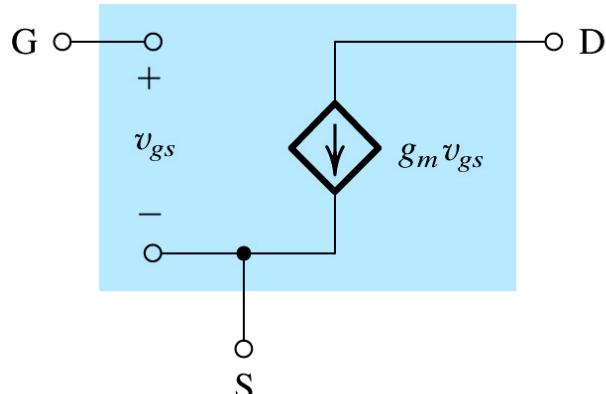
$$A_v = \frac{\partial v_{DS}}{\partial v_{GS}} = \frac{v_{ds}}{v_{gs}} = -g_m R_D = -k_n V_{OV} R_D$$

$$v_{ds} = -R_D i_d = -R_D g_m v_{gs}$$

# MOSFET

## SMALL SIGNAL MODEL – SATURATION REGION

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(a)

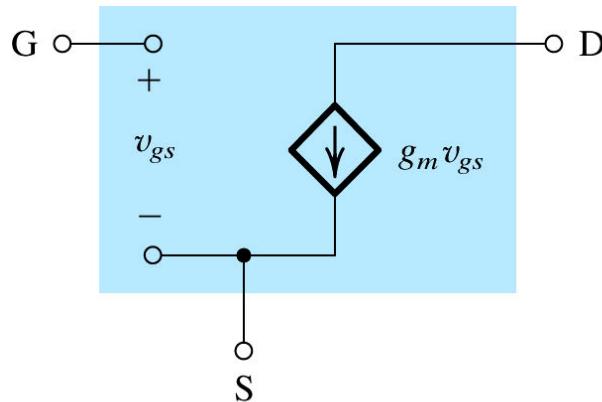
$$g_m = k_n(V_{GS} - V_t) = \frac{2I_D}{V_{GS} - V_t}$$

# MOSFET

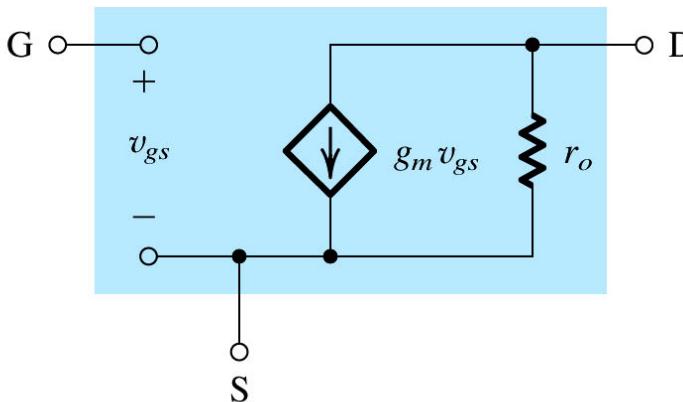


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## SMALL SIGNAL MODEL – SATURATION REGION



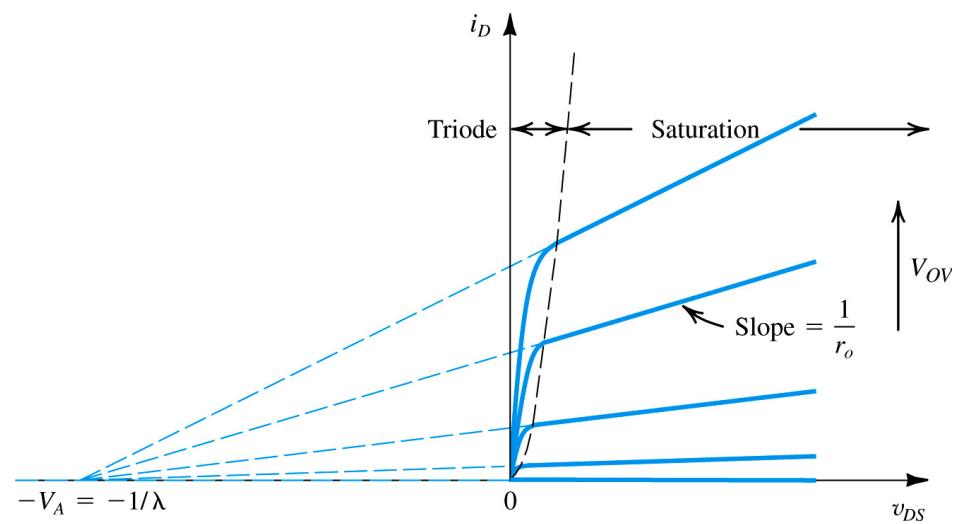
(a)



(b)

$$g_m = k_n(V_{GS} - V_t) = \frac{2I_D}{V_{GS} - V_t}$$

$$r_o = \frac{|V_A|}{I_D}$$

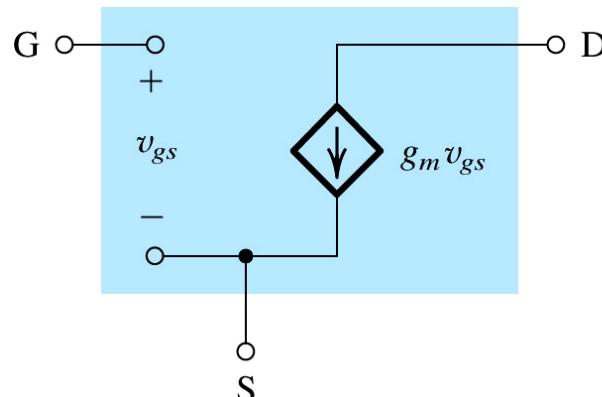
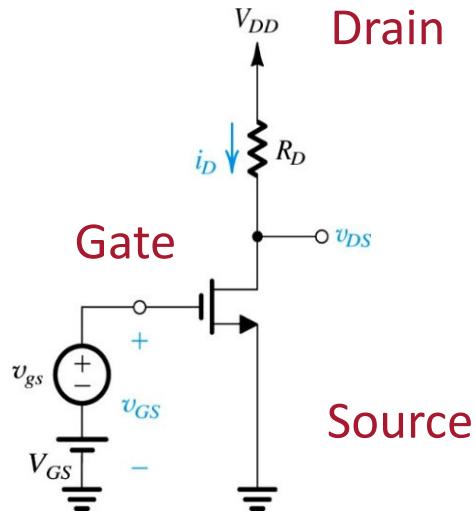


# Class Exercise



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What is the input resistance of this circuit?



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What is the input resistance?

Zero

Infinite

Unknown

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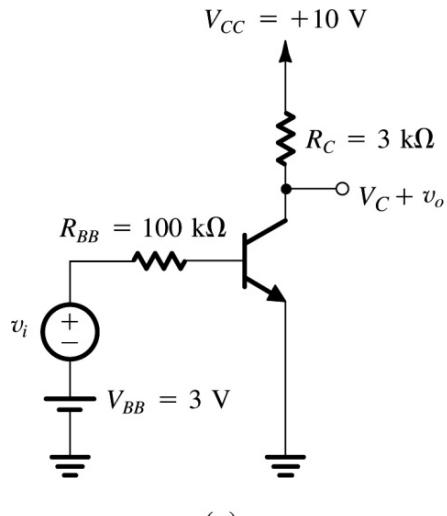
# Reminder - BJTs



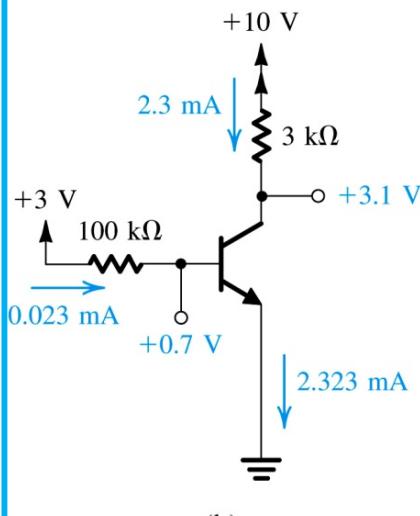
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## SMALL SIGNAL FOR BJTS

Amplifier circuit



Step 1: DC operating point

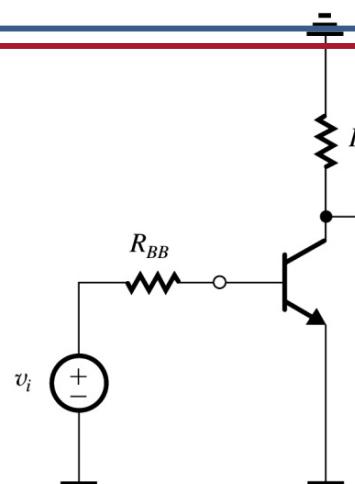


Step 2: Small signal parameters

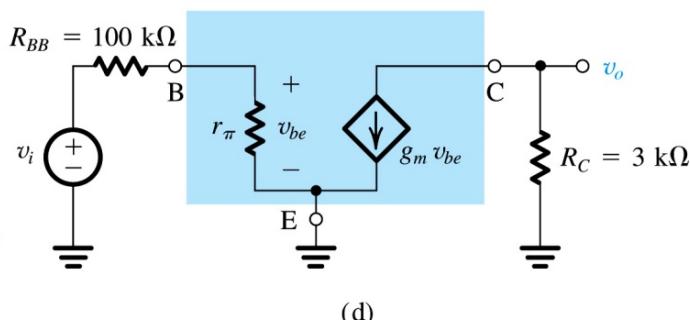
$$g_m = \frac{I_C}{V_T} = \frac{2.3 \text{ mA}}{25 \text{ mV}} = 92 \text{ mA/V}$$

$$r_\pi = \frac{\beta}{g_m} = \frac{100}{92} \text{ k}\Omega = 1.09 \text{ k}\Omega$$

$$r_e = \frac{V_T}{I_E} = \frac{25 \text{ mV}}{(2.3 / 0.99) \text{ mA}} = 10.8 \Omega$$



Step 3: Small signal equiv. circuit



Step 4: Small signal analysis

$$v_{be} = v_i \frac{r_\pi}{R_{BB} + r_\pi} = 0.011v_i$$

$$v_o = -R_C g_m v_{be} = -3.04v_i$$

$$A_v = \frac{v_o}{v_i} = -3.04$$



# Lecture 6

- 1. MOSFET Amplifiers**
- 2. MOSFET Small Signals**
- 3. Examples**

# Class Exercise

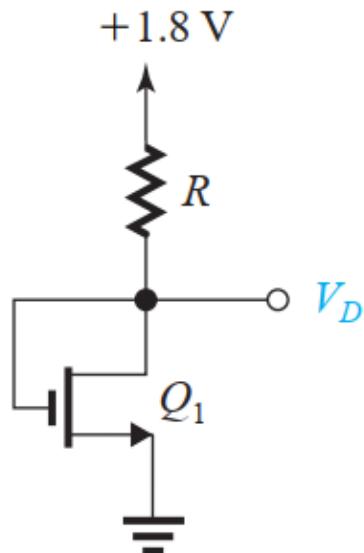


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Which operating mode is this transistor in if  $V_D > V_t$ ?

When poll is active, respond at [pollev.com/davidpayne187](http://pollev.com/davidpayne187)

**Which operating mode is the transistor in?**



Cutoff

Triode

Saturation

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# MOSFET

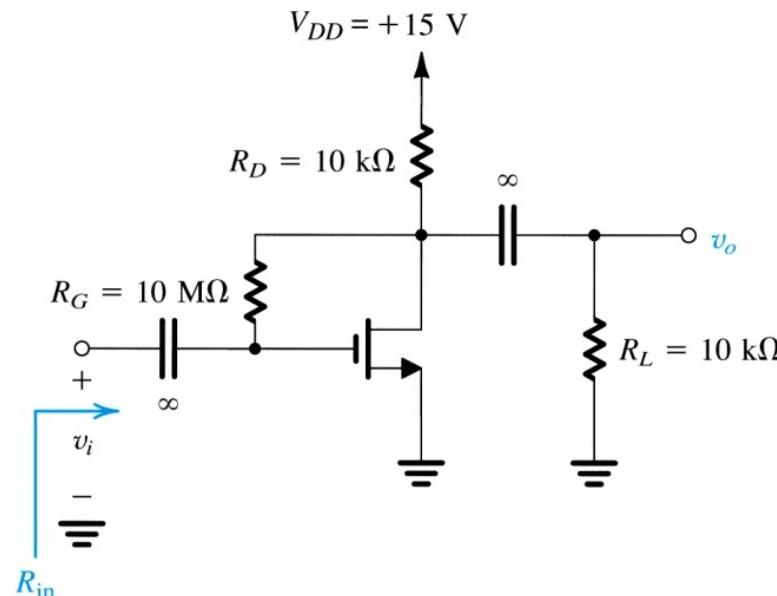
## EXAMPLE-2



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The circuit below shows a discrete common-source MOSFET amp, using a drain-to-gate resistor  $R_g$  for biasing. The input signal  $v_i$  is coupled to the gate via a large capacitor, the output signal at the drain is coupled to the load resistance  $R_L$  via another large capacitor. We want to analyze this amplifier to determine its **small-signal voltage gain, its input resistance and the largest allowable input**.

The transistor has  $V_t = 1.5V$ ,  $K'_n (W/L) = 0.25mA/V^2$  and  $V_A = 50V$ . Assume the coupling capacitors to be large enough act as short circuits at the frequencies of interest.



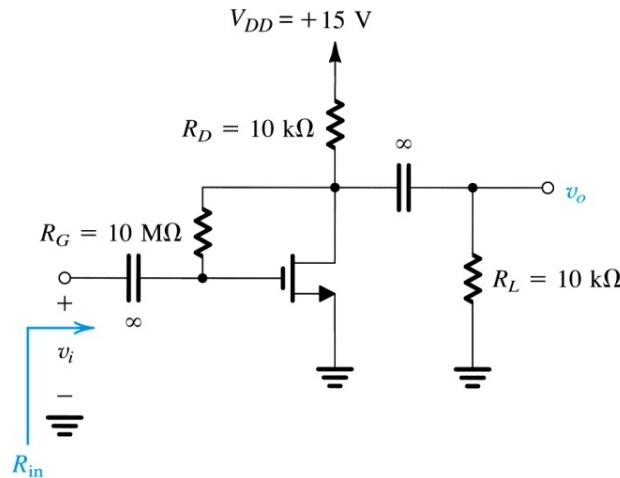
(a)

Amplifier Circuit

# MOSFET

## EXAMPLE – STEP 1 – DC EQUIVALENT

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(a)

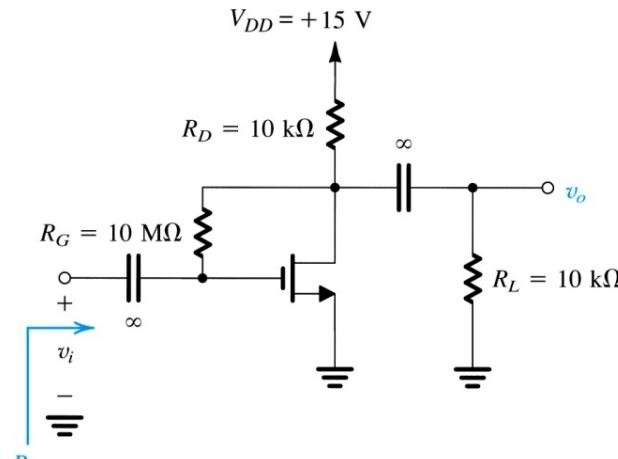
Amplifier Circuit

# MOSFET



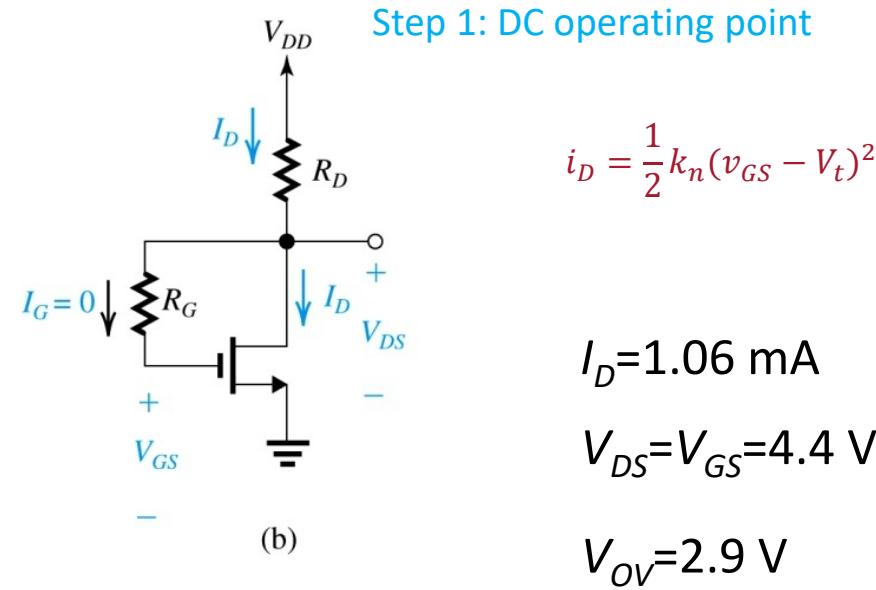
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## EXAMPLE – STEP 1 – DC EQUIVALENT



(a)

Amplifier Circuit



$$i_D = \frac{1}{2} k_n (v_{GS} - V_t)^2$$

$$I_D = 1.06 \text{ mA}$$

$$V_{DS} = V_{GS} = 4.4 \text{ V}$$

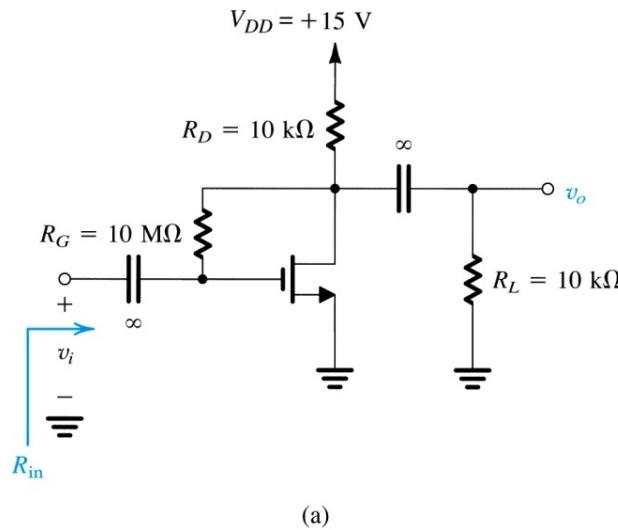
$$V_{OV} = 2.9 \text{ V}$$

# MOSFET

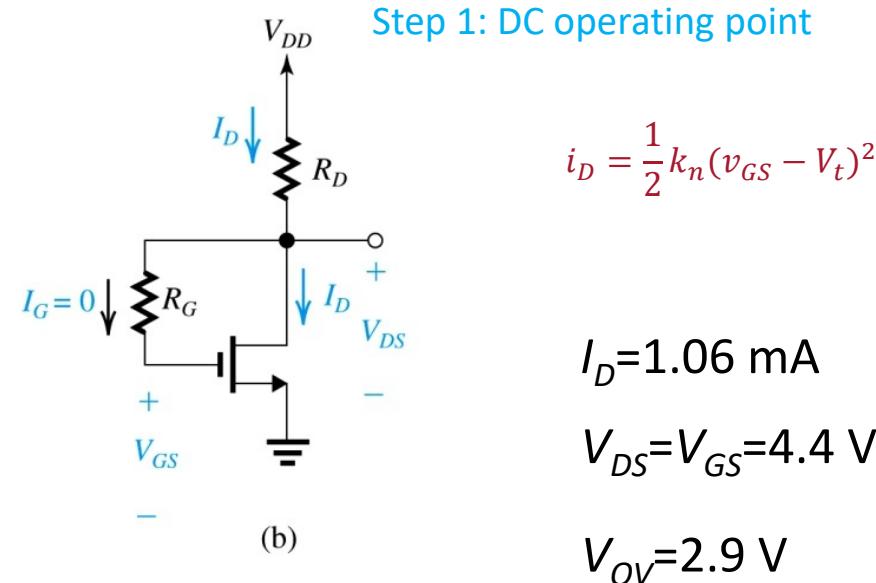


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## EXAMPLE – STEP 2 – SMALL SIGNAL PARAMETERS



Amplifier Circuit



dc equivalent circuit

Step 2: Small signal parameters

$$g_m = k_n V_{OV} = 0.725 \text{ mA/V}$$

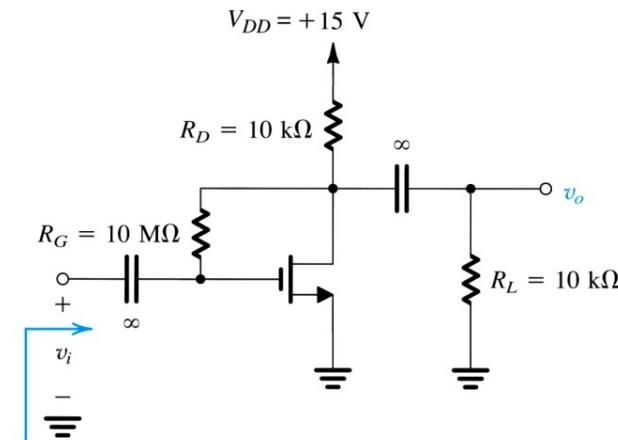
$$r_o = V_A / I_D = 50 \text{ V} / 1.06 \text{ mA} = 47 \text{ k}\Omega$$

# MOSFET



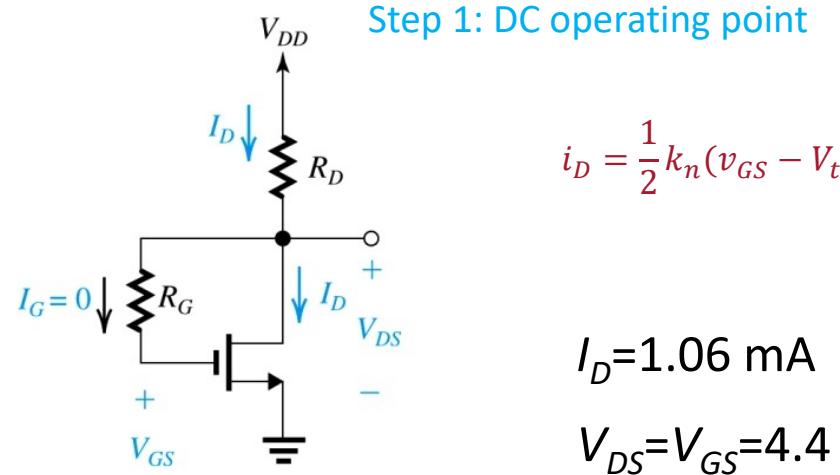
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## EXAMPLE – STEP 3 – AC EQUIVALENT



(a)

Amplifier Circuit



(b)

dc equivalent circuit

Step 1: DC operating point

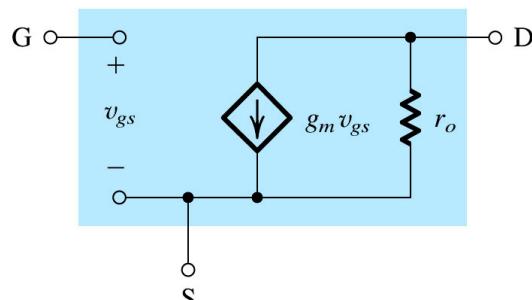
$$I_D = \frac{1}{2} k_n (v_{GS} - V_t)^2$$

$$I_D = 1.06 \text{ mA}$$

$$V_{DS} = V_{GS} = 4.4 \text{ V}$$

$$V_{OV} = 2.9 \text{ V}$$

MOSFET small signal model  
(from formula sheet)



(b)

Step 2: Small signal parameters

$$g_m = k_n V_{OV} = 0.725 \text{ mA/V}$$

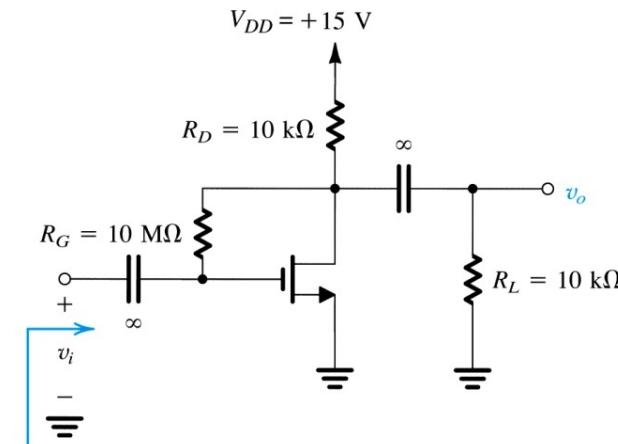
$$r_o = V_A / I_D = 50 \text{ V} / 1.06 \text{ mA} = 47 \text{ k}\Omega$$

# MOSFET



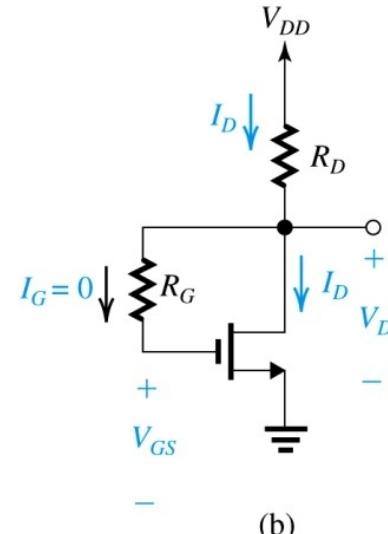
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## EXAMPLE – STEP 3 – AC EQUIVALENT



(a)

Amplifier Circuit



(b)

dc equivalent circuit

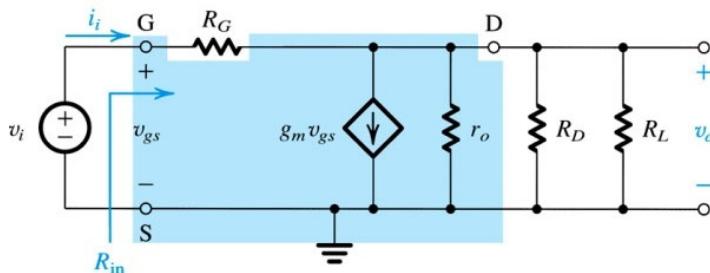
$$i_D = \frac{1}{2} k_n (v_{GS} - V_t)^2$$

$$I_D = 1.06 \text{ mA}$$

$$V_{DS} = V_{GS} = 4.4 \text{ V}$$

$$V_{OV} = 2.9 \text{ V}$$

### Step 3: Small signal equiv. circuit



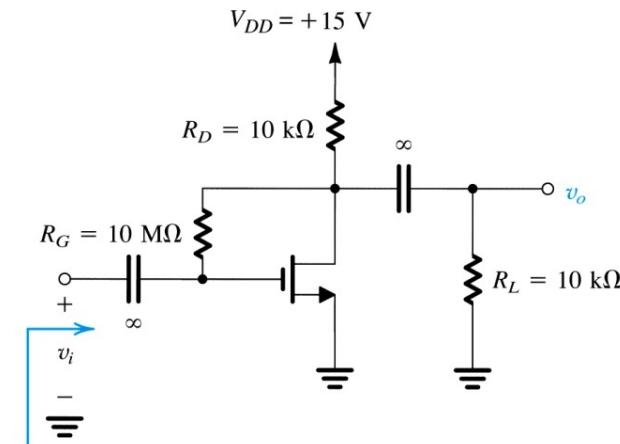
Ac small-signal equivalent circuit

# MOSFET



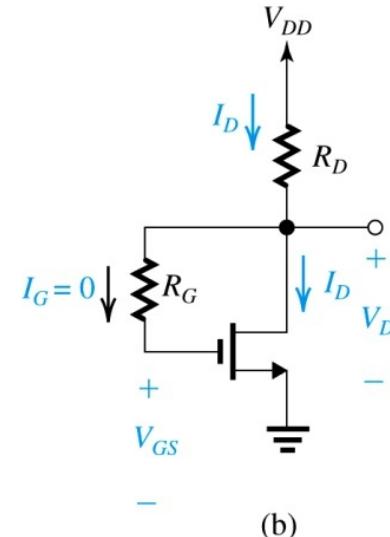
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## EXAMPLE – STEP 3 – AC EQUIVALENT SIMPLIFIED



(a)

Amplifier Circuit



(b)

$$I_D = \frac{1}{2} k_n (v_{GS} - V_t)^2$$

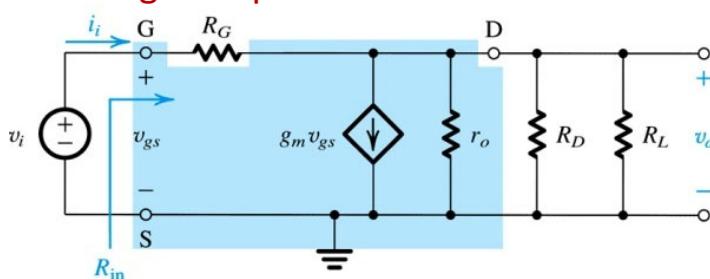
$$I_D = 1.06 \text{ mA}$$

$$V_{DS} = V_{GS} = 4.4 \text{ V}$$

$$V_{OV} = 2.9 \text{ V}$$

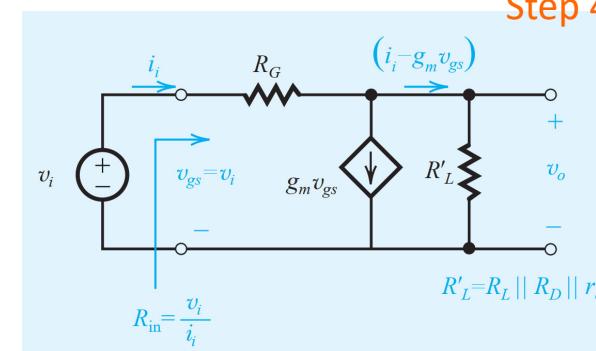
dc equivalent circuit

Step 3: Small signal equiv. circuit



Ac small-signal equivalent circuit

Step 4: Small signal analysis



$$R'_L = 4.52 \text{ k}\Omega$$

Ac small-signal equivalent circuit (simplified)

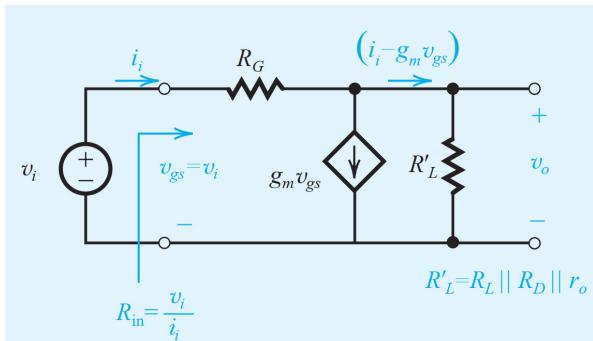
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## EXAMPLE – STEP 3 – AC EQUIVALENT SIMPLIFIED

Step 4: Small signal analysis



$$V_o = (i_i - g_m v_{gs}) R' L \quad i_i = \frac{V_{gs} - V_o}{R_G}$$

$$\therefore V_o = \left( \frac{V_{gs} - V_o}{R_G} - g_m v_{gs} \right) R' L$$

Rearrange to get  $V_o = \frac{R' L V_{gs} - V_o R' L}{R_G} - g_m v_{gs} R' L$

$$V_o (R_G + R'_L) = R' L V_{gs} - g_m v_{gs} R' L R_G$$

$$\therefore V_o = \frac{R' L V_{gs} - g_m v_{gs} R' L R_G}{R_G + R'_L} = R' L \frac{\frac{V_{gs}}{R_G} - g_m v_{gs}}{1 + \frac{R' L}{R_G}}$$

$$V_{gs} = V_i \quad A_v = \frac{V_o}{V_i}$$

# MOSFET



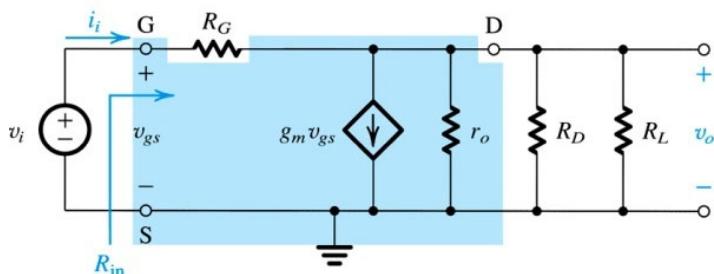
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## EXAMPLE – STEP 4 – AC ANALYSIS

$$A_v = \frac{v_o}{v_i} = -g_m R'_L \frac{1 - \left( \frac{1}{g_m R_G} \right)}{1 + \left( \frac{R'_L}{R_G} \right)} \approx -g_m R'_L = -3.3$$

$$R_i = \frac{R_G}{1 - A_v} = \frac{R_G}{1 + g_m R'_L} = 2.33 \text{ M}\Omega$$

$$\text{Max, allowed swing at input: } \hat{v}_i = \frac{V_t}{|A_v| + 1} = 0.35 \text{ V}$$



Ac small-signal equivalent circuit

$$R_{in} = \frac{V_i}{i_i} = \frac{V_i}{\frac{V_i - V_o}{R_G}} = \frac{R_G V_i}{V_i - V_o} = \frac{R_G}{1 - A_v}$$

condition  $V_{DS} > V_V$  so at edge of region:

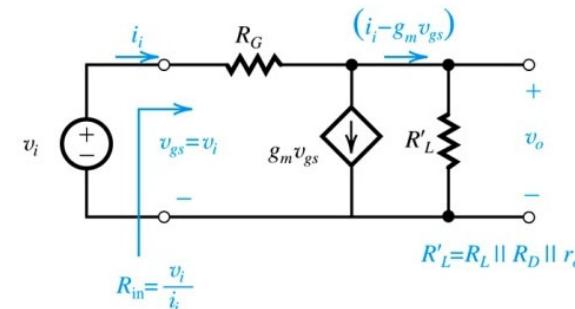
$$V_{DS,\min} = V_{GS,\max} - V_L \quad \text{Original Circuit}$$

$$V_{DS} - A_v V_i = V_{GS} + V_i - V_L$$

$$\therefore -A_v V_i = V_i - V_t$$

$$A_v V_i + V_i = V_L$$

$$V_i (A_v + 1) = V_L$$



Ac small-signal equivalent circuit (simplified)

# MOSFET

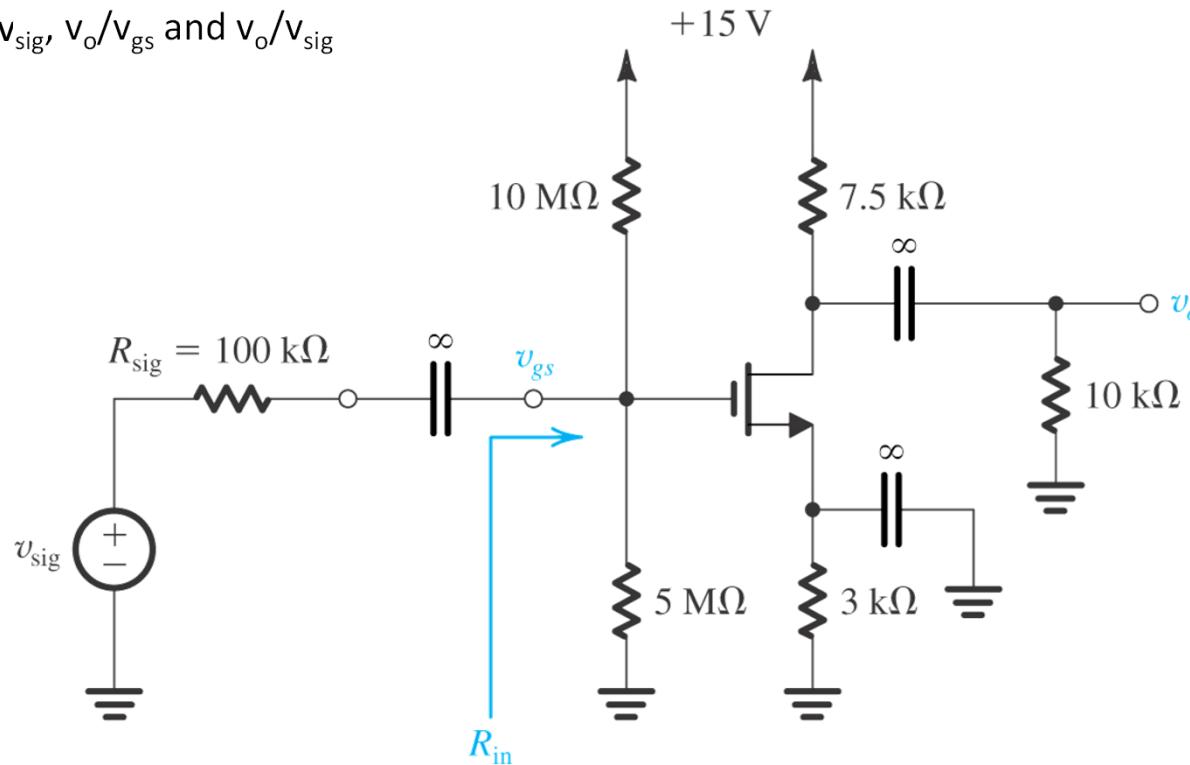


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## EXAMPLE 3

A discrete-circuit amplifier is shown below. The input signal  $v_{\text{sig}}$  is coupled to the gate through a very large capacitor. The transistor source is connected to ground at signal frequencies via a very large capacitor. The output voltage signal at the drain is coupled to a load resistance via a very large capacitor.

- If the transistor has  $V_t = 1\text{V}$  and  $K_n = 2\text{mA/V}^2$ , verify that the bias circuit establishes  $V_{GS} = 2\text{V}$ ,  $I_D = 1\text{mA}$  and  $V_D = +7.5\text{ V}$ . That is, assume these values and verify that they are consistent.
- Find  $g_m$  and  $r_o$  if  $V_A = 100\text{V}$
- Draw the small signal equivalent circuit
- Find  $R_{\text{in}}$ ,  $v_{gs}/v_{\text{sig}}$ ,  $v_o/v_{gs}$  and  $v_o/v_{\text{sig}}$

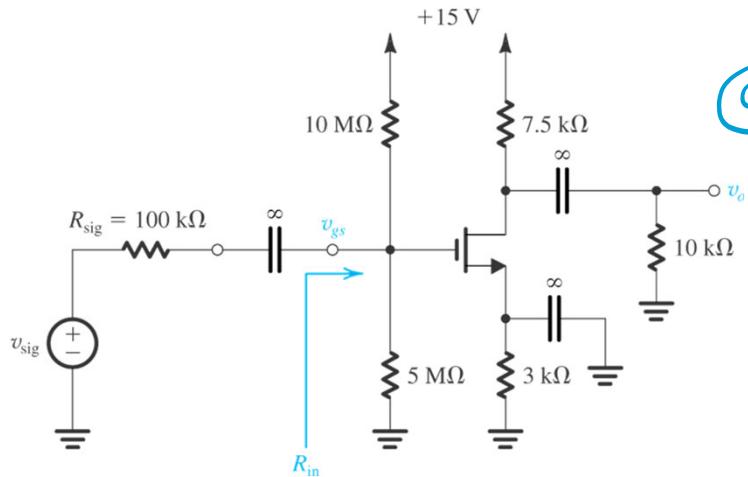


# MOSFET

## EXAMPLE 3

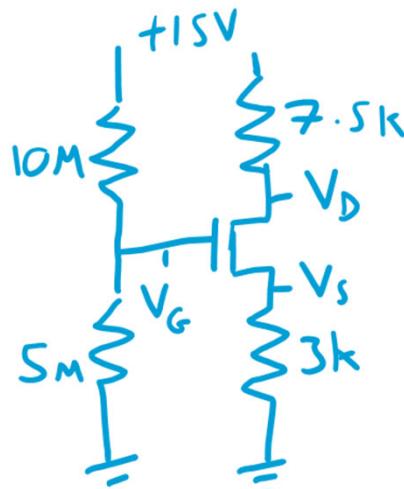


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a)

Step 1: Draw DC equivalent



$$V_G = \frac{S}{15} \times 15 = SV$$

↖ simple potential divider or no current to gate

$$V_S = 3 \times I_D$$

$$I_D = \frac{1}{2} k_n (S - 3I_D - 1) \approx \frac{1}{2} k_n (4 - 3I_D)^2$$

we get a quadratic for  $I_D$ :  $I_D = aI_D^2 - 2SI_D + 16 = 0$

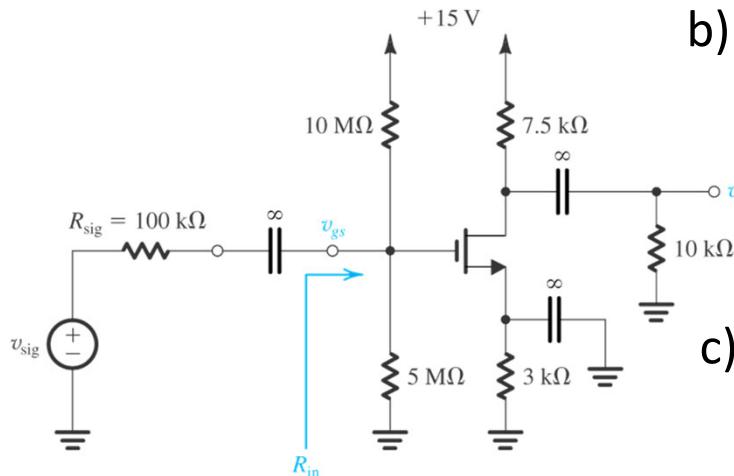
$$\begin{aligned} I_D &\rightarrow 1.77mA \\ I_D &\rightarrow 1mA \end{aligned}$$

$$V_S = 3 \times I_D = 3V \quad V_{GS} = S - 3 = 2V$$

$$V_D = 15 - 7.5 = 7.5V$$

# MOSFET

### EXAMPLE 3

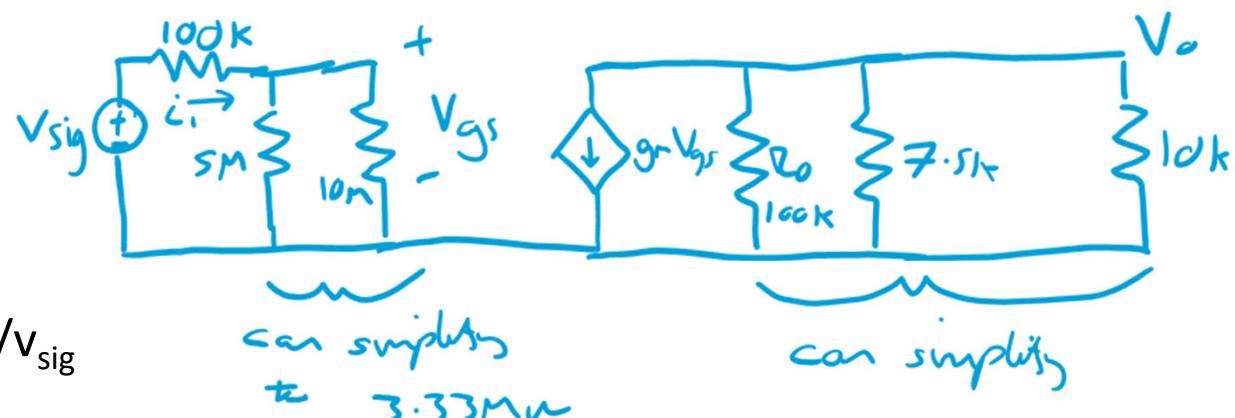


b) Find  $g_m$  and  $r_o$  if  $V_A = 100V$

## b) Small signal parameters

$$g_m = k_n V_{ov} = 2mAN \quad r_o = \frac{V_A}{I_D} = 100k\Omega$$

c) Draw the small signal equivalent circuit



d) Find  $R_{in}$ ,  $v_{gs}/v_{sig}$ ,  $v_o/v_{gs}$  and  $v_o/v_{sig}$

$$R_{in} = \frac{V_{gs}}{I_1}$$

$$i_i = \frac{V_{sig} - V_{gs}}{100k} \quad \text{so} \quad R_{in} = \frac{V_{gs} \times 100k}{V_{sig} - V_{gs}}$$

$$= \frac{100k}{\frac{V_{sig} - 1}{V_{gs}}} V_{gs} = V_g - V_s = V_g = \frac{3.3M}{3.4M} V_{sig}$$

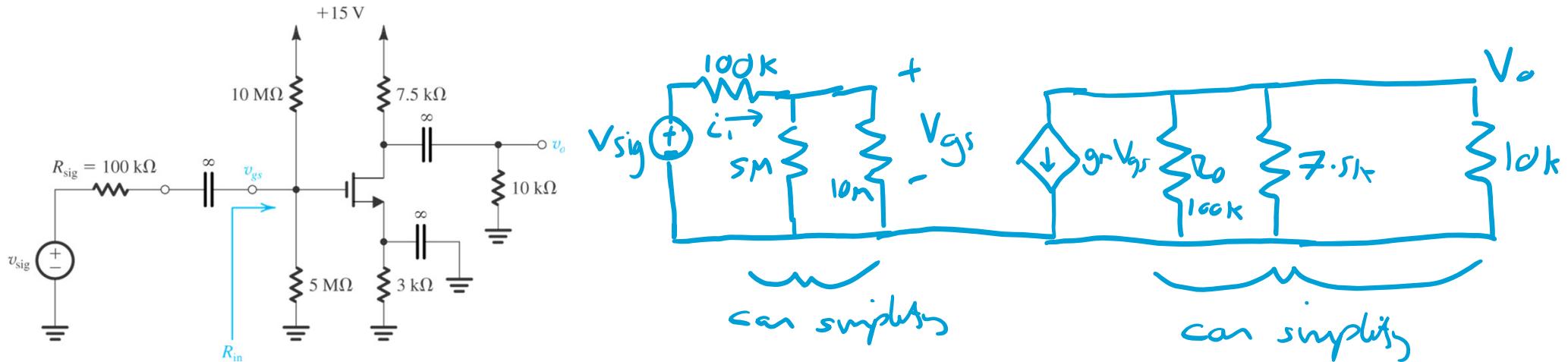
So  $R_{in} = \frac{100}{\frac{3.4}{3.3} - 1} = 3.33 M\Omega$

# MOSFET

## EXAMPLE 3



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④ cont'd

$$\frac{V_{gs}}{V_{sig}} = \frac{3.3}{3.4} \frac{V_{sig}}{V_{sig}} = 0.97 \frac{V}{V}$$

$$\frac{V_o}{V_{gs}} = -g_m (\Gamma_0 || R_D || R_L) = -8.2$$

$$\frac{V_o}{V_{sig}} = -8.2 \times 0.97 = -7.95$$

# This week



## LAB + SECOND ASSIGNMENT

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### MOSFET Lab:

- The worksheet is available on iLearn
- You will use the AD2 and your kit to make a MOSFET amplifier with biasing
- Complete as much of the lab as you can before attending the support session (prelab part is the minimum requirement)

### 2nd Assignment!

- The second assignment will be posted on Friday
- Submission is due early in week 8 (you have the semester break to work on it)